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FM31276/FM31278

64-Kbit/256-Kbit Integrated Processor Companion with F-RAM

Features

- 64-Kbit/256-Kbit ferroelectric random access memory (F-RAM) ❐ Logically organized as 8K × 8 (FM31276)/32K × 8 (FM31278)
	- **□ High-endurance 100 trillion (10¹⁴) read/writes**
	- ❐ 151-year data retention (See [Data Retention and Endurance](#page-28-0) [on page 28](#page-28-0))
	- ❐ NoDelay™ writes
	- ❐ Advanced high-reliability ferroelectric process
- High Integration Device Replaces Multiple Parts
	- ❐ Serial nonvolatile memory
	- ❐ Real time clock (RTC)
	- ❐ Low voltage reset
	- ❐ Watchdog timer
	- ❐ Early power-fail warning/NMI
	- ❐ Two 16-bit event counter
	- ❐ Serial number with write-lock for security
- Real-time Clock/Calendar
	- \Box Backup current at 2 V: 1.15 μ A at +25 °C
	- ❐ Seconds through centuries in BCD format
	- ❐ Tracks leap years through 2099
	- ❐ Uses standard 32.768 kHz crystal (6 pF/12.5 pF)
	- ❐ Software calibration
	- ❐ Supports battery or capacitor backup
- Processor Companion
	- \Box Active-low reset output for V_{DD} and watchdog
	- \Box Programmable low-V_{DD} reset trip point
	- ❐ Manual reset filtered and debounced
	- ❐ Programmable watchdog timer
	- ❐ Dual Battery-backed event counter tracks system intrusions or other events
	- ❐ Comparator for power-fail interrupt
	- ❐ 64-bit programmable serial number with lock
- **E** Fast 2-wire serial interface (I^2C)
	- ❐ Up to 1-MHz frequency
	- ❐ Supports legacy timings for 100 kHz and 400 kHz
	- □ RTC, Supervisor controlled via I²C interface
	- ❐ Device select pins for up to 4 memory devices
- Low power consumption ❐ 1.5 mA active current at 1 MHz
	- \Box 150 μA standby current
- Operating voltage: V_{DD} = 4.0 V to 5.5 V
- Industrial temperature: -40 °C to +85 °C
- 14-pin small outline integrated circuit (SOIC) package
- Restriction of hazardous substances (RoHS) compliant
- Underwriters laboratory (UL) recognized

Functional Description

The FM31276/FM31278 device integrates F-RAM memory with the most commonly needed functions for processor-based systems. Major features include nonvolatile memory, real time clock, low- V_{DD} reset, watchdog timer, nonvolatile event counter, lockable 64-bit serial number area, and general purpose comparator that can be used for a power-fail (NMI) interrupt or any other purpose.

The FM31276/FM31278 is a 64-Kbit/256-Kbit nonvolatile memory employing an advanced ferroelectric process. A ferroelectric random access memory or F-RAM is nonvolatile and performs reads and writes similar to a RAM. This memory is truly nonvolatile rather than battery backed. It provides reliable data retention for 151 years while eliminating the complexities, overhead, and system-level reliability problems caused by other nonvolatile memories. The FM31276/FM31278 is capable of supporting 10¹⁴ read/write cycles, or 100 million times more write cycles than EEPROM.

The real time clock (RTC) provides time and date information in BCD format. It can be permanently powered from an external backup voltage source, either a battery or a capacitor. The timekeeper uses a common external 32.768 kHz crystal and provides a calibration mode that allows software adjustment of timekeeping accuracy.

The processor companion includes commonly needed CPU support functions. Supervisory functions include a reset output signal controlled by either a low V_{DD} condition or a watchdog timeout. $\overline{\text{RST}}$ goes active when V_{DD} drops below a programmable threshold and remains active for 100 ms after V_{DD} rises above the trip point. A programmable watchdog timer runs from 100 ms to 3 seconds. The watchdog timer is optional, but if enabled it will assert the reset signal for 100 ms if not restarted by the host before the timeout. A flag-bit indicates the source of the reset.

A comparator on PFI compares an external input pin to the onboard 1.2 V reference. This is useful for generating a power-fail interrupt (NMI) but can be used for any purpose. The family also includes a programmable 64-bit serial number that can be locked making it unalterable. Additionally it offers a dual battery-backed event counter that tracks the number of rising or falling edges detected on a dedicated input pin.

For a complete list of related documentation, click [here](http://www.cypress.com/?rID=76639).

