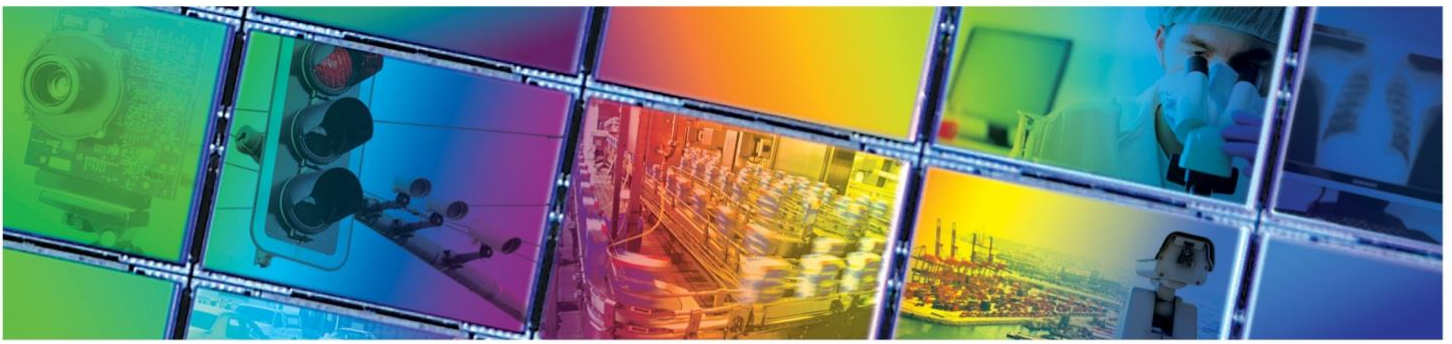


**ON Semiconductor®**



**KAC-06040 IMAGE SENSOR**  
**2832 (H) X 2128 (V) CMOS IMAGE SENSOR**



**JUNE 13, 2014**  
**DEVICE PERFORMANCE SPECIFICATION**  
**REVISION 1.1 PS-1593**



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## Summary Specification

### KAC-06040 Image Sensor

#### DESCRIPTION

The KAC-06040 Image Sensor is a high-speed 6 megapixel CMOS image sensor in a 1" optical format based on a 4.7  $\mu\text{m}$  5T CMOS platform. The image sensor features very fast frame rate, excellent NIR sensitivity, and flexible readout modes with multiple regions of interest (ROI). The readout architecture enables use of 8, 4, or 2 LVDS output banks for full resolution readout of 160 frames per second.

Each LVDS output bank consists of up to 8 differential pairs operating at 200 MHz DDR for a 400 Mbps data rate per pair. The pixel architecture allows rolling shutter operation for motion capture with optimized dynamic range or global shutter for precise still image capture.

The image sensor has a pre-configured QHD (4 x 720p, 16:9) video mode, fully programmable, multiple ROI for windowing, programmable sub-sampling, and reverse readout (flip and mirror). The two ADCs can be configured for 8-bit, 10-bit, 12-bit or 14-bit conversion and output.

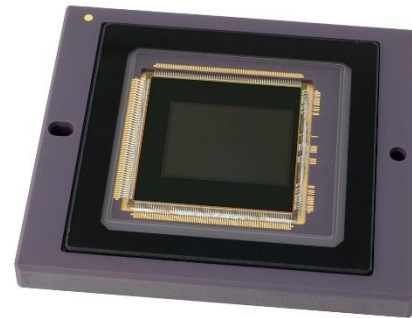
Additional features include interspersed video streams (dual-video), on-chip responsivity calibration, black clamping, overflow pixel for blooming reduction, black-sun correction (anti-eclipse), column and row noise correction, and integrated timing generation with SPI control, 4:1 and 9:1 averaging decimation modes.

#### FEATURES

- Global shutter and rolling shutter
- Very fast frame rate
- High NIR sensitivity
- Multiple regions of interest
- Interspersed video streams

#### APPLICATIONS

- Machine Vision
- Intelligent Transportation Systems
- Surveillance



Parameter	Typical Value
Architecture	5T Global Shutter CMOS
Resolution	6 megapixels
Aspect Ratio	4:3
Pixel Size	4.7 $\mu\text{m}$ (H) x 4.7 $\mu\text{m}$ (V)
Total Number of Pixels	3024 (H) x 2320 (V)
Number of Effective Pixels	2848 (H) x 2144 (V)
Number of Active Pixels	2832 (H) x 2128 (V)
Active Image Size	13.1 mm (H) x 10.0 mm (V) 16.65 mm (diag.), 1" optical format
Master Clock Input Speed	5 MHz to 50 MHz
Maximum Pixel Clock Speed	200 MHz DDR LVDS, 400 Mbps
Number of LVDS Outputs	64 differential pairs
Number of Output Banks	8, 4, or 2
Frame Rate, 6 MP	1 - 160 fps 10 bits
Charge Capacity	17,000 electrons
Quantum Efficiency	
KAC-06040-CBA	40%, 47%, 45% (470, 540, 620 nm)
KAC-06040-ABA	53%, 15%, 10% (500, 850, 900 nm)
Read Noise (at maximum LVDS clock)	3.4 $e^-$ rms, Rolling Shutter 25 $e^-$ rms, Global Shutter
Dynamic Range	74 dB, Rolling Shutter 57 dB, Global Shutter
Blooming Suppression	>10,000x
Image Lag	1.6 electron
Digital Core Supply	2.0 V
Analog Core Supply	1.8 V
Pixel Supply	2.8 V & 3.5 V
Power Consumption	2.3 W for 6 Mp @ 160 fps 10 bits
Package	267 pin ceramic micro-PGA
Cover Glass	AR Coated, 2-sides

All parameters are specified at T = 40 °C unless otherwise noted



## Ordering Information

### KAC-06040 IMAGE SENSOR

Catalog Number	Product Name	Description	Marking Code
4H2284	KAC-06040-ABA-JD-AA	Monochrome, micro-PGA Package, Sealed Clear Cover Glass with AR coating(both sides), Standard Grade	KAC-06040-ABA Serial Number
4H2286 (1)	KAC-06040-ABA-JD-AE	Monochrome, micro-PGA Package, Sealed Clear Cover Glass with AR coating(both sides), Engineering Grade	KAC-06040-ABA Serial Number
4H2287	KAC-06040-CBA-JD-AA	Bayer (RGB) Color Filter Pattern, micro-PGA Package, Sealed Clear Cover Glass with AR coating(both sides), Standard Grade	KAC-06040-CBA Serial Number
4H2289 (1)	KAC-06040-CBA-JD-AE	Bayer (RGB) Color Filter Pattern, micro-PGA Package, Sealed Clear Cover Glass with AR coating(both sides), Engineering Grade	KAC-06040-CBA Serial Number

Notes:

1. Engineering Grade samples might not meet final production testing limits, especially for cosmetic defects such as clusters, but also possibly column and row artifacts. Overall performance is representative of final production parts.

## EVALUATION SUPPORT

Catalog Number	Product Name	Description
4H2306	KEK-TBD-KAC-06040-CB	Evaluation Hardware for KAC-06040 Image Sensor (Bayer). Includes Image Sensor.
4H2307	KEK-TBD-KAC-06040-AB	Evaluation Hardware for KAC-06040 Image Sensor (monochrome). Includes Image Sensor.
4H2211	Lens Mount Kit	Lens Mount Kit that supports C, CS, and F mount lenses. Includes IR cut-filter for color imaging.

See Application Note Product Naming Convention for a full description of the naming convention used for image sensors. For reference documentation, including information on evaluation kits, please visit our web site at [www.truesenseimaging.com](http://www.truesenseimaging.com).

Please address all inquiries and purchase orders to:

Truesense Imaging, Inc.  
1964 Lake Avenue  
Rochester, New York 14615

Phone: (585) 784-5500  
E-mail: [info@truesenseimaging.com](mailto:info@truesenseimaging.com)

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## Device Description

### ARCHITECTURE

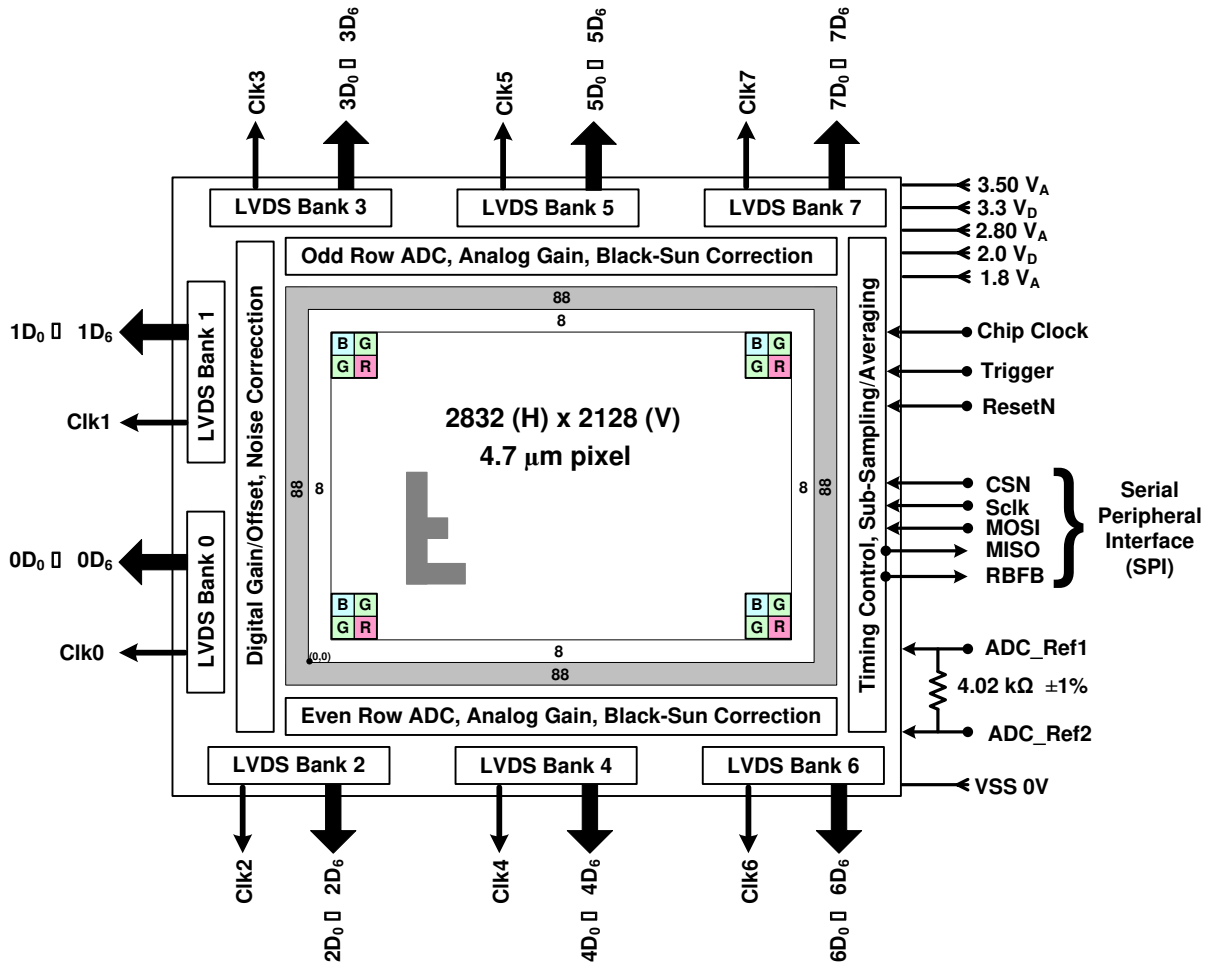


Figure 1: Block Diagram



**PHYSICAL ORIENTATION**

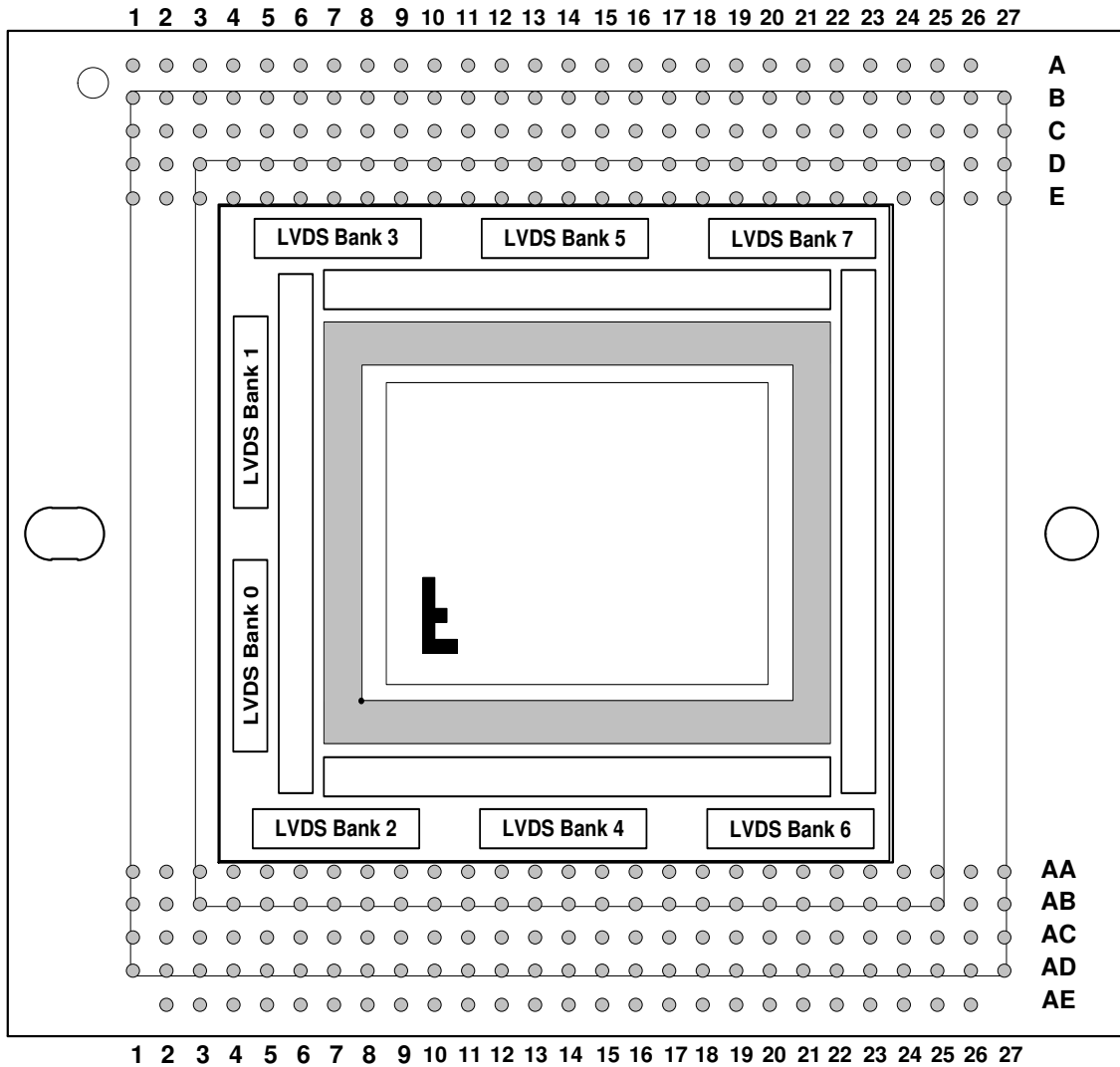


Figure 2: Package Pin Orientation – Top x-ray view

Notes:

1. The center of the pixel array is aligned to the physical package center.
2. The region under the sensor die is clear of pins enabling the use of a heat sink.
3. Non-symmetric mounting holes provide orientation and mounting precision.
4. Non-symmetric pins prevent incorrect placement in PCB.
5. Letter "F" indicator shows default readout direction relative to package pin 1





## PRIMARY PIN DESCRIPTION

Pin	Name	Type	Description
AB09	RESETN	DI	Sensor reset (0 V = Reset State)
E07	CLK_In1	DI	Sensor Input Clk_In1 (5 – 50 MHz)
D08	CLK_In2	DI	Sensor Input Clk_In2 (connect to Clk1)
AB08	TRIGGER	DI	Trigger input (optional)
AA05	SCLK	DI	SPI Master Clock
AA06	CSN	DI	SPI Chip Select (0 V = Selected)
AA07	MISO	DO	SPI Master Input, Slave Output
AA08	MOSI	DI	SPI Master Output Slave Input
AB05	FB	DO	SPI register read feedback
D07	SPI_MS	DI	SPI CPOL/CPHA mode select
AA14	ADC_Ref1	AO	4.02 kΩ ±1% resistor between Ref1 & Ref2
AA15	ADC_Ref2	AO	4.02 kΩ ±1% resistor between Ref1 & Ref2
AB06	FLO	DO	Flash output sync (optional)
AB07	MSO	DO	Mechanical Shutter output sync (optional)
E05	FEN	DO	Frame ENable reference output (optional)
E06	LEN	DO	Line ENable reference output (optional)

### Notes:

1. DI = Digital Input, DO = Digital Output, AO = Analog Output
2. Tie unused DI pins to Ground, No Connect (NC) unused DO pins
3. By default Clk\_In2 should equal Clk\_In1 and should be the same source clock.
4. The RESETN pin has a 62 kΩ internal pull-up resistor, so if left floating the chip will not be in reset mode.
5. The TRIGGER pin has an internal 100 kΩ pull down resistor. If left floating (and at default polarity) then the sensor state will not be affected by this pin (ie defaults to 'not triggered' mode if floated)
6. All of the DI and DO pins nominally operate at 0 V → 2.0 V and are associated with the VDD\_DIG power supply.
7. The SPI\_MS pin has an internal 100 kΩ pull down resistor. If left floating the CPOL/CPHA will be compatible with CPOL=CPHA=0 or CPOL=CPHA=1

## POWER PIN DESCRIPTION

Name	Voltage	Pins	Description
VDD_LVDS	3.3V D	C04, C05, C23, C24, D04, D24, E04, E24, AA04, AA24, AB04, AB24, AC04, AC05, AC23, AC24	LVDS output supply
VDD_DIG	2.0V D	C18, C19, C20, C21, C22, D18, D19, D20, D21, D22, D23, E08, E18, E20, E21, E22, AA18, AA20, AA21, AA22, AB18, AB19, AB20, AB21, AB22, AB23, AC18, AC19, AC20, AC21, AC22, AB15	Digital core supply
VDD_HV	3.5V A	C11, D11, E11, AA11, AB11, AC11, C10, D10, E10, AA10, AB10, AC10	Pixel supply 1
Vref_P	2.8V A	C13, D13, E13, AA13, AB13, AC13	Pixel supply 2
AVDD_LV	1.8V A	C17, D16, D17, E17, AA17, AB16, AB17, AC17	Analog low voltage supply
Vpixel_low	0 V	E09	Pixel Supply 3. Combine with VSS for normal operation. Can be pulsed for Extended Dynamic Range Operation
VSS	0 V	A02, A14, A26, B14, C03, C06, C12, C14, C25, D03, D12, D14, D25, E03, E12, E19, E23, E25, AA03, AA12, AA19, AA23, AA25, AB03, AB12, AB14, AB25, AC03, AC06, AC12, AC14, AC25, AD14, AE02, AE14, AE26, D15, E15, AA09	Sensor ground reference
No Connect	NA	A01, E14, E16, C09, D09, D05, D06, AA16, AC09	Unused and test-only pins. These pins must be floated.



