

# LM627/LM637 Precision Operational Amplifiers

## **General Description**

The LM627/LM637 series feature extremely low noise and excellent precision along with high speed. Voltage noise is a low 3 nV/ $\sqrt{\text{Hz}}$  in the flat band and rises to only 3.5 nV/ $\sqrt{\text{Hz}}$  at

10 Hz. The A grades offer guaranteed specifications of 25  $\mu$ V offset voltage and 0.3  $\mu$ V/°C drift, and their *guaranteed* 126 dB CMRR, 120 dB PSRR and voltage gain of 5 Million ensure an ultra-low V<sub>OS</sub> under all conditions.

The unity-gain stable LM627 is nearly twice as fast as the OP-27 with a slew rate of 4.5 V/ $\mu$ s and a 14 MHz gain-bandwidth product. Stable at gains of 5 or more, the decompensated LM637 is considerably faster.

Other enhancements of the LM627/LM637 include a guaranteed  $600\Omega$  load drive capability over temperature:  $\pm\,10V$  output swing at voltage gains over one million. Bias current has been reduced to 10 nA for the A and B grades and 25 nA for the C grade. Furthermore the LM627 may be overcompensated to allow it to drive capacitive loads up to 2000 pF while maintaining its superb dc specs.

## **Features**

■ Low Noise

3 nV/√Hz@1 kHz

3.5 nV/√Hz@ 10 Hz

■ Low V<sub>OS</sub>

25 μV Max 0.3 μV/°C Max

■ Offset Drift 100% Tested (A and B grades)

Offiset Drift 100% Tested (A and B grades)

Noise Voltage 100% Tested (A and B grades)High Gain5 Million Min

■ High CMRR
■ High PSRR

126 dB Min 120 dB Min

■ High Speed LM627:

14 MHz Gain-Bandwidth

LM637:

4.5 V/µs Slew Rate 65 MHz Gain-Bandwidth

14 V/µs Slew Rate

■ Guaranteed 600Ω drive over temperature

■ Wide Power Supply Range

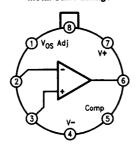
 $\pm$  4V to  $\pm$  18V

Overcompensation Pin

Allows driving high CL

## **Connection Diagrams**

### Metal Can Package



TL/H/9212-1

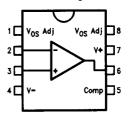
# Ordering Information

## LM627

**Top View** 

Package	Temperat	NSC				
	Military	Military Commercial				
TO-99	LM627AMH LM627BMH	LM627ACH LM627BCH LM627CH	H08C			
8-Pin Molded DIP		LM627CN	N08E			

### **DIP Packages**



TL/H/9212-2

#### LM637

Package	Temperat	NSC							
rackage	Military	Commercial	Drawing						
TO-99	LM637AMH LM637BMH	LM637ACH LM637BCH LM637CH	H08C						
8-Pin Molded DIP		LM637CN	N08E						

## **Absolute Maximum Ratings (Note 1)**

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

Differential Input Overdrive Current (Note 7)

Supply Voltage

Supply Voltage Input Voltage Output Short Circuit to Gnd Continuous

Power Dissipation (Note 8)

Molded DIP 1300 mW Metal Can 830 mW Storage Temperature Range -65°C to +150°C Lead Temperature (Soldering, 5 sec.) Maximum Junction Temperature ESD Rating  $C_{ZAP} = 100 \text{ pF}, R_{ZAP} = 1.5 \text{ k}\Omega$ 

## **Operating Ratings**

Temperature Range (Note 8)

 $-55^{\circ}C \leq T_{J} \leq \, +\, 125^{\circ}C$ AM and BM grades AC, BC, and C grades  $-25^{\circ}\text{C} \leq \bar{\text{T}}_{\text{J}} \leq +85^{\circ}\text{C}$ 

260°C

150°C

3 kV

Electrical Characteristics All limits guaranteed for T<sub>J</sub> = 25°C, V<sub>CM</sub> = 0, V<sub>O</sub> = 0 and ±15V supplies unless otherwise specified. Boldface limits apply at operating temperature extremes.

