USB-C Controller + Power Switch with HV Protection

General Description

The RT9752 is a USB Type-C Downstream Facing Port (DFP) controller with back to back power switch which offers 28V tolerance on VOUT pin at off state and 3A rated supply current for VBUS for USB Type-C applications. The RT9752 have Configuration Channel (CC) function to determine when USB device is attached. If the Upstream Facing Port (UFP) is attached using an E-marked cable, the RT9752 also applies power to VBUS and VCONN power to the cable CC pin. The RT9752 also identifies when Type-C audio or debug accessories are attached. Reverse voltage protection is provided to prevent reverse current flowing back to input power supply in a power source swap application.

Features

- USB Type-C Compliant DFP Controller
- Standard USB/1.5A/3A Capability Advertisement on CC
- VBUS Application and Discharge
- VCONN Application to Electronically Marked Cable
- VIN Supply Voltage Range from 4.5V to 5.5V
- Reverse Voltage Protection with Ultra Fast RVP Recovery
- Low On-Resistance of the Power FETs
- Programmable Current Limit Level
- High Reliability of High Voltage Power Switch
- UL Approved-E219878
- Nemko Approved-NO110026

Applications

- Type-C USB Chargers
- PC Products Base on Type-C Host Port

Simplified Application Circuit



Copyright©2022 Richtek Technology Corporation. All rights reserved. RICHTEK is a registered trademark of Richtek Technology Corporation.



Ordering Information

Varaian	Discharge	EN F	unction	Marking	Paakaga Typa	
version	Function	Active High	Active Low	Information	rackaye i ype	
RT9752AGQWF	Yes	V		0B=		
RT9752BGQWF	Yes		V	0A=		
RT9752ANGQWF	No	V		0D=	WQFN-201L 3X4 (FC)	
RT9752BNGQWF	No		V	0C=		

Note :

Richtek products are :

- RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.
- Suitable for use in SnPb or Pb-free soldering processes.

Pin Configuration



WQFN-20TL 3x4 (FC)

Functional Pin Description

Pin No.	Pin Name	Pin Function
1	FAULT#	Fault event indicator. Open-drain logic output that asserts low to indicate current limit or thermal shutdown event due to over temperature, in which the pull-up resistor is recommended from $10k\Omega$ to $100k\Omega$.
2, 3, 4	VIN	Input power for VBUS and VCONN.
5	VCC	Bias voltage for control logic. The bias voltage of VCC should be tied to same voltage source of VIN pin (5V). For avoiding noise disturbance, the supplied bias voltage must be stable, Beside, a RC filter (R = $2.2\Omega/0603$ and C = 1μ F/0603) from bias voltage to VCC pin is necessary which should be placed as close as physically possible to VCC pin.
6	EN	Enable control input pin (Active High). If the voltage of EN pin is risen higher than 1.15V, the RT9752 is ready to work. If the voltage of EN pin falls below its falling 0.65V, the RT9752 is shut down.
7	CHG	Charge logic input for standard USB (2.0/3.0) or Type-C selection.
8	СНС_НІ	High-charge logic input for current sourcing capability selection (1.5A/3A) of Type-C.
9	NC	No internal connection.
10	NC	No internal connection.
11	CC1	Connect to Type-C receptacle CC pin. Suggest to place a equal-value output capacitor 22μ F (25V/1206/X5R) as close to the CC1 and GND pins as possible.
12	GND	Ground pin. Directly solder to the large PGND plane and use thermal vias to connect PGND of other layers for thermal resistor reduction.
13	CC2	Connect to Type-C receptacle CC pin. Suggest to place a equal-value output capacitor 22μ F (25V/1206/X5R) as close to the CC2 and GND pins as possible.
14, 15	OUT	Power switch output VBUS.
16	DEBUG#	Open-drain logic output. It asserts when entry Type-C Debug accessory mode, in which the pull-up resistor is recommended from $10k\Omega$ to $100k\Omega$.
17	AUDIO#	Open-drain logic output. It asserts when entry Type-C Audio accessory mode, in which the pull-up resistor is recommended from $10k\Omega$ to $100k\Omega$.
18	POL#	Polarity open-drain logic output that signals which Type-C CC pin is connected to the CC line, in which the pull-up resistor is recommended from $10k\Omega$ to $100k\Omega$.
19	UFP#	Open-drain logic output. It asserts when entry Type-C UFP mode, in which the pull-up resistor is recommended from $10k\Omega$ to $100k\Omega$.
20	LD_DET#	Load-detect open-drain logic output. It asserts when set to source Type-C 3A current, and VBUS is sourcing over 1.8 A nominal, In which the pull-up resistor is recommended from $10k\Omega$ to $100k\Omega$.