### LVDS PIN DESCRIPTION

Pin	Name	Descr	Pin	Name	Descr	Pin	Name	Descr	Pin	Name	Descr
E01	1DCLK+	Bank 1 LVDS Clock	C07	3DCLK+	Bank 3 LVDS Clock	C15	5DCLK+	Bank 5 LVDS Clock	A22	7DCLK+	Bank 7 LVDS Clock
E02	1DCLK-		C08	3DCLK-		C16	5DCLK-		B22	7DCLK-	
D01	1DATA0+	Bank 1 LVDS Data	A07	3DATA0+	Bank 3 LVDS Data	A15	5DATA0+	Bank 5 LVDS Data	A23	7DATA0+	Bank 7 LVDS Data
D02	1DATA0-		B07	3DATA0-		B15	5DATA0-		B23	7DATA0-	
C01	1DATA1+		A08	3DATA1+		A16	5DATA1+		A24	7DATA1+	
C02	1DATA1-		B08	3DATA1-		B16	5DATA1-		B24	7DATA1-	
B01	1DATA2+		A09	3DATA2+		A17	5DATA2+		A25	7DATA2+	
B02	1DATA2-		B09	3DATA2-		B17	5DATA2-		B25	7DATA2-	
A03	1DATA3+		A10	3DATA3+		A18	5DATA3+		B27	7DATA3+	
B03	1DATA3-		B10	3DATA3-		B18	5DATA3-		B26	7DATA3-	
A04	1DATA4+		A11	3DATA4+		A19	5DATA4+		C27	7DATA4+	
B04	1DATA4-		B11	3DATA4-		B19	5DATA4-		C26	7DATA4-	
A05	1DATA5+		A12	3DATA5+		A20	5DATA5+		D27	7DATA5+	
B05	1DATA5-		B12	3DATA5-		B20	5DATA5-		D26	7DATA5-	
A06	1DATA6+		A13	3DATA6+		A21	5DATA6+		E27	7DATA6+	
B06	1DATA6-		B13	3DATA6-		B21	5DATA6-		E26	7DATA6-	

Pin	Name	Descr	Pin	Name	Descr	Pin	Name	Descr	Pin	Name	Descr
AA01	0DCLK+	Bank 0 LVDS Clock	AC07	2DCLK+	Bank 2 LVDS Clock	AC15	4DCLK+	Bank 4 LVDS Clock	AE22	6DCLK+	Bank 6 LVDS Clock
AA02	0DCLK-		AC08	2DCLK-		AC16	4DCLK-		AD22	6DCLK-	
AB01	0DATA0+	Bank 0 LVDS Data	AE07	2DATA0+	Bank 2 LVDS Data	AE15	4DATA0+	Bank 4 LVDS Data	AE23	6DATA0+	Bank 6 LVDS Data
AB02	0DATA0-		AD07	2DATA0-		AD15	4DATA0-		AD23	6DATA0-	
AC01	0DATA1+		AE08	2DATA1+		AE16	4DATA1+		AE24	6DATA1+	
AC02	0DATA1-		AD08	2DATA1-		AD16	4DATA1-		AD24	6DATA1-	
AD01	0DATA2+		AE09	2DATA2+		AE17	4DATA2+		AE25	6DATA2+	
AD02	0DATA2-		AD09	2DATA2-		AD17	4DATA2-		AD25	6DATA2-	
AE03	0DATA3+		AE10	2DATA3+		AE18	4DATA3+		AD26	6DATA3+	
AD03	0DATA3-		AD10	2DATA3-		AD18	4DATA3-		AD27	6DATA3-	
AE04	0DATA4+		AE11	2DATA4+		AE19	4DATA4+		AC26	6DATA4+	
AD04	0DATA4-		AD11	2DATA4-		AD19	4DATA4-		AC27	6DATA4-	
AE05	0DATA5+		AE12	2DATA5+		AE20	4DATA5+		AB26	6DATA5+	
AD05	0DATA5-		AD12	2DATA5-		AD20	4DATA5-		AB27	6DATA5-	
AE06	0DATA6+		AE13	2DATA6+		AE21	4DATA6+		AA26	6DATA6+	
AD06	0DATA6-		AD13	2DATA6-		AD21	4DATA6-		AA27	6DATA6-	

Notes:

1. All LVDS Data and Clock lines must be routed with 100 Ω differential transmission line traces.
2. All the traces for a single LVDS Bank should be the same physical length to minimize skew between the clock and data lines.
3. In 2 Bank mode, only LVDS banks 0 and 1 are active.
4. In 4 Bank mode, only LVDS bank 0, 1, 2, and 3 are active.
5. Float the pins of unused LVDS Banks to conserve power.
6. Unused pins in active banks (due to ADC bit depth <14) are automatically tri-stated to save power, but these can also be floated.



## Imaging Performance

### TYPICAL OPERATION CONDITIONS

Unless otherwise noted, the Imaging Performance Specifications are measured using the following conditions.

Description	Condition	Notes
Light Source	Continuous red, green and blue LED illumination	1
Temperature	Measured die temperature: 40 °C and 27 °C	
Integration Time	16.6 msec (1400d LL, register 0201h)	
Readout Mode	Dual-Scan, Global Shutter, 320 MHz PLL2	
Clamps	Column/Row Noise Correction active, Frame Black Level Clamp active	
ADC Bit Depth	10 bit	
Analog Gain	Unity gain or referred back to unity gain.	

Notes:

1. For monochrome sensor, only green LED used.

### PERFORMANCE SPECIFICATIONS ALL CONFIGURATIONS

Description	Symbol	Min.	Nom.	Max.	Units	Sampling Plan	Temperature Tested At (°C)	Test	Notes
Photodiode Charge Capacity	PNe	-	17	-	ke <sup>-</sup>	Die	27 & 40	16	
Read Noise	ne-T	-	3.4 RS/GR DS 3.7 RS/GR TS 2.5 GS DS/TS	-	e <sup>-</sup> rms	Die	27	8	
Total Pixelized Noise	TPN	-	3.6 RS/GR DS 3.9 RS/GR TS 2.5 GS DS/TS	-	e <sup>-</sup> rms	Die	27	19	
Dynamic Range	DR	-	74 RS/GR DS 73 RS/GR TS 57 GS DS/TS	-	dB	Die	27		3
Column Noise	Cn	-	0.4 RS/GR DS/TS 2.4 GS DS/TS	-	e <sup>-</sup> rms	Die	27	9	5
Row Noise	Rn	-	0.4 RS/GR DS 0.7 RS/GR TS 2.7 GS DS/TS	-	e <sup>-</sup> rms	Die	27	10	6
Dark Field Local Non-Uniformity Floor	DSNU_flr	-	1.3 RS/GR DS 1.7 RS/GR TS 10 GS DS/TS	-	e <sup>-</sup> rms	Die	27 & 40	1	4
Bright Field Local Photoresponse Non-Uniformity	PRNU_1	-	1.1 Mono 1.5 Bayer	-	%rms	Die	27 & 40	2	1
Bright Field Global Photoresponse Non-Uniformity	PRNU_2	-	3.7 Mono 5.4 Bayer	-	%pp	Die	27 & 40	3	1
Maximum Photoresponse Nonlinearity	NL	-	5.4	-	%	Die	27 & 40	11	2
Maximum Gain Difference Between Outputs	ΔG	-	0.3	-	%	Die	27 & 40	12	7
Photodiode Dark Current	I <sub>pd</sub>	-	6.6	-	e <sup>-</sup> /p/s	Die	40	13	8
Storage Node Dark Current	I <sub>vd</sub>	-	1490	-	e <sup>-</sup> /p/s	Die	40	14	4
Image Lag	Lag	-	1.6	-	e <sup>-</sup>	Design	27, 40	15	
Black-Sun Anti-Blooming	Xab	-	15 >10,000	-	W/cm <sup>2</sup> xIllumSat	Design	27	7	13
Parasitic Light Sensitivity	PLS	-	728	-	-	Design	27	6	9
Dual-Video WDR		-	140 RS 120 GS	-	dB	Design	27		10, 11
Pulsed Pixel WDR (GS only)		-	100	-	dB	Design	27		12, 11

RS = Rolling Shutter operation mode, GS = Global Shutter operation mode, GR = Global Reset

DS = Dual-Scan, TS = Tri-Scan



**KAC-06040-ABA Configuration (Monochrome)**

Description	Symbol	Wavelength (nm)	Nom.	Units	Sampling Plan	Temperature Tested At (°C)	Notes	Test
Peak Quantum Efficiency	Green NIR1 NIR2	550 850 900	52 15 9.0	%	Design	27		
Responsivity			83	$\frac{ke^-}{Lux * s}$	Design	27		20
Responsivity			7.3	$\frac{V}{Lux * s}$	Design	27		21

**KAC-06040-CBA Configuration (Bayer RGB)**

Description	Symbol	Wavelength h (nm)	Nom.	Units	Sampling Plan	Temperature Tested At (°C)	Notes	Test
Peak Quantum Efficiency	Gree n	470 540 620 850 900	42 47 44 16 9.8	%	Design	27		
Responsivity		Blue Green Red	18 36 39	$\frac{ke^-}{Lux * s}$	Design	27		20
Responsivity		Blue Green Red	1.6 3.1 3.4	$\frac{V}{Lux * s}$	Design	27		21

Notes

1. Measured per color, worst of all colors reported
2. Value is over the range of 10% to 90% of photodiode saturation, Green response used.
3. Uses 20LOG(PNe/ ne-T)
4. Photodiode dark current made negligible
5. Column Noise Correction active
6. Row Noise Correction active
7. Measured at ~70% illumination
8. Storage node dark current made negligible
9. GSE (Global Shutter Efficiency) = 1-1/PLS
10. Min vs Max integration time at 30 fps
11. WDR measures expanded exposure latitude from linear mode DR
12. Min/Max responsivity in a 30 fps image
13. Saturation Illumination referenced to a three line time integration.



## Typical Performance Curves

### QUANTUM EFFICIENCY

#### Monochrome with Microlens

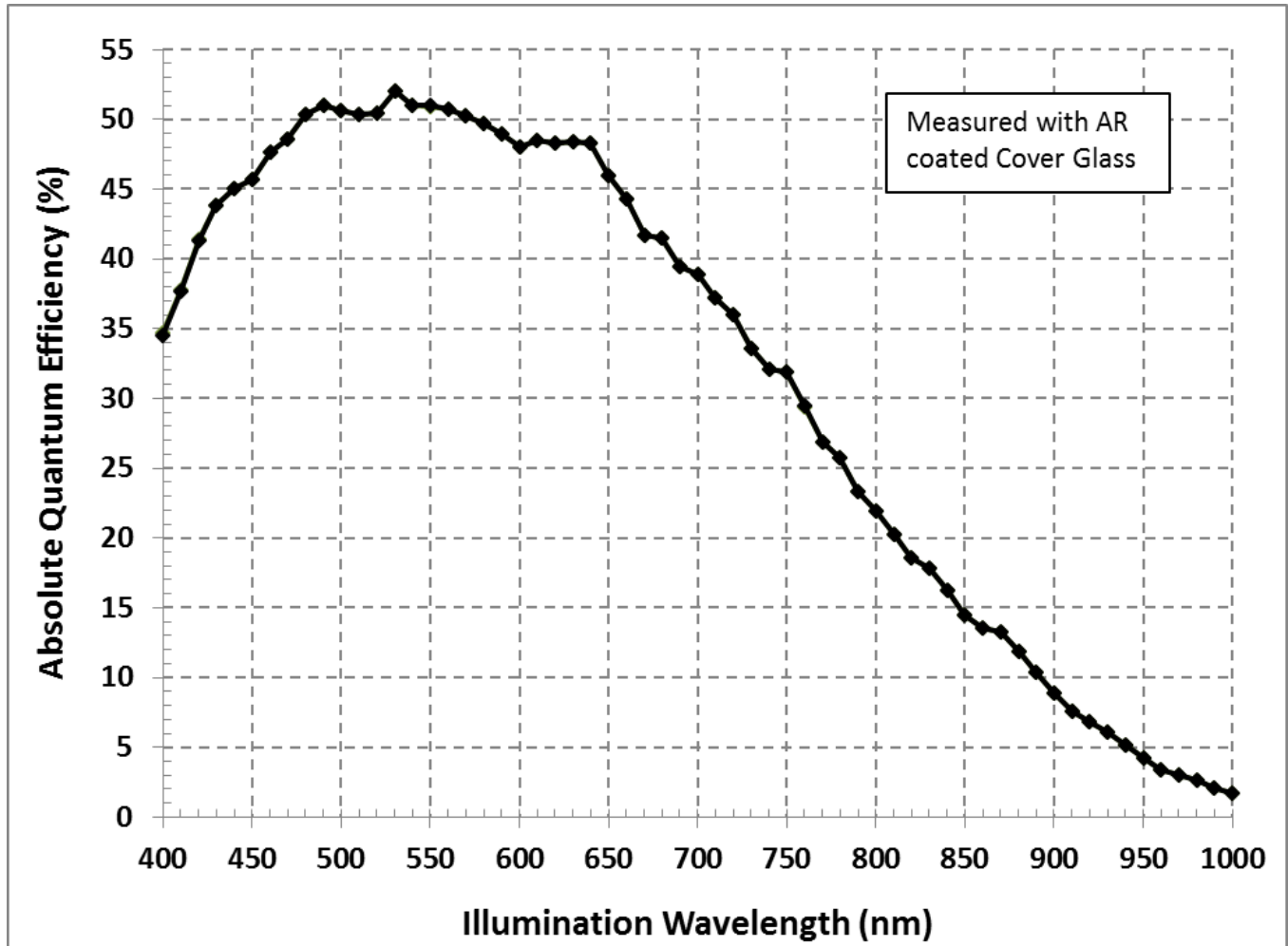


Figure 3: Monochrome QE (with Microlens)



### Color (Bayer RGB) with Microlens

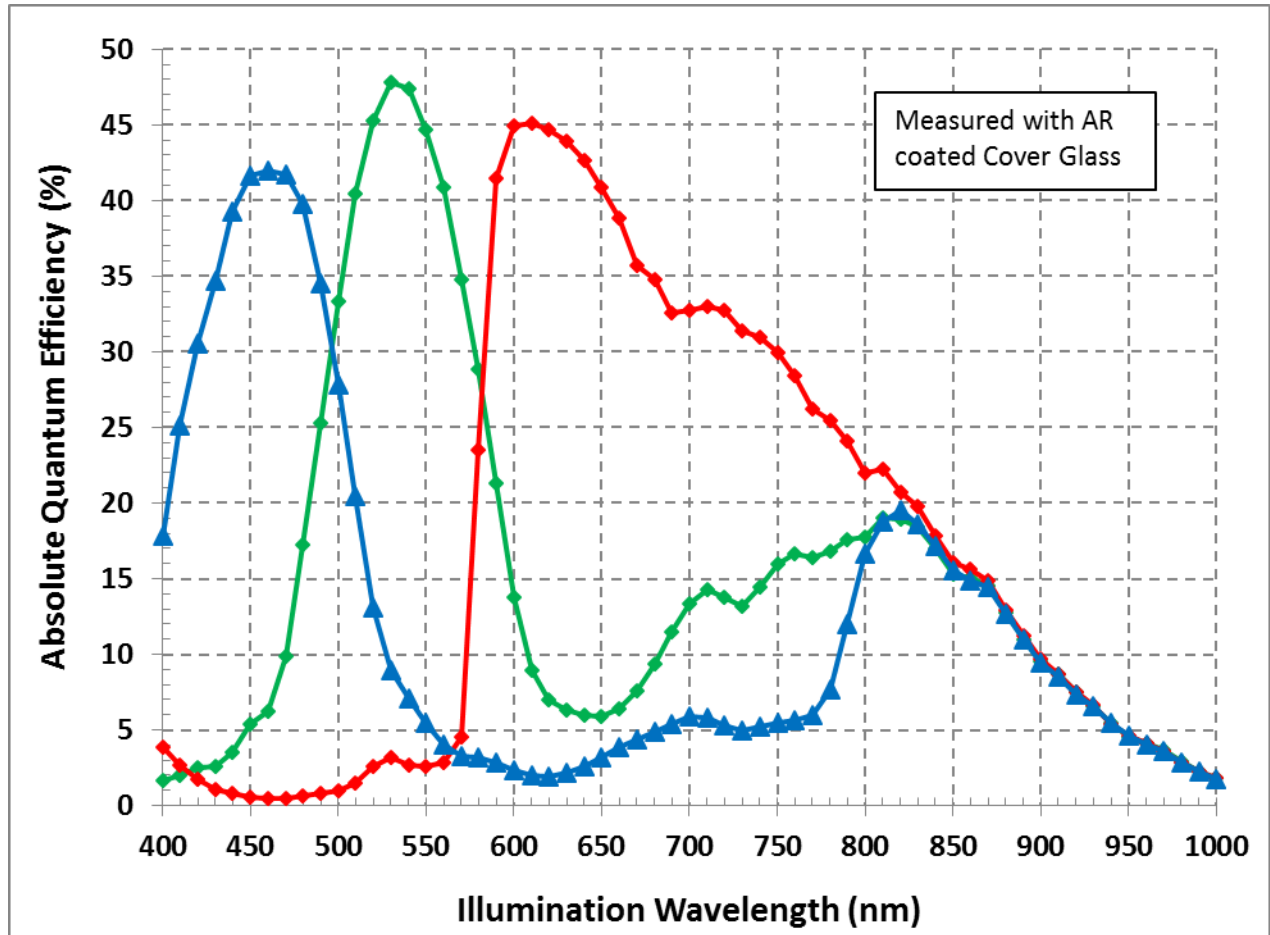


Figure 4: Bayer QE (with Microlens)



### ANGULAR QUANTUM EFFICIENCY

For the curves marked "Horizontal", the incident light angle is varied along the wider array dimension.

