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September 2004



LM4880 Boomer[®] Audio Power Amplifier Series Dual 250 mW Audio Power Amplifier with Shutdown Mode

General Description

The LM4880 is a dual audio power amplifier capable of delivering typically 250mW per channel of continuous average power to an 8Ω load with 0.1% THD+N using a 5V power supply.

Boomer audio power amplifiers were designed specifically to provide high quality output power with a minimal amount of external components using surface mount packaging.

Since the LM4880 does not require bootstrap capacitors or snubber networks, it is optimally suited for low-power portable systems.

The LM4880 features an externally controlled, low-power consumption shutdown mode, as well as an internal thermal shutdown protection mechanism.

The unity-gain stable LM4880 can be configured by external gain-setting resistors.

Key Specifications

 THD+N at 1kHz at 200mW continuous average output power into 8Ω: 0.1% (max)

Connection Diagram

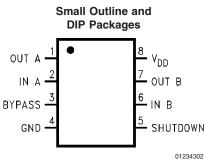
- THD+N at 1kHz at 85mW continuous average output power into 32Ω: 0.1% (typ)
- Output power at 10% THD+N at 1kHz into 8Ω: 325mW (typ)
- Shutdown current: 0.7µA (typ)
- 2.7V to 5.5V supply voltage range

Features

- No bootstrap capacitors or snubber circuits are necessary
- Small Outline (SO) and DIP packaging
- Unity-gain stable
- External gain configuration capability

Applications

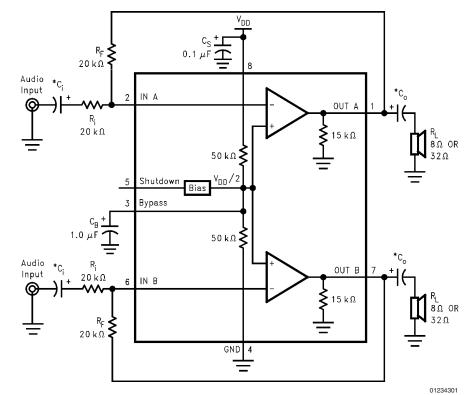
- Headphone Amplifier
- Personal Computers
- CD-ROM Players



Top View Order Number LM4880M or LM4880N See NS Package Number M08A for SO or NS Package Number N08E for DIP

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Typical Application



*Refer to the Application Information section for information concerning proper selection of the input and output coupling capacitors.

FIGURE 1. Typical Audio Amplifier Application Circuit

Absolute Maximum Ratings (Note 2)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

Supply Voltage	6.0V
Storage Temperature	-65°C to +150°C
Input Voltage	–0.3V to V _{DD} + 0.3V
Power Dissipation (Note 3)	Internally limited
ESD Susceptibility (Note 4)	2000V
ESD Susceptibility (Note 5)	200V
Junction Temperature	150°C
Soldering Information Small Outline Package	
Vapor Phase (60 sec.)	215°C
Infrared (15 sec.)	220°C

See AN-450 "Surface Mounting and their Effects on Product Reliability" for other methods of soldering surface mount devices.

Thermal Resistance

θ_{JC} (DIP)	37°C/W
θ_{JA} (DIP)	107°C/W
$\theta_{\rm JC}$ (SO)	35°C/W
θ_{JA} (SO)	170°C/W

Operating Ratings

Temperature Range	
T _{MIN} ≤T _A ≤T _{MAX}	–40°C≤T _A ≤+85°C
Supply Voltage	2.7V≤V _{DD} ≤5.5V

Electrical Characteristics (Notes 1, 2)

The following specifications apply for V_{DD} = 5V unless otherwise specified. Limits apply for T_A = 25°C.

Symbol	Parameter	Conditions	LM4	LM4880	
			Typical	Limit	(Limits)
			(Note 6)	(Note 7)	
V _{DD}	Supply Voltage			2.7	V (min)
				5.5	V (max)
I _{DD}	Quiescent Power Supply Current	V _{IN} =0V, I _O =0A	3.6	6.0	mA
					(max)
I _{SD}	Shutdown Current	V _{PIN5} =V _{DD}	0.7	5	µA (max)
Vos	Output Offset Voltage	V _{IN} =0V	5	50	mV
					(max)
Po	Output Power	THD=0.1% (max); f=1 kHz;			
		$R_L=8\Omega$	250	200	mW
					(min)
		$R_L=32\Omega$	85		mW
		THD+N=10%; f=1 kHz			
		$R_L=8\Omega$	325		mW
		$R_L=32\Omega$	110		mW
THD+N	Total Harmonic Distortion+Noise	R _L =8Ω, P _O =200 mW;	0.03		%
		R _L =32Ω, P _O =75 mW;	0.02		%
		f=1 kHz			
PSRR	Power Supply Rejection Ratio	C _B = 1.0 μF,	50		dB
		V _{RIPPLE} =200 mVrms, f = 100 Hz			