Logic Block Diagram

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Pinout

Figure 1. 14-pin SOIC pinout

Pin Definitions

Functional Overview

The FM31276/FM31278 device combines a serial nonvolatile RAM with a real time clock (RTC) and a processor companion. The companion is a highly integrated peripheral including a processor supervisor, a comparator used for early power-fail warning, nonvolatile event counters, and a 64-bit serial number. The FM31276/FM31278 integrates these complementary but distinct functions under a common interface in a single package. The product is organized as two logical devices. The first is a memory and the second is the companion which includes all the remaining functions. From the system perspective they appear to be two separate devices with unique IDs on the serial bus.

The memory is organized as a standalone nonvolatile I^2C memory using standard device ID value. The real time clock and supervisor functions are accessed with a separate I²C device ID. This allows clock/calendar data to be read while maintaining the most recently used memory address. The clock and supervisor functions are controlled by 25 special function registers. The RTC and event counter circuits are maintained by the power source on the V_{BAK} pin, allowing them to operate from battery or backup capacitor power when V_{DD} drops below a set threshold. Each functional block is described below.

Memory Architecture

The FM31276/FM31278 device is available in memory size 64-Kbit/256-Kbit. The device uses two-byte addressing for the memory portion of the chip. This makes the device software compatible with its standalone memory counterparts, but makes them compatible within the entire family.

The memory array is logically organized as 8,192 × 8 bits/

 $32,768 \times 8$ bits and is accessed using an industry-standard I²C interface. The memory is based on F-RAM technology. Therefore it can be treated as RAM and is read or written at the speed of the I^2C bus with no delays for write operations. It also offers effectively unlimited write endurance unlike other nonvolatile memory technologies. The I²C protocol is described in [I2C Interface on page 20](#page-20-0).

The memory array can be write-protected by software. Two bits in the processor companion area (WP1, WP0 in register 0Bh) control the protection setting. Based on the setting, the protected addresses cannot be written and the I^2C interface will not acknowledge any data to protected addresses. The special function registers containing these bits are described in detail below.

Processor Companion

In addition to nonvolatile RAM, the FM31276/FM31278 incorporates a real time clock and highly integrated processor companion. The companion includes a low- V_{DD} reset, a programmable watchdog timer, a battery-backed event counters, a comparator for early power-fail detection or other purposes, and a 64-bit serial number.

Processor Supervisor

Supervisors provide a host processor two basic functions: detection of power supply fault conditions and a watchdog timer to escape a software lockup condition. The FM31276/FM31278 has a reset pin (RST) to drive a processor reset input during power faults, power-up, and software lockups. It is an open drain output with a weak internal pull-up to V_{DD} . This allows other reset sources to be wire-OR'd to the RST pin. When V_{DD} is above the programmed trip point, RST output is pulled weakly to V_{DD} . If V_{DD} drops below the reset trip point voltage level (V_{TP}), the RST pin will be driven LOW. It will remain LOW until V_{DD} falls too low for circuit operation which is the V_{RST} level. When V_{DD} rises again above V_{TP} RST continues to drive LOW for at least 100 ms (t_{RPU}) to ensure a robust system reset at a reliable V_{DD} level. After t_{RPU} has been met, the RST pin will return to the weak HIGH state. While RST is asserted, serial bus activity is locked out even if a transaction occurred as V_{DD} dropped below V_{TP} . A memory operation started while V_{DD} is above V_{TP} will be completed internally.

[Table 2](#page-5-4) below shows how bit VTP controls the trip point of the low-V_{DD} reset. They are located in register 0Bh, bits 1 and 0. The reset pin will drive LOW when V_{DD} is below the selected V_{TP} voltage, and the I²C interface and F-RAM array will be locked out. Note that the bit 1 location is a don't care. [Figure 2](#page-5-5) illustrates the reset operation in response to a low V_{DD} .

Table 2. VTP setting

Figure 2. Low V_{DD} Reset

A watchdog timer can also be used to drive an active reset signal. The watchdog is a free-running programmable timer. The timeout period can be software programmed from 100 ms to 3 seconds in 100 ms increments via a 5-bit nonvolatile register.

All programmed settings are minimum values and vary with temperature according to the operating specifications. The watchdog has two additional controls associated with its operation, a watchdog enable bit (WDE) and timer restart bits (WR). Both the enable bit must be set and the watchdog must timeout in order to drive RST active. If a reset event occurs, the timer will automatically restart on the rising edge of the reset pulse. If WDE = '0', the watchdog timer runs but a watchdog fault will not cause RST to be asserted LOW. The WTR flag will be set, indicating a watchdog fault. This setting is useful during software development if the developer does not want RST to drive. Note that setting the maximum timeout setting (11111b) disables the counter to save power. The second control is a nibble that restarts the timer preventing a reset. The timer should be restarted after changing the timeout value.