For the curves marked "Vertical", the incident light angle is varied along the shorter array dimension.

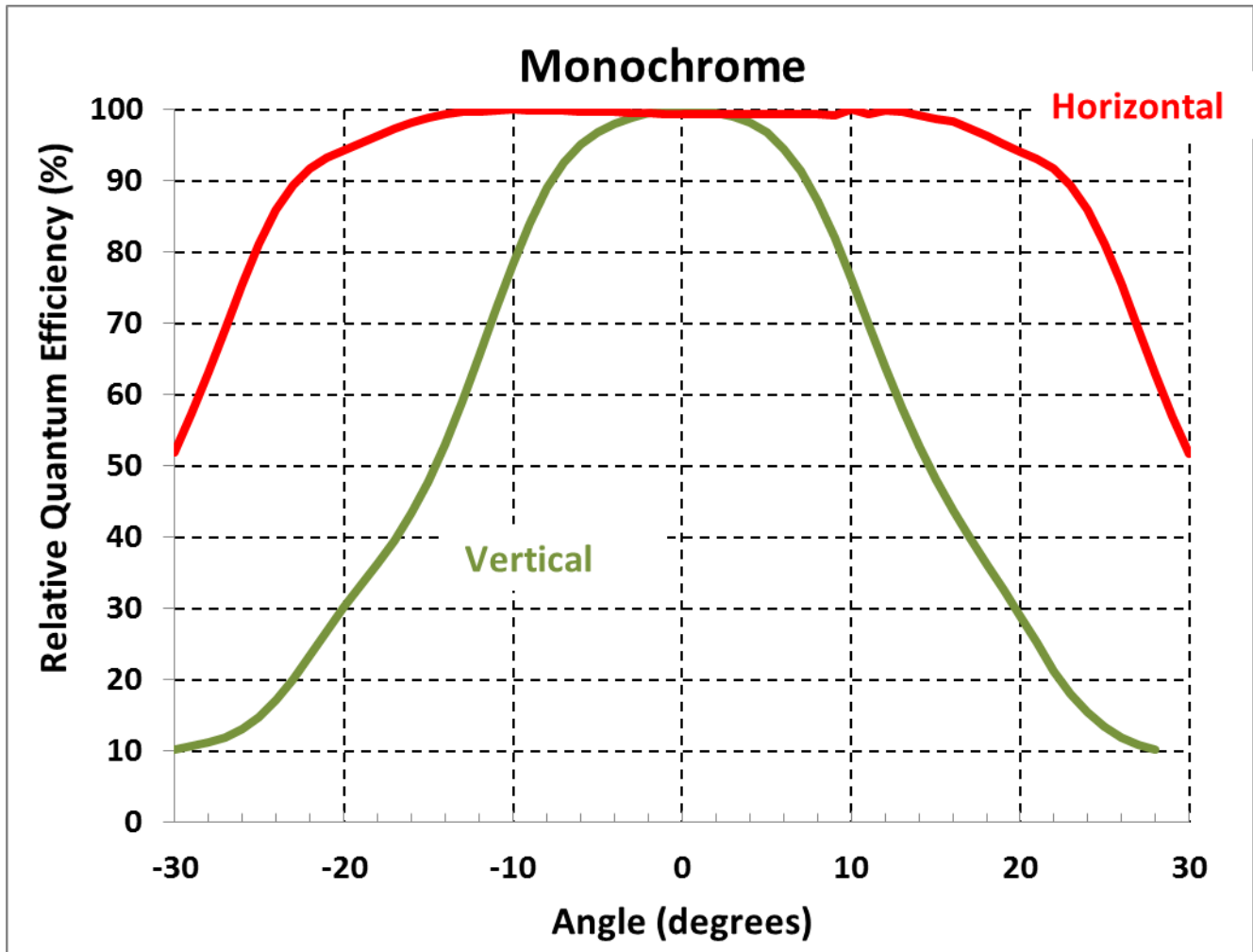


Figure 5: Monochrome Relative Angular QE (with Microlens)

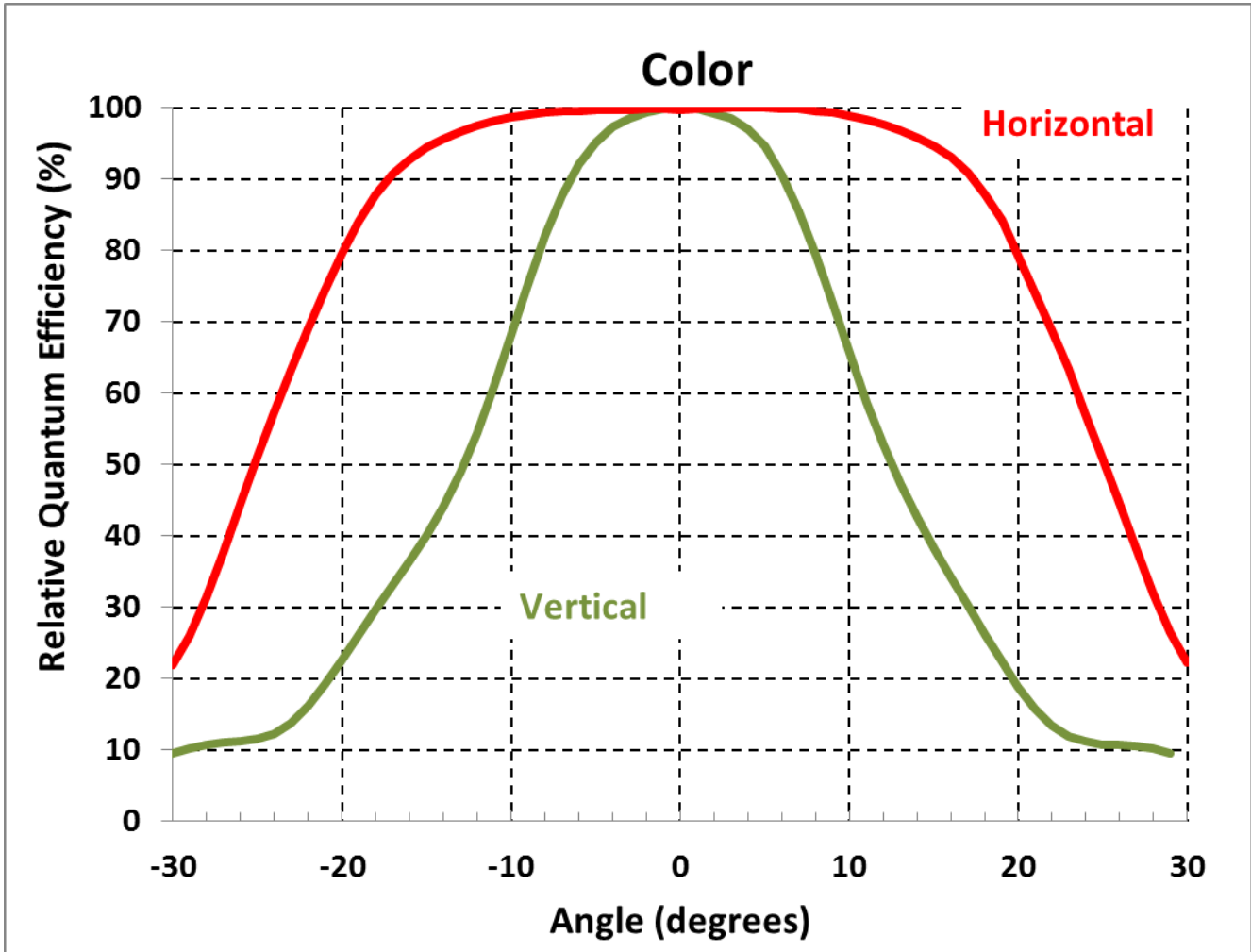


Figure 6: Monochrome Relative Angular QE (with Microlens)





### DARK CURRENT VERSUS TEMPERATURE

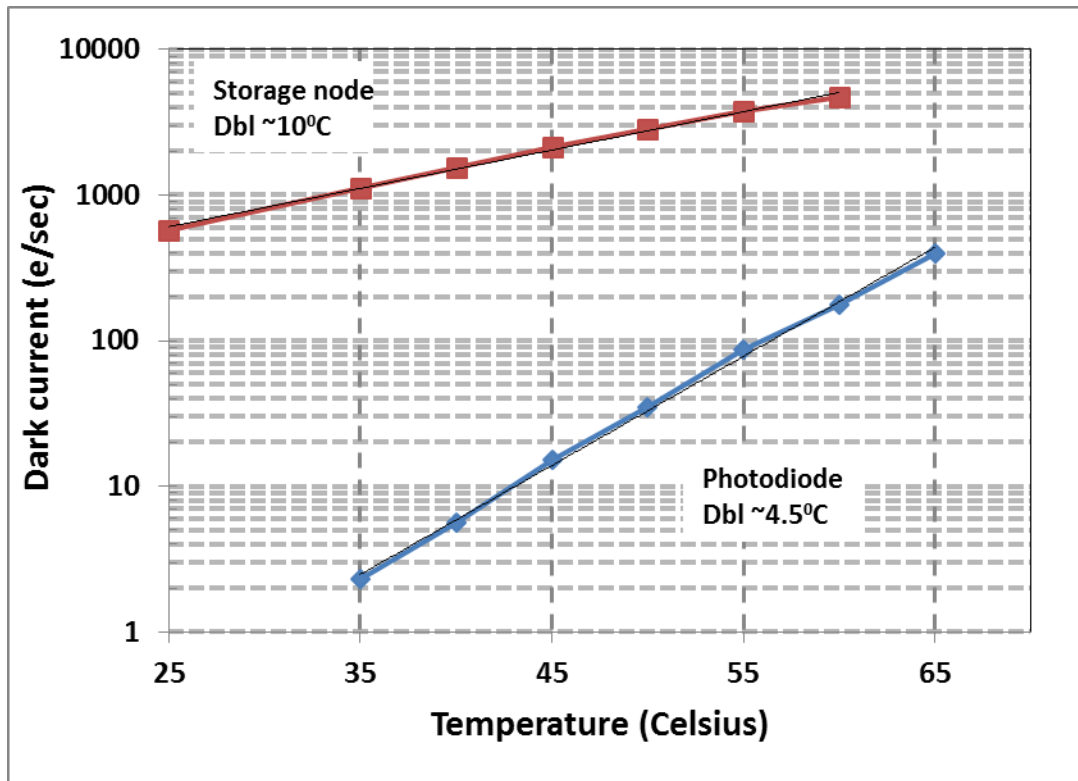


Figure 7: Dark Current vs Temperature

Note: "Dbt" denotes an approximate doubling temperature for the dark current for the displayed temperature range.



### POWER VS FRAME RATE

The most effective method to set the frame rate is to use vertical blanking (Register 01F1h). Unnecessary chip operations are suspended during vertical blanking conserving significant power consumption and also minimizing the image storage time on the storage node when in Global Shutter Operation. Tri-scan can reach higher frame rates, but consumes more power at all frame rates. It is recommended use Dual-Scan unless the frame rate required can only be reached with Tri-Scan. The LVDS clock is 1/2 the PLL2 clock frequency.

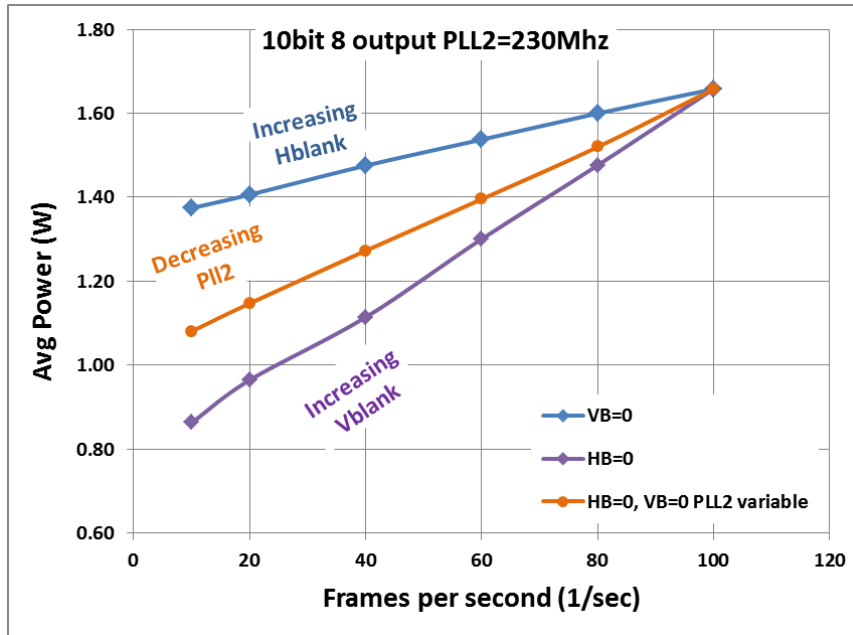


Figure 8: Dual-Scan Power vs Frame Rate, 10 bit mode

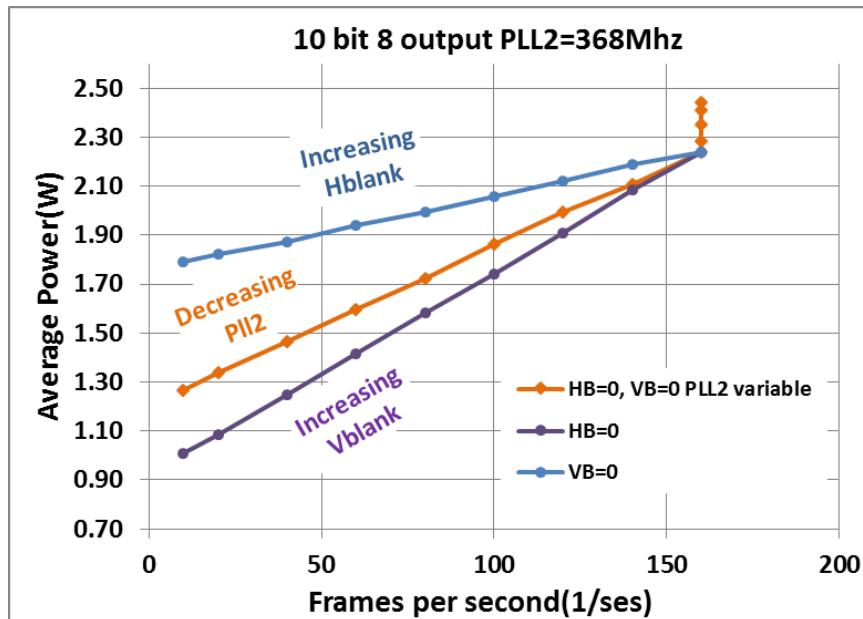


Figure 9: Tri-Scan Power vs Frame Rate, 10 bit mode



### Power and Frame Rate vs ADC Bit depth

Increasing the ADC bit depth impacts the frame rate by changing the ADC conversion time. The following figure shows the power and Frame rate range for several typical cases. For optimum image quality and power consumption the PLL2 and vertical blanking have been optimized for each bit depth and target frame rate. Because of the different parameters impacting the line time, Tri-Scan only has significant benefit at 10 bit operation. At 8 bit operation the LVDS readout time dominates the line time; and at 12 and 14 bit the ADC time dominates the line time and the pixel time is not significant. But at 10 bit operation Tri-Scan can almost halve the line time at the cost of additional power consumption.