Note 1: All voltages are measured with respect to the ground pin, unless otherwise specified.

Note 2: Absolute Maximum Ratings indicate limits beyond which damage may occur. Operating Ratings indicate conditions for which the device is functional, but do not guarantee specific performance limits. Electrical Characteristics state DC and AC electrical specifications under particular test conditions which guarantee specific performance limits. This assumes that the device is within the Operating Ratings. Specifications are not guaranteed for parameters where no limit is given, however, the typical value is a good indication of device performance.

Note 3: The maximum power dissipation must be derated at elevated temperatures and is dictated by T_{JMAX} , θ_{JA} , and the ambient temperature T_A . The maximum allowable power dissipation is $P_{DMAX} = (T_{JMAX} - T_A)/\theta_{JA}$ or the number given in the Absolute Maximum Ratings, whichever is lower. For the LM4880, $T_{JMAX} = 150^{\circ}$ C, and the typical junction-to-ambient thermal resistance is 170°C/W for package M08A and 107°C/W for package N08E.

Note 4: Human body model, 100 pF discharged through a 1.5 $k\Omega$ resistor.

Note 5: Machine model, 220 pF-240 pF discharged through all pins.

Note 6: Typicals are measured at 25°C and represent the parametric norm.

Note 7: Limits are guaranteed to National's AOQL (Average Outgoing Quality Level).

Automatic Shutdown Circuit

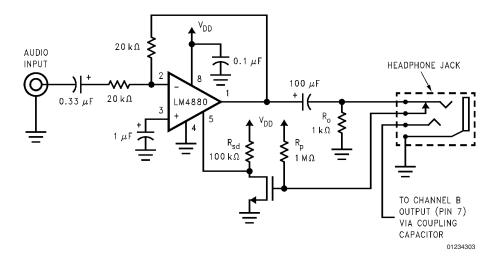
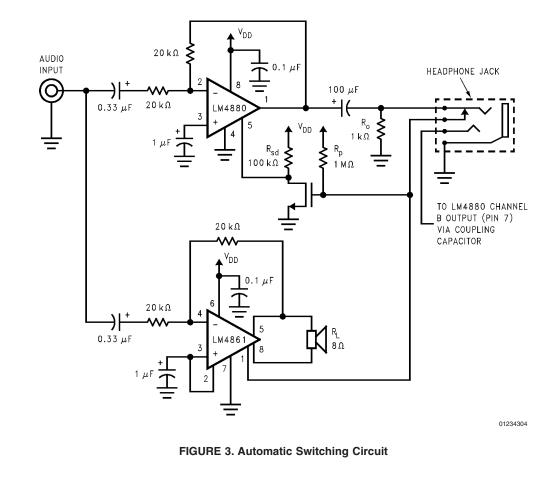


FIGURE 2. Automatic Shutdown Circuit

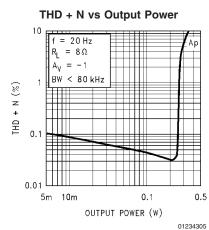
Automatic Switching Circuit



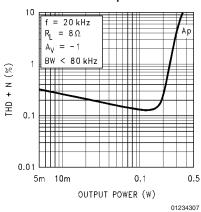
External Components Description (Figure 1)

Components		Functional Description				
1. R _i		Inverting input resistance which sets the closed-loop gain in conjunction with R _F . This resistor				
		also forms a high pass filter with C_i at $f_c = 1/(2\pi R_i C_i)$.				
2.	Ci	Input coupling capacitor which blocks the DC voltage at the amplifier's input terminals. Also				
		creates a high pass filter with R_i at $f_c = 1/(2\pi R_i C_i)$. Refer to the section, Proper Selection of				
		External Components, for an explanation of how to determine the value of C _i .				
3.	R _F	Feedback resistance which sets closed-loop gain in conjunction with R _i .				
4.	Cs	Supply bypass capacitor which provides power supply filtering. Refer to the Application				
		Information section for proper placement and selection of the supply bypass capacitor.				
5.	C _B	Bypass pin capacitor which provides half-supply filtering. Refer to the section, Proper				
		Selection of External Components, for information concerning proper placement and				
		selection of C _B .				
6.	C _o	Output coupling capacitor which blocks the DC voltage at the amplifier's output. Forms a high				
		pass filter with R_{I} at $f_{o} = 1/(2\pi R_{I} C_{o})$.				