The watchdog timeout value is located in register 0Ah, bits 4:0, and the watchdog enable is bit 7. The watchdog is restarted by writing the pattern 1010b to the lower nibble of register 09h. Writing this pattern will also cause the timer to load new timeout values. Writing other patterns to this address will not affect its operation. Note the watchdog timer is free-running. Prior to enabling it, users should restart the timer as described above. This assures that the full timeout period will be set immediately after enabling. The watchdog is disabled when V_{DD} is below V_{TP} . The following table summarizes the watchdog bits. A block diagram follows.

Figure 3. Watchdog Timer

Manual Reset

The RST is a bi-directional signal allowing the FM31276/FM31278 to filter and de-bounce a manual reset switch. The RST input detects an external low condition and responds by driving the RST signal LOW for 100 ms.

Figure 4. Manual Reset

Note The internal weak pull-up eliminates the need for additional external components.

Reset Flags

In case of a reset condition, a flag bit will be set to indicate the source of the reset. A low- V_{DD} reset is indicated by the POR flag, register 09h bit 6. A watchdog reset is indicated by the WTR flag, register 09h bit 7. Note that the flags are internally set in response to reset sources, but they must be cleared by the user. When the register is read, it is possible that both flags are set if both have occurred since the user last cleared them.

Early Power Fail Comparator

An early power fail warning can be provided to the processor well before V_{DD} drops out of spec. The comparator is used to create a power fail interrupt (NMI). This can be accomplished by connecting the PFI pin to the unregulated power supply via a resistor divider. An application circuit is shown below.

Figure 5. Comparator as a Power-Fail Warning

The voltage on the PFI input pin is compared to an onboard 1.2 V reference. When the PFI input voltage drops below this threshold, the comparator will drive the CAL/PFO pin to a LOW state. The comparator has 100 mV (max) of hysteresis to reduce noise sensitivity, only for a rising PFI signal. For a falling PFI edge, there is no hysteresis.

The comparator is a general purpose device and its application is not limited to the NMI function.

The comparator is not integrated into the special function registers except as it shares its output pin with the CAL output. When the RTC calibration mode is invoked by setting the CAL bit (register 00h, bit 2), the CAL/PFO output pin will be driven with a 512 Hz square wave and the comparator will be ignored. Since most users only invoke the calibration mode during production, this should have no impact on system operations using the comparator.

Note The maximum voltage on the comparator input PFI is limited to 3.75 V under normal operating conditions.

Event Counter

The FM31276/FM31278 offers the user two battery-backed event counters. Input pins CNT1 and CNT2 are programmable edge detectors. Each clocks a 16-bit counter. When an edge occurs, the counters will increment their respective registers. Counter 1 is located in registers 0Dh and 0Eh, Counter 2 is located in registers 0Fh and 10h. These register values can be read anytime V_{DD} is above V_{TP} , and they will be incremented as
long as a valid V_{BAK} power source is provided. To read, set the long as a valid V_{BAK} power source is provided. To read, set the
RC bit, register 0Ch bit 3 to 1. This takes a snapshot of all four counter bytes allowing a stable value even if a count occurs during the read. The registers can be written by software allowing the counters to be cleared or initialized by the system. Counts are blocked during a write operation. The two counters can be cascaded to create a single 32-bit counter by setting the CC control bit (register 0Ch, bit 2). When cascaded, the CNT1 input will cause the counter to increment. CNT2 is not used in this mode and should be tied to ground.

Figure 6. Event Counter

The control bits for event counting are located in register 0Ch. Counter 1 Polarity is bit C1P, bit 0; Counter 2 Polarity is C2P, bit 1; the Cascade Control is CC, bit 2; and the Read Counter bit is RC, bit 3.

The polarity bits must be set prior to setting the counter value(s). If a polarity bit is changed, the counter may inadvertently increment. If the counter pins are not being used, tie them to ground.

Serial Number

A memory location to write a 64-bit serial number is provided. It is a writeable nonvolatile memory block that can be locked by the user once the serial number is set. The 8 bytes of data and the lock bit are all accessed via the device ID for the Processor Companion. Therefore the serial number area is separate and distinct from the memory array. The serial number registers can be written an unlimited number of times, so these locations are general purpose memory. However, once the lock bit is set, the values cannot be altered and the lock cannot be removed. Once locked the serial number registers can still be read by the system.

The serial number is located in registers 11h to 18h. The lock bit is SNL (register 0Bh, bit 7). Setting the SNL bit to a '1' disables writes to the serial number registers, and the SNL bit cannot be cleared.

