January 2014



FNA40860

Motion SPM® 45 Series

Features

- UL Certified No. E209204 (UL1557)
- 600 V 8 A 3-Phase IGBT Inverter with Integral Gate Drivers and Protection
- · Low Thermal Resistance Using Ceramic Substrate
- · Low-Loss, Short-Circuit Rated IGBTs
- Built-In Bootstrap Diodes and Dedicated Vs Pins Simplify PCB Layout
- · Built-In NTC Thermistor for Temperature Monitoring
- Separate Open-Emitter Pins from Low-Side IGBTs for Three-Phase Current Sensing
- Single-Grounded Power Supply
- · Optimized for 5 kHz Switching Frequency
- Isolation Rating: 2000 V_{rms} / min.

Applications

· Motion Control - Home Appliance / Industrial Motor

Related Resources

- AN-9070 Motion SPM® 45 Series Users Guide
- AN-9071 Motion SPM® 45 Series Thermal Performance Information
- AN-9072 Motion SPM® 45 Series Mounting Guidance
- RD-344 Reference Design (Three Shunt Solution)
- RD-345 Reference Design (One Shunt Solution)

General Description

FNA40860 is a Motion SPM® 45 module providing a fully-featured, high-performance inverter output stage for AC Induction, BLDC, and PMSM motors. These modules integrate optimized gate drive of the built-in IGBTs to minimize EMI and losses, while also providing multiple on-module protection features including under-voltage over-current shutdown, lockouts. monitoring, and fault reporting. The built-in, high-speed HVIC requires only a single supply voltage and translates the incoming logic-level gate inputs to the high-voltage, high-current drive signals required to properly drive the module's robust short-circuit-rated IGBTs. Separate negative IGBT terminals are available for each phase to support the widest variety of control algorithms.

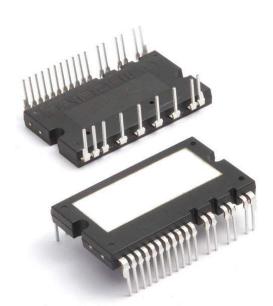


Figure 1. Package Overview

Package Marking and Ordering Information

Device	Device Marking	Package	Packing Type	Quantity
FNA40860	FNA40860	SPMAA-A26	Rail	12

Integrated Power Functions

• 600 V - 8 A IGBT inverter for three-phase DC / AC power conversion (please refer to Figure 3)

Integrated Drive, Protection, and System Control Functions

- For inverter high-side IGBTs: gate drive circuit, high-voltage isolated high-speed level shifting control circuit Under-Voltage Lock-Out (UVLO) protection
- For inverter low-side IGBTs: gate drive circuit, Short-Circuit Protection (SCP)
 control supply circuit Under-Voltage Lock-Out (UVLO) protection
- · Fault signaling: corresponding to UVLO (low-side supply) and SC faults
- Input interface: active-HIGH interface, works with 3.3 / 5 V logic, Schmitt trigger input

