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TPS3840 Nano Power, High Input Voltage Supervisor With MR and Programmable Delay

Technical [Documents](http://www.ti.com/product/TPS3840?dcmp=dsproject&hqs=td&#doctype2)

1 Features

- Wide operating voltage: 1.5 V to 10 V
- Nano supply current: 300 nA (Typ), 700 nA (Max)
- Fixed threshold voltage (V_{1T-})
	- Threshold from 1.6 V to 4.9 V in 0.1-V steps
	- High accuracy: 1% (Typ), 1.5% (Max)
	- $-$ Built-in hysteresis (V_{1T+})
		- 1.6 V < V_{IT} ≤ 3.0 V = 100 mV (Typical)
		- $-$ 3.1 V ≤ V_{IT-} < 4.9 V = 200 mV (Typical)
- Start-up delay (t_{STRT}) : 220 µs (Typ), 350 µs (Max)
- Programmable reset time delay (t_D) :
	- $-$ 50 µs (no capacitor) to 6.2 s (10-µF)
- Active-low manual reset (MR)
- Three output topologies:
	- TPS3840DL: open-drain, active-low (RESET), requires pull-up resistor
	- TPS3840PL: push-pull, active-low (RESET)
	- TPS3840PH: push-pull, active-high (RESET)
- Wide temperature range: -40° C to $+125^{\circ}$ C
- Package: SOT23-5 (DBV)

2 Applications

- Grid infrastructure: circuit breaker, smart meter, other monitoring and protection equipment
- Factory automation: field transmitter, PLC.
- • Building automation: fire safety, smoke detector, and HVAC
- Electronic point of sale
- Portable, battery-powered systems

TPS3840DL49 VDD GND MR RESET CT 9V μ F $\frac{1}{1}$ **DC/DC** Vin Vout EN 5V

3 Description

Tools & [Software](http://www.ti.com/product/TPS3840?dcmp=dsproject&hqs=sw&#desKit)

Wide Vin allows monitoring 9V rails or batteries without external components and 24V rails with external resistors. Nano-Iq extends battery life for low power applications and minimizes current consumption when using external resistors. Fast start-up delay allows the detection of a voltage fault before the rest of the system powers up providing maximum safety in hazardous start-up fault conditions. Low Power-on-Reset (V_{POR}) prevents false resets, premature enable or turn-on of next device, and proper transistor control during power-up and power-down.

Support & **[Community](http://www.ti.com/product/TPS3840?dcmp=dsproject&hqs=support&#community)**

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Reset output signal is asserted when the voltage at V_{DD} drops below the negative voltage threshold (V_{IT}) or when manual reset (MR) is pulled to a low logic $(V_{\overline{MR}})$. Reset signal is cleared when V_{DD} rise above V_{1T-} plus hysteresis (V_{1T+}) and manual reset is floating or above $V_{\overline{MR} H}$ and the reset time delay (t_D) expires. Reset time delay can be programmed by connecting a capacitor between CT pin and ground. For a fast reset CT pin can be left floating.

Additional features: Built-in glitch immunity protection for MR and V_{DD} , built-in hysteresis, low open-drain output leakage current $(I_{LKG(OD)})$.

Device Information[\(1\)](#page-0-0)

(1) For package details, see the mechanical drawing addendum at the end of the data sheet.

Typical Application Circuit TPS3840 Typical Supply Current

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4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

5 Device Comparison

[Figure 1](#page-2-3) shows the device nomenclature to determine the device variant. Other voltages from [Table 3](#page-28-2) at the end of datasheet can be sample upon request, please contact TI sales representative for details.

Figure 1. Device Nomenclature

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6 Pin Configuration and Functions

Pin Functions

7 Specifications

7.1 Absolute Maximum Ratings

over operating free-air temperature range, unless otherwise noted⁽¹⁾

(1) Stresses beyond those listed under *Absolute Maximum Rating* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Condition*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

If the logic signal driving MR is less than V_{DD} , then additional current flows into V_{DD} and out of MR. V_{MR} should not be higher than V_{DD} .

(3) As a result of the low dissipated power in this device, it is assumed that $T_{\rm J} = T_{\rm A}$.

7.2 ESD Ratings

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

7.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

(1) If the logic signal driving MR is less than V_{DD} , then additional current flows into V_{DD} and out of MR. V_{MR} should not be higher than V_{DD}

7.4 Thermal Information

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](http://www.ti.com/lit/SPRA953) application report.