Typical Performance Characteristics

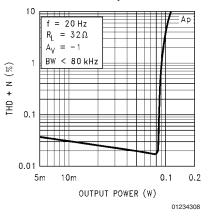


THD + N vs Output Power



10 f = 1 kHzR_L = 8Ω $A_V = -1$ BW < 80 kHz 1 THD + N (%) 0.1 0.01 5m 10m 0.1 0.5 OUTPUT POWER (W) 01234306

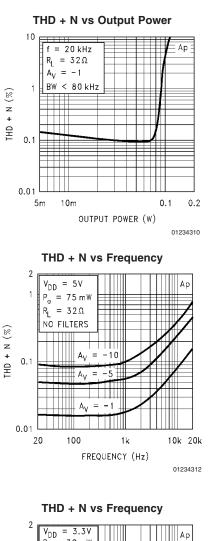
THD + N vs Output Power

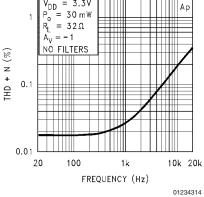


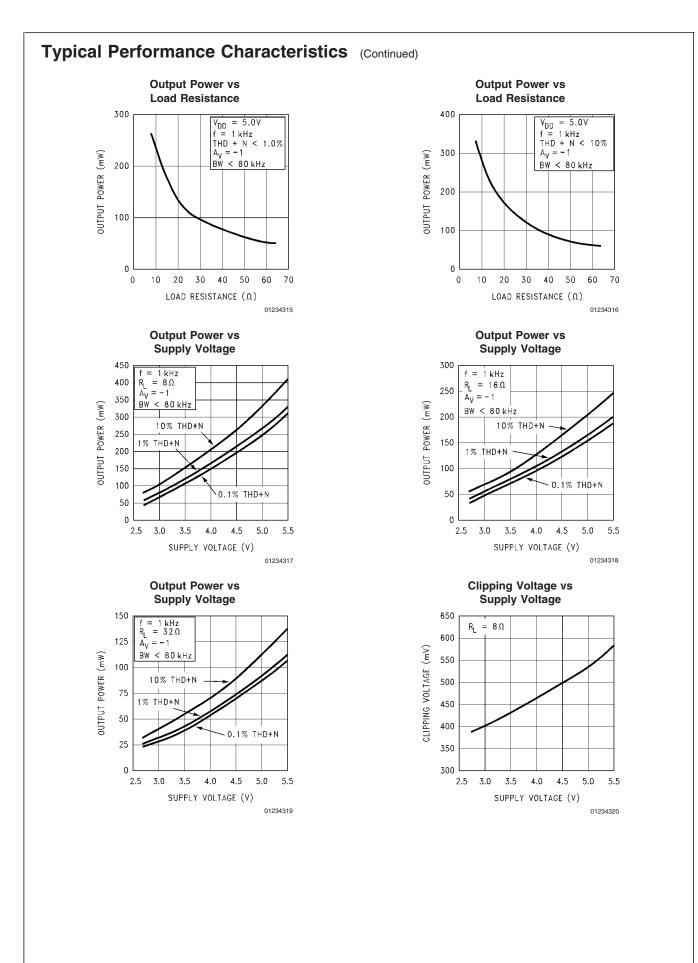
THD + N vs Output Power

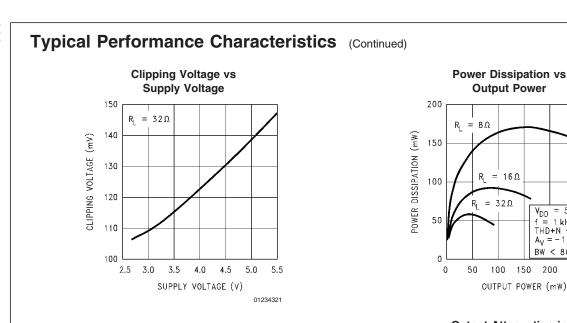
LM4880

Typical Performance Characteristics (Continued) THD + N vs Output Power 10 Ap f = 1 kHz $R_L = 32\Omega$ $A_{V} = -1$ BW < 80 kHz 1 THD + N (%) ++++ 0.1 0.01 5m 1**0**m 0.1 0.2 OUTPUT POWER (W) 01234309 THD + N vs Frequency 5 $V_{DD} = 5V$ Ap $P_0 = 200 \text{ mW}$ $R_L = 8 \Omega$ 1 NO FILTERS THD + N (%) A_V = -10 0.1 F∓∐<u>III</u>I Av -5 [||||||| A_v = 0.01 20 100 10k 20k 1k FREQUENCY (Hz) 01234311 THD + N vs Frequency 0.1 ff $V_{DD} = 5V$ Ap = 4 V_{PP} V_o RĽ $= 10 k\Omega$ Āv = -1 ВŴ < 80 kHz THD + N (%) 0.01 0.001 20 100 1k 10k 20k FREQUENCY (Hz) 01234313

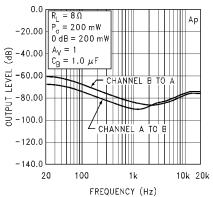




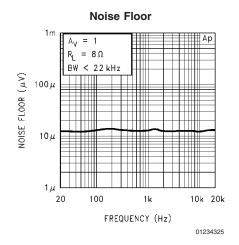


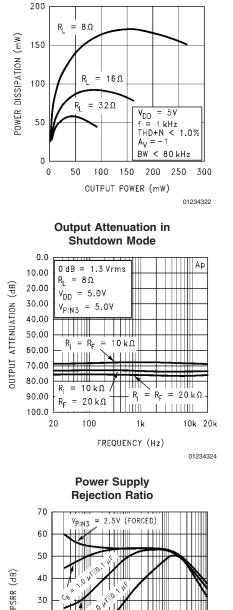




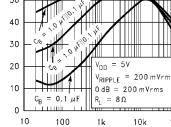


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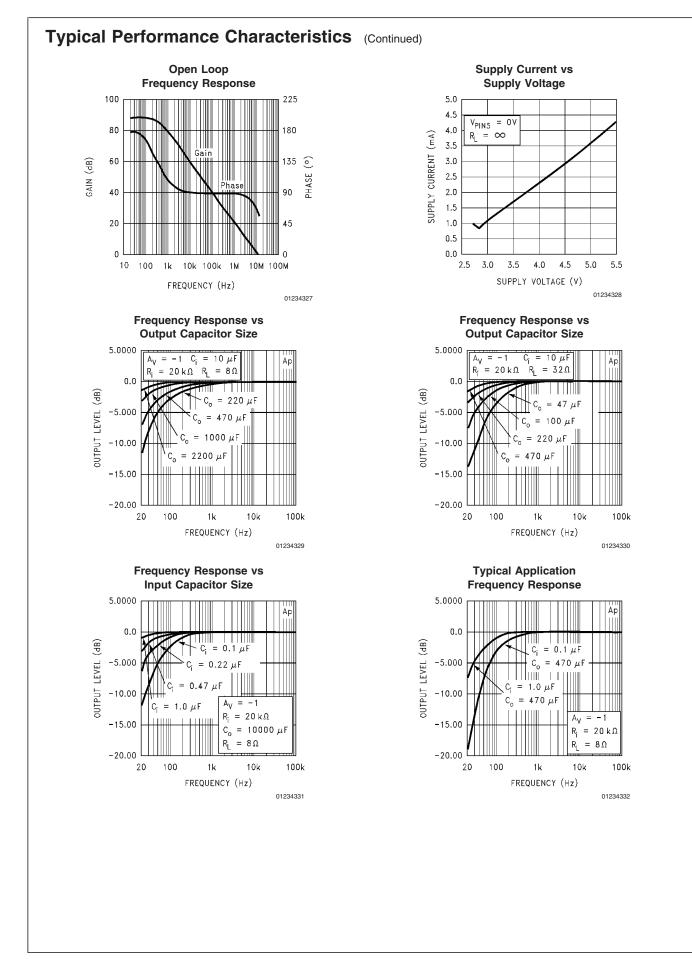
Output Power



