

## Description

The DZDH0401DW is intended to drive a P-channel enhancement MOSFET configured as an ideal diode. The device operates as a differential amplifier and PMOS controller to minimize forward current losses when  $V_{IN} > V_{OUT}$  and provide high isolation when  $V_{IN} < V_{OUT}$ . The circuit compares the voltage between IN and OUT. If the differential is greater than  $\sim 34\text{mV}$  (typ.)  $V_{BIAS}$  will fall and the PMOS will turn on. If the differential is less than  $\sim 34\text{mV}$   $V_{BIAS}$  will rise and the PMOS will turn off, isolating IN from OUT.

## Applications

- High Side Gate Driving PMOS
- High Side Disconnect Switch
- Battery Discharge Protection
- Emergency Lighting
- Active OR'ing Redundant Power Supplies

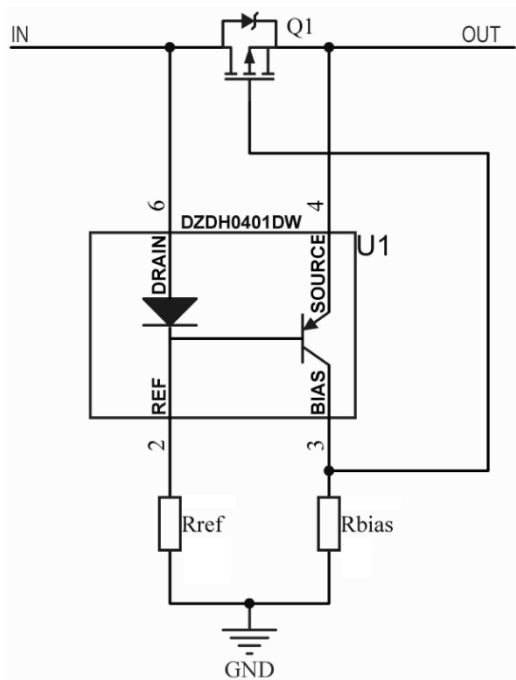
## Features

- Max Input Voltage: 40V
- Peak Bias Current: -300mA
- Max Reverse Voltage Protection: 50V
- **Totally Lead-Free & Fully RoHS Compliant (Notes 1 & 2)**
- **Halogen and Antimony Free. "Green" Device (Note 3)**
- **For automotive applications requiring specific change control (i.e. parts qualified to AEC-Q100/101/200, PPAP capable, and manufactured in IATF 16949 certified facilities), please [contact us](mailto:contact@diodes.com) or your local Diodes representative.**

<https://www.diodes.com/quality/product-definitions/>

## Mechanical Data

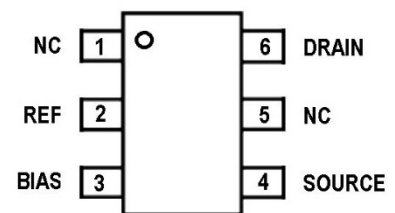
- Case: SOT363
- Case Material: Molded Plastic, "Green" Molding Compound. UL Flammability Classification Rating 94V-0
- Moisture Sensitivity: Level 1 per J-STD-020
- Terminals: Finish—Matte Tin Finish. Solderable per MIL-STD-202, Method 208
- Weight: 0.006 grams (Approximate)



Typical Configuration



SOT363 Top View



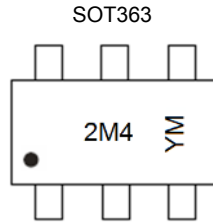
Pinout

## Ordering Information (Note 4)

Part Number	Marking	Reel Size (inches)	Tape Width (mm)	Quantity per Reel
DZDH0401DW-7	2M4	7	8	3,000

- Notes:
1. No purposely added lead. Fully EU Directive 2002/95/EC (RoHS), 2011/65/EU (RoHS 2) & 2015/863/EU (RoHS 3) compliant.
  2. See <https://www.diodes.com/quality/lead-free/> for more information about Diodes Incorporated's definitions of Halogen- and Antimony-free, "Green" and Lead-free.
  3. Halogen- and Antimony-free "Green" products are defined as those which contain <900ppm bromine, <900ppm chlorine (<1500ppm total Br + Cl) and <1000ppm antimony compounds.
  4. For packaging details, go to our website at <https://www.diodes.com/design/support/packaging/diodes-packaging/>.

