



The Future of Analog IC Technology®

# MP2678

## Single-Cell, Li-Ion Battery Charger Protection Circuit with Low-Dropout Mode

### DESCRIPTION

The MP2678 is a high-performance, single-cell Li-ion and Li-polymer battery charger protection circuit with low-dropout mode. Its integrated high-voltage input protection allows the IC to tolerate input surges up to 30V.

The device operates similar to a linear regulator, and maintains a regulated 5V output at input voltages up to the over-voltage protection (OVP) threshold (10.4V).

Full protection features include input OVP, MPS's proprietary battery OVP, and over-current protection (OCP). If a fault event occurs, the internal N-channel MOSFET turns off to disconnect the charging circuit from the input. Fault indication is also provided to indicate when a fault event occurs.

Thermal monitoring and thermal shutdown are provided to guarantee safe operation. The MP2678 monitors the internal temperature. If the die temperature exceeds the thermal shutdown threshold (typically 140°C), then the N-channel MOSFET bridging the input and output turns off.

The MP2678 requires a minimal number of standard, external components, and is available in a QFN-8 (2mmx2mm) package.

### FEATURES

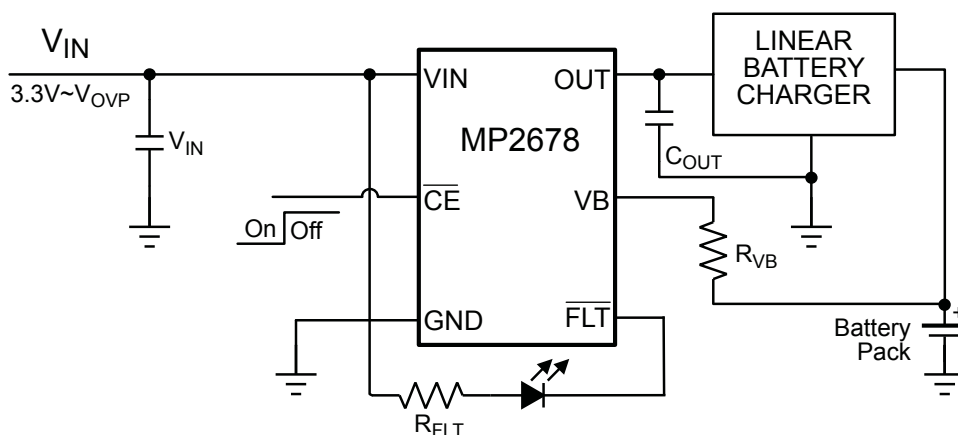
- Input Surge Up to 30V
- 5V Regulated Output
- Input Over-Voltage Protection (OVP)
- MPS Proprietary Battery OVP
- Output Short-Circuit Protection (SCP)
- Soft-Stop to Prevent Voltage Spikes
- Up to 1.7A Load Current
- Thermal Monitoring
- Thermal Shutdown
- Enable (EN) Function
- Fault Indication
- Available in a QFN-8 (2mmx2mm) Package

### APPLICATIONS

- Cell Phones
- Smartphones
- Personal Digital Assistants (PDAs)
- MP3 Players
- Digital Cameras
- Low-Power Handheld Devices

All MPS parts are lead-free, halogen-free, and adhere to the RoHS directive. For MPS green status, please visit the MPS website under Quality Assurance. "MPS", the MPS logo, and "Simple, Easy Solutions" are trademarks of Monolithic Power Systems, Inc. or its subsidiaries.

### TYPICAL APPLICATION CIRCUIT



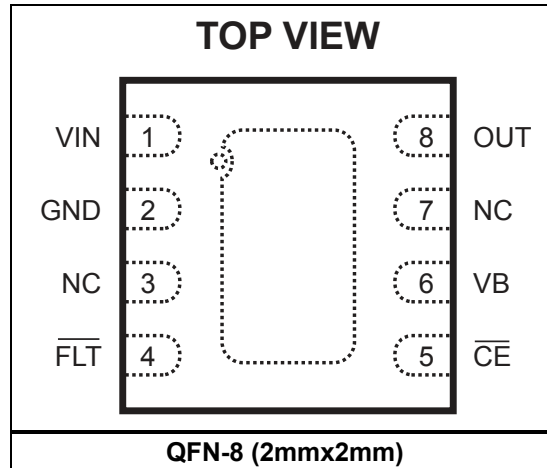
### ORDERING INFORMATION

Part Number*	OVP Voltage	Package	Top Marking	Free Air Temperature (T <sub>A</sub> )
MP2678EG	10.4V	QFN-8 (2mmx2mm)	AK	-20°C to +85°C

For Tape & Reel, add suffix -Z (e.g. MP2678EG-Z).

