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**FAIRCHILL** SEMICONDUCTOR

# **FXMHD103 — HDMI Voltage Translator**

### **Features**

- CEC, DDC, and HPD Level Shifting without a Direction Pin
- Host Port Voltage Supply ( $V_{\text{CCA}}$ ): 1.6V 3.6V
- HDMI Port Voltage Supply ( $V_{\text{CCC}}$ ): 4.8V  $-$  5.3V
- Long HDMI Cable Support with Integrated DDC  $(I<sup>2</sup>C)$  Edge Rate Accelerators
- Supports DDC (I<sup>2</sup>C) Clock Stretching
- Pin Out Tailored for PCB Trace Routing to HDMI Type D Connectors
- Back Drive Protection
- Non-Preferential Power-Up/Down Sequencing between VCCA and VCCC
- Operating Temperature Range: -40°C to 85°C
- ESD Protection:

- 8kV HBM (per JESD22-A114)

- 2kV CDM (per JESD22-C101)

### **Applications**

- Smart Phones
- Multimedia Phones
- Digital Camcorders
- Digital Still Cameras
- Portable Game Consoles
- Notebooks
- MP3 Players
- PC and Consumer Electronics

## **Description**

The FXMHD103 is a reduced-pin-count, low-power, High-Definition Multimedia Interface (HDMI), voltage translator for the Data Display Channel (DDC), Consumer Electronic Control (CEC), and Hot Plug Detect (HPD) control lines.

There are three non-inverting bi-directional voltage translation circuits for the DDC serial data (SDA)/clock (SCL) lines and CEC lines. Each line has a common power rail ( $V_{\text{CCA}}$ ) on the host side from 1.6V to 3.6V. On the HDMI connector side, the SCL\_C and SDA\_C pins each have an internal 1.75K $\Omega$  pull-up connected to the HDMI 5V rail,  $V_{\text{CCC}}$ . The SCL and SDA pins exceed the HDMI specification for driving up to 800pF loads. The CEC C pin has an internal 27K $\Omega$  pull-up to an internal 3.3V supply  $(V_{REG})$ .

The HPD\_C path is uni-directional. The direction is from the HDMI connector port to the host port. HPD\_H references  $V_{\text{CCA}}$ , and HPD\_C references  $V_{\text{CCC}}$ . HPD\_C offers hysteresis to avoid false detection due to bouncing while inserting the HDMI plug.

The FXMHD103 device can be powered down if the OE pin is LOW. If OE is HIGH, the HPD path is enabled. If an HDMI sink asserts the HPD\_C pin HIGH, the DDC and CEC paths are enabled. OE references  $V_{\text{CCA}}$ .

Back drive protection is provided on pins facing the HDMI connector.



### **Ordering Information**







**Figure 3. Circuit Block Diagram** 





#### **Note:**

2. SCL\_C and SDA\_C internally pulled up to V<sub>CCC</sub>. CEC\_C is 0V because V<sub>REG</sub> is disabled. This is required for HDMI compliance testing. The VOUT<sub>DIS</sub> parameter captures this requirement.

**FXMHD103 — HDMI Voltage Translator** 

FXMHD103 - HDMI Voltage Translator

# **Pin Configuration**  2 1 7 8 3 | | 4 | | 5 | | 6  $\begin{array}{ccc} \tau_1 & \tau_2 & \tau_1 & \tau_2 \\ \hline \tau_2 & \tau_3 & \tau_2 & \tau_1 \\ \hline \tau_1 & \tau_2 & \tau_1 & \tau_2 \\ \hline \tau_2 & \tau_1 & \tau_2 & \tau_1 \end{array}$ SCL\_H SDA\_H GND VCCC SDA\_C **VCCA**  $HPD$ <sup>H</sup> HPD\_C

OE



SCL\_C

CEC\_C

# **Pin Definitions**



**FXMHD103 — HDMI Voltage Translator** 

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## **Absolute Maximum Ratings**

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.