Pin Configuration

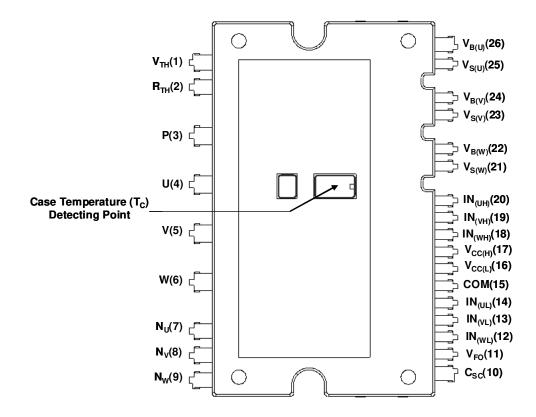


Figure 2. Top View

Pin Descriptions

Pin Number	Pin Name	Pin Description
1	V _{TH}	Thermistor Bias Voltage
2	R _{TH}	Series Resistor for the Use of Thermistor (Temperature Detection)
3	Р	Positive DC-Link Input
4	U	Output for U-Phase
5	V	Output for V-Phase
6	W	Output for W-Phase
7	N _U	Negative DC-Link Input for U-Phase
8	N _V	Negative DC-Link Input for V-Phase
9	N _W	Negative DC-Link Input for W-Phase
10	C _{SC}	Capacitor (Low-Pass Filter) for Short-circuit Current Detection Input
11	V _{FO}	Fault Output
12	IN _(WL)	Signal Input for Low-Side W-Phase
13	IN _(VL)	Signal Input for Low-Side V-Phase
14	IN _(UL)	Signal Input for Low-Side U-Phase
15	COM	Common Supply Ground
16	V _{CC(L)}	Low-Side Common Bias Voltage for IC and IGBTs Driving
17	V _{CC(H)}	High-Side Common Bias Voltage for IC and IGBTs Driving
18	IN _(WH)	Signal Input for High-Side W-Phase
19	IN _(VH)	Signal Input for High-Side V-Phase
20	IN _(UH)	Signal Input for High-Side U-Phase
21	V _{S(W)}	High-Side Bias Voltage Ground for W-Phase IGBT Driving
22	V _{B(W)}	High-Side Bias Voltage for W-Phase IGBT Driving
23	V _{S(V)}	High-Side Bias Voltage Ground for V-Phase IGBT Driving
24	V _{B(V)}	High-Side Bias Voltage for V-Phase IGBT Driving
25	V _{S(U)}	High-Side Bias Voltage Ground for U-Phase IGBT Driving
26	V _{B(U)}	High-Side Bias Voltage for U-Phase IGBT Driving

Internal Equivalent Circuit and Input/Output Pins

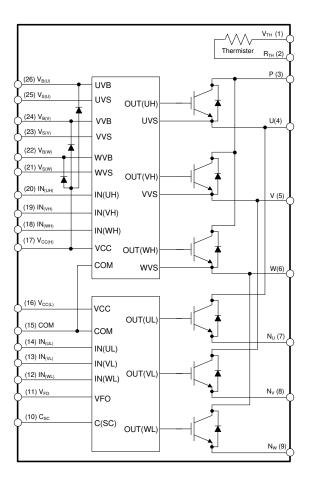


Figure 3. Internal Block Diagram

1st Notes:

- 1. Inverter high-side is composed of three IGBTs, freewheeling diodes, and one control IC for each IGBT.
- 2. Inverter low-side is composed of three IGBTs, freewheeling diodes, and one control IC for each IGBT. It has gate drive and protection functions.
- 3. Inverter power side is composed of four inverter DC-link input terminals and three inverter output terminals.

Absolute Maximum Ratings ($T_J = 25$ °C, unless otherwise specified.)

Inverter Part

Symbol	Parameter	Parameter Conditions		Unit
V _{PN}	Supply Voltage	Applied between P - N _U , N _V , N _W	450	V
V _{PN(Surge)}	Supply Voltage (Surge)	Applied between P - N _U , N _V , N _W	500	V
V _{CES}	Collector - Emitter Voltage		600	V
± I _C	Each IGBT Collector Current	$T_{C} = 25^{\circ}C, T_{J} < 150^{\circ}C$	8	Α
± I _{CP}	Each IGBT Collector Current (Peak)	$T_C = 25$ °C, $T_J < 150$ °C, Under 1 ms Pulse Width	16	Α
P _C	Collector Dissipation	T _C = 25°C per Chip	32	W
T _J	Operating Junction Temperature	(2nd Note 1)	-40 ~ 150	°C

2nd Notes:

Control Part

Symbol	Parameter	Conditions	Rating	Unit
V _{CC}	Control Supply Voltage	Applied between V _{CC(H)} , V _{CC(L)} - COM	20	٧
V_{BS}	High - Side Control Bias Voltage		20	٧
V _{IN}	Input Signal Voltage	$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	-0.3 ~ V _{CC} + 0.3	V
V_{FO}	Fault Output Supply Voltage	Applied between V _{FO} - COM	-0.3 ~ V _{CC} + 0.3	٧
I _{FO}	Fault Output Current	Sink Current at V _{FO} pin	1	mA
V_{SC}	Current-Sensing Input Voltage	Applied between C _{SC} - COM	$-0.3 \sim V_{CC} + 0.3$	>