## Marking Information



2M4 = Product Type Marking Code  
 YM = Date Code Marking  
 Y = Year (ex: 1 = 2021)  
 M = Month (ex: 3 = March)

### Date Code Key

Year	2020	2021	2022	2023	2024	2025	2026	2027	2028	2029	2030	2031
Code	H	I	J	K	L	M	N	O	P	R	S	T

Month	Jan	Feb	Mar	Apr	May	Jun	Jul	Aug	Sep	Oct	Nov	Dec
Code	1	2	3	4	5	6	7	8	9	O	N	D

## Absolute Maximum Ratings (@ T<sub>A</sub> = +25°C unless otherwise specified.)

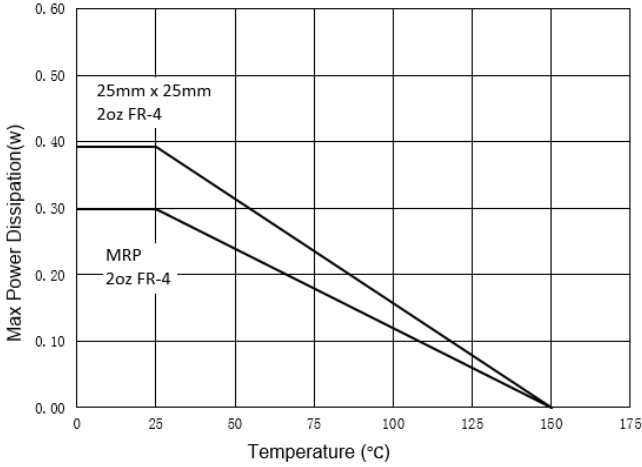
Characteristic	Symbol	Value	Unit
DRAIN BIAS Voltage	V <sub>DRAIN-BIAS</sub>	40	V
SOURCE DRAIN Voltage	V <sub>SOURCE-DRAIN</sub>	50	V
BIAS Current	I <sub>BIAS</sub>	-300	mA
DRAIN Current	I <sub>DRAIN</sub>	300	mA

## Thermal Characteristics – Total Device (@ T<sub>A</sub> = +25°C unless otherwise specified.)

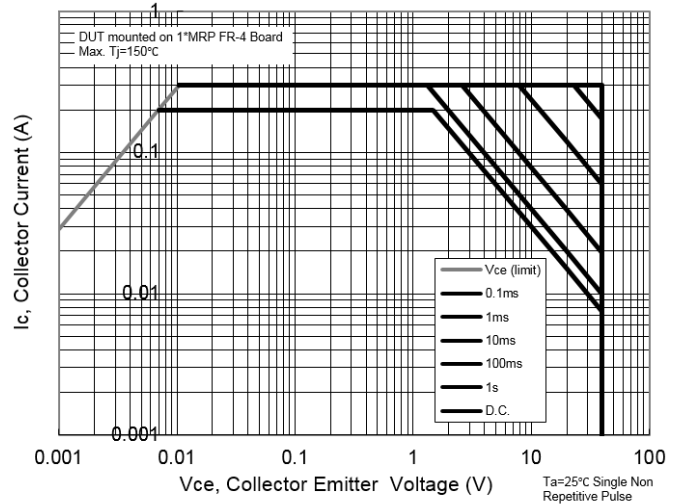
Characteristic	Symbol	Value	Unit
Power Dissipation (Note 5)	P <sub>D</sub>	300	mW
Thermal Resistance, Junction to Ambient (Note 5)	R <sub>θJA</sub>	424	°C/W
Thermal Resistance, Junction to Case (Note 5)	R <sub>θJC</sub>	111	°C/W
Operating and Storage Temperature Range	T <sub>J</sub> , T <sub>STG</sub>	-65 to +150	°C

Note: 5. For a device mounted on minimum recommended pad layout with 1oz copper that is on a single-sided 1.6mm FR4 PCB; the device is measured under still air conditions whilst operating in a steady-state.

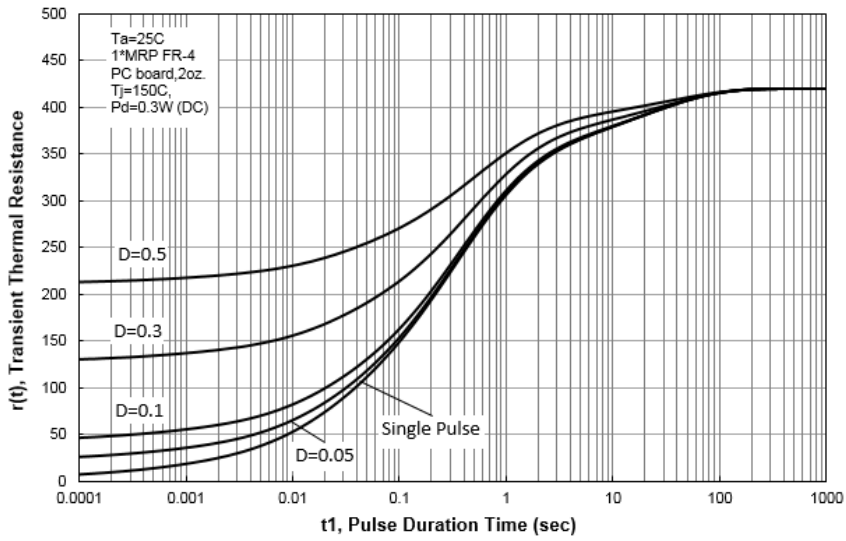
**Thermal Characteristics – Total Device**



