

### **Boundary Conduction Mode**

# **Power Factor Correction Controller IC**

### BD7693FJ BD7694FJ

#### **General Description**

BD7693FJ and BD7694FJ are Power Factor Correction IC for AC/DC supplies the system which is suitable for all the products needing power factor improvement. The PFC adopts boundary conduction mode (BCM), and switching loss reduction and noise reduction are possible by Zero Current Detection (ZCD). This IC incorporates a circuit lowering total harmonics distortion (THD) and can support IEC61000-3-2 Class-C.

#### **Features**

- Boundary Conduction Mode PFC
- Low THD Circuit Incorporation
- Low Power Consumption
- VCC UVLO Function
- ZCD by Auxiliary Winding
- Static OVP by the VS Pin
- Error Amplifier Input Short Protection
- Stable MOSFET Gate Drive
- Soft Start

### **Applications**

Lighting Equipment, AC Adopter, TV, Refrigerator, etc.

### **Key Specifications**

■ Input VCC Voltage Range: 10 V to 38 V
 ■ Operating Current: 0.58 mA (Typ)
 ■ Operating Temperature Range: -40 °C to +105 °C

Package

SOP-J8

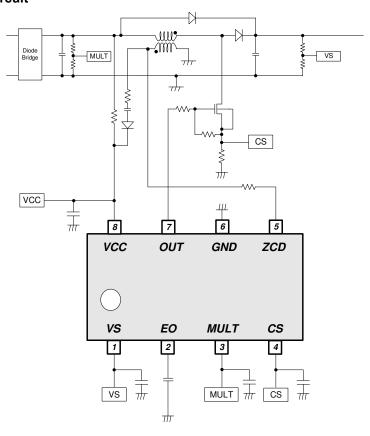
W (Typ) x D (Typ) x H (Max) 4.9 mm x 6.0 mm x 1.65 mm



#### Lineup

Product name	Brown Out
BD7693FJ-E2	-
BD7694FJ-E2	0

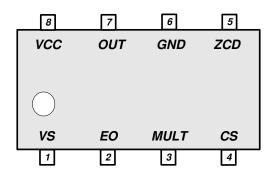
### **Typical Application Circuit**



OProduct structure : Silicon integrated circuit OThis product has no designed protection against radioactive rays.

### **Pin Configuration**

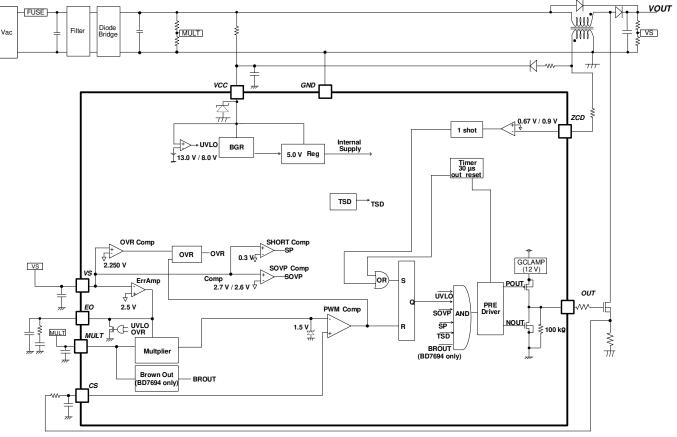
### (TOP VIEW)



### **Pin Description**

Pin No.	Pin Name	I/O	Function	ESD I	Diode
PIII NO.	Pin Name	Ď	Function	VCC	GND
1	VS	1	Feedback input pin	-	0
2	EO	0	Error amplifier output pin	-	0
3	MULT		Multiplier input pin	-	0
4	CS		Over current detection pin	-	0
5	ZCD		Zero current detection pin	-	0
6	GND	-	GND pin	0	-
7	OUT	0	External MOSFET driver pin	-	0
8	VCC		Power supply pin	-	0





### **Description of Blocks**

#### 1 VCC Protection

This IC has VCC UVLO (Under Voltage Lock Out) of the VCC pin. Switching stops at the time of VCC voltage drop. In addition, when the VCC voltage becomes higher than the  $V_{CC\_DIS1}$  (38 V Typ) voltage, it increases operating current and suppresses the rise in VCC voltage. When the VCC voltage lowers than the  $V_{CC\_DIS2}$  (34 V Typ) voltage, the operating current becomes usual. This function assumes the case that the VCC voltage rises by startup resistance.

#### 2 PFC: Power Factor Correction

The power factor improvement circuit is a voltage control method of Boundary Conduction Mode. The outline operation circuit diagram is shown in Figure 1. The switching operation is shown in Figure 2.

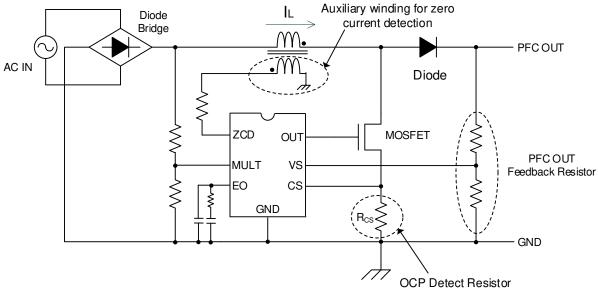


Figure 1. Operation Circuit Outline

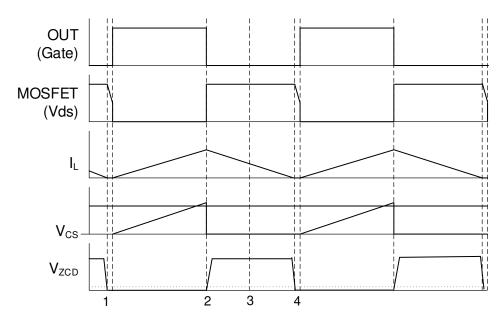


Figure 2. Switching Operation Timing Chart

### **Switching Operation**

- 1. MOSFET is turned on, and I<sub>L</sub> increases.
- 2. The IC compares Multiplier out with Vcs slope, and MOSFET is off when the Vcs voltage higher than Multiplier out.
- MOSFET is off, and I<sub>L</sub> decreases.
- 4. The ZCD pin detects a zero point of the I<sub>L</sub> and turns on MOSFET.