Real-time Clock Operation

The real-time clock (RTC) is a timekeeping device that can be battery or capacitor backed for permanently-powered operation. It offers a software calibration feature that allows high accuracy.

The RTC consists of an oscillator, clock divider, and a register system for user access. It divides down the 32.768 kHz time-base and provides a minimum resolution of seconds (1 Hz). Static registers provide the user with read/write access to the time values. It includes registers for seconds, minutes, hours, day-of-the-week, date, months, and years. A block diagram ([Figure 7 on page 8](#page-8-2)) illustrates the RTC function.

The user registers are synchronized with the timekeeper core using R and W bits in register 00h described below. Changing the R bit from '0' to '1' transfers timekeeping information from the core into holding registers that can be read by the user. If a timekeeper update is pending when R is set, then the core will be updated prior to loading the user registers. The registers are frozen and will not be updated again until the R bit is cleared to '0'. R is used for reading the time.

Setting the W bit to '1' locks the user registers. Clearing it to '0' causes the values in the user registers to be loaded into the timekeeper core. W bit is used for writing new time values. Users should be certain not to load invalid values, such as FFh, to the timekeeping registers. Updates to the timekeeping core occur continuously except when locked.

Figure 7. Real-time Clock Core Block Diagram

Backup Power

The real-time clock/calendar is intended to be permanently powered. When the primary system power fails, the voltage on the V_{DD} pin will drop. When V_{DD} is less than 2.5 V, the RTC (and event counters) will switch to the backup power supply on V_{BAK} . The clock operates at extremely low current in order to maximize battery or capacitor life. However, an advantage of combining a clock function with F-RAM memory is that data is not lost regardless of the backup power source.

The I_{BAK} current varies with temperature and voltage (see [DC](#page-26-2) [Electrical Characteristics on page 26\)](#page-26-2). The following graph shows I_{BAK} as a function of V_{BAK}. These curves are useful for calculating backup time when a capacitor is used as the V_{BAK} source.

The minimum V_{BAK} voltage varies linearly with temperature. The user can expect the minimum V_{BAK} voltage to be 1.23 V at +85 °C and 1.90 V at -40 °C. The tested limit is 1.55 V at +25 °C.

Note The minimum V_{BAK} voltage has been characterized at -40 °C and +85 °C but is not 100% tested.

Figure 9. V_{BAK}(min.) vs Temperature

Trickle Charger

To facilitate capacitor backup the V_{BAK} pin can optionally provide a trickle charge current. When the VBC bit (register 0Bh, bit 2) is set to '1', the V_{BAK} pin will source approximately 80 µA until V_{BAK} reaches 3.75 V. In 5 V systems, this charges the capacitor to V_{DD} without an external diode and resistor charger and also prevents the user from exceeding the V_{BAK} maximum voltage specification. There is a Fast Charge mode which is enabled by the FC bit (register 0Bh, bit 5). In this mode the trickle charger current is set to approximately 1 mA, allowing a large backup capacitor to charge more quickly.

In the case where no battery is used, the V_{BAK} pin should be tied to V_{SS} . V_{BAK} should not be tied to 5 V since the V_{BAK} (max) specification will be exceeded. Be sure to turn off the trickle charger (VBC = '0'), otherwise charger current will be shunted to ground from V_{DD} .

Note Systems using lithium batteries should clear the VBC bit to '0' to prevent battery charging. The V_{BAK} circuitry includes an internal 1 K Ω series resistor as a safety element. The trickle charger is UL Recognized.

Calibration

When the CAL bit in the register 00h is set to '1', the clock enters calibration mode. In calibration mode, the CAL/PFO output pin is dedicated to the calibration function and the power fail output is temporarily unavailable. Calibration operates by applying a digital correction to the counter based on the frequency error. In this mode, the CAL/PFO pin is driven with a 512 Hz (nominal) square wave. Any measured deviation from 512 Hz translates into a timekeeping error. The user converts the measured error in ppm and writes the appropriate correction value to the calibration register. The correction factors are listed in the table below. Positive ppm errors require a negative adjustment that removes pulses. Negative ppm errors require a positive correction that adds pulses. Positive ppm adjustments have the CALS (sign) bit set to '1', whereas negative ppm adjustments have CALS = '0'. After calibration, the clock will have a maximum error of ±2.17 ppm or ±0.09 minutes per month at the calibrated temperature.

The calibration setting is stored in F-RAM so it is not lost should the backup source fail. It is accessed with bits CAL(4:0) in register 01h. This value can be written only when the CAL bit is set to a '1'. To exit the calibration mode, the user must clear the CAL bit to a '0'. When the CAL bit is '0', the CAL/PFO pin will revert to the power fail output function.