For RoHS-compliant packaging, add suffix -LF-Z (e.g. MP2678EG-LF-Z).

### PACKAGE REFERENCE



## PIN FUNCTIONS

Pin #	Name	Description
1	VIN	<b>Input power source.</b> The VIN pin can withstand up to a 30V input.
2	GND Exposed Pad	<b>System ground.</b> Connect the exposed pad to the GND plane for optimized thermal performance.
3, 7	NC	<b>No connect.</b> Float NC.
4	$\overline{\text{FLT}}$	<b>Open-drain output.</b> If a fault event occurs, the $\overline{\text{FLT}}$ pin goes low.
5	$\overline{\text{CE}}$	<b>Active-low enable (EN).</b> Pull the $\overline{\text{CE}}$ pin below 0.4V to turn the IC on; pull $\overline{\text{CE}}$ above 1.5V to turn it off.
6	VB	<b>Battery voltage (V<sub>BATT</sub>) monitoring input.</b> Connect the VB pin to the battery pack's positive terminal via an isolation resistor.
8	OUT	<b>Output.</b> The OUT pin is the input pin of the protected charger.

### ABSOLUTE MAXIMUM RATINGS <sup>(1)</sup>

V <sub>IN</sub> to GND .....	-0.3V to +30V
V <sub>OUT</sub> to GND .....	-0.3V to +7V
All other pins to GND .....	-0.3V to +7V
Continuous power dissipation (T <sub>A</sub> = 25°C) <sup>(2)</sup> .....	1.25W
Junction temperature .....	-20°C to +150°C
Storage temperature .....	-65°C to +150°C

### ESD Ratings <sup>(3)</sup>

HBM (human-body model) .....	2kV
MM (machine model) .....	200V

### Recommended Operating Conditions <sup>(4)</sup>

Supply input voltage (V <sub>IN</sub> ) .....	3.3V to V <sub>OVP</sub>
Output current (I <sub>OUT</sub> ) .....	1.5A
Maximum junction temp (T <sub>J</sub> ) .....	125°C

<b>Thermal Resistance <sup>(5)</sup></b>	<b>θ<sub>JA</sub></b>	<b>θ<sub>JC</sub></b>
QFN-8 (2mmx2mm) .....	80	60 ... °C/W

#### Notes:

- Exceeding these ratings may damage the device.
- The maximum allowable power dissipation is a function of the maximum junction temperature T<sub>J</sub> (MAX), the junction-to-ambient thermal resistance θ<sub>JA</sub>, and the ambient temperature T<sub>A</sub>. The maximum allowable continuous power dissipation at any ambient temperature is calculated by P<sub>D</sub> (MAX) = (T<sub>J</sub> (MAX) - T<sub>A</sub>) / θ<sub>JA</sub>. Exceeding the maximum allowable power dissipation can cause excessive die temperature, and the device may go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
- Devices are ESD sensitive. Handling precaution recommended.
- The device is not guaranteed to function outside of its operating conditions.
- Measured on JESD51-7, 4-layer PCB.

## ELECTRICAL CHARACTERISTICS

$V_{IN} = 5.5V$ ,  $T_J = -20^{\circ}C$  to  $+125^{\circ}C$ , unless otherwise noted.