#### 3 About ErrAMP

#### 3.1 GmAMP

The VS pin monitors a divided point for resistance of the output voltage. The ripple voltage of AC frequency (50 Hz / 60 Hz) overlaps with the VS pin. GmAMP removes this ripple voltage. GmAMP compares  $V_{AMP1}$  (2.500 V Typ) with the removed voltage, GmAMP controls the EO voltage by this gap. When the EO pin voltage rises, ON width of the OUT pin becomes wide. When the EO voltage less than  $V_{BURST}$  (1.9 V Typ), the IC stops switching. Therefore, it can stop switching operation when the EO pin connects to the GND.

Also, you must set the error amplifier constant so that the AC frequency does not overlap on the EO pin. And, please confirm it by an actual board.

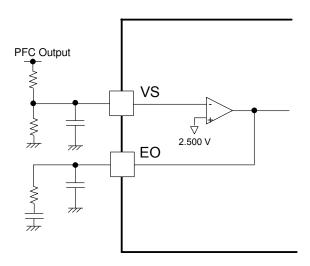


Figure 3. GmAMP Block Diagram

#### 3.2 VS Short Protection

The VS pin has a short protection function.

A state of the VS pin voltage <  $V_{SHORT}$  (0.3 V Typ) continues  $t_{VS\_SH}$  (150  $\mu s$  Typ) or more, it stops switching. Figure 4 shows the operation.

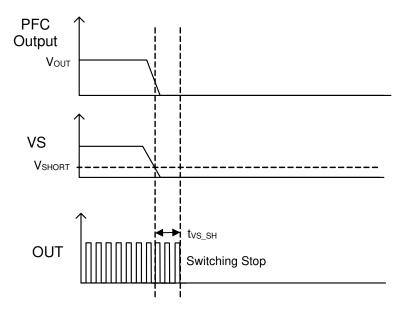


Figure 4. Operation of VS Short Protection

#### 3 About ErrAMP - continued

### 3.3 VS Overvoltage Protection Function (SOVP)

The VS pin voltage rises from  $V_{\text{OVP1}}$  (2.7 V Typ), it stops switching immediately. The VS pin voltage less than  $V_{\text{OVP2}}$  (2.6 V Typ), it starts switching. Figure 5 shows the operation.

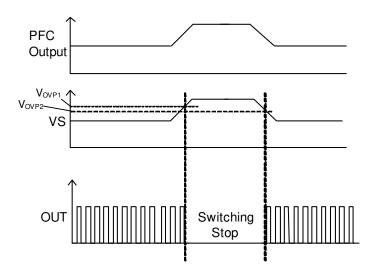


Figure 5. VS Overvoltage Protection Operation

### 3.4 Over Voltage Reduce Function at Start Up (OVR)

When the VS pin voltage performs a rise in startup to  $V_{\text{OVR}}$  (2.25 V Typ) (equivalent to -10 % of output voltage), it discharges the EO voltage to the  $V_{\text{BURST}}$  forcibly. OUT pulse width is narrows when the EO voltage falls, through rate of output voltage becomes slow and reduces over voltage in the startup. This function is effective only once after VCC UVLO cancellation.

### 4 ZCD pin

The zero current detection circuit is a function to detect a zero cross of the inductor current ( $I_L$ ) (Figure 6, 7). If the voltage at the ZCD pin becomes lower than  $V_{ZCD2}$  (0.67 V Typ) after becoming higher than  $V_{ZCD1}$  (0.9 V Typ), the OUT output becomes High after the ZCD output delay time ( $t_{ZCD}$  260 ns Typ) has elapsed. When the ZCD voltage does not reach  $V_{ZCD1}$  (0.9 V Typ), it becomes the restart timer operation. After the OUT output became Low, OUT becomes High after  $t_{REST}$  (30  $\mu$ s Typ) progress (Figure 8).

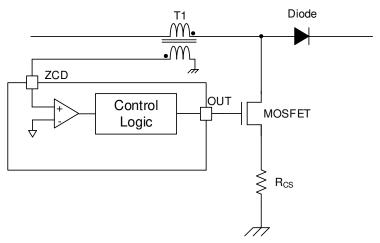


Figure 6. Zero Current Detection Circuit

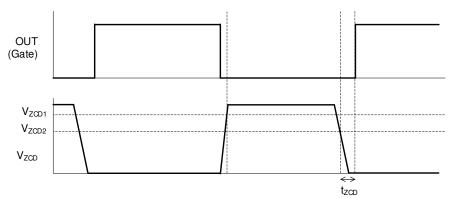


Figure 7. Zero Current Detection

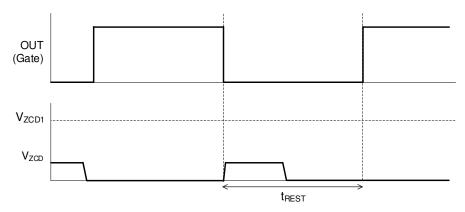


Figure 8. Restart Timer

#### 5 MULTIPLIER

The ON width of the OUT pin is fixed in Multiplier out and  $V_{\rm CS}$  as it showed in Figure 2.  $V_{\rm CS}$  is expressed in the following formula.

