NSSHNBO

R1286K SERIES

2ch. PWM Step-up / Inverting DC/DC Converter with Synchronous Rectifier for AMOLED / LCD

NO.EA-283-191114

OUTLINE

The R1286K 2ch DC/DC converter is designed for AMOLED display power source. It contains a step up DC/DC converter and an inverting DC/DC converter.

Step up DC/DC converter generates boosted output voltage to 4.6 V to 5.8 V (Selectable). Inverting DC/DC converter generates negative voltage down to -2.0 V to -6.0 V (Selectable) that is dynamically adjustable with single wire interface signal. R1286K consist of a voltage reference, error amplifiers, an oscillator, PWM control circuits, over current protection circuits, short protection circuits, an under voltage lockout circuit (UVLO), thermal shutdown circuit, a NMOS driver and a synchronous PMOS switch for boost converter, a PMOS driver and a synchronous NMOS switch for inverting converter, and so on. High efficiency boost and inverting DC/DC converters can be composed with two external inductors and three capacitors.

FEATURES

•	Operating Voltage ·····	2.3 V to 5.5 V
[Ste	p-up DC/DC Converter (CH1)]	
•	Selectable Output Voltage (Vourp) ······	R1286KxxxX ⁽¹⁾ : 4.6 V to 5.8 V (0.1V Step)
•	Externally Adjustable Output Voltage	R1286K001B: 4.6 V to 5.8 V
•	Maximum Output Current	R1286K0xxX ⁽¹⁾ / R1286K001B: 250 mA
		R1286K1xxX ⁽¹⁾ : 300 mA
•	VOUTP Voltage Load Regulation	Typ.± 5 mV
•	VOUTP Voltage Line Transient Response ·······	Typ. ± 10 mV
[lnv	erting DC/DC Converter (CH2)]	
•	Dynamically Adjustable Output Voltage (V_{OUTN})	-2.0 V to -6.0 V (Fixed Rate: 3.0 V, 0.1 V Step)
•	Selectable Single Wire (S-Wire) I/F	R1286KxxxX ⁽¹⁾ : Default value (0.1 V Step)
•	Externally Adjustable Output Voltage	R1286K001B: -2.0 V to -6.0 V
•	Maximum Output Current	R1286K0xxX ⁽¹⁾ / R1286K001B: 250 mA
		R1286K1xxX ^{(1):} 300 mA
•	VOUTN Voltage Load Regulation	Typ.± 5 mV
•	VOUTN Voltage Line Transient Response	
	ntroller]	
•	Internal Start-up Sequence Control with Soft-start	Operation
•	Auto Discharge Operation for Both Outputs	
•	Short circuit protection	
•	Internal timer-latch protection	Typ. 16 ms or 40 ms
•	Maximum duty cycle	Тур. 85% (СН1) / Тур. 90% (СН2)

⁽¹⁾ X : A to N (Provided except B and I)

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- LX peak current limit ······ R1286K0xxX⁽¹⁾ : Typ. 1.0 A (CH1), 1.5 A (CH2)
 - R1286K1xxX⁽¹⁾: Typ. 1.1 A (CH1)、1.8 A (CH2)
- UVLO(Under voltage lock out) protection Typ. 2.05 V
- Thermal Shutdown Typ. 150°C
- Operating Frequency 1750kHz
- Package DFN(PL)2730-12

APPLICATION

- Fixed voltage power supply for portable equipment
- Fixed voltage power supply for AMOLED, LCD

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SELECTION GUIDE

The inverting output voltage (V_{OUTN}), the positive output voltage (V_{OUTP}) and the versions of the inverting output voltage are user-selectable options.

Product Name	Package	Quantity per Reel	Pb Free	Halogen Free
R1286K\$xx*-TR	DFN(PL)2730-12	5,000pcs	Yes	Yes

- \$: Specify the delay time for timer latch ⁽¹⁾.
 - (0) Typ.16msec
 - (1) Typ.40msec

xx: Specify the set output voltages (V_{SET}) for default value of V_{OUTx} and V_{ONDEF}^{(2)}

- *: Specify setting methods for V_{OUTN} and V_{OUTP}.
 - VONDEF : VOUTN default value⁽³⁾ (Internal fixed value at shipping)
 - VONMIN : VOUTN minimum value with S-Wire