**Fig.1 Derating Curve**



**Fig.2 Safe Operation Area**

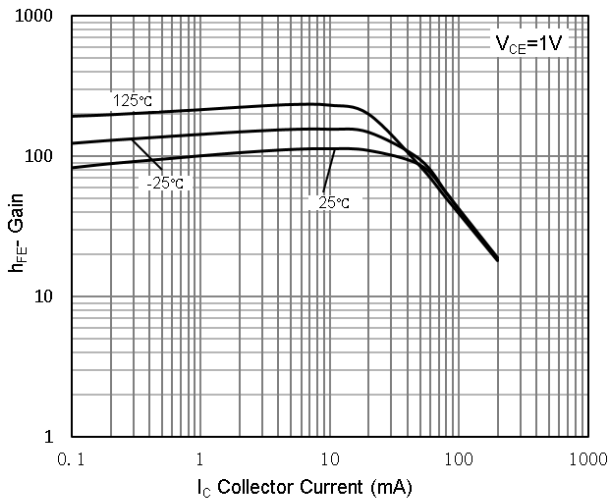


**Fig.3 Transient Thermal Resistance**

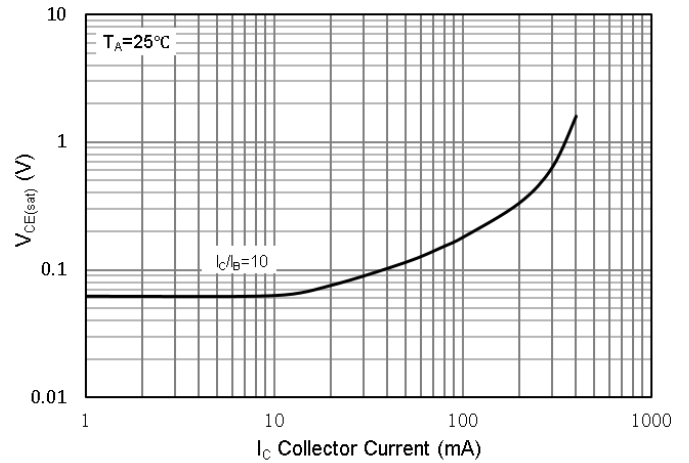
**Electrical Characteristics** (@  $T_A = +25^\circ\text{C}$  unless otherwise specified)

Characteristic	Symbol	Min	Typ	Max	Unit	Test Condition
DRAIN - BIAS Voltage	$V_{\text{DRAIN-BIAS}}$	40	78	—	V	$I_{\text{DRAIN}}=100\mu\text{A}$
SOURCE - DRAIN Voltage	$V_{\text{SOURCE-DRAIN}}$	50	84	—	V	$I_{\text{SOURCE}}=100\mu\text{A}$
DRAIN - REF Voltage	$V_{\text{DRAIN-REF}}$	—	588	—	mV	$I_{\text{DRAIN}}=100\mu\text{A}$
SOURCE Current	$I_{\text{SOURCE}}$	—	11.6	—	$\mu\text{A}$	$V_{\text{SOURCE-REF}}=0.56\text{V}$
REF-SOURCE Voltage	$V_{\text{REF-SOURCE}}$	—	-554	—	mV	$I_{\text{REF}}=-10\mu\text{A}$
Turn-Off Differential Voltage	$V_T$	5	34	80	mV	$I_{\text{DRAIN}}=100\mu\text{A}; I_{\text{SOURCE}}=10\mu\text{A}$
REF-SOURCE Voltage ( $V_{\text{BIAS low}}$ )	$V_{\text{REF-SOURCE}}$	-250	-472	—	mV	$V_{\text{BIAS-SOURCE}}=-5\text{V}; I_{\text{BIAS}}=-1\mu\text{A}$
		-300	-541	—	mV	$V_{\text{BIAS-SOURCE}}=-5\text{V}; I_{\text{BIAS}}=-10\mu\text{A}$
REF-SOURCE Voltage ( $V_{\text{BIAS high}}$ )	$V_{\text{REF-SOURCE}}$	—	-601	-800	mV	$V_{\text{BIAS-SOURCE}}=-0.5\text{V}; I_{\text{BIAS}}=-100\mu\text{A}$
		—	-663	-850	mV	$V_{\text{BIAS-SOURCE}}=-0.5\text{V}; I_{\text{BIAS}}=-1\text{mA}$

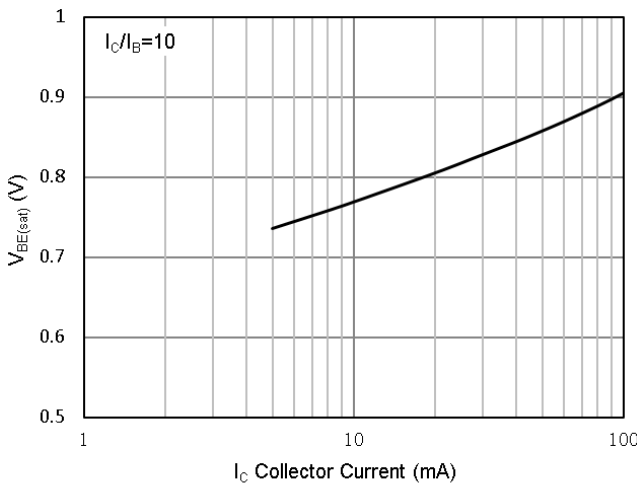
**Typical Electrical Characteristics** (@  $T_A = +25^\circ\text{C}$  unless otherwise specified.)