Parameter	Conditions	Тур		27AM 37AM	LM6: LM6:			
			Tested Limit (Note 5)	Design Limit (Note 6)	Tested Limit (Note 5)	Design Limit (Note 6)	Units	
Input Offset Voltage	(Note 2)	15	25 <b>55</b>		50 1 1 0		μV Max	
Input Offset Voltage Drift	(Note 3)	0.2	0.3		0.6		μV/°C Max	
Input Offset Voltage Long Term Stability	(Note 4)	0.2					μV/mo Max	
Input Bias Current		3	10 <b>20</b>		10 <b>20</b>		nA Max	
Input Offset Current		2	10 <b>20</b>		10 <b>20</b>		nA Max	
Input Noise Voltage	0.1 to 10Hz	0.08		0.18		0.18	μV p-p Max	
Input Noise Voltage Density	f = 10Hz f = 30Hz f = 1kHz	3.5 3.1 3.0	5.5 4.5 3.8		5.5 4.5 3.8		nV/√H Max	
Input Noise Current Density	f = 10Hz f = 30Hz f = 1kHz	1.7 1.0 0.4					pA/√H Max	
Input Resistance	Common-Mode	20					GΩ	
Input Voltage Range		±12	±11 ±10.3		±11 10.3		V Min	
Common-Mode Rejection Ratio	V <sub>CM</sub> = ±11V V <sub>CM</sub> = ±10V	140	126 <b>120</b>		126 <b>120</b>		dB Min	
Power Supply Rejection Ratio	$V_S = \pm 4V \text{ to } \pm 18V$ $V_S = \pm 4.5V \text{ to } \pm 18V$	140	120 117		120 <b>117</b>		dB Min	
Large-Signal Voltage Gain	$V_O = \pm 12V$ $R_L \ge 2 k\Omega$	10000	5000 <b>3000</b>		5000 <b>2000</b>			
	$V_O = \pm 10V$ $R_L \ge 1 k\Omega$	7000	4000 <b>2000</b>		3500 <b>1500</b>		V/m¹ Min	
	$R_{L} \ge 600\Omega$ $V_{O} = \pm 10V$	6000	3000 <b>1500</b>		2000 <b>1000</b>			

# **Electrical Characteristics** (Continued)

Parameter	Conditions	Тур		27AM B7AM	LM6: LM6:		
			Tested Limit (Note 5)	Design Limit (Note 6)	Tested Limit (Note 5)	Design Limit (Note 6)	Units
Output Voltage Swing	$R_L \ge 2 k\Omega$ $R_L \ge 600\Omega$	± 13.8 ± 12.5	±13 ±12.5 ±11 ±10.5		±13 ±12.5 ±11 ±10.5		V Min
Slew Rate	LM627 LM637 R <sub>L</sub> = 2k	4.5 14		3 10		3 10	V/μs <b>M</b> in
Gain-Bandwidth Product	LM627 LM637 f = 10 kHz	14 65		10 45		10 45	MHz Min
Output Resistance	Open Loop	50					Ω
Supply Current		3	4.5 <b>5.5</b>		4.5 <b>5.5</b>		mA Max
Offset Adjust Range	$R_P \ge 10  k\Omega$	±2					mV

**Electrical Characteristics** All limits guaranteed for  $T_J=25^{\circ}C$ ,  $V_{CM}=0$ ,  $V_{O}=0$  and  $\pm 15V$  supplies unless otherwise specified. **Boldface limits apply at operating temperature extremes.** 

Parameter	Conditions		LM627AC LM637AC		LM627BC LM637BC		LM627C LM637C		
		Тур	Tested Limit (Note 5)	Design Limit (Note 6)	Tested Limit (Note 5)	Design Limit (Note 6)	Tested Limit (Note 5)	Design Limit (Note 6)	Units
Input Offset Voltage	(Note 2)	15	25 <b>50</b>		50 110		100	210	μV Max
Input Offset Voltage Drift	(Note 3)	0.2	0.6		1.0		_	1.8	μV/°C <b>M</b> ax
Input Offset Voltage Long Term Stability	(Note 4)	0.2							μV/mo Max
Input Bias Current		3	10	20	10	20	25	50	nA Max
Input Offset Current		2	10	20	10	20	25	50	nA Max
Input Noise Voltage	0.1 to 10 Hz	0.08		0.18		0.18		0.25	μV p-p Max
Input Voltage Noise Density	f = 10 Hz f = 30 Hz f = 1 kHz	3.5 3.1 3.0	5.5 4.5 3.8		5.5 4.5 3.8			8.0 5.6 4.5	nV/√Hz Max
Input Noise Current Density	f = 10 Hz f = 30 Hz f = 1 kHz	1.7 1.0 0.4							pA/√Hz Max
Input Resistance	Common Mode	20							GΩ
Input Voltage Range		±12	±11	± 10.3	±11	± 10.3	±11	± 10.3	V Min
Common-Mode Rejection Ratio	V <sub>CM</sub> = ±11V V <sub>CM</sub> = ±10V	140	126	120	126	120	120	116	dB Min
Power Supply Rejection Ratio	$V_S = \pm 4V \text{ to } \pm 18V$ $V_S = \pm 4.5V \text{ to } \pm 18V$	140	120	117	120	117	110	108	dB Min