**Note:** 

3. The input and output voltage ratings may be exceeded if the input and output clamp-current ratings are observed.

### **Recommended Operating Conditions**

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to Absolute Maximum Ratings. Unless otherwise noted, values are across the recommended operating free-air temperature range.



# **Thermal Properties**

Junction-to-ambient thermal resistance is a function of application and board layout. This data is measured with fourlayer 2s2p boards in accordance to JEDEC standard JESD51. Special attention must be paid not to exceed junction temperature T**J** (maximum) at a given ambient temperature.



# **DC Electrical Characteristics (I<sub>CC</sub>)**

Unless otherwise specified,  $T_A$ =-40 to 85°C.



# **Back Drive Current**

Unless otherwise specified,  $T_A$ =-40 to 85°C.



## **Voltage Level Shifter: SCL, SDA Lines (Host/Connector Ports)**

Unless otherwise specified,  $T_A$ =-40 to 85°C.



### **Voltage Level Shifter: CEC Lines (Host/Connector Ports)**

Unless otherwise specified,  $T_A$ =-40 to 85°C.



### **Voltage Level Shifter: HPD Lines (Host/Connector Ports)**

 $T_A$ =-40 to 85°C unless otherwise specified.



# **AC Electrical Characteristics**(4)

Unless otherwise specified,  $T_A$ =-40 to 85°C. Typical values  $T_A$ = 25°C.

### **Voltage Level Shifter: SCL, SDA Lines (Host and Connector Ports); V<sub>CCA</sub>=1.8V**



## **Voltage Level Shifter: CEC Line (Host and Connector Ports); V<sub>CCA</sub>=1.8V**



### **Voltage Level Shifter: HPD Line (Host and Connector Ports); V<sub>CCA</sub>=1.8V**



### **I/O Capacitance**

 $T_A$ = 25°C unless otherwise specified.



#### **Note:**

4. AC Characteristics are guaranteed by Design and Characterization, not production tested.

# **AC Parameter Measurement Information**(5,6,7,8,9)



**Figure 5. Device Under Test Setup** 

#### **Table 2. AC Load**



#### **Notes:**

5.  $R_T$  termination resistance should be equal to  $Z_{OUT}$  of the pulse generator.<br>6.  $C_L$  includes probe and jig capacitance.

6.  $C_L$  includes probe and jig capacitance.<br>
7. All input pulses supplied by generators

All input pulses supplied by generators have the following characteristics: PRR  $\leq$  10MHz, Z<sub>o</sub>=50 $\Omega$ , slew rate  $\geq$ 1V/ns.

8. The outputs are measured one at a time, with one transition per measurement.

9.  $t_{PLH}$  and  $t_{PHL}$  are the same as  $t_{PD}$ .

**V<sub>CCA</sub>** 

GND

VOL



- Input  $t_R = t_F = 2.5$ ns, 10% to 90%, at  $V_{IN} = 3.0V$  to 3.6V only; Input  $t_R = t_F = 2.5$ ns, 10% to 90%, at  $V_{IN} = 4.5V$  to 5.5 only. 11.  $V_{\text{CCI}}=V_{\text{CCA}}$  for control pin OE or  $V_{\text{mi}}=(V_{\text{CCA}}/2)$ .
- 12. DDC Rise Times 30% 70%, CEC & HPD Rise Times 10% 90%
- 13. DDC Fall Times 30% 70%, CEC & HPD Fall Times 10% 90%
- 14.  $V_{\text{CCI}}$  is the V<sub>CC</sub> associated with the input side. V<sub>CCO</sub> is the V<sub>CC</sub> associated with the output side.

V<sub>CCO</sub>

Time

GND

Vcco

GND

Vmo

tfall

VOL

### **Application Information**

### **Power Down**

The FXMHD103 can be powered down if either  $V_{\text{CCA}}$  or  $V_{\text{CCC}}$  equals 0V, or if OE is LOW.

