

[ADIN1100](http://www.analog.com//ADIN1100.html)

Robust, Industrial, Low Power 10BASE-T1L PHY

FEATURES

- ► 10BASE-T1L IEEE Standard 802.3cg-2019 compliant
- \triangleright Cable reach up to 1700 m with 1.0 V p-p and 2.4 V p-p
- ► Supports 1.0 V p-p and 2.4 V p-p transmit levels
- ► Supports intrinsic safety applications
- ► Low power consumption: 39 mW (dual supply, 1.0 V p-p)
- ► Diagnostics
	- ► Cable fault detection with TDR
	- \blacktriangleright Link quality indicator with MSE
	- ► Frame generator and checker
	- ► Multiple loopback modes
	- ► IEEE test mode support
- ► MII, RMII, and RGMII MAC interfaces
- ► MDIO management interface
- ► Unmanaged configuration using pin strapping
- ► 25 MHz crystal or external clock input (50 MHz for RMII)
- ► Single or dual supply with 1.8 V or 3.3 V operation
- ► 3.3 V, 2.5 V, or 1.8 V MAC interface VDDIO supply
- ► Integrated power supply monitoring and POR
- ► EMC test standards
	- ► IEC 61000-4-4 EFT (±4 kV)
	- ► IEC 61000-4-2 ESD (±4 kV contact discharge)
	- \blacktriangleright IEC 61000-4-2 ESD (\pm 8 kV air discharge)
	- ► IEC 61000-4-5 surge (±4 kV)
	- \blacktriangleright IEC 61000-4-6 conducted immunity (10 V/m)
	- ► IEC 61000-4-3 radiated immunity (Class A)
	- ► EN 55032 radiated emissions (Class B)
- ► Small package: [40-lead, 6 mm × 6 mm LFCSP](#page-79-0)
- ► Temperature range
	- ► Industrial: −40°C to +85°C
	- ► Extended: −40°C to +105°C

APPLICATIONS

- ► Process control
- ► Factory automation
- ► Building automation
- ► Field instruments and switches

FUNCTIONAL BLOCK DIAGRAM

Figure 1.

GENERAL DESCRIPTION

The ADIN1100 is a low power, single port, 10BASE-T1L transceiver designed for industrial Ethernet applications and is compliant with the IEEE® 802.3cg-2019™ Ethernet standard for long reach 10 Mbps single pair Ethernet (SPE). The ADIN1100 integrates an Ethernet PHY core with all the associated analog circuitry, input and output clock buffering, the management interface control register and subsystem registers, as well as the MAC interface and control logic to manage the reset, clock control, and pin configuration.

The ADIN1100 supports cable reach of up to 1700 meters with autonegotiation enabled and has ultra low power consumption of 39 mW.

The PHY core supports the 1.0 V p-p operating mode and the 2.4 V p-p operating mode defined in the IEEE 802.3cg standard and can operate from a single power supply rail of 1.8 V or 3.3 V, with the lower voltage option supporting the 1.0 V p-p transmit voltage level.

The ADIN1100 has an integrated voltage supply monitoring circuit and power-on reset (POR) circuitry to improve system level robustness.

The MDIO interface is a 2-wire serial interface for communication between a host processor or MAC and the ADIN1100, thereby allowing access to control and status information in the PHY core management registers. This interface is compatible with both the IEEE 802.3 Standard Clause 22 and Clause 45 management frame structures.

Rev. A

[DOCUMENT FEEDBACK](https://form.analog.com/Form_Pages/feedback/documentfeedback.aspx?doc=ADIN1100.pdf&product=ADIN1100&rev=A) [TECHNICAL SUPPORT](http://www.analog.com/en/content/technical_support_page/fca.html)

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REVISION HISTORY

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9/2021—Revision 0: Initial Version

SPECIFICATIONS

AVDD_H = AVDD_L = VDDIO = 3.3 V; DVDD_1P1 from internal low dropout (LDO) regulator (DVDD_1P1 = DLDO_1P1); all specifications at −40°C to +105°C, unless otherwise noted.

SPECIFICATIONS

Table 1. General Specifications (Continued)

¹ Load capacitance $(C_L) = ((C1 \times C2)/(C1 + C2) + C_{STRAY})$, where C_{STRAY} is the stray capacitance including routing and package parasitics.

 2 R_P and C_P are the values of the equivalent parallel RC circuit to ac ground (R_P||C_P), modeling the driving point impedance of the XTAL_I/CLK_IN pin.

Table 2. 10BASE-T1L Specifications

SPECIFICATIONS

Table 2. 10BASE-T1L Specifications (Continued)

TIMING CHARACTERISTICS

POWER-UP TIMING

¹ The minimum time interval is referenced to the last supply to reach its rising threshold. There is no specific power supply sequencing required.

Figure 2. Power-Up Timing

MANAGEMENT INTERFACE TIMING

Table 4.

Figure 3. Management Interface Timing

ABSOLUTE MAXIMUM RATINGS

 T_A = 25°C, unless otherwise noted.

Table 5.

See the [Pin Configuration and Function Descriptions](#page-8-0) section for the full list of MAC interface pins.

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

THERMAL RESISTANCE

Thermal performance is directly linked to printed circuit board (PCB) design and operating environment. Careful attention to PCB thermal design is required.

 θ_{JA} is the natural convection junction to ambient thermal resistance measured in a one cubic foot sealed enclosure. θ_{JC} is the junction to case thermal resistance.

Table 6. Thermal Resistance

 $1 - \theta_{JA}$ is specified for the worst case conditions, that is, a device soldered in a circuit board for surface-mount packages.

² Test Condition 1: thermal impedance simulated values are based on a JE-DEC 2S2P thermal test board with thermal vias. See JEDEC JESD51.

ELECTROSTATIC DISCHARGE (ESD) RATINGS

Human body model (HBM) per ANSI/ESDA/JEDEC JS-001.

Field induced charged device model (FICDM) per ANSI/ESDA/JE-DEC JS-002.

ESD Ratings

Table 7. ADIN1100, 40-Lead LFCSP

ESD CAUTION

ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

Figure 4. Pin Configuration

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

Table 8. Pin Function Descriptions (Continued)

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

Table 8. Pin Function Descriptions (Continued)

¹ Where a pin is shared between a functional signal and a hardware configuration pin signal, the hardware configuration pin signal is listed last.

² PU is pull-up, and PD is pull-down. The internal pull-up or pull-down resistor is predefined and non-configurable. N/A means not applicable.

TYPICAL PERFORMANCE CHARACTERISTICS

Figure 5. Power vs. Temperature, 1.8 V Single Supply, Internal LDO Circuit, 10BASE-T1L Mode

Figure 6. Power vs. Temperature, 3.3 V Single Supply, Internal LDO Circuit, 10BASE-T1L Mode

Figure 7. Power vs. Temperature, AVDD_H = 3.3 V, AVDD_L = 3.3 V, VDDIO = 1.8 V, Internal LDO Circuit, 10BASE-T1L Mode

Figure 8. Power vs. Temperature, AVDD_H = 3.3 V, VDDIO = 3.3 V, AVDD_L = 1.8 V, Internal LDO Circuit, 10BASE-T1L Mode

Figure 9. Power vs. Temperature, AVDD_H = 3.3 V, AVDD_L = VDDIO = 1.8 V, Internal LDO Circuit, 10BASE-T1L Mode

The ADIN1100 is a low power, single port 10BASE-T1L Ethernet PHY device, compliant with the IEEE 802.3cg Ethernet standard for long reach, 10 Mbps single pair Ethernet.

The ADIN1100 integrates the following features:

- ► PHY core with common analog circuitry
- ► Input and output clock buffering
- ► MDIO interface to control subsystem registers, clock, and software resets
- ► MAC interface control logic
- ► Hardware configuration pins
- ► Configurable hardware interrupt pin
- ► Two configurable LED pins

POWER SUPPLY DOMAINS

The ADIN1100 has four power supply domains and requires a minimum of one supply rail.

- ► AVDD_H is the analog power supply input for the analog front end (AFE) circuitry in the ADIN1100.
- ► AVDD L is the analog supply voltage for the internal LDO circuits. AVDD_L can be connected to the AVDD_H rail in single supply mode, or to an alternative lower voltage rail in dual supplies mode for lower power consumption.
- ► DVDD 1P1 is the 1.1 V digital core power supply input. It can operate from the internal 1.1 V LDO output available on the DLDO 1P1 pin. Alternatively, DVDD 1P1 can be driven from an external 1.1 V supply for lower power consumption.
- ► VDDIO is the digital power supply input for the ADIN1100 MAC interface, MDIO, and digital inputs/outputs (I/Os). It can be connected directly to the AVDD_L rail or to an external power rail.

The [System Level Power Management](#page-33-0) section describes various application circuits that can be used as reference.

Single-Supply Applications

In a single-supply application, connect AVDD_H and AVDD_L to VDDIO, and in the ADIN1100, use the internal 1.1 V LDO circuit for DVDD 1P1. The appropriate supply voltage used depends on the end application and cable length. A recommended circuit is shown in the [Single-Supply Configuration](#page-33-0) section.

Long Reach and Trunk/Spur Applications

The 1.0 V p-p transmit operating mode supports the spur use case and can operate at a lower AVDD_H supply voltage of 1.8 V. The ADIN1100 in this mode supports intrinsic safety applications.

The higher transmit operating mode of 2.4 V p-p supports trunk applications and requires a higher AVDD_H supply voltage of 3.3 V. This mode can be used for longer cable lengths in industrial Ethernet environments with higher noise levels.

MAC INTERFACE

The ADIN1100 provides the option of MII, RMII, or RGMII MAC interfaces. The MAC interface is selected using the hardware configuration pins (RX_CLK/RXC/MACIF_SEL0 and RX_ER/MA-CIF SEL1) or via the management interface (MDIO). The two hardware configuration pins for MAC interface selection strap the ADIN1100 configuration after power-up, hardware reset, or software reset.

By default (hardware configuration pins floating), the MAC interface is configured in RMII mode. See the [MAC Interface Selection](#page-24-0) for more details on how to configure the MAC interface using the hardware configuration pins.

It is recommended to use the hardware configuration pins for the MAC interface selection because the supported interfaces have different clock and pin mapping requirements.

MII Interface Mode

For the MII receive interface, the ADIN1100 generates a 2.5 MHz reference clock on RX_CLK to synchronize the RXD_3 to RXD_0 receive data signals. \overline{RX} DV indicates to the MAC that valid data is present on the RXD 3 to RXD 0 signals. RX ER is driven high by the ADIN1100 if an error was detected in the frame that was received from the MDI interface and is being transmitted to the MAC interface.

For the MII transmit interface, the ADIN1100 generates a 2.5 MHz reference clock on TX_CLK. The MAC transmits data on TXD_3 to TXD 0 that is synchronized with TX CLK. The MAC asserts TX EN high to indicate to the ADIN1100 that transmission data is available on the TXD_3 to TXD_0 signal lines.

Figure 10. MII MAC PHY Interface Signals

RMII Interface Mode

RMII mode requires an external 50 MHz clock, which can be sourced externally from a MAC chip or a reference clock and

applied to the ADIN1100 XTAL I/CLK IN pin. The 50 MHz clock is used for both the transmit and receive interfaces.

The receive data, RXD_1 to RXD_0, transitions synchronously to the reference clock (REF_CLK). The carrier sense and received data valid signal (CRS \overline{DV}) is a combination of the carrier sense and RX_DV signals, and is asserted while the receive medium is not idle. CRS_DV is asserted asynchronously to REF_CLK and deasserted synchronously.

RX ER is also synchronous to the reference clock signal (REF_CLK) and asserted when an error is detected in the received frame or when a false carrier is detected. RX_ER assertion on a false carrier can be disabled by software.

Do not configure the MAC interface to RMII in software without ensuring the required 50 MHz clock is available. See [MAC Interface](#page-63-0) [Configuration Register](#page-63-0) for more details on RMII interface parameters.

Figure 11. RMII MAC PHY Interface Signals

RGMII Interface Mode

For the RGMII receive interface, the ADIN1100 generates a 2.5 MHz RXC signal to synchronize RXD 3 to RXD 0. RX CTL is a combination of the RX_DV and RX_ER signals (as described in [Table 8](#page-8-0)) using both edges of the RXC signal. The ADIN1100 transmits the RX_DV signal on the positive edge of RXC and a combination (XOR function) of RX_DV and RX_ER on the negative edge of RXC.

For the RGMII transmit interface, the ADIN1100 generates a 2.5 MHz reference clock on TX_CLK. The MAC transmits the TXD_3 to TXD 0 data on both edges of TXC. TX_CTL is a combination of the TX_EN and TX_ER signals using both edges of TXC. TX_EN is transmitted on the positive edge of TXC, and a combination (XOR function) of TX_EN and TX_ER is transmitted on the negative edge of TXC.

Figure 12. RGMII MAC PHY Interface Signals

TRANSMIT AMPLITUDE CONFIGURATION

The ADIN1100 supports two transmit amplitude modes of operation, as follows:

- \blacktriangleright 1.0 V p-p and 2.4 V p-p mode (high level)
- ► 1.0 V p-p only mode

The high level transmit operating mode allows the ADIN1100 to support both voltage levels. The operating level can then be automatically configured during autonegotiation (if enabled) based on the link partner capabilities.

The mode of operation is configured through the TX2P4 EN hardware configuration pin signal (see the [Transmit Amplitude](#page-23-0) section). The ADIN1100 also configures the default value for the transmit level register bits used for the autonegotiation process based on the level configured with the RX D0/ TX2P4 EN pin (see the [Transmit](#page-14-0) [Amplitude Advertisement](#page-14-0) section).

The ADIN1100 is configured in high level transmit operating mode by default if the RX D0/TX2P4 EN pin is left floating (internal pull-down resistor).

MASTER/SLAVE CONFIGURATION

The 10BASE-T1L standard uses a master/slave clock scheme. This scheme is commonly used in full duplex transceiver standards with echo cancellation. One PHY is designated as the master, and the other PHY as the slave. Autonegotiation is used to determine which PHY is the master and which is the slave. Master and slave assignment does not generally matter.

Hardware Configuration

The ADIN1100 can be configured to prefer slave or prefer master through the RXD_1/MS_SEL hardware configuration pin. The recommendation is to select prefer slave or prefer master (see the [Master/Slave Preference](#page-23-0) section). If autonegotiation is disabled, the MS SEL hardware configuration pin signal sets the default

master/slave selection. The ADIN1100 is configured in prefer slave by default if the pin is left floating (internal pull-down resistor).

Software Configuration

The master and slave configuration bit (CFG_MST) is automatically updated after power-up, hardware reset, or software reset based on the MS SEL hardware configuration pin signal level. This bit is only used when autonegotiation is disabled. Otherwise, this bit is set or reset during the autonegotiation process (see the Autonegotiation section).

AUTONEGOTIATION

The ADIN1100 uses the autonegotiation capability in accordance with IEEE 802.3 Clause 98, providing a mechanism for exchanging information between the local device and link partners to agree to a common mode of operation. Single twisted pair autonegotiation is performed using differential Manchester encoding (DME) pages exchanged between the local device and its link partner. At a high level, the autonegotiation provides the following functions:

- ► Transmit
- ► Receive
- ► Half duplex
- ► Arbitration

During the autonegotiation process, the local device advertises its own capabilities and compares them to those received from the link partner. The arbitration mechanism defines the operating mode selected so that the transmit amplitude mode and master/slave selection are configured for the linked devices.

If the link is dropped, the autonegotiation process restarts automatically. An autonegotiation restart can also be requested by writing to the autonegotiation restart bit (AN_RESTART) in the autonegotiation control register (AN_CONTROL).

The autonegotiation process can take time to complete, depending on the number of pages exchanged, but is always the fastest way to bring up a link. Clause 98 of the IEEE 802.3 standard details the sequence timers and DME pages timing related to autonegotiation.

Autonegotiation is enabled by default for the ADIN1100, and it is strongly recommended to always keep it enabled.