## **Electrical Characteristics (Continued)**

Parameter	Conditions		LM627AC LM637AC		LM627BC LM637BC		LM627C LM637C		
		Тур	Tested Limit (Note 5)	Design Limit (Note 6)	Tested Limit (Note 5)	Design Limit (Note 6)	Tested Limit (Note 5)	Design Limit (Note 6)	Units
Large-Signal Voltage Gain	$V_O = \pm 12V$ $R_L \ge 2 k\Omega$ $V_O = \pm 10V$ $R_L \ge 1 k\Omega$ $R_L \ge 600\Omega$	7000 6000	5000 4000 3000	3000 2500 2000	5000 3500 2000	3000 2000 1500	4000 2500 1500	2500 1500 1000	V/mV Min
Output Voltage Swing	$R_L \ge 2 k\Omega$ $R_L \ge 600\Omega$	±13.8 ±12.5	±13 ±11	± 12.5 ± 10.5	±13 ±11	± 12.5 ± 10.5	± 13 ± 10.5	± 12.5 ± 10	V Min
Slew Rate	LM627 LM637 R <sub>L</sub> = 2k	4.5 14		3 10		3 10		3 10	V/μs Min
Gain-Bandwidth Product	LM627 LM637 f = 10 kHz	14 65		10 45		10 45		10 45	MHz Min
Output Resistance	Open Loop	50							Ω
Supply Current		3	4.5	5	4.5	5	4.8	5.2	mA Max
Offset Adjust Range	R <sub>P</sub> ≥ 10 kΩ	±2							m∨

Note 1: Absolute maximum ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but do not guarantee specific performance limits. For guaranteed specifications and test conditions, see the Electrical Characteristics. The guaranteed specifications apply only for the test conditions listed.

Note 2: Input offset voltage for A and B grades is tested and guaranteed with the device fully warmed up. See Figure 1 in the Application Hints for test circuit. Warmup drift is typically 5 µV settling out in 5 minutes. The LM627C/LM637C offset voltage is measured by automated test equipment within 200 ms of applying power.

Note 3: Input offset voltage drift is defined as  $(V_{OS}(85^{\circ}C) - V_{OS}(-25^{\circ}C))/110^{\circ}C$  for the industrial temperature range. For the military temperature range, the input offset voltage drift is measured from room temperature to both extremes: both  $(V_{OS}(25^{\circ}C) - V_{OS}(-55^{\circ}C))/80^{\circ}C$  and  $(V_{OS}(125^{\circ}C) - V_{OS}(25^{\circ}C))/100^{\circ}C$ .

Note 4: Input offset voltage long term stability refers to the average trend line of V<sub>OS</sub> vs. time over extended periods of time after the first 30 days of operation. Excluding the initial hour of operation, changes in V<sub>OS</sub> during the first 30 days are typically 2 μV.

Note 5: Guaranteed and 100% production tested. These limits are used to calculate outgoing quality levels.

Note 6: Guaranteed but not 100% production tested. These limits are not used to calculate outgoing quality levels.

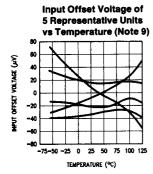
Note 7: Inputs are protected by back-to-back diodes to prevent zener breakdown of the input transistors. Series limiting resistors have not been included since they degrade noise performance. Excessive current may flow if a differential voltage in excess of 0.7V is applied.

Note 8: For operation above 25°C, the maximum power dissipation specification must be derated. Typical junction-to-ambient thermal resistance of the molded DIP is 105°C/W. The metal can package has a typical junction-to-ambient thermal resistance of 15°C/W and a typical junction-to-case thermal resistance of 17°C/W.

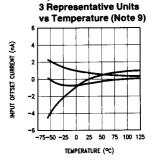
Note 9: These units selected to illustrate the types of variations that may be encountered. (This note refers to particular curves within the Typical Performance Characteristics.)