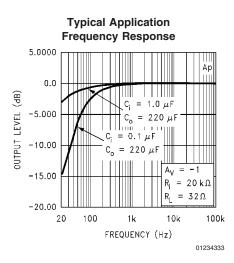
FREQUENCY (Hz)

01234326

100k



Typical Performance Characteristics (Continued)



Application Information

SHUTDOWN FUNCTION

In order to reduce power consumption while not in use, the LM4880 contains a shutdown pin to externally turn off the amplifier's bias circuitry. This shutdown feature turns the amplifier off when a logic high is placed on the shutdown pin. The trigger point between a logic low and logic high level is typically half supply. It is best to switch between ground and the supply to provide maximum device performance. By switching the shutdown pin to $V_{\rm DD}$, the LM4880 supply current draw will be minimized in idle mode. While the device will be disabled with shutdown pin voltages less than $V_{\rm DD}$, the idle current may be greater than the typical value of 0.7 μ A. In either case, the shutdown pin floating may result in an unwanted shutdown condition.

In many applications, a microcontroller or microprocessor output is used to control the shutdown circuitry which provides a quick, smooth transition into shutdown. Another solution is to use a single-pole, single-throw switch in conjunction with an external pull-up resistor. When the switch is closed, the shutdown pin is connected to ground and enables the amplifier. If the switch is open, then the external pull-up resistor will disable the LM4880. This scheme guarantees that the shutdown pin will not float which will prevent unwanted state changes.

POWER DISSIPATION

Power dissipation is a major concern when using any power amplifier and must be thoroughly understood to ensure a successful design. *Equation (1)* states the maximum power dissipation point for a single-ended amplifier operating at a given supply voltage and driving a specified output load.

$$P_{\rm DMAX} = (V_{\rm DD})^2 / (2\pi^2 R_{\rm L})$$
(1)

Since the LM4880 has two operational amplifiers in one package, the maximum internal power dissipation point is twice that of the number which results from *Equation (1)*. Even with the large internal power dissipation, the LM4880 does not require heat sinking over a large range of ambient temperatures. From *Equation (1)*, assuming a 5V power supply and an 8Ω load, the maximum power dissipation point is 158 mW per amplifier. Thus the maximum package

Power Derating Curve 1250 Package N08E 1000 POWER DISSIPATION (mW) 750 500 250 Package M08A 0 -25 0 25 50 75 100 125 150 AMBIENT TEMPERATURE (°C) 01234334

dissipation point is 317 mW. The maximum power dissipation point obtained must not be greater than the power dissipation that results from *Equation (2)*:

$$\mathsf{P}_{\mathsf{DMAX}} = (\mathsf{T}_{\mathsf{JMAX}} - \mathsf{T}_{\mathsf{A}})/\theta_{\mathsf{JA}} \tag{2}$$

For the LM4880 surface mount package, $\theta_{JA} = 170^{\circ}$ C/W and $T_{JMAX} = 150^{\circ}C$. Depending on the ambient temperature, T_A , of the system surroundings, Equation (2) can be used to find the maximum internal power dissipation supported by the IC packaging. If the result of Equation (1) is greater than that of Equation (2), then either the supply voltage must be decreased, the load impedance increased, or the ambient temperature reduced. For the typical application of a 5V power supply, with an 8Ω load, the maximum ambient temperature possible without violating the maximum junction temperature is approximately 96°C provided that device operation is around the maximum power dissipation point. Power dissipation is a function of output power and thus, if typical operation is not around the maximum power dissipation point, the ambient temperature may be increased accordingly. Refer to the Typical Performance Characteristics curves for power dissipation information for lower output powers.

POWER SUPPLY BYPASSING

As with any power amplifier, proper supply bypassing is critical for low noise performance and high power supply rejection. The capacitor location on both the bypass and power supply pins should be as close to the device as possible. As displayed in the **Typical Performance Characteristics** section, the effect of a larger half supply bypass capacitor is improved low frequency PSRR due to increased half-supply stability. Typical applications employ a 5V regulator with 10 μ F and a 0.1 μ F bypass capacitors which aid in supply stability, but do not eliminate the need for bypassing the supply nodes of the LM4880. The selection of bypass capacitors, especially C_B, is thus dependant upon desired low frequency PSRR, click and pop performance as explained in the section, system cost, and size constraints.