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7.5 Electrical Characteristics

At 1.5 V ≤ V_{DD} ≤ 10 V, CT = MR = Open, RESET pull-up resistor (R_{pull-up}) = 100 kΩ to VDD, output reset load (C_{LOAD}) = 10 pF and over the operating free-air temperature range – 40°C to 125°C, unless otherwise noted. Typical values are at T $_J$ = 25°C.

(1) V_{IT} threshold voltage range from 1.6 V to 4.9 V in 100 mV steps, for released versions see Device Voltage Thresholds table.

(2) V_{IT+} = V_{HYS} + V_{IT-}
(3) If the logic signal driving MR is less than VDD, then additional current flows into VDD and out of MR

(4) V_{POR} is the minimum V_{DD} voltage level for a controlled output state. V_{DD} slew rate ≤ 100mV/µs

7.6 Timing Requirements

At 1.5 V ≤ V_{DD} ≤ 10 V, CT = MR = Open, RESET pull-up resistor (R_{pull-up}) = 100 kΩ to VDD, output reset load (C_{LOAD}) = 10 pF and over the operating free-air temperature range – 40°C to 125°C, VDD slew rate < 100mV / us, unless otherwise noted. Typical values are at $T_1 = 25^{\circ}C$.

(1) When VDD starts from less than the specified minimum V_{DD} and then exceeds V_{IT+}, reset is release after the startup delay (t_{STRT}), a capacitor at CT pin will add t_D delay to t_{STRT} time

(2) t_{P. HL} measured from threhold trip point (V_{IT-}) to V_{OL} for active low variants and V_{OH} for active high variants.
(3) The MIN and MAX reset time delay with external capacitor depends on R_{CT} and is calculated The MIN and MAX reset time delay with external capacitor depends on R_{CT} and is calculated using Equation 5 and Equation 6 in Section 8.3.2

(4) Overdrive % = $[(V_{DD}/V_{IT.}) - 1] \times 100\%$

(1) t_{D (no cap)} is included in t_{STRT} time delay. If t_D delay is programmed by an external capacitor connected to CT pin then t_D programmed time will be added to the startup time, VDD slew rate = 100 mV / μ s.

(2) Open-Drain timing diagram assumes pull-up resistor is connected to RESET

(3) RESET output is undefined when VDD is $<$ V_{POR}

Figure 4. Timing Diagram TPS3840DL (Open-Drain Active-Low)

(4) t_D (no cap) is included in t_{STRT} time delay. If t_D delay is programmed by an external capacitor connected to CT pin, then t_D programmed time will be added to the startup time. VDD slew rate = 100 mV / μ s.

(5) RESET output is undefined when VDD < V_{POR} and limited to V_{OL} for VDD slew rate = 100 mV / μs

(6) t_D (no cap) is included in t_{STRT} time delay. If t_D delay is programmed by an external capacitor connected to CT pin, then t_D programmed time will be added to the total startup time. VDD slew rate = 100 mV / μ s.

Figure 6. Timing Diagram TPS3840PH (Push-Pull Active-High)

7.7 Typical Characteristics

Typical characteristics show the typical performance of the TPS3840 device. Test conditions are $T_J = 25^{\circ}C$, $V_{DD} = 3.3 V$, R_{pull} . $_{up}$ = 100 kΩ, C_{Load} = 50 pF, unless otherwise noted.

Typical characteristics show the typical performance of the TPS3840 device. Test conditions are $T_J = 25^{\circ}C$, $V_{DD} = 3.3 V$, R_{pull-} $_{up}$ = 100 kΩ, C_{Load} = 50 pF, unless otherwise noted.

Typical characteristics show the typical performance of the TPS3840 device. Test conditions are $T_J = 25^{\circ}C$, $V_{DD} = 3.3 V$, R_{pull-} $_{up}$ = 100 kΩ, C_{Load} = 50 pF, unless otherwise noted.

Typical characteristics show the typical performance of the TPS3840 device. Test conditions are $T_J = 25^{\circ}C$, $V_{DD} = 3.3 V$, R_{pull} . $_{up}$ = 100 kΩ, C_{Load} = 50 pF, unless otherwise noted.