Transmit Amplitude Advertisement

High Voltage Transmit Ability Advertisement

The B10L TX LVL HI ABLE bit configures the default values for the autonegotiation advertisement parameters. This bit is read only and configured based on the TX2P4_EN hardware configuration pin signal as described in the [Transmit Amplitude s](#page-23-0)ection. The transmit amplitude advertisement parameters are defined with the following bits:

- ► AN_ADV_B10L_TX_LVL_HI_ABL: advertisement of the 10BASE-T1L high level transmit operating mode ability bit
- ► AN_ADV_B10L_TX_LVL_HI_REQ: advertisement of the 10BASE-T1L high level transmit operating mode request bit
- ► B10L_TX_LVL_HI: 10BASE-T1L transmit voltage amplitude control bit

Table 9. AN_ADV_B10L_TX_LVL_HI_ABL Settings

1 Request 2.4 V p-p transmit level

High Voltage Transmit Level Request Advertisement

The ADIN1100 can be configured to advertise a request for the 2.4 V p-p transmit level using the 10BASE-T1L high level transmit operating mode ability advertisement bit (AN_ADV_B10L_TX_LVL_HI_ABL). Note that the 2.4 V p-p transmit level must be enabled using the $\overline{\text{TX2P4}}$ EN hardware configuration pin signal to enable the high voltage request advertisement.

Link Partner Transmit Level Advertisement

The high level transmit information advertised from the link partner can be read using the following bits:

- ► AN_LP_ADV_B10L_TX_LVL_HI_ABL: Link Partner 10BASE-T1L high level transmit operating mode ability
- ▶ AN_LP_ADV_B10L_TX_LVL_HI_REQ: Link Partner 10BASE-T1L high level transmit operating mode request

These bits are updated during the autonegotiation process and are valid when the autonegotiation complete register bit (AN_COM-PLETE) is set.

Transmit Amplitude Resolution

Overview

Autonegotiation supports the following features to define the transmit amplitude to use between a local node and its link partner:

- ► Advertise the high voltage transmit ability from the local node
- ► Request to use the high voltage transmit level from the local node
- ► Read the link partner transmit level ability and transmit level request
- ► Autonegotiation and transmit level operating mode selection

If the ADIN1100 is configured in high voltage mode, the autonegotiation process determines the level to use based on the link partner capabilities.

Determination of Transmit Level Resolution

For a 10BASE-T1L link, if either the local or remote PHY advertises that it is not capable of supporting the high level (2.4 V p-p) transmit operating mode or if neither the local nor remote PHY advertises a request for high level (2.4 V p-p) transmit operating mode, the result is operation at the 1.0 V p-p transmit level.

If both the local and remote PHYs advertise that they are capable of transmitting in the high level (2.4 V p-p) transmit operating mode

Table 11. Determination of Transmit Level by Autonegotiation

and if either the local or remote PHY advertises a request for high level (2.4 V p-p) transmit operating mode, the result is operation at the 2.4 V p-p transmit level.

Thus, a PHY can ensure that the device must operate at the 1.0 V p-p transmit level. But it can only request operation at the 2.4 V p-p transmit level.

Software Configuration

The ADIN1100 transmit level can also be configured in software using the following 10BASE-T1L autonegotiation advertisement bits:

- ▶ AN_ADV_B10L_TX_LVL_HI_ABL: high level transmit operating mode ability bit
- ► AN_ADV_B10L_TX_LVL_HI_REQ: high level transmit operating mode request register bits

The higher transmit level must be enabled with the TX2P4 EN hardware configuration pin signal to configure the two autonegotiation advertisement bits through software. See the [Transmit Ampli](#page-23-0)[tude](#page-23-0) section for details.

If it is required to only operate the PHY at 1.0 V p-p transmit level operation, clear the AN_ADV_B10L_TX_LVL_HI_ABL bit, so that 2.4 V p-p transmit level operation is not advertised. In this case, autonegotiation can only resolve to 1.0 V p-p transmit level operation, irrespective of the setting that the remote PHY advertises.

¹ X means don't care

Master/Slave Advertisement

On a 10BASE-T1L link, a local node and its remote link partner advertise their role capabilities whether they can operate as master, slave, prefer master, or prefer slave.

The ADIN1100 PHY provides the following functions:

- ► Master/slave configuration advertisement
- ► Forced master/slave configuration advertisement
- ► Read link partner master/slave configuration

Master/Slave Configuration Advertisement

The master/slave configuration register bit (AN_ADV_MST) is used to configure the PHY to advertise its master/slave configuration. Note that this bit is configured after power-up, hardware reset, or software reset based on the MS_SEL configuration pin signal status and can be overridden by software via the MDIO interface.

Forced Master/Slave Configuration Advertisement

The ADIN1100 PHY can be forced to operate as a master/slave. Forced configuration must be used with caution because a configuration fault can occur if the link partner is also set in forced mode as a master/slave, resulting in autonegotiation failure. The ADIN1100 can be forced to operate as a master or slave using the force master/slave configuration bit (AN_ADV_FORCE_MS) in the BASE-T1 autonegotiation advertisement register, Bits[15:0] (AN_ADV_ABILITY_L).

Read Link Partner Master/Slave Configuration

The link partner advertised master/slave setting can be read using the following bits:

Table 12. Determination of Master/Slave by Autonegotiation

- ► AN_LP_ADV_FORCE_MS: link partner force master/slave configuration register bit
- ► AN_LP_ADV_MST: link partner master/slave configuration register bit

These bits are updated during the autonegotiation process and are valid when the autonegotiation complete register bit (AN_COM-PLETE) is set.

Master/Slave Resolution

Determination of Master/Slave Configuration

On a 10BASE-T1L link, when the local and remote PHYs have the same preferred configuration, for example, both slave or both master, autonegotiation randomly assigns compatible configurations to the local and remote PHYs. When one PHY has a forced configuration, its master/slave configuration is given priority over a PHY with a preferred setting.

Master/Slave Configuration Resolution

The ADIN1100 master/slave configuration defined by the autonegotiation can be checked using the master/slave resolution result register bits (AN_MS_CONFIG_RSLTN). The bits indicate if the PHY is configured as master or slave or if there was a configuration fault.

These bits are updated during the autonegotiation process and are valid when the autonegotiation complete register bit (AN_COM-PLETE) is set.

¹ X means don't care.

Autonegotiation Fail

On a 10BASE-T1L link, the autonegotiation can fail and can be caused by various scenarios.

- ► Configuration fault such as invalid master/slave configuration
- ► Link quality issues
- ► Timeout in pages transmission

If the autonegotiation fails, the link remains down until the autonegotiation process is restarted and completed.

MANAGEMENT INTERFACE

The MII management interface provides a 2-wire serial interface between a host controller or external MAC chip and the ADIN1100 allowing access to control and status information in the subsystem and PHY core management registers.

The MII management interface consists of the following:

- ► MDC: clock line
- ► MDIO: bidirectional data line
- ► PHYAD 0, PHYAD 1, PHYAD 2: PHY address selection
- \triangleright $\overline{\text{INT}}$: configurable management interrupt

The interface is compatible with both IEEE Standard 802.3 Clause 22 and Clause 45 management frame structures (see the [MDIO](#page-37-0) [Interface](#page-37-0) section).

MDI Interface

The MDI connects the ADIN1100 to the Ethernet network via a twisted wire pair.

The ADIN1100 requires an external hybrid circuit between the TXN/TXP and RXN/RXP pins and the 10BASE-T1L twisted wire pair. This function of the hybrid is to remove the local transmitted signal from the combined signal on the cable, thereby allowing full duplex operation on the 10BASE-T1L twisted wire pair.

An example circuit with the recommended topology and components values is shown in Figure 13. Consider the size, power, and voltage rating of these components in the context of other system requirements, for example, requirements for intrinsic safety.

Figure 13. Recommended Hybrid for the ADIN1100

Hardware Interrupt (INT)

The ADIN1100 can generate a hardware interrupt to a host controller or external MAC chip using the INT pin.

PHY Status Interrupts

The following conditions can be selected to generate an interrupt:

- ► MAC interface frame checker/generator interrupt
- ► MAC interface buffers overflow/underflow interrupt
- ► Autonegotiation status change interrupt
- ► Link status change interrupt

Those conditions can be set to enable an interrupt on the $\overline{\text{INT}}$ pin using PHY_SUBSYS_IRQ_MASK (see the [PHY Subsystem](#page-71-0) [Interrupt Mask Register](#page-71-0) section).

Following a hardware interrupt on the $\overline{\text{INT}}$ pin, the interrupt source can be checked using [PHY](#page-70-0)_SUBSYS_IRQ_STATUS (see the PHY [Subsystem Interrupt Status Register](#page-70-0) section).

Hardware Reset Interrupt

The ADIN1100 can also be configured to generate a hardware interrupt after a hardware reset (RESET pin pulled low) by setting the enable hardware reset interrupt bit (CRSM_HRD_RST_IRQ_EN) in CRSM_IRQ_MASK (see the [System Interrupt Mask Register](#page-62-0) section).

Following a hardware interrupt on the INT pin, the interrupt source can be checked with the CRSM_HRD_RST_IRQ_LH bit in CRSM_IRQ_STATUS (see the [System Interrupt Status Register](#page-62-0) section).

Software Requested Interrupt

For system validation with an external host controller, the ADIN1100 can be requested to generate a hardware interrupt on the $\overline{\text{INT}}$ pin using the CRSM_SW_IRQ_REQ bit in CRSM_IRQ_MASK (see the [System Interrupt Mask Register](#page-62-0) section).

Following a hardware interrupt on the $\overline{\text{INT}}$ pin, the interrupt source can be checked with the CRSM_SW_IRQ_LH bit in CRSM_IRQ_STATUS (see the [System Interrupt Status Register](#page-62-0) section).

System Error Interrupts

The ADIN1100 can also generate system errors interrupts. The interrupt flags are located within the reserved bit sections of CRSM_IRQ_STATUS (see the [System Interrupt Status Register](#page-62-0) section).

CRSM_IRQ_MASK (see the [System Interrupt Mask Register](#page-62-0) section) must be configured to allow system error interrupts. Refer to [Table 94](#page-62-0) for details on the interrupts mask. The ADIN1100 must be hardware reset to recover from a system error interrupt (CRSM_IRQ_STATUS reserved bits read as 1).

RESET OPERATIONS

Overview

The ADIN1100 supports the following chip resets:

- ► Power-on reset
- ► Hardware reset
- ► Software reset
- ► MAC interface reset
- ► PHY subsystem reset

All of these resets put the ADIN1100, including the PHY core, into a known state. Whenever the PHY core is reset, the ADIN1100 MAC interface output pins are driven to a low state.

Power-On Reset

The ADIN1100 includes a power supply monitoring circuit to ensure that the chip has the correct voltage supply before initiating the power-up sequence. During power-up, the ADIN1100 is held in hardware reset until each of the supplies has crossed its minimum rising threshold value and the power is considered good.

Hardware Reset

A hardware reset is initiated by the power-on reset circuitry or by asserting the RESET pin low for a minimum of 10 µs. The ADIN1100 includes a deglitch circuitry on this pin to reject pulses shorter than 1 µs.

When the RESET pin is deasserted, all the input/output (I/O) pins are held in tristate mode, the hardware configuration pins are latched, and the I/O pins are configured to their functional mode. When all the external and internal supplies are valid and stable, the crystal oscillator circuit is enabled. After the crystal has started up and stabilized, the phase-locked loop (PLL) is enabled. After a delay of 50 ms (maximum) from the deassertion of the RESET pin, all the internal clocks are valid, the internal logic is released from

reset, and all the management interface registers are accessible over the MDIO interface.

The CLK25_REF clock output stays low while the RESET pin is asserted low and remains low for another 70 ms (maximum) after the RESET pin is deasserted.

Software Reset

A full chip software reset is initiated by setting the software reset bit (CRSM_SFT_RST). When this bit is set, the chip fully initializes, almost equivalent to a hardware reset except that it does not go through the voltage supply validation sequence. The I/O pins are held in tristate mode, the hardware configuration pins are latched, and then the I/O pins are configured to their functional mode. The crystal oscillator circuit is enabled, and after the crystal has started up and stabilized, the PLL is enabled. Approximately 10 ms (maximum) after setting the CRSM_SFT_RST bit, the internal logic is released from reset and all the management interface registers are accessible. The system ready bit (CRSM_SYS_RDY) indicates that the start-up sequence is complete and the system is ready for normal operation.

The CLK25 REF clock output remains low for 25 ms (maximum) following a software reset.

PHY Subsystem Reset

A PHY subsystem reset is initiated on the ADIN1100 by setting the PHY subsystem reset register bit (CRSM_PHY_SUBSYS_RST) to 1. The reset is applied for typically 1.2 µs, and then this bit self clears. All of the PHY digital circuitry is reset, and any available active link drops. The PHY subsystem reset does not alter the values of the management registers, which remain accessible throughout the sequence. The subsystem reset is a short reset and can be used to put the device into a known state while retaining the internal register contents.

MAC Interface Reset

A MAC interface reset is initiated on the ADIN1100 by setting the PHY MAC interface reset register bit (CRSM_MAC_IF_RST) to 1. The reset is applied for typically 1.2 μ s, and then this bit self clears. A reset sequence is provided to the ADIN1100 MAC interface, but without dropping the available active link. This reset interrupts any packet transmission or reception on the MAC interface, but does not drop an existing active link nor prevent a link from being established. The MAC interface reset does not alter the values of the management registers, which remain accessible throughout the sequence.

STATUS LEDS

Overview

The LED 0 and LED 1 is only available in ADIN1100 pins can be used to connect external LEDs to indicate the ADIN1100 link status and transmit or receive activity. The activity assigned to each LED is configurable through LED_CNTRL (see the [LED Control Register](#page-67-0) section).

The LED pins are suitable for ultra low power LEDs. The maximum output current for the LED 0 and LED 1 pins is 8 mA with a VDDIO = 3.3 V. For higher LED power requirements, the use of an external transistor is recommended, as described in the [Transistor](#page-34-0) [Controlled LED](#page-34-0) section.

The LED x pins can also be connected to a host microcontroller GPIO (configured as a pulse-width modulated input or hardware interrupt). This configuration can be useful in applications where the user interface must be fully handled by an external host controller (for example, an external LED module or display).

If the LED 0 and LED 1 pins are directly connected to a host controller, it is recommended to place a low value resistance in series between the ADIN1100 LED x pins and the host controller to avoid any potential current surge. The resistor value must be defined based on host controller capabilities and the ADIN1100 LED x pin output current capabilities listed in [Table 1](#page-3-0).

LED Pin Multiplexing

For the LED 1 pin only, an internal multiplexer must be configured to enable the LED 1 signal on the pin. LED 1 is disabled by default and can be enabled using the DIGIO_LED1_PINMUX bits (see the [Pin Mux Configuration 1 Register](#page-65-0) section).

The LED 0 pin does not need multiplexing.

LED Polarity

The LED 0 and LED 1 pins can be configured to support various LED circuit polarities through the LED polarity mode feature (see

Table 13. LED_x Pins Configuration Summary

the [LED Polarity Register](#page-69-0) section). Three polarity modes are available for each LED, as follows:

- ► Autosense (default)
- ► Active high
- ► Active low

In autosense mode, the ADIN1100 automatically senses the pin at power-up or reset to select the appropriate polarity configuration.

In active high mode, the ADIN1100 is configured to drive the LED from the anode side.

In active low mode, the ADIN1100 is configured to drive the LED from the cathode side.

Example circuits are described in the [LED Circuit Examples](#page-33-0) section.

LED Function

LED 0 and LED 1 can be configured to display various activities of the ADIN1100 using the LED function feature. The LED function is configurable using the LED0_FUNCTION and LED1_FUNCTION bits (see the [LED Control Register](#page-67-0) section).

Note that the 7, 8, 9, and 10 (decimal) bit settings for LEDx_FUNC-TION are not available in LED Mode 2.

LED Mode

LED 0 and LED 1 activity behavior can be configured using the two LED modes, as follows:

- \blacktriangleright LED Mode 1: blink duty cycle defined using the LED0_BLINK_TIME_CNTRL (see the [LED_0 On/Off Blink Time](#page-66-0) [Register](#page-66-0) section)
- ► LED Mode 2: blink duty cycle automatically defined by the ADIN1100 based on activity level (%)

LED_x Pins Configuration Summary

See Table 13 for the configuration options of the LED x pins.

