

Data Sheet

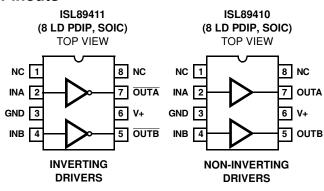
July 1, 2009

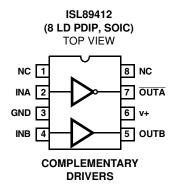
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FN6798.1
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High Speed, Dual Channel Power MOSFET Drivers

The ISL89410, ISL89411, ISL89412 ICs are similar to the EL7202, EL7212, EL7222 series but with greater VDD ratings. These are very high speed matched dual drivers capable of delivering peak currents of 2.0A into highly capacitive loads. The high speed performance is achieved by means of a proprietary "Turbo-Driver" circuit that speeds up input stages by tapping the wider voltage swing at the output. Improved speed and drive capability are enhanced by matched rise and fall delay times. These matched delays maintain the integrity of input-to-output pulse-widths to reduce timing errors and clock skew problems. This improved performance is accompanied by a 10-fold reduction in supply currents over bipolar drivers, yet without the delay time problems commonly associated with CMOS devices. Dynamic switching losses are minimized with non-overlapped drive techniques.

Pinouts





Manufactured under U.S. Patent Nos. 5,334,883, #5,341,047

Features

- · Industry Standard Driver Replacement
- Improved Response Times
- Matched Rise and Fall Times
- Reduced Clock Skew
- Low Output Impedance
- Low Input Capacitance
- High Noise Immunity
- Improved Clocking Rate
- Low Supply Current
- · Wide Operating Voltage Range
- Pb-Free Available (RoHS compliant)

Applications

- Clock/line Drivers
- CCD Drivers
- Ultra-Sound Transducer Drivers
- Power MOSFET Drivers
- · Switch Mode Power Supplies
- Class D Switching Amplifiers
- · Ultrasonic and RF Generators
- · Pulsed Circuits

Pin Descriptions

SYMBOL	PIN DESCRIPTIONS
V ₊	Power voltage from 4.5V to 18V.
GND	Power voltage return
INA, INB	Logic inputs.
OUTA OUTA	Non-inverted ouput for ISL89410. Inverted output for ISL89411 and ISL89412.
OUTB OUTB	Non-inverted output for ISL89410 and ISL89412. Inverted output for ISL89411.
NC	These pins must be left unconnected.

Ordering Information

PART NUMBER	PART MARKING	TEMP. RANGE (°C)	PACKAGE	PKG. DWG. #
ISL89410IP	ISL 89410IP	-40 to +85	8 Ld PDIP	E8.3
ISL89410IPZ (Note)	89410 IPZ	-40 to +85	8 Ld PDIP** (Pb-free)	E8.3
ISL89410IBZ (Note)	89410 IBZ	-40 to +85	8 Ld SOIC (Pb-free)	MDP0027
ISL89410IBZ-T13* (Note)	89410 IBZ	-40 to +85	8 Ld SOIC (Tape and Reel) (Pb-free)	MDP0027
ISL89411IP	ISL 89411IP	-40 to +85	8 Ld PDIP	E8.3
ISL89411IPZ (Note)	ISL 89411IPZ	-40 to +85	8 Ld PDIP** (Pb-free)	E8.3
ISL89411IBZ (Note)	89411 IBZ	-40 to +85	8 Ld SOIC (Pb-free)	MDP0027
ISL89411IBZ-T13* (Note)	89411 IBZ	-40 to +85	8 Ld SOIC (Tape and Reel) (Pb-free)	MDP0027
ISL89412IP	ISL 89412IP	-40 to +85	8 Ld PDIP	E8.3
ISL89412IPZ	89412 IPZ	-40 to +85	8 Ld PDIP** (Pb-free)	E8.3
ISL89412IBZ (Note)	89412 IBZ	-40 to +85	8 Ld SOIC (Pb-free)	MDP0027
ISL89412IBZ-T13* (Note)	89412 IBZ	-40 to +85	8 Ld SOIC (Tape and Reel) (Pb-free)	MDP0027

*Please refer to TB347 for details on reel specifications.