Crystal Oscillator

The crystal oscillator is designed to use a 6 pF/12.5 pF crystal without the need for external components, such as loading capacitors. The FM31276/FM31278 device has built-in loading capacitors that are optimized for use with 6 pF crystals, but which work well with 12.5 pF crystals. For either crystal, no additional external loading capacitors are required nor suggested.

If a 32.768 kHz crystal is not used, an external oscillator may be connected to the FM31276/FM31278. Apply the oscillator to the X1 pin. Its high and low voltage levels can be driven rail-to-rail or amplitudes as low as approximately 500 mV p-p. To ensure proper operation, a DC bias must be applied to the X2 pin. It should be centered between the high and low levels on the X1 pin. This can be accomplished with a voltage divider.

Figure 10. External Oscillator

In the example, R1 and R2 are chosen such that the X2 voltage is centered around the X1 oscillator drive levels. If you wish to avoid the DC current, you may choose to drive X1 with an external clock and X2 with an inverted clock using a CMOS inverter.

Layout Recommendations

The X1 and X2 crystal pins employ very high impedance circuits and the oscillator connected to these pins can be upset by noise or extra loading. To reduce RTC clock errors from signal switching noise, a guard ring must be placed around these pads and the guard ring grounded. SDA and SCL traces should be routed away from the X1 / X2 pads. The X1 and X2 trace lengths should be less than 5 mm. The use of a ground plane on the backside or inner board layer is preferred. See layout example. Red is the top layer, green is the bottom layer.

Layout for Surface Mount Crystal (red = top layer, green = bottom layer)

Figure 11. Layout Recommendations

Layout for Through Hole Crystal (red = top layer, green = bottom layer)

Table 3. Digital Calibration Adjustments

Table 3. Digital Calibration Adjustments (continued)

Register Map

The RTC and processor companion functions are accessed via 25 special function registers, which are mapped to a separate I^2C device ID. The interface protocol is described on [I2C Interface on page 20.](#page-20-0) The registers contain timekeeping data, control bits, and information flags. A description of each register follows the summary table.

Table 4. Register Map Summary Table

Nonvolatile = \Box Battery-backed = \Box

Note When the device is first powered up and programmed, all timekeeping registers must be written because the battery-backed register values cannot be guaranteed. The table below shows the default values of the non-volatile registers. All other register values should be treated as unknown.

Table 5. Default Register Values

Table 6. Register Description

I ²C Interface

The FM31276/FM31278 employs an industry standard I²C bus that is familiar to many users. This product is unique since it incorporates two logical devices in one chip. Each logical device can be accessed individually. Although monolithic, it appears to the system software to be two separate products. One is a memory device. It has a Slave Address (Slave ID = 1010b) that operates the same as a stand-alone memory device. The second device is a real-time clock and processor companion which have a unique Slave Address (Slave ID = 1101b).

By convention, any device that is sending data onto the bus is the transmitter while the target device for this data is the receiver. The device that is controlling the bus is the master. The master is responsible for generating the clock signal for all operations. Any device on the bus that is being controlled is a slave. The FM31276/FM31278 is always a slave device.

The bus protocol is controlled by transition states in the SDA and SCL signals. There are four conditions including START, STOP, data bit, or acknowledge. [Figure 12](#page-20-4) and [Figure 13](#page-20-5) illustrates the signal conditions that specify the four states. Detailed timing diagrams are shown in the electrical specifications section.

STOP Condition (P)

A STOP condition is indicated when the bus master drives SDA from LOW to HIGH while the SCL signal is HIGH. All operations using the FM31276/FM31278 should end with a STOP condition. If an operation is in progress when a STOP is asserted, the operation will be aborted. The master must have control of SDA in order to assert a STOP condition.

START Condition (S)

A START condition is indicated when the bus master drives SDA from HIGH to LOW while the SCL signal is HIGH. All commands should be preceded by a START condition. An operation in progress can be aborted by asserting a START condition at any time. Aborting an operation using the START condition will ready the FM31276/FM31278 for a new operation.

If during operation the power supply drops below the specified V_{TP} minimum, any 1^2C transaction in progress will be aborted and the system should issue a START condition prior to performing another operation.

Figure 12. START and STOP Conditions

Data/Address Transfer

All data transfers (including addresses) take place while the SCL signal is HIGH. Except under the three conditions described above, the SDA signal should not change while SCL is HIGH.

Figure 13. Data Transfer on the I2C Bus

Acknowledge/No-acknowledge

The acknowledge takes place after the 8th data bit has been transferred in any transaction. During this state the transmitter should release the SDA bus to allow the receiver to drive it. The receiver drives the SDA signal LOW to acknowledge receipt of the byte. If the receiver does not drive SDA LOW, the condition is a no-acknowledge and the operation is aborted.