$$V_{CS} = K \times V_{MULT}(V_{EO} - V_{BURST})$$

K: MULTIPLIER GAIN
VMULT: MULT pin voltage
VEO: EO pin voltage
VBURST: Burst voltage

AC voltage information is input into  $V_{\text{MULT}}$ . The IC improves a power factor by controlling AC current with the AC voltage. In addition,  $V_{\text{CS}}$  in AC voltage 0 V ( $V_{\text{MULT}}$  = 0 V) is expressed in the following formula.

$$V_{CS} = K \times V_{MULT}(V_{EO} - V_{BURST}) + V_{OFFSET} = V_{OFFSET}$$

The ON width of the OUT pin at the age of AC voltage 0 V ( $V_{MULT} = 0 V$ ) becomes long by adding  $V_{OFFSET}$  (25 mV Typ). Because ON width gets longer, diode bridge output voltage is discharged. As a result, an AC current distortion is improved without the current supply from a diode bridge stopping (Figure 9).

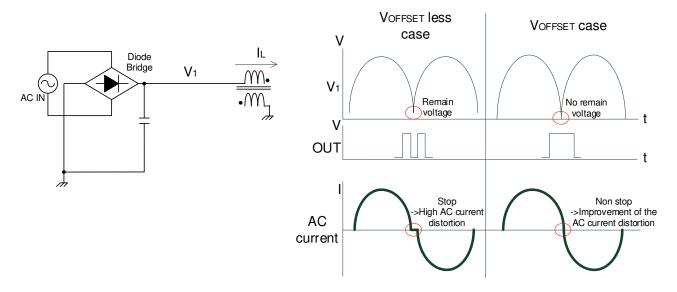


Figure 9. Improvement of the AC Current Distortion

### 6 MULT pin

When the state that the MULT pin voltage is lower than V<sub>BROUT1</sub> (0.8 V Typ) continues t<sub>BROUT</sub> (160 ms Typ) or more, the IC stops switching by a brown out function (only in BD7694).

When the MULT pin voltage becomes higher than V<sub>BROUT2</sub> (0.97 V Typ), the IC switches again.

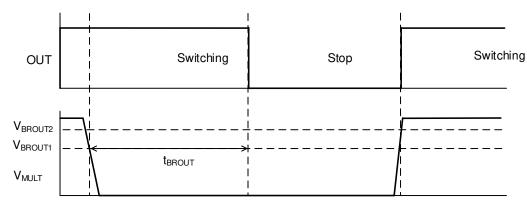


Figure 10. Brown Out

### 7 CS pin

In normal operation, turn OFF of the switching is usually decided by ON width by the EO pin and the MULT pin voltage. However, the IC is off in a pulse by pulse in overcurrent protection when the CS pin rises than  $V_{CS}$  (1.5 V Typ). By this protection, it prevents an overcurrent to MOSFET.

The overcurrent protection function limits ON width. When this protection becomes the working PFC load, PFC output voltage decreases. You must decide sense resistance of PFC so that this protection does not work in rated load with the minimum input voltage at the time of the application design.

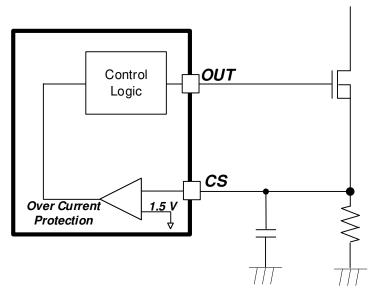


Figure 11. Current Limit

### **Operation Mode of Protection Circuit**

Table 1 showed the operation mode of each protection function.

Table 1. Operation Mode of Each Protective Circuit

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			Protection	n Mode		
Parameter	Contents	Detection Method	Detection Operation	Cancellation Method	Cancellation Operation	
VCC UVLO	Under Voltage Lock Out on the VCC pin	VCC < 8 V (Typ) (VCC drop)	OUT OFF EO discharge	VCC > 13 V (Typ) (VCC rise)	Startup Operation	
CS OCP	Over Current Protection on the CS pin	CS > 1.5 V (Typ) (CS rise)	OUT OFF	CS < 1.5 V (Typ) (CS drop)	Normal Operation	
VS Short	Short Protection on the VS pin	VS < 0.3 V (Typ) (VS drop)	OUT OFF EO discharge	VS > 0.3 V (Typ) (VS rise)	Normal Operation	
VS Static OVP	Over Voltage Protection on the VS pin	VS > 2.7 V (Typ) (VS rise)	OUT OFF	VS < 2.6 V (Typ) (VS drop)	Normal Operation	
Brown Out (Only BD7694)	Low Voltage Protection on the MULT pin	MULT < 0.8 V (Typ) (MULT drop)	OUT OFF EO discharge	MULT > 0.97 V (Typ) (MULT rise)	Normal Operation	

Absolute Maximum Ratings (Ta = 25 °C)

Parameter	Symbol	Rating	Unit	Condition
Maximum Voltage 1	$V_{MAX1}$	-0.3 to +40	V	VCC
Maximum Voltage 2	$V_{MAX2}$	-0.3 to +14	V	OUT
Maximum Voltage 3	$V_{MAX3}$	-0.3 to +6.5	V	CS, MULT, VS, EO
Maximum Current 1	I <sub>ZCD1</sub>	-10 to +10	mA	ZCD current
OUT Pin Output Peak Current 1	I <sub>OUT1</sub>	-0.5	Α	Source current
OUT Pin Output Peak Current 2	I <sub>OUT2</sub>	+1	Α	Sink current
Maximum Junction Temperature	Tjmax	+150	°C	
Storage Temperature Range	Tstg	-55 to +150	°C	

Caution 1: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

Caution 2: Should by any chance the maximum junction temperature rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, design a PCB with thermal resistance taken into consideration by increasing board size and copper area so as not to exceed the maximum junction temperature rating.

