ICS8432-111

700MHZ/350MH^Z

DIFFERENTIAL-TO-3.3V LVPECL FREQUENCY SYNTHESIZER

PRODUCT DISCONTINUATION NOTICE - LAST TIME BUY EXPIRES JULY 31, 2015

GENERAL DESCRIPTION

The ICS8432-111 is a general purpose, dual output

Differential-to-3.3V LVPECL High Frequency Synthesizer. The ICS8432-111 has a selectable differential CLK, nCLK pair or LVCMOS/LVTTL TEST_CLK. The TEST_CLK input accepts LVCMOS or LVTTL input levels and translates them to 3.3V LVPECL levels. The CLK, nCLK pair can accept most standard differential input levels.The VCO operates at a frequency range of 200MHz to 700MHz. The VCO frequency is programmed in steps equal to the value of the input differential or single ended reference frequency. Output frequencies up to 700MHz for FOUT and 350MHz for FOUT/2 can be programmed using the serial or parallel interfaces to the configuration logic. The low phase noise characteristics and the multiple frequency outputs of the ICS8432-111 makes it an ideal clock source for Fibre Channel 1 and 2, and Infiniband applications.

FEATURES

- Dual differential 3.3V LVPECL outputs
- Selectable differential CLK, nCLK pair or LVCMOS TEST_ **CLK**
- CLK, nCLK pair can accept the following differential input levels: LVPECL, LVDS, LVHSTL, SSTL, HCSL
- TEST_CLK can accept the following input types: LVCMOS or LVTTL
- Maximum FOUT frequency: 700MHz Maximum FOUT/2 frequency: 350MHz
- CLK, nCLK or TEST_CLK input frequency: 40MHz
- VCO range: 250MHz to 700MHz
- Parallel or serial interface for programming counter and VCO frequency multiplier and dividers
- RMS period jitter: 5ps (maximum)
- Cycle-to-cycle jitter: 40ps (maximum)
- 3.3V supply voltage
- 0°C to 70°C ambient operating temperature
- Available in lead-free (RoHS 6) package
- For functional replacement device use 8432DY-101LF

BLOCK DIAGRAM PIN ASSIGNMENT

FUNCTIONAL DESCRIPTION

NOTE: The functional description that follows describes operation using a 25MHz clock input. Valid PLL loop divider values for different input frequencies are defined in the Input Frequency Characteristics, Table 5, NOTE 1.

The ICS8432-111 features a fully integrated PLL and therefore requires no external components for setting the loop bandwidth. A differential clock input is used as the input to the ICS8432-111. This input is fed into the phase detector. A 25MHz clock input provides a 25MHz phase detector reference frequency. The VCO of the PLL operates over a range of 250MHz to 700MHz. The output of the M divider is also applied to the phase detector.

The phase detector and the M divider force the VCO output frequency to be M times the reference frequency by adjusting the VCO control voltage. Note, that for some values of M (either too high or too low), the PLL will not achieve lock. The output of the VCO is scaled by a divider prior to being sent to each of the LVPECL output buffers. The divider provides a 50% output duty cycle.

The programmable features of the ICS8432-111 support two input modes to program the PLL M divider and N output divider. The two input operational modes are parallel and serial. Figure 1 shows the timing diagram for each mode. In parallel mode, the nP_LOAD input is initially LOW. The data on inputs M0 through M8 and N0 and N1 is passed directly to the M divider and N output divider. On the LOW-to-HIGH transition of the nP_LOAD input, the data is latched and the M divider remains loaded until the next LOW transition on nP LOAD or until a serial

event occurs. As a result, the M and N bits can be hardwired to set the M divider and N output divider to a specific default state that will automatically occur during power-up. The TEST output is LOW when operating in the parallel input mode. The relationship between the VCO frequency, the input frequency and the M divider is defined as follows: $\mathsf{fVCO} = \mathsf{f}_{\mathsf{IN}} \times \mathsf{M}$

The M value and the required values of M0 through M8 are shown in Table 3B, Programmable VCO Frequency Function Table. When the input clock is at 25MHz, the valid M values for which the PLL will achieve lock are defined as $10 \le M \le 28$. The frequency out is defined as follows: fOUT = $\frac{\text{fVCO}}{\text{N}} = f_{\text{IN}} \times \frac{\text{M}}{\text{N}}$ N N

Serial operation occurs when nP_LOAD is HIGH and S_LOAD is LOW. The shift register is loaded by sampling the S_DATA bits with the rising edge of S_CLOCK. The contents of the shift register are loaded into the M divider and N output divider when S LOAD transitions from LOW-to-HIGH. The M divide and N output divide values are latched on the HIGH-to-LOW transition of S_LOAD. If S_LOAD is held HIGH, data at the S_DATA input is passed directly to the M divider and N output divider on each rising edge of S_CLOCK. The serial mode can be used to program the M and N bits and test bits T1 and T0. The internal registers T0 and T1 determine the state of the TEST output as follows:

FIGURE 1. PARALLEL & SERIAL LOAD OPERATIONS

^{*}NOTE: The NULL timing slot must be observed.