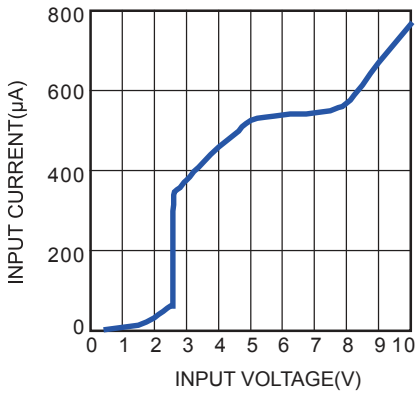
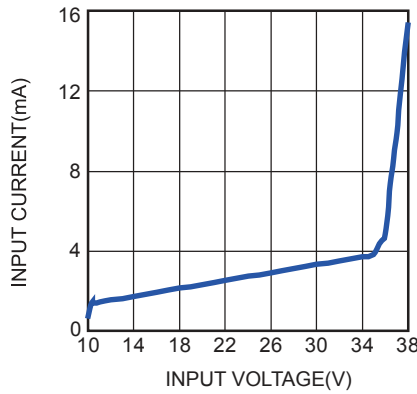
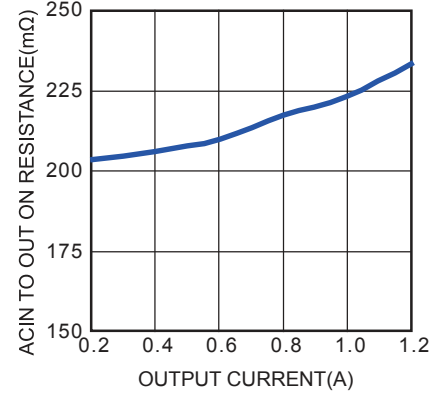
Parameter	Symbol	Condition	Min	Typ	Max	Units
<b>Power-On Reset</b>						
Output voltage	$V_{OUT}$	$5.5V < V_{IN} < V_{OVP}$ , $I_{OUT} = 1mA$	4.6	5	5.4	V
$V_{IN}$ rising threshold	$V_{POR}$		2.4		2.8	V
POR hysteresis				150		mV
Supply current	$I_{IN}$	$\overline{CE}$ is low, $V_{IN} = 5V$ , no load			600	$\mu A$
		$\overline{CE}$ is high, $V_{IN} = 5.5V$			100	$\mu A$
Input power-on blanking time	$T_{REC(Vout)}$	$V_{IN}$ rising to $V_{OUT}$ rising		8		ms
<b>Protection</b>						
Input over-voltage protection (OVP)	$V_{OVP}$	MP2678EG	9.9	10.4	10.8	V
Input OVP hysteresis				150		mV
Input OVP propagation delay <sup>(6)</sup>				1		$\mu s$
Input OVP recovery time	$T_{REC(OVP)}$			8		ms
Over-current protection (OCP)	$I_{OCP}$	$3V < V_{IN} < V_{OVP}$	1.2	1.5	1.7	A
OCP blanking time	$BT_{OCP}$			170		$\mu s$
OCP recovery time	$T_{REC(OCP)}$			50		ms
Battery OVP threshold	$V_{BOVP}$		4.23	4.35	4.5	V
Battery OVP hysteresis				150		mV
Battery OVP blanking time	$BT_{BOVP}$			176		$\mu s$
VB Leakage Current		$T_J = 25^{\circ}C$			100	nA
Over-Temperature Protection (OTP) Rising Threshold				140	150	$^{\circ}C$
OTP Falling Threshold				20		$^{\circ}C$
<b>Logic</b>						
$\overline{FLT}$ output logic low		5mA sink current		0.2		V
$\overline{FLT}$ output logic high leakage current					10	$\mu A$
$\overline{CE}$ logic low threshold	$V_{IH}$				0.4	V
$\overline{CE}$ logic high threshold	$V_{IL}$		1.5			V
<b>Input to Output Characteristics</b>						
Dropout voltage	$V_{DO}$	$V_{IN} = V_{OUT(NOM)} - 0.1V$ , $I_{OUT} = 1A$			330	mV
Q1 off-state leakage current	$I_{OFF}$	$\overline{CE}$ is high, $V_{IN} = 5.5V$			10	$\mu A$

**Note:**

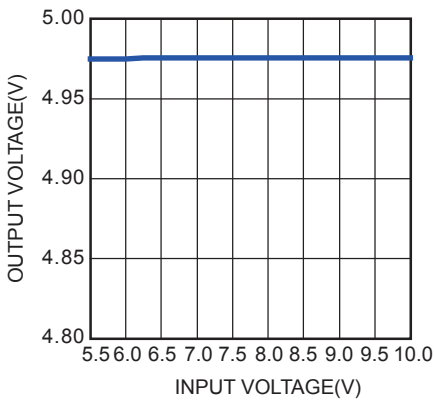
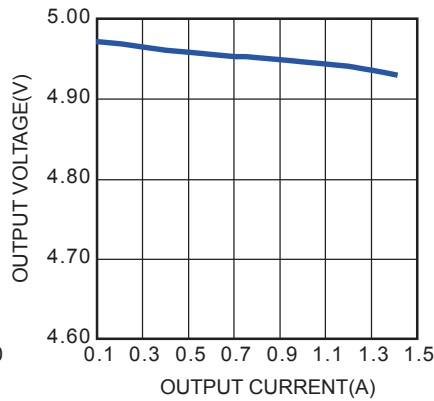
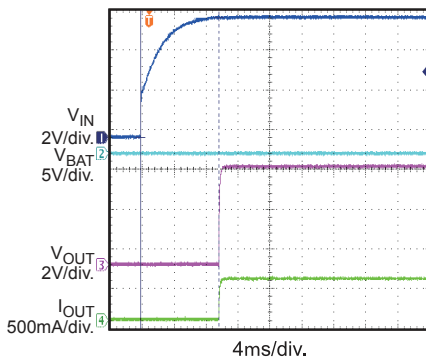
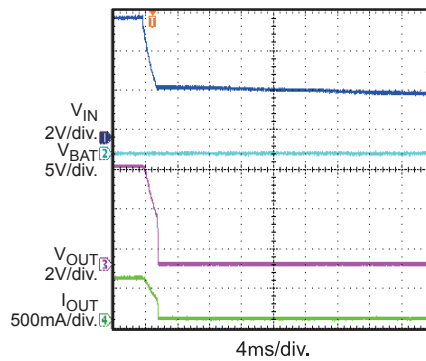
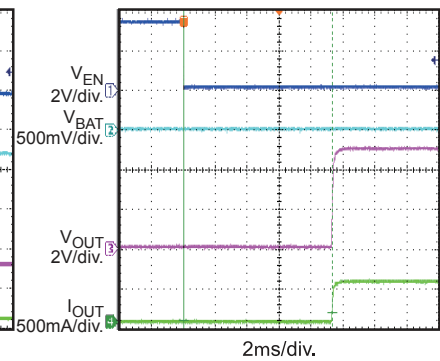
6) Guarantee by design.

