

# FMS6146 Low-Cost Six-Channel 4th-Order Standard Definition Video Filter Driver

### **Features**

- Six 4th-order 8MHz (SD) filters
- Drives single, AC- or DC-coupled, video loads ( $2V_{pp}$ , 150 $\Omega$ )
- Drives dual, AC- or DC-coupled, video loads (2V<sub>pp</sub>, 75Ω)
- Transparent input clamping
- AC- or DC-coupled inputs
- AC- or DC-coupled outputs
- DC-coupled outputs eliminate AC-coupling capacitors
- 5V only
- Robust 8kV ESD protection
- Lead-free TSSOP-14 package

## Applications

- Cable set-top boxes
- Satellite set-top boxes
- DVD players
- HDTV
- Personal Video Recorders (PVR)
- Video On Demand (VOD)

# Description

The FMS6146 Low-Cost Video Filter (LCVF) is intended to replace passive LC filters and drivers with a low-cost integrated device. Six 4th-order filters provide improved image quality compared to typical 2nd or 3rd-order passive solutions.

The FMS6146 may be directly driven by a DC-coupled DAC output or an AC-coupled signal. Internal diode clamps and bias circuitry may be used if AC-coupled inputs are required (see *Applications* section for details).

The outputs can drive AC- or DC-coupled single  $(150\Omega)$  or dual  $(75\Omega)$  loads. DC coupling the outputs removes the need for output coupling capacitors. The input DC levels are offset approximately +280mV at the output (see *Applications* section for details).



## **Ordering Information**

Part Number	Package	Operating Temperture Range	Packaging Method		
FMS6146MTC14	TSSOP-14	-40°C to +85°C	Tube		
FMS6146MTC14X	TSSOP-14	-40°C to +85°C	Tape and Reel		
All pagkagag arg load free par JEDEC: J STD 020P standard					

All packages are lead free per JEDEC: J-STD-020B standard.

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## **Pin Assignments**

Pin#	Pin Name	Туре	Description	
1	IN1	Input	Video input, Channel 1	
2	IN2	Input	Video input, Channel 2	
3	IN3	Input	Video input, Channel 3	
4	V <sub>CC</sub>	Input	+5V supply, do not float	
5	IN4	Input	Video input, Channel 4	
6	IN5	Input	Video input, Channel 5	
7	IN6	Input	Video input, Channel 6	
8	OUT6	Output	Filtered video output, Channel 6	
9	OUT5	Output	Filtered video output, Channel 5	
10	OUT4	Output	Filtered video output, Channel 4	
11	GND	Output	Must be tied to ground, do not float	
12	OUT3	Output	iltered video output, Channel 3	
13	OUT2	Output	Filtered video output, Channel 2	
14	OUT1	Output	Filtered video output, Channel 1	

# **Absolute Maximum Ratings**

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

Parameter	Min.	Max.	Unit
DC Supply Voltage	-0.3	6	V
Analog and Digital I/O	-0.3	V <sub>CC</sub> + 0.3	V
Output Channel - Any One Channel (Do Not Exceed)		50	mA

## **Reliability Information**

Symbol	Parameter	Min.	Тур.	Max.	Unit
TJ	Junction Temperature			+150	°C
T <sub>STG</sub>	Storage Temperature Range	-65		+150	°C
TL	Lead Temperature (Soldering, 10 seconds)			+300	°C
$\Theta_{JA}$	Thermal Resistance, JEDEC Standard Multi-layer Test Boards, Still Air		90		°C/W

# **Recommended Operating Conditions**

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to absolute maximum ratings.

Parameter	Min.	Тур.	Max.	Unit
Operating Temperature Range	-40		+85	°C
V <sub>CC</sub> Range	+4.75	+5.0	+5.25	V

## **DC Electrical Characteristics**

 $T_A = 25^{\circ}$ C,  $V_{CC} = 5$ V,  $R_{SOURCE} = 37.5\Omega$ ; all inputs are AC coupled with 0.1µF; all outputs are AC coupled with 220µF into 150Ω loads; unless otherwise noted.

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Units
I <sub>cc</sub>	Supply Current <sup>(1)</sup>	FMS6146 (No Load)		35	55	mA
V <sub>IN</sub>	Video Input Voltage Range	Referenced to GND if DC coupled		1.4		Vpp
PSRR	Power Supply Rejection	DC (All Channels)		-50		dB

### Note:

1. 100% tested at 25°C.

### **AC Electrical Characteristics**

 $T_A = 25^{\circ}C$ ,  $V_{IN} = 1V_{pp}$ ,  $V_{CC} = 5V$ ,  $R_{SOURCE} = 37.5\Omega$ ; all inputs are AC coupled with  $0.1\mu$ F; all outputs are AC coupled with  $220\mu$ F into  $150\Omega$  loads; unless otherwise noted.

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Units
AV	Channel Gain <sup>(1)</sup>	All Channels	6.0	6.2	6.4	dB
f <sub>1dB</sub>	-1dB Bandwidth <sup>(1)</sup>	All Channels	4.5	6.8		MHz
f <sub>c</sub>	-3dB Bandwidth	All Channels		7.8		MHz
f <sub>SB</sub>	Attenuation (Stopband Reject)	All Channels at f = 27MHz		48		dB
dG	Differential Gain	All Channels		0.3		%
dø	Differential Phase	All Channels		0.6		0
THD	Output Distortion (All Channels)	V <sub>OUT</sub> = 1.8V <sub>pp</sub> , 1MHz		0.4		%
X <sub>TALK</sub>	Crosstalk (Channel-to-Channel)	at 1MHz		-60		dB
SNR	Signal-to-Noise Ratio	All Channels NTC-7 Weighting: 100kHz to 4.2MHz		75		dB
<sup>t</sup> pd	Propagation Delay	Delay from Input-to-Output, 4.5MHz		59		ns

Note:

1.100% tested at 25°C.