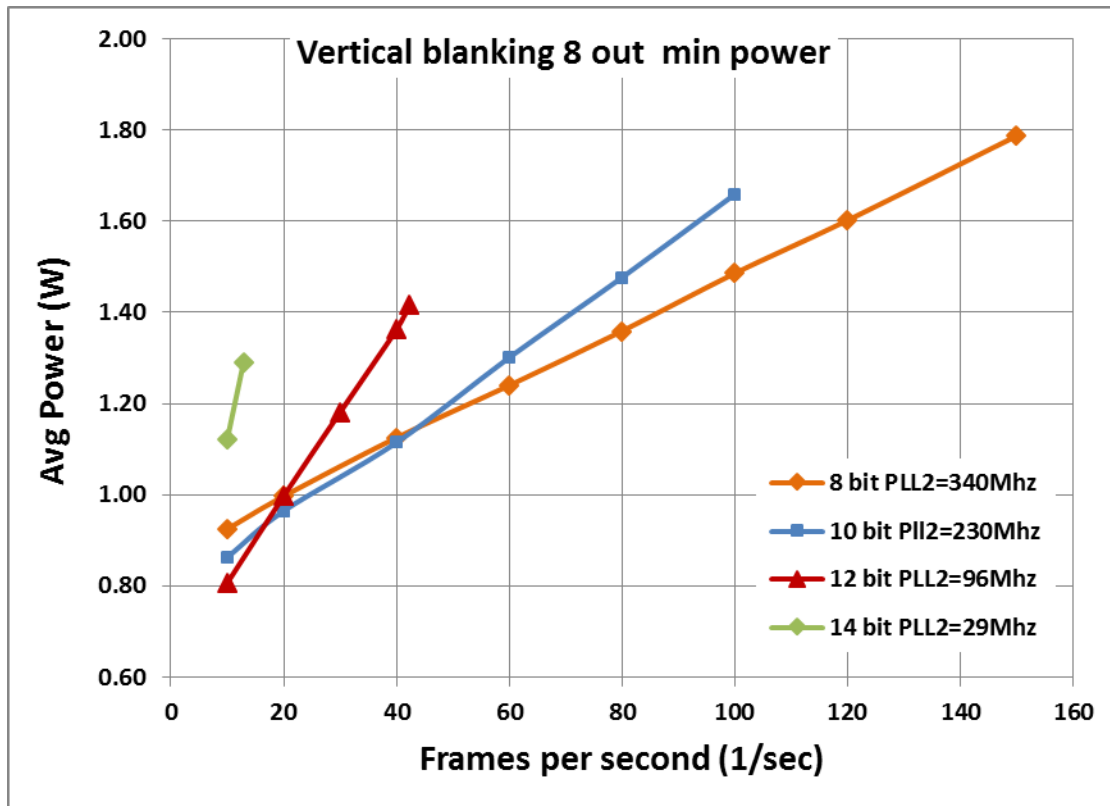


Figure 10: Dual-Scan ADC Bit Depth impact on Frame Rate and Power

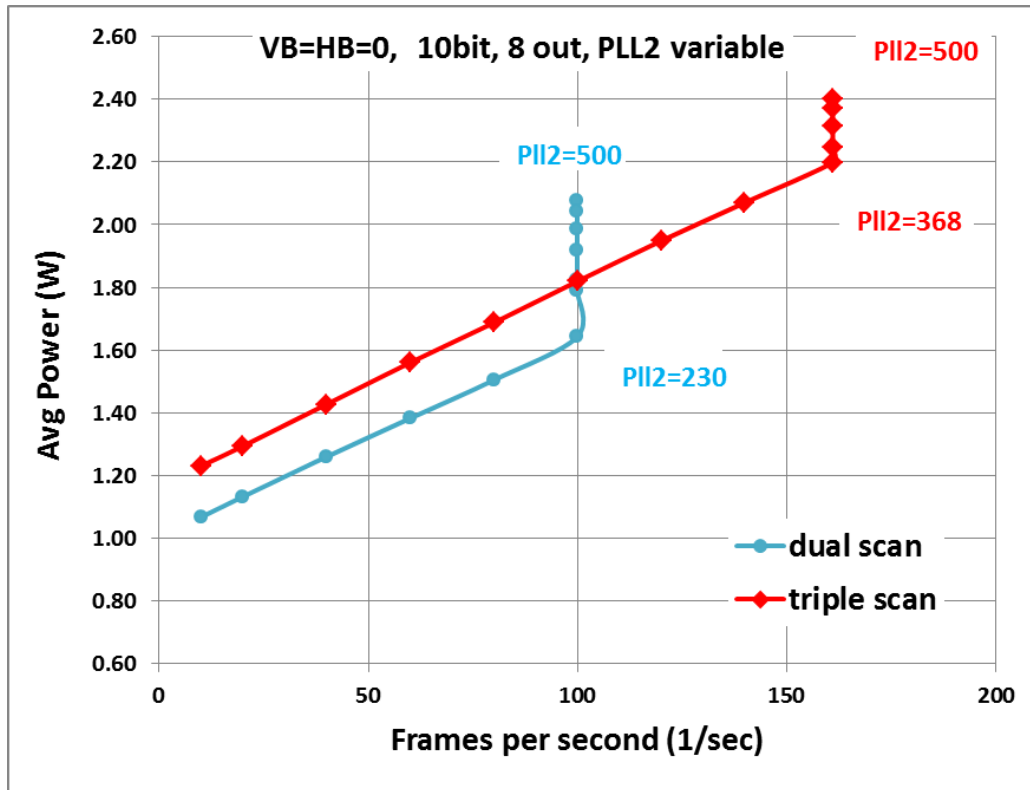


Figure 11: Tri-Scan vs Dual-Scan Power



## Defect Definitions

### OPERATION CONDITIONS FOR DEFECT TESTING

Description	Condition	Notes
Operational Mode	10 bit ADC, 8 LVDS outputs, Global Shutter and Rolling Shutter modes, Dual-Scan, Black Level Clamp on, Column/Row Noise Correction on, 1x Analog Gain, 1x Digital Gain	
Pixels Per Line	2832	
Lines Per Frame	2128	
Line Time	6.875 μsec	
Frame Time	8.25 msec	
Photodiode Integration Time	33 msec	
Storage Readout Time	7.85 msec	
Temperature	40 °C and 30 °C	
Light Source	Continuous red, green and blue LED illumination (green only for monochrome sensor)	
Operation	Nominal operating voltages and timing, PLL1 = 320 MHz, PLL2 = 410 MHz	

### DEFECT DEFINITIONS FOR TESTING

Description	Definition	Limit	Test	Notes	
Dark Field Defective Pixel	29 °C RS: Defect ≥ 20 dn GS: Defect ≥ 180 dn	40 °C RS: Defect ≥ 30 dn GS: Defect ≥ 240 dn	60	4	3, 4
Bright Field Defective Pixel	Defect ≥ ±12% from local mean			5	1, 4
Cluster Defect	A group of 2 to 10 contiguous defective pixels, but no more than 3 adjacent defects horizontally		11		2
Column/Row Major Defect	A group of more than 10 contiguous defective pixels along a single column or row		0		
Dark Field Faint Column/Row Defect	RS: 3 dn threshold GS: 10 dn threshold		0	17	
Bright Field Faint Column/Row Defect	RS: 12 dn threshold GS: 18 dn threshold		0	18	

Notes:

RS = Rolling Shutter, GS = Global Shutter

1. For the color devices, all bright defects are defined within a single color plane, each color plane is tested
2. Cluster defects are separated by no less than two good pixels in any direction.
3. Rolling Shutter Dark Field points are dominated by photodiode integration time, Global Shutter Dark Field defects are dominated by the readout time.
4. The net sum of all bright and dark field pixel defects in rolling and global shutter are combined and then compared to the test limit



## Test Definitions

### TEST REGIONS OF INTEREST

Image Area ROI: Pixel (0, 0) to Pixel (2847, 2143)

Active Area ROI: Pixel (8, 8) to Pixel (2839, 2135)

Only the Active Area ROI pixels are used for performance and defect tests.

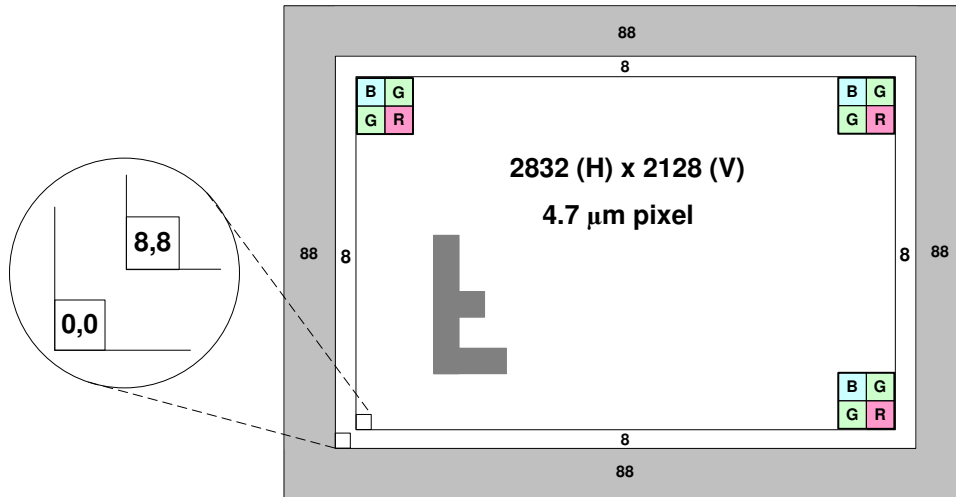


Figure 12: Regions of Interest

## TESTS

### 1) Dark Field Local Non-Uniformity Floor (DSNU\_flr)

This test is performed under dark field conditions. A 4 frame average image is collected. This image is partitioned into 180 subregions of interest, each of which is 190 by 178 pixels in size. For each sub-region the standard deviation of all its pixels is calculated. The dark field local non-uniformity is the largest standard deviation found from all the sub regions of interest. Units: e<sup>-</sup> rms (electrons rms)

### 2) Bright Field Local Photoresponse Non-Uniformity (PRNU\_1)

The sensor illuminated to 70% of saturation (~700 dn). In this condition a 4 frame average image is collected. From this 4 frame average image a 4 frame average dark image is subtracted. The Active Area Standard Deviation is the standard deviation of the resultant image and the Active Area Signal is the average of the resultant image.

$$PRNU_1 = 100 * \left( \frac{\text{Active Area Standard Deviation}}{\text{Active Area Signal}} \right) \text{ Units: \%rms}$$



### 3) Bright Field Global Non-Uniformity (PRNU\_2)

This test is performed with the sensor uniformly illuminated to 70% of saturation (~700 dn), a 4 frame average image is collected and a 4 frame averaged dark image is subtracted. The resultant image is partitioned into 180 subregions of interest, each of which is 190 by 178 pixels in size. The average signal level of each sub regions of interest (sub-ROI) is calculated.

The highest sub-ROI average (Maximum Signal) and the lowest sub-ROI average (Minimum Signal) are then used in the following formula to calculate PRNU\_2.

$$\text{PRNU}_2 = 100 * \frac{\text{Maximum Signal} - \text{Minimum Signal}}{\text{Active Area Signal}} \quad \text{Units: \%pp}$$

### 4) Dark Field Defect Test

This test is performed under dark field conditions. The sensor is partitioned into 390 subregions of interest, each of which is 128 by 128 pixels in size. In each region of interest, the median value of all pixels is found. For each region of interest, a pixel is marked defective if it is greater than or equal to the median value of that region of interest plus the defect threshold specified in the Defect Definition Table section.

### 5) Bright Field Defect Test

This test is performed with the imager illuminated to a level such that the output is at approximately 700 dn. The average signal level of all active pixels is found. The bright and dark thresholds are set as:

Dark defect threshold = Active Area Signal \* threshold

Bright defect threshold = Active Area Signal \* threshold

The sensor is then partitioned into 390 subregions of interest, each of which is 128 by 128 pixels in size. In each region of interest, the average value of all pixels is found. For each region of interest, a pixel is marked defective if it is greater than or equal to the median value of that region of interest plus the bright threshold specified or if it is less than or equal to the median value of that region of interest minus the dark threshold specified.

Example for bright field defective pixels:

- Average value of all active pixels is found to be 700 dn
- Lower defect threshold:  $700 \text{ dn} * 12 \% = 84 \text{ dn}$
- A specific 128 x 128 ROI is selected:
  - Median of this region of interest is found to be 690 dn.
  - Any pixel in this region of interest that is  $\leq (690 - 84 \text{ dn})$  in intensity will be marked defective.
  - Any pixel in this region of interest that is  $\geq (690 + 84 \text{ dn})$  in intensity will be marked defective
- All remaining 299 sub regions of interest are analyzed for defective pixels in the same manner.



## 6) Parasitic Light Sensitivity (PLS)

Parasitic Light Sensitivity is the ratio of the light sensitivity of the photodiode to the light sensitivity of the storage node in Global Shutter. There is no equivalent distortion in Rolling Shutter. A low PLS value can provide distortion of the image on the storage node by the scene during readout.

$$PLS = \frac{\text{Photodiode Responsivity}}{\text{Storage Node Responsivity}} \quad (\text{unitless ratio})$$

GSE (Global Shutter Efficiency) is a related unit.  $GSE = \left(1 - \frac{1}{PLS}\right) \%$

Detailed method: Photodiode Responsivity:

The sensor is set in global shutter serial mode (integration time not overlapping readout) and the FLO signal is used to control a 550 nm normal incident (or large f# focused) illumination source so that the sensor is illuminated only during photodiode integration time (not illuminated during readout time). The integration time is not critical but should be large enough to create a measurable mean during this time. A 16 frame-average illuminated photodiode image is recorded. A 16 frame-average dark frame using the same sensor settings is captured and is subtracted from the illuminated image.

Detailed method: Storage Node Responsivity:

The sensor is set to a special characterization mode where the PD signal is discarded and does not impact the storage node. A long total frame time (storage node exposure time) is used to increase the storage node signal. A 16 frame-average dark frame is captured. The sensor is illuminated by the same 550nm incident light source used for the photodiode responsivity. A 16 frame-average illuminated photodiode image is recorded; the dark frame image is subtracted from this. The integration time is not critical but should be set such that a significant response is detected, typically several orders of magnitude greater than the photodiode integration time.

## 7) Black-Sun Anti-blooming

A typical CMOS image sensor has a light response profile that goes from 0 dn to saturation (1023 dn for KAC-06040 in 10 bit ADC mode) and, with enough light, back to 0 dn. The sensor reaching 0 dn at very bright illumination is often called the "Black-sun" artifact and is undesirable. Black-sun artifact is typically the dominant form of anti-blooming image distortion. For the KAC-06040 the Black-sun artifact threshold is measured at the onset of saturation distortion, not at the point where the output goes to 0 dn. To first order the onset of black-sun artifact for the KAC-06040 is not proportional to the integration time or readout time.

The sensor is placed in the dark at unity gain and illuminated with a 532 nm laser with the intensity of about 26 W/cm<sup>2</sup> at the center of the sensor. The laser is strong enough to make the center of the laser spot below 1020 dn without any ND filters. ND filters are added to adjust the laser intensity until the signal in the region at the center of the spot increases to >1020 dn.

This illumination intensity at this ND filter is recorded (W/cm<sup>2</sup>) as the Black-Sun Anti-blooming.

The 'xllumSat' unit is calculated using and integration time of 100 μsec.

Exposing the sensor to very strong illumination for extended periods of time will permanently alter the sensor performance in that localized region.





## 8) Read Noise

This test is performed with no illumination and one line of integration time. The read noise is defined as one standard deviation of the frequency histogram containing the values of all pixels after the excessively deviant pixels ( $\pm$  three standard deviations) are removed.

## 9) Column Noise

After all rows are averaged together. Shading (low frequency change wrt column address) is removed. A frequency histogram is constructed of the resulting column values. The column noise is the standard deviation of the frequency histogram of the column values. This Metric includes both temporal and FPN.

## 10) Row Noise

All columns are averaged together. Shading (low frequency change wrt row address) is removed. A frequency histogram is constructed of the resulting row values. The row noise is the standard deviation of the frequency histogram of the row values. This Metric includes both temporal and FPN.

## 11) Maximum Photoresponse Non-Linearity

The photoresponse nonlinearity is defined as the deviation from the best fit of the sensor response using 70% of saturation and zero signal as the reference points. The different signal levels are determined by varying the integration time. The sensor saturation level is (1023-dark offset). The dark offset is subtracted from the image for the following  $M_{avg}$  and  $L_{avg}$ .

- The integration time is varied until the integration time required to reach the 70% saturation is determined.  $M_{avg}$  = the active array mean at the 70% saturation integration time.
- The integration is set to 1/14 (5% exposure point).  $L_{avg}$  = meant at the 5% exposure point
- $PRNL$  (@ 5% saturation) =  $((L_{avg}/M_{avg}) * (14/1) - 1) * 100$

## 12) Maximum Gain Difference Between Outputs

The LVDS outputs contain no gain or offset error since these are purely digital segmentations. The predominant output mismatch comes from the pixel array readout segmentation. The sensor contains two ADC banks and four channels of analog line stores in its highest frame rate configuration, Tri-Scan. The sensor is factory calibrated to match the gain differences between all four possible gain channels. The gain variations are manifest as an every 4<sup>th</sup> row gain pattern. In tri-scan, and an even/odd row gain difference in Dual-Scan. The sensor is factory calibrated to match the four possible row gains. This test is performed in Tri-Scan mode to test the worst case gain error including all possible 4 row gains after the calibration has been applied. The sensor is illuminated at 70% of saturation. The entire test frame ROI into 4 groups of every 4<sup>th</sup> row. The first row group(average) is used as a reference and the following three row groups are compared to the first. The largest error is reported.

$$((\text{Second Row Average}/\text{First Row Average}) - 1) * 100$$

$$((\text{Third Row Average}/\text{First Row Average}) - 1) * 100$$

$$((\text{Fourth Row Average}/\text{First Row Average}) - 1) * 100$$



### 13) Photodiode Dark Current

The photodiode dark current is measured in rolling shutter read out mode using 105 msec integration time and an analog gain = 8. The value is converted to electrons/pix/sec using the formula:

$$\text{Photodiode dark current} = \text{average signal (DN)} * \text{el-per-DN (gain=8)} / .105 \text{ seconds}$$

where 'average signal (DN)' is the average of all pixels in the sensor array, and 'el-per-DN(gain=8)' is measured on each sensor using the photon transfer method.

### 14) Storage Node Dark Current

The storage node dark current is measured in global shutter read out mode using a special timing mode to prevent the photodiode dark current from being transferred to the storage node. In global shutter mode, the integration time of the storage node is the time it takes to read out a frame. The sensor analog gain is set to 2:

$$\text{Storage node dark current} = \text{average signal (DN)} * \text{el-per-DN (gain=2)} / .0138 \text{ seconds}$$

where 'average signal (DN)' is the average of all pixels in the sensor array and 'el-per-DN(gain=2)' is measured on each sensor using the photon transfer method.

### 15) Lag

Lag is measured as the number of electrons left in the photodiode after readout when the sensor is illuminated at 70% of Photodiode Charge Capacity.

Analog gain is set to 8. With no illumination a 64 average dark image is recorded (Dark\_ref). The 'el-per-DN' is measured using the photon transfer method.