The receiver would fail to acknowledge for two distinct reasons. First is that a byte transfer fails. In this case, the no-acknowledge ceases the current operation so that the device can be addressed again. This allows the last byte to be recovered in the event of a communication error.

Second and most common, the receiver does not acknowledge to deliberately end an operation. For example, during a read operation, the FM31276/FM31278 will continue to place data onto the bus as long as the receiver sends acknowledges (and clocks). When a read operation is complete and no more data is needed, the receiver must not acknowledge the last byte. If the receiver acknowledges the last byte, this will cause the FM31276/FM31278 to attempt to drive the bus on the next clock while the master is sending a new command such as STOP.

Slave Address

The first byte that the FM31276/FM31278 expects after a START condition is the slave address. As shown in [Figure 15](#page-21-3) and [Figure 16,](#page-21-2) the slave address contains the device type or slave ID, the device select address bits, and a bit that specifies if the transaction is a read or a write.

The FM31276/FM31278 has two Slave Addresses (Slave IDs) associated with two logical devices. Bits 7–4 are the device type (slave ID) and should be set to 1010b for the memory device. The other logical device within the FM31276/FM31278 is the real-time clock and companion. Bits 7–4 are the device type (slave ID) and should be set to 1101b for the RTC and companion. A bus transaction with this slave address will not affect the memory in any way. The figures below illustrate the two Slave Addresses.

Bits 2–1 are the device select address bits. They must match the corresponding value on the external address pins to select the device. Up to four FM31276/FM31278 devices can reside on the same I^2C bus by assigning a different address to each. Bit 0 is the read/write bit (R/W). R/W = '1' indicates a read operation and R/\overline{W} = '0' indicates a write operation.

Figure 15. Memory Slave Device Address

Figure 16. Companion Slave Device Address

Addressing Overview - Memory

After the FM31276/FM31278 (as receiver) acknowledges the slave address, the master can place the memory address on the bus for a write operation. The address requires two bytes. The complete 15-bit address is latched internally. Each access causes the latched address value to be incremented automatically. The current address is the value that is held in the latch; either a newly written value or the address following the last access. The current address will be held for as long as V_{DD} > V_{TP} or until a new value is written. Reads always use the current address. A random read address can be loaded by beginning a write operation as explained below.

After transmission of each data byte, just prior to the acknowledge, the FM31276/FM31278 increments the internal address latch. This allows the next sequential byte to be accessed with no additional addressing. After the last address (7FFFh) is reached, the address latch will roll over to 0000h. There is no limit to the number of bytes that can be accessed with a single read or write operation.

Addressing Overview - RTC & Companion

The RTC and Processor Companion operate in a similar manner to the memory, except that it uses only one byte of address. Addresses 00h to 18h correspond to special function registers. Attempting to load addresses above 18h is an illegal condition; the FM31276/FM31278 will return a NACK and abort the I^2C transaction.

Data Transfer

After the address bytes have been transmitted, data transfer between the bus master and the FM31276/FM31278 can begin. For a read operation the FM31276/FM31278 will place 8 data bits on the bus then wait for an acknowledge from the master. If the acknowledge occurs, the FM31276/FM31278 will transfer the next sequential byte. If the acknowledge is not sent, the FM31276/FM31278 will end the read operation. For a write operation, the FM31276/FM31278 will accept 8 data bits from the master then send an acknowledge. All data transfer occurs MSB (most significant bit) first.

Memory Operation

The FM31276/FM31278 is designed to operate in a manner very similar to other 1^2C interface memory products. The major differences result from the higher performance write capability of F-RAM technology. These improvements result in some differences between the FM31276/FM31278 and a similar configuration EEPROM during writes. The complete operation for both writes and reads is explained below.

The memory address for FM31276 range from 0x0000 to 0x1FFFF, and for FM31278, they range from 0x0000 to 0x7FFF. Memory functionality is described with respect to FM31278 in the following sections.

Memory Write Operation

All writes begin with a slave address, then a memory address. The bus master indicates a write operation by setting the LSB of the slave address (R/W bit) to a '0'. After addressing, the bus master sends each byte of data to the memory and the memory generates an acknowledge condition. Any number of sequential bytes may be written. If the end of the address range is reached internally, the address counter will wrap from 7FFFh to 0000h.

Unlike other nonvolatile memory technologies, there is no effective write delay with F-RAM. Since the read and write access times of the underlying memory are the same, the user experiences no delay through the bus. The entire memory cycle occurs in less time than a single bus clock. Therefore, any operation including read or write can occur immediately following a write. Acknowledge polling, a technique used with EEPROMs to determine if a write is complete is unnecessary and will always return a ready condition.