VONMAX : VOUTN maximum value with S-Wire

tTRA : Variable time per 0.1V with S-Wire⁽⁴⁾

*	Designation for Settings of VOUTX		V _{ONMIN}	VONMAX	t _{TRA}
Α	VOUTP / VOUTN Fixed Output Voltage type ⁽⁵⁾	-5.4 V to -2.4 V	-5.4 V	-2.4 V	10 ms
В	VOUTP / VOUTN Adjustable Output Voltage type	-	-	-	-
С		-5.0 V to -2.4 V	-5.0 V	-2.0 V	10 ms
D		-5.2 V to -2.4 V	-5.2 V	-2.2 V	10 ms
Ε		-5.6 V to -2.6 V	-5.6 V	-2.6 V	10 ms
F		-5.8 V to -2.8 V	-5.8 V	-2.8 V	10 ms
G		-6.0 V to -3.0 V	-6.0 V	-3.0 V	10 ms
н	VOUTP / VOUTN Fixed Output Voltage type	-5.0 V to -2.4 V	-5.0 V	-2.0 V	360 µs
J		-5.4 V to -2.4 V	-5.4 V	-2.4 V	360 µs
κ		-5.6 V to -2.6 V	-5.6 V	-2.6 V	360 µs
L		-5.8 V to -2.8 V	-5.8 V	-2.8 V	360 µs
Μ		-6.0 V to -3.0 V	-6.0 V	-3.0 V	360 µs
Ν		-5.2 V to -2.4 V	-5.2 V	-2.2 V	360 µs

⁽¹⁾ Fixed Output Voltage type only can select the delay time of 40 msec (Typ).

⁽²⁾ Refer to *Voltage Combination List* for details.

⁽³⁾ Selectable in 0.1V step

⁽⁴⁾ Refer to the *TIMING CHART of S-Wire* for details.

⁽⁵⁾ Dynamically adjustable output voltage with S-Wire

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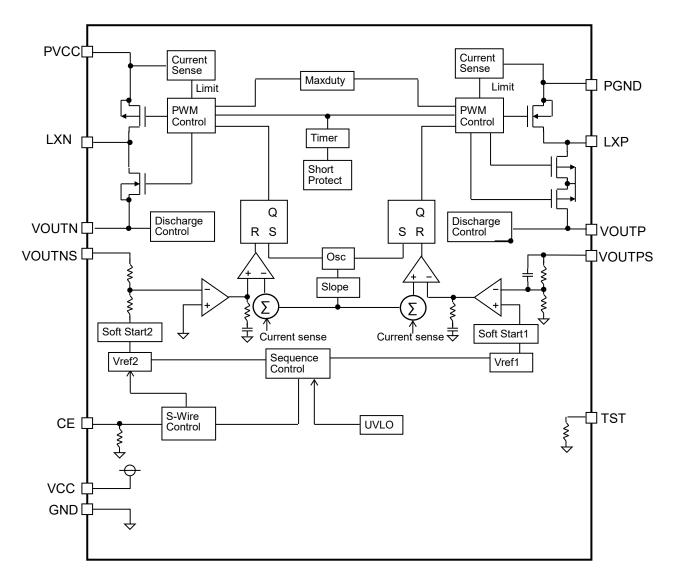
Output voltage combination list

V _{SET} codes (xx)	V _{OUTP}	
01	Setting by external resistor	Setting by external resistor
02	4.6 V	-4.9 V
03	5.8 V	-6.0 V
04	4.8 V	-4.9 V
05	5.4 V	-5.4 V
06	5.0 V	-5.0 V
07	5.0 V	-3.5 V
08	5.6 V	-5.6 V
09	5.8 V	-5.8 V
10	5.5 V	-5.5 V
11	4.6 V	-4.4 V

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BLOCK DIAGRAMS

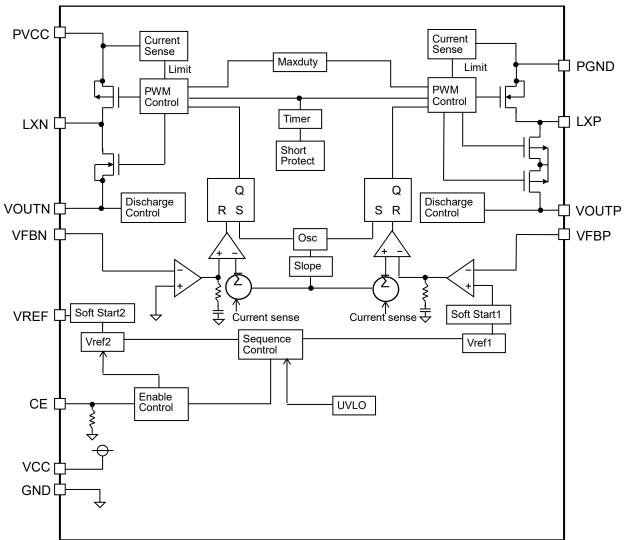
R1286KxxxX⁽¹⁾ (Fixed Output Voltage Type)