Bootstrap Diode Part

Symbol	Parameter	Conditions	Rating	Unit
V_{RRM}	Maximum Repetitive Reverse Voltage		600	V
I _F	Forward Current	$T_{C} = 25^{\circ}C, T_{J} < 150^{\circ}C$	0.50	Α
I _{FP}	Forward Current (Peak)	T_{C} = 25°C, T_{J} < 150°C, Under 1 ms Pulse Width	1.50	Α
TJ	Operating Junction Temperature		-40 ~ 150	°C

Total System

Symbol	Parameter	Conditions	Rating	Unit
V _{PN(PROT)}	Self-Protection Supply Voltage Limit (Short-Circuit Protection Capability)	$V_{CC} = V_{BS} = 13.5 \sim 16.5 \text{ V}$ T _J = 150°C, Non-Repetitive, < 2 μs	400	V
T _{STG}	Storage Temperature		-40 ~ 125	°C
V _{ISO}	Isolation Voltage	60 Hz, Sinusoidal, AC 1 Minute, Connect Pins to Heat Sink Plate	2000	V _{rms}

Thermal Resistance

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit
R _{th(j-c)Q}	Junction to Case Thermal Resistance	Inverter IGBT Part (per 1 / 6 module)	-	-	3.8	°C / W
R _{th(j-c)F}		Inverter FWDi Part (per 1 / 6 module)	-	-	4.8	°C / W

^{1.} The maximum junction temperature rating of the power chips integrated within the Motion SPM 8 45 product is 150°C.

^{2.} For the measurement point of case temperature (T $_{\rm C}$), please refer to Figure 2.

Electrical Characteristics ($T_J = 25^{\circ}C$, unless otherwise specified.)

Inverter Part

S	ymbol	Parameter	Cond	itions	Min.	Тур.	Max.	Unit
V	CE(SAT)	Collector - Emitter Saturation Voltage	$V_{CC} = V_{BS} = 15 \text{ V}$ $V_{IN} = 5 \text{ V}$	I _C = 8 A, T _J = 25°C	-	1.7	2.2	V
	V_{F}	FWDi Forward Voltage	$V_{IN} = 0 V$	$I_F = 8 A, T_J = 25^{\circ}C$	-	1.7	2.2	V
HS	t _{ON}	Switching Times	$V_{PN} = 300 \text{ V}, V_{CC} = V_{E}$	_{BS} = 15 V, I _C = 8 A	0.45	0.75	1.25	μS
	t _{C(ON)}		$T_J = 25^{\circ}C$ $V_{vv} = 0 \text{ V} \leftrightarrow 5 \text{ V}$ Induce	tive I oad	-	0.20	0.45	μS
	t _{OFF}		(2nd Note 3)	$V_{IN} = 0 \ V \leftrightarrow 5 \ V$, Inductive Load (2nd Note 3)		0.80	1.30	μS
	t _{C(OFF)}					0.30	0.55	μS
	t _{rr}				-	0.15	-	μS
LS	t _{ON}		$V_{PN} = 300 \text{ V}, V_{CC} = V_{E}$	_{3S} = 15 V, I _C = 8 A	0.45	0.75	1.25	μS
	t _{C(ON)}		$T_J = 25^{\circ}C$ $V_{IN} = 0 V \leftrightarrow 5 V$, Induce	tive I and	-	0.20	0.45	μS
	t _{OFF}		(2nd Note 3)	Silve Load	-	0.80	1.30	μS
	t _{C(OFF)}			, , , , , , , , , , , , , , , , , , , ,		0.30	0.55	μS
	t _{rr}				-	0.15	-	μS
	I _{CES}	Collector - Emitter Leakage Current	V _{CE} = V _{CES}		-	-	1	mA

^{3.} t_{ON} and t_{OFF} include the propagation delay of the internal drive IC. $t_{\text{C(ON)}}$ and $t_{\text{C(OFF)}}$ are the switching time of IGBT itself under the given gate driving condition internally. For the detailed information, please see Figure 4.