Illumination is adjusted blink every other frame such that the mean image output is 70% of the Photodiode Charge Capacity for even frames, and with no illumination for odd frames. A 64 frame average of Odd Dark Frames is recorded as Dark\_Lag.

$$\text{Lag} = (\text{Dark\_Lag} - \text{Dark\_Ref}) * \text{el-per-DN}' \text{ Units: electrons rms}$$

### 16) Photodiode Charge Capacity

The sensor analog gain is reduced to <1 to prevent ADC clipping at 1023 dn. The 'el-per-DN' is measured using the photon transfer method. The sensor is illuminated at a light level ~1.5x the illumination at which the pixel output no longer linearly changes with illumination level. The Photodiode Charge Capacity is equal to the average signal (DN) \* el-per-DN. Units: electrons rms.

### 17) Dark Field Faint Column/Row Defect

A 4 frame average, no illumination image is acquired at one line time of integration. Major defective pixels are removed (> 5 Sigma). All columns or rows are averaged together. The average of the local ROI of 128 columns or rows about the column/row being tested is determined. Any columns/rows greater than the local average by more than the threshold are identified.



### **18) Bright Field Faint Column/Row Defect**

A 4 frame average, 70% illumination image is acquired at one line time of integration. Major defective pixels are removed ( $> 5$  Sigma). All columns or rows are averaged together. The average of the local ROI of 128 columns or rows about the column/row being tested is determined. Any columns/rows greater than the local average by more than the threshold are identified.

### **19) Total Pixelized Noise**

This test is performed with no illumination and one line of integration time. A single image is captured including both Temporal and Fixed Pattern Noise (FPN). A spatial low pass filter is applied to remove shading and excessively deviant pixels ( $\pm$  three standard deviations) are removed. The Total Pixelized Noise is defined as one standard deviation of the frequency histogram.

### **20) Responsivity ke/lux-sec**

This number is calculated by integrating the multiplication of the sensor QE by the human photopic response assuming a 3200K light source with a QT100 IR filter. This is a sharp 650nm cutoff filter. If the IR filter is removed a higher response value will result.

### **21) Responsivity V/lux-sec**

Voltage levels are not output from the sensor. This value uses the pixel output before analog gain to match the ADC input range. Including the ADC matching gain will result in a larger responsivity value.



## Operation

This section is a brief discussion of the most common features and functions assuming default conditions. See the *KAC-06040 User Guide* for a full explanation of the sensor operation modes, options, and registers.

## REGISTER ADDRESSES

The last bit of any register address is a Read/Write bit. Most references in this document refer to the Write address. All SPI reads are to an even address, all SPI writes are to an odd address.

## SENSOR STATES

Figure 13 shows the sensor states, see the *KAC-06040 User Guide* for detailed explanation of the States.

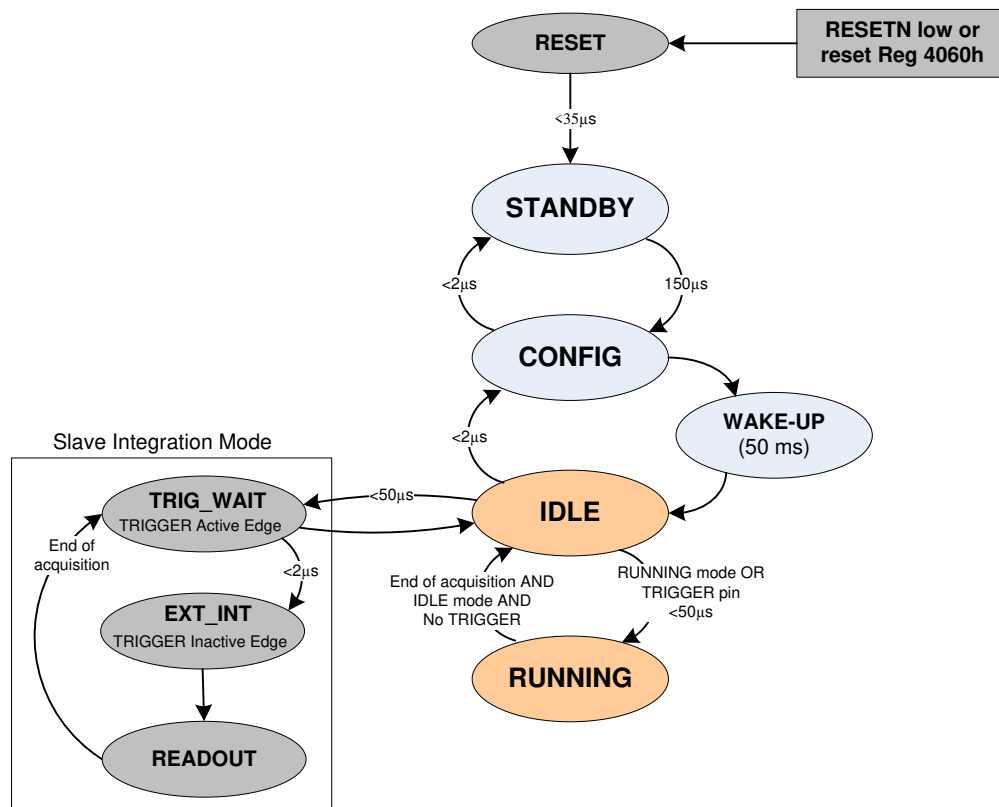


Figure 13: Sensor State Diagram



### ENCODED SYNCs

To facilitate system acquisition synchronization the KAC-06040 places synchronization words (SW) at the beginning and at the end of each output row as indicated in the following Figure 14. This is performed for each of the 8 LVDS output banks providing frame, line, and output synchronization. See the *KAC-06040 User Guide* for additional detail on LVDS and Encoded Sync output.

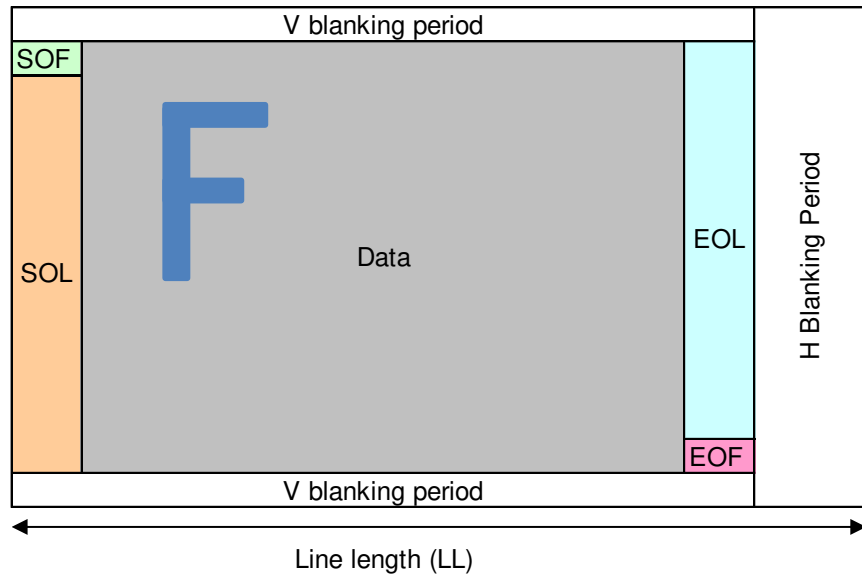


Figure 14: Encoded Frame Syncs



### LINE TIME

This Datasheet presumes the recommended startup script that is defined in the *KAC-06040 User Guide* has been applied. The KAC-06040 defaults to Dual-Scan mode. In this mode the LVDS data readout overlaps the pixel readout and ADC conversion time. The Pixel read time is fixed, and the ADC Conversion Time is dependent on the ADC bit depth selected. The LVDS time will be dependent on the PLL2 frequency selected. Depending on the ADC bit depth and the PLL2 frequency the LVDS readout or the (Pixel + ADC conversion) may limit the minimum possible line time. The Line Time is not impacted by the selection of Rolling Shutter or Global Shutter mode.

Tri-scan mode can be used in for shorter line times and faster frame rates (at elevated power consumption). Tri-scan is of most value when the Pixel time and ADC conversion time and LVDS readout time are similar in size. For full resolution this corresponds to 8 LVDS bank and 10 bit ADC bit depth. In Tri-Scan mode the longest of the three components will define the minimum line time.

The KAC-06040 architecture always outputs two rows at once, one row from the top ADC, and one from the bottom ADC. Each ADC then divides up the pixel into 1 → 4 parallel pixel output LVDS Banks. The default is 4 output banks per ADC for a total of 8 parallel pixel outputs to minimize the LVDS data output time. Since the sensor always outputs 2 rows at a time the timing and registers are based on a Line Time (LT) or Line Length (LL) where one LT = the time to readout 2 rows in parallel (one even row and one odd row).

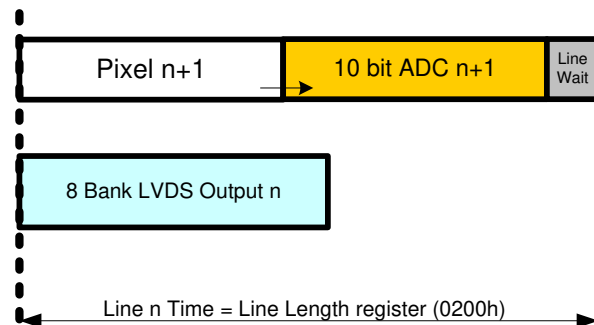


Figure 15: Dual-Scan Line Time Relationship

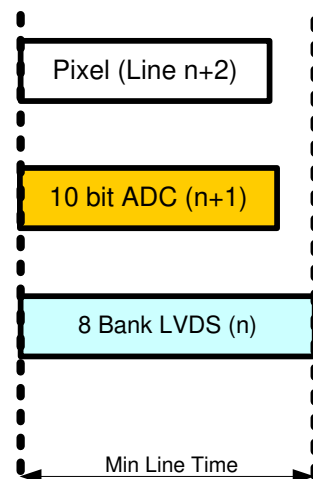


Figure 16: Tri-Scan Line Time Relationship



## FRAME TIME

The frame time is defined in units of Line Time. 1 Line Time unit = 2 output rows. To first-order the frame rate is not directly impacted by selection of Global Shutter, Rolling Shutter, Dual-Scan, or Tri-Scan.

The Frame Time is made up of three phases:

1. Integration Phase
2. Readout Phase
3. Frame Wait Phase (Vertical Blanking, Vblank)

By default the Integration Phase overlaps the Readout and Frame Wait Phases. If the Integration Phase is larger than the Readout + Frame Wait time, then the Integration Phase will determine the video frame rate. Otherwise the frame rate will be set by the Readout + Frame Wait time. In other words, if the programmed integration time is larger than the minimum readout time (and vertical blanking) then extra vertical blanking will be added and the frame rate will slow to accommodate the requested integration time.

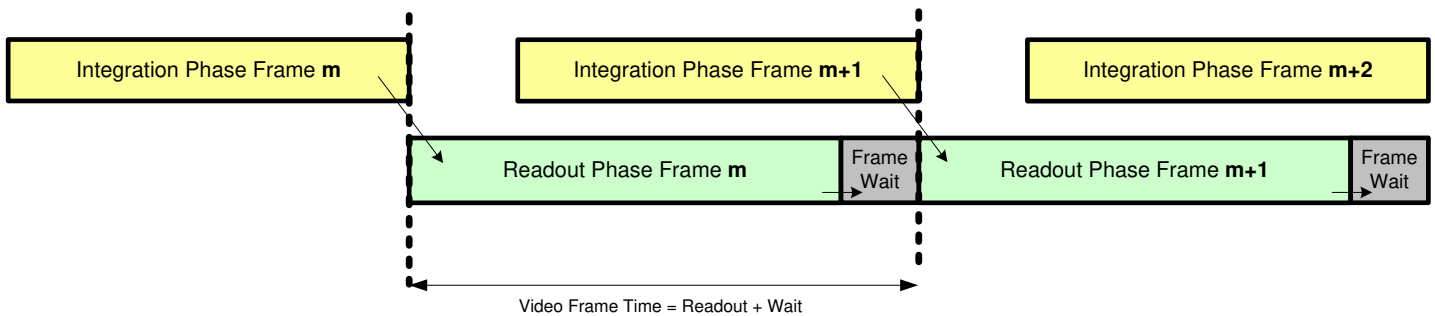


Figure 17: Default frame time configuration (Frame A)

If the Integration Phase is less than the Readout Phase then the start of integration is automatically delayed to minimize the storage time and dark current.

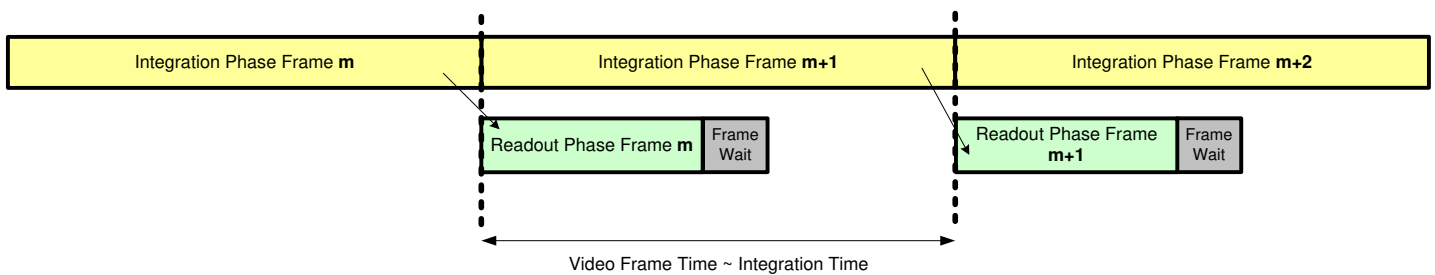


Figure 18: Frame time with extended integration time.

If the Readout Phase (+ Vblanking) is less than the Integration Phase, then the readout occurs as soon the integration is complete to minimize the storage time and dark current.

See the *KAC-06040 User Guide* for detailed calculation of the Integration Phase, Readout Phase, and Frame Wait.

To first-order the Readout Phase is equal to the number of rows \* row\_time.



### GLOBAL SHUTTER READOUT

Global Shutter readout provides the maximum precision for freezing scene motion. Any motion artifacts will be 100% defined by an ideal integration time edge. Every pixel in the array starts and stops integration at the same time.

Figure 19 illustrates a Global Shutter Frame readout assuming the recommended Start-up Script defined in the *KAC-06040 User Guide* (8 LVDS banks, Dual-Scan, 8.75 μSec line time). The Frame Wait Phase is not shown due to its small default size (1 LL) and for clarity.

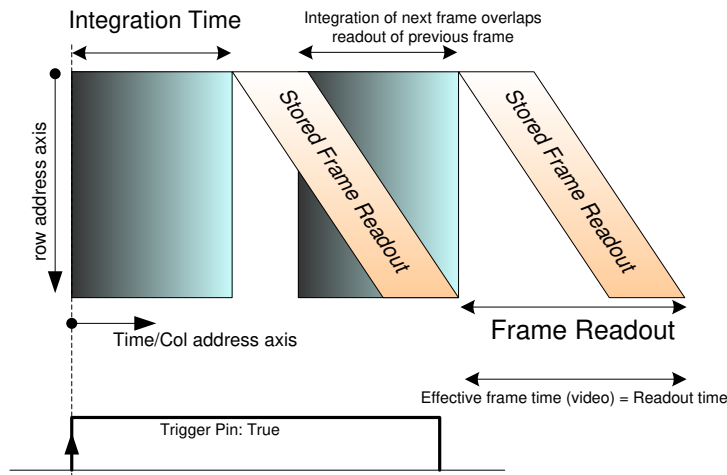


Figure 19: Illustration of frame time for Global Shutter readout

Global Shutter readout mode is selected using Bits [1:0] of Register 01D1h.

Images can be initiated by setting and holding the TRIGGER input pin or by placing the sensor into RUNNING mode by writing 03d to register 4019h. If the TRIGGER input pin is true when at the start of the integration time for the next frame then the sensor will complete an additional frame integration and readout. In the case shown in Figure 19 two frames will be output.





### ROLLING SHUTTER READOUT

The KAC-06040 high speed Rolling Shutter readout provides the maximum dynamic range while still providing excellent motion capture. In Rolling Shutter the readout more closely matches a film camera shutter. Each row of the image receives the same integration time, but each row starts and ends at a different time as the shutter travels from the top of the array to the bottom. In the Figure 20 frame time illustration this 'moving shutter' displays as a sloped edge for the blue pixel array region, just as the readout edge is sloped.

The Figure 20 illustration shows a 2 frame output sequence using the external TRIGGER pin.

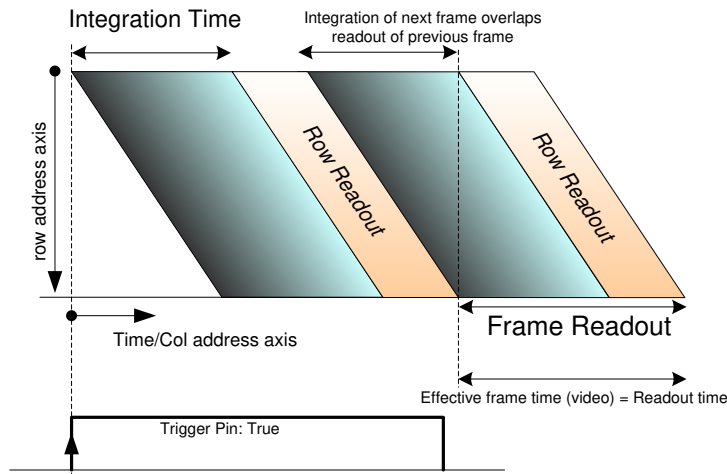


Figure 20: Illustration of Frame time for Rolling Shutter readout

Rolling Readout mode can be selected using Bits [1:0] of Register 01D1h.

Images can be initiated by setting and holding the TRIGGER input pin or by placing the sensor into RUNNING mode by writing 03d to register 4019h. If the TRIGGER input pin is True when at the start of the integration time for the next frame then the sensor will complete an additional frame integration and readout.