## TYPICAL PERFORMANCE CHARACTERISTICS

$V_{IN} = 6V$ ,  $V_{BATT} = GND$ ,  $R_{VB} = 10k\Omega$ ,  $R_{FLT} = 6.04k\Omega$ ,  $C_{IN} = 1\mu F$ ,  $C_{OUT} = 1\mu F$ ,  $T_A = 25^\circ C$ , unless otherwise noted.

**Supply Current vs.  $V_{IN}$** 

**Breakdown Voltage**

**ACIN to OUT On Resistance vs.  $I_{OUT}$** 

**Line Regulation**

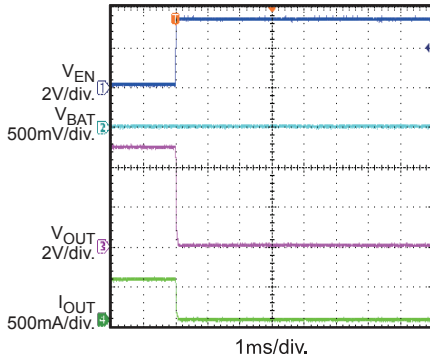
No Load


**Load Regulation**

**Power On**
 $R_{OUT}=10\Omega$ 

**Power Off**
 $R_{OUT}=10\Omega$ 

**En On**
 $R_{OUT}=10\Omega$ 


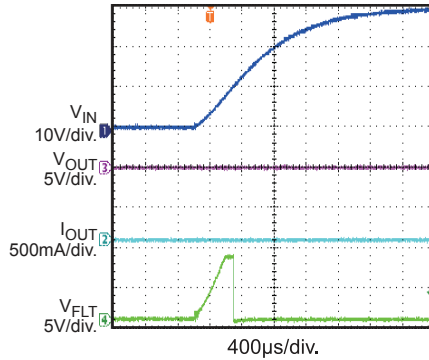
**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

$V_{IN} = 6V$ ,  $V_{BATT} = GND$ ,  $R_{VB} = 10k\Omega$ ,  $R_{FLT} = 6.04k\Omega$ ,  $C_{IN} = 1\mu F$ ,  $C_{OUT} = 1\mu F$ ,  $T_A = 25^\circ C$ , unless otherwise noted.

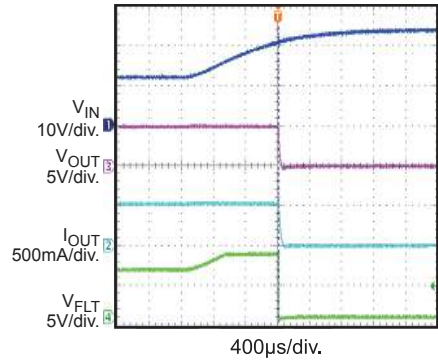
**En Off**  
 $R_{OUT}=10\Omega$



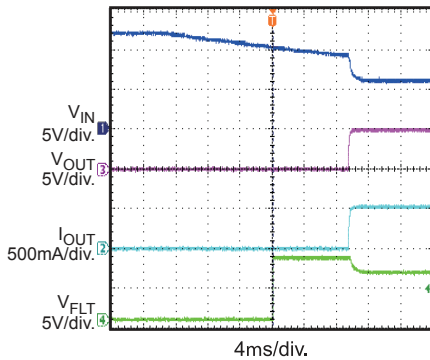
**Input OVP Start Up**  
 $V_{IN}$  jumps from 0V to 30V  
 $R_{OUT}=10\Omega$



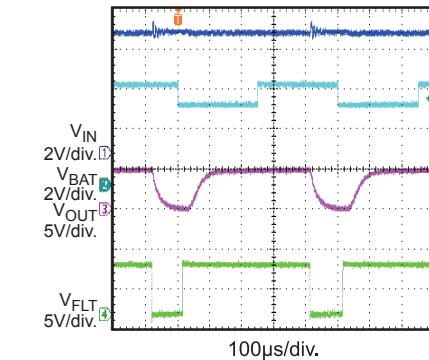
**Input OVP**  
 $V_{IN}$  steps from 6V to 12V  
 $R_{OUT}=10\Omega$