¹ The 7, 8, 9, and 10 (decimal) settings in the LEDx_FUNCTION bits are not available in Mode 2.

² See [Table 1](#page-3-0) for details.

LINK STATUS PIN

Overview

The link status pin (LINK_ST/PHYAD_2) is asserted high when the link status bit (AN_LINK_STATUS) is asserted and indicates that the link between the ADIN1100 and its link partner is active.

The LINK_ST/PHYAD_2 pin has a weak internal pull-down resistor. The pin is also used as a hardware configuration pin signal (PHYAD_2) during power-up, hardware reset, or software reset.

Typical Use

The link status pin can be used to connect an external LED or can be connected to a host microcontroller GPIO (configured as a pulse-width modulated input or hardware interrupt).

By default, the LINK_ST signal is active high and can be configured as either active high or low using the link status polarity bit (DGIO_LINK_ST_POLARITY). See the [Pin Mux Configuration 1](#page-65-0) [Register](#page-65-0) section.

The link status pin is not intended to source current. Use the circuit recommendation in the [Transistor Controlled LED](#page-34-0) section as a reference to interface an LED on this pin. If the link status pin is directly connected to a host controller, it is recommended to place a low value resistance in series between the ADIN1100 link status pin and the host controller to avoid any potential current surge. The resistor value must be defined based on host controller capabilities.

POWER-DOWN MODES

The ADIN1100 supports two power-down modes.

- ► Hardware power-down
- ► Software power-down

Hardware Power-Down Mode

The hardware power-down mode can be used when no operation is required on the ADIN1100 and the power consumption needs to be minimized.

The device enters hardware power-down mode when the RESET pin is asserted and held low. In this mode, all analog and digital circuits are disabled, the clocks are gated off, and all the I/O pins are held in tristate mode.

In this mode, the ADIN1100 power consumption is equivalent to the internal circuit leakage. The internal registers are not accessible in this mode.

Software Power-Down Mode

The software power-down mode can be used to configure the ADIN1100 registers before bringing a link up. In this mode, the analog and digital circuits are in a low power state, and the PLL is active and can provide output clocks if configured to do so. Any signals exposed to the MDI pins (TXP, TXN, RXP, RXN) are ignored and any active link is dropped. The MAC interface output pins are asserted low and internal registers are accessible using the MDIO interface.

The device can be configured to automatically enter software power-down mode after power-up, hardware reset, or software reset using the **SWPD_EN** hardware configuration pin signal. The ADIN1100 can also be instructed to enter software power-down mode by setting the software power-down bit (CRSM_SFT_PD).

The software power-down status bit (CRSM_SFT_PD_RDY) indicates that the device is in software power-down mode.

The ADIN1100 exits software power-down mode when the CRSM_SFT_PD bit is cleared. After exiting software power-down and if autonegotiation is completed, the device attempts to bring a link up.

HARDWARE CONFIGURATION PINS

OVERVIEW

The ADIN1100 can operate in unmanaged or managed configurations with the use of the hardware configuration pins.

The hardware configuration pins are standard pins with an alternate bootstrap function. The ADIN1100 reads the configuration pin level immediately after power-up, hardware reset, or software reset, and configures the PHY settings accordingly. When active, the ADIN1100 immediately attempts to bring up a link and the hardware configuration pins can then be used with their main pin function. These pins can be used in unmanaged or managed configuration.

The unmanaged configuration refers to the ADIN1100 operating in standalone mode. This mode can be used when the system requires a static configuration of the ADIN1100 without the need for software control and an external host controller.

The managed configuration refers to the use of an external host such as a microcontroller to control and manage the ADIN1100 by software over the MDIO interface. The configuration pins can be connected to the external host or hardware configured using pull-up/pull-down resistors. When active, the host controller can override any of the ADIN1100 hardware configurations set by the hardware pins after power-up, hardware reset, or software reset.

UNMANAGED APPLICATIONS

In unmanaged applications, it is possible to configure the desired operation of the ADIN1100 using hardware configuration pins without any software intervention. The hardware configuration pins set the default values of the corresponding management registers after power-up, hardware reset, or software reset.

Software power-down after reset must be disabled for unmanaged applications or the ADIN1100 remains in power-down indefinitely because the device can only exit power-down from register operation using the management interface (see the [Software Power-](#page-21-0)[Down Mode](#page-21-0) section).

MANAGED APPLICATIONS

In managed applications, the ADIN1100 can be configured by a host controller via the management interface. The host controller can dynamically configure the device as required by the application.

In managed applications, the software power-down after reset functionality can be enabled as the host controller brings the ADIN1100 to active mode using the management interface.

HARDWARE CONFIGURATION PINS FUNCTIONS

Overview

The following functions are configurable from the ADIN1100 hardware configuration pins:

► PHY address

- ► Software power-down mode after reset
- ► Transmit amplitude configuration
- ► Master/slave selection
- ► MAC interface selection (RGMII/RMII/MII)
- ► Media converter operation

All of the hardware configuration pins have internal pull-down resistors. The default mode of operation is shown in Table 14. If an alternative mode of operation is required, refer to Table 15 for the suggested external pin control.

Table 15. Recommended Control for Hardware Configuration Pins

¹ A low value series resistor is recommended.

² External pull-down resistor is recommended.

Media Converter

The ADIN1100 can operate as a media converter, which allows a 10BASE-T PHY to be connected directly to the ADIN1100 via the RMII or RGMII. The ADIN1100 can connect to a 10BASE-T1L remote PHY via the MDI pins.

In RMII mode, the MEDIA CNV hardware configuration strap pin can be used to enable the media converter functionality by default after reset or power cycle. Alternatively, the media converter can be configured via the CRSM_RMII_MEDIA_CNV_EN bits (see the [MAC Interface Configuration Register](#page-63-0) section).

In RGMII mode, the ADIN1100 can be directly connected to another RGMII device without any media configuration requirements.

The MEDIA CNV hardware configuration pin signal is shared with TXD 3, which is not used in RGMII mode (see the [MAC Inter](#page-12-0)[face](#page-12-0) section). The TXD_3/MEDIA_CNV pin has a weak internal pull-down resistor. Thus, by default, the ADIN1100 is configured for normal PHY operation (for example, external MAC chip connected on the MAC interface). An external pull-up resistor or external host

HARDWARE CONFIGURATION PINS

control via a GPIO must be used to select the media converter operation in RMII mode.

Table 16. Media Converter Selection (Hardware Configuration)

PHY Address Configuration

The ADIN1100 PHY address can be configured using three pins:

- ► RXD_2/PHYAD_0
- ► RXD_3/PHYAD_1
- ► LINK_ST/PHYAD_2

These are two-level configuration pins, which means that it is possible to configure the ADIN1100 to any of the eight available PHY addresses. The PHY address pins have weak internal pull-down resistors. Thus, by default, the ADIN1100 is configured with PHY Address 0x0.

Particular attention is required if these three pins are used in managed applications and connected to an external host controller because the PHY address is set after power-up, hardware reset, or software reset.

Software Power-Down After Reset

The SWPD_EN hardware configuration pin signal is shared with the RX DV signal and is used to enable or disable the software power-down after reset feature. The power-down bit (CRSM_SFT_PD) is set based on the SWPD_EN signal status during power-up, hardware reset, or software reset. CRSM_SFT_PD can also be set using the MDIO interface to enable software power-down after reset.

The RX_DV/RX_CTL/SWPD_EN pin has a weak internal pull-down resistor. Thus, by default, the ADIN1100 is configured with software power-down after reset enabled.

When software power-down after reset is enabled, the ADIN1100 enters software power-down after power-up, hardware reset, or software reset. Software power-down provides a lower power mode where most of the ADIN1100 internal modules are turned off.

The ADIN1100 can be configured to exit power-down by setting the CRSM_SFT_PD bit to 0 using the MDIO interface.

Following a power-up, hardware reset, or software reset, and if autonegotiation is enabled and software power-down after reset is disabled, the ADIN1100 starts autonegotiation and tries to bring up a link.

Master/Slave Preference

The MS_SEL hardware configuration pin signal is shared with the RXD 1 signal and configures the default master/slave selection. If MS SEL is pulled low during power-up, hardware reset, or software reset, the device is configured by default to prefer slave. If MS_SEL is pulled high during power-up, hardware reset, or software reset, the device is configured by default to prefer master.

The RXD_1/MS_SEL pin has a weak internal pull-down resistor. Thus, by default, the ADIN1100 is configured to prefer slave.

The MS SEL hardware configuration pin signal configures the default setting of the autonegotiation master/slave configuration register bit (AN_ADV_MST). MS_SEL also configures the default setting of the master slave configuration register bit (CFG_MST), which is used when autonegotiation is disabled.

The AN_ADV_MST and CFG_MST bits can be modified using the MDIO interface but return to their default values set by the MS_SEL hardware configuration pin signal after power-up, hardware reset, or software reset.

Transmit Amplitude

The TX2P4_EN hardware configuration pin signal is shared with the RXD_0 signal and is used to configure the default transmit amplitude mode. The transmit amplitude mode is defined by the pin status during power-up, hardware reset, or software reset as defined in [Table 19](#page-24-0).

The TX2P4 EN hardware configuration pin signal configures the default setting of the high voltage transmit ability bit (B10L_TX_LVL_HI_ABLE).

The RXD_0/TX2P4_EN pin has a weak internal pull-down resistor. Thus, by default, the ADIN1100 is configured to support both 1.0 V p-p and 2.4 V p-p voltage levels.

If the RXD $0/\overline{TX2P4}$ EN pin is strapped high (1.0 V p-p only), the associated register cannot be changed through the MDIO interface. That is, 2.4 V p-p operation is not possible if the ADIN1100 has been hardware pin configured only for 1.0 V p-p.

HARDWARE CONFIGURATION PINS

Table 19. Transmit Amplitude Selection (Hardware Configuration)

If TX2P4 EN is pulled low during power-up, hardware reset, or software reset, the 2.4 V p-p transmit operating mode is enabled and the value of B10L_TX_LVL_HI_ABLE is set to 1.

If TX2P4_EN is pulled high during power-up, hardware reset, or software reset, the 2.4 V p-p transmit operating mode is disabled, and the value of B10L_TX_LVL_HI_ABLE is set to 0.

The B10L_TX_LVL_HI_ABLE bit reports whether the PHY is capable of operating in the 10BASE-T1L high transmit voltage mode, as described in Table 20.

Table 20. B10L_TX_LVL_HI_ABLE Settings

MAC Interface Selection

The MAC interface hardware configuration pin signals (MA-CIF SELx) are shared with the RX CLK/RXC and RX ER signals and can be configured according to Table 21. The RX_CLK/RXC/ MACIF_SEL0 and RX_ER/MACIF_SEL1 pins have weak internal pull-down resistors. Thus, by default, the ADIN1100 is configured in RMII mode. External pull-up/pull-down resistors or host control via a GPIO must be used to select the RGMII or MII MAC interface mode.

Table 21. MAC Interface Selection (Hardware Configuration)

BRINGING UP 10BASE-T1L LINKS

The following sections provide some recommendations on how to bring a link up between the ADIN1100 and a remote link partner. The sections cover various configurations and some may not be relevant to the intended application. Refer to the [Theory of Operation](#page-12-0) section for more detailed explanations.

UNMANAGED PHY OPERATION

For an unmanaged PHY where there is no control of the ADIN1100 over the management interface, the hardware configuration pins determine the operating mode. See the [Hardware Configuration](#page-22-0) [Pins](#page-22-0) section for more details on how to use the hardware configuration pins. The following sections describe the steps required to bring up a link in unmanaged applications.

Set the PHY Address

The PHY address can be selected by asserting the PHY address pins: RXD_2/PHYAD_0, RXD_3/PHYAD_1, and LINK_ST/ PHYAD_2.

When it exits reset, the ADIN1100 starts autonegotiation and tries to bring up a link after autonegotiation completes.

See the [PHY Address Configuration](#page-23-0) section for details on how to use the address pins.

Disable Software Power-Down Mode After Reset

The software power-down mode must be disabled in unmanaged applications. Otherwise, the ADIN1100 remains in software powerdown indefinitely. Assert the RX_DV/RX_CTL/SWPD_EN pin high during power-up and reset so that the PHY does not enter software power-down mode when it exits reset.

See the [Software Power-Down After Reset](#page-23-0) section for details on how to configure the software power-down after reset function.

Master/Slave Selection

The RXD 1/MS SEL pin is used to configure the PHY to advertise prefer slave or prefer master.

See the [Master/Slave Preference](#page-23-0) section for details on how to configure the master/slave setting.

Set Transmit Amplitude Level

The RXD_0/TX2P4_EN pin configures the PHY to advertise the support of both 1.0 \overline{V} p-p and 2.4 \overline{V} p-p transmit level operation or to only advertise support of 1.0 V p-p transmit level operation.

By default, the ADIN1100 is configured to support 1.0 V p-p and 2.4 V p-p transmit levels due to the internal pull-down resistor. Assert the relevant pin high or low to disable the support for the 2.4 V p-p transmit level.

See the [Transmit Amplitude s](#page-23-0)ection for details on how to configure the transmit amplitude level.

Select the MAC Interface (RGMII/RMII/MII)

The MAC interface type can be selected using the RX_CLK/RXC/ MACIF_SEL0 and RX_ER/MACIF_SEL1 pins.

See the [MAC Interface Selection](#page-24-0) for details on how to select the MAC interface.

Enable the Media Converter Functionality

Assert the TXD 3/MEDIA CNV pin high if media converter function is required (RMII mode only).

See the [Media Converter](#page-22-0) section for details on how to enable the media converter functionality.

MANAGED PHY OPERATION

In a managed PHY application, a host controller such as a microcontroller is used to configure the ADIN1100 operation in software via the management interface (MDIO).

Similar to the unmanaged PHY operation, the hardware configuration pins can be used to set up the controlled ADIN1100 (see the Unmanaged PHY Operation section for details). Alternatively, the hardware configuration pins can directly be controlled by the host (for example, GPIO) via an external pull-up or pull-down resistor or both.

In managed applications, the software power-down after reset can be enabled. The ADIN1100 stays in software power-down mode until the software has configured the PHY to be active. When active, the PHY can then start autonegotiation and try to bring up a link.

Power-Up and Reset Complete

A typical way for software to verify that the device has completed the power-up and reset sequence and is available for normal operation is to read the management register that has the IEEE organizationally unique identifier (OUI), model, and revision numbers. The value of this register is unique to each PHY vendor and is a nonzero value. If the device has not completed the power-up, the value does not read correctly. In legacy BASE-T PHYs, this value is at management interface Register Address 0x2 and Register Address 0x3.

In the ADIN1100, the OUI, model number, and revision numbers can also be read at Device Address 0x1F Clause 45 only), Register Address 0x2, and Register Address 0x3 (Clause 22 and Clause 45).

MMD1 DEV ID1 contains the OUI, Bits[3:18] (see the [Vendor](#page-60-0) [Specific 1 MMD Identifier High Register](#page-60-0) section).

BRINGING UP 10BASE-T1L LINKS

MMD1 DEV ID2 contains the OUI, Bits[19:24] (MMD1_DEV_ID2_OUI), the model number (MMD1_MOD-EL_NUM), and the revision number (MMD1_REV_NUM). See the [Vendor Specific 1 MMD Identifier Low Register](#page-61-0) section.

Table 22. ADIN1100 Unique Identifier Values

When a valid read of the IEEE OUI is done, the system ready bit (CRSM_SYS_RDY) can also be read to verify that the start-up sequence is complete and the system is ready for normal operation.

The software power-down status bit (CRSM_SFT_PD_RDY) can be read to check if the device is in the software power-down state. This bit is also controlled by the SWPD_EN hardware configuration pin signal.

Configuring the Device for Linking

After power-up or reset, configure the ADIN1100 for the desired operation for linking. The ADIN1100 may already be configured as required by the hardware configuration pins, but greater control is available using the management registers.

The autonegotiation process is used to match the operating mode between a local and remote PHY. For example, autonegotiation is used to ensure that the modes agree between the two devices on which PHY operates as master and which as slave. Autonegotiation is also used to match the transmit level between the two PHYs.