Functional Block Diagram



Operation

Supply Power

The RT9752 has two input power pins; VIN and VCC. VIN is the power source connection to the VOUT, CC1 and CC2 through the power FETs inside chip for VBUS, VCONN with the Type-C charging current. VCC supply provides power to internal circuits in the chip.

Enable and Disable

The RT9752 includes an EN pin for sequence control (independent control). If the voltage of EN pin rises above rising threshold, the device is turned on or enabled. On the contrary, the RT9752 is shut down when the voltage of EN pin falls below its falling threshold.

Under-Voltage Lockout

A voltage-sense circuit monitors the input voltage. When the input voltage VIN is above 4V, UVLO turns on the MOSFET switch if CC pins configured the UFP device attached.

USB Type-C DFP Output Control

The RT9752 is a DFP Type-C port controller with integrated power switch for VCONN and VBUS. The RT9752 output for VBUS or VCONN supply decided by CC1 and CC2 pin configuration defined by USB Type-C standard. Table 1 lists the RT9752 output response to various port attachments.

RT9752	001	000	RT9752 Output			
Type-C Port Configuration		002	VOUT	VCONN Supply		
Nothing Attached	OPEN	OPEN	OPEN	NO		
LIED Connected	Rd	OPEN	VIN	NO		
	OPEN	Rd	VIN	NO		
Doworod Cable without LIED Connected	OPEN	Ra	OPEN	NO		
	Ra	OPEN	OPEN	NO		
Doworod Cable with LIED Connected	Rd	Ra	VIN	CC2		
	Ra	Rd	VIN	CC1		
Debug Accessory Connected	Rd	Rd	VIN	NO		
Audio Accessory Connected	Ra	Ra	OPEN	NO		

Table 1. RT9752 Output with Port Configuration

The Rd and Ra impedance can refer to Universal Serial Bus Type-C Cable and Connector Specification.

Table 2. CC Termination (Rd) Requirements

Rd Implementation	Nominal Value	Max Voltage On Pin
+-20% resistor to GND	5.1kΩ	2.18V

Table 3. CC Termination (Ra) Requirements

Ra Implementation	Minimum Impedance	Maximum Impedance
Ra	800Ω	1. 2k Ω

Current Capability and Over-Current Protection

The RT9752 supports current rating defined by USB Type-C standard with corresponding over-current protection which is configured by the CHG and CHG_HI pins setting. The RT9752 has an internal current limit circuit that protects the device during overload condition or shorting events. After output voltage is successfully powered up and closed to input voltage, the output could draw the current until trip current threshold(I_{TRIP}). When the loading is higher than I_{TRIP} , output voltage will drop and approach to GND. The current limit will change to I_{LIM} level to protect the system. Besides, the current limit is set at I_{LIM} level when output voltage is not established.

The RT9752 further protects device by sequentially switching off and on hiccup cycles if large output voltage drop under current limit condition or the thermal protection engages. The current capability is shown in Table 4.

CHG	CHG_HI	Current Advertisement	Current Limit Threshold
Low	Low /High	Standard USB2.0/3.0	1.7A
High	Low	1.5A	1.7A
High	High	3A	3.4A

Table 4. USB Type-C Current Advertisement and Current Limit Threshold

Reverse-Voltage Protection (RVP) and High Voltage Tolerance

The RVP circuit is integrated to prevent leakage current from output voltage to input voltage. If there is any high voltage plugging in output voltage, the output voltage exceeds the input voltage around 100mV instantly, the reverse voltage protection circuitry will turn internal power FET off immediately to protect the input power supply. The power FET will turn on again when output voltage returns to the same level with input voltage. In addition, to avoid the existing voltage of VBUS higher than 5.25V, the RT9752 provides dynamic gate driver control loop to implement the reverse-voltage protection. Device will always try to regulate the output voltage, lower than intput voltage. If output voltage is higher than input voltage when enabling the device, the power FET will never turn on. The device will always do pre-check before switching on the power MOSFETs. The RT9752 offers 28 V tolerance on VOUT pin and 24V tolerance on CC pins to ensure the device is able to work on a USB PD port.