**Pb-free PDIPs can be used for through-hole wave solder processing only. They are not intended for use in Reflow solder processing applications NOTE: These Intersil Pb-free plastic packaged products employ special Pb-free material sets, molding compounds/die attach materials, and 100% matter tin plate plus anneal (e3 termination finish, which is RoHS compliant and compatible with both SnPb and Pb-free soldering operations). Intersil Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.

Absolute Maximum Ratings

Supply (V+ to GND)	V
Input Pins0.3V to +0.3V above V	/+
Combined Peak Output Current	A

Maximum Recommended Operating Conditions

Recommended Operating V+ Range	. 4.5V to 18.0V
Input Pins	0V to V+

Thermal Information

Operating Junction Temperature
8 Ld SOIC
8 Ld PDIP*
Storage Temperature Range
Ambient Operating Temperature40°C to +85°C
Pb-Free Reflow Profile
http://www.intersil.com/pbfree/Pb-FreeReflow.asp
*Pb-free PDIPs can be used for through-hole wave solder
processing only. They are not intended for use in Reflow solder
processing applications.

CAUTION: Do not operate at or near the maximum ratings listed for extended periods of time. Exposure to such conditions may adversely impact product reliability and result in failures not covered by warranty.

IMPORTANT NOTE: All parameters having Min/Max specifications are guaranteed. Typical values are for information purposes only. Unless otherwise noted, all tests are at the specified temperature and are pulsed tests, therefore: $T_J = T_C = T_A$

DC Electrical Specifications

 $T_A = +25^{\circ}$ C, V = 18V unless otherwise specified; Parameters with MIN and/or MAX limits are 100% tested at +25°C, unless otherwise specified. Temperature limits established by characterization and are not production tested.

PARAMETER	DESCRIPTION	TEST CONDITIONS	MIN	ТҮР	MAX	UNITS
INPUT					1	1
V _{IH}	Logic "1" Input Voltage		2.4			V
IIH	Logic "1" Input Current	@V+		0.1	10	μA
V _{IL}	Logic "0" Input Voltage				0.8	V
IIL	Logic "0" Input Current	@0V		0.1	10	μA
V _{HVS}	Input Hysteresis			0.3		V
OUTPUT		I			1	1
R _{OH}	Pull-Up Resistance	I _{OUT} = -100mA		3	6	Ω
R _{OL}	Pull-Down Resistance	I _{OUT} = +100mA		4	6	Ω
I _{PK}	Peak Output Current	Source		2		А
		Sink		2		А
IDC	Continuous Output Current	Source/Sink	100			mA
POWER SUPPLY		I			1	1
IS	Power Supply Current	Inputs High/ISL89410		4.5	7.5	mA
		Inputs High/ISL89411		1	2.5	mA
		Inputs High/ISL89412		2.5	5.0	mA
V _S	Operating Voltage		4.5		18	V

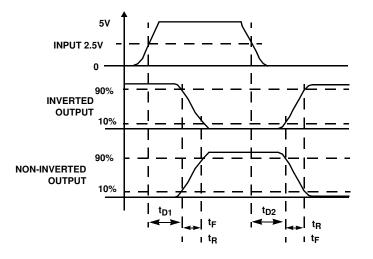
AC Electrical Specifications $T_A = +25^{\circ}C$, V = 18V unless otherwise specified.

PARAMETER	DESCRIPTION	DESCRIPTION TEST CONDITIONS N		TYP	MAX	UNITS			
SWITCHING CHARACTERISTICS									
t _R	Rise Time (Note 1)	C _L = 500pF		7.5		ns			
		C _L = 1000pF		10	20	ns			
t _F	Fall Time (Note 1)	C _L = 500pF		10		ns			
		C _L = 1000pF		13	20	ns			
t _{D1}	Turn-On Delay Time (Note 1)	See "Timing Table" on page 4		18	25	ns			
t _{D2}	Turn-Off Delay Time (Note 1)	See"Timing Table" on page 4		20	25	ns			

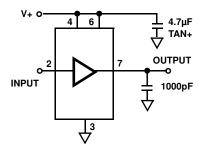
NOTE:

1. Limits established by characterization and are not production tested.

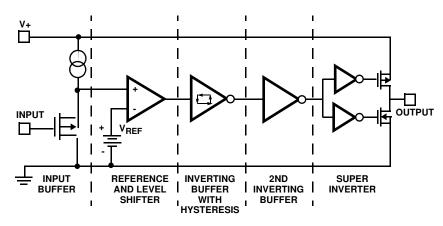
Timing Table



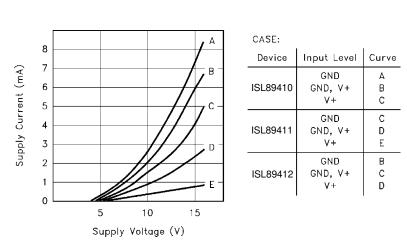
Standard Test Configuration

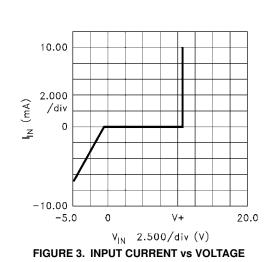


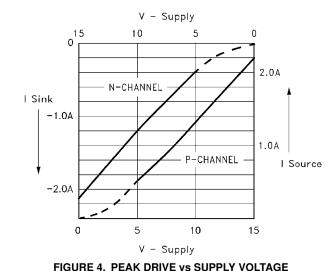
Simplified Schematic



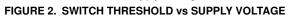


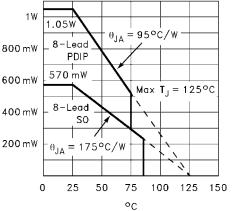




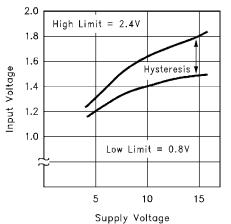


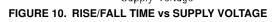


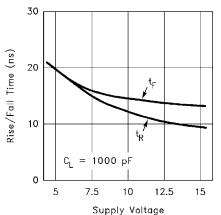




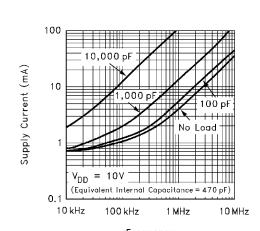
Typical Performance Curves







Frequency FIGURE 8. AVERAGE SUPPLY CURRENT vs CAPACITIVE LOAD



'ON" Resistance (N) 2 0 5 10 15 Supply Voltage FIGURE 6. "ON" RESISTANCE vs SUPPLY VOLTAGE

Measured at

100 mA

Pull Down

Typical Performance Curves (Continued)

Pull Up

8

6

4

ISL89410, ISL89411, ISL89412

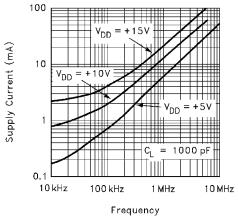


FIGURE 7. AVERAGE SUPPLY CURRENT vs VOLTAGE AND FREQUENCY

V_{DD} 0\ 80 60 40 20 100 1,000 10,000 Load Capacitance (pF) FIGURE 9. RISE/FALL TIME vs LOAD

100 Rise/Fall Time (ns)

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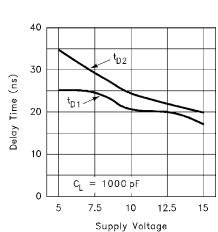


FIGURE 11. PROPAGATION DELAY vs SUPPLY VOLTAGE

Typical Performance Curves (Continued)

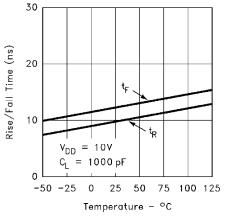


FIGURE 12. RISE/FALL TIME vs TEMPERATURE

ISL89411 Macro Model

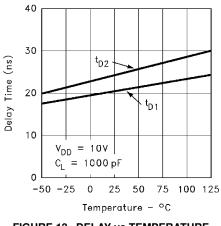
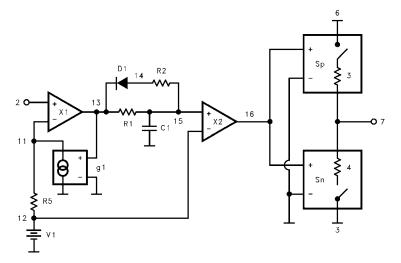