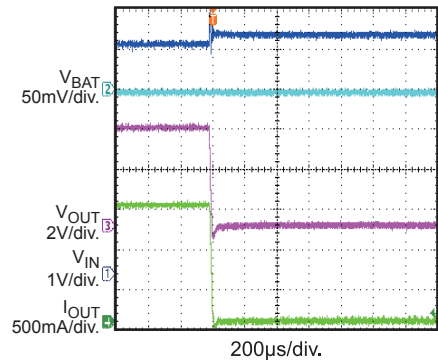
**Input OVP Recovery**  
 $V_{IN}$  drops from 12V to 6V  
 $R_{OUT}=10\Omega$



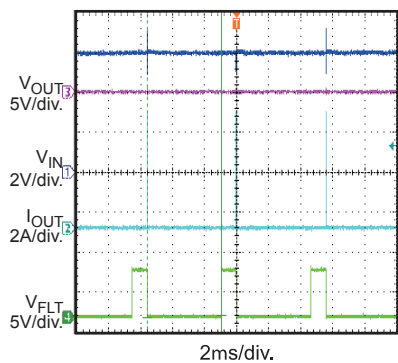
**Battery OVP**  
 $V_{BATT}$  switches between 4V and 5V  
 $R_{OUT}=10\Omega$



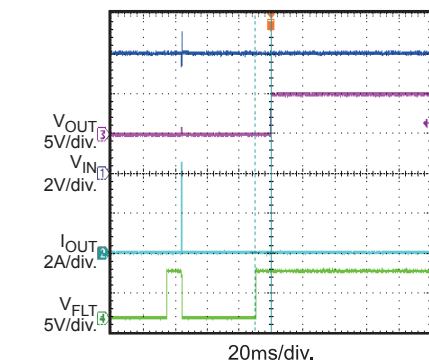
**OCP**  
Increase  $I_{OUT}$  to OCP point  
after input start up