Autonegotiation is enabled by default for the ADIN1100, and it is strongly recommended to always keep Autonegotiation enabled. Autonegotiation is defined by the IEEE standard and includes a number of mechanisms to ensure robust linking operation between PHYs and is the fastest way to bring up a link.

Configuration of Transmit Level Mode

Overview

The ADIN1100 can support transmit level operation at either 1.0 V p-p or 2.4 V p-p if the B10L_TX_LVL_HI_ABLE bit is set to 1 and a 3.3 V supply is provided on the AVDD_H pins. The higher transmit level can support longer reach but also has higher power consumption.

The ADIN1100 can support 1.0 V p-p transmit level operation with a 1.8 V supply on the AVDD H pins at very low power consumption.

The ADIN1100 can either be configured to advertise support of both 1.0 V p-p and 2.4 V p-p transmit level operation or to advertise

support of only 1.0 V p-p transmit level operation. Refer to the Transmit Level Mode Advertisement section for more details.

1.0 V p-p transmit level operation is required for intrinsically safe operation.

Enable High Voltage Transmit Ability

The high voltage transmit ability is set using the TX2P4_EN hardware configuration pin signal, which internally sets the high voltage transmit ability bit, B10L_TX_LVL_HI_ABLE (read only), as descri-bed in the [Transmit Amplitude s](#page-23-0)ection.

Enable 1.0 V p-p and 2.4 V p-p Transmit Levels

To allow both 1.0 V p-p and 2.4 V p-p transmit level operation, set AN_ADV_B10L_TX_LVL_HI_ABL to 1 to indicate that the device is capable of $2.\overline{4}$ V p-p transmit level operation (a 3.3 V supply is required on the AVDD H pins).

Set 2.4 V p-p Transmit Level as Preferred

If 2.4 V p-p transmit level operation is preferred, set AN_ADV_B10L_TX_LVL_HI_REQ to 1 to request 2.4 V p-p transmit level operation. Note that autonegotiation determines the transmit level that the link operates at.

Set 1.0 V p-p Transmit Level as Preferred

If 1.0 V p-p transmit level operation is preferred, set AN_ADV_B10L_TX_LVL_HI_REQ to 0. Autonegotiation determines the transmit level that the link operates at.

Enable 1.0 V p-p Transmit Level Only

If it is required to only operate the PHY at 1.0 V p-p transmit level operation, set AN_ADV_B10L_TX_LVL_HI_ABL to 0 so that 2.4 V p-p transmit level operation is not advertised.

In this case, autonegotiation can only resolve to 1.0 V p-p transmit level operation, irrespective of the setting that the remote PHY advertises. For very long cable lengths, depending on the characteristics of the cable, it may not be possible to bring up a link at 1.0 V p-p operation.

When the high level transmit is disabled through the RXD D0/ TX2P4 EN pin, the AVDD H supply can be supplied from either 1.8 V or 3.3 V for 1.0 V p-p transmit level operation.

Transmit Level Mode Advertisement

Enable High Voltage Transmit Ability

The AVDD H power rail must be provided with a 3.3 V supply for the ADIN1100 to support the 2.4 V p-p transmit level.

The high voltage transmit ability is enabled on the ADIN1100 by setting the $\overline{TX2P4}$ EN hardware configuration pin signal low during power-up, hardware reset, or software reset. The transmit ability

BRINGING UP 10BASE-T1L LINKS

bit, B10L_TX_LVL_HI_ABLE (read only), is set automatically to the defined hardware configuration as follows:

- \triangleright B10L TX LVL HI ABLE = 0: 1.0 V p-p only ability
- \triangleright B10L TX LVL HI ABLE = 1: 1.0 V p-p and 2.4 V p-p ability

See the [Configuration of Transmit Level Mode](#page-26-0) section for more details.

Advertise High Voltage Transmit Ability

Set the AN_ADV_B10L_TX_LVL_HI_ABL bit to 1 to advertise the high level transmit mode to the link partner during autonegotiation. This bit can only be set if the ADIN1100 has the ability to transmit in high voltage mode (B10L TX LVL HI ABLE = 1).

High voltage transmit ability only enables the ADIN1100 to advertise support for both 2.4 V p-p and 1.0 V p-p levels. The selected level is determined by autonegotiation with the link partner.

See the [Transmit Amplitude Advertisement](#page-14-0) section for more details.

Advertise a Request for High Voltage Transmit Level

Set the AN_ADV_B10L_TX_LVL_HI_REQ bit to 1 to advertise a request for 2.4 V p-p transmit level operation during autonegotiation. This bit can only be set if the ADIN1100 has the ability to transmit in high voltage mode (B10L TX LVL HI ABLE = 1).

See the [Transmit Amplitude Advertisement](#page-14-0) section for more details.

Read Link Partner Advertised Transmit Level

The link partner advertised transmit information can be read using the link partner high level transmit operating mode ability bit (AN_LP_ADV_B10L_TX_LVL_HI_ABL) and the link partner high level transmit operating mode request bit (AN_LP_ADV_B10L_TX_LVL_HI_REQ). These bits are valid when the autonegotiation is completed (AN_COMPLETE = 1).

See the [Transmit Amplitude Advertisement](#page-14-0) section for more details.

Completion of Autonegotiation

When autonegotiation has completed, the autonegotiation complete indication register bit (AN_LINK_GOOD) is set. This bit indicates the completion of the autonegotiation sequence and that the enabled PHY link is setting up or active.

When autonegotiation has completed and the link is up, the autonegotiation complete register bit (AN_COMPLETE) is set to 1 and the contents of the following registers are valid:

- ► BASE-T1 autonegotiation advertisement registers
	- ► AN_ADV_ABILITY_L: Bits[15:0]
	- ► AN_ADV_ABILITY_M: Bits[31:16]
	- ► AN_ADV_ABILITY_H: Bits[47:32]
- ► BASE-T1 autonegotiation link partner base page ability registers ► AN LP ADV ABILITY L: Bits[15:0]
	- ► AN_LP_ADV_ABILITY_M: Bits[31:16]
	- ► AN_LP_ADV_ABILITY_H: Bits[47:32]

Link Status

The status of the link can be determined by reading the link status register bit (AN_LINK_STATUS). This bit latches low.

When read as 1, this bit indicates that a valid link has been established.

If this bit reads 0, it means that the link has failed since the last time it was read. This bit latches low. Thus, if a 0 is read, this bit must be read a second time to determine if the link status has come up in the interim (see the [Latch Low Registers](#page-39-0) section).

If the link is dropped, the autonegotiation process restarts automatically. Autonegotiation can be restarted by request through a write to the autonegotiation restart bit (AN_RESTART) in the AN_CON-TROL register (see the [BASE-T1 Autonegotiation Control Register](#page-53-0) section).

LOOPBACK MODES

The PHY core provides the following loopback modes:

- ► Physical medium attachment (PMA) loopback
- ► Physical coding sublayer (PCS) loopback
- ► MAC interface loopback
- ► MAC interface remote loopback

These loopback modes test and verify various functional blocks within the PHY. The use of a frame generator and frame checkers allows completely self contained in-circuit testing of the digital and analog data paths within the PHY core.

PMA Loopback

In PMA loopback, the MDI interface must be left open circuit, thereby transmitting into an unterminated connector or cable. In this mode, the signal transmitted from the ADIN1100 is echoed back from the open 10BASE-T1L MDI. This test mode is an implementation of the PMA local loopback function defined in Subclause 146.5.6 of the IEEE Standard 802.3cg. Remove any cable connected to the MDI interface to improve the test mode accuracy.

In PMA loopback mode, the device must be configured in forced link configuration mode (autonegotiation disabled). To enable PMA loopback, set the 10BASE-T1L PMA loopback enable bit (B10L_LB_PMA_LOC_EN) to 1 (see the [10BASE-T1L PMA Con](#page-49-0)[trol Register](#page-49-0) section).

PCS Loopback

PCS loopback mode loops the transmit data back to the receiver within the PCS block at the input stage of the PHY digital block. Setting the B10L_LB_PCS_EN bit (B10L_PCS_CNTRL register) enables PCS loopback.

When the PCS loopback mode is enabled, no signal is transmitted to the MDI pins.

MAC Interface Loopback

MAC interface loopback mode loops the data received on the MAC interface transmit data pins (TXD_x signals) back to the receive data pins (RXD_x signals) and can therefore be used to verify MAC interface connectivity. Set the MAC_IF_LB_EN bit to 1 (see the [MAC Interface Loopbacks Configuration Register](#page-77-0) section) to enable MAC interface loopback.

If the MAC_IF_LB_TX_SUP_EN bit within the same register is set (enabled by default), the transmission of the signal received on the MAC interface is not transferred to the ADIN1100 PHY core.

MAC Interface Remote Loopback

MAC interface remote loopback requires a link up with a remote PHY and enables looping of the data received on the ADIN1100 to the remote PHY. This linking allows a remote PHY to verify a complete link by ensuring that the PHY receives the proper data. Set the MAC_IF_REM_LB_EN bit to 1 (see the [MAC Interface](#page-77-0) [Loopbacks Configuration Register](#page-77-0)) to enable MAC interface remote loopback.

If the MAC_IF_REM_LB_RX_SUP_EN bit (see the [MAC Interface](#page-77-0) [Loopbacks Configuration Register](#page-77-0) section) is set (set by default), the data received by the ADIN1100 from the MDI pins is not transferred through the MAC interface.

External MII/RMII Loopback

Overview

When configured in external MII/RMII loopback mode, the ADIN1100 echoes back the data received from a remote PHY. The external MII/RMII loopback is performed by physically connecting the MAC transmission signal pins to the MAC reception signal pins. The wiring configuration required is described in Table 23.

Table 23. External MII/RMII Loopback Wiring

 1 MII only.

² MII reference name.

³ RMII reference name.

Software Configuration

For the external RMII loopback, the RMII TXD check enable bit, RMII_TXD_CHK_EN (see the [RMII Configuration Register](#page-77-0) section), must be set to 1 so that CRS DV can be connected to TX EN.

The external MII loopback does not require any particular register bits to be set to be enabled.

Figure 14. ADIN1100 Loopback Modes

FRAME GENERATOR AND CHECKER

Overview

The ADIN1100 can be configured to generate frames and to check received frames (see [Figure 15](#page-30-0)). The generating and checking functions can be used together or independently. If the ADIN1100 transmitted frames are looped back at the remote end, the frame checker can be used to check the echoed self generated frames.

Frame Generator

When the frame generator is enabled, the MAC interface is ignored and the frame generator data is used for transmission on the MDI pins. To use the frame generator, the diagnostic clock must also be enabled using the [CRSM](#page-64-0)_DIAG_CLK_EN bit (see the CRSM [Diagnostics Clock Control Register](#page-64-0) section).

The frame generator control registers configure the type of frames to be sent (for example, random data, all 1s), the frame length, and the number of frames to be generated.

The generation of the requested frames starts by enabling the frame generator by setting the FG EN bit (see the [Frame Genera](#page-74-0)[tor Enable Register](#page-74-0) section).

When the generation of the frames is completed, the frame generator done bit, FG_DONE, is set (see the [Frame Generator Done](#page-76-0) [Register](#page-76-0) section).

Frame Checker

The frame checker is enabled by setting the frame checker enable bit, FC_EN (see the [Frame Checker Enable Register](#page-71-0) section). The frame checker can be configured to check and analyze received frames from either the MAC interface or the PHY using the frame checker transmit select bit, FC_TX_SEL (see th[eFrame Checker](#page-72-0) [Transmit Select Register](#page-72-0) section). The frame checker reports the number of frames received, cyclic redundancy check (CRC) errors, and various frame errors. The frame checker frame counter registers and the frame checker error counter registers count these events.

Error Counters

The frame checker counts the number of CRC errors, and these errors are reported in the receive error counter register (RX_ERR_CNT). To ensure synchronization between the frame checker error counter and frame checker frame counters, all of the counters are latched when the receive error counter register is read. Therefore, when using the frame checker, read the receive error counter first, and then read all other frame counters and error counters. A latched copy of the receive frame counter register is available in FC_FRM_CNT_H (see the [Frame Checker Count High](#page-72-0) [Register](#page-72-0) section) and FC_FRM_CNT_L (see the [Frame Checker](#page-72-0) [Count Low Register](#page-72-0) section).

In addition to CRC errors, the frame checker counts frame length errors, frame alignment errors, symbol errors, oversized frames errors, and undersized frame errors.

The frame checker also counts frames with an odd number of nibbles in the frame, and counts packets with an odd number of nibbles in the preamble.

The frame checker also counts the number of false carrier events, which is a count of the number of times the bad start of stream delimiter (SSD) state is entered.

FRAME GENERATOR AND CHECKER LINK TEST

Using the ADIN1100 and a second PHY device, the user can configure a convenient, self contained validation setup of the PHY to PHY connection. [Figure 15](#page-30-0) shows an overview of how each PHY is configured. An external cable is connected between both devices, and PHY 1 is generating frames using the frame generator. PHY 2 has MAC interface remote loopback enabled using MAC_IF_REM_LB_EN (see the [MAC Interface Loopbacks](#page-77-0) [Configuration Register](#page-77-0) section).

- **1.** The PHY 1 frames (frame generator) are sent over the 10BASE-T1L single pair cable.
- **2.** PHY 2 receives frames on the PHY 2 MDI pins.
- **3.** The PHY 2 MAC interface loops the frame back.
- **4.** The PHY 2 frames (looped back) are sent over the 10BASE-T1L single pair cable.
- **5.** PHY 1 receives the PHY 2 frames (looped back) on the MDI pins.
- **6.** The PHY 1 frame checker checks the received frames.

Figure 15. Remote Loopback Used Across Two PHYs for Self Check Purposes

TEST MODES

The ADIN1100 provides several test modes as described in Subclause 146.5.2 from the IEEE 802.3cgTM-2019 standard that allows testing of the transmitter waveform, distortion, jitter, and droop. These test modes change only the data symbols provided to the transmitter circuitry and do not alter the electrical and jitter characteristics of the transmitter and receiver from the normal operation.

Additionally, the ADIN1100 supports the transmit disable mode as described in Subclause 45.2.1.186a.2.

Enable the PMA Test Mode 1 to Test Mode 3

The ADIN1100 can be configured in one of the PMA test modes (Test Mode 1 to Test Mode 3) using the following procedure:

- **1.** Enter software power-down mode by writing a 1 to the CRSM_SFT_PD bit in the CRSM_SFT_PD_CNTRL register (see the [Software Power-Down Control Register](#page-62-0) section).
- **2.** Check that the ADIN1100 has entered software power-down mode by reading the CRSM_SFT_PD_RDY bit in the CRSM_STAT register (see the [System Status Register](#page-63-0)).
- **3.** Disable autonegotiation by writing a 0 to the AN_EN bit in the AN_CONTROL register (see the [BASE-T1 Autonegotiation](#page-53-0) [Control Register](#page-53-0) section).
- **4.** Set autonegotiation forced mode by writing a 1 to the AN_FRC_MODE_EN bit in the AN_FRC_MODE_EN register (see the [Autonegotiation Forced Mode Enable Register](#page-59-0)).
- **5.** Select the desired test mode by writing the appropriate value to the B10L_TX_TEST_MODE bits in the B10L_TEST_MODE_CNTRL register (see the [10BASE-T1L](#page-50-0) [Test Mode Control Register](#page-50-0) section). Table 25 outlines the bit settings for each PMA test mode.

6. Exit software power-down mode by writing 0 to the CRSM_SFT_PD bit in the CRSM_SFT_PD_CNTRL register (see the [Software Power-Down Control Register](#page-62-0) section).