### **"Hot Plug" Detect Operation**

After VCCA and VCCC have powered up to valid levels, and OE enabled (HIGH) the HPD path is enabled. The internal 3.3V voltage regulator and the CEC & DDC blocks are disabled due to the internal weak pull-down resistor (100kΩ to GND) on HPD C. When the HDMI sink recognizes a valid 5V signal on the HDMI connector, to inform the HDMI source there is a valid HDMI sink connected to the HDMI connector; the sink typically ties the HPD\_C signal to the HDMI 5V supply through a 1KΩ resistor. A HIGH on HPD C, in turn, enables the internal voltage regulator, as well as the DDC & CEC paths. The HDMI link is active between the HDMI source and the HDMI sink.

When HPD C is LOW, the respective resistor pullups (RPUs) on the host and connector sides of the DDC paths remain coupled to their respective voltage references. Likewise, when HPD\_C is LOW, the RPUs on the host and connector sides of the CEC path remain coupled to their respective voltage references. Since HPD C disables  $V_{RFG}$ and  $V_{REG}$  is the CEC\_C voltage reference, CEC\_C is held to 0V by a weak (50nA) current source when HPD\_C is LOW. This is captured by the VOUT<sub>DIS</sub> parameter.

### **Backdrive Protection**

Backdrive-current protection is available on all FXMHD103 signals interfacing with the HDMI connector, including VCCC, SCL\_C, SDA\_C, CEC\_C, and HPD\_C. If the FXMHD103 is powered down,  $V_{CCA}$ =0V or  $V_{CCC}$ =0V and the HDMI sink forces  $0V - 5V$  onto any of the HDMI connector-facing pins (VCCC, SCL\_C, SDA\_C, CEC C & HPD C). The maximum current flow from the FXMHD103 is only 5µA, with the exception of 1.8µA (maximum) on CEC\_C.

### **DDC Channel Description**

The HDMI specification implements the Video Electronics Standards Association (VESA) Display Data Channel (DDC) for communication between a single HDMI source and a single HDMI sink. The DDC is used by the HDMI source to read the HDMI sinkís Enhanced Extended Display Identification Data (E-EDID) to discover the sink's configuration or capabilities. DDC must meet the <sup>2</sup>C specification, version 2.1, for Standard Mode devices. Because the HDMI application is meant for high-definition Transition-Minimized Differential Signaling (TMDS) video transport across a cable, the HDMI specification requires the DDC signals (SCL & SDA) be able to drive a minimum capacitance of 800pF (source 50pF + cable assembly 700pF + sink 50pF). The  $I^2C$  specification requires a minimum of 400pF capacitance.



### **Edge Rate Accelerators**

The FXMHD103 DDC channel is designed for highperformance  $I^2C$  level shifting. Figure 13 shows that each bi-directional channel contains an Npassgate and two dynamic drivers. This hybrid architecture is highly beneficial in an  $I^2C$  application with large capacitive loads and where auto-direction is necessary.

For example, during the following  $I^2C$  protocol events the bus direction needs to change from "Source-to-Sink" to "Sink-to-Source" without the occurrence of an edge:

- Clock Stretching
- Slave's ACK Bit (9th bit=0) following a Master's Write Bit (8th bit=0)
- Clock Synchronization and Multi Master Arbitration

If there is an  $I^2C$  translator between the source and sink in these examples, the  $I^2C$  translator must change direction when both A and C ports are LOW. The Npassgate can accomplish this efficiently because, when both A and C ports are LOW, the Npassgate acts as a low resistive short between the (A and C) ports.

Due to the  $I^2C$  open-drain topology,  $I^2C$  drivers are not push/pull devices. Logic LOWs are "pulled down" (Isink), while logic HIGHs are "let go" (3-state). For example, when the source lets go of SCL (SCL always comes from the source), the rise time of SCL is largely determined by the RC time constant, where R=RPU and C=the bus capacitance. If the FXMHD103 is attached to the source [on the A port] and there is a source on the C port, the Npassgate acts as a low-resistive short between both ports until either of the port's  $V_{CC/2}$ thresholds is reached. After the RC time constant has reached the  $V_{CC/2}$  threshold of either port, the port's edge detector triggers both dynamic drivers to drive their respective ports in the LOW-to-HIGH (LH) direction, accelerating the rising edge. The resulting rise time resembles the scope shot in Figure 14. Effectively, two distinct slew rates appear in rise time. The first slew rate (slower) is the RC time constant of the bus. The second slew rate (much faster) is the dynamic driver accelerating the edge.