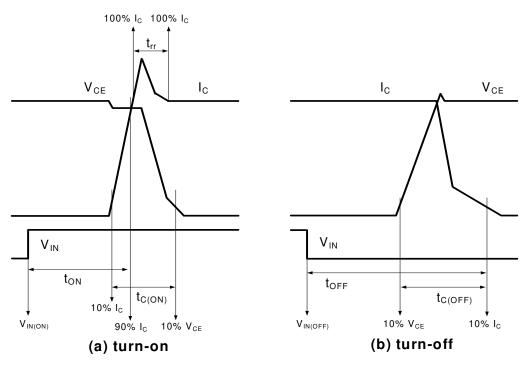


Figure 4. Switching Time Definition

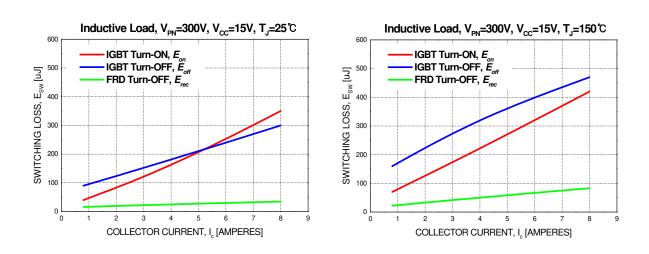


Figure 5. Switching Loss Characteristics (Typical)

Control Part

Symbol	Parameter	Conditions	1	Min.	Тур.	Max.	Unit
I _{QCCH}	Quiescent V _{CC} Supply	V _{CC(H)} = 15 V, IN _(UH,VH,WH) = 0 V	V _{CC(H)} - COM	-	-	0.10	mA
I _{QCCL}	Current	V _{CC(L)} = 15 V, IN _(UL,VL, WL) = 0 V	V _{CC(L)} - COM	-	-	2.65	mA
I _{PCCH}	Operating V _{CC} Supply Current	$V_{CC(L)} = 15 \text{ V}, f_{PWM} = 20 \text{ kHz}, duty = 50\%, Applied to One PWM Signal Input for High-Side}$	V _{CC(H)} - COM	-	-	0.15	mA
I _{PCCL}		$V_{\text{CC(L)}} = 15 \text{ V}, f_{\text{PWM}} = 20 \text{ kHz}, \text{ duty}$ = 50%, Applied to One PWM Signal Input for Low-Side	V _{CC(L)} - COM	-	-	3.65	mA
I _{QBS}	Quiescent V _{BS} Supply Current	V _{BS} = 15 V, IN _(UH, VH, WH) = 0 V	$V_{B(U)} - V_{S(U)}, V_{B(V)} - V_{S(V)}, V_{B(W)} - V_{S(W)}$	-	-	0.30	mA
I _{PBS}	Operating V _{BS} Supply Current	$V_{\rm CC} = V_{\rm BS} = 15 \text{ V}, f_{\rm PWM} = 20 \text{ kHz},$ Duty = 50%, Applied to One PWM Signal Input for High-Side	$\begin{matrix} V_{B(U)} - V_{S(U)}, V_{B(V)} - \\ V_{S(V)}, V_{B(W)} - V_{S(W)} \end{matrix}$	-	-	2.00	mA
V_{FOH}	Fault Output Voltage	$V_{SC} = 0 \text{ V}, V_{FO} \text{ Circuit: } 10 \text{ k}\Omega \text{ to } 5 \text{ V}$	/ Pull-up	4.5	-	-	V
V_{FOL}		$V_{SC} = 1 \text{ V}, V_{FO} \text{ Circuit: } 10 \text{ k}\Omega \text{ to } 5 \text{ V}$	/ Pull-up	-	-	0.5	V
V _{SC(ref)}	Short-Circuit Current Trip Level	V _{CC} = 15 V (2nd Note 4)		0.45	0.50	0.55	V
UV _{CCD}		Detection level		10.5	-	13.0	V
UV _{CCR}	Supply Circuit Under-Voltage	Reset level		11.0	-	13.5	V
UV _{BSD}	Protection	Detection level		10.0	-	12.5	٧
UV _{BSR}		Reset level		10.5	-	13.0	V
t _{FOD}	Fault-Out Pulse Width			30	-	-	μS
V _{IN(ON)}	ON Threshold Voltage	Applied between IN _(UH) , IN _(VH) , IN _(WH) , IN _(UL) , IN _(VL) ,		-	-	2.6	V
V _{IN(OFF)}	OFF Threshold Voltage	IN _(WL) - COM	N _(WL) - COM		-	-	V
R _{TH}	Resistance of	@T _{TH} = 25°C, (2nd Note 5)		-	47	-	kΩ
	Thermister	@T _{TH} = 100°C		-	2.9	-	kΩ