Table 25. PMA Test Modes Configuration

Enable Transmit Disable Mode

[System Status Register](#page-63-0)The ADIN1100 can be configured in transmit disable mode using the following procedure:

- **1.** Enter software power-down mode by writing a 1 to the CRSM_SFT_PD bit in the CRSM_SFT_PD_CNTRL register (see the [Software Power-Down Control Register](#page-62-0) section).
- **2.** Check that the ADIN1100 has entered the software powerdown mode by reading the CRSM_SFT_PD_RDY bit in the CRSM STAT register (see the [System Status Register](#page-63-0) section).
- **3.** Disable autonegotiation by writing a 0 to the AN_EN bit in the AN_CONTROL register (see the [BASE-T1 Autonegotiation](#page-53-0) [Control Register](#page-53-0) section).
- **4.** Set autonegotiation forced mode by writing a 1 to the AN_FRC_MODE_EN bit in the AN_FRC_MODE_EN register (see the [Autonegotiation Forced Mode Enable Register](#page-59-0) section).
- **5.** Set the transmit disable mode by writing a 1 to the B10L_TX_DIS_MODE_EN bit in the B10L_PMA_CNTRL register (see the [10BASE-T1L PMA Control Register](#page-49-0) section).
- **6.** Exit software power-down mode by writing 0 to the CRSM_SFT_PD bit in the CRSM_SFT_PD_CNTRL register (see the [Software Power-Down Control Register](#page-62-0) section).

TIME DOMAIN REFLECTOMETRY (TDR)

Given that the 10BASE-T1L compliant PHY enables communication over long cables, debugging a faulty cable can become costly and difficult without the right tools. To help with this, Analog Devices 10BASE-T1L products provide a TDR engine that enables cable fault detection, distance to fault, and cable length estimation.

The diagnostics solution is the combination of a highly accurate on-chip TDR engine and a set of algorithms that run on a host microcontroller, allowing maximum flexibility for a wide variety of cables and more advanced cable diagnostic capabilities.

Figure 16. ADIN1100 TDR Engine

Fault Detection with the TDR Engine

The Analog Devices algorithm has a time resolution of 8.3 ns, which translates to a length resolution of less than 1 m and a maximum of 1600 m, with an accuracy of 2%.

This fault detector algorithm is capable of finding open and short fault conditions even when the ADIN1100 is physically connected to another PHY through their MDI, which implies that the link partner PHY is potentially transmitting DME pages. Traditional TDR methods struggle to find faults if other signal sources or noise is also present in the same link. This is not the case of the Analog Devices solution, which makes it suitable for debugging when there is no control over the remote end.

The fault detector algorithm is provided as a C-code library containing the high-level functions required for diagnostics. These functions have been optimized to not utilize any advanced processing so that they can be executed by any low-power microcontroller.

A single function call is sufficient to execute the fault detector. The function returns the type of fault and the distance to the fault in meters from the MDI connector.

TDR Offset Calibration

The library includes a function to calibrate the offset of the TDR measurement. This particular function in the library is useful given that different MDI circuits may introduce variable delays in the signal path, which can contribute to the offset of the length measurement. For instance, an isolation transformer on the MDI is highly likely to introduce a signal delay that corresponds to a couple of meters in length.

This calibration is not required to run the fault detector, and an average value is provided by default. However, it is recommended for short cables if accuracy is required. If this calibration is required, it can be done once in the lab for a specific MDI circuit implementation, and the offset value can then be stored in nonvolatile memory for future use.

To perform this calibration, the MDI port must be left open or shorted. No load or cable can be connected to the MDI port.

Cable Calibration

By default, the algorithm is optimized to support long reach cables compliant with the IEEE 802.3cg standard. However, given the wide variety of cable types, which have different insertion loss, return loss, and signal delay characteristics, the library includes a calibration function that optimizes the algorithm to operate with any cable, and estimates its nominal velocity of propagation (NVP) for more accurate length estimations. The length accuracy mainly depends on the accuracy of the NVP value.

To run this calibration, a cable with a known length must be attached to the MDI port, and its end must be left open or shorted. NVP values are generally between 0.5 and 0.9 and are a property of

the construction of the cable. In general, an average NVP value of approximately 0.65 can be assumed. This calibration is not required to run the fault detector, unless higher length accuracy is needed or if nonstandard cables are utilized. This calibration can be done once in the laboratory for a given cable, and the values can be stored in nonvolatile memory.

Refer to the C-code driver for more information related to the usage of these functions.

Length/Distance to Fault Accuracy

The accuracy of the distance to a fault, or length measurements, mainly depends on the NVP value, which is determined by the accuracy of the cable length used to perform the NVP calibration.

Table 26 provides results for induced faults and distance-to-fault measurements for different cables and lengths. In all cases, the algorithm was successful finding the open or short conditions induced during the test. The NVP value for the Profibus PA cable used in this test was roughly estimated, and the same was used for the Cat5E and Cat6 cables.

APPLICATIONS INFORMATION

SYSTEM LEVEL POWER MANAGEMENT

Transmit Level = 1.0 V p-p

The 1.0 V p-p transmit operating mode supports the spur use case and can operate at a lower AVDD_H supply voltage of 1.8 V. This mode supports intrinsic safe applications.

For applications where the ADIN1100 must operate in 1.0 V p-p transmit operating mode, the RXD 0/TX2P4 EN pin must be tied high via a 4.7 kΩ resistor (see Figure 17). This configuration forces the ADIN1100 to only operate in 1.0 V p-p transmit operating mode and enables the ADIN1100 to operate from a signal supply voltage at a lower voltage rail (for example, 1.8 V), allowing the user to minimize power dissipation in the system.

Figure 17. Supplies and Capacitors for Forced 1 V p-p Transmit Mode

Transmit Level = 2.4 V p-p

The higher transmit operating mode of 2.4 V p-p supports trunk applications and requires a higher AVDD_H supply voltage of 3.3 V. This mode can be used for longer cable lengths in industrial Ethernet environments with high noise levels.

For the ADIN1100 to be able to operate at 2.4 V p-p, the RXD_0/ TX2P4 EN pin must be held low (pin has an internal pull-down resistor). This mode of operation still allows the 1.0 V p-p operating mode to be selected via MDIO or via autonegotiation.

Multiple Supplies Configuration

Figure 18 shows an overview of the proposed power configuration. Note that this configuration requires that AVDD $H = 3.3$ V even if the link is established in 1.0 V p-p transmit operating mode via MDIO or autonegotiation.

Figure 18. Supplies and Capacitors for Multiple Supply 2.4 V p-p and 1.0 V p-p Transmit Mode

Single-Supply Configuration

For single-supply operation, the same rail can be used to supply the ADIN1100 AVDD H, AVDD L, and VDDIO supply rails. The DVDD 1P1 1.1 V rail can be derived internally or alternatively provided by an external 1.1 V rail. This configuration is shown in Figure 19.

Figure 19. Supplies and Capacitors for Single-Supply 2.4 V p-p Transmit Mode

LED CIRCUIT EXAMPLES

The LED_0 and LED_1 pins can be used in various circuit configurations depending on the LED polarity mode selected (see the [LED](#page-69-0) [Polarity Register](#page-69-0) section). The circuits described in the following sections provide examples for three polarity modes that are available for each LED.

- ► Autosense (default)
- ► Active high
- ► Active low

APPLICATIONS INFORMATION

The output current for the LED 0 and LED 1 pins is 8 mA with a VDDIO = 3.3 V (see the [Specifications](#page-3-0) section for details). For higher current requirements, consider using the circuit described in the Transistor Controlled LED section.

Active High LED Polarity

In active high configuration, the LED x pin can drive an external LED from the anode side. Select the R0 and R1 resistors to control the LED current (refer to the LED specifications in [Table 1](#page-3-0) for information). External pull-down resistors (R_{PD0}, R_{PD1}) with a value of 4.7 kΩ are recommended.

Figure 20. Active High LED Polarity Configuration

Active Low LED Polarity

In active low configuration, the LED x pin can drive an external LED from the cathode side. Select the R0 and R1 resistors to control the LED current (refer to the LED specifications in [Table 1](#page-3-0) for information). External pull-up resistors (R_{P10}, R_{P11}) with a value of 4.7 kΩ are recommended.

Figure 21. Active Low LED Polarity Configuration

Transistor Controlled LED

The following circuit displays a typical configuration where the LED current required is higher than what the LED_0 and LED_1 pins can supply. The circuit operates using the active high LED mode. An external transistor such as an N channel MOSFET transistor can be used. The transistor must be selected so that the gate input capacitance is not sinking current above the maximum rating of the LED x pin during the actuation. Refer to the transistor technical

specifications for information. If required, the inrush current can be reduced by placing a resistance between the transistor gate and the ADIN1100 pin and/or adding a parallel capacitor between the GND and the LED x pin. The additional resistor and capacitor values must be defined based on the transistor selection.

Select the R0 and R1 resistors to control the LED current (refer to the selected LED and transistor specifications of the manufacturer for information).

External pull-down resistors (R_{PD0}, R_{PD1}) with a value of 4.7 kΩ are recommended. In Figure 22, VCC is the power supply used to supply the LEDs.

Figure 22. Transistor Controlled LED Configuration

Autosense Polarity

In autosense mode, the polarity of the LED is automatically detected during power-up, hardware reset, or software reset. LED_0 (internal pull-up) and LED_1 (internal pull-down) have different autosense behaviors due to their internal pull-up and pull-down configurations. Use one of the configurations described in the Active High LED Polarity, Active Low LED Polarity, and Transistor Controlled LED sections so that the two LEDs can be controlled the same way.

COMPONENT RECOMMENDATIONS

The ADIN1100 requires an external 25 MHz or 50 MHz clock, which can be sourced from an external crystal oscillator (25 MHz) or an external single-ended clock (25 MHz or 50 MHz). The RMII requires an external 50 MHz clock as described in the [External 50 MHz](#page-35-0) [Clock Input for RMII Mode](#page-35-0) section.

The signal voltage on the XTAL_I/CLK_IN pin $(V_{CLK|IN})$ must be a sine or filtered square wave signal with a peak-to-peak voltage range from 0.8 V to 2.5 V. For the single-ended clock option, a $V_{CLK~IN}$ with a 1.0 V p-p swing is recommended to achieve best performance.

Various circuit configurations are proposed in the following sections. A common circuit topology can be used across these options with a change to the passive component values.

APPLICATIONS INFORMATION

Note that during normal operation, a 25 MHz reference clock generated from the external clock source input (crystal, 25 MHz or 50 MHz clock) is provided on the CLK25_REF output pin. This pin can be used as a reference clock for other circuits, such as another 10BASE-T1L device. CLK25_REF is disabled in reset mode.

External Crystal Oscillator for RMII and RGMII Modes

The typical connection for an external crystal (XTAL) is shown in Figure 23.

To ensure minimum current consumption and to minimize stray capacitance, make connections between the crystal, capacitors, and ground as close to the ADIN1100 as possible. Consult individual crystal vendors for recommended load information and crystal performance specifications.

The crystal load capacitance (C_L) is defined by the crystal vendor. C_{PCB1} and C_{PCB2} are the parasitic capacitance between the XTAL_I/CLK_IN and XTAL_O tracks and the ground plane beneath, respectively. C_{X1} and C_{X2} are the two external load capacitors required for the oscillator to operate.

Assuming the following:

 \triangleright C_{PCB1} \approx C_{PCB2} \approx C_{PCBx}

$$
\blacktriangleright C_{X1} \approx C_{X2} \approx C_{Xx}
$$

Then, $C_{Xx} = 2 \times C_1 - C_{PCBx} - 3$ pF

Choose precision capacitors for C_{Xx} with low appreciable temperature coefficient to minimize frequency errors.

To ensure minimum current consumption and to minimize stray capacitance, make the connections between the crystal, capacitors, and ground as close to the ADIN1100 as possible.

Figure 23. Crystal Oscillator Connection

External 25 MHz Clock Input for MII and RGMII Modes

A single-ended 25 MHz reference clock on XTAL_I/CLK_IN can be used for MII or RGMII mode. The clock source must be dc-coupled with the ADIN1100 XTAL_I/CLK_IN pin input, and the XTAL_O pin must be left open circuit.

With 0.8 V \leq V_{CLK} IN p-p \leq 2.5 V, the following results:

- ► For 0.8 V ≤ V_S p-p ≤ 1.0 V, the following is true:
	- $▶$ R1 = 50 Ω
	- ► R2 is not required
- \triangleright For 1.0 V < V_S p-p < 1.8 V, the following is true:
	- \triangleright For best performance, set V_{CLK} IN to 1.0 V p-p
	- ► 500 Ω ≤ R1 ≤ 2 kΩ
	- ► 1 kΩ ≤ R2 ≤ 2 kΩ

$$
V_S p-p - V_{CLK_N} p-p > 0.2 V
$$

\n
$$
R2 = \frac{V_{CLK_N} p-p \times R1}{V_{S} p-p - V_{CLK_N} p-p}
$$

$$
WZ = V_{S\,p} - p - V_{CLK_IN}\,p - p
$$

► For 1.8 V ≤ V_S p-p, the following is true:

$$
R1 = 2 k\Omega
$$

$$
\triangleright R2 = 2 k\Omega
$$

Figure 24. External 25 MHz Clock Input Circuit for MII and RGMII Modes

External 50 MHz Clock Input for RMII Mode

RMII mode requires a single-ended 50 MHz reference clock signal on XTAL/CLK IN with the XTAL O pin left open circuit.

The PCB parasitic impedance requires particular attention, and it is recommended to match the clock trace lengths between the ADIN1100 and the external MAC (for example, the meandered trace) for RMII timing purposes.
APPLICATIONS INFORMATION

For best performance, set the ADIN1100 $V_{CLK IN}$ signal to 1.0 V p-p.

Figure 25. External 50 MHz Clock Input Circuit for RMII Mode

For 0.8 V \leq V_S p-p \leq 1 V (dc coupling), the following results:

 \triangleright C1 = 50 Ω (replace capacitor with resistor)

► C2 not required

For V_S p-p > 1.0 V (ac coupling) and C1 and C2 placed near the ADIN1100 XTAL_I/CLK_IN pin, the following results:

- \triangleright C1 = 70 pF maximum
- ► C2 = 10 pF
- \triangleright C_{PIN} = 3 pF (ADIN1100 pin capacitance)
- \triangleright C_{STRAY2} \approx 0 pF
- V_S p-p V_{CLK} IN p-p > 0.2 V

$$
\triangleright C1 = \frac{V_{CLK_IN} \times (C2 + C_{STRAY2} + C_{PIN})}{V_S - V_{CLK_IN}}
$$

Table 28. C1 and C2 Values for Different V_S p-p Values with V_{CLK} _{IN} = 1 V p-p

The capacitor tolerance from the manufacturer must be included in the calculations to provide a reasonable margin with regard to the $V_{\text{CLK IN}}$ target voltage.

Note that additional filtering can be required for the external MAC module. The selected clock source must be able to supply the current required by the circuit.

PCB Parasitic Capacitance Considerations

The parasitic capacitance of the clock traces can have a nonnegligible impact on the signal. The PCB stack and the trace impedance must be selected carefully to reduce parasitic impedance and provide the best timing performance, especially if the RMII is selected (50 MHz clock).

Considering the ac coupling RMII scenario described in the [Exter](#page-35-0)[nal 50 MHz Clock Input for RMII Mode](#page-35-0) section, Table 29 displays some basic calculations of the PCB parasitic capacitance using the following PCB parameters:

- ► Transmit line type: Microstrip Layer 1 and Layer 2
- \triangleright Clock trace width: W = 127 µm (5 mils)
- \triangleright PCB L1 copper foil thickness after plating: T = 35 µm (1.38 mils)
- \blacktriangleright L1 to L2 ground plane distance: H = 0.116 mm (4.55 mils)
- \blacktriangleright Substrate relative permittivity: ε_{r} = 4.6 (FR4)

In addition, see the IPC-2141 standard for models and guidelines.

Table 29. Trace Parasitic Capacitance Examples

ELECTROMAGNETIC COMPATIBILITY (EMC) AND ELECTROMAGNETIC IMMUNITY (EMI)

The ADIN1100 was tested at the system level for EMC and EMI. Table 30 summarizes the results.

Table 30. EMC/EMI Tests Conducted on ADIN1100 at System Level

MDIO INTERFACE

The management interface provides a 2-wire serial interface (MDIO) between a host controller (such as micro-controller or an external MAC chip) and the ADIN1100, allowing read and write operations in the management registers.

The management interface of the ADIN1100 is compatible with both IEEE Standard 802.3 Clause 22 and IEEE Standard 802.3 Clause 45.