Internally, an actual memory write occurs after the 8th data bit is transferred. It will be complete before the acknowledge is sent. Therefore, if the user desires to abort a write without altering the memory contents, this should be done using START or STOP condition prior to the 8th data bit. The FM31276/FM31278 uses no page buffering.

[Figure 17](#page-23-2) and [Figure 18](#page-23-3) below illustrate a single-byte and multiple-byte write cycles.

Memory Read Operation

There are two basic types of read operations. They are current address read and selective address read. In a current address read, the FM31276/FM31278 uses the internal address latch to supply the address. In a selective read, the user performs a procedure to set the address to a specific value.

Current Address & Sequential Read

As mentioned above the FM31276/FM31278 uses an internal latch to supply the address for a read operation. A current address read uses the existing value in the address latch as a starting place for the read operation. The system reads from the address immediately following that of the last operation.

To perform a current address read, the bus master supplies a slave address with the LSB set to a '1'. This indicates that a read operation is requested. After receiving the complete slave address, the FM31276/FM31278 will begin shifting out data from the current address on the next clock. The current address is the value held in the internal address latch.

Beginning with the current address, the bus master can read any number of bytes. Thus, a sequential read is simply a current address read with multiple byte transfers. After each byte the internal address counter will be incremented.

Note Each time the bus master acknowledges a byte, this indicates that the FM31276/FM31278 should read out the next sequential byte.

There are four ways to properly terminate a read operation. Failing to properly terminate the read will most likely create a bus contention as the FM31276/FM31278 attempts to read out additional data onto the bus. The four valid methods are:

- 1. The bus master issues a no-acknowledge in the 9th clock cycle and a STOP in the 10th clock cycle. This is illustrated in the diagrams below. This is preferred.
- 2. The bus master issues a no-acknowledge in the 9th clock cycle and a START in the 10th.
- 3. The bus master issues a STOP in the 9th clock cycle.
- 4. The bus master issues a START in the 9th clock cycle.

If the internal address reaches 7FFFh, it will wrap around to 0000h on the next read cycle. [Figure 19](#page-24-1) and [Figure 20](#page-24-2) below show the proper operation for current address reads.

Figure 19. Current Address Read

Selective (Random) Read

There is a simple technique that allows a user to select a random address location as the starting point for a read operation. This involves using the first three bytes of a write operation to set the internal address followed by subsequent read operations.

To perform a selective read, the bus master sends out the slave address with the LSB (R/\overline{W}) set to '0'. This specifies a write operation. According to the write protocol, the bus master then sends the address bytes that are loaded into the internal address latch. After the FM31276/FM31278 acknowledges the address, the bus master issues a START condition. This simultaneously aborts the write operation and allows the read command to be issued with the slave address LSB set to a '1'. The operation is now a current address read.

Figure 21. Selective (Random) Read

RTC/Companion Write Operation

All RTC and Companion writes operate in a similar manner to memory writes. The distinction is that a different device ID is used and only one byte address is needed instead of two byte address. [Figure 22](#page-25-3) illustrates a single byte write to this device.

Note Although not required, it is recommended that A5-A7 in the register address byte are zeros in order to preserve compatibility with future devices.

RTC/Companion Read Operation

As with writes, a read operation begins with the Slave Address. To perform a register read, the bus master supplies a Slave Address with the LSB set to '1'. This indicates that a read operation is requested. After receiving the complete Slave Address, the FM31276/FM31278 will begin shifting data out from the current register address on the next clock. Auto-increment operates for the special function registers as with the memory address. A current address read for the registers look exactly like the memory except that the device ID is different.

The FM31276/FM31278 contains two separate address registers, one for the memory address and the other for the register address. This allows the contents of one address register to be modified without affecting the current address of the other register. For example, this would allow an interrupted read to the memory while still providing fast access to an RTC register. A subsequent memory read will then continue from the memory address where it previously left off, without requiring the load of a new memory address. However, a write sequence always requires an address to be supplied.

Addressing FRAM Array in the FM31276/FM31278 Family

The FM31276/FM31278 family includes 64-Kbit and 256-Kbit memory densities. The following 2-byte address field is shown for each density.

Maximum Ratings

Exceeding maximum ratings may shorten the useful life of the device. These user guidelines are not tested.

Operating Range

SDA inputs which do not employ a diode to $\mathsf{V}_{\mathsf{DD}}.$

DC Electrical Characteristics

Over the [Operating Range](#page-26-1)

Notes

1. Typical values are at 25 °C, $V_{DD} = V_{DD}(typ)$. Not 100% tested.

2. Full complete operation. Supervisory circuits, RTC, etc operate to lower voltages as specified.

3. The V_{BAK} trickle charger automatically regulates the maximum voltage on this pin for capacitor backup applications.