## Typical DC Performance Characteristics (LM627, LM637)

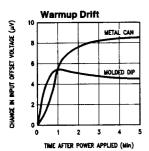
 $V_S = \pm 15V$ ,  $T_A = 25$ °C,  $R_L = 2k$  unless othewise indicated.

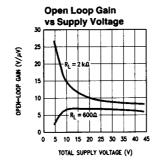


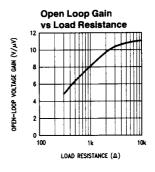
**Input Bias Current of** 

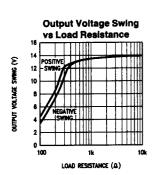


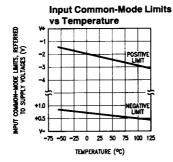
Input Offset Current of

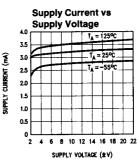








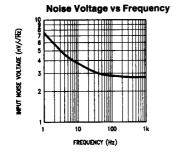


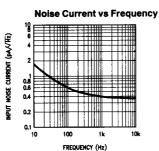


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# Typical Noise Characteristics (LM627, LM637)

 $V_S = \pm 15V, T_A = 25^{\circ}C$ 



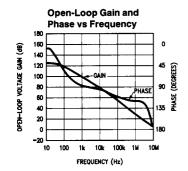


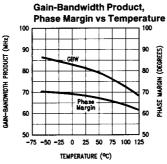
TL/H/9212-10

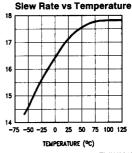
#### Typical AC Performance Characteristics (LM627) $V_S = \pm 15 V, T_A = 25^{\circ}C, R_L = 2k$ Open-Loop Gain and Gain-Bandwidth Product, Phase Margin vs Temperature Slew Rate vs Temperature Phase vs Frequency 180 20 160 GAIN-BANDWIDTH PRODUCT (MHz) OPEN-LOOP VOLTAGE GAIN (dB) 5.5 140 18 MARGIN (DEGREES) SLEW RATE (V/µs) 120 PHASE (DEGREES) 5.0 100 16 80 4.5 Phase Maroin 60 4.0 40 20 12 3.5 0 -20 -75 -50 **-**25 0 25 50 75 100 125 10k -75 -50 -25 0 25 50 75 100 125 10 100 1k TEMPERATURE (°C) FREQUENCY (Hz) TEMPERATURE (°C) Common-Mode Rejection **Power Supply Rejection Maximum Output Swing** vs Frequency Ratio vs Frequency Ratio vs Frequency 160 140 POWER SUPPLY REJECTION RATIO (dB) COMMON-MODE REJECTION RATIO (4B) 140 120 PEAK-TO-PEAK AMPLITUDE (Y) 120 100 NEGATIVE 20 100 80 POSITIVE 80 15 60 10 40 20 20 1k 10k 100k 100 100k 100 FREQUENCY (Hz) FREQUENCY (Hz) FREQUENCY (Hz) TL/H/9212-11 Large-Signal Pulse **Small-Signal Pulse** Response, $A_V = +1$ Response, $A_V = +1$ +50 mV+ 10<math>V0 0 -100-50 mV5 A S TIME (5 µs/DIV) TIME (500 ns/DIV) TL/H/9212-13 TL/H/9212-12

# Typical AC Performance Characteristics (LM637)

 $V_S = \pm 15V$ ,  $T_A = 25$ °C,  $R_L = 2k$ 

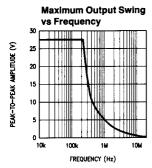




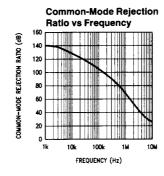


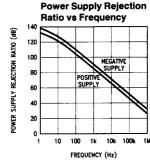
SLEW RATE (V/µs)

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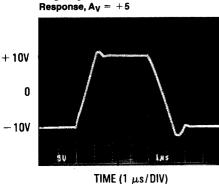


Large-Signal Pulse

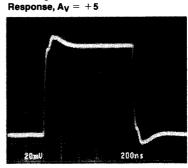




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Small-Signal Pulse

TIME (200 ns/DIV)

TL/H/9212-17

TL/H/9212-16

## **Application Hints**

#### **OFFSET VOLTAGE**

Offset voltage of the LM627/637 is internally trimmed to a very low value. The data sheet VOS specification applies at  $T_J = 25$ °C,  $V_{CM} = 0$  and  $\pm 15$ V supplies. For other temperatures, common-mode voltages, and supply voltages, temperature drift, common-mode rejection and power-supply rejection errors must be taken into account.