⁽¹⁾ X : A to N (Provided, except "B" and "I")

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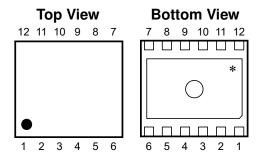


R1286K001B (Adjustable Output Voltage Type)

R1286K001B Block Diagram

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PIN DESCRIPTION



R1286K (DFN(PL)2730-12) Pin Configuration

R1286K Pin Description

Pin	Syr	nbol	Description	
No.	R1286KxxxX ⁽¹⁾	R1286K001B	Description	
1	VOUTNS	VFBN	Feed Back Pin for Inverting DC/DC	
2	VOUTN	VOUTN	Outout Pin for Inverting DC/DC	
3	LXN	LXN	Switching Pin for Inverting DC/DC	
4	PVCC	PVCC	Power Input Pin	
5	VCC	VCC	Analog Power Input Pin	
6	GND	GND	Analog GND Pin	
7	PGND	PGND	Power GND Pin	
8	LXP	LXP	Switching Pin for Step up DC/DC	
9	VOUTP	VOUTP	Output Pin for Step up DC/DC	
10	VOUTPS	VFBP	Feed Back Pin for Step up DC/DC	
11	CE	CE	Chip Enable and S-Wire Control Input Pin (R1286KxxxX) Chip Enable Pin (R1286KxxxB)	
12	TST	VREF	TEST Pin ⁽²⁾ (R1286KxxxX) Reference Voltage Output Pin for Inverting DC/DC (R1286KxxxB)	

* The tab on the bottom of the package is substrate level (GND). It is recommended that the tab be connected to the ground plane on the board.

⁽¹⁾ X : A to N (Provided, except "B" and "I")

⁽²⁾ TEST pin must be connected to the GND or leaving it open.

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ABSOLUTE MAXIMUM RATINGS

		(GND = PGN	D = 0 V)
Symbol	Parameter	Rating	Unit
Vcc	VCC / PVCC Pin Voltage	-0.3 to 6.0	V
VCE	CE Pin Voltage	-0.3 to 6.0	V
VLXP	LXP Pin Voltage	-0.3 to 6.5	V
Voutp(s)	VOUTP Pin Voltage	-0.3 to 6.5	V
V _{LXN}	LXN Pin Voltage	V_{CC} - 14 to V_{CC} + 0.3	V
Voutn(s)	VOUTN Pin Voltage	VCC - 14 to 0.3	V
V _{TST}	TST Pin Voltage [R1286kxxxx ⁽¹⁾]	-0.3 to 6.0	V
VFBP	VFBP Pin Voltage [R1286K001B]	-0.3 to 6.0	V
V _{FBN}	VFBN Pin Voltage [R1286K001B]	-0.3 to VCC + 0.3	V
VREF	VREF Pin Voltage [R1286K001B]	-0.3 to VCC + 0.3	V
PD	Power Dissipation ⁽²⁾ (DFN(PL)2730-12, JEDEC STD. 51-7)	3100	mW
Tj	Junction Temperature Range	-40 to 125	°C
Tstg	Storage Temperature Range	-55 to 125	°C

ABSOLUTE MAXIMUM RATINGS

Electronic and mechanical stress momentarily exceeded absolute maximum ratings may cause the permanent damages and may degrade the lifetime and safety for both device and system using the device in the field. The functional operation at or over these absolute maximum ratings is not assured.

RECOMMENDED OPERATING CONDITIONS

Symbol	Parameter	Rating	Unit
Vcc	Operating Input Voltage	2.3 to 5.5	V
Та	Operating Temperature Range	−40 to 85	°C

RECOMMENDED OPERATING CONDITIONS

All of electronic equipment should be designed that the mounted semiconductor devices operate within the recommended operating conditions. The semiconductor devices cannot operate normally over the recommended operating conditions, even if when they are used over such conditions by momentary electronic noise or surge. And the semiconductor devices may receive serious damage when they continue to operate over the recommended operating conditions.