FIGURE 13. DELAY vs TEMPERATURE



** ISL89411 Model ****

input gnd Vsupply Vout 7 .subckt M89411 2 3 6 V1 12 3 1.6 R1 13 15 1k R2 14 15 5k R5 11 12 100 C1 15 3 43.3 pF D1 14 13 dmod X1 13 11 2 3 comp1 X2 16 12 15 3 comp1 sp 6 7 16 3 spmod sn 7 3 16 3 snmod g1 11 0 13 0 938µ .model dmod d .model spmod vswitch ron3 roff2meg von1 voff1.5 .model snmod vswitch ron4 roff2meg von3 voff2 .ends M89411

7

.subckt comp1 out inp inm vss e1 out vss table { (v(inp) v(inm))* 5000} (0,0) (3.2,3.2) Rout out vss 10meg Rinp inp vss 10meg Rinm inm vss 10meg .ends comp1

Application Guidelines

It is important to minimize inductance to the power FET by keeping the output drive current loop as short as possible. Also, the decoupling capacitor, Cq, should be a high quality ceramic capacitor with a Q that should be a least 10x the gate Q of the power FET. A ground plane under this circuit is also recommended.

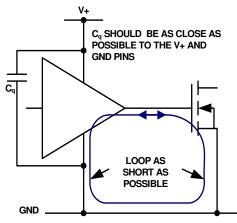


FIGURE 14. RECOMMENDED LAYOUT METHODS

In applications where it is difficult to place the driver very close to the power FET (which may result with excessive parasitic inductance), it then may be necessary to add an external gate resistor to dampen the inductive ring. If this resistor must be too large in value to be effective, then as an alternative, Schottky diodes can be added to clamp the ring voltage to V+ or GND.

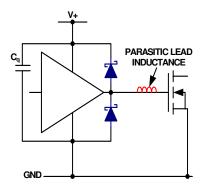


FIGURE 15. SUGGESTED CONFIGURATION FOR DRIVING INDUCTIVE LOADS

Where high supply voltage operation is required (15V to 18V), input signals with a minimum of 3.3V input drive is suggested and a minimum rise/fall time of 100ns. This is recommended to minimize the internal bias current power dissipation.

Excessive power dissipation in the driver can result when driving highly capacitive FET gates at high frequencies. These gate power losses are defined by Equation 1:

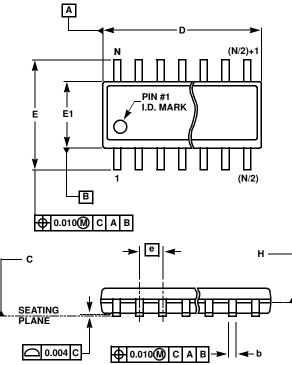
$$P = 2 \bullet Q_{C} \bullet V_{gs} \bullet f_{SW}$$
(EQ. 1)

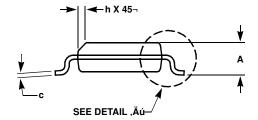
where:

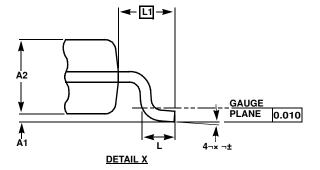
 $\label{eq:power} \begin{array}{l} \mathsf{P} = \mathsf{Power} \\ \mathsf{Q}_{c} = \mathsf{Charge of the Power FET at } \mathsf{V}_{gs} \\ \mathsf{V}_{gs} = \mathsf{Gate drive voltage (V+)} \\ \mathsf{f}_{SW} = \mathsf{switching Frequency} \end{array}$

Adding a gate resistor to the output of the driver will transfer some of the driver dissipation to the resistor. Another possible solution is to lower the gate driver voltage which also lowers Q_c .