## 8 Bank LVDS Data Readout

### LVDS BANKS

The KAC-06040 provides 8 parallel pixel banks, each consisting of 8 LVDS differential pairs (7 data pairs + 1 clock pair). This allows the output of 8 pixels per LVDS clock period. All 7 data pairs, of each bank, are used only in 14 bit operation mode. By default only 5 data pairs are used for 10 bit mode (D4 → D0). The unused pairs are held in low-power high impedance mode.

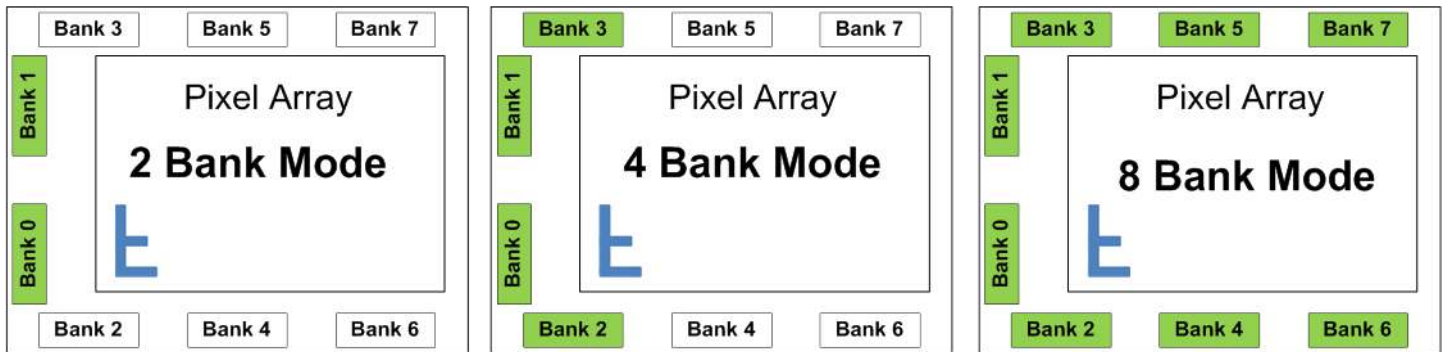


Figure 21: LVDS Bank labeling

The number of output banks used is independent of the ADC bit depth chosen. By default the KAC-06040 uses all 8 output banks for maximum frame rate. If technical restrictions prevent the use of 8 LVDS banks, the sensor can be programmed to use 4 or 2 banks, however this can result in reduced frame rate and reduction of image quality. It is recommended that 8 banks be used when possible. Only the 8 bank option is discussed in detail in this specification, see the *KAC-06040 User Guide* for additional detail on 4 and 2 bank mode.

In order to minimize the LVDS clock rate (and power) for a given data rate the pixels are output in DDR (Double Data Rate) where the MSB is always sent first (on rising edge) and the LSB second (falling edge) This is not programmable.

### PORTS PER LVDS BANK

The MSB comes out first on the falling edge, followed by the LSB on the net rising edge.

Bit Depth	Edge of DATA_CLK	Data0	Data1	Data2	Data3	Data4	Data5	Data6
14 bits	Falling (MSB nibble)	D7	D8	D9	D10	D11	D12	D13
	Rising (LSB nibble)	D0	D1	D2	D3	D4	D5	D6
12 bits	Falling (MSB nibble)	D6	D7	D8	D9	D10	D11	HiZ
	Rising (LSB nibble)	D0	D1	D2	D3	D4	D5	HiZ
10 bits	Falling (MSB nibble)	D5	D6	D7	D8	D9	HiZ	HiZ
	Rising (LSB nibble)	D0	D1	D2	D3	D4	HiZ	HiZ
8 bits	Falling (MSB nibble)	D4	D5	D6	D7	HiZ	HiZ	HiZ
	Rising (LSB nibble)	D0	D1	D2	D3	HiZ	HiZ	HiZ

Figure 22: Number of LVDS pairs (ports) used vs. bit depth



### 8 BANK PIXEL ORDER

The KAC-06040 always processes two rows at a time. Even row decodes are sent to the bottom ADC and LVDS output banks (0, 2, 4, 6). Odd rows are sent to the top ADC and LVDS banks (1, 3, 5, 7). The ROI must be (and is internally forced to) an even size and always starting on an even row decode.

The rows are read out progressively left to right (small column address to large). Eight pixels are sent out of the chip at once, one pixel per LVDS bank per LVDS clock cycle.

Pixel Readout order:

1. Two rows are selected, the even row is sent to the bottom ADC and the odd row to the top ADC.
2. Each ADC converts its row of pixel data at once and stores the result in a line buffer.
3. At default settings there are 4 output LVDS banks for each ADC.
4. Each LVDS Bank outputs one pixel per clock cycle, so 4 pixels of each row are output each full LVDS clock cycle, two rows in parallel for 8 pixels per clock cycle total.
5. The pixels are sent out from left to right (low column number to high column number). So the first 4 pixels are sent out on clock cycle 1, and the next 4 pixels to the right are sent out on clock cycle 2.
6. To conserve the number of wires per port, the 10 bits per pixel are sent out DDR (Dual Data Rate) over 5 ports. On the falling edge the upper 5 MSB bits are sent out, and on the rising edge the lower 5 bits LSB are sent out. Completing one full LVDS clock cycle and one set of eight pixels.

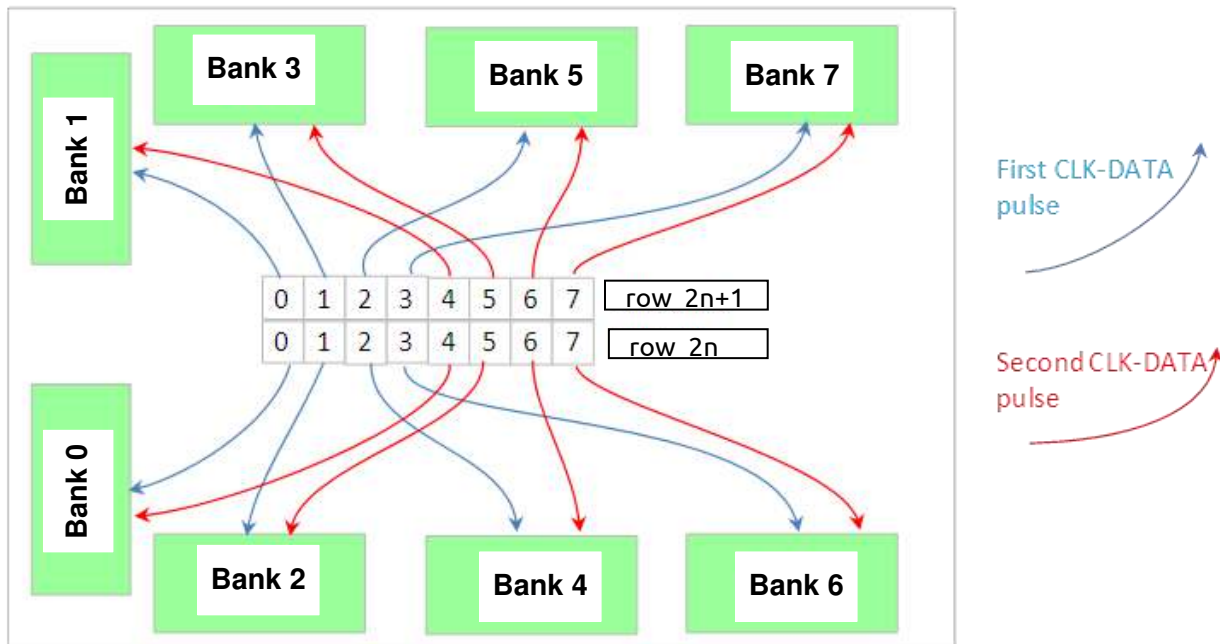


Figure 23: Pixel readout order diagram



LVDS Bank	Row	Pixel number				
Bank 0	2 n (even)	0	4	8	12	16
Bank 2	2 n (even)	1	5	9	13	17
Bank 4	2 n (even)	2	6	10	14	18
Bank 6	2 n (even)	3	7	11	15	19
Bank 1	2 n+1 (odd)	0	4	8	12	16
Bank 3	2 n+1 (odd)	1	5	9	13	17
Bank 5	2 n+1 (odd)	2	6	10	14	18
Bank 7	2 n+1 (odd)	3	7	11	15	19
LVDS Clock Cycle		1	2	3	4	5

Figure 24: Pixel readout order table

### DE-SERIALIZER SETTINGS

Figure 25 shows the data stream of one LVDS bank for 10 bit resolution.

Data serialization is fixed at 2 cycle DDR for all bit depths. Data output order is MSB first on the falling edge, and LSB following on the rising edge.

Four pixel values per synchronization word are embedded into the video stream per LVDS bank.

The SOL/SOF synchronization words are sent out of each LVDS bank before the first valid pixel data from that bank. Each bank outputs all 4 syncs of the SOF or SOL.

And each of the active LVDS banks each output all 4 sync codes for the EOL/EOF.

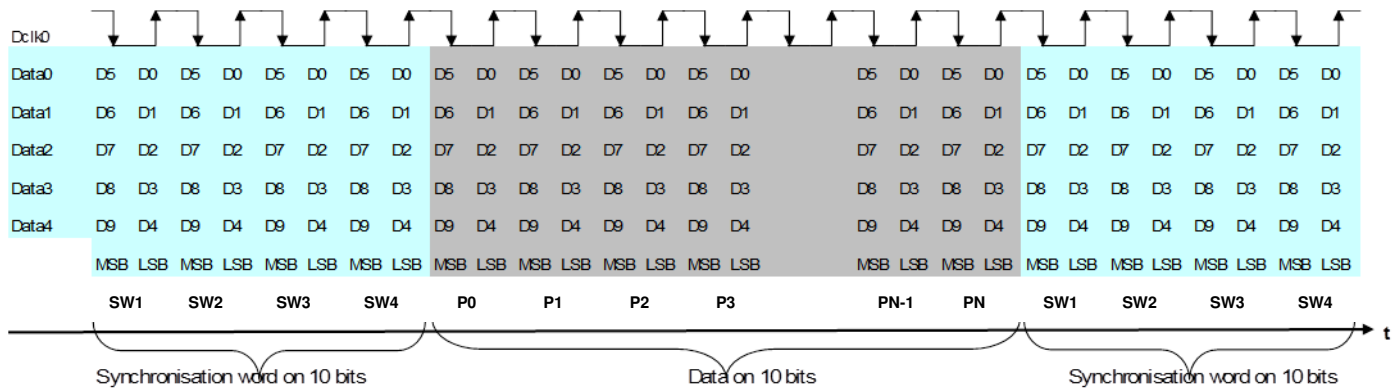


Figure 25: Data stream of one LVDS Bank for 10bits ADC resolution



## Register Definition

16 bit addr (Hex)	Reset Value Hex/Dec	SPI State	Register Name	
0001	0d	Any	Frame A ROI y1	
0009	2144d		Frame A ROI h1	
0011	0d		Frame A ROI x1	
0019	2848d		Frame A ROI w1	
0021	0d		Frame A sub-ROI y2	
0029	0d		Frame A sub-ROI h2	
0031	0d		Frame A sub-ROI x2	
0039	0d		Frame A sub-ROI w2	
0041	0d		Frame A sub-ROI y3	
0049	0d		Frame A sub-ROI h3	
0051	0d		Frame A sub-ROI x3	
0059	0d		Frame A sub-ROI w3	
0061	0d		Frame A sub-ROI y4	
0069	0d		Frame A sub-ROI h4	
0071	0d		Frame A sub-ROI x4	
0079	0d		Frame A sub-ROI w4	
0081	0011h		Frame A Decimation	
0089	0d		Frame A Video Blanking	
0091	3430d		Frame A Integration Rows	
0099	0d		Frame A Integration Sub-Row	
00A1	10d		Frame A Black Level	
00A9	001Fh		Frame A Gain	
00E9	344d		Any	Frame B ROI y1
00F1	1456d			Frame B ROI h1
00F9	136d			Frame B ROI x1
0101	2576d			Frame B ROI w1
0109	0d			Frame B sub-ROI y2
0111	0d	Frame B sub-ROI h2		
0119	0d	Frame B sub-ROI x2		
0121	0d	Frame B sub-ROI w2		
0129	0d	Frame B sub-ROI y3		
0131	0d	Frame B sub-ROI h3		
0139	0d	Frame B sub-ROI x3		
0141	0d	Frame B sub-ROI w3		
0149	0d	Frame B sub-ROI y4		
0151	0d	Frame B sub-ROI h4		
0159	0d	Frame B sub-ROI x4		
0161	0d	Frame B sub-ROI w4		
0169	0011h	Frame B Decimation		
0171	0d	Frame B Video Blanking		
0179	3430d	Frame B Integration Rows		
0181	0d	Frame B Integration Sub-Row		
0189	10d	Frame B Black Level		
0191	001Fh	Frame B Gain		

16 bit addr (Hex)	Reset Value Hex/Dec	SPI State	Description
01D1	FC10h	CONFIG only	Config1
01D9	0500h	CONFIG or IDLE	Config2
01E1	00AAh	CONFIG or IDLE	Analog/Digital Power Mode
01E9	0000h	CONFIG or IDLE	Dual-Video Repetition
01F1	0d	CONFIG or IDLE	Vertical Blanking
01F9	3431d	CONFIG or IDLE	Fixed Frame Period
0201	1400d	CONFIG or IDLE	Line Length (LL)
0209	002Dh	CONFIG or IDLE	ADC Bit Depth
0211	0000h	CONFIG or IDLE	FLO Edge Delay
0219	0000h	CONFIG or IDLE	MSO Edge Delay
0708	0000h	Any	Sensor Type FB
0710	0000h	Any	Temperature Sensor FB
0718	0000h	Any	General Feedback
0720	0000h	Any	Minimum LL FB
0730	0000h	Any	User OTP1 FB
0738	0000h	Any	User OTP2 FB
2059	0300h	CONFIG only	Output Bank Select 1
2099	2877h	CONFIG only	PLL1 Setting
20A1	0861h	CONFIG only	PLL2 Setting
2449	1C32h	CONFIG only	Sub-LVDS Enable
2479	10B8h	Any	BSC Clamp Threshold A
2481	20C7h	Any	BSC Clamp Threshold B
2499	0000h	CONFIG or IDLE	Test Pattern Control 1
24A1	536d	CONFIG or IDLE	Test Pattern Control 2
24B9	202d	CONFIG only	Companding Slope 1 Length
24C1	101d	CONFIG only	Companding Slope 2 Length
24C9	101d	CONFIG only	Companding Slope 3 Length
24D1	101d	CONFIG only	Companding Slope 4 Length
24D9	101d	CONFIG only	Companding Slope 5 Length
24E1	420d	CONFIG only	Companding Slope 6 Length
24E9	0083h	CONFIG only	Companding Slope 1/2 Gain
24F1	038Fh	CONFIG only	Companding Slope 3/4 Gain
24F9	0FBFh	CONFIG only	Companding Slope 5/6 Gain
2501	1F9Fh	CONFIG only	Companding Slope 7 Gain
2559	7804h	Any	Defect Avoidance Threshold
2561	003Fh	Any	Defect Avoidance Enable
25C1	0003h	CONFIG or IDLE	Encoded Sync Config
2619	000Bh	CONFIG only	Output Bank Select 2
4000	4100h	Any	Chip Revision Code
4008	0011h	Any	Chip ID Code MSB
4010	0080h	Any	Chip ID Code LSB
4019	0000h	Any	Set Sensor State
4021	0000h	CONFIG or IDLE	OTP Address
4029	0000h	CONFIG or IDLE	OTP Write Data
4031	0000h	CONFIG or IDLE	Command_Done_FB
4041	0000h	CONFIG or IDLE	OTP Read Data
4061	0000h	CONFIG or IDLE	Soft Reset

See Notes on following page.



Notes – SPI State (the Sensor State from which the register can be set):

1. "Any": Can be written from any state (including RUNNING)
2. "CONFIG or IDLE": These registers can be changed in IDLE or CONFIG states
3. "CONFIG only": Sensor must be in CONFIG state to set these registers
4. Only Register 4018h and 4060h may be set when the sensor is in STANDBY state
5. FB = Feedback, a read-only register that provides some error or status

Notes - Decimal, hexadecimal, binary values:

1. "b" denotes a binary number, a series of bits: MSB is on the left, LSB is on the right.
2. "h" or "hex" denotes a hexadecimal number (Base 16, 1-9, A-F). The letters in a hex number are always capitalized
3. "d" denotes a decimal number.
4. Note that "0" and "1" are the same value in all number base systems and sometimes the base notation is omitted.

The KAC-06040 features an embedded microprocessor by Cortus.

## Absolute Maximum Ratings

For Supplies and Inputs the maximum rating is defined as a level or condition that should not be exceeded at any time. If the level or the condition is exceeded, the device will be degraded and may be damaged. Operation at these values will reduce Mean Time to Failure (MTTF).

## SUPPLIES

Parameter	Value
AVDD_LV, VDD_DIG	-0.25 V; 2.3 V
AVDD_HV, Vref_P, VDD_LVDS	-0.25 V; 4 V
DC Input voltage at any input pin	-0.25 V; VDD_DIG +0.25 V

## CMOS INPUTS

Parameter	Symbol	Min	Typical	Max	Unit
Input voltage low level	$V_{IL}$	-0.3		0.35 VDD_DIG	V
Input voltage high level	$V_{IH}$	0.65 VDD_DIG		VDD_DIG +0.3	V



## Operating Ratings

### INPUT CLOCK CONDITIONS

Parameter	Min	Typical	Max	Unit
Frequency for Clk_In1 and Clk_In2	5	48	50	MHz
Duty cycle for Clk_In1 and Clk_In2	40	50	60	%
RESETN	10			ns
TRIGGER pin minimum pulse width	20			ns

TRIGGER must be active at least 4 periods of PLL1 (~12.5 ns at 320 MHz) to start a capture cycle. The polarity of the active level is configurable by Register 01D8h Bit 0; the default is active high (ie pin = VDD\_DIG = trigger request).