⁽¹⁾ X : A to N (Provided, except "B" and "I")

⁽²⁾ Refer to POWER DISSIPATION for detailed information.

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ELECTRICAL CHARACTERISTICS

The specifications surrounded by \square are guaranteed by Design Engineering at - 40°C \leq Ta \leq 85°C.

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit
lcc	VCC Consumption Current (at no switching)	Vcc=5.5V		1.2		mA
ISTANDBY	Standby Current	$V_{CC}=V_{LXP}=5.5V$, $V_{CE}=V_{LXN}=0V$		0.1	5	μA
V_{UVLO1}	UVLO Detection Voltage	Falling	1.95	2.05	2.15	V
VUVLO2	UVLO Release Voltage	Rising		V _{UVLO1} +0.10	2.28	V
fosc	Oscillator Frequency	Vcc=3.7V	1500	1750	2000	kHz
VCEH	CE Pin Input Voltage, high	Vcc=5.5V	1.2			V
VCEL	CE Pin Input Voltage, low	Vcc=2.3V			0.4	V
RCE	CE Pin Pull-down Resistance	Vcc=3.7V		160		kΩ
T _{TSD}	Thermal Shutdown Detection Temperature	V _{IN} =3.7V		150		°C
T _{TSR}	Thermal Shutdown Release Temperature	VIN=3.7V		125		°C
[R1286K	0xxx]					
t DLY	Delay Time for Protection	Vcc=3.7V	8	16	24	ms
[R1286K	1xxX ⁽¹⁾]					
t _{DLY}	Delay Time for Protection	Vcc=3.7V	32	40	48	ms
■ Set-up	DC/DC Converter (CH1)					
Maxduty1	Maximum Duty Cycle 1	Vcc=3.7V		85		%
IVOUTP	VOUTP Discharge Current	Vcc=3.7V, Voutp=0.1V		1.1		mA
tssp	CH1 Soft-start Time	Vcc=3.7V	1.6	2.4	3.0	ms
RLXP	LXP Pin On-resistance	Vcc=3.7V		400		mΩ
RSYNCP	Synchronous SW Pch.On- resistance	Vcc=3.7V		700		mΩ
[R1286K	0xxx]					
ILIMLXP	LXP Pin Limit Current	Vcc=3.7V		1.0		Α
[R1286K	1xxX]					
ILIMLXP	LXP Pin Limit Current	Vcc=3.7V		1.1		Α
[R1286K	xxxX]					
Voutp	VOUTP Voltage Tolerance	Vcc=3.7V	×0.991	Vset	×1.009	V

⁽¹⁾ X : A to N (Provided, except "B" and "I")

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The specifications surrounded by \square are guaranteed by Design Engineering at - 40°C \leq Ta \leq 85°C.

R1286K EI	ectrical Characteristics (Co	ntinued)			(Ta =	= 25°C
Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit
[R1286K	001B]					
VFBP	VFBP Voltage Tolerance	Vcc=3.7V	0.985	1.000	1.015	V
IFBP	VFBP Input Current	Vcc=5.5V, VFBP=0V or 5.5V	-0.1		0.1	μA
Inverti	ing DC/DC Converter (CH2)					
Maxduty2	Maximum Duty Cycle 2	Vcc=3.7V		90		%
IVOUTN	VOUTN Discharge Current	Vcc=3.7V, Voutn=-0.1		0.3		mA
R _{LXN}	LXN Pin On-resistance	Vcc=3.7V		400		mΩ
RSYNCN	Synchronous SW Nch.On- resistance	Vcc=3.7V		600		mΩ
[R1286K	0xxx]					
	LXN Pin Limit Current	Vcc=3.7V		1.5		А
[R1286K	1xxX]					
LIMLXN	LXN Pin Limit Current	Vcc=3.7V		1.8		А
[R1286K	xxxX]					
VONDEF	Voutn Default Voltage Tolerance	V _{CC} =3.7V, selectable between V _{ONMIN} and V _{ONMAX} at shipping	V _{SET} -70	VSET	V _{SET} +70	mV
Vonmin	Voutn Minimum Voltage Tolerance	Vcc=3.7V, selectable between -2.0V and -3.0V at shipping	V _{SET} -70	V _{SET}	V _{SET} +70	mV
V _{ONMAX}	Voute Maximum Voltage Tolerance	Vcc=3.7V	V _{SET} -70	V _{ONMIN} + 3.0V	V _{SET} +70	mV
Voutn	Vou™ Voltage Tolerance (S-Wire)	Vcc=3.7V (Guaranteed by design engineering)	V _{SET} -80	Vset	V _{SET} +80	mV
tssn	Soft-start Time for CH2	Vcc=3.7V	1.6x Vondef/ -4.9	2.3x Vondef/ -4.9	3.0x Vondef/ -4.9	ms
[R1286K	001B]					
V_{FBN}	VFBN Voltage Tolerance	Vcc=3.7V	-25	0	25	mV
VREF	VREF Voltage Tolerance	Vcc=3.7V	1.18 +V _{FBN}	1.2 +V _{FBN}	1.22 +V _{FBN}	V
I _{FBN}	VFBN Input Current	Vcc=5.5V, V _{FBN} = 0V or 5.5V	-0.1		0.1	μA
tssn	Soft-start Time for CH2	Vcc=3.7V	1.6	2.8	3.6	ms