### Thermal Resistance<sup>(Note 1)</sup>

Parameter		Thermal Resi	l lmi4	
		1s <sup>(Note 3)</sup>	2s2p <sup>(Note 4)</sup>	Unit
SOP-J8				
Junction to Ambient	θја	149.3	76.9	°C/W
Junction to Top Characterization Parameter <sup>(Note 2)</sup>	$\Psi_{JT}$	18	11	°C/W

(Note 1) Based on JESD51-2A(Still-Air)

(Note 2) The thermal characterization parameter to report the difference between junction temperature and the temperature at the top center of the outside surface of the component package.

(Note 3) Using a PCB board based on JESD51-3.

(Note 4) Using a PCB board based on JESD51-7.

Layer Number of Measurement Board	Material	Board Size
Single	FR-4	114.3 mm x 76.2 mm x 1.57 mmt
Тор		
Copper Pattern	Thickness	
Footprints and Traces	70 µm	

Layer Number of Measurement Board	Material	Board Size
4 Layers	FR-4	114.3 mm x 76.2 mm x 1.6 mmt

Тор		2 Internal Lay	ers	Bottom		
Copper Pattern	Thickness	Copper Pattern	Thickness	Copper Pattern	Thickness	
Footprints and Traces	70 µm	74.2 mm x 74.2 mm	35 µm	74.2 mm x 74.2 mm	70 µm	

**Recommended Operating Conditions** 

Parameter	Symbol	Min	Тур	Max	Unit	Condition
Supply Voltage	VCC	10	15	38	V	VCC Voltage
Operating Temperature	Topr	-40	+25	+105	°C	

Recommended Range of the External Component (Ta = 25 °C)

Parameter	Symbol	Rating	Unit
VCC Pin Connection Capacity	Cvcc	22 or more	μF

Electrical Characteristics (Unless otherwise specified VCC = 15 V, Ta = -40 °C to +105 °C)