### OPERATING TEMPERATURE

Description	Symbol	Minimum	Maximum	Units	Notes
Operating Temperature	TOP	-50	+80	°C	1

Notes:

- Under conditions of no condensation on the sensor

### CMOS IN/OUT

CMOS IN/OUT characteristics

Parameter	Symbol	Min	Typical	Max	Unit
Output voltage low level	$V_{OL}$			0.45	V
Output voltage high level	$V_{OH}$	VDD_DIG-0.45			V
Input Hysteresis voltage	$V_{TH}$		0.25		
Pull-up resistor value for RESETN pin	$R_{PU}$	62			k $\Omega$
Pull-down resistor value for TRIGGER pin	$R_{PD}$	100			k $\Omega$
Current on ADC_REF pin	$I_{ADC\_REF}$		100		$\mu$ A



## SUPPLIES

Parameter	Symbol	Min	Typical	Max	Unit	Comment
LVDS IO supply	VDD_LVDS	3.15	3.30	3.63	V	
Pixel high voltage supply	AVDD_HV	3.40	3.50	3.60	V	
Pixel low voltage supply	Vref_P	2.71	2.80	2.88	V	
Analog power supply	AVDD_LV	1.71	1.80	1.89	V	
Digital power supply	VDD_Dig	1.9	2.00	2.10	V	
AVDD_HV – Vref_P			0.5		V	

Parameter	Symbol	Min	Typical	Max	Unit	Comment
Power in STANDBY State			2		mW	
Current in STANDBY State						
	VDD_LVDS		< 0.5		mA	
	AVDD_HV		< 0.5		mA	
	AVDD_LV		< 0.5		mA	
	Vref_P		< 0.5		mA	
	VDD_DIG		1		mA	
Power in CONFIG State			320		mW	
Current in CONFIG State						
	VDD_LVDS		< 0.5		mA	
	AVDD_HV		< 0.5		mA	
	AVDD_LV		< 0.5		mA	
	Vref_P		< 0.5		mA	
	VDD_DIG		162		mA	
Power in IDLE State			510		mW	
Current in IDLE State						
	VDD_LVDS		< 0.5		mA	
	AVDD_HV		20		mA	
	AVDD_LV		< 0.5		mA	
	Vref_P		< 0.5		mA	
	VDD_DIG		222		mA	
Power in RUNNING State			2.26		W	
Current in RUNNING State						
	VDD_LVDS in Sub-LVDS mode		115		mA	
	AVDD_HV		100		mA	
	AVDD_LV		20		mA	
	Vref_P		20		mA	
	VDD_DIG		721		mA	

Notes:

1. Voltages relative to VSS. Current measurements made in darkness.
2. Max Frame Rate (and thus maximum current mode)
  - a. Tri-Scan mode
  - b. 10 bit ADC
  - c. PLL2 = Max spec MHz,
  - d. No horizontal or vertical blanking and 8 active LVDS banks).





## SPI (Serial Peripheral Interface)

The SPI communication interface lets the application system to control and configure the sensor. The sensor has an embedded slave SPI interface. The application system is the master of the SPI bus.

Name	Sensor I/O direction	Description
CSN	I	SPI chip select - Active low, this input activates the slave interface in the sensor
SCK	I	SPI clock – toggled by the master
MISO	O	SPI master serial data input – slave(sensor) serial data output
MOSI	I	SPI master serial data output – slave(sensor) serial data input

Parameter	Min	Typ	Max	Unit
SPI SCK	5	25	50	MHz
Duty cycle on SPI SCK	40	50	60	%

### CLOCK POLARITY AND PHASE

CPOL(Clock POLarity) and CPHA(Clock PHase) are commonly defined in SPI protocol such as to define SCK clock phase and polarity. The KAC-06040 defaults to expecting the master to be configured with CPOL=1 (the base value of the clock is VDD\_DIG) and CPHA=1 (data is valid on the clock rising edge).

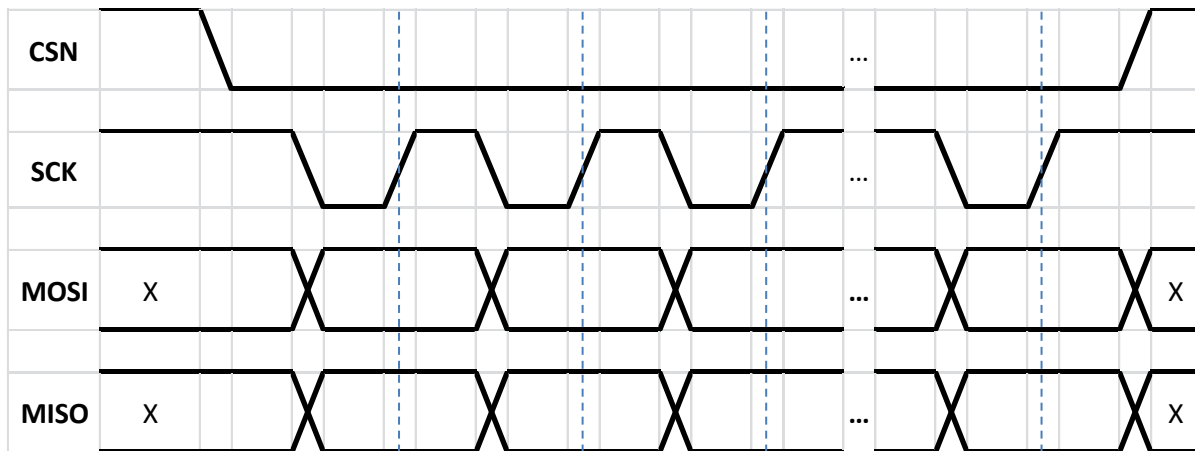


Figure 26: CPOL = 1 and CPHA = 1 configuration



## SPI PROTOCOL

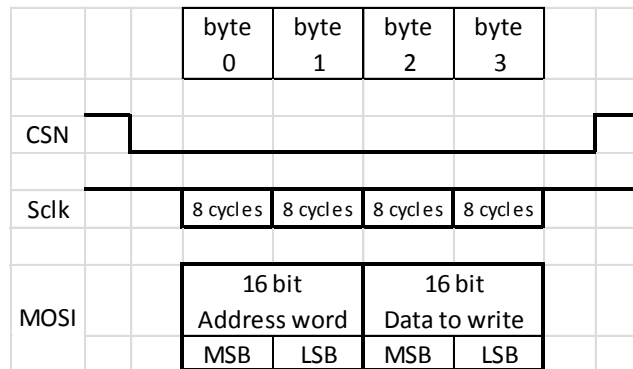


Figure 27: SPI Write byte order

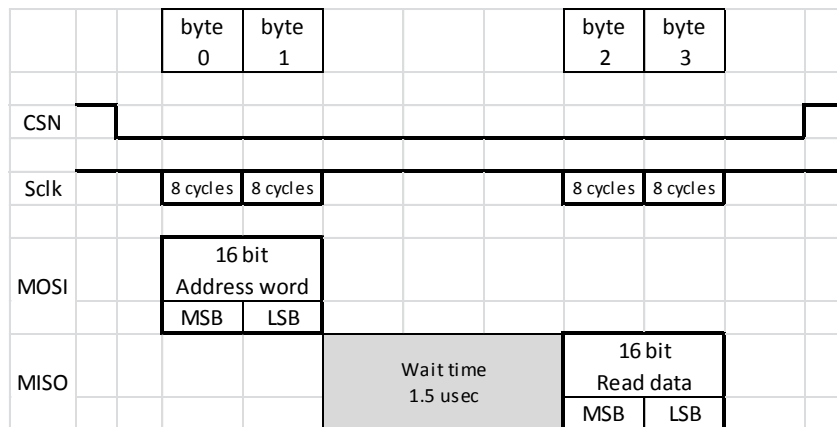


Figure 28: SPI Read byte order

There is a delay during readback between presenting the address to be read on the MOSI and being able to read the register contents on the MISO. This delay is not the same for all registers. Some are available immediately, some require a longer fetch time. The 1.5  $\mu$ sec shown in Figure 28 is the maximum time to fetch a register’s value when in CONFIG state (the recommended state for changing registers). Some registers can be adjusted during RUNNING state (see the register summary on page 37). If performing a readback during RUNNING state, the delay could be as long as 4.5  $\mu$ sec depending on when in the row the request was sent and the sensor’s microcontroller activity at that moment.

The SPI FB pin can be used to dynamically adjust the wait time for a register contents to be fetched. Figure 29 illustrates the use of the FB pin. The FB output will be low (VSS) until the requested register contents are ready to be clocked out of the MOSI pin. Once the FB pin goes high (VDD\_DIG) then clocking the Sclk will transmit the requested register contents. The SPI FB pin is inactive by default, this function is enabled in register 4041h.

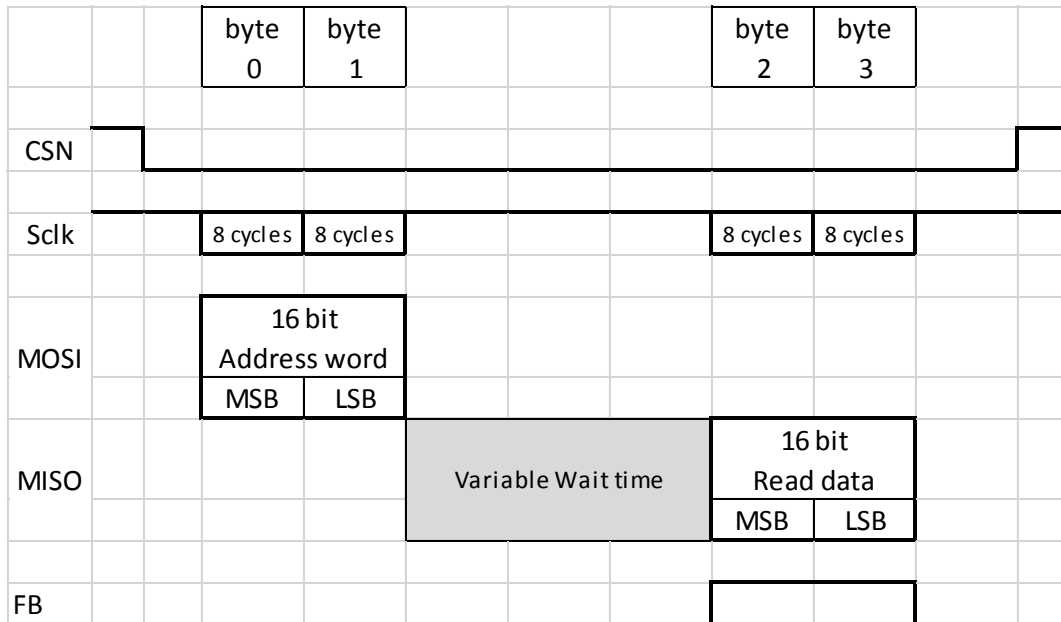


Figure 29: SPI Read with FBRB handshaking

Note that readback does not provide the actual register value being used, but reflects the next value to be used. All new register writes are placed in a shadow memory until they can be updated into the active memory. This active memory update occurs at the start of the next frame or upon entering the state listed in the Register Summary table on page 37. Register reads access this shadow memory not the active memory. For instance if the sensor is in RUNNING mode and you adjust the LL in register 200h. You can read back and confirm that your register change was received by the sensor; however, the LL will not change since register 200h can only be changed in CONFIG state. If you change the sensor state to CONFIG and then back to RUNNING, then the new LL will take effect.



### SPI INTERFACE

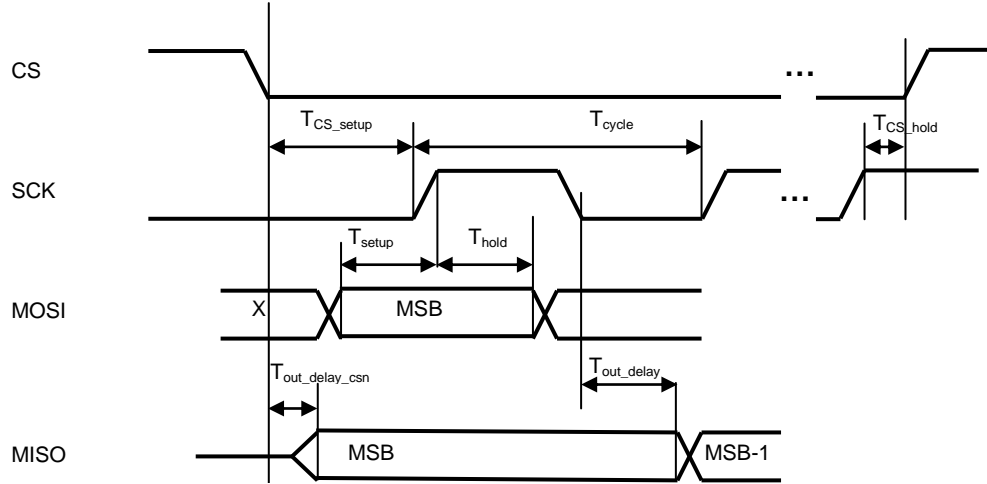


Figure 30: SPI timing chronogram

### SPI TIMING SPECIFICATION

Symbol	Min Value	Max Value
$T_{cycle}$	25 ns	
$T_{setup}$		2.9 ns
$T_{hold}$	0.8 ns	
$T_{cs\_setup}$	12.5 ns	
$T_{cs\_hold}$	12.5 ns	
$T_{out\_delay\_csn}$	3.1 ns	4.7 ns
$T_{out\_delay}$	4.9 ns	8.7 ns



## LVDS Interface

The data output can be configured to follow standard TIA/EIA-644-A LVDS specification or a low power mode compatible with common Sub-LVDS definition used in FPGA industry. (Please refer to the *KAC-06040 User Guide* for more information).

Unless otherwise noted, min/max characteristics are for  $T = -40\text{ }^{\circ}\text{C}$  to  $+85\text{ }^{\circ}\text{C}$ , output termination resistance  $R_L = 100\ \Omega \pm 1\%$ , Typical values are at  $V_{DD\_LVDS} = 3.3\text{ V}$ .

Use register 2449h to select standard or Sub-LVDS. This document assumes that Sub-LVDS is active for all power measurements. Standard LVDS can increase the average power consumption as much as 200 mW in the case of minimum horizontal and vertical blanking.

### STANDARD LVDS CHARACTERISTICS

Parameter	Symbol	Min	Typical	Max	Unit
Differential output voltage	VOD	250	355	450	mV
VOD variation between complementary output states	$\Delta V_{OD}$	-20		+20	mV
Common mode output voltage	VOCM	1.235	1.259	1.275	V
VOCM variation between complementary output states	$\Delta V_{OCM}$	-25		+25	mV
High impedance leakage current	IOZD	-1		+1	$\mu\text{A}$
Output short circuit current: when D+ or D- connected to ground when D+ or D- connected to 3.3V	IOSD	2.9 12.25		4.3 30.47	mA
Output capacitance	CDO		1.3		pF
Maximum Transmission Capacitive Load Expected (for 260 MHz LVDS clock)				10	pF

### SUB-LVDS CHARACTERISTICS

Parameter	Symbol	Min	Typical	Max	Unit
Differential output voltage	$V_{OD}$	140	180	220	mV
VOD variation between complementary output states	$\Delta V_{OD}$	-5		+5	mV
Common mode output voltage	$V_{OCM}$	0.88	0.9	0.92	V
VOCM variation between complementary output states	$\Delta V_{OCM}$	-10		+10	mV
High impedance leakage current	$I_{OZD}$	-1		+1	$\mu\text{A}$
Output short circuit current: when D+ or D- connected to ground when D+ or D- connected to 3.3V	$I_{OSD}$	1.4 10.21		2.2 30.47	mA
Output capacitance	$C_{DO}$		1.3		pF
Maximum Transmission Capacitive Load Expected (for 260 MHz LVDS clock)				10	pF

Parameter	Min	Typical	Max	Unit
LVDS_CLK	50	160	160	MHz
Duty cycle on LVDS_CLK		50		%



### IN-BLOCK LVDS TIMING SPECIFICATION

The table below gives LVDS timing specification for one group of LVDS for nominal frequency of 260 MHz. There is no skew specification between groups.

Parameter	Symbol	Min	Typical	Max	Unit
Minimum time between data change and clock rising edge	$t_{SDLH}$	600			ps
Minimum time between clock rising edge and data change	$t_{HDLH}$	600			ps
Minimum time between data change and clock falling edge	$t_{SDHL}$	600			ps
Minimum time between clock falling edge and data change	$t_{HDHL}$	600			ps
Maximum differential skew between the 7 data pairs	$t_{SKD}$		200	700	ps

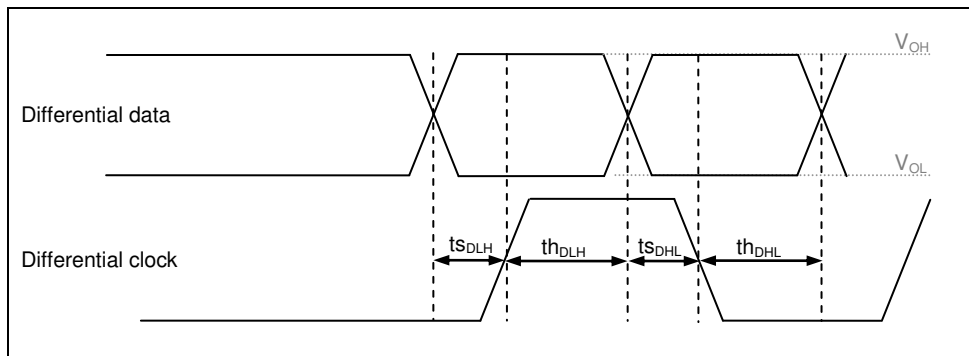


Figure 31: LVDS timing chronogram

### INTER-BLOCK LVDS TIMING SPECIFICATION

Parameter	Min	Typical	Max	Unit
Inter-block skew		6	12	LVDS Clock periods



## Storage and Handling

### STORAGE CONDITIONS

Description	Symbol	Minimum	Maximum	Units	Notes
Storage Temperature	TST	-40	+80	°C	1
Humidity	RH	5	90	%	2

#### Notes:

1. Long-term storage toward the maximum temperature will accelerate color filter degradation.
2. T=25 °C. Excessive humidity will degrade MTTF.