^{4.} Short-circuit protection is functioning only at the low-sides.

^{5.} T_{TH} is the temperature of thermister itselt. To know case temperature (T_C), please make the experiment considering your application.

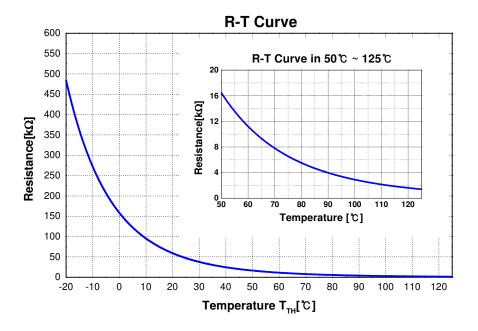


Figure. 6. R-T Curve of The Built-In Thermistor

Bootstrap Diode Part

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit
V _F	Forward Voltage	$I_F = 0.1 \text{ A}, T_C = 25^{\circ}\text{C}$	-	2.5	-	V
t _{rr}	Reverse-Recovery Time	$I_F = 0.1 \text{ A}, T_C = 25^{\circ}\text{C}$	-	80	-	ns

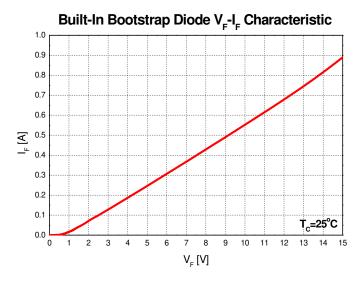


Figure 7. Built-In Bootstrap Diode Characteristic

2nd Notes:

6. Built-in bootstrap diode includes around 15 $\,\Omega$ resistance characteristic.

Recommended Operating Conditions

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit
V _{PN}	Supply Voltage	Applied between P - N _U , N _V , N _W	-	300	400	V
V _{CC}	Control Supply Voltage	Applied between V _{CC(H)} , V _{CC(L)} - COM	13.5	15.0	16.5	٧
V _{BS}	High-Side Bias Voltage	Applied between $V_{B(U)}$ - $V_{S(U)}$, $V_{B(V)}$ - $V_{S(V)}$, $V_{B(W)}$ - $V_{S(W)}$	13.0	15.0	18.5	V
dV_{CC} / dt , dV_{BS} / dt	Control Supply Variation		- 1	-	1	V / μs
t _{dead}	Blanking Time for Preventing Arm-Short	For each input signal	1.5	-	-	μS
f _{PWM}	PWM Input Signal	- 40°C < T _J < 150°C	-	-	20	kHz
V _{SEN}	Voltage for Current Sensing	Applied between N _U , N _V , N _W - COM (Including Surge-Voltage)	- 4		4	V
P _{WIN(ON)}	Minimun Input Pulse	(2nd Note 7)	0.5	-	-	μS
P _{WIN(OFF)}	Width		0.5	-	-	

2nd Notes:

7. This product might not make response if input pulse width is less than the recommanded value.

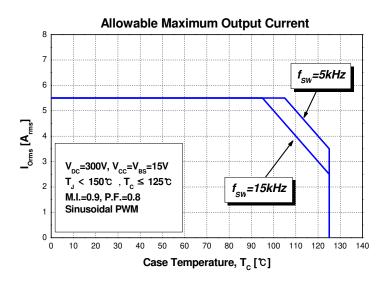


Figure 8. Allowable Maximum Output Current

2nd Notes:

8. This allowable output current value is the reference data for the safe operation of this product. This may be different from the actual application and operating condition.