Parameter	Symbol	Min	Тур	Max	Unit	Condition
[Circuit Current]				T		
Circuit Current (ON) 1	I <sub>ON1</sub>	-	0.58	1.20	mA	VS = 0 V
Circuit Current (ON) 2	I <sub>ON2</sub>	-	0.95	2.00	mA	50 kHz switching
Circuit Current (ON) 3	Іомз	4.5	9.0	13.5	mA	VCC discharge Switching stop
Start Up Current	ISTART	-	100	200	μA	VCC = 12 V
VCC Pin Protection]					· · ·	
/CC UVLO Voltage1	V <sub>UVLO1</sub>	12	13	14	V	VCC rise
/CC UVLO Voltage2	V <sub>UVLO2</sub>	7	8	9	V	VCC drop
/CC UVLO Hysteresis	V <sub>UVLO3</sub>	3.8	5.0	6.2	V	V <sub>UVLO3</sub> = V <sub>UVLO1</sub> -V <sub>UVLO2</sub>
CC Discharge Voltage1	VCC_DIS1	0.0	38	0.2	V	VCC rise
				-		
/CC Discharge Voltage2	V <sub>CC_DIS2</sub>	-	34	-	V	VCC drop
Gm Amplifier Block]	I					1
'S Pin Pull-up Current	Ivs	-	0.1	0.5	μA	VS = 0 V
Gm Amplifier	\/ <u>.</u>	2.465	2.500	2.535	V	Ta = 25 °C
Reference Voltage1	V <sub>AMP1</sub>	∠.405	∠.500	2.535	V	14-20 6
Gm Amplifier Reference Voltage2	V <sub>AMP2</sub>	2.44	-	2.54	V	Ta = -40 °C to +105 °C
6m Amplifier Line Regulation	V <sub>AMP_LINE</sub>	_	1	10	mV	VCC = 10 V to 38 V
Sm Amplifier						
rans Conductance	Tvs	80	100	130	μA/V	EO = 2.5 V, Ta = 25 °C
			40	00		VO 0.0 V
Sm Amplifier Source Current	leo_source	5	10	20	μA	VS = 2.3 V
6m Amplifier Sink Current	IEO_SINK	5	10	20	μΑ	VS = 2.7 V
EO Block]						
EO L Voltage	V <sub>EOL</sub>	-	1.6	1.8	V	VS = 2.7 V
Burst Voltage	V <sub>BURST</sub>	1.8	1.9	-	V	
EO Discharge Current	I <sub>EO</sub>	0.8	1.8	3.0	mA	VCC = 12 V, EO = 1.0 V
MULT Block]	·LO	0.0	1.0	0.0		12 1, 20 1.0 1
MULT Pin Pull-up Current	len n =		0.1	0.5		MULT = 0 V
	IMULT	04-05		-	μA	MOLI = 0 V
MULT Pin Dynamic Range	V <sub>MULT</sub>	0 to 2.5	0 to 3.5	-	V	
		VBURST	VBURST			
EO Pin Dynamic Range	$V_{EOD}$	to	to	-	V	
		2.9	3.4			
MULTIPLIER Gain	K	0.43	0.65	0.87	1/V	MULT = 0.5 V, EO = 3.0 V
						MULT drop
Brown Out Detect Voltage1	V <sub>BROUT1</sub>	0.7	0.8	0.9	V	BD7694FJ Only
-						•
Brown Out Detect Voltage2	V <sub>BROUT2</sub>	0.87	0.97	1.07	V	MULT rise
						BD7694FJ Only
Brown Out Detect Timer	<b>t</b> brout	80	160	320	ms	BD7694FJ Only
7CD Blocks						
						ZCD rise
	V <sub>ZCD1</sub>	8.0	0.9	1.0	V	20D 113C
CD Threshold Voltage1	V <sub>ZCD1</sub>				V	
ZCD Threshold Voltage1 ZCD Threshold Voltage2	V <sub>ZCD2</sub>	0.55	0.67	0.79	V	ZCD drop
ZCD Threshold Voltage1 ZCD Threshold Voltage2 ZCD Output Delay	V <sub>ZCD2</sub>	0.55 -	0.67 260		V ns	ZCD drop
ZCD Threshold Voltage1 ZCD Threshold Voltage2 ZCD Output Delay nput H Clamp Voltage	Vzcd2 tzcd VIH	0.55 - 6.1	0.67 260 6.7	0.79 520 -	V ns V	ZCD drop  Isink = 3 mA
ZCD Threshold Voltage1 ZCD Threshold Voltage2 ZCD Output Delay nput H Clamp Voltage nput L Clamp Voltage	Vzcd2 tzcd VIH VIL	0.55 - 6.1 -0.3	0.67 260 6.7 -0.1	0.79 520 - -	V ns V V	ZCD drop
ZCD Threshold Voltage1 ZCD Threshold Voltage2 ZCD Output Delay nput H Clamp Voltage nput L Clamp Voltage Restart Timer	Vzcd2 tzcd VIH	0.55 - 6.1	0.67 260 6.7	0.79 520 -	V ns V	ZCD drop  Isink = 3 mA
ZCD Threshold Voltage1 ZCD Threshold Voltage2 ZCD Output Delay nput H Clamp Voltage nput L Clamp Voltage Restart Timer VS Protection Block]	Vzcd2 tzcd VIH VIL	0.55 - 6.1 -0.3	0.67 260 6.7 -0.1	0.79 520 - -	V ns V V	ZCD drop  Isink = 3 mA
CCD Threshold Voltage1 CCD Threshold Voltage2 CCD Output Delay nput H Clamp Voltage nput L Clamp Voltage Restart Timer VS Protection Block]	Vzcd2 tzcd Vih Vil trest	0.55 - 6.1 -0.3 15	0.67 260 6.7 -0.1 30	0.79 520 - - 45	V ns V V µs	ZCD drop  Isink = 3 mA
CCD Threshold Voltage1 CCD Threshold Voltage2 CCD Output Delay nput H Clamp Voltage nput L Clamp Voltage Restart Timer VS Protection Block] /S Short Protection	Vzcd2 tzcd VIH VIL	0.55 - 6.1 -0.3	0.67 260 6.7 -0.1	0.79 520 - -	V ns V V	ZCD drop  Isink = 3 mA
ZCD Threshold Voltage1 ZCD Threshold Voltage2 ZCD Output Delay nput H Clamp Voltage nput L Clamp Voltage Restart Timer VS Protection Block] /S Short Protection Detection Voltage	Vzcd2 tzcd Vih Vil trest	0.55 - 6.1 -0.3 15	0.67 260 6.7 -0.1 30	0.79 520 - - 45	V ns V V µs	ZCD drop  Isink = 3 mA
CCD Threshold Voltage1 CCD Threshold Voltage2 CCD Output Delay nput H Clamp Voltage nput L Clamp Voltage Restart Timer VS Protection Block] /S Short Protection Detection Voltage /S Shortstop Protection	Vzcd2 tzcd Vih Vil trest	0.55 - 6.1 -0.3 15	0.67 260 6.7 -0.1 30	0.79 520 - - 45	V ns V V µs	ZCD drop  Isink = 3 mA
ZCD Threshold Voltage1 ZCD Threshold Voltage2 ZCD Output Delay nput H Clamp Voltage nput L Clamp Voltage Restart Timer VS Protection Block] /S Short Protection Detection Voltage /S Shortstop Protection Detection Time	Vzcd2 tzcd Vih Vil trest	0.55 - 6.1 -0.3 15	0.67 260 6.7 -0.1 30 0.3	0.79 520 - - 45	V ns V V µs	ZCD drop  Isink = 3 mA
ZCD Threshold Voltage1 ZCD Threshold Voltage2 ZCD Output Delay nput H Clamp Voltage nput L Clamp Voltage Restart Timer VS Protection Block] //S Short Protection Detection Voltage //S Shortstop Protection Detection Time Over Voltage Reduce Detection	Vzcd2 tzcd Vih Vil trest  Vshort  tvs_sh	0.55 - 6.1 -0.3 15	0.67 260 6.7 -0.1 30 0.3 150 0.9 x	0.79 520 - - 45	V ns V V µs	ZCD drop  Isink = 3 mA
ZCD Threshold Voltage1 ZCD Threshold Voltage2 ZCD Output Delay nput H Clamp Voltage nput L Clamp Voltage Restart Timer VS Protection Block] //S Short Protection Detection Voltage //S Shortstop Protection Detection Time Dver Voltage Reduce Detection //oltage	Vzcd2 tzcd Vih Vil trest	0.55 - 6.1 -0.3 15 0.2 50	0.67 260 6.7 -0.1 30 0.3 150 0.9 x VAMP1	0.79 520 - - 45 0.4 300	V ns V V µs	ZCD drop  Isink = 3 mA
CCD Threshold Voltage1 CCD Threshold Voltage2 CCD Output Delay Input H Clamp Voltage Input L Clamp Voltage Restart Timer VS Protection Block] //S Short Protection Detection Voltage //S Shortstop Protection Detection Time Diver Voltage Reduce Detection //oltage //S Overvoltage Protection	Vzcd2 tzcd VIH VIL trest  Vshort  tvs_sh Vovr	0.55 - 6.1 -0.3 15	0.67 260 6.7 -0.1 30 0.3 150 0.9 x VAMP1 1.080 x	0.79 520 - - 45 0.4 300 - 1.095 x	V ns V V µs V	Isink = 3 mA Isource = -3 mA
ZCD Block] ZCD Threshold Voltage1 ZCD Threshold Voltage2 ZCD Output Delay nput H Clamp Voltage nput L Clamp Voltage Restart Timer VS Protection Block] VS Short Protection Detection Voltage VS Shortstop Protection Detection Time Dver Voltage Reduce Detection Voltage VS Overvoltage Protection Detection Voltage	Vzcd2 tzcd Vih Vil trest  Vshort  tvs_sh	0.55 - 6.1 -0.3 15 0.2 50	0.67 260 6.7 -0.1 30 0.3 150 0.9 x VAMP1	0.79 520 - - 45 0.4 300	V ns V V µs	ZCD drop  Isink = 3 mA
ZCD Threshold Voltage1 ZCD Threshold Voltage2 ZCD Output Delay nput H Clamp Voltage nput L Clamp Voltage Restart Timer VS Protection Block] VS Short Protection Detection Voltage VS Shortstop Protection Detection Time Dver Voltage Reduce Detection Voltage VS Overvoltage Protection	Vzcd2 tzcd VIH VIL trest  Vshort  tvs_sh Vovr	0.55 - 6.1 -0.3 15 - 0.2 - 1.065 x	0.67 260 6.7 -0.1 30 0.3 150 0.9 x VAMP1 1.080 x	0.79 520 - - 45 0.4 300 - 1.095 x	V ns V V µs V	Isink = 3 mA Isource = -3 mA