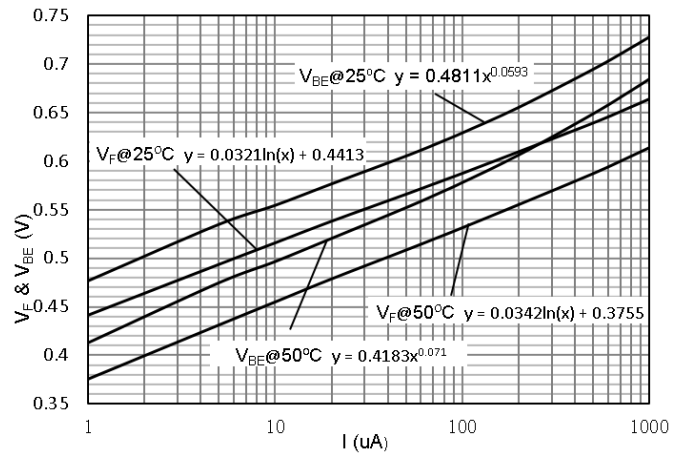
**Fig.4  $h_{FE}$  vs  $I_C$**



**Fig.5  $V_{CE(sat)}$  vs  $I_C$**

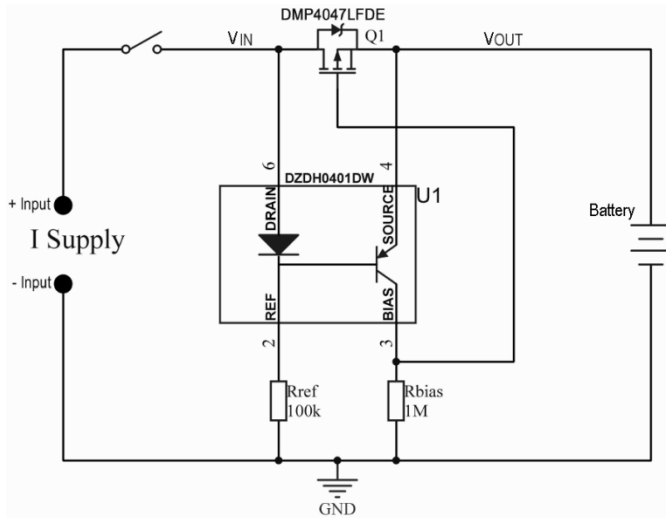


**Fig.6  $V_{BE(sat)}$  vs  $I_C$**



**Fig.7  $I$  vs  $V_F$  &  $V_{BE}$**

## Typical Application Circuit/ Pin Out Details/ Functional Description



**Typical Application Circuit**

Pin	Name	Function
1	NC	No internal connection
2	REF	Reference current to set $V_F$
3	BIAS	Reference current to set $V_{BE}$ and control Gate
4	SOURCE	$V_{OUT}$ sense voltage
5	NC	No internal connection
6	DRAIN	$V_{IN}$ sense voltage
n/a	Rref	This resistor sets the turn off speed of the FET. The lower the resistance, the more base drive to the transistor, the faster the transistor shorts out the gate to turn off the FET.
n/a	Rbias	This resistor sets the turn on speed of the FET. When the ideal diode circuit is turning on the PNP is held off by the diode and FET voltage drops. It is Rbias that pulls the gate low and turns on the FET.
n/a	Rbias : Rref	Ideal diode Turn-Off threshold voltage $V_T \propto Rbias / Rref$
n/a	Vref	Voltage across Rref

### Functional Description (Refer to typical application circuit above)

**Supply Connect:**

As a +Input is applied, the body drain diode of Q1 becomes forward biased. U1 diode holds U1 transistor base at  $V_{IN} - V_F$ , and so  $V_{BE}$  is too low to turn on U1 transistor. As Q1 gate capacitance charges through Rbias, Q1 turns on and  $R_{DS}$  decreases causing  $V_{DS}$  to decrease and  $V_{BE}$  to increase until U1 transistor starts to conduct. This process continues until Q1  $R_{DS}$  reaches its minimum value and U1 transistor  $V_{BE}$  cannot increase and  $I_C$  reaches its maximum.  $V_{GS}$  should be high enough at this point to ensure linear operation.