TABLE 1. PIN DESCRIPTIONS

NOTE: Pullup and Pulldown refers to internal input resistors. See Table 2, Pin Characteristics, for typical values.

TABLE 2. PIN CHARACTERISTICS

TABLE 3A. PARALLEL AND SERIAL MODE FUNCTION TABLE

 $NOTE: L = LOW$

 $H = HIGH$

 $X = Don't care$

↑= Rising edge transition

↓= Falling edge transition

TABLE 3B. PROGRAMMABLE VCO FREQUENCY FUNCTION TABLE

NOTE 1: These M count values and the resulting frequencies correspond to differential input or TEST_CLK input frequency of 25MHz.

TABLE 3C. PROGRAMMABLE OUTPUT DIVIDER FUNCTION TABLE

ABSOLUTE MAXIMUM RATINGS

NOTE: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the DC Characteristics or AC Characteristics is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

TABLE 4A. POWER SUPPLY DC CHARACTERISTICS, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 5\%$ **,** $T_A = 0^{\circ}C$ **to 70°C**

TABLE 4B. LVCMOS / LVTTL DC CHARACTERISTICS, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 5\%$ **, TA = 0°C to 70°C**

NOTE 1: Outputs terminated with 50Ω to V_{cco}/2. See Parameter Information, 3.3V Output Load Test Circuit.

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TABLE 4C. DIFFERENTIAL DC CHARACTERISTICS, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 5\%$ **,** $T_A = 0^{\circ}C$ **to 70°C**

NOTE 1: Common mode voltage is defined as V_{H} .

NOTE 2: For single ended applications, the maximum input voltage for CLK, nCLK is V_{cc} + 0.3V.

TABLE 4D. LVPECL DC CHARACTERISTICS, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 5\%$ **,** $T_A = 0^{\circ}C$ **to 70°C**

NOTE 1: Outputs terminated with 50 W to V_{cco} - 2V.

TABLE 5. INPUT FREQUENCY CHARACTERISTICS, $V_{CC} = V_{CCA} = V_{cco} = 3.3V \pm 5\%$ **, TA = 0°C to 70°C**

NOTE 1: For the input frequency range, the M value must be set for the VCO to operate within the 250MHz to 700MHz range. Using the minimum input frequency of 10MHz, valid values of M are $25 \le M \le 70$. Using the maximum frequency of 40MHz, valid values of M are $7 \le M \le 17$.

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TABLE 6. AC CHARACTERISTICS, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 5\%$, $T_A = 0^{\circ}C$ to 70°C

NOTE 1: Defined as skew between outputs at the same supply voltage and with equal load conditions.

Measured at the output differential cross points.

NOTE 2: This parameter is defined in accordance with JEDEC Standard 65.

PARAMETER MEASUREMENT INFORMATION

APPLICATION INFORMATION

POWER SUPPLY FILTERING TECHNIQUES

As in any high speed analog circuitry, the power supply pins are vulnerable to random noise. The ICS8432-111 provides separate power supplies to isolate any high switching noise from the outputs to the internal PLL. $\mathsf{V}_{_{cc^{\boldsymbol{c}}}}, \ \mathsf{V}_{_{cc\mathsf{A}}},$ and $V_{\rm cco}$ should be individually connected to the power supply plane through vias, and bypass capacitors should be used for each pin. To achieve optimum jitter performance, power supply isolation is required. Figure 2 illustrates how a 10Ω resistor along with a 10µF and a .01µF bypass capacitor should be connected to each V_{cca} pin.

FIGURE 2. POWER SUPPLY FILTERING

WIRING THE DIFFERENTIAL INPUT TO ACCEPT SINGLE ENDED LEVELS

Figure 3 shows how the differential input can be wired to accept single ended levels. The reference voltage V_REF = $V_{cc}/2$ is generated by the bias resistors R1, R2 and C1. This bias circuit should be located as close as possible to the input pin. The ratio

of R1 and R2 might need to be adjusted to position the V_REF in the center of the input voltage swing. For example, if the input clock swing is only 2.5V and V_{cc} = 3.3V, V_REF should be 1.25V and $R2/R1 = 0.609$.

 T **ERMINATION FOR LVPECL OUTPUTS**
The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

FOUT and nFOUT are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to drive 50Ω transmission lines. Matched impedance techniques

should be used to maximize operating frequency and minimize signal distortion. Figures 4A and 4B show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.

FIGURE 4A. LVPECL OUTPUT TERMINATION FIGURE 4B. LVPECL OUTPUT TERMINATION

DIFFERENTIAL CLOCK INPUT INTERFACE

The CLK /nCLK accepts LVDS, LVPECL, LVHSTL, SSTL, HCSL and other differential signals. Both Vswing and VoH must meet the VPP and VCMR input requirements. Figures 5A to 5E show interface examples for the CLK/nCLK input driven by the most common driver types. The input interfaces suggested here are examples only. Please consult with the vendor of the driver

component to confirm the driver termination requirements. For example in Figure 5A, the input termination applies for LVHSTL drivers. If you are using an LVHSTL driver from another vendor, use their termination recommendation.