Electrical Characteristics (Unless otherwise specified VCC = 15 V, Ta = -40 °C to +105 °C) - continued

Parameter	Symbol	Min	Тур	Max	Unit	Condition
[CS Block]	1	I				1
CS Threshold Voltage	Vcs	1.3	1.5	1.8	V	
Output Delay Time	t <sub>DELAY</sub>	-	150	300	ns	
CS Pin Pull-up Current	Ics	-	0.15	1.00	μΑ	CS = 0 V
CS Offset Voltage	Voffset	-	25	-	mV	MULT = 0 V
[OUT Block]					-1	
OUT H Voltage	V <sub>POUTH</sub>	9.0	10.2	11.4	V	OUT = -20 mA
OUT L Voltage	V <sub>POUTL</sub>	-	-	0.8	V	OUT = +20 mA
Rise Time	tr	-	50	-	ns	OUT load capacitor = 1000 pF OUT L Voltage to 5 V
Fall Time	tf	-	50	-	ns	OUT load capacitor = 1000 pF OUT H Voltage to 5 V
OUT Pull-down Resistance	RPDOUT	50	100	150	kΩ	

### **Typical Performance Curves**

(Reference data)

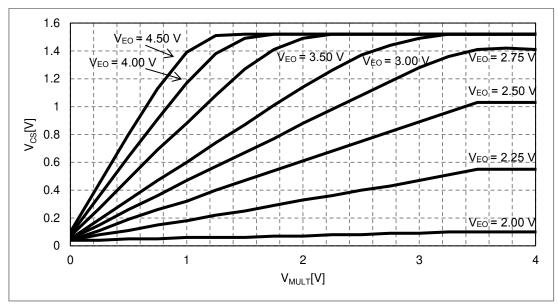


Figure 12. Vcs vs V<sub>MULT</sub>

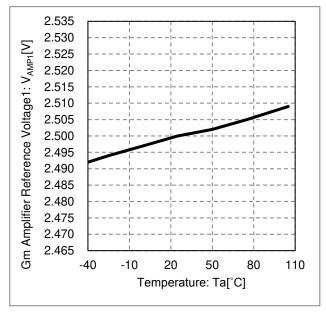


Figure 13. Gm Amplifier Reference Voltage1 vs Temperature

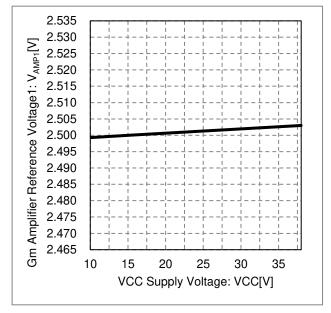


Figure 14. Gm Amplifier Reference Voltage1 vs VCC

### **Typical Performance Curves - continued**

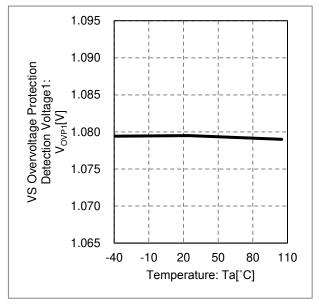


Figure 15. VS Overvoltage Protection Detection Voltage1 vs Temperature

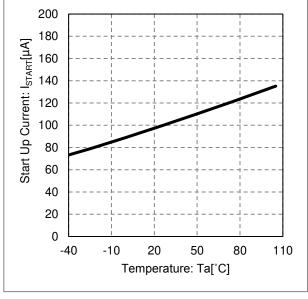


Figure 16. Start Up Current vs Temperature

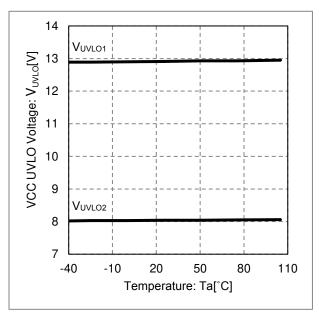


Figure 17. VCC UVLO Voltage vs Temperature

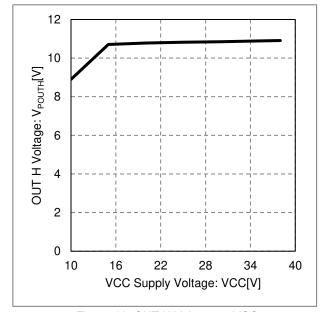


Figure 18. OUT H Voltage vs VCC

### **Application Example**

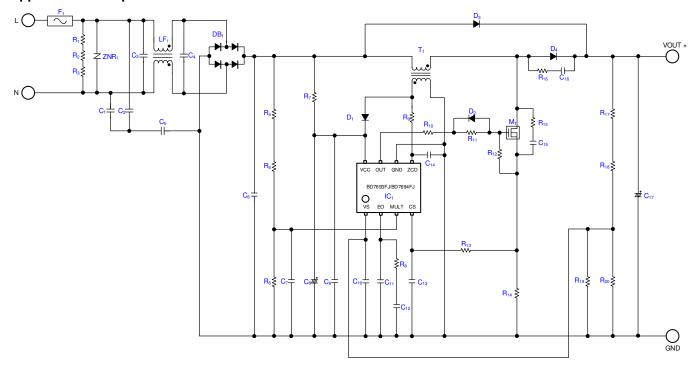


Figure 19. Application Example

#### 1 Output Voltage Setting

The output voltage is decided on feedback resistance by the VS pin.

$$V_{OUT} = \left(1 + \frac{(R_{17} + R_{18})}{(R_{19} / / R_{20})}\right) \times V_{AMP} = \left(1 + \frac{1582 \, k\Omega}{10 \, k\Omega}\right) \times 2.5 \, V = 398 \, \text{[V]}$$

 $R_{17}+R_{18}$ : Upper side resister of the output feedback  $R_{19}//R_{20}$ : Bottom side resister of the output feedback

 $V_{AMP}$ : Gm amplifier reference voltage1

#### 2 Calculation of the Inductance

Reference value in case of Vout = 400 V, Output power = 200 W