4. V_{BAK} will source current when trickle charge is enabled (VBC bit = '1'), V_{DD} > V_{BAK}, and V_{BAK} < V_{BAK} max.

DC Electrical Characteristics (continued)

Over the Operating Range

-
-
- **Notes**
5. The minim<u>um</u> V_{DD} to guarantee the level of RST remains a valid V_{OL} level. <u>—</u>
6. Includes RST input detection of external reset condition to trigger driving of RST signal by FM31276/FM31278.

Data Retention and Endurance

Capacitance

Thermal Resistance

AC Test Loads and Waveforms

AC Test Conditions

Notes

7. This parameter is characterized and not 100% tested.

8. The crystal attached to the X1/X2 pins must be rated as 6 pF/12.5 pF.

Supervisor Timing

Over the [Operating Range](#page-26-1)

Parameter	Description	Min	Max	Units
t _{RPU}	$\overline{\text{RST}}$ active (LOW) after V_{DD} > V_{TP}	100	200	ms
t_{RNR} ^[9]	RST response time to V_{DD} < V_{TP} (noise filter)	10	25	μS
$t_{VR}^{[9, 10]}$	V_{DD} power-up ramp rate	50		μ s/V
$t_{\rm VF}$ [9, 10]	V _{DD} power-down ramp rate	100		μ s/V
t_{WDP} [11]	Pulse width of RST for watchdog reset	100	200	ms
t _{WDOG} [11]	Timeout of watchdog	t _{DOG}	$2 \times t_{DOG}$	ms
f_{CNT}	Frequency of event counters	Ω	10	MHz
tosc	RTC Oscillator time to start		2	s

Figure 24. RST Timing

Notes

9. This parameter is characterized and not 100% tested.

10. Slope measured at any point on V_{DD} waveform.

11. t_{DOG} is the programmed time in register in register 0Ah, V $_{\text{DD}}$ > V $_{\text{TP}}$, and t_{RPU} satisfied.

AC Switching Characteristics

Over the [Operating Range](#page-26-1)

Parameter ^[12]									
Cypress Parameter	Alt. Parameter	Description	Min	Max	Min	Max	Min	Max	Unit
f_{SCL}		SCL clock frequency	0	100	0	400	Ω	1000	kHz
t_{SU ; STA		Start condition setup for repeated Start	4.7	$\overline{}$	0.6	—	0.25		μS
t _{HD;STA}		Start condition hold time	4.0	$\overline{}$	0.6	—	0.25		μS
t _{LOW}		Clock LOW period	4.7	—	1.3		0.6		μS
^t HIGH		Clock HIGH period	4.0	—	0.6		0.4		μS
$t_{\text{SU;DAT}}$	t _{SU;DATA}	Data in setup	250	—	100	—	100		ns
^t HD;DAT	^I HD;DATA	Data in hold	0	-	0		Ω		ns
t_{DH}		Data output hold (from SCL $@V_{\text{II}}$)	0		0		Ω		ns
t_R ^[13]	t_{r}	Input rise time	—	1000		300		300	ns
t_F ^[13]	t _f	Input fall time	—	300	—	300	$\overline{}$	100	ns
t _{SU;STO}		STOP condition setup	4	—	0.6		0.25		μS
t _{AA}	t _{VD;DATA}	SCL LOW to SDA Data Out Valid		3		0.9		0.55	μS
t_{BUF}		Bus free before new transmission	4.7	—	1.3		0.5		μS
t_{SP}		Noise suppression time constant on SCL, SDA		50		50		50	ns

Figure 25. Read Bus Timing Diagram

Figure 26. Write Bus Timing Diagram

Notes

12. Test conditions assume a signal transition time of 10 ns or less, timing reference levels of 0.5 × V_{DD}, input pulse levels of 10% to 90% of V_{DD}, and output loading of the specified I_{OL}/I_{OH} and 100 pF load capacitance shown in [Figure 23 on page 28](#page-28-7).

13. This parameter is characterized and not 100% tested.

Ordering Information

All these parts are Pb-free. Contact your local Cypress sales representative for availability of these parts.

Ordering Code Definitions

Package Diagram

Figure 27. 14-pin SOIC (150 Mils) Package Outline, 51-85067

DIMENSIONS IN INCHESIMMJ MIN. MAX. REFERENCE JEDEC MS-012

51-85067 *E

Acronyms Document Conventions

Units of Measure

Document History Page

Document History Page (continued)

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