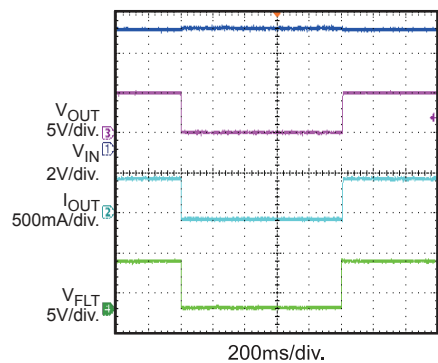
**SCP**  
Short the OUT to GND



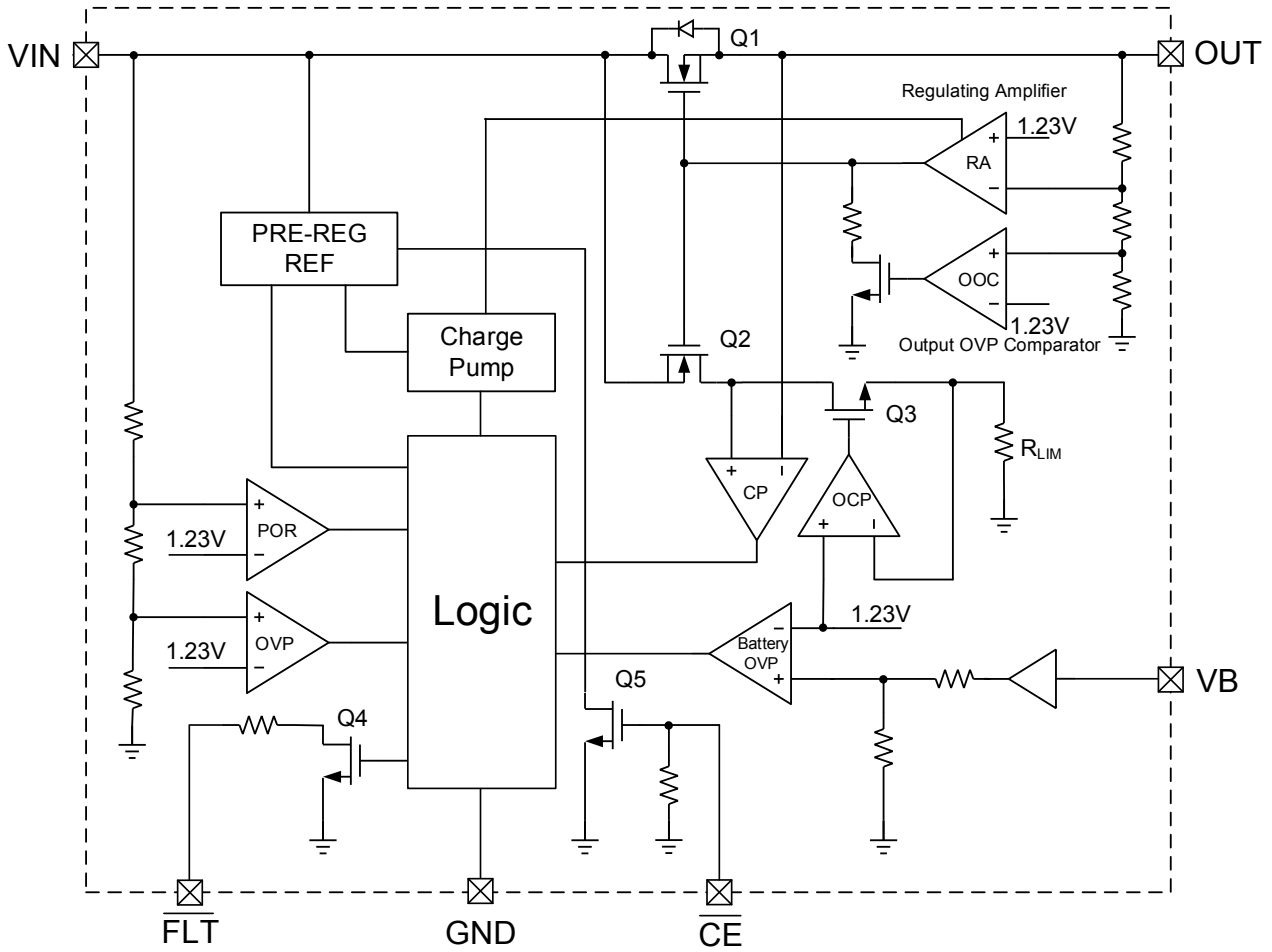
**SCP Recovery**  
Remove the short wire



**OTP**  
Increase  $T_A$  till the part shuts down  
 $R_{OUT}=10\Omega$



**FUNCTIONAL BLOCK DIAGRAM**



**Figure 1: Functional Block Diagram**

## OPERATION

The MP2678 is a high-performance, single-cell Li-ion and Li-polymer battery charger protection IC. Its integrated high-voltage input protection allows the IC to tolerate input surges up to 30V.

The device operates similar to a linear regulator, and maintains a regulated 5V output at input voltages up to the over-voltage protection (OVP) threshold (10.4V).

Full protection features include input OVP, MPS proprietary battery OVP, and over-current protection (OCP). If a fault event occurs, the internal N-channel MOSFET turns off to disconnect the charging circuit from the input. Fault indication is also provided to indicate when a fault event occurs.

Thermal monitoring and thermal shutdown are provided to guarantee safe operation. The MP2678 monitors the internal temperature. If the die temperature exceeds the thermal shutdown threshold (typically 140°C), then the N-channel MOSFET bridging the input and output turns off.

For guaranteed safe operation, the MP2678 monitors its internal temperature and turns off the N-MOSFET bridging VIN and OUT when the die temperature exceeds 140°C.

### Power-On Reset

The MP2678 has a power-on reset (POR) threshold of 2.8V, with a built-in hysteresis of 150mV. If the input voltage ( $V_{IN}$ ) drops below the POR threshold, then the internal N-channel MOSFET turns off. The IC resets itself and is followed by an 8ms delay. If  $V_{IN}$  and battery voltage ( $V_{BATT}$ ) are at safe levels, then the N-channel MOSFET turns on again. During a hot insertion of the power supply, the 8ms delay allows any transient to settle before the IC resumes normal operation.

### Input Over-Voltage Protection (OVP)

The IC's internal comparator monitors  $V_{IN}$  continuously. If the input voltage exceeds the OVP threshold ( $V_{OVP}$ ), then the internal N-channel MOSFET turns off within 1 $\mu$ s,

preventing the high  $V_{IN}$  from damaging the electronics in the handheld system. Input OVP has a typical threshold of 10.4V. Once the over-voltage (OV) condition is removed, then the N-channel MOSFET turns on again. Due to the 8ms delay, the output is not enabled during a fault condition where  $V_{IN}$  quickly exceeds  $V_{OVP}$ .

### Battery Over-Voltage Protection (OVP)

The battery OVP threshold ( $V_{BOVP}$ ) is set internally to 4.35V. The threshold has an integrated 150mV hysteresis. The VB pin monitors  $V_{BATT}$ . If  $V_{BATT}$  exceeds  $V_{BOVP}$ , then VB issues an OV signal to turn off the N-channel MOSFET. The internal comparator has a built-in blanking time (176 $\mu$ s) to prevent any transient voltage from triggering an OV fault. If the OV condition is still present after the blanking time, then the power MOSFET turns off.