Typical characteristics show the typical performance of the TPS3840 device. Test conditions are $T_J = 25^{\circ}C$, $V_{DD} = 3.3 V$, R_{pull} . $_{up}$ = 100 kΩ, C_{Load} = 50 pF, unless otherwise noted.

Typical characteristics show the typical performance of the TPS3840 device. Test conditions are $T_J = 25^{\circ}C$, $V_{DD} = 3.3 V$, R_{pull} $_{\text{up}}$ = 100 kΩ, C_{Load} = 50 pF, unless otherwise noted.

Typical characteristics show the typical performance of the TPS3840 device. Test conditions are $T_J = 25^{\circ}C$, $V_{DD} = 3.3 V$, R_{pull-} $_{up}$ = 100 kΩ, C_{Load} = 50 pF, unless otherwise noted.

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8 Detailed Description

8.1 Overview

The TPS3840 is a family of wide VDD and nano-quiescent current voltage detectors with fixed threshold voltage. TPS3840 features include programable reset time delay using external capacitor, active-low manual reset, 1% typical monitor threshold accuracy with hysteresis and glitch immunity.

Fixed negative threshold voltages (V_{IT}) can be factory set from 1.6 V to 4.9 V (see the *[Device Comparison](#page-2-0)* for available options). TPS3840 is available in SOT-23 5 pin industry standard package.

8.2 Functional Block Diagram

8.3 Feature Description

8.3.1 Input Voltage (VDD)

VDD pin is monitored by the internal comparator to indicate when VDD falls below the fixed threshold voltage. VDD also functions as the supply for the internal bandgap, internal regulator, state machine, buffers and other control logic blocks. Good design practice involve placing a 0.1 uF to 1 uF bypass capacitor at VDD input for noisy applications to ensure enough charge is available for the device to power up correctly.

Feature Description (continued)

8.3.1.1 VDD Hysteresis

The internal comparator has built-in hysteresis to avoid erroneous output reset release. If the voltage at the VDD pin falls below V_{IT} the output reset is asserted. When the voltage at the VDD pin goes above V_{IT} plus hysteresis (V_{HYS}) the output reset is deasserted after t_D delay.

Figure 45. Hysteresis Diagram

8.3.1.2 VDD Transient Immunity

The TPS3840 is immune to quick voltage transients or excursion on VDD. Sensitivity to transients depends on both pulse duration and overdrive. Overdrive is defined by how much VDD deviates from the specified threshold. Threshold overdrive is calculated as a percent of the threshold in question, as shown in [Equation 1](#page-16-2).

Overdrive = $|(V_{\text{DD}}/V_{\text{IT}}-1) \times 100\%|$ (1)

Figure 46. Overdrive vs Pulse Duration

8.3.2 User-Programmable Reset Time Delay

The reset time delay can be set to a minimum value of 50 us by leaving the CT pin floating, or a maximum value of approximately 6.2 seconds by connecting 10 μ F delay capacitor. The reset time delay (t_D) can be programmed by connecting a capacitor no larger than 10 µF between CT pin and GND.

The relationship between external capacitor ($C_{CT\,EXT}$) in Farads at CT pin and the time delay (t_D) in seconds is given by [Equation 2](#page-16-3).

$$
t_D = -\ln(0.29) \times R_{CT} \times C_{CT_EXT} + t_D \text{ (no cap)}
$$
 (2)

[Equation 2](#page-16-3) is simplified to [Equation 3](#page-16-4) by plugging R_{CT} and t_{D(no cap)} given in *[Electrical Characteristics](#page-5-0)* section:

$$
t_{D} = 618937 \times C_{CT_EXT} + 50 \,\mu s \tag{3}
$$

[Equation 4](#page-16-5) solves for external capacitor value (C_{CT_EXT}) in units of Farads where t_D is in units of seconds

$$
C_{CT_EXT} = (t_D - 50 \text{ }\mu\text{s}) \div 618937 \tag{4}
$$

The reset delay varies according to three variables: the external capacitor variance (C_{CT}), CT pin internal resistance (R_{CT}) provided in the Electrical Characteristics table, and a constant. The minimum and maximum variance due to the constant is shown in [Equation 5](#page-16-6) and [Equation 6](#page-16-7).