FIGURE 5A. CLK/nCLK INPUT DRIVEN BY LVHSTL DRIVER

FIGURE 5C. CLK/nCLK INPUT DRIVEN BY 3.3V LVPECL DRIVER

FIGURE 5E. CLK/nCLK INPUT DRIVEN BY 3.3V LVPECL DRIVER WITH AC COUPLE

FIGURE 5D. CLK/nCLK INPUT DRIVEN BY 3.3V LVDS DRIVER

POWER CONSIDERATIONS

This section provides information on power dissipation and junction temperature for the ICS8432-111. Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the ICS8432-111 is the sum of the core power plus the power dissipated in the load(s). The following is the power dissipation for $V_{cc} = 3.3V + 5\% = 3.465V$, which gives worst case results. **NOTE:** Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)_{MAX} = V_{CC_MAX} ^{*} I_{EE_MAX} = 3.465V * 140mA = 485.1mW
- Power (outputs)_{MAX} = 30mW/Loaded Output pair If all outputs are loaded, the total power is 2 * 30mW = **60mW**

 Total Power $_{\text{max}}$ (3.465V, with all outputs switching) = 485.1mW + 60mW = 545.1mW

2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad and directly affects the reliability of the device. The maximum recommended junction temperature for the devices is 125°C.

The equation for Tj is as follows: $Tj = \theta_{JA} * Pd_total + T_A$

 Tj = Junction Temperature

 θ JA = junction-to-ambient thermal resistance

Pd_total = Total device power dissipation (example calculation is in section 1 above)

 T_A = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance $\theta_{\rm JA}$ must be used. Assuming a moderate air flow of 200 linear feet per minute and a multi-layer board, the appropriate value is 42.1°C/W per Table 7 below.

Therefore, Tj for an ambient temperature of 70°C with all outputs switching is: 70° C + 0.545W $*$ 42.1 $^{\circ}$ C/W = 93 $^{\circ}$ C. This is well below the limit of 125 $^{\circ}$ C.

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow, and the type of board (single layer or multi-layer).

Table 7. THERMAL RESISTANCE θ**JA FOR 32-PIN LQFP, FORCED CONVECTION**

3. Calculations and Equations.

The purpose of this section is to derive the power dissipated into the load.

LVPECL output driver circuit and termination are shown in Figure 6.

To calculate worst case power dissipation into the load, use the following equations which assume a 50Ω load, and a termination voltage of V_{cco} - 2V.

• For logic high, $V_{\text{OUT}} = V_{\text{OH MAX}} = V_{\text{CCO MAX}} - 0.9V$

 $(V_{\text{CCO_MAX}} - V_{\text{OH_MAX}}) = 0.9V$

• For logic low, $V_{\text{OUT}} = V_{\text{OL}_{\text{MAX}}} = V_{\text{CCO}_{\text{MAX}}} - 1.7V$

 $(V_{\text{CCO MAX}} - V_{\text{OL MAX}}) = 1.7V$

Pd_H is power dissipation when the output drives high. Pd_L is the power dissipation when the output drives low.

 $\sf{Pd_H} = [(\sf{V_{OH_MAX}} - (\sf{V_{CCO_MAX}} - 2V))/R_L]^\star\;(\sf{V_{CCO_MAX}} - \sf{V_{OH_MAX}}) = [(2\sf{V} - (\sf{V_{CCO_MAX}} - \sf{V_{OH_MAX}}))/R_L]^\star\;(\sf{V_{CCO_MAX}} - \sf{V_{OH_MAX}}) = [(2\sf{V} - (\sf{V_{CCO_MAX}} - \sf{V_{OH_MAX}}))/R_L]^\star\;(\sf{V_{CCO_MAX}} - \sf{V_{OH_MAX}})$ $[(2V - 0.9V)/50\Omega] * 0.9V = 19.8mW$

 $\sf{Pd_L} = [(V_{\sf OL_MAX} - (V_{\sf CO_MAX} \text{-} 2V)) / R_{\sf L}]^* \text{ (}V_{\sf CO_MAX} \text{-} V_{\sf OL_MAX}) = [(2V - (V_{\sf CO_MAX} \text{-} V_{\sf OL_MAX}) / R_{\sf L}]^* \text{ (}V_{\sf CO_MAX} \text{-} V_{\sf OL_MAX}) = [(V_{\sf CO_MAX} \text{-} V_{\sf CO_MAX}]^* \text{ (}V_{\sf CO_MAX} \text{-} V_{\sf CO_MAX}]^*]$ [(2V - 1.7V)/50Ω] * 1.7V = 10.2mW

Total Power Dissipation per output pair = $Pd_H + Pd_L = 30$ mW

RELIABILITY INFORMATION

TABLE 8. θ_{JA} VS. AIR FLOW TABLE FOR 32 LEAD LQFP

TRANSISTOR COUNT

The transistor count for ICS8432-111 is: 3765

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PACKAGE OUTLINE - Y SUFFIX FOR 32 LEAD LQFP

TABLE 9. PACKAGE DIMENSIONS

Reference Document: JEDEC Publication 95, MS-026

TABLE 10. ORDERING INFORMATION

NOTE: "LF" suffix to the part number are the PB-free configuration, RoHS compliant

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