### Over-Current Protection (OCP)

The internal N-channel MOSFET's current is limited. This prevents charging the battery with an excessive current. The OCP threshold is set to 1.5A by default. If an over-current (OC) condition occurs, the  $\overline{FLT}$  pin is pulled low and the timer ( $t_{REC(OCP)}$ ) begins. Once  $t_{REC(OCP)}$  expires,  $\overline{FLT}$  is pulled high, and the part restarts again after an 8ms delay.

### Thermal Shutdown

The MP2678 monitors its die temperature to prevent any thermal failure. When the internal temperature reaches 140°C, the internal N-channel MOSFET is turned off and the  $\overline{FLT}$  pin is pulled low. The IC does not resume operation until the internal temperature drops below 120°C.

### Enable (EN) Function

The MP2678 has an active-low  $\overline{CE}$  pin that enables and disables the IC. Pull  $\overline{CE}$  high to turn the internal N-channel MOSFET off; pull  $\overline{CE}$  low to turn it on and begin the start-up sequence.  $\overline{CE}$  has an internal pull-down resistor that can remain unconnected.



**Fault Indication**

The  $\overline{\text{FLT}}$  pin is an open-drain output that indicates a low signal when any of the following fault events occur:

1. Output short-circuit condition
2. Input OV condition
3. Battery OV condition
4. Thermal shutdown

If  $\overline{\text{CE}}$  goes high, then  $\overline{\text{FLT}}$  is pulled high.

## APPLICATION INFORMATION

For safe and effective charging, certain requirements have to be satisfied while charging Li-Ion batteries, such as a high, precise power source (4.2V  $\pm$ 50mV). For charging, the accuracy should be above 1%. For highly used capacity,  $V_{BATT}$  should not be charged to above 4.2V; otherwise, the performance and battery life can suffer from over-charging.

Linear battery chargers require pre-charge for depleted batteries, charge voltage, charge current, temperature monitoring, and thermal shutdown. The typical output of an MPS linear charger has an I-V curve, which provides OCP, battery OVP, input OVP, and over-temperature protection (OTP). The function of the MP2678 is to add a redundant protection layer such that, under any fault condition, the charging system output does not exceed the I-V limits required by the battery.

Additionally, the MP2678 provides full protection features for chargers that do not have complete protection (particularly for those without input surge voltage sustainability). The MP2678 guarantees the charge system's safety with its full protection features, including OVP, battery OVP, OCP, and OTP.

An internal N-channel MOSFET regulates the output voltage ( $V_{OUT}$ ) at 5V while  $V_{IN}$  reaches  $V_{OVP}$ .

The MP2678 requires a minimal number of standard, external components, in addition to the linear charger circuit as shown in the Typical Application Circuit on page 1.

### Selecting the Input Capacitor ( $C_{IN}$ ) and Output Capacitor ( $C_{OUT}$ )

The input capacitor ( $C_{IN}$ ) is used for decoupling. A higher  $C_{IN}$  capacitance reduces the voltage drop and the overshoot during transients. The AC adapter is inserted live (i.e. hot insertion). The sudden step down of the current may cause  $V_{IN}$  to overshoot. During an input OV condition, the N-channel MOSFET turns off in less than 1 $\mu$ s, which can to significant overshooting. A higher  $C_{IN}$  capacitance reduces this type of overshoot. However, overshoot caused by a hot insertion is not entirely dependent on the decoupling capacitance.

It is recommended that  $C_{IN}$  be a dielectric ceramic capacitor between 1 $\mu$ F to 4.7 $\mu$ F.

The MP2678 output and charging circuit input typically share one decoupling output capacitor ( $C_{OUT}$ ). The selection of  $C_{OUT}$  is predominately determined by the charging circuit requirement. When using the MP2602 family chargers, it is recommended that  $C_{OUT}$  be a 1 $\mu$ F to 4.7 $\mu$ F ceramic capacitor.

### Selecting the VB Resistor ( $R_{VB}$ )

$R_{VB}$  limits the current from the VB pin to the battery terminal in case the MP2678 fails. It is recommended that the  $R_{VB}$  resistance be between 200k $\Omega$  to 1M $\Omega$ . For example, during a failure operation, the worst-case scenario for the current flowing from VB to the charger output can be calculated with Equation (1):

$$(V_{VB} - V_{BATT}) / R_{VB} = 130\mu A \quad (1)$$

Where  $R_{VB}$  is 200k $\Omega$ , the VB pin voltage ( $V_{VB}$ ) is 30V, and  $V_{BATT}$  is 4.2V.