Feature Description (continued)

The recommended maximum delay capacitor for the TPS3840 is limited to 10 µF as this ensures there is enough time for the capacitor to fully discharge when the reset condition occurs. When a voltage fault occurs, the previously charged up capacitor discharges, and if the monitored voltage returns from the fault condition before the delay capacitor discharges completely, the delay capacitor will begin charging from a voltage above zero volts and the reset delay will be shorter than expected. Larger delay capacitors can be used so long as the capacitor has enough time to fully discharge during the duration of the voltage fault.

8.3.3 Manual Reset (MR) Input

The manual reset (MR) input allows a processor GPIO or other logic circuits to initiate a reset. A logic low on MR with pulse duration longer than t_{MR_RES} will causes reset output to assert. After MR returns to a logic high (V_{MR_H}) and VDD is above V_{IT+} , reset is deasserted after the user programmed reset time delay (t_D) expires.

If \overline{MR} is not controlled externally, then \overline{MR} can be left disconnected. If the logic signal controlling \overline{MR} is less than VDD, then additional current flows from VDD into MR internally. For minimum current consumption, drive MR to either VDD or GND. V_{MR} should not be higher than VDD voltage.

Figure 47. Timing Diagram MR and RESET (TPS3840DL)

8.3.4 Output Logic

8.3.4.1 RESET Output, Active-Low

RESET (Active-Low) applies to TPS3840D**L** (Open-Drain) and TPS3840P**L** (Push-Pull) hence the "**L**" in the device name. RESET remains high (deasserted) as long as VDD is above the negative threshold (V_{II}) and the \overline{MR} pin is floating or above $V_{\overline{MR}}$. If VDD falls below the negative threshold (V_{IT-}) or if \overline{MR} is driven low, then RESET is asserted.

When \overline{MR} is again logic high or floating and VDD rise above V_{IT+} , the delay circuit will hold \overline{RESET} low for the specified reset time delay (t_D) . When the reset time delay has elapsed, the RESET pin goes back to logic high voltage (V_{OH}) .

The TPS3840DL (Open-Drain) version, denoted with "**D**" in the device name, requires a pull-up resistor to hold RESET pin high. Connect the pull-up resistor to the desired pull-up voltage source and RESET can be pulled up to any voltage up to 10 V independent of the VDD voltage. To ensure proper voltage levels, give some consideration when choosing the pull-up resistor values. The pull-up resistor value determines the actual V_{OL} , the output capacitive loading, and the output leakage current $(I_{LKG(OD)})$.

The Push-Pull variants (TPS3840**P**L and TPS3840**P**H), denoted with "**P**" in the device name, does not require a pull-up resistor

Feature Description (continued)

8.3.4.2 RESET Output, Active-High

RESET (active-high), denoted with no bar above the pin label, applies only to TPS3840PH push-pull active-high version. RESET remains low (deasserted) as long as VDD is above the threshold (V_{IT}) and the manual reset signal (MR) is logic high or floating. If VDD falls below the negative threshold (V_{IT}) or if MR is driven low, then RESET is asserted driving the RESET pin to high voltage (V_{OH}) .

When \overline{MR} is again logic high and VDD is above V_{IT+} the delay circuit will hold RESET high for the specified reset time delay (t_D). When the reset time delay has elapsed, the RESET pin goes back to low voltage (V_{OL})

8.4 Device Functional Modes

[Table 1](#page-18-1) summarizes the various functional modes of the device. Logic high is represented by "H" and logic low is represented by "L".

Table 1. Truth Table

(1) When V_{DD} falls below V_{DD(MIN)}, undervoltage-lockout (UVLO) takes effect and output reset is held asserted until V_{DD} falls below V_{POR}.

8.4.1 Normal Operation ($V_{DD} > V_{DD(min)}$)

When VDD is greater than $V_{DD(min)}$, the reset signal is determined by the voltage on the VDD pin with respect to the trip point (V_{IT}) and the logic state of \overline{MR} .

- MR high: the reset signal corresponds to VDD with respect to the threshold voltage.
- MR low: in this mode, the reset is asserted regardless of the threshold voltage.

8.4.2 VDD Between VPOR and V_{DD(min)}

When the voltage on VDD is less than the $V_{DD(min)}$ voltage, and greater than the power-on-reset voltage (V_{POR}), the reset signal is asserted.

8.4.3 Below Power-On-Reset (V_{DD} < V_{POR})

When the voltage on VDD is lower than V_{POR} , the device does not have enough bias voltage to internally pull the asserted output low or high and reset voltage level is undefined.