Rref and Rbias set the currents through U1 diode and U1 collector respectively so that  $V_{F(DIODE)}$  is greater than  $V_{BE(on)}$ .

**Supply Disconnect:**

As the +Input is removed,  $V_{DS} < V_T$ , Q1 is on and  $V_{IN} = V_{OUT}$ , causing  $V_{REF}$  to fall and U1  $V_{BE} > V_{BE(on)}$  so U1 transistor discharges Q1 gate capacitance and Q1 turns off causing  $V_{IN}$  to fall to 0V.

**Quiescent Current and Isolation:**

With a battery connected at Supply Out, there are two leakage paths back to the Supply In. One is straight through Q1 and the other is through U1 emitter-anode. The high reverse breakdown voltage of U1 diode provides a high isolation path. The Rref & Rbias currents bias U1 transistor on which keeps Q1 off. These resistors' values are chosen to minimize quiescent current operation of the circuit.

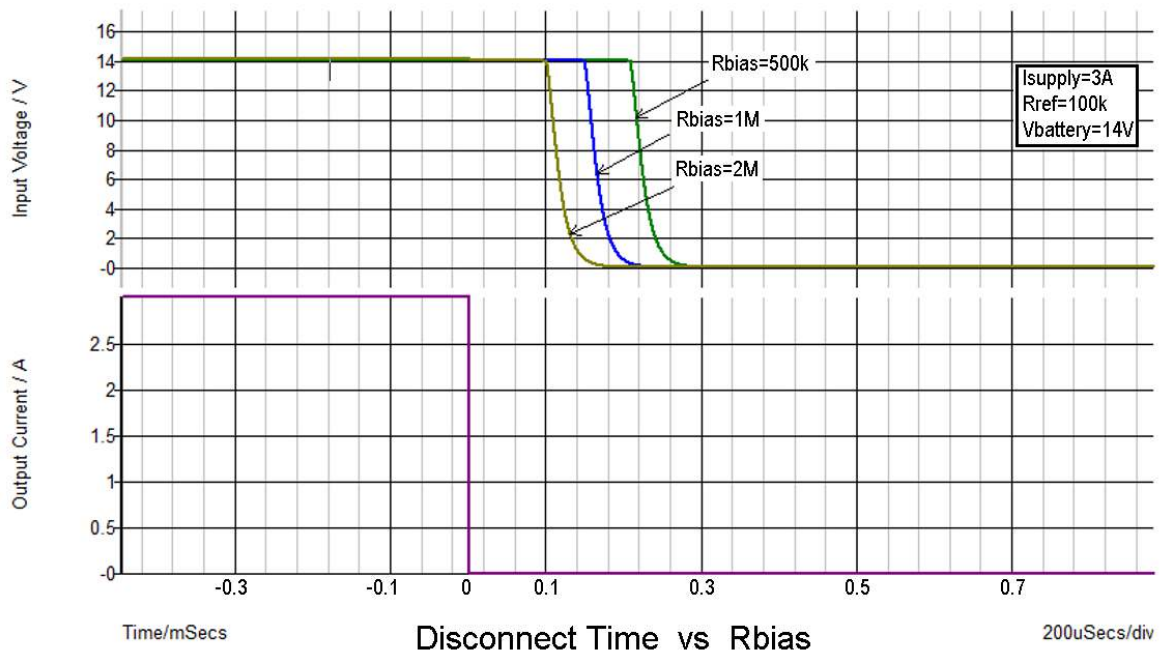
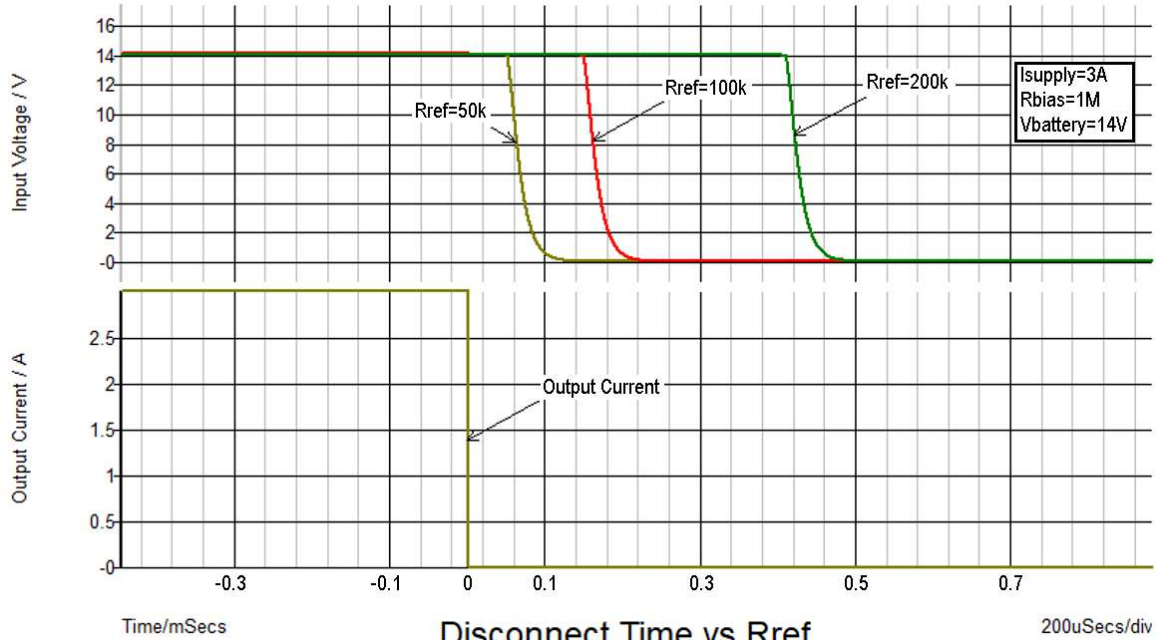
Typical Charging Conditions. (Ta=25°C Vbatt=14V switch closed, Isupply=3A)			
Parameter	Symbol	Typ	Unit
Input Voltage	$V_{IN}$	14.1	V
Input current	$I_{IN}$	3	A
Output Voltage	$V_{OUT}$	14	V
Output Current	$I_{OUT}$	3.0	A
Diode Forward Voltage	$V_F$	0.6	V
Diode forward current	$I_F$	135	uA
Reference Voltage	$V_{REF}$	13.4	V
Reference Current	$I_{REF}$	136.6	uA
Base Current	$I_B$	1.6	uA
Emitter Current	$I_E$	12.1	uA
Bias Voltage	$V_{BIAS}$	10.5	V
Collector Current	$I_C$	10.5	uA
Operating Current	$I_{CC}$	147	uA