Over-Voltage Protection (OVP)

The RT9752A also provides the over-voltage protection. When the output voltage exceeds over-voltage threshold (6V, Typ.), the over-voltage protection circuitry will turn off internal power FET immediately to protect the input power supply. The power FET will turn on again when output voltage is lower than voltage threshold (6V, Typ.). Besides, when OVP is triggered, the output voltage is almost the same as input voltage (VOUT – VIN < 100mV). However, OVP behavior is different from RVP (VOUT – VIN > 100mV).

Thermal Shutdown

The RT9752 continuously monitors the operating temperature of the power switch for over-temperature protection. The RT9752 turns off the power switch to prevent the device from damage if the junction temperature rises to approximately 150°C due to over-current or short circuit conditions. The pass element turns on again after the junction temperature cools down to 130°C.

Indicator LD_DET# Response

The load detect function enabled when port configures 3A rated VBUS charging (CHG = CHG_HI = High). LD_DET# pin asserts (active low) when the device monitored that current draw by UFP device exceeds 1.9A (Typ.). Connect LD_DET# pin with a pull-high resistor with 100k Ω to VCC. LD_DET# can be left open or tied to GND when not used.

Indicator FAULT# Response

The FAULT# pin is an open drain output that asserts (active low) when device VOUT current exceeds its programmed value with large output voltage drop after VOUT power

ready or under over-temperature protection. The FAULT# signal remains asserted until the fault condition is removed and the device resumes normal operation. Connect FAULT# with a pull-high resistor with $100k\Omega$ to VCC. FAULT# can be left open or tied to GND when not used.

Indicator AUDIO# / DEBUG# / UFP# / POL# Response

The AUDIO#, DEBUG#, POL# and UFP# pins are all open drain output that asserts (active low) respectively with different CC1 and CC2 pin configured settings. The AUDIO# pin asserts when port configured audio accessory mode (both CC1 and CC2 pin = Ra). Similarly, the DEBUG# pin asserts when port configured debug accessory mode (both CC1 and CC2 pin = Rd). The UFP# pin drives low when port recognizes an UFP device inserted, and the POL# pin asserts if reverse plug orientation UFP# is detected. All of these pins are connected to pull-high resistors with 100k Ω to VCC respectively when used. Tie to GND or leave open when not used.



Absolute Maximum Ratings (Note 1)

Supply Input Voltage, VIN, VCC	–0.3V to 7V
Output Voltage, VOUT	–0.3V to 28V
Output Voltage, CC1, CC2	–0.3V to 24V
Other I/O Voltages	–0.3V to 7V
• Power Dissipation, $P_D \textcircled{0} T_A = 25^{\circ}C$	
WQFN-20TL 3x4 (FC)	1.54W
Package Thermal Resistance (Note 2)	
WQFN-20TL 3x4 (FC), θ _{JA}	64.6°C/W
WQFN-20TL 3x4 (FC), θ _{JC}	8.4°C/W
• Lead Temperature (Soldering, 10 sec.)	260°C
• Junction Temperature	150°C
Storage Temperature Range	$-65^{\circ}C$ to $150^{\circ}C$
• ESD Susceptibility (Note 3)	
HBM (Human Body Model)	2kV
CDM (Charged Device Model)	500V

Recommended Operating Conditions (Note 4)

Supply Input Voltage VIN	4.5V to 5.5V
Supply Input Voltage VCC	4.5V to 5.5V
Junction Temperature Range	$-10^\circ C$ to $105^\circ C$

Electrical Characteristics

(VIN = VCC = 5V, $V_{EN} = V_{CHG} = V_{CHG_{HI}} = VCC$, typical values are referenced to $T_A = T_J = 25^{\circ}C$, Min and Max values are referenced to $T_A = T_J$ from $-10^{\circ}C$ to $105^{\circ}C$, unless other noted)

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Input Power Supply						
UVLO Threshold	VTH_UVLO_VIN	Rising threshold voltage for UVLO		4		V
UVLO Hysteresis	VTH_UVLO_HYS	Hysteresis		100		mV
Supply Current	IVIN_SHDN	Disabled supply current, $V_{EN} = 0V$			1	μA
	IVIN_CCOPEN	Enabled with CC lines open, V _{EN} = VCC			2	μA
	IVIN_ACC	Enabled with accessory mode or dangling e-marked cable on CC lines ,V _{EN} = VCC, CC = Ra/Ra			10	μΑ
	IVIN_UFP1	Enabled with UFP attached V_CHG = 0V, 0V \leq VCCx \leq 1.5V		200	250	μA
	IVIN_UFP2	Enabled with UFP attached V _{CHG} = VIN and V _{CHG_HI} = 0V, 0V \leq VCCx \leq 1.5V		200	250	μA
	IVIN_UFP3	Enabled with UFP attached 0V \leq VCCx \leq 2.45V		200	250	μA