9 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

The following sections describe in detail how to properly use this device, depending on the requirements of the final application.

9.2 Typical Application

9.2.1 Design 1: Dual Rail Monitoring with Power-Up Sequencing

A typical application for the TPS3840 is voltage rail monitoring and power-up sequencing as shown in [Figure 48](#page-19-3). The TPS3840 can be used to monitor any rail above 1.6 V. In this design application, two TPS3840 devices monitor two separate voltage rails and sequences the rails upon power-up. The TPS3840PL30 is used to monitor the 3.3-V main power rail and the TPS3840DL16 is used to monitor the 1.8-V rail provided by the LDO for other system peripherals. The RESET output of the TPS3840PL30 is connected to the ENABLE input of the LDO. A reset event is initiated on either voltage supervisor when the VDD voltage is less than V_{IT} or when \overline{MR} is driven low by an external source.

Figure 48. TPS3840 Voltage Rail Monitor and Power-Up Sequencer Design Block Diagram

Typical Application (continued)

9.2.1.1 Design Requirements

This design requires voltage supervision on two separate rails: 3.3-V and 1.8-V rails. The voltage rail needs to sequence upon power up with the 3.3-V rail coming up first followed by the 1.8-V rail at least 25 ms after.

9.2.1.2 Detailed Design Procedure

The primary constraint for this application is choosing the correct device to monitor the supply voltage of the microprocessor. The TPS3840 can monitor any voltage between 1.6 V and 10 V and is available in 0.1 V increments. Depending on how far away from the nominal voltage rail the user wants the voltage supervisor to trigger determines the correct voltage supervisor variant to choose. In this example, the first TPS3840 triggers when the 3.3-V rail falls to 3.0 V. The second TPS3840 triggers a reset when the 1.8-V rail falls to 1.6 V. The secondary constraint for this application is the reset time delay that must be at least 25 ms to allow the microprocessor, and all other devices using the 3.3-V rail, enough time to startup correctly before the 1.8-V rail is enabled via the LDO. Because a minimum time is required, the user must account for capacitor tolerance. For applications with ambient temperatures ranging from -40° C to +125°C, C_{CT} can be calculated using R_{CT} and solving for C_{CT} in [Equation 2](#page-16-3). Solving Equation 2 for 25 ms gives a minimum capacitor value of 0.04 μ F which is rounded up to a standard value 0.047 μ F to account for capacitor tolerance.

A 1-µF decoupling capacitor is connected to the VDD pin as a good analog design practice. The pull-up resistor is only required for the Open-Drain device variants and is calculated to maintain the RESET current within the ±5 mA limit found in the *[Recommended Operating Conditions](#page-4-3)*: R_{Pull-up} = V_{Pull-up} ÷ 5 mA. For this design, a standard 10-kΩ pull-up resistor is selected to minimize current draw when RESET is asserted. Keep in mind the lower the pull-up resistor, the higher V_{OL} . The MR pin can be connected to an external signal if desired or left floating if not used due to the internal pull-up resistor to VDD.

9.2.1.3 Application Curves

9.2.2 Design 2: Battery Voltage and Temperature Monitor

A typical application for the TPS3840 is battery voltage and temperature monitoring. The TPS3840 is offered in active-low or active-high output topologies and can operate above or below the voltage threshold meaning the device can be used as an undervoltage monitor as shown in [Figure 50](#page-21-1) or overvoltage monitor as shown in [Figure 51.](#page-21-2) The TPS3840 can be used to monitor any rail above 1.6 V. In this design application, one TPS3840DL30 monitors the 3.3-V battery voltage rail and triggers an active-low reset fault condition if the battery voltage falls below the 3-V threshold. For overvoltage monitoring, another TPS3840DL30 monitors a 2.8-V battery and triggers a logic high at the 3-V threshold plus 100 mV hysteresis so at 3.1 V. Both applications monitor the battery temperature using TMP303, a push-pull, active-high temperature switch. A temperature fault is triggered if the battery temperature falls outside of a defined window temperature range set by the TMP303 variant chosen.