Such small currents can be absorbed easily by the bias current of other components. A larger-value resistor reduces the worst-case scenario current. However, a larger-value resistor can increase the error of the battery's OVP threshold (4.35V).

The typical VB leakage current is 20nA. the error of  $V_{BOVP}$  can be calculated with Equation (2):

$$4.35V + 20nA \times R_{VB} \quad (2)$$

With a 200k $\Omega$  resistor, the worst-case scenario for the additional error is 4mV. With a 1M $\Omega$  resistor, the worst-case additional error is 20mV.

### Selecting the $\overline{FLT}$ Resistor ( $R_{FLT}$ )

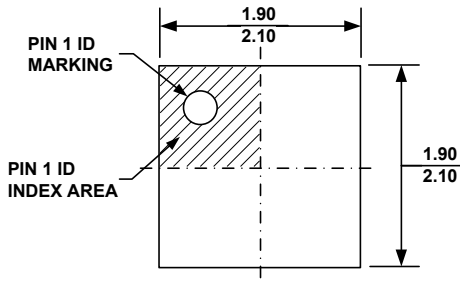
If a fault events occurs, the  $\overline{FLT}$  pin is pulled low and the pull-up resistor ( $R_{FLT}$ ) limits the sink current ( $I_{SINK}$ ). The maximum  $I_{SINK}$  should not exceed 5mA during the worst-case scenario, which means that  $V_{IN}$  should be 30V. Then the  $R_{FLT}$  can be calculated with Equation (3):

$$R_{FLT} > 30V / 5mA = 6k\Omega \quad (3)$$

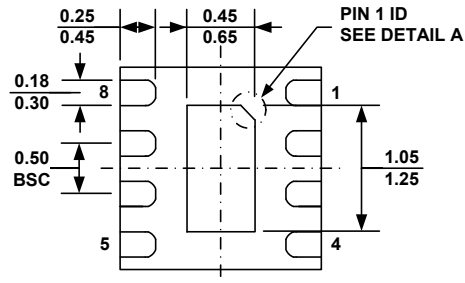
It is recommended that  $R_{FLT}$  be between 6k $\Omega$  to 200k $\Omega$ . When using an LED driver to indicate the status, select a smaller-value resistor (e.g. 6.04k $\Omega$ ) to drive the LED.

PACKAGE INFORMATION

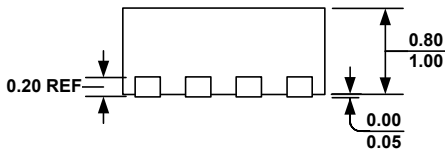
QFN-8 (2mmx2mm)



TOP VIEW

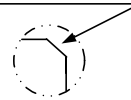


BOTTOM VIEW

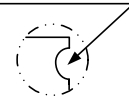


SIDE VIEW

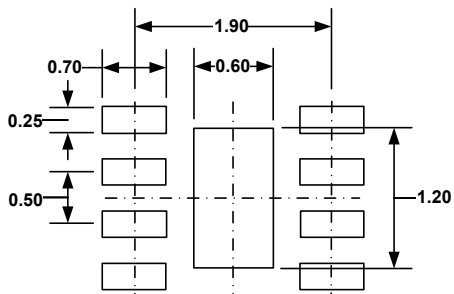
PIN 1 ID OPTION A  
0.30x45° TYP.



PIN 1 ID OPTION B  
R0.20 TYP.



DETAIL A



RECOMMENDED LAND PATTERN

**NOTE:**

- 1) ALL DIMENSIONS ARE IN MILLIMETERS.
- 2) EXPOSED PADDLE SIZE DOES NOT INCLUDE MOLD FLASH.
- 3) LEAD COPLANARITY SHALL BE 0.10 MILLIMETER MAX.
- 4) DRAWING CONFORMS TO JEDEC MO-229, VARIATION VCCD-3.
- 5) DRAWING IS NOT TO SCALE.



## REVISION HISTORY

Revision #	Revision Date	Description	Pages Updated
1.0	1/6/2011	Initial Release	-
1.1	7/19/2021	Updated the Description section; updated the footnote below the Applications section	1
		Updated the part number from “MP2678EG-104-LF-Z” to “MP2678EG-LF-Z”; removed “-XXX” suffix option; added the package information to the Package Reference section	2
		Added equation numbers	10
		Grammatical, formatting, and clerical updates; updated headers and footers; updated pagination	All

**Notice:** The information in this document is subject to change without notice. Please contact MPS for current specifications. Users should warrant and guarantee that third-party Intellectual Property rights are not infringed upon when integrating MPS products into any application. MPS will not assume any legal responsibility for any said applications.