AUTOMATIC SHUTDOWN CIRCUIT

As shown in *Figure 2*, the LM4880 can be set up to automatically shutdown when a load is not connected. This circuit is based upon a single control pin common in many headphone jacks. This control pin forms a normally closed

Application Information (Continued)

switch with one of the output pins. The output of this circuit (the voltage on pin 5 of the LM4880) has two states based on the state of the switch. When the switch is open, signifying that headphones are inserted, the LM4880 should be enabled. When the switch is closed, the LM4880 should be off to minimize power consumption.

The operation of this circuit is rather simple. With the switch closed, $R_{\rm p}$ and $R_{\rm o}$ form a resistor divider which produces a gate voltage of less than 5 mV. This gate voltage keeps the NMOS inverter off and $R_{\rm sd}$ pulls the shutdown pin of the LM4880 to the supply voltage. This places the LM4880 in shutdown mode which reduces the supply current to 0.7 μA typically. When the switch is open, the opposite condition is produced. Resistor $R_{\rm p}$ pulls the gate of the NMOS high which turns on the inverter and produces a logic low signal on the shutdown pin of the LM4880. This state enables the LM4880 and places the amplifier in its normal mode of operation.

This type of circuit is clearly valuable in portable products where battery life is critical, but is also benefical for power conscious designs such as "Green PC's".

AUTOMATIC SWITCHING CIRCUIT

A circuit closely related to the **Automatic Shutdown Circuit** is the **Automatic Switching Circuit** of *Figure 3*. The **Automatic Switching Circuit** utilizes both the input and output of the NMOS inverter to toggle the states of two different audio power amplifiers. The LM4880 is used to drive stereo single ended loads, while the LM4861 drives bridged internal speakers.

In this application, the LM4880 and LM4861 are never on at the same time. When the switch inside the headphone jack is open, the LM4880 is enabled and the LM4861 is disabled since the NMOS inverter is on. If a headphone jack is not present, it is assumed that the internal speakers should be on and thus the voltage on the LM4861 shutdown pin is low and the voltage at the LM4880 pin is high. This results in the LM4880 being shutdown and the LM4861 being enabled.

Only one channel of this circuit is shown in *Figure 3* to keep the drawing simple but the typical application would a LM4880 driving a stereo external headphone jack and two LM4861's driving the internal stereo speakers. If only one internal speaker is required, a single LM4861 can be used as a summer to mix the left and right inputs into a single mono channel.

PROPER SELECTION OF EXTERNAL COMPONENTS

Selection of external components when using integrated power amplifiers is critical to optimize device and system performance. While the LM4880 is tolerant of external component combinations, care must be exercised when choosing component values.

The LM4880 is unity-gain stable which gives a designer maximum system flexibility. The LM4880 should be used in low gain configurations to minimize THD + N values, and maximize the signal to noise ratio. Low gain configurations require large input signals to obtain a given output power. Input signals equal to or greater than 1 Vrms are available from sources such as audio codecs. Please refer to the section, **Audio Power Amplifier Design**, for a more complete explanation of proper gain selection.

Besides gain, one of the major design considerations is the closed-loop bandwidth of the amplifier. To a large extent, the

bandwidth is dictated by the choice of external components shown in *Figure 1*. Both the input coupling capacitor, C_i , and the output coupling capacitor, C_o , form first order high pass filters which limit low frequency response. These values should be chosen based on needed frequency response for a few distinct reasons.

Selection of Input and Output Capacitor Size

Large input and output capacitors are both expensive and space hungry for portable designs. Clearly a certain sized capacitor is needed to couple in low frequencies without severe attenuation. But in many cases the transducers used in portable systems, whether internal or external, have little ability to reproduce signals below 100 Hz–150 Hz. Thus using large input and output capacitors may not increase system performance.

In addition to system cost and size, click and pop performance is effected by the size of the input coupling capacitor, C_i. A larger input coupling capacitor requires more charge to reach its quiescent DC voltage (normally 1/2 V_{DD} .) This charge comes from the output via the feedback and is apt to create pops upon device enable. Thus, by minimizing the capacitor size based on necessary low frequency response, turn-on pops can be minimized.