RT9752

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Input Power Supply	VCC					
	IVCC_SHDN	Disabled supply current, V _{EN} = 0V			2	μA
Supply Current	IVCC_CCOPEN	Enabled with CC lines open, $V_{EN} = VCC$			3	μA
	Ivcc_acc	Enabled with accessory mode or dangling e-marked cable on CC lines ,V _{EN} = VCC, CC = Ra/Ra		240	288	μA
	IVCC_UFP1	Enabled with UFP attached V_{CHG} = 0V, 0V \leq VCCx \leq 1.5V		330	396	μA
Supply Current	IVCC_UFP2	Enabled with UFP attached V _{CHG} = V _{IN} and V _{CHG_HI} = 0V, 0V \leq V _{CCx} \leq 1.5V		430	516	μA
	IVCC_UFP3	Enabled with UFP attached, $0V \le VCCx \le 2.45V$		580	696	μΑ
Logic Threshold						
EN/CHG/CHG_HI	Vтн_н	Rising edge		0.93	1.15	V
	V _{TH_L}	Falling edge	0.65	0.88		V
	VTH_HYS	Hysteresis		70		mV
EN	I _{EN}	Input current ,V _{EN} = 0V or 6.5V	-0.5		0.5	μA
	Vol	Output low voltage, IFAULT = 1mA			350	mV
FAULT#	I _{OFF}	Off-state leakage, V _{FAULT} = VCC			1	μA
	Vol	Output low voltage, IFAULT = 1mA	-		350	mV
	IOFF	Off-state leakage, V _{FAULT} = VCC			1	μA
LD_DET#	Ітн	VOUT sourcing, rising threshold current for load detect, $T_J = 25^{\circ}C$	1.7	1.9	2.1	А
	Ітн_нүs	Hysteresis, T _J = 25°C		150		mA
UFP#/POL#	V _{OL}	Output low voltage, I _{FAULT} = 1mA			250	mV
AUDIO#/DEBUG#	IOFF	Off-state leakage, V _{FAULT} = VCC			1	μA
VOUT – VBUS Powe	er Switch					
On-Resistance	R _{DS(ON)}	T _J = 25°C, I _{OUT} = 3A		34	37	mΩ
Reverse Leakage Current	IREV	V_{OUT} = 6.5V, VIN = V_{EN} = 0V, $-40^\circ C \leq T_J \leq 85^\circ C, \ I_{REV}$ is current out of VIN pin		0	3	μΑ
Over Current Trip		V _{CHG} = 0V or V _{CC} and V _{CHG_HI} = 0V T _J = 25°C , SR = 100A/s	2	3	3.4	А
Threshold		$V_{CHG} = V_{CC}$ and $V_{CHG_HI} = V_{CC}$ T _J = 25°C , SR = 100A/s	3.6	5	5.8	А