The hardware configuration pins determine the default value of some registers after power-up, hardware reset, or software reset. For those registers, the reset value referenced in the register bit description tables is listed as pin dependent. This dependency allows the ADIN1100 to be configured in hardware without the need for software operations over the MDIO interface (unmanaged application).

The [Hardware Configuration Pins](#page-22-0) section provides full details on configuration pin setup for managed and unmanaged applications.

Table 31. Clause 22 Frame Format

The access permissions of the registers are as follows:

- ► R/W: read/write
- ► R: read only
- ► R LL: read only, latch low
- ► R LH: read only, latch high
- ► R/W SC: read/write, self clear
- ► R SC: read only, self clear

CLAUSE 22

IEEE Standard 802.3 Clause 22 allows access to up to 32 registers in 32 different PHY addresses.

The IEEE Clause 22 MMD register access format is shown in Table 31 and Table 32.

Table 32. Clause 22 Input Register Decode

MDIO INTERFACE

CLAUSE 45

The clause 45 registers are made up of four device address groupings (see Table 33) based on the MDIO manageable device (MMD). Within each device address space, IEEE standard registers are located in register addresses between 0x0000 and 0x7FFF and vendor specific registers are located in register addresses from 0x8000 to 0xFFFF.

This setup allows access to up to 32 PHYs consisting of up to 32 MMDs through a single MDIO interface.

The IEEE Clause 45 MMD register access format is shown in Table 34 and Table 35.

Clause 45 operations differ from Clause 22 operations where a single frame specifies the register address and data to read or

Table 34. Clause 45 Frame Format

write. In Clause 45, a first frame is sent to specify the device address and register address to access. A second frame is then sent to perform the read or write operation on the selected device address and register specified in the first frame.

Table 35. Clause 45 Input Register Decode

MDIO INTERFACE

RECOMMENDED REGISTER OPERATION

Many of the ADIN1100 registers are defined and compliant with the IEEE 802.3 standard. The register behaviors defined by the standard are not always obvious and are described in the Latch Low Registers, IEEE Duplicated Registers, and Read Modify Write Operation sections, including the recommended operation and use of the registers.

Latch Low Registers

The IEEE Standard 802.3-2018 requires certain MDIO accessible registers to exhibit latch low behavior. This behavior allows software that only intermittently reads these registers to detect conditions that may be transitory or short lived. For example, the AN_LINK_STATUS bit is required to latch low. When the device exits from a reset or power-down state, the latching condition is not active and the value of the AN_LINK_STATUS bit reflects the current status of the link. However, if the link comes up and then drops, the latching condition becomes active. In this case, the AN_LINK_STATUS bit reads as 0 even if the link has come back up again in the interim. The latching condition is only cleared after the AN_LINK_STATUS bit is read to ensure that software has had the opportunity to observe that the link dropped.

This latch low behavior means that the software must perform two reads of the AN_LINK_STATUS bit back to back to determine the current status of the link. The first read is needed to clear any active latching condition.

It is important that the software take account of the interaction between MDIO accessible bits that share a register address. For example, the AN_PAGE_RX and AN_LINK_STATUS bits reside at the same register address. As a result, reading the AN_PAGE_RX bit clears any active latching condition associated with the AN_LINK_STATUS bit.

IEEE Duplicated Registers

IEEE Standard 802.3-2018 covers a wide range of definitions and speeds from 10 Mbps to 40 Gbps and higher, and includes an extensive number of clauses. Many of the registers defined by this standard are associated with various clauses, and different PHYs may include different clauses and/or combinations of clauses. Thus, the registers for common functions like software reset, software power-down, and loopback tend to be implemented in multiple clauses.

In the ADIN1100, the physical implementation of these registers is in a single location, but they can be accessed at multiple addresses. For example, the software reset bit can be read or written in all the IEEE MMD locations and vendor specific register locations listed in Table 36.

In this example, these locations are the PMA/PMD, PCS, autonego-tiation, and Vendor Specific 1 device address locations (per [Table](#page-38-0) [33](#page-38-0)).

The ADIN1100 has multiple address locations for the same register to match the IEEE standard.

The ADIN1100 data sheet only mentions a single recommended address location for each of these IEEE registers to simplify the operation and use of the device. In general, the registers introduced in the 802.3cg (10BASE-T1L) section of the standard are recommended over older (equivalent) registers. Registers in a vendor specific address are recommended, in particular where a register brings a number of useful IEEE register bits into a single register address. The ADIN1100 correctly responds to register accesses to all the IEEE register address locations covered by the 10BASE-T1L standard when the start-up sequence is complete after power-up, hardware reset, or software reset.

Read Modify Write Operation

It is strongly recommended that all register write operations be performed as read modify write, especially when modifying individual register bits. If this is not followed, the value of register bits can be inadvertently changed.

ETHERNET CLAUSE 22 REGISTER DETAILS

Table 37. ADIN1100 Register Summary

MII Control Register

Address: 0x0, Reset: 0x1100, Name: MI_CONTROL

This address corresponds to the MII control register specified in Clause 22.2.4.1 of Standard 802.3.

Table 38. Bit Descriptions for MI_CONTROL

MII Status Register

Address: 0x1, Reset: 0x1009, Name: MI_STATUS

This address corresponds to the MII status register specified in Clause 22.2.4.2 of Standard 802.3.

Table 39. Bit Descriptions for MI_STATUS

PHY Identifier 1 Register

Address: 0x2, Reset: 0x0283, Name: MI_PHY_ID1

The PHY Identifier 1 address allows 16 bits of the OUI to be observed.

Table 40. Bit Descriptions for MI_PHY_ID1

PHY Identifier 2 Register

Address: 0x3, Reset: 0xBC81, Name: MI_PHY_ID2

The PHY Identifier 2 address allows six bits of the OUI, and the model and revision number to be observed.

Table 41. Bit Descriptions for MI_PHY_ID2

MMD Access Control Register

Address: 0xD, Reset: 0x0000, Name: MMD_ACCESS_CNTRL

This address corresponds to the MMD access control register specified in Clause 22.2.4.3.11 of IEEE Standard 802.3-2018.

MMD Access Register

Address: 0xE, Reset: 0x0000, Name: MMD_ACCESS

This address corresponds to the MMD access address data register specified in Clause 22.2.4.3.12 of IEEE Standard 802.3-2018.

The MMD_ACCESS register is used in conjunction with the MMD_ACCESS_CNTRL register to provide access to the MMD address space using the interface and mechanisms defined in Clause 22.2.4.

Table 43. Bit Descriptions for MMD_ACCESS

ETHERNET CLAUSE 45 REGISTER DETAILS

Table 44. Ethernet Clause 45 Register Summary

Table 44. Ethernet Clause 45 Register Summary (Continued)

Table 44. Ethernet Clause 45 Register Summary (Continued)

PMA/PMD Control 1 Register

Device Address: 0x01; Register Address: 0x0000, Reset: 0x0000, Name: PMA_PMD_CNTRL1

This address corresponds to PMA/PMD Control Register 1 specified in Clause 45.2.1.1 of Standard 802.3. Note that the reset value of this register is dependent on the hardware configuration pin settings.

Table 45. Bit Descriptions for PMA_PMD_CNTRL1

PMA/PMD Status 1 Register

Device Address: 0x01; Register Address: 0x0001, Reset: 0x0002, Name: PMA_PMD_STAT1

This address corresponds to PMA/PMD Status Register 1 specified in Clause 45.2.1.2 of Standard 802.3.

Table 46. Bit Descriptions for PMA_PMD_STAT1

PMA/PMD MMD Devices in Package 1 Register

Device Address: 0x01; Register Address: 0x0005, Reset: 0x008B, Name: PMA_PMD_DEVS_IN_PKG1

Table 47. Bit Descriptions for PMA_PMD_DEVS_IN_PKG1

PMA/PMD MMD Devices in Package 2 Register

Device Address: 0x01; Register Address: 0x0006, Reset: 0xC000, Name: PMA_PMD_DEVS_IN_PKG2

Table 48. Bit Descriptions for PMA_PMD_DEVS_IN_PKG2

PMA/PMD Control 2 Register

Device Address: 0x01; Register Address: 0x0007, Reset: 0x003D, Name: PMA_PMD_CNTRL2

Table 49. Bit Descriptions for PMA_PMD_CNTRL2

Table 49. Bit Descriptions for PMA_PMD_CNTRL2 (Continued)

Table 49. Bit Descriptions for PMA_PMD_CNTRL2 (Continued)

PMA/PMD Status 2 Register

Device Address: 0x01; Register Address: 0x0008, Reset: 0x8301, Name: PMA_PMD_STAT2

Table 50. Bit Descriptions for PMA_PMD_STAT2

PMA/PMD Transmit Disable Register

Device Address: 0x01; Register Address: 0x0009, Reset: 0x0000, Name: PMA_PMD_TX_DIS

This address corresponds to the PMD transmit disable register specified in Clause 45.2.1.8 of Standard 802.3.

Table 51. Bit Descriptions for PMA_PMD_TX_DIS

PMA/PMD Extended Abilities Register

Device Address: 0x01; Register Address: 0x000B, Reset: 0x0800, Name: PMA_PMD_EXT_ABILITY

PMA/PMD extended abilities.

Table 52. Bit Descriptions for PMA_PMD_EXT_ABILITY

BASE-T1 PMA/PMD Extended Ability Register

Device Address: 0x01; Register Address: 0x0012, Reset: 0x0004, Name: PMA_PMD_BT1_ABILITY

This address corresponds to the BASE-T1 PMA/PMD extended ability register specified in Clause 45.2.1.16 of Standard 802.3. This register is read only, and writes have no effect.

Table 53. Bit Descriptions for PMA_PMD_BT1_ABILITY

BASE-T1 PMA/PMD Control Register

Device Address: 0x01; Register Address: 0x0834, Reset: 0x8002, Name: PMA_PMD_BT1_CONTROL

This address corresponds to the BASE-T1 PMA/PMD control register specified in Clause 45.2.1.185 of Standard 802.3.

Table 54. Bit Descriptions for PMA_PMD_BT1_CONTROL

10BASE-T1L PMA Control Register

Device Address: 0x01; Register Address: 0x08F6, Reset: 0x0000, Name: B10L_PMA_CNTRL

This address corresponds to the 10BASE-T1L PMA control register specified in Clause 45.2.1.186a of Standard 802.3cg.

Table 55. Bit Descriptions for B10L_PMA_CNTRL

10BASE-T1L PMA Status Register

Device Address: 0x01; Register Address: 0x08F7, Reset: 0x2800, Name: B10L_PMA_STAT

This address corresponds to the 10BASE-T1L PMA status register specified in Clause 45.2.1.186b of Standard 802.3cg.

Table 56. Bit Descriptions for B10L_PMA_STAT

10BASE-T1L Test Mode Control Register

Device Address: 0x01; Register Address: 0x08F8, Reset: 0x0000, Name: B10L_TEST_MODE_CNTRL

This address corresponds to the 10BASE-T1L PMA test mode control register specified in Clause 45.2.1.186c of Standard 802.3cg. The default value of this register selects normal operation without management intervention as the initial state of the device.

Table 57. Bit Descriptions for B10L_TEST_MODE_CNTRL

10BASE-T1L PMA Link Status Register

Device Address: 0x01; Register Address: 0x8302, Reset: 0x0000, Name: B10L_PMA_LINK_STAT

This address can be read to determine the 10BASE-T1L PMA link status. Reading B10L_PMA_LINK_STAT clears the latching condition of the bits in Table 58.

Table 58. Bit Descriptions for B10L_PMA_LINK_STAT

MSE Value Register

Device Address: 0x01; Register Address: 0x830B, Reset: 0x0000, Name: MSE_VAL

Table 59. Bit Descriptions for MSE_VAL

PCS Control 1 Register

Device Address: 0x03; Register Address: 0x0000, Reset: 0x0000, Name: PCS_CNTRL1

This address corresponds to PCS Control Register 1 specified in Clause 45.2.3.1 of Standard 802.3.

PCS Status 1 Register

Device Address: 0x03; Register Address: 0x0001, Reset: 0x0002, Name: PCS_STAT1

Table 61. Bit Descriptions for PCS_STAT1

PCS MMD Devices in Package 1 Register

Device Address: 0x03; Register Address: 0x0005, Reset: 0x008B, Name: PCS_DEVS_IN_PKG1

Table 62. Bit Descriptions for PCS_DEVS_IN_PKG1

PCS MMD Devices in Package 2 Register

Device Address: 0x03; Register Address: 0x0006, Reset: 0xC000, Name: PCS_DEVS_IN_PKG2

Vendor Specific Device 1 and Vendor Specific Device 2 MMDs are present.

Table 63. Bit Descriptions for PCS_DEVS_IN_PKG2

PCS Status 2 Register

Device Address: 0x03; Register Address: 0x0008, Reset: 0x8000, Name: PCS_STAT2

Table 64. Bit Descriptions for PCS_STAT2

10BASE-T1L PCS Control Register

Device Address: 0x03; Register Address: 0x08E6, Reset: 0x0000, Name: B10L_PCS_CNTRL

This address corresponds to the 10BASE-T1L PCS control register specified in Clause 45.2.3.68a of Standard 802.3cg.

Table 65. Bit Descriptions for B10L_PCS_CNTRL

10BASE-T1L PCS Status Register

Device Address: 0x03; Register Address: 0x08E7, Reset: 0x0000, Name: B10L_PCS_STAT

This address corresponds to the 10BASE-T1L PCS status register specified in Clause 45.2.3.68b of Standard 802.3cg.

Table 66. Bit Descriptions for B10L_PCS_STAT

Autonegotiation MMD Devices in Package 1 Register

Device Address: 0x07; Register Address: 0x0005, Reset: 0x008B, Name: AN_DEVS_IN_PKG1

Clause 22 registers and PMA/PMD, PCS, and autonegotiation MMDs are present.

Table 67. Bit Descriptions for AN_DEVS_IN_PKG1

Autonegotiation MMD Devices in Package 2 Register

Device Address: 0x07; Register Address: 0x0006, Reset: 0xC000, Name: AN_DEVS_IN_PKG2

Vendor Specific Device 1 and Vendor Specific Device 2 MMDs are present.

Table 68. Bit Descriptions for AN_DEVS_IN_PKG2

BASE-T1 Autonegotiation Control Register

Device Address: 0x07; Register Address: 0x0200, Reset: 0x1000, Name: AN_CONTROL

This address corresponds to the BASE-T1 autonegotiation control register specified in Clause 45.2.7.19 of Standard 802.3.

Table 69. Bit Descriptions for AN_CONTROL

BASE-T1 Autonegotiation Status Register

Device Address: 0x07; Register Address: 0x0201, Reset: 0x0008, Name: AN_STATUS

This address corresponds to the BASE-T1 autonegotiation status register specified in Clause 45.2.7.20 of Standard 802.3.

Table 70. Bit Descriptions for AN_STATUS

Table 70. Bit Descriptions for AN_STATUS (Continued)

BASE-T1 Autonegotiation Advertisement Register, Bits[15:0]

Device Address: 0x07; Register Address: 0x0202, Reset: 0x0001, Name: AN_ADV_ABILITY_L

This address corresponds to the BASE-T1 autonegotiation advertisement register, Bits[15:0], specified in Clause 45.2.7.21 of Standard 802.3.

Table 71. Bit Descriptions for AN_ADV_ABILITY_L

BASE-T1 Autonegotiation Advertisement Register, Bits[31:16]

Device Address: 0x07; Register Address: 0x0203, Reset: 0x4000, Name: AN_ADV_ABILITY_M

This address corresponds to the BASE-T1 autonegotiation advertisement register, Bits[31:16], specified in Clause 45.2.7.21 of Standard 802.3.

Table 72. Bit Descriptions for AN_ADV_ABILITY_M

BASE-T1 Autonegotiation Advertisement Register, Bits[47:32]

Device Address: 0x07; Register Address: 0x0204, Reset: 0x0000, Name: AN_ADV_ABILITY_H

This address corresponds to the BASE-T1 autonegotiation advertisement register, Bits[47:32], specified in Clause 45.2.7.21 of Standard 802.3.