Besides minimizing the input and output capacitor sizes, careful consideration should be paid to the bypass capacitor size. The bypass capacitor, C_B, is the most critical component to minimize turn-on pops since it determines how fast the LM4880 turns on. The slower the LM4880's outputs ramp to their quiescent DC voltage (nominally 1/2 V_{DD}), the smaller the turn-on pop. Choosing C_B equal to 1.0 μ F along with a small value of C_i (in the range of 0.1 μ F to 0.39 μ F), should produce a virtually clickless and popless shutdown function. While the device will function properly, (no oscillations or motorboating), with C_B equal to 0.1 μ F, the device will be much more susceptible to turn-on clicks and pops. Thus, a value of C_B equal to 1.0 μ F or larger is recommended in all but the most cost sensitive designs.

AUDIO POWER AMPLIFIER DESIGN

Design a Dual 200 mW/8 Ω Audio Amplifier Given:

Power Output: 200 mWrms Load Impedance: 8Ω Input Level: 1 Vrms (max) Input Impedance: 20 k Ω Bandwidth: 100 Hz-20 kHz ± 0.50 dB

A designer must first determine the needed supply rail to obtain the specified output power. Calculating the required supply rail involves knowing two parameters, V_{opeak} and also the dropout voltage. As shown in the Typical Performance Curves, the dropout voltage is typically 0.5V. V_{opeak} can be determined from Equation (3).

$$V_{opeak} = \sqrt{(2R_LP_o)}$$

For 200 mW of output power into an 8 Ω load, the required V_{opeak} is 1.79V. Since this is a single supply application, the minimum supply voltage is twice the sum of V_{opeak} and V_{od}. Since 5V is a standard supply voltage in most applications, it is chosen for the supply rail. Extra supply voltage creates headroom that allows the LM4880 to reproduce peaks in excess of 200 mW without clipping the signal. At this time, the designer must make sure that the power supply choice along with the output impedance does not violate the conditions explained in the **Power Dissipation** section. Remem-

(3)

Application Information (Continued)

ber that the maximum power dissipation value from Equation (1) must be multiplied by two since there are two independent amplifiers inside the package.

Once the power dissipation equations have been addressed, the required gain can be determined from Equation (4).

$$|A_V| \ge \sqrt{(P_0 R_L)} / (V_{IN}) = V_{orms} / V_{inrms}$$

$$A_V = -R_F / R_i$$
(4)
(5)

From Equation (4), the minimum gain is:: $A_V = -1.26$

Since the desired input impedance was 20 k Ω , and with a gain of –1.26, a value of 27 k Ω is designated for R_f, assuming 5% tolerance resistors. This combination results in a nominal gain of –1.35. The final design step is to address the bandwidth requirements which must be stated as a pair of –3 dB frequency points. Five times away from a –3 dB point is 0.17 dB down from passband response assuming a single pole roll-off. As stated in the **External Components** section,

both R_i in conjunction with C_i, and C_o with R_L, create first order high pass filters. Thus to obtain the desired frequency low response of 100 Hz within ± 0.5 dB, both poles must be taken into consideration. The combination of two single order filters at the same frequency forms a second order response. This results in a signal which is down 0.34 dB at five times away from the single order filter –3 dB point. Thus, a frequency of 20 Hz is used in the following equations to ensure that the response if better than 0.5 dB down at 100 Hz.

 $C_i \ge 1/(2\pi^*20k\Omega^*20Hz) = 0.397 \ \mu\text{F}$; use 0.39 μF

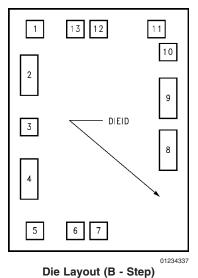
 $C_o \ge 1/(2\pi^* 8\Omega^* 20 Hz) = 995 \ \mu\text{F}$; use 1000 μF

The high frequency pole is determined by the product of the desired high frequency pole, f_H , and the closed-loop gain, A_V. With a closed-loop gain magnitude of 1.35 and $f_H = 100$ kHz, the resulting GBWP = 135 kHz which is much smaller than the LM4880 GBWP of 12.5 MHz. This figure displays that if a designer has a need top design an amplifier with a higher gain, the LM4880 can still be used without running into bandwidth limitations.