Mechanical Characteristics and Ratings

Parameter	Co	Min.	Тур.	Max.	Unit	
Device Flatness	See Figure 9		0	-	+ 120	μm
Mounting Torque	Mounting Screw: M3	Recommended 0.7 N • m	0.6	0.7	0.8	N•m
	See Figure 10	Recommended 7.1 kg • cm	6.2	7.1	8.1	kg • cm
Weight			-	11	-	g

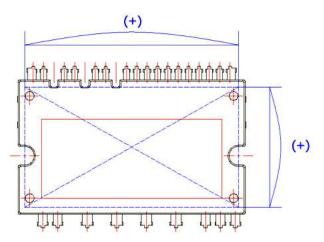


Figure 9. Flatness Measurement Position

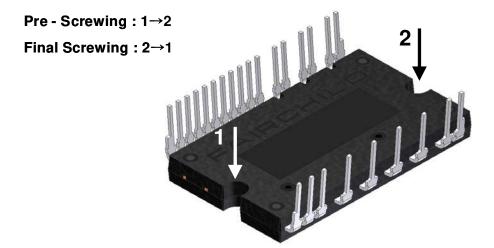
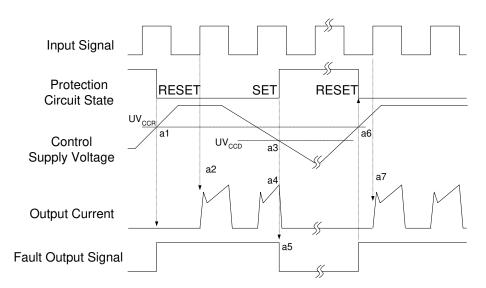


Figure 10. Mounting Screws Torque Order

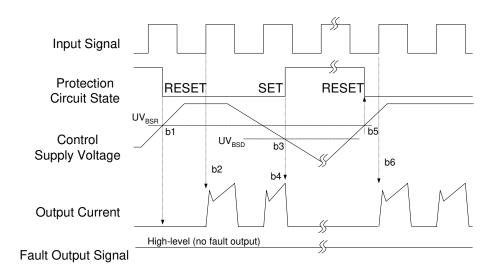
- 9. Do not make over torque when mounting screws. Much mounting torque may cause ceramic cracks, as well as bolts and AI heat-sink destruction.
- 10. Avoid one side tightening stress. Figure 10 shows the recommended torque order for mounting screws. Uneven mounting can cause the ceramic substrate of the SPM[®] 45 package to be damaged. The pre-screwing torque is set to 20 ~ 30% of maximum torque rating.

Time Charts of Protective Function



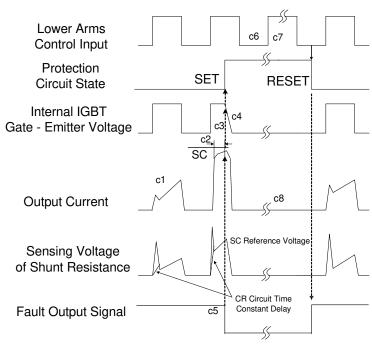
- a1 : Control supply voltage rises: after the voltage rises UV_{CCR} , the circuits start to operate when next input is applied.
- a2: Normal operation: IGBT ON and carrying current.
- a3 : Under-voltage detection (UV_{CCD}).
- a4: IGBT OFF in spite of control input condition.
- a5 : Fault output operation starts.
- a6 : Under-voltage reset (UV $_{CCR}$).
- a7: Normal operation: IGBT ON and carrying current.

Figure 11. Under-Voltage Protection (Low-Side)



- b1 : Control supply voltage rises: after the voltage reaches UV_{BSR}, the circuits start to operate when next input is applied.
- b2: Normal operation: IGBT ON and carrying current.
- b3 : Under-voltage detection (UV_{BSD}).
- b4: IGBT OFF in spite of control input condition, but there is no fault output signal.
- b5 : Under-voltage reset (UV_{BSB}).
- b6: Normal operation: IGBT ON and carrying current.