Figure 50. Low Battery Voltage and Window Temperature Monitoring Solution

9.2.2.1 Design Requirements

This design requires voltage and temperature supervision on a battery voltage rail and the requirements may differ depending on if undervoltage or overvoltage monitoring is required. For this design, both requirements are considered to show the flexibility of the TPS3840 device. The first application example shown in [Figure 50](#page-21-1) uses TPS3840DL30, an open-drain active-low voltage supervisor to monitoring undervoltage and TMP303, a push-pull active-high window temperature switch to monitor under and over temperature. For the undervoltage application, the TPS3840DL30 is operating in the inactive logic high region so an overvoltage fault occurs when the battery voltage falls below $V_{IT} = 3.0$ V or when the battery temperature is outside the range from 0°C to 60°C. The second application example uses TPS3840DL30 operating in the active-low region to monitor overvoltage and TMP303 to monitor under and over temperature. For the overvoltage requirement, the fault occurs when the battery voltage rises above 3.1 V or when the battery temperature is outside the range from 0°C to 60°C.

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9.2.2.2 Detailed Design Procedure

The primary constraint for this application is choosing the correct device to monitor the battery supply voltage. The TPS3840 can monitor any voltage between 1.6 V and 10 V and is available in 0.1 V increments. Depending on how far away from the nominal voltage rail the user wants the voltage supervisor to trigger determines the correct voltage supervisor variant to choose. In this design example, the TPS3840DL30 is chosen for both the undervoltage and overvoltage monitoring. For undervoltage monitoring, the undervoltage fault occurs when the 3.3-V rail falls to 3 V and for the overvoltage monitoring, the overvoltage fault occurs when the 2.8-V rail rises above the 3-V threshold (V_{IT}) plus 100mV hysteresis (V_{HYS}). It's important to note that in the undervoltage application, the TPS3840 RESET output is logic high during normal conditions whereas in the overvoltage application, the TPS3840 RESET output is logic low during normal conditions which is the reason a single device can be used for either type of monitoring depending on the logic required at the output. The opposite RESET output logic is offered in the push-pull, active-high device TPS3840PH noted with the RESET output. The secondary constraint for this application is the battery temperature monitoring accomplished by the TMP303A. Typical Lithium Ion battery discharge temperature range is 0°C to 60°C which is accomplished by the 'A' variant of TMP303A. The TMP303A triggers a fault to the MR pin of the TPS3840 or directly to the battery charger whenever the temperature is outside of the temperature range. The TMP303A offers 1°C resolution to meet the high resolution requirement. Because the undervoltage monitor design uses TMP303A, a push-pull active-high output device, an additional inverter is required before the MR pin because during normal operation, the TMP303 output is low but the MR pin must be logic high during normal operation. If using two TPS3840 devices for both undervoltage and overvoltage monitoring on the same battery, only one single temperature monitoring device is required. The last constraint is the RESET/RESET time delay set by C_{CT} . For applications with ambient temperatures ranging from –40°C to +125°C, C_{CT} can be calculated using R_{CT} and solving for C_{CT} in [Equation 2.](#page-16-3) By choosing a standard 10% capacitor value of 10 µF ensures the RESET/RESET time delay will be at least 6 seconds. Note: active-low devices use the output label RESET and active-high devices use the output label RESET.

A 0.1-µF decoupling capacitor is connected to the VDD pin as a good analog design practice. The pull-up resistor is only required for the Open-Drain device variants and is calculated to maintain the RESET current within the ±5 mA limit found in the *[Recommended Operating Conditions](#page-4-3)*: R_{Pull-up} = V_{Pull-up} ÷ 5 mA. For this design, a 1-MΩ pull-up resistor is selected to minimize current draw when RESET is asserted and to prevent the battery from unnecessary discharge. Keep in mind the lowering the pull-up resistor, increases V_{OL} and I_{OUT} . The \overline{MR} pin is used for a second fault condition provided by the temperature switch.

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9.2.3 Design 3: Fast Start Undervoltage Supervisor with Level-shifted Input