If both the A and C ports of the translator are HIGH, a high-impedance path exists between the A and C ports because both the Npassgates are turned off. If a source or sink device decides to pull SCL or SDA LOW, that device's driver pulls down (Isink) SCL or SDA until the edge reaches the A or C port  $V_{\text{CC}}/2$  threshold. When either the A or C port threshold is reached, the port's edge detector triggers both dynamic drivers to drive their respective ports in the HIGH-to-LOW (HL) direction, accelerating the falling edge.

### **Driving a Capacitive Load**

The FXMHD103 dynamic drivers have enough current sourcing capability to drive an 800pF capacitive bus. The Figure 14 scope shot is of an FXMHD103 driving a lumped load of 600pF. Notice the (30% - 70%) rise time is only 112ns ( $R_{PU}$ =5K $\Omega$ ). This is well below the maximum rise time of 1000ns in Standard Mode (100KHz) or 300ns in Fast Mode (400KHz).

### **VOL vs. VIL & IOL**

The  $I^2C$  specification mandates a maximum  $V_{IL}$  ( $I_{OL}$  of 3mA) of  $V_{CC}$  x 0.3 for an I<sup>2</sup>C receiver and a maximum  $V_{OL}$  of 0.4V for an  $I^2C$  transmitter. If there is an HDMI source on the A port of an  $I^2C$  translator with a  $V_{CC}$  of 1.8V and an HDMI sink on the  $I^2C$  translator C port with a V<sub>CC</sub> of 5.0V, the maximum V<sub>IL</sub> of the source is (1.6V x 0.3) 480mV. Meanwhile, the sink could transmit a valid logic LOW of 0.4V to the source. 80mV is not very much margin between the maximum transmitted  $V_{OL}$  of 400mV (HDMI sink) to the maximum received  $V_{II}$  of 480mV (HDMI source). This appears to be an oversight in the  $I^2C$  specification, but there is an explanation. The <sup>2</sup>C specification assumes transmitters and receivers share the same  $V_{CC}$ . The  $I^2C$  specification does call out separate  $V_{OL}$  requirements vs.  $V_{CC}$  conditions where  $V_{OL1}=0.4V$  when  $V_{CC}$  is > 2.0V and  $V_{OL3}=0.2$  x  $V_{CC}$ , when  $V_{CC}$  is < 2.0V. When there is  $V_{CC}$  alignment between  $1^2C$  transmitters and receivers, the  $1^2C$ specification provides adequate  $V_{IL}$  vs.  $V_{OL}$  margins. However, when you have a transmitter operating at 5V and a receiver operating at 1.6V through a translator or level shifter, the  $V_{OL}$  vs.  $V_{IL}$  margin gets very tight, as in the above example. Therefore, the voltage drop across the  $I^2C$  translator must be as low as possible.

In general, if the  $I^2C$  translator's channel resistance is too high, the voltage drop across the translator could present a  $V_{\parallel}$  to a receiver greater than the receiver's maximum  $V_{IL}$ . To complicate matters, the  $I^2C$ specification states that 6mA of  $I_{OL}$  is recommended for bus capacitances approaching 400pF in Fast Mode. More  $I_{OL}$  increases the voltage drop across the I<sup>2</sup>C translator. The  $I^2C$  application benefits when  $I^2C$ translators exhibit low  $V_{OL}$  performance. Table 3 depicts the FXMHD103 targeted  $V_{OL}$  performance vs.  $V_{IL}/I_{OL}$ when the direction is from C side to A side,  $V_{\text{CCC}}=5.0V$ and  $V_{\text{CCA}}$ =1.6V.



# **PCB Layout Recommendation**



### **Figure 15. PCB Routing Example (Molex HDMI Type-D Connector)**



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