## **Typical Performance Characteristics**

 $T_A = 25^{\circ}C$ ,  $V_{CC} = 5V$ ,  $R_{SOURCE} = 37.5\Omega$ ; all inputs AC coupled with  $0.1\mu$ F; all outputs are AC coupled with  $220\mu$ F into  $150\Omega$  loads; unless otherwise noted.



Figure 1. Frequency Response



Figure 3. Noise vs. Frequency



Figure 5. Differential Phase



Figure 2. Group Delay vs. Frequency



Figure 4. Differential Gain

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## **Typical Application Diagrams**

The following circuit may be used for direct DC-coupled drive by DACs with an output voltage range of 0V to 1.4V. AC-coupled or DC-coupled outputs may be used with AC-coupled outputs, offering slightly lower power dissipation.



## **Application Information**

### **Application Circuits**

The FMS6146 Low Cost Video Filter (LCVF) provides 6dB gain from input to output. In addition, the input is slightly offset to optimize the output driver performance. The offset is held to the minimum required value to decrease the standing DC current into the load. Typical voltage levels are shown in the diagram below:







#### Figure 7. Typical Voltage Levels

The FMS6146 provides an internal diode clamp to support AC-coupled input signals. If the input signal does not go below ground, the input clamp does not operate. This allows DAC outputs to directly drive the FMS6146 without an AC coupling capacitor. When the input is AC coupled, the diode clamp sets the sync tip (or lowest voltage) just below ground. The worst-case sync tip compression due to the clamp cannot exceed 7mV. The input level set by the clamp, combined with the internal DC offset, keeps the output within its acceptable range.

For symmetric signals like Chroma, U, V, Pb, and Pr, the average DC bias is fairly constant and the inputs can be AC coupled with the addition of a pull-up resistor to set the DC input voltage. DAC outputs can also drive these same signals without the AC coupling capacitor. A conceptual illustration of the input clamp circuit is shown in Figure 8.



Figure 8. Input Clamp Circuit

### I/O Configurations

For a DC-coupled DAC drive with DC-coupled outputs, use the configuration shown in Figure 9.



Figure 9. DC-Coupled Inputs and Outputs

Alternatively, if the DAC's average DC output level causes the signal to exceed the range of 0V to 1.4V, it can be AC coupled as shown in Figure 10.



### Figure 10. AC-Coupled Inputs, DC-Coupled Outputs

When the FMS6146 is driven by an unknown external source or a SCART switch with its own clamping circuitry, the inputs should be AC coupled as shown in Figure 11.





The same method can be used for biased signals, with the addition of a pull-up resistor to make sure the clamp never operates. The internal pull-down resistance is  $800k\Omega \pm 20\%$ , so the external resistance should be 7.5M $\Omega$  to set the DC level to 500mV as shown in Figure 12.



Figure 12. Biased SCART with DC-Coupled Outputs

The same circuits can be used with AC-coupled outputs if desired.



Figure 14. AC-Coupled Inputs and Outputs

## **Layout Considerations**

General layout and supply bypassing play major roles in high-frequency performance and thermal characteristics. Fairchild offers a demonstration board, FMS6146DEMO, to guide layout and aid device testing and characterization. The FMS6146DEMO is a four-layer board with full power and ground planes. Following this layout configuration provides the optimum performance and thermal characteristics. For optimum results, follow the guidelines below as a basis for high-frequency layout:



**NOTE:** The video tilt or line time distortion is dominated by the AC-coupling capacitor. The value may need to be increased beyond  $220\mu$ F to obtain satisfactory operation in some applications.

### **Power Dissipation**

The FMS6146 output drive configuration must be considered when calculating overall power dissipation. Care must be taken not to exceed the maximum die junction temperature. The following example can be used to calculate the FMS6146's power dissipation and internal temperature rise.

$\Gamma_j = T_A + P_d \cdot \theta_{JA}$	EQ. 1
where: $P_d = P_{CH1} + P_{CH2} + P_{CH3}$ and	EQ. 2
$P_{CHx} = V_{CC} \cdot I_{CH} - (V_O^2/R_L)$	EQ. 3
where: $V_0 = 2V_{IN} + 0.280V$	EQ. 4

$_{CH} = (I_{CC}/3) + (V_O/R_L)$	EQ. 5
	L Q. 0

V<sub>IN</sub> = RMS value of input signal

 $I_{CC} = 35 \text{mA}$ 

V<sub>CC</sub> = 5V

R<sub>L</sub> = channel load resistance

Board layout can also affect thermal characteristics. Refer to *Layout Considerations* for more information.

The FMS6146 is specified to operate with output currents typically less than 50mA, more than sufficient for a dual  $(75\Omega)$  video load. Internal amplifiers are current limited to a maximum of 100mA and should withstand brief-duration short-circuit conditions; this capability is not guaranteed.

- Include 1µF and 0.1µF ceramic bypass capacitors.
- Place the 1µF capacitor within 0.75 inches of the power pin.
- Place the 0.1µF capacitor within 0.1 inches of the power pin.
- For multi-layer boards, use a large ground plane to help dissipate heat.
- For two-layer boards, use a ground plane that extends beyond the device by at least 0.5 inches.
- Minimize all trace lengths to reduce series inductances.







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