A typical application for the TPS3840 is a fast startup undervoltage supervisor that operates with an input power supply higher than the recommended maximum of 10 V through the use of a resistor divider at the input as shown in [Figure 52.](#page-23-0) The TPS3840 can be used to monitor any rail above 1.6 V and only requires maximum 350 µs upon startup before the device can begin monitoring a voltage. In this design application, a TPS3840 monitors a 12-V rail and triggers a reset fault condition if the voltage rail voltage drops below 10 V using a TPS3840 device with V_{1T} of 4.9 V. This design also accounts for a wide input range in the case the 12-V rail rises higher, the resistor divider is set so that the voltage at the VDD pin never exceeds 10 V. The resistor values must not be so large that the external resistor divider affects the accuracy or operation of the device. TPS3840 is available in both active-low and active-high topologies providing the flexibility to monitor undervoltage or overvoltage with either output logic. This design uses the active-low, open-drain TPS3840DL49 variant so that when the undervoltage condition occurs, that is when the voltage at VDD pin falls below the voltage threshold set by the external resistor divider, the output transitions to logic-low and can be used to flag an undervoltage condition or used to connect to the ENABLE of the next device to shut it off as a logic low on an ENABLE pin typically disables the device. In this design, the output of the TPS3840 simply connects to a MCU to flag an undervoltage condition.

Figure 52. Fast Start Undervoltage Supervisor with Level-shifted Input

9.2.3.1 Design Requirements

This design requires voltage supervision on a 12-V power supply voltage rail with possibility of the 12-V rail rising up as high as 18 V. The undervoltage fault occurs when the power supply voltage drops below 10 V.

9.2.3.2 Detailed Design Procedure

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The primary constraint for this application is monitoring a 12-V rail while preventing the VDD pin on TPS3840 from exceeding the recommended maximum of 10 V. This is accomplished by sizing the resistor divider so that when the 12-V rail drops to 10 V, the VDD pin for TPS3840 will be at 4.9 V which is the V_{IT} threshold for triggering a undervoltage condition for TPS3840DL49 as shown in [Equation 7.](#page-24-0)

$$
V_{\text{tail_trigger}} = V_{IT} \times (R_{\text{top}} + R_{\text{bottom}}) \div R_{\text{bottom}} \tag{7}
$$

where V_{rail_trigger} is the trigger voltage of the rail being monitored, V_{IT-} is the falling threshold on the VDD pin of TPS3840, and R_{top} and R_{bottom} are the top and bottom resistors of the external resistor divider. Be sure to size the resistor values such that the current through the external resistor divider is much greater than IDD to preserve voltage monitoring accuracy. V_{IT} is fixed per device variant and is 4.9 V for TPS3840DL49. Substituting in the values from [Figure 52,](#page-23-0) the undervoltage trigger threshold for the rail is set to 10.045 V.

Since the undervoltage trigger of 10 V on the rail corresponds to 4.9 V undervoltage threshold trigger of the TPS3840 device, there is plenty of room for the rail to rise up while maintaining less than 10 V on the VDD pin of the TPS3840. [Equation 8](#page-24-1) shows the maximum rail voltage that still meets the 10 V maximum at the VDD pin for TPS3840.

$$
V_{\text{tail_max}} = 10 \times (10,500 + 10,000) \div 10,000 = 20.5 \text{ V}
$$
\n(8)

This means the monitored voltage rail can go as high as 20.5 V and still not violate the recommended maximum for the VDD pin on TPS3840. This is useful when monitoring a voltage rail that has a wide range that may go much higher than the nominal rail voltage such as in this case with the specification that the 12-V rail can go as high as 18 V. Notice that the resistor values chosen are less than 100kΩ to preserve the accuracy set by the internal resistor divider. Good design practice recommends using a 0.1-µF capacitor on the VDD pin and this capacitance may need to increase when using an external resistor divider.

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9.2.4 Design 4: Voltage Monitor with Back-up Battery Switchover

A typical application for the TPS3840 is to monitor a voltage rail and switch the power to a back-up battery if the main supply is in undervoltage condition. Because systems that utilize a back-up battery tend to require low quiescent current, TPS3840 serves as the perfect solution as this device only requires 350 nA typically. The TPS3840 monitors the main power rail via the VDD pin and when the main power rail falls, the RESET output asserts causing a switch to close on the back-up battery rail. The diodes provide an ORing logic function to prevent reverse leakage and to allow either rail to connect to the output depending on the status of the main voltage rail.

Figure 53. Voltage Monitor with Back-up Battery Switchover Solution

9.2.4.1 Design Requirements

This design requires voltage supervision on a 5-V main supply voltage rail and when the main rail fails, switch to a back-up battery supply to prevent complete power loss in the system. The System Output must remain above 1.8 V even when the main supply completely fails. The design requires less than 500 nA of total current consumption and must prevent battery leakage when the battery is not being used. When the system is using the back-up battery and the main supply voltage rail comes back up, the system must switch back to the main power supply in less than 100 us to save battery power.