Figure 12. Under-Voltage Protection (High-Side)



(with the external shunt resistance and CR connection)

- c1: Normal operation: IGBT ON and carrying current.
- c2 : Short-circuit current detection (SC trigger).
- c3: Hard IGBT gate interrupt.
- c4: IGBT turns OFF.
- c5 : Input "LOW": IGBT OFF state.
- c6 : Input "HIGH": IGBT ON state, but during the active period of fault output, the IGBT doesn't turn ON.
- c7: IGBT OFF state.

Figure 13. Short-Circuit Protection (Low-Side Operation Only)

Input/Output Interface Circuit

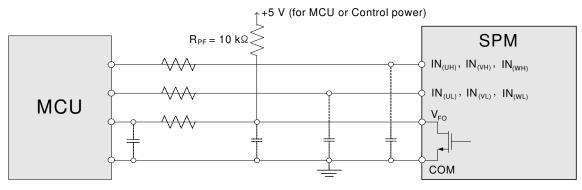


Figure 14. Recommended MCU I/O Interface Circuit

^{11.} RC coupling at each input (parts shown dotted) might change depending on the PWM control scheme in the application and the wiring impedance of the application's printed circuit board. The input signal section of the Motion SPM[®] 45 product integrates a 5 kΩ (typ.) pull-down resistor. Therefore, when using an external filtering resistor, pay attention to the signal voltage drop at input terminal.

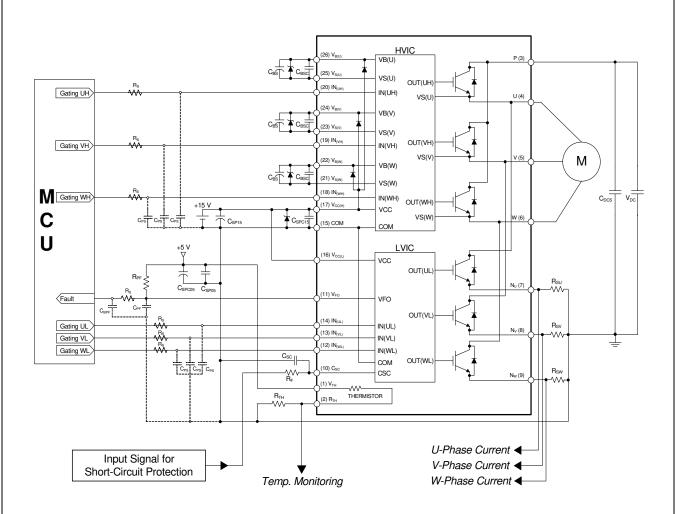
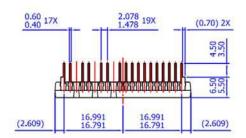


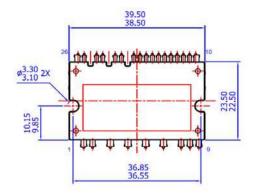
Figure 15. Typical Application Circuit

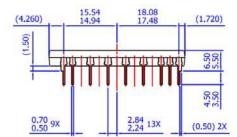
3rd Notes:

- 1) To avoid malfunction, the wiring of each input should be as short as possible (less than 2 3 cm).
- 2) By virtue of integrating an application-specific type of HVIC inside the Motion SPM[®] 45 product, direct coupling to MCU terminals without any optocoupler or transformer isolation is possible.
- 3) V_{FO} output is open-drain type. This signal line should be pulled up to the positive side of the MCU or control power supply with a resistor that makes I_{FO} up to 1 mA (please refer to Figure 14).
- 4) C_{SP15} of around seven times larger than bootstrap capacitor C_{BS} is recommended.
- 5) Input signal is active-HIGH type. There is a 5 k Ω resistor inside the IC to pull down each input signal line to GND. RC coupling circuits is recommanded for the prevention of input signal oscillation. R_SC_{PS} time constant should be selected in the range 50 ~ 150 ns (recommended R_S = 100 Ω , C_{PS} = 1 nF).
- 6) To prevent errors of the protection function, the wiring around R_F and C_{SC} should be as short as possible.
- 7) In the short-circuit protection circuit, please select the R_FC_{SC} time constant in the range 1.5 \sim 2 $\mu s.$
- 8) The connection between control GND line and power GND line which includes the N_U, N_V, N_W must be connected to only one point. Please do not connect the control GND to the power GND by the broad pattern. Also, the wiring distance between control GND and power GND should be as short as possible.
- 9) Each capacitor should be mounted as close to the pins of the Motion SPM 45 product as possible.
- 10) To prevent surge destruction, the wiring between the smoothing capacitor and the P & GND pins should be as short as possible. The use of a high-frequency non-inductive capacitor of around 0.1 ~ 0.22 µF between the P and GND pins is recommended.
- 11) Relays are used in almost every systems of electrical equipment in home appliances. In these cases, there should be sufficient distance between the MCU and the relays.
- 12) The zener diode or transient voltage suppressor should be adopted for the protection of ICs from the surge destruction between each pair of control supply terminals (recommanded zener diode is 22 V / 1 W, which has the lower zener impedance characteristic than about 15 \,\Omega\$).
- 13) Please choose the electrolytic capacitor with good temperature characteristic in C_{BS} . Also, choose 0.1 ~ 0.2 μF R-category ceramic capacitors with good temperature and frequency characteristics in C_{BSC} .
- 14) For the detailed information, please refer to the AN-9070, AN-9071, AN-9072, RD-344, and RD-345.

Detailed Package Outline Drawings

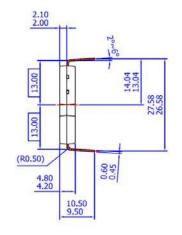


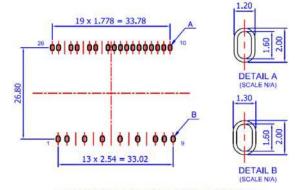




NOTES: UNLESS OTHERWISE SPECIFIED

- A) THIS PACKAGE DOES NOT COMPLY TO ANY CURRENT PACKAGING STANDARD
- B) ALL DIMENSIONS ARE IN MILLIMETERS
- C) DIMENSIONS ARE EXCLUSIVE OF BURRS, MOLD FLASH, AND TIE BAR EXTRUSIONS
- D) () IS REFERENCE
- E) [] IS ASS'Y QUALITY
- F) DRAWING FILENAME: MOD26AAREV2.0





LAND PATTERN RECOMMENDATIONS

Package drawings are provided as a service to customers considering Fairchild components. Drawings may change in any manner without notice. Please note the revision and/or data on the drawing and contact a FairchildSemiconductor representative to verify or obtain the most recent revision. Package specifications do not expand the terms of Fairchild's worldwide therm and conditions, specifically the the warranty therein, which covers Fairchild products.

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http://www.fairchildsemi.com/dwg/MO/MOD26AA.pdf





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Dual Cool™ Making Small Speakers Sound Louder

EcoSPARK® and Better™ EfficientMax™ MegaBuck^{TI} MICROCOUPLER™ ESBCTh

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Programmable Active Droop™

OFFT OSTM Quiet Series™ RapidConfigure™

Saving our world, 1mW/W/kW at a time™

SignalWise** SmartMax™ SMART START™

Solutions for Your Success™

SPM® STEALTH* SuperFET® SuperSOT™-3 SuperSOT™-6 SuperSOT™-8 SupreMOS® SyncFET™

Sync-Lock™ SYSTEM STERNER TinyBoost[®] TinyBuck[®] TinyCalc™ TinyLogic[®] TINYOPTO™ TinyPower¹¹ TinyPWM™ TinyWire™ TranSiC™ TriFault Detect™ TRUECURRENT®*

UHC Ultra FRFET™ UniFET™ VCX** VisualMax™ VoltagePlus™ XS™

uSerDes™

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FPS™

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