Application Information (Continued)

LM4880 MDA MWA

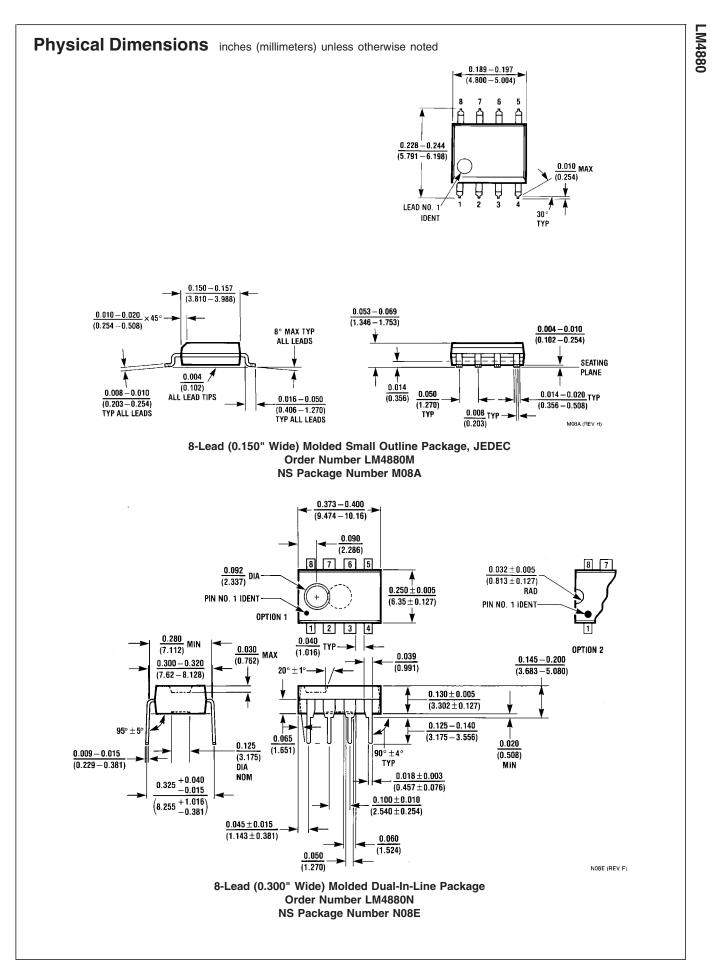
DUAL 250 MW AUDIO POWER AMPLIFIER WITH SHUTDOWN MODE



		Die/Wa	afer Characte	ristics			
	Fabrication At	tributes			General	Die Informat	ion
Physical Die Ide	nysical Die Identification LM4880B		Bond Pad Opening Size (min)		ze (min)	86µm x 86µm	
Die Step			Bond Pa	Bond Pad Metalization		ALUMINUM	
	Physical Attr	ibutes		Passiva	tion		NITRIDE
Wafer Diameter		150mm	ı	Back Sid	de Metal		Bare Back
Dise Size (Draw	'n)		952µm x 1283µm Back Side Connection 37mils x 51mils		1	GND	
Thickness		254µm	Nominal				
Min Pitch		117µm	Nominal				
Special Assem	bly Requirements:	ł					
	ie size is rounded to	o the nearest mid	cron.				
		Die Bond Pad C	oordinate Loc	ations (B -	Sten)		
	(Referen	ced to die center,				ion	
			X/Y COORDINATES		PAD SIZE		
SIGNAL NAME	PAD# NUMBER	Х	Y		Х		Y
BYPASS	1	-322	523		86	x	86
GND	2	-359	259		86	x	188
NC	3	-359	5		86	x	86
GND	4	-359	-259		86	х	188
SHUTDOWN	5	-323	-523		86	x	86
INPUT B	6	-109	-523		86	х	86
OUTPUT B	7	8	-523		86	x	86
VDD	8	358	-78		86	x	188
GND	9	358	141		86	х	188
OUTPUT A	10	359	406		86	х	86
INPUT A	11	323	523		86	х	86
NC	12	8	523		86	х	86
NC	13	-109	523		86	x	86

Application Information (Continued)

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Fax:	1 207 541 6140
IN EUROPE	
Tel:	49 (0) 8141 351492 / 1495
Fax:	49 (0) 8141 351470
IN ASIA PACIFIC	
Tel:	(852) 27371701
IN JAPAN	
Tel:	81 043 299 2308



Notes

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- 2. A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

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