9.2.4.2 Detailed Design Procedure

The primary constraints for this application are choosing the correct device variant for the monitored voltage and deciding the preferred solution to switch the back-up battery in and out of the system. For this design, the TPS3840PL30 provides an active-low, push-pull output topology that turns on the PFET when the 5-V rail monitored by VDD drops to 3.0 V. The diodes logically OR the power supply with the back-up battery and prevents reverse current leakage. Using this solution, the System Output remains above 1.8 V in all circumstances unless both the 5-V rail and back-up battery fail. The System Output voltage will follow the 5-V rail minus a diode drop until the 5-V rail drops to 3 V then the back-up battery switches into the system providing 3.3 V minus a diode drop to the System Output. When the 5-V rail comes back above 3.1 V accounting for hysteresis, the PFET turns off to disconnect the back-up battery from the system. Since this design disconnects the battery when not being used, this solution maximizes battery life.

9.2.5 Application Curve: TPS3840EVM

These application curves are taken with the *[TPS3840EVM](http://www.ti.com/tool/TPS3840EVM)*. Please see the *[TPS3840EVM User Guide](http://www.ti.com/lit/pdf/SNVU611)* for more information.

Figure 56. TPS3840EVM RESET Time Delay (t_D) with 1-µF Capacitor

10 Power Supply Recommendations

These devices are designed to operate from an input supply with a voltage range between 1.5 V and 10 V. TI recommends an input supply capacitor between the VDD pin and GND pin. This device has a 12-V absolute maximum rating on the VDD pin. If the voltage supply providing power to VDD is susceptible to any large voltage transient that can exceed 12 V, additional precautions must be taken.

11 Layout

11.1 Layout Guidelines

Make sure that the connection to the VDD pin is low impedance. Good analog design practice recommends placing a minimum 0.1-µF ceramic capacitor as near as possible to the VDD pin. If a capacitor is not connected to the CT pin, then minimize parasitic capacitance on this pin so the rest time delay is not adversely affected.

- Make sure that the connection to the VDD pin is low impedance. Good analog design practice is to place a >0.1-µF ceramic capacitor as near as possible to the VDD pin.
- If a C_{CT} capacitor is used, place these components as close as possible to the CT pin. If the CT pin is left unconnected, make sure to minimize the amount of parasitic capacitance on the pin to <5 pF.
- Place the pull-up resistors on RESET pin as close to the pin as possible.
- For V_{DD} slew rate >100mV/ μ s, increase input capacitor and pull-up resistor for OD variants.

11.2 Layout Example

The layout example in shows how the TPS3840 is laid out on a printed circuit board (PCB) with a user-defined delay.

Vias used to connect pins for application-specific connections

Figure 57. TPS3840 Recommended Layout

12 Device and Documentation Support

12.1 Device Nomenclature

[Table 2](#page-28-3) shows how to decode the function of the device based on its part number

Table 2. Device Naming Convention

[Table 3](#page-28-2) shows the possible variants of the TPS3840. Contact Texas Instruments for details and availability of other options shown; minimum order quantities apply.

Table 3. Device Threshold

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Table 3. Device Threshold (continued)

12.2 Support Resources

[TI E2E™ support forums](http://e2e.ti.com) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](http://www.ti.com/corp/docs/legal/termsofuse.shtml).

12.3 Trademarks

E2E is a trademark of Texas Instruments.

All other trademarks are the property of their respective owners.

12.4 Electrostatic Discharge Caution

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

12.5 Glossary

[SLYZ022](http://www.ti.com/lit/pdf/SLYZ022) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures. "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the ≤ 1000 ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE OPTION ADDENDUM

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

OTHER QUALIFIED VERSIONS OF TPS3840 :

• Automotive : [TPS3840-Q1](http://focus.ti.com/docs/prod/folders/print/tps3840-q1.html)

NOTE: Qualified Version Definitions:

• Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

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TEXAS

TAPE AND REEL INFORMATION

ISTRUMENTS

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

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PACKAGE OUTLINE

DBV0005A SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR

NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. Refernce JEDEC MO-178.
- 4. Body dimensions do not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.25 mm per side.
- 5. Support pin may differ or may not be present.

EXAMPLE BOARD LAYOUT

DBV0005A SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

DBV0005A SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

9. Board assembly site may have different recommendations for stencil design.

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