### ESD

1. This device contains limited protection against Electrostatic Discharge (ESD). ESD events may cause irreparable damage to an image sensor either immediately or well after the ESD event occurred. Failure to protect the sensor from electrostatic discharge may affect device performance and reliability.
2. Devices should be handled in accordance with strict ESD procedures. Devices are shipped in static-safe containers and should only be handled at static-safe workstations.
3. See Application Note *Image Sensor Handling Best Practices* for proper handling and grounding procedures. This application note also contains workplace recommendations to minimize electrostatic discharge.
4. Store devices in containers made of electro-conductive materials.

### COVER GLASS CARE AND CLEANLINESS

1. The cover glass is highly susceptible to particles and other contamination. Perform all assembly operations in a clean environment.
2. Touching the cover glass must be avoided.

3. Improper cleaning of the cover glass may damage these devices. Refer to Application Note *Image Sensor Handling Best Practices*.

### ENVIRONMENTAL EXPOSURE

1. Extremely bright light can potentially harm image sensors. Do not expose to strong sunlight for long periods of time, as the color filters and/or microlenses may become discolored. In addition, long time exposures to a static high contrast scene should be avoided. Localized changes in response may occur from color filter/microlens aging.
2. Exposure to temperatures exceeding maximum specified levels should be avoided for storage and operation, as device performance and reliability may be affected.
3. Avoid sudden temperature changes.
4. Exposure to excessive humidity may affect device characteristics and may alter device performance and reliability, and therefore should be avoided.
5. Avoid storage of the product in the presence of dust or corrosive agents or gases, as deterioration of lead solderability may occur. It is advised that the solderability of the device leads be assessed after an extended period of storage, over one year.

### SOLDERING RECOMMENDATIONS

1. For manual soldering the soldering iron tip temperature is not to exceed 370 °C. Higher temperatures may alter device performance and reliability.
2. Re-Flow soldering method is not recommended. Wave soldering may be employed, however solder should not touch the cover glass.





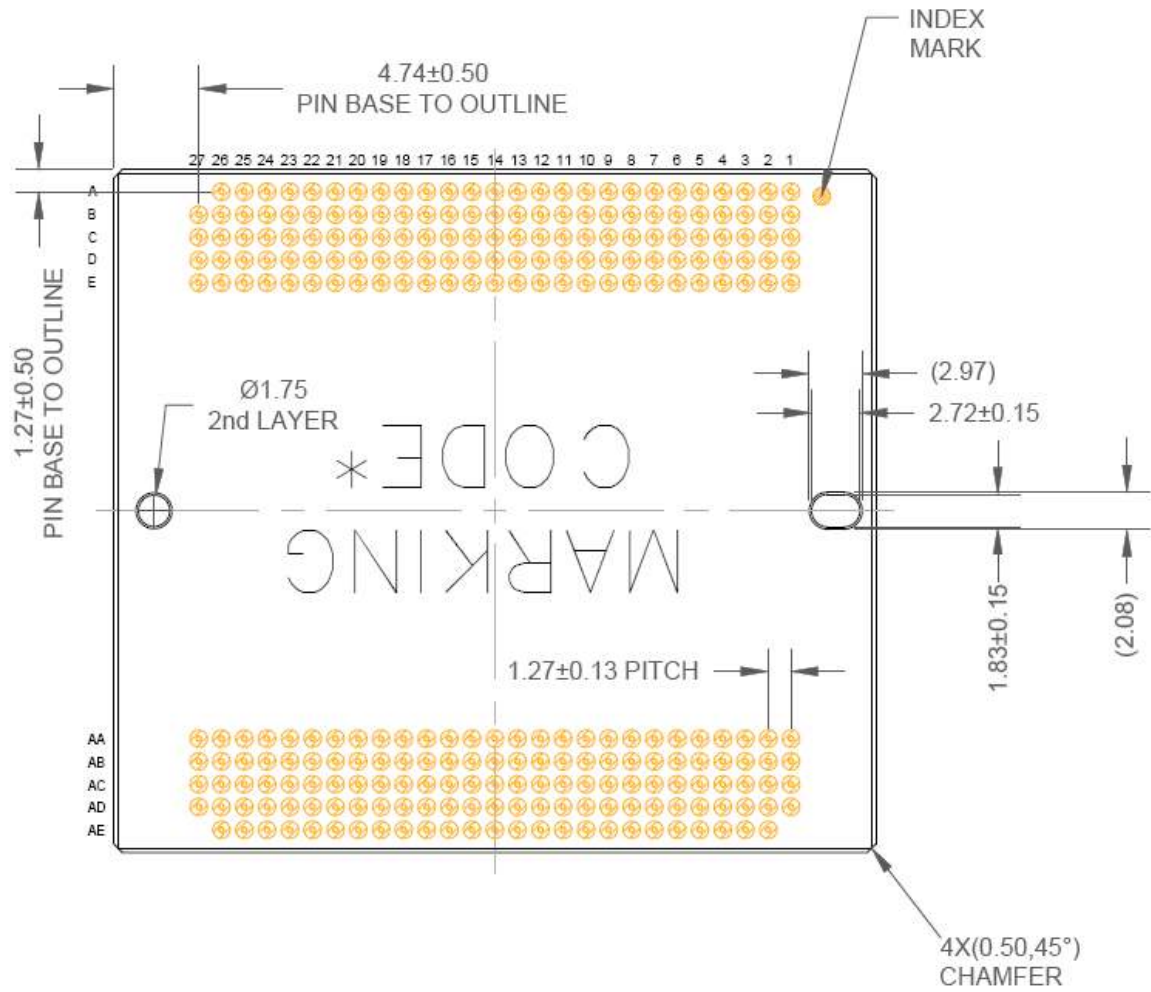


Figure 33: Completed Assembly (2 of 5)

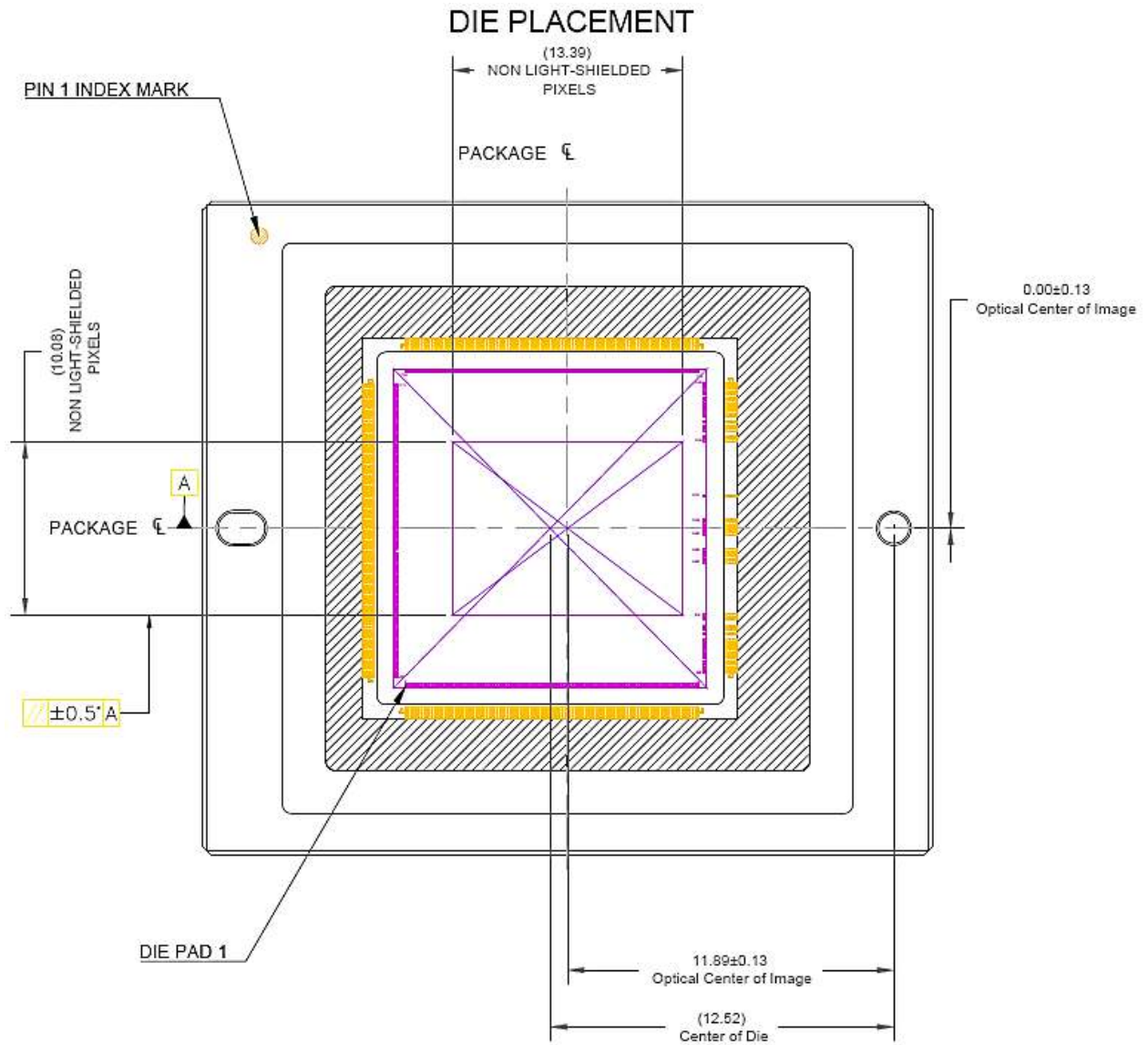


Figure 34: Completed Assembly (3 of 5)

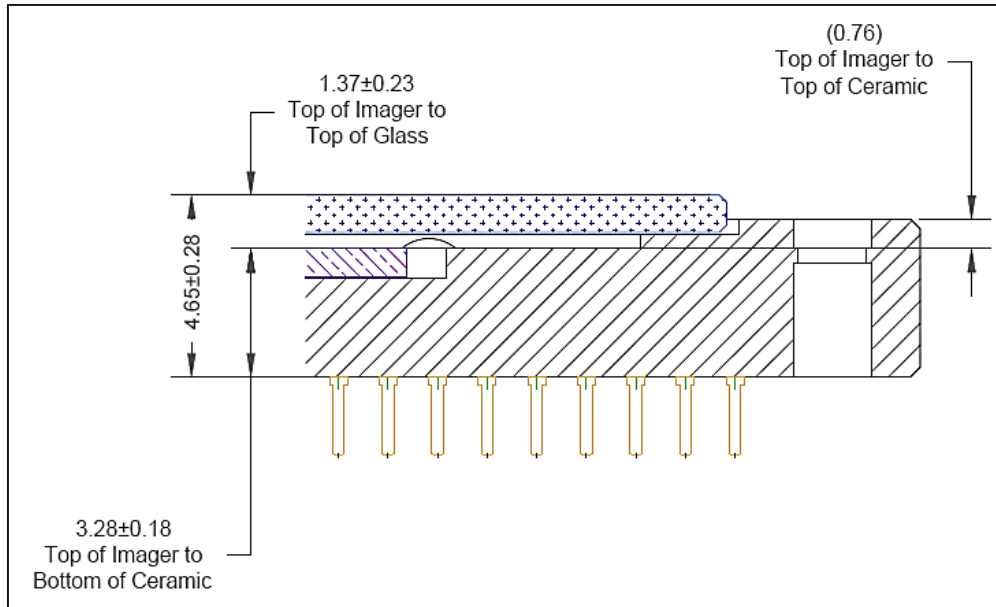


Figure 35: Completed Assembly (4 of 5)

### IMAGE ORIENTATION

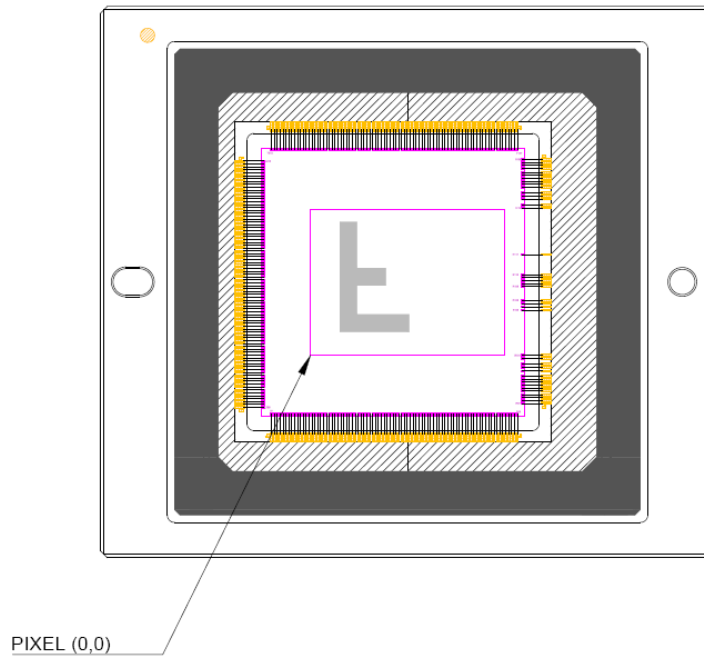


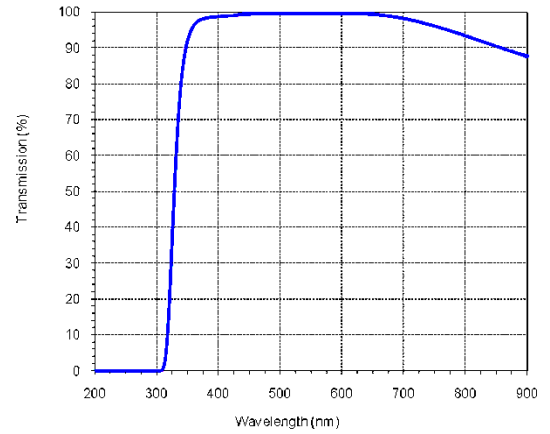
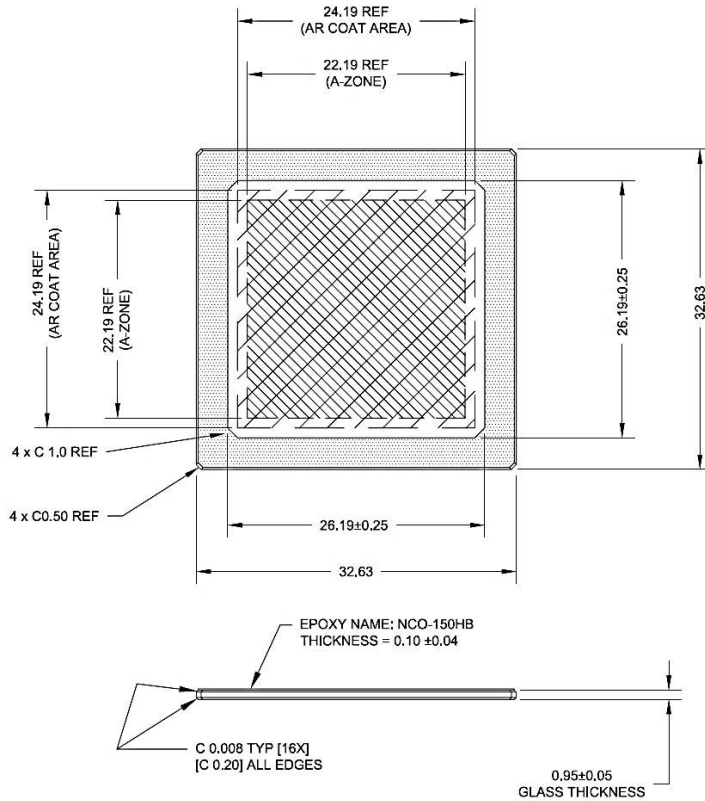
Figure 36: Completed Assembly (5 of 5)



## MAR (MULTI-LAYER ANTI-REFLECTIVE COATING) COVER GLASS

Notes:

1. Units: IN [MM]
2. A-Zone Dust/Scratch Spec: 10 µm Maximum
3. Index of refraction: 1.5231



405-450 nm T > 98%  
 450-650 nm T > 99%  
 650-690 nm T > 98%  
 690-770 nm T > 94%  
 770-870 nm T > 88%

Figure 37: MAR Cover Glass Specification



## Quality Assurance and Reliability

### QUALITY AND RELIABILITY

All image sensors conform to the specifications stated in this document. This is accomplished through a combination of statistical process control and visual inspection and electrical testing at key points of the manufacturing process, using industry standard methods. Information concerning the quality assurance and reliability testing procedures and results are available from ON Semiconductor upon request. For further information refer to Application Note *Quality and Reliability*.

### REPLACEMENT

All devices are warranted against failure in accordance with the *Terms of Sale*. Devices that fail due to mechanical and electrical damage caused by the customer will not be replaced.

### LIABILITY OF THE SUPPLIER

A reject is defined as an image sensor that does not meet all of the specifications in this document upon receipt by the customer. Product liability is limited to the cost of the defective item, as defined in the *Terms of Sale*.

### LIABILITY OF THE CUSTOMER

Damage from mishandling (scratches or breakage), electrostatic discharge (ESD), or other electrical misuse of the device beyond the stated operating or storage limits, which occurred after receipt of the sensor by the customer, shall be the responsibility of the customer.

### TEST DATA RETENTION

Image sensors shall have an identifying number traceable to a test data file. Test data shall be kept for a period of 2 years after date of delivery.

### MECHANICAL

The device assembly drawing is provided as a reference.

ON Semiconductor reserves the right to change any information contained herein without notice. All information furnished by ON Semiconductor is believed to be accurate.

## Life Support Applications Policy

ON Semiconductor image sensors are not authorized for and should not be used within Life Support Systems without the specific written consent of ON Semiconductor.



## Revision Changes

Revision Number	Description of Changes
1.0	<ul style="list-style-type: none"> <li>Initial Release</li> </ul>
1.1	<ul style="list-style-type: none"> <li>Added SPI FB pin</li> <li>Added cover glass index of refraction</li> <li>Updated branding</li> </ul>

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