Table 73. Bit Descriptions for AN_ADV_ABILITY_H

BASE-T1 Autonegotiation Link Partner Base Page Ability Register, Bits[15:0]

Device Address: 0x07; Register Address: 0x0205, Reset: 0x0000, Name: AN_LP_ADV_ABILITY_L

This address corresponds to the BASE-T1 autonegotiation base page ability register, Bits[15:0], of the link partner specified in Clause 45.2.7.22 of Standard 802.3. Note that the value of the AN_LP_ADV_ABILITY_M and AN_LP_ADV_ABILITY_H registers is latched when AN_LP_ADV_ABILITY_L is read.

Table 74. Bit Descriptions for AN_LP_ADV_ABILITY_L

BASE-T1 Autonegotiation Link Partner Base Page Ability Register, Bits[31:16]

Device Address: 0x07; Register Address: 0x0206, Reset: 0x0000, Name: AN_LP_ADV_ABILITY_M

This address corresponds to the BASE-T1 autonegotiation base page ability register, Bits[31:16], of the link partner specified in Clause 45.2.7.22 of Standard 802.3. Note that the value of this register is latched when AN_LP_ADV_ABILITY_L is read. Reading this register returns the latched value rather than the current value.

Table 75. Bit Descriptions for AN_LP_ADV_ABILITY_M

BASE-T1 Autonegotiation Link Partner Base Page Ability Register, Bits[47:32]

Device Address: 0x07; Register Address: 0x0207, Reset: 0x0000, Name: AN_LP_ADV_ABILITY_H

This address corresponds to the BASE-T1 autonegotiation base page ability register, Bits[47:32], of the link partner specified in Clause 45.2.7.22 of Standard 802.3. Note that the value of this register is latched when AN_LP_ADV_ABILITY_L is read. Reading this register returns the latched value rather than the current value.

Bits Bit Name Description Reset Access 15 RESERVED Reserved. Reserved. And the set of 14 AN_LP_ADV_B10L_EEE Link Partner 10BASE-T1L EEE Ability. This bit reports if the link partner is capable of using 10BASE-T1L energy efficient Ethernet. $0x0$ R 13 AN_LP_ADV_B10L_TX_LVL_HI_ABL Link Partner 10BASE-T1L High Level Transmit Operating Mode Ability. This bit reports whether the link partner is capable of transmitting in the high level (2.4 V p-p) transmit operating mode. This bit is used with AN_LP_ADV_B10L_TX_LVL_HI_REQ to configure the 10BASE-T1L transmission level (2.4 V p-p or 1.0 V p-p); see the AN_ADV_B10L_TX_LVL_HI_REQ bit for more details. $0x0$ R 12 AN_LP_ADV_B10L_TX_LVL_HI_REQ Link Partner 10BASE-T1L High Level Transmit Operating Mode Request. This bit reports whether the link partner is requesting that the high level (2.4 V p-p) transmit operating mode be used. See the AN_ADV_B10L_TX_LVL_HI_REQ bit for more details. $0x0$ R 11 AN_LP_ADV_B10S_HD Link Partner 10BASE-T1S Half-Duplex Ability. This bit reports if the link partner is capable of using 10BASE-T1S half duplex. $0x0$ R [10:0] RESERVED Reserved. 0x0 R

Table 76. Bit Descriptions for AN_LP_ADV_ABILITY_H

BASE-T1 Autonegotiation Next Page Transmit Register, Bits[15:0]

Device Address: 0x07; Register Address: 0x0208, Reset: 0x2001, Name: AN_NEXT_PAGE_L

This address corresponds to the BASE-T1 autonegotiation next page transmit register, Bits[15:0], specified in Clause 45.2.7.23 of Standard 802.3. On power-up or autonegotiation reset, this register contains the default value, which represents a message page with the message code set to null. Write AN_NEXT_PAGE_M and AN_NEXT_PAGE_H before AN_NEXT_PAGE_L.

Table 77. Bit Descriptions for AN_NEXT_PAGE_L

BASE-T1 Autonegotiation Next Page Transmit Register, Bits[31:16]

Device Address: 0x07; Register Address: 0x0209, Reset: 0x0000, Name: AN_NEXT_PAGE_M

This address corresponds to the BASE-T1 autonegotiation next page transmit register, Bits[31:16], specified in Clause 45.2.7.23 of Standard 802.3. On power-up or autonegotiation reset, this register contains the default value, which represents a message page with the message code set to null. Write AN_NEXT_PAGE_M and AN_NEXT_PAGE_H before AN_NEXT_PAGE_L.

Table 78. Bit Descriptions for AN_NEXT_PAGE_M

BASE-T1 Autonegotiation Next Page Transmit Register, Bits[47:32]

Device Address: 0x07; Register Address: 0x020A, Reset: 0x0000, Name: AN_NEXT_PAGE_H

This address corresponds to the BASE-T1 autonegotiation next page transmit register, Bits[47:42], specified in Clause 45.2.7.23 of Standard 802.3. On power-up or autonegotiation reset, this register contains the default value, which represents a message page with the message code set to null. Write AN_NEXT_PAGE_M and AN_NEXT_PAGE_H before AN_NEXT_PAGE_L.

Table 79. Bit Descriptions for AN_NEXT_PAGE_H

BASE-T1 Autonegotiation Link Partner Next Page Ability Register, Bits[15:0]

Device Address: 0x07; Register Address: 0x020B, Reset: 0x0000, Name: AN_LP_NEXT_PAGE_L

This address corresponds to the BASE-T1 autonegotiation next page ability register, Bits[15:0], of the link partner specified in Clause 45.2.7.24 of Standard 802.3. The values of AN_LP_NEXT_PAGE_M and AN_LP_NEXT_PAGE_H are latched when this register is read.

Table 80. Bit Descriptions for AN_LP_NEXT_PAGE_L

BASE-T1 Autonegotiation Link Partner Next Page Ability Register, Bits[31:16]

Device Address: 0x07; Register Address: 0x020C, Reset: 0x0000, Name: AN_LP_NEXT_PAGE_M

This address corresponds to the BASE-T1 autonegotiation next page ability register, Bits[31:16], of the link partner specified in Clause 45.2.7.24 of Standard 802.3. The values of this register are latched when AN_LP_NEXT_PAGE_L is read. Reading this register returns the latched value rather than the current value.

Table 81. Bit Descriptions for AN_LP_NEXT_PAGE_M

BASE-T1 Autonegotiation Link Partner Next Page Ability Register, Bits[47:32]

Device Address: 0x07; Register Address: 0x020D, Reset: 0x0000, Name: AN_LP_NEXT_PAGE_H

This address corresponds to the BASE-T1 autonegotiation link partner's next page ability register, Bits[47:32], specified in Clause 45.2.7.24 of Standard 802.3. The values of this register are latched when AN_LP_NEXT_PAGE_L is read. Reading this register returns the latched value rather than the current value.

Table 82. Bit Descriptions for AN_LP_NEXT_PAGE_H

10BASE-T1 Autonegotiation Control Register

Device Address: 0x07; Register Address: 0x020E, Reset: 0x8000, Name: AN_B10_ADV_ABILITY

This address corresponds to the 10BASE-T1 autonegotiation control register specified in Clause 45.2.7.25 of Standard 802.3cg.

Table 83. Bit Descriptions for AN_B10_ADV_ABILITY

Table 83. Bit Descriptions for AN_B10_ADV_ABILITY (Continued)

10BASE-T1 Autonegotiation Status Register

Device Address: 0x07; Register Address: 0x020F, Reset: 0x0000, Name: AN_B10_LP_ADV_ABILITY

This address corresponds to the 10BASE-T1 autonegotiation status register specified in Clause 45.2.7.26 of Standard 802.3cg.

Autonegotiation Forced Mode Enable Register

Device Address: 0x07; Register Address: 0x8000, Reset: 0x0000, Name: AN_FRC_MODE_EN

The effect of this register is superseded by the AN_EN bit, which enables the autonegotiation process. If autonegotiation is disabled (AN_EN = 0) and AN_FRC_MODE_EN = 1, forced mode is enabled.

Table 85. Bit Descriptions for AN_FRC_MODE_EN

Extra Autonegotiation Status Register

Device Address: 0x07; Register Address: 0x8001, Reset: 0x0000, Name: AN_STATUS_EXTRA

This register is provided in addition to AN_STATUS.

PHY Instantaneous Status Register

Device Address: 0x07; Register Address: 0x8030, Reset: 0x0010, Name: AN_PHY_INST_STATUS

This register address provides access to instantaneous status indications. These values are not latched. The set of indications returned by this register is a consistent set, that is, a set of values in effect at the time the register address is read.

Table 87. Bit Descriptions for AN_PHY_INST_STATUS

Vendor Specific 1 MMD Identifier High Register

Device Address: 0x1E; Register Address: 0x0002, Reset: 0x0283, Name: MMD1_DEV_ID1

This address corresponds to the Vendor Specific 1 MMD identifier register specified in Clause 45.2.11.1 of Standard 802.3 and allows 16 bits of the organizationally unique identifier (OUI) to be observed.

Table 88. Bit Descriptions for MMD1_DEV_ID1

Vendor Specific 1 MMD Identifier Low Register

Device Address: 0x1E; Register Address: 0x0003, Reset: 0xBC81, Name: MMD1_DEV_ID2

This address corresponds to the Vendor Specific 1 MMD identifier register specified in Clause 45.2.11.1 of Standard 802.3 and allows six bits of the OUI along with the model number and revision number to be observed.

Table 89. Bit Descriptions for MMD1_DEV_ID2

Vendor Specific 1 MMDs in Package Register

Device Address: 0x1E; Register Address: 0x0005, Reset: 0x008B, Name: MMD1_DEVS_IN_PKG1

Clause 22 registers and PMA/PMD, PCS, and autonegotiation MMDs are present.

Table 90. Bit Descriptions for MMD1_DEVS_IN_PKG1

Device Address: 0x1E; Register Address: 0x0006, Reset: 0xC000, Name: MMD1_DEVS_IN_PKG2

Vendor Specific 1 and Vendor Specific 2 MMDs are present.

Table 91. Bit Descriptions for MMD1_DEVS_IN_PKG2

Vendor Specific 1 MMD Status Register

Device Address: 0x1E; Register Address: 0x0008, Reset: 0x8000, Name: MMD1_STATUS

This address corresponds to the Vendor Specific 1 MMD status register specified in Clause 45.2.11.2 of Standard 802.3.

System Interrupt Status Register

Device Address: 0x1E; Register Address: 0x0010, Reset: 0x1000, Name: CRSM_IRQ_STATUS

This address can be used to check which interrupt requests have been triggered since the last time it was read. Each bit goes high when the associated event occurs and then latches high until it is unlatched by reading. The bits of CRSM_IRQ_STATUS go high even when the associated interrupts are not enabled. A reserved interrupt being triggered indicates a system error, which requires a hardware reset.

Table 93. Bit Descriptions for CRSM_IRQ_STATUS

System Interrupt Mask Register

Device Address: 0x1E; Register Address: 0x0020, Reset: 0x1FFE, Name: CRSM_IRQ_MASK

Controls whether or not the interrupt signal is asserted in response to various events.

Table 94. Bit Descriptions for CRSM_IRQ_MASK

Software Reset Register

Device Address: 0x1E; Register Address: 0x8810, Reset: 0x0000, Name: CRSM_SFT_RST

Table 95. Bit Descriptions for CRSM_SFT_RST

Software Power-Down Control Register

Device Address: 0x1E; Register Address: 0x8812, Reset: 0x0000, Name: CRSM_SFT_PD_CNTRL

Table 96. Bit Descriptions for CRSM_SFT_PD_CNTRL

PHY Subsystem Reset Register

Device Address: 0x1E; Register Address: 0x8814, Reset: 0x0000, Name: CRSM_PHY_SUBSYS_RST

Table 97. Bit Descriptions for CRSM_PHY_SUBSYS_RST

PHY MAC Interface Reset Register

Device Address: 0x1E; Register Address: 0x8815, Reset: 0x0000, Name: CRSM_MAC_IF_RST

Table 98. Bit Descriptions for CRSM_MAC_IF_RST

System Status Register

Device Address: 0x1E; Register Address: 0x8818, Reset: 0x0000, Name: CRSM_STAT

Table 99. Bit Descriptions for CRSM_STAT

CRSM Power Management Control Register

Device Address: 0x1E; Register Address: 0x8819, Reset: 0x0000, Name: CRSM_PMG_CNTRL

Table 100. Bit Descriptions for CRSM_PMG_CNTRL

MAC Interface Configuration Register

Device Address: 0x1E; Register Address: 0x882B, Reset: 0x0000, Name: CRSM_MAC_IF_CFG

Configure the MAC interface only by pins. Do not change by software.

Table 101. Bit Descriptions for CRSM_MAC_IF_CFG

Table 101. Bit Descriptions for CRSM_MAC_IF_CFG (Continued)

CRSM Diagnostics Clock Control Register

Device Address: 0x1E; Register Address: 0x882C, Reset: 0x0002, Name: CRSM_DIAG_CLK_CTRL

CRSM diagnostics clock control.

Table 102. Bit Descriptions for CRSM_DIAG_CLK_CTRL

Package Configuration Values Register

Device Address: 0x1E; Register Address: 0x8C22, Reset: 0x0002, Name: MGMT_PRT_PKG

The MGMT_PRT_PKG_VAL address allows reading of the package configuration values.

Table 103. Bit Descriptions for MGMT_PRT_PKG

MDIO Control Register

Device Address: 0x1E; Register Address: 0x8C30, Reset: 0x0000, Name: MGMT_MDIO_CNTRL

Table 104. Bit Descriptions for MGMT_MDIO_CNTRL

Pin Mux Configuration 1 Register

Device Address: 0x1E; Register Address: 0x8C56, Reset: 0x00FE, Name: DIGIO_PINMUX

Table 105. Bit Descriptions for DIGIO_PINMUX

Pin Mux Configuration 2 Register

Device Address: 0x1E; Register Address: 0x8C57, Reset: 0x00FF, Name: DIGIO_PINMUX2

Table 106. Bit Descriptions for DIGIO_PINMUX2

LED_0 On/Off Blink Time Register

Device Address: 0x1E; Register Address: 0x8C80, Reset: 0x3636, Name: LED0_BLINK_TIME_CNTRL

LED on blink time = LED0 ON N4MS \times 4 ms.

LED off blink time = LED0 OFF $N4MS \times 4 ms$.

If LEDx MODE = 0 and LEDx FUNCTION is set to blink, the LED activity starts with an LED off sequence, followed by an LED on sequence, and then repeats.

If LEDx MODE = 1 and LEDx FUNCTION is set to blink, the LED activity starts with an LED on sequence, followed by an LED off sequence, and then repeats.

If LED x OFF N4MS = LED x ON N4MS = 0, this is a special case whereby the internal activity signal as selected by LED x FUNCTION can be monitored live.

If LEDx FUNCTION is programmed to a combination of a link and activity signal, the LED is on while the link is up and with no activity. The LED switches off for either loss of link or receipt of activity.

If LED_x FUNCTION is programmed to an activity signal, the LED is off with no activity. The LED switches on upon receipt of activity.

Table 107. Bit Descriptions for LED0_BLINK_TIME_CNTRL

LED_1 On/Off Blink Time Register

Device Address: 0x1E; Register Address: 0x8C81, Reset: 0x3636, Name: LED1_BLINK_TIME_CNTRL

LED on blink time = LED1 ON N4MS \times 4ms.

LED off blink time = LED1 OFF N4MS \times 4 ms.

If LEDx MODE = 0 and LEDx FUNCTION is set to blink, the LED activity starts with an LED off sequence followed by an LED on sequence, and then repeats.

If LEDx MODE = 1 and LEDx FUNCTION is set to blink, the LED activity starts with an LED on sequence, followed by an LED Off sequence, and then repeats.

If LEDx $OFN4MS = LEDx ON N4MS = 0$, this is a special case whereby the internal activity signal as selected by LEDx $FUNCTION can be$ monitored live.

If LEDx FUNCTION is programmed to a combination of a link and activity signal, the LED is on while the link is up and with no activity. The LED switches off for either loss of link or receipt of activity.

If LEDx FUNCTION is programmed to an activity signal, the LED is off with no activity. The LED switches on upon receipt of activity.