Since the LM627/LM637C offset voltage is measured within 200 ms of applying power, the 5 µV typical warmup drift is not accounted for in the measurement. Fortunately, the warmup drift is a small fraction of its 100 µV max offset. For the 25 µV A and 50µV B grades, the offset voltage is measured with the circuit of Figure 1 approximately 5 minutes after applying power.

To measure  $V_{OS}$  with high accuracy,  $V_{OS}$  must be amplified right at the device as shown; otherwise the offset voltage can be obscured by noise and thermoelectric voltages. Thermocouples occur in the devices, the IC socket and the resistor across the device inputs (R2), all of which must be held isothermal. Usually best results are obtained by placing the circuit in a box or chamber to minimize airflow and employing a long thermal soak time. R2 should be mounted symmetrically with respect to potential thermal gradients: e.g. not perpendicular to the board but instead parallel to the board and the device socket. In addition, R2 should have low thermal EMF. Cermet or nichrome metal film types are acceptable; avoid tin-oxide resistors.

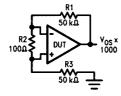
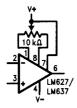


FIGURE 1. Offset Voltage Test Circuit

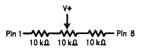
### **OFFSET NULLING**

This is usually not required on the LM627/637 family since its offset voltage is internally trimmed. An offset adjust range of approximately ±2 mV is available using a single 10 or 20 k $\Omega$  potentiometer as shown in Figure 2. With these values, the adjustment is relatively linear over the entire range. If a 100 k $\Omega$  potentiomer is used, the adjustment becomes very coarse at the extremes (above 700 µV) but fine in the center, which makes it easier to precisely null the offset. For even more sensitivity, employ a pot in conjunction with two fixed resistors. The circuit of Figure 3, which uses this technique, has an adjustment range of  $\pm 200 \mu V$ . Because adjusting the offset voltage of an LM627/637 will alter its offset voltage temperature drift, caution is advised.



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FIGURE 2. Offset Adjust Circuit



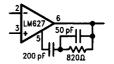
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FIGURE 3. Improved Sensitivity Offset Adjust

Every 100μV of offset will produce a 0.33 μV/°C drift component. For this reason the offset adjust potentiometer should not be used to null out a sensor offset if system temperature drift is important; rather a stable voltage reference must be added to the sensor voltage. Offset voltage drift is guaranteed by design for the LM627C either with or without external nulling. The higher precision A and B grades are 100% drift tested and guaranteed without nulling only.

## **OVERCOMPENSATION**

Without any external compensation, the LM627 is stable at unity gain and up to 500 pF load capacitance. It has a slew rate of 4.5 V/µs and a gain-bandwidth product of 14 MHz. If desired, the amplifier may be overcompensated by adding external components as shown in Figure 4. This increases maximum capacitive loading to 2000 pF while decreasing slew rate to 1.5 V/µs and bandwidth to 1.5 MHz. If overcompensation of the LM627 (or the LM637) is not desired, pin 5 should be left open.

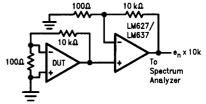


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FIGURE 4. Overcompensation

### NOISE

When measuring spot noise voltage, a circuit as shown in Figure 5 is recommended. The DUT running at a gain of 100 will not roll off until approximately 140 kHz. Adding the second gain of 100 amplifier brings total DUT-input-referred gain up to 10,000, which minimizes to minimize sensitivity to EMI in the environment. When measuring spot noise at 30 Hz, it is recommended that the spectrum analyzer bandwidth be 20 Hz or less to minimize pickup at line frequency.



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**FIGURE 5. Spot Noise Test Circuit** 

# **Application Hints** (Continued)

The circuit used to measure peak-to-peak noise voltage in the 0.1 to 10 Hz range is shown in Figure 6. The device should be warmed up for about 2 minutes and shielded from air currents to minimize warmup drift and thermoelectric voltages. The test time should be limited to only 10 seconds, as this limits noise contributions below 0.1 Hz, as does the single zero rolloff. The measuring equipment must be flat down to 0.1 Hz. DC coupling must be employed to ensure this. Certain types of X-Y plotters may not be usable because of severe rolloff above a few Hz.

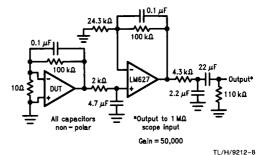


FIGURE 6. 0.1 Hz to 10 Hz Noise Test Circuit