 $L = 250 [\mu H]$ 

Setting a large value of inductance will reduce the THD but increase the component size.

#### 3 External Parts of VCC

The VCC pin can reduce VCC voltage change at the time of the switching by attaching capacitor.

This IC drives gate capacitor of the external MOSFET by the OUT pin. The VCC capacitor recommends electric field capacitor 22  $\mu$ F or more withstand pressure 50 V or more.

In addition, you must confirm VCC voltage evaluation at the time of startup and the protection detection with an actual board when VCC is generated by startup resistance and the auxiliary winding of the transformer.

Because the consumption current of the IC decreases when an IC becomes the switching stop state after startup, the VCC voltage may rise by startup resistance. The overvoltage destruction of VCC is prevented by VCC voltage discharge function. The startup resistor value makes small by this function, boot-time becomes fast.

### Attention in the Board Design

About parts placement

You must locate the parts in the Figure 20 inside dot line near the IC. In addition, please do parts placement to avoid the interference with switching lines and high current lines such as inductor, DRAIN.

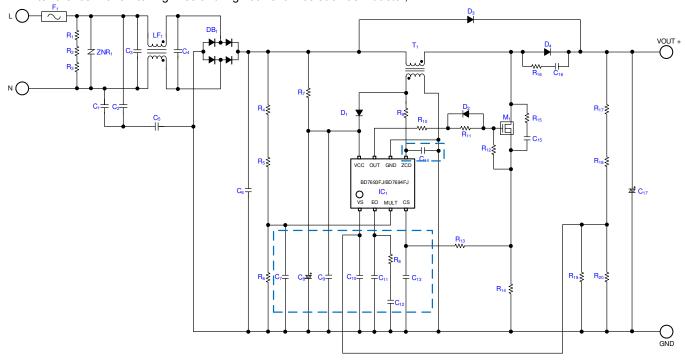


Figure 20. Parts Placement

### About GND wiring guidance

The red line of Figure 21 is the GND lines which large current flows. Draw each line as an independent wire. In addition, pull the wiring thick and short. The blue line is the GND of the IC. Make the GND of the IC and the GND of the peripheral parts common.

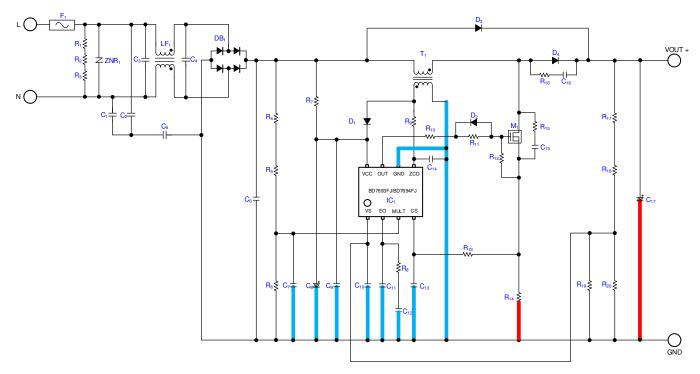


Figure 21. GND Line Layout

### Attention in the Board - continued

### About large current line

Large circuit current flows through the part of the red line of Figure 22. You must wire it short and thickly. Do not place IC and high impedance line near the red line because it has large noise.

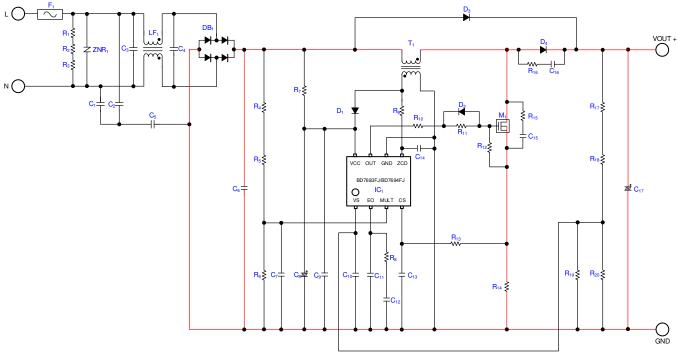
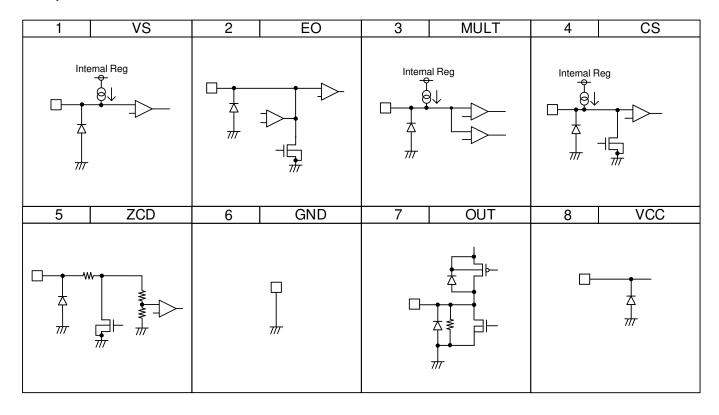