Small Outline Package Family (SO)







MDP0027

SMALL OUTLINE PACKAGE FAMILY (SO)

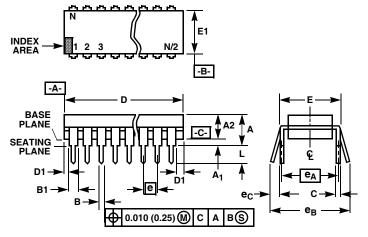
			INCHES						
SYMBOL	SO-8	SO-14	SO16 (0.150")	SO16 (0.300") (SOL-16)	SO20 (SOL-20)	SO24 (SOL-24)	SO28 (SOL-28)	TOLERANCE	NOTES
А	0.068	0.068	0.068	0.104	0.104	0.104	0.104	MAX	-
A1	0.006	0.006	0.006	0.007	0.007	0.007	0.007	±0.003	-
A2	0.057	0.057	0.057	0.092	0.092	0.092	0.092	±0.002	-
b	0.017	0.017	0.017	0.017	0.017	0.017	0.017	±0.003	-
С	0.009	0.009	0.009	0.011	0.011	0.011	0.011	±0.001	-
D	0.193	0.341	0.390	0.406	0.504	0.606	0.704	±0.004	1, 3
Е	0.236	0.236	0.236	0.406	0.406	0.406	0.406	±0.008	-
E1	0.154	0.154	0.154	0.295	0.295	0.295	0.295	±0.004	2, 3
е	0.050	0.050	0.050	0.050	0.050	0.050	0.050	Basic	-
L	0.025	0.025	0.025	0.030	0.030	0.030	0.030	±0.009	-
L1	0.041	0.041	0.041	0.056	0.056	0.056	0.056	Basic	-
h	0.013	0.013	0.013	0.020	0.020	0.020	0.020	Reference	-
Ν	8	14	16	16	20	24	28	Reference	-

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NOTES:

- 1. Plastic or metal protrusions of 0.006" maximum per side are not included.
- 2. Plastic interlead protrusions of 0.010" maximum per side are not included.
- 3. Dimensions "D" and "E1" are measured at Datum Plane "H".
- 4. Dimensioning and tolerancing per ASME Y14.5M-1994

Dual-In-Line Plastic Packages (PDIP)



NOTES:

- 1. Controlling Dimensions: INCH. In case of conflict between English and Metric dimensions, the inch dimensions control.
- 2. Dimensioning and tolerancing per ANSI Y14.5M-1982.
- 3. Symbols are defined in the "MO Series Symbol List" in Section 2.2 of Publication No. 95.
- 4. Dimensions A, A1 and L are measured with the package seated in JEDEC seating plane gauge GS-3.
- D, D1, and E1 dimensions do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.010 inch (0.25mm).
- 6. E and e_A are measured with the leads constrained to be perpendicular to datum -C-.
- 7. e_B and e_C are measured at the lead tips with the leads unconstrained. e_C must be zero or greater.
- 8. B1 maximum dimensions do not include dambar protrusions. Dambar protrusions shall not exceed 0.010 inch (0.25mm).
- 9. N is the maximum number of terminal positions.
- Corner leads (1, N, N/2 and N/2 + 1) for E8.3, E16.3, E18.3, E28.3, E42.6 will have a B1 dimension of 0.030 - 0.045 inch (0.76 - 1.14mm).

E8.3 (JEDEC MS-001-BA ISSUE D) 8 LEAD DUAL-IN-LINE PLASTIC PACKAGE

	INC	HES	MILLIM		
SYMBOL	MIN	MAX	MIN	MIN MAX	
А	-	0.210	-	5.33	4
A1	0.015	-	0.39	-	4
A2	0.115	0.195	2.93	4.95	-
В	0.014	0.022	0.356	0.558	-
B1	0.045	0.070	1.15	1.77	8, 10
С	0.008	0.014	0.204	0.355	-
D	0.355	0.400	9.01	10.16	5
D1	0.005	-	0.13	-	5
E	0.300	0.325	7.62	8.25	6
E1	0.240	0.280	6.10	7.11	5
е	0.100	BSC	2.54	BSC	-
e _A	0.300	BSC	7.62 BSC		6
е _В	-	0.430	-	10.92	7
L	0.115	0.150	2.93	3.81	4
Ν	8	3		В	9

Rev. 0 12/93

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