Table 108. Bit Descriptions for LED1_BLINK_TIME_CNTRL

LED Control Register

Device Address: 0x1E; Register Address: 0x8C82, Reset: 0x8480, Name: LED_CNTRL

LED control register

Table 109. Bit Descriptions for LED_CNTRL

Table 109. Bit Descriptions for LED_CNTRL (Continued)

LED Polarity Register

Device Address: 0x1E; Register Address: 0x8C83, Reset: 0x0000, Name: LED_POLARITY

Allows the LED polarity to be automatically sensed by the internal logic or allows reconfiguration by the user.

Vendor Specific MMD 2 Device Identifier High Register

Device Address: 0x1F; Register Address: 0x0002, Reset: 0x0283, Name: MMD2_DEV_ID1

Table 111. Bit Descriptions for MMD2_DEV_ID1

Vendor Specific MMD 2 Device Identifier Low Register

Device Address: 0x1F; Register Address: 0x0003, Reset: 0xBC81, Name: MMD2_DEV_ID2

Table 112. Bit Descriptions for MMD2_DEV_ID2

Vendor Specific 2 MMDs in Package Register

Device Address: 0x1F; Register Address: 0x0005, Reset: 0x008B, Name: MMD2_DEVS_IN_PKG1

Clause 22 registers and PMA/PMD, PCS, and autonegotiation MMDs are present.

Table 113. Bit Descriptions for MMD2_DEVS_IN_PKG1

Device Address: 0x1F; Register Address: 0x0006, Reset: 0xC000, Name: MMD2_DEVS_IN_PKG2

Vendor Specific 1 and Vendor Specific 2 MMDs are present.

Table 114. Bit Descriptions for MMD2_DEVS_IN_PKG2

Vendor Specific MMD 2 Status Register

Device Address: 0x1F; Register Address: 0x0008, Reset: 0x8000, Name: MMD2_STATUS

This address corresponds to the Vendor Specific MMD 2 status register.

Table 115. Bit Descriptions for MMD2_STATUS

PHY Subsystem Interrupt Status Register

Device Address: 0x1F; Register Address: 0x0011, Reset: 0x0000, Name: PHY_SUBSYS_IRQ_STATUS

This address can be read to check which interrupt events have occurred since the last time it was read. Each bit goes high when the associated event occurs and then latches high until it is unlatched by reading. The bits of PHY_SUBSYS_IRQ_STATUS go high even when the associated bits in PHY_SUBSYS_IRQ_MASK are not set. A reserved interrupt being triggered indicates a fatal error in the system.

Table 116. Bit Descriptions for PHY_SUBSYS_IRQ_STATUS (Continued)

PHY Subsystem Interrupt Mask Register

Device Address: 0x1F; Register Address: 0x0021, Reset: 0x2402, Name: PHY_SUBSYS_IRQ_MASK

Controls whether or not the interrupt signal is asserted in response to various events.

Table 117. Bit Descriptions for PHY_SUBSYS_IRQ_MASK

Frame Checker Enable Register

Device Address: 0x1F; Register Address: 0x8001, Reset: 0x0001, Name: FC_EN

This register is used to enable the frame checker. The frame checker analyzes the received frames from either the MAC interface or the PHY (see the FC_TX_SEL register) to report the number of frames received, CRC errors, and various other frame errors. The frame checker frame and error counter registers count these events.

Table 118. Bit Descriptions for FC_EN

Frame Checker Interrupt Enable Register

Device Address: 0x1F; Register Address: 0x8004, Reset: 0x0001, Name: FC_IRQ_EN

This register is used to enable the frame checker interrupt. An interrupt is generated when a receive error occurs. Enable the frame checker/generator interrupt in the PHY_SUBSYS_IRQ_MASK register. Set the MAC_IF_FC_FG_IRQ_EN bit.

The status can be read via the MAC_IF_FC_FG_IRQ_LH bit in the PHY_SUBSYS_IRQ_STATUS register.

Frame Checker Transmit Select Register

Device Address: 0x1F; Register Address: 0x8005, Reset: 0x0000, Name: FC_TX_SEL

This register is used to select the transmit side or receive side for frames to be checked. If set, frames received from the MAC interface to be transmitted are checked. The frame checker can be used to verify that correct data is received over the MAC interface and is also useful if remote loopback is enabled (see the MAC_IF_REM_LB_EN bit in the MAC_IF_LOOPBACK register) because it can be used to check the received data after it is looped back at the MAC interface.

Table 120. Bit Descriptions for FC_TX_SEL

Receive Error Count Register

Device Address: 0x1F; Register Address: 0x8008, Reset: 0x0000, Name: RX_ERR_CNT

The receive error counter register is used to access the receive error counter associated with the frame checker in the PHY.

Table 121. Bit Descriptions for RX_ERR_CNT

Frame Checker Count High Register

Device Address: 0x1F; Register Address: 0x8009, Reset: 0x0000, Name: FC_FRM_CNT_H

This register is a latched copy of Bits[31:16] of the 32-bit receive frame counter register. When the receive error counter (RX_ERR_CNT) is read, the receive frame counter register is latched so that the error count and the receive frame count are synchronized.

Table 122. Bit Descriptions for FC_FRM_CNT_H

Frame Checker Count Low Register

Device Address: 0x1F; Register Address: 0x800A, Reset: 0x0000, Name: FC_FRM_CNT_L

This register is a latched copy of Bits[15:0] of the 32-bit receive frame counter register. When the receive error counter (RX_ERR_CNT) is read, the receive frame counter register is latched so that the error count and receive frame count are synchronized.

Table 123. Bit Descriptions for FC_FRM_CNT_L

Frame Checker Length Error Count Register

Device Address: 0x1F; Register Address: 0x800B, Reset: 0x0000, Name: FC_LEN_ERR_CNT

This register is a latched copy of the frame length error counter register. This register is a count of received frames with a length error status. When the receive error counter (RX_ERR_CNT) is read, the frame length error counter register is latched, which ensures that the frame length error count and receive frame count are synchronized.

Table 124. Bit Descriptions for FC_LEN_ERR_CNT

Frame Checker Alignment Error Count Register

Device Address: 0x1F; Register Address: 0x800C, Reset: 0x0000, Name: FC_ALGN_ERR_CNT

This register is a latched copy of the frame alignment error counter register. This register is a count of received frames with an alignment error status. When the receive error counter (RX_ERR_CNT) is read, the alignment error counter is latched, which ensures that the frame alignment error count and the receive frame count are synchronized.

Table 125. Bit Descriptions for FC_ALGN_ERR_CNT

Frame Checker Symbol Error Count Register

Device Address: 0x1F; Register Address: 0x800D, Reset: 0x0000, Name: FC_SYMB_ERR_CNT

This register is a latched copy of the symbol error counter register. This register is a count of received frames with both RX_ER and RX_DV set. When the receive error counter (RX_ERR_CNT) is read, the symbol error count is latched, which ensures that the symbol error count and the frame receive count are synchronized.

Table 126. Bit Descriptions for FC_SYMB_ERR_CNT

Frame Checker Oversized Frame Count Register

Device Address: 0x1F; Register Address: 0x800E, Reset: 0x0000, Name: FC_OSZ_CNT

This register is a latched copy of the oversized frame error counter register. This register is a count of receiver frames with a length greater than 1522 bytes. When the receive error counter (RX_ERR_CNT) is read, the oversized frame counter register is latched, which ensures that the oversized error count and the receive frame count are synchronized.

Table 127. Bit Descriptions for FC_OSZ_CNT

Frame Checker Undersized Frame Count Register

Device Address: 0x1F; Register Address: 0x800F, Reset: 0x0000, Name: FC_USZ_CNT

This register is a latched copy of the undersized frame error counter register. This register is a count of received frames with less than 64 bytes. When the receive error counter (RX_ERR_CNT) is read, the undersized frame error counter is latched, which ensures that the undersized frame error count and the receive frame count are synchronized.

Table 128. Bit Descriptions for FC_USZ_CNT

Frame Checker Odd Nibble Frame Count Register

Device Address: 0x1F; Register Address: 0x8010, Reset: 0x0000, Name: FC_ODD_CNT

This register is a latched copy of the odd nibble frame register. This register is a count of received frames with an odd number of nibbles in the frame. When the receive error counter (RX_ERR_CNT) is read, the odd nibble frame counter register is latched, which ensures that the odd nibble frame count and the receive frame count are synchronized.

Table 129. Bit Descriptions for FC_ODD_CNT

Frame Checker Odd Preamble Packet Count Register

Device Address: 0x1F; Register Address: 0x8011, Reset: 0x0000, Name: FC_ODD_PRE_CNT

This register is a latched copy of the odd preamble packet counter register. This register is a count of received packets with an odd number of nibbles in the preamble. When the receive error counter (RX_ERR_CNT) is read, the odd preamble packet counter register is latched, which ensures that the odd preamble packet count and the receive frame count are synchronized.

Table 130. Bit Descriptions for FC_ODD_PRE_CNT

Frame Checker False Carrier Count Register

Device Address: 0x1F; Register Address: 0x8013, Reset: 0x0000, Name: FC_FALSE_CARRIER_CNT

This register is a latched copy of the false carrier events counter register. This register is a count of the number of times the bad SSD state is entered. When the receive error counter (RX_ERR_CNT) is read, the false carrier events counter register is latched, which ensures that the false carrier events count and the receive frame count are synchronized.

Table 131. Bit Descriptions for FC_FALSE_CARRIER_CNT

Frame Generator Enable Register

Device Address: 0x1F; Register Address: 0x8020, Reset: 0x0000, Name: FG_EN

This register is used to enable the frame generator. When the frame generator is enabled, the source of data for the PHY comes from the frame generator and not the MAC interface.

Table 132. Bit Descriptions for FG_EN

Frame Generator Control/Restart Register

Device Address: 0x1F; Register Address: 0x8021, Reset: 0x0001, Name: FG_CNTRL_RSTRT

This register controls the frame generator. The FG_CNTRL bit field specifies data field type used by the frame generator, for example, random or all zeros. The FG_RSTRT bit restarts the frame generator.

Table 133. Bit Descriptions for FG_CNTRL_RSTRT

Frame Generator Continuous Mode Enable Register

Device Address: 0x1F; Register Address: 0x8022, Reset: 0x0000, Name: FG_CONT_MODE_EN

This register is used to put the frame generator into continuous mode. The default mode of operation is burst mode, where the number of frames generated is specified by the FG_NFRM_H and FG_NFRM_L registers.

Table 134. Bit Descriptions for FG_CONT_MODE_EN

Frame Generator Interrupt Enable Register

Device Address: 0x1F; Register Address: 0x8023, Reset: 0x0000, Name: FG_IRQ_EN

This register is used to enable the frame generator interrupt. An interrupt is generated when the requested number of frames has been generated. Enable the frame checker/generator interrupt in the PHY_SUBSYS_IRQ_MASK register. Set the MAC_IF_FC_FG_IRQ_EN bit.

The interrupt status can be read via the MAC_IF_FC_FG_IRQ_LH bit in the PHY_SUBSYS_IRQ_STATUS register.

Table 135. Bit Descriptions for FG_IRQ_EN

Frame Generator Frame Length Register

Device Address: 0x1F; Register Address: 0x8025, Reset: 0x006B, Name: FG_FRM_LEN

This register specifies the data field frame length in bytes. In addition to the data field, six bytes are added for the source address, six bytes for the destination address, two bytes for the length field, and four bytes for the frame check sequence (FCS). The total length is the data field length plus 18.

Table 136. Bit Descriptions for FG_FRM_LEN

Frame Generator Interframe Gap Length Register

Device Address: 0x1F; Register Address: 0x8026, Reset: 0x000C, Name: FG_IFG_LEN

This register specifies the length in bytes of the interframe gap to be inserted between frames by the frame generator.

Table 137. Bit Descriptions for FG_IFG_LEN

Frame Generator Number of Frames High Register

Device Address: 0x1F; Register Address: 0x8027, Reset: 0x0000, Name: FG_NFRM_H

This register is Bits[31:16] of a 32-bit register that specifies the number of frames to be generated each time the frame generator is enabled or restarted.

Table 138. Bit Descriptions for FG_NFRM_H

Frame Generator Number of Frames Low Register

Device Address: 0x1F; Register Address: 0x8028, Reset: 0x0100, Name: FG_NFRM_L

This register is Bits[15:0] of a 32-bit register that specifies the number of frames to be generated each time the frame generator is enabled or restarted.

Table 139. Bit Descriptions for FG_NFRM_L

Frame Generator Done Register

Device Address: 0x1F; Register Address: 0x8029, Reset: 0x0000, Name: FG_DONE

This register is used to indicate that the frame generator has completed the generation of the number of frames requested in the FG_NFRM_H and FG_NFRM_L registers.

Table 140. Bit Descriptions for FG_DONE

RMII Configuration Register

Device Address: 0x1F; Register Address: 0x8050, Reset: 0x0006, Name: RMII_CFG

Table 141. Bit Descriptions for RMII_CFG

MAC Interface Loopbacks Configuration Register

Device Address: 0x1F; Register Address: 0x8055, Reset: 0x000A, Name: MAC_IF_LOOPBACK

MAC interface loopbacks configuration.

Table 142. Bit Descriptions for MAC_IF_LOOPBACK

MAC Start of Packet (SOP) Generation Control Register

Device Address: 0x1F; Register Address: 0x805A, Reset: 0x001B, Name: MAC_IF_SOP_CNTRL

Table 143. Bit Descriptions for MAC_IF_SOP_CNTRL

PCB LAYOUT RECOMMENDATIONS

LAND PATTERN

The LFCSP has an exposed pad underneath the package that must be soldered to the PCB ground for mechanical, electrical, and thermal reasons.

For thermal impedance performance and to maximize heat transfer to the PCB, the use of a 4 × 4 array of thermal vias beneath the exposed ground pad is recommended. Via tenting is also recommended.

COMPONENT PLACEMENT AND ROUTING

Prioritization of the critical traces and components helps simplify the routing exercise. Place and orient the critical traces and components first to ensure an effective layout. The critical components are the crystal and load capacitors, the CEXT_2 and CEXT_3 capacitors, and all bypass capacitors local to the ADIN1100 device. Prioritize these components for placement and routing.

- ► Place the decoupling capacitor as close as possible to their input pins.
- \blacktriangleright Minimize traces turns, and use a 45 $^{\circ}$ corner.
- ► Avoid traces crossing power planes on adjacent layers.
- ► Avoid stubs.
- ► Keep the MDI traces (RXP, RXN, TXP, and TXN) as short as possible.
- ► Avoid vias on a high speed signal. Place ground vias next to the signal vias to improve the return current path.

CRYSTAL PLACEMENT AND ROUTING

Particular attention is required on the crystal placement and routing to ensure minimum current consumption, reduce stray capacitance, and improve noise immunity.

- ► Place the crystal, capacitors as close as possible to the ADIN1100 XTAL_I/CLK_IN and XTAL_O pins.
- ► Place the load capacitors close to each other.
- ► Use a local GND plane (copper island) for the crystal and load capacitors with a single point connection to the main GND.
- ► Reduce parasitic capacitance by keeping the XTAL I and XTAL O traces away from each other.
- ► Adding a copper keepout on the layer beneath the crystal can also reduce the parasitic capacitance.

PCB STACK

Follow these recommendations for the PCB stack:

- ► Use a PCB stack with a minimum of four layers. Consider six layers or more with external layers used as ground planes to improve EMI issues (optional).
- ► Define copper layer thickness based on the application and power requirements.
- ► Use internal layers for the power and ground planes.
- ► Use external layers for the signal traces.
- ► Use via stitching to improve ground and reduce EMI. The stitching pattern and via to via gaps are defined based on the application.

OUTLINE DIMENSIONS

COMPLIANT TO JEDEC STANDARDS MO-220-WJJD-5

Figure 26. 40-Lead Lead Frame Chip Scale Package [LFCSP] 6 mm × 6 mm Body and 0.75 mm Package Height (CP-40-29) Dimensions shown in millimeters

Updated: July 27, 2021

ORDERING GUIDE

¹ Z = RoHS Compliant Part.

EVALUATION BOARDS

 $1 Z =$ RoHS Compliant Part.