Figure 22. High Current Line Layout

### I/O Equivalence Circuits



#### **Operational Notes**

#### 1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

#### 2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

#### 3. Ground Voltage

Except for pins the output and the input of which were designed to go below ground, ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

#### 4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

#### 5. Recommended Operating Conditions

The function and operation of the IC are guaranteed within the range specified by the recommended operating conditions. The characteristic values are guaranteed only under the conditions of each item specified by the electrical characteristics.

#### 6. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

#### 7. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

#### 8. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

### 9. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

#### **Operational Notes - continued**

### 10. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode. When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

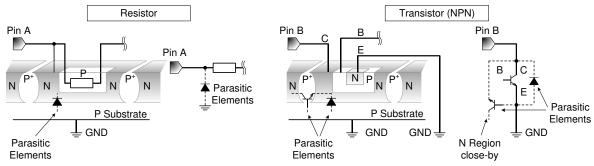


Figure 23. Example of Monolithic IC Structure

### 11. Ceramic Capacitor

When using a ceramic capacitor, determine a capacitance value considering the change of capacitance with temperature and the decrease in nominal capacitance due to DC bias and others.

#### 12. Thermal Shutdown Circuit (TSD)

This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's maximum junction temperature rating. If however the rating is exceeded for a continued period, the junction temperature (Tj) will rise which will activate the TSD circuit that will turn OFF power output pins. When the Tj falls below the TSD threshold, the circuits are automatically restored to normal operation.

Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

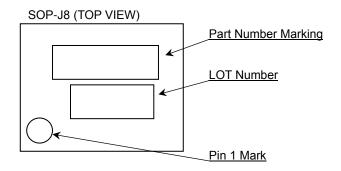
### 13. Over Current Protection Circuit (OCP)

This IC incorporates an integrated overcurrent protection circuit that is activated when the load is shorted. This protection circuit is effective in preventing damage due to sudden and unexpected incidents. However, the IC should not be used in applications characterized by continuous operation or transitioning of the protection circuit.

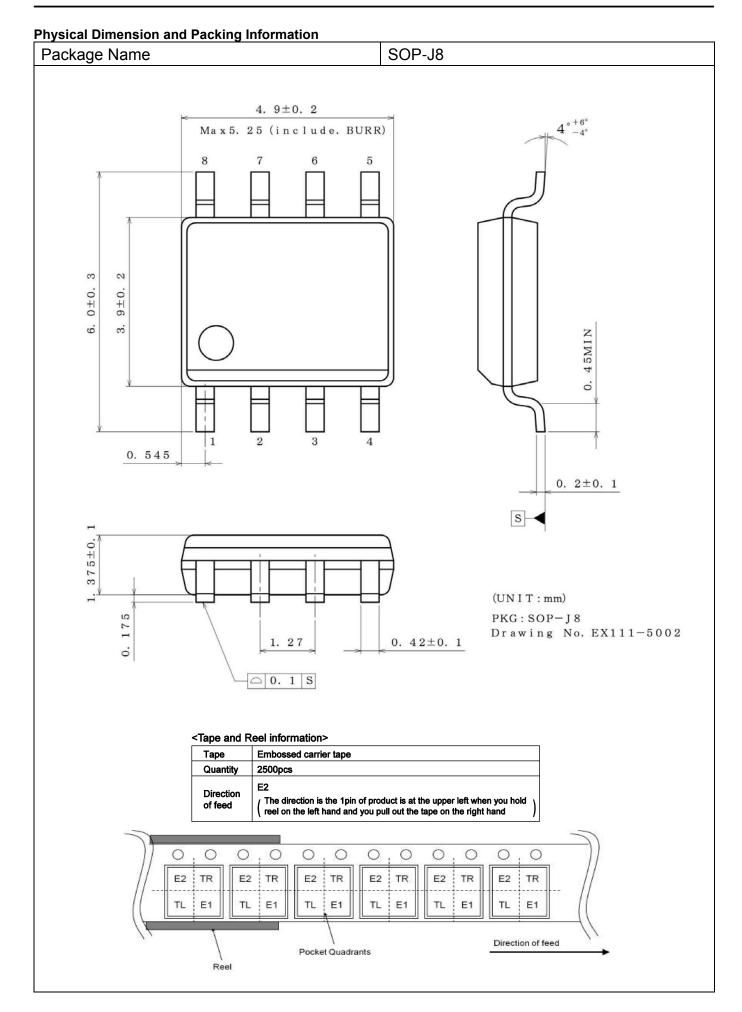
### **Ordering Information**



### **Marking Diagram**



Product name	Part Number Marking	
BD7693FJ-E2	D7693	
BD7694FJ-E2	D7694	



**Revision History** 

Date	Revision	Changes
25.Nov.2020	001	New Release

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JAPAN	USA	EU	CHINA
CLASSⅢ	CLASSIII	CLASS II b	CLASSIII
CLASSIV		CLASSⅢ	

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  - [h] Use of the Products in places subject to dew condensation
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