Typical Non-Charging Conditions. (Ta=25°C Vbatt=14V switch open, Isupply=3A)			
Parameter	Symbol	Typ	Unit
Input Voltage	$V_{IN}$	-	uV
Input current	$I_{IN}$	--	A
Output Voltage	$V_{OUT}$	14	V
Output Current	$I_{OUT}$	--	A
Diode Forward Voltage	$V_F$	--	V
Diode forward current	$I_F$	0	uA
Reference Voltage	$V_{REF}$	13.3	V
Reference Current	$I_{REF}$	133	uA
Base Current	$I_B$	133	uA
Emitter Current	$I_E$	145	uA
Bias Voltage	$V_{BIAS}$	13.94	V
Bias Current	$I_{BIAS}$	13.94	uA
Operating Current	$I_{CC}$	147	uA

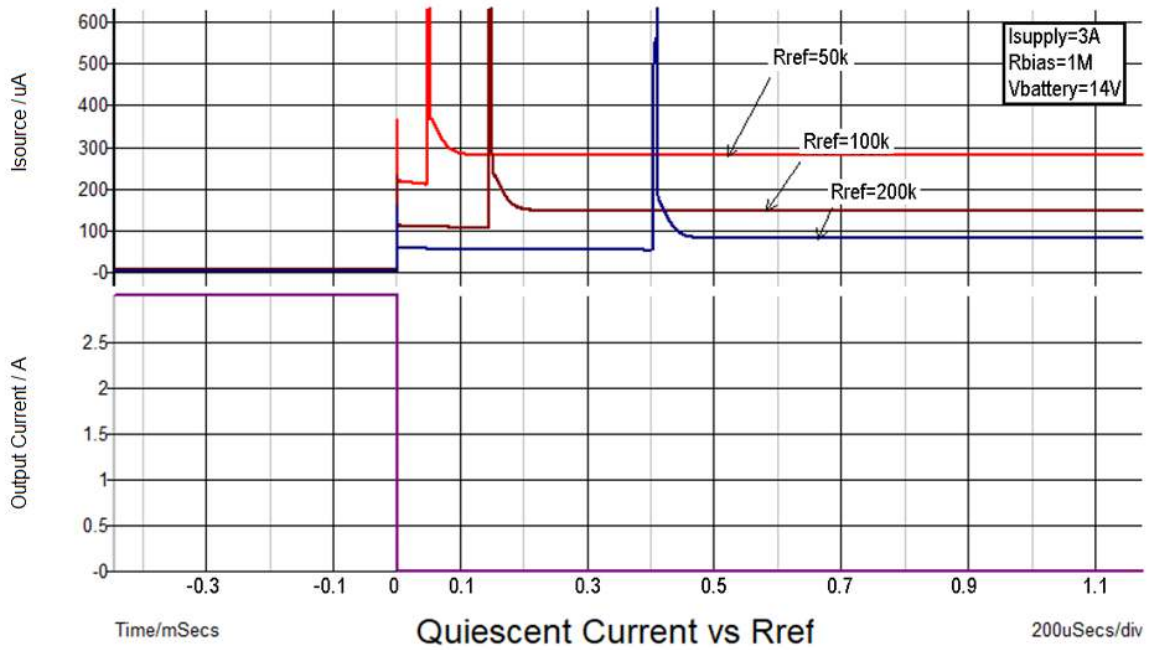
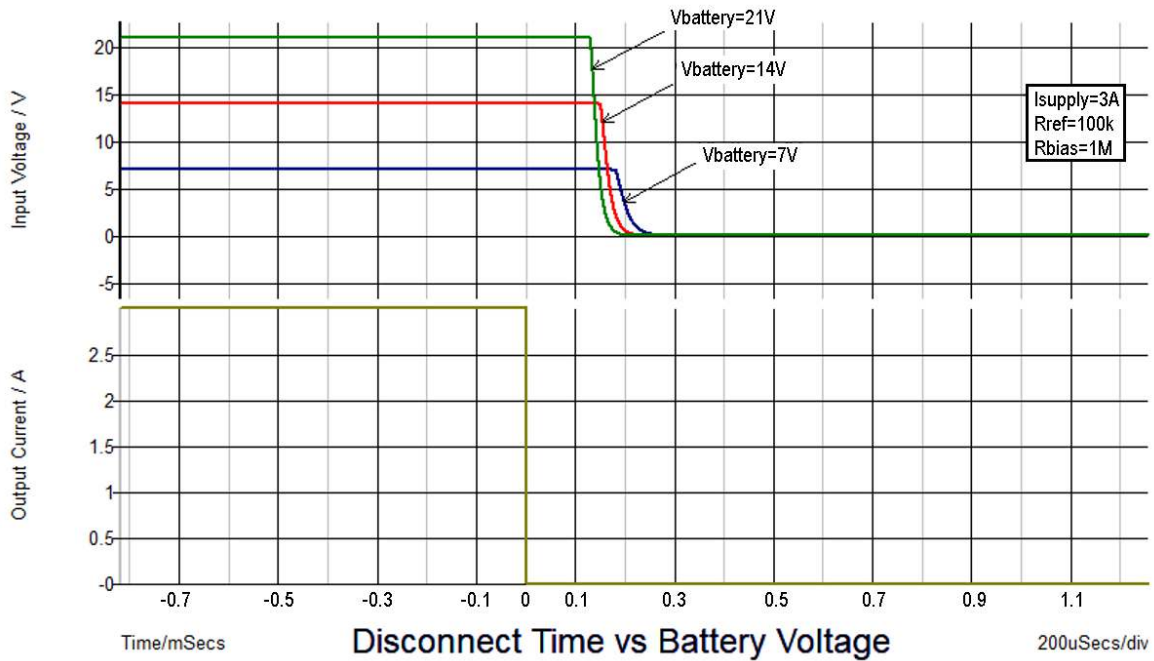
**Typical Application Circuit/ Pin Out Details/ Functional Description** (continued)

**Timing**

Switching speed is affected by PMOS characteristics, Rbias, Rref, and operating voltage. Using the typical application circuit, we can see how modifying values can affect the timing in the simulations below



**Typical Application Circuit/ Pin Out Details/ Functional Description** (continued)



**Typical Application Circuit/ Pin Out Details/ Functional Description** (continued)

**Ideal Diode Power Saving**

The typical voltage drop across a standard diode rectifier means higher power dissipation and more heat to manage. This both wastes power and significantly drops the potential on low voltage rails.

Example:

A diode rectifier with a typical forward voltage  $V_F = 0.55V$  carrying 3A current would dissipate 1.65W ( $I \times V_F$ ).