RT9752



Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Current Limit		V_{CHG} = 0V or VCC and V_{CHG} HI = 0V TJ = 25°C, V_{OUT} < 1V	1.54	1.7	1.86	А
Level	ILIM	V_{CHG} = VCC and V_{CHG_HI} = VCC T _J = 25°C, V_{OUT} < 1V	3.09	3.4	3.71	А
Discharge Resistance	RDISCHG	V _{OUT} = 4V, UFP removed from CC lines, time < tw_DCHG	400	500	600	Ω
Bleed Discharge Resistance	RDISCHG_BLEED	VOUT = 4V, No UFP on CC lines, time > tw_DCHG	100	150	250	kΩ
Rise Time	t _R	VIN1 = 5V, C _L = 1 μ F, R _L = 100 Ω		2.5		ms
Fall Time	tF	(measured from 10% to 90% of final value)		0.55		ms
CC1/CC2 - VCONN F	Power Switch /Cu	Irrent Limit				
On-Resistance	R _{DS(ON)}	T _J = 25°C, I _{OUT} = 250mA		570	1000	mΩ
Over Current Trip Threshold	I _{TRIP}	T _J = 25°C, SR = 100A/s	330	420	510	mA
Current Limit	I _{LIM}	$T_J = 25^{\circ}C, V_{CCx} < 1V$	300	355	410	mA
Rise Time	t _R	VIN1 = 5V, C _L = 1 μ F, R _L = 100 Ω		45		μs
Fall Time	tF	value)		220		μs
CC1/CC2 – Connect	Management – D	Dangling E-Marked Cable Mode				
Coursing Current	ISRC	Current on the pass through CC Line $0V \le VCCx \le 1.5V$	74	93	112	μA
Sourcing Current		Current on the Ra CC line 0V \leq VCCx \leq 1.5V	74	93	112	μA
CC1/CC2 – Connect	Management – A	Accessory Mode				
Councilor Current		CCx Sourcing current (CC2- Audio, CC1-Debug), V _{CHG} = V _{CHG_HI} = 0V		80		μA
Sourcing Current	ISRC	CCx Sourcing current (CC1- Audio, CC2-Debug), V _{CHG} = V _{CHG_HI} = 0V		80		μA
CC1/CC2 – Connect	Management – U	IFP Mode				
		Current with VIN in UVLO 0V \leq VCCx \leq 1.5V VIN1 $<$ VTH_UVLO_VIN	64	80	96	μA
Sourcing Current		V_{CHG} = 0V and V_{CHG_HI} = 0V, 0V \leq VCCx \leq 1.5V	70	80	90	μA
Sourcing Current	ISRC	V _{CHG} = VCC and V _{CHG_HI} = 0V, 0V \leq VCCx \leq 1.5V	158	180	202	μA
		V_{CHG} = VCC and $V_{CHG_{HI}}$ = VCC, 0V \leq VCCx \leq 2.45V	290	330	370	μA

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Over-Temperature S	hut Down					
Thermal Shutdown Threshold	T _{SD}	Rising threshold temperature for device shutdown		150		°C
Thermal Shutdown Hysteresis	ΔT_{SD}			20		°C
Asserting / De-Asserting Timing						
FAULT#	tDEGA	Asserting deglitch due to over-current and VOUT $< 1V$ (Note 5)	1.6	2.5	3.4	ms
	tDEGAD	De-asserting deglitch	5.5	8.2	10.7	ms
	t _{DEGA}	Asserting deglitch	45	65	85	ms
LD_DET#	tDEGAD	De-asserting deglitch	1.5	2.15	2.9	s
UFP# / POL#	t _{DEGA}	Asserting deglitch	100	150	200	ms
AUDIO#/DEBUG#	tDEGAD	De-asserting deglitch	7.9	12.5	17.7	ms

- Note 1. Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.
- **Note 2.** θ_{JA} is measured under natural convection (still air) at $T_A = 25^{\circ}C$ with the component mounted on a high effectivethermal-conductivity four-layer test board on a JEDEC 51-7 thermal measurement standard.
- Note 3. Devices are ESD sensitive. Handling precautions are recommended.
- Note 4. The device is not guaranteed to function outside its operating conditions.
- Note 5. Function is enabled after VOUT power is ready.



Typical Application Circuit



Typical Operating Characteristics







VBUS Power Off from EN

























Copyright ©2022 Richtek Technology Corporation. All rights reserved. RICHTEK is a registered trademark of Richtek Technology Corporation.

14

Application Information

The RT9752 is a Type-C DFP load switch and controller that supports all Type-C DFP required functions and further integrated HV protection in internal power switches for VCONN and VBUS. The RT9752 can only be used in a USB2.0 or USB3.0 port implementation, but BC1.2, because of no interacting with USB D+/D-. As UFP is attached detecting by CC pins, RT9752 allows power delivering to VBUS. Conversely, as UFP is detached, power is removed by RT9752. Besides, the RT9752 applies three current advertisements for Type-C through CHG and CHG HI pin setting, and provides over-current limit protection for device. Apart from delivering power to UFP, the RT9752 also supports Audio and Debug accessory modes.

Input and Output Filter Capacitor

Although the RT9752 provides RVP function to protect any abnormal electrical over-stress happening, there are risks of input voltage overshoot and output undershoot to damage the device.

For the input voltage overshoot, as VBUS is suddenly short circuit while RT9752 is turn-off, the stored energy in input trace inductor is going to release to VIN terminal and cause VIN abnormal oscillation. For avoiding VIN abnormal rising over 7V (absolute maximum rating), a large input capacitor is required to reduce voltage overshoot during output short circuit. Hence, it's strongly suggested to put a 10μ F to 22μ F low-ESR ceramic capacitor as input filter capacitor. Moreover, for eliminating local noise decoupling, a 0.1µF or greater ceramic bypass capacitor connected between VIN to GND is recommended and place it as close to the VIN terminal as possible.

The output undershoot is caused by output trace inductance after short circuit happening at VBUS. As output short circuit occurring, the RT9752 is turned off to reduce output current and finally the stored energy in trace inductance of output drives the output voltage down to negative level. That negative voltage on output would cause unexpected circuit behavior or damage. However, for preventing any negative voltage influences, the output capacitor is required, which is sufficient to control voltage undershoot. According to USB standard, 120µF minimum

output capacitance is required to meet the minimum drop voltage of VBUS (330mV), but the maximum output capacitance is constrained from the applications. Too much output capacitor induces large inrush current during power on period and large VIN spike as output short circuit happens. Thus, the recommended output capacitance is from 120μ F to 220μ F.

Thermal Considerations

The junction temperature should never exceed the absolute maximum junction temperature $T_{J(MAX)}$, listed under Absolute Maximum Ratings, to avoid permanent damage to the device. The maximum allowable power dissipation depends on the thermal resistance of the IC package, the PCB layout, the rate of surrounding airflow, and the difference between the junction and ambient temperatures. The maximum power dissipation can be calculated using the following formula :

$\mathsf{P}_{\mathsf{D}(\mathsf{MAX})} = (\mathsf{T}_{\mathsf{J}(\mathsf{MAX})} - \mathsf{T}_{\mathsf{A}}) / \theta_{\mathsf{JA}}$

where $T_{J(MAX)}$ is the maximum junction temperature, T_A is the ambient temperature, and θ_{JA} is the junction-to-ambient thermal resistance.

For continuous operation, the maximum operating junction temperature indicated under Recommended Operating Conditions is 125°C. The junction-to-ambient thermal resistance, θ_{JA} , is highly package dependent. For a WQFN-20TL 3x4 (FC) package, the thermal resistance, θ_{JA} , is 64.6°C/W on a standard JEDEC 51-7 high effectivethermal-conductivity four-layer test board. The maximum power dissipation at $T_A = 25^{\circ}C$ can be calculated as below :

 $P_{D(MAX)} = (125^{\circ}C - 25^{\circ}C) / (64.6^{\circ}C/W) = 1.54W$ for a WQFN-20TL 3x4 (FC) package.

The maximum power dissipation depends on the operating ambient temperature for the fixed $T_{J(MAX)}$ and the thermal resistance, θ_{JA} . The derating curves in Figure 1 allows the designer to see the effect of rising ambient temperature on the maximum power dissipation.

Copyright ©2022 Richtek Technology Corporation. All rights reserved. RICHTEK is a registered trademark of Richtek Technology Corporation.

RT9752





Figure 1. Derating Curve of Maximum Power Dissipation

Layout consideration

In order to meet the voltage drop, droop, and EMI requirements, careful PCB layout is necessary. The following guidelines must be followed :

Locate the ceramic bypass capacitors as close as possible to the VIN pins of the RT9752.

- Place a ground plane under all circuitry to lower both resistance and inductance, and improve DC and transient performance (Use a separate ground and power plans if possible).
- Avoid vias as much as possible. If vias are necessary, make them as large as feasible.
- Place cuts in the ground plane between ports to help reduce the coupling of transients between ports.



Figure 2. PCB Layout Guide

Outline Dimension



Symbol	Dimensions In Millimeters		Dimensions In Inches		
	Min	Max	Min	Max	
A	0.700	0.800	0.028	0.031	
A1	0.000	0.050	0.000	0.002	Tolerance
A3	0.175	0.250	0.007	0.010	±0.050

W-Type 20TL QFN 3x4 (FC) Package



Footprint Information



Package	Number of Pin	Tolerance
V/W/U/XQFN3x4-20T(FC)	20	±0.05

Richtek Technology Corporation

14F, No. 8, Tai Yuen 1st Street, Chupei City Hsinchu, Taiwan, R.O.C. Tel: (8863)5526789

RICHTEK

Richtek products are sold by description only. Richtek reserves the right to change the circuitry and/or specifications without notice at any time. Customers should obtain the latest relevant information and data sheets before placing orders and should verify that such information is current and complete. Richtek cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Richtek product. Information furnished by Richtek is believed to be accurate and reliable. However, no responsibility is assumed by Richtek or its subsidiaries for its use; nor for any infringements of patents or other rights of third parties which may result from its use. No license is granted by implication or otherwise under any patent or patent rights of Richtek or its subsidiaries.

18