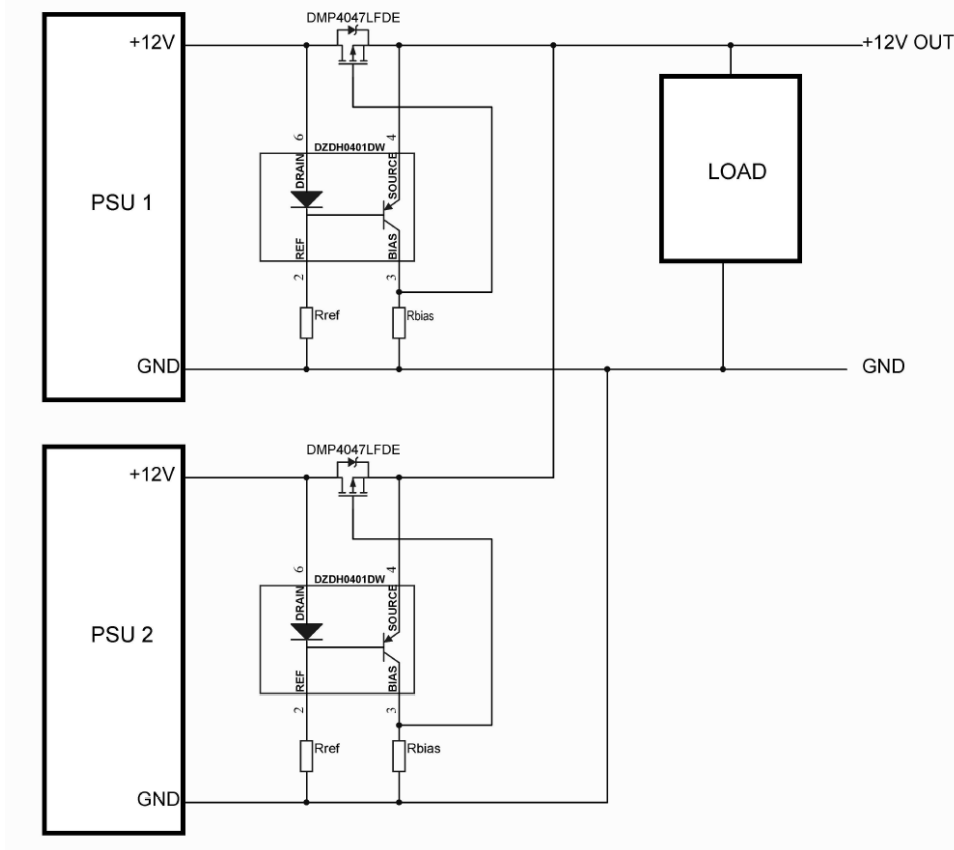
Whereas with P-MOSFET such as the DMP4047LFDE that has an  $R_{DS(on)}$  of 33mΩ, the power dissipation reduces to only 0.29W ( $I^2 \times R$ ).

Hence, very low  $R_{DS(on)}$  Power MOSFETs can replace the standard rectifiers and the DZDH0401DW controls the MOSFET as an ideal diode.

**N+1 redundancy OR'ing controller**

Critical systems require a fault-tolerant power supply that can be achieved by paralleling two or more PSUs into an (N+1) redundancy configuration.

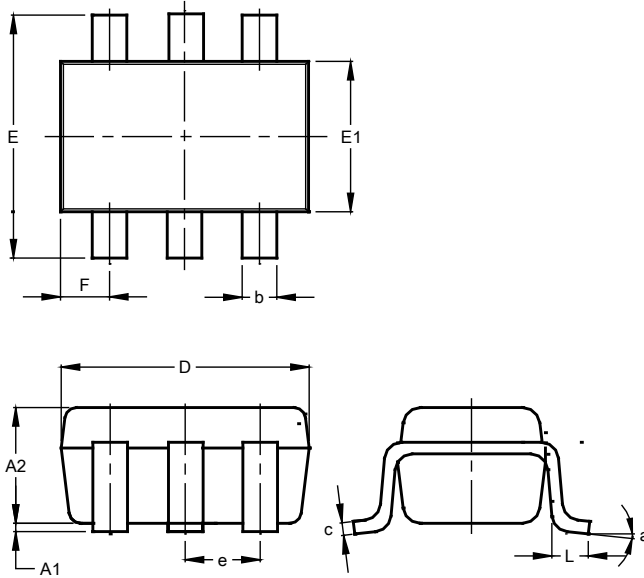
During normal operation, usually all PSUs equally share the load for maximum reliability. If one of the PSUs is unplugged or fails, then the other PSUs fully support the load. To avoid the faulty PSU from affecting the common bus, an OR'ing rectifier blocks the reverse current flow into the faulty PSU. Likewise during hot-swapping, the OR'ing rectifiers isolate a PSU's discharged output capacitors from the common bus.





**Package Outline Dimensions**

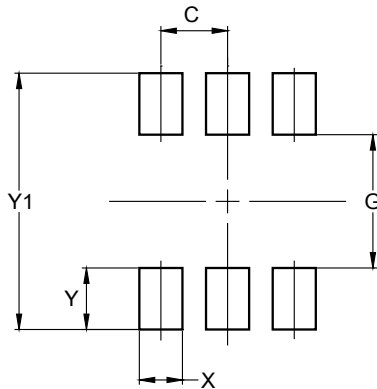
Please see <http://www.diodes.com/package-outlines.html> for the latest version.



SOT363			
Dim	Min	Max	Typ
A1	0.00	0.10	0.05
A2	0.90	1.00	1.00
b	0.10	0.30	0.25
c	0.10	0.22	0.11
D	1.80	2.20	2.15
E	2.00	2.20	2.10
E1	1.15	1.35	1.30
e	0.650 BSC		
F	0.40	0.45	0.425
L	0.25	0.40	0.30
a	0°	8°	--
<b>All Dimensions in mm</b>			

**Suggested Pad Layout**

Please see <http://www.diodes.com/package-outlines.html> for the latest version.



Dimensions	Value (in mm)
C	0.650
G	1.300
X	0.420
Y	0.600
Y1	2.500

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