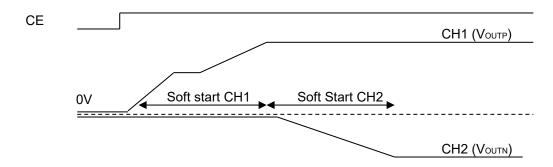
All test items listed under *Electrical Characteristics* are done under the pulse load condition (Tj≈Ta=25°C).

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THEORY OF OPERATION

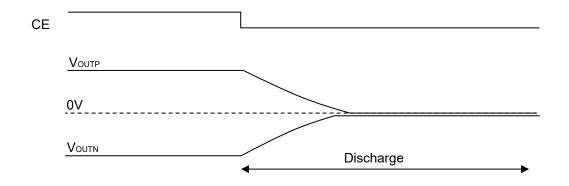
Start-up Sequence

When CE level turns from 'L' to 'H' level, the softstart of CH1 starts the operation. After detecting output voltage of CH1(Voutp) as the nominal level, the soft start of CH2 starts the operation.



Auto Discharge Function

When CE level turns from 'H' to 'L' level, the R1286K goes into standby mode and switching of the outputs of L_{XP} and L_{XN} will stop. Then dischage switsh between V_{OUTN} and GND and switch between V_{OUTP} and GND turn on and discharge the negative output voltage and positive output voltage. The positive and negative output voltage is discharged to 0V in standby mode. If Vcc voltage became lower than UVLO detect voltage , discharge switches also turn on and discharge output voltage(Voutn and VoutP). In case of timer latch protection, discharge switches will keep off .



Thermal Shutdown Protection

If the over temparature is detected, internal Mosfet will turn-off soon. And when the temparature get lower than the release temparature, IC is reset and restart the operation.

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Overcurrent Protection and Short-circuit Protection Circuit Timer

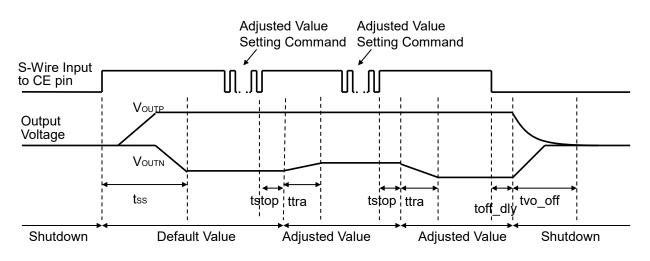
The over current protection circuit supervises the peak current of the inductor (The current passing through NMOS transistor of CH1 and PMOS transistor of CH2) with respect to each switching cycle. If the peak current exceeds the Lx current limit (ILIMLXP or ILIMLXN), the over current protection circuit turns off the NMOS transistor of CH1 or PMOS transistor of CH2. If the over current continues more than the protection delay time (T_{DLY}), the short current protection circuit latches the built-in driver at OFF state and stops the operation of DC/DC converter.

* Lx limit current (I_{LIMLXP} or I_{LIMLXN}) and the protection delay time (T_{DLY}) can be easily affected by self-heating and ambient environment. The drastic drop of output voltage or the unstable output voltage caused by the short-circuiting may affect the protection operation and the delay time.

To release the latch over current protection, reset the IC by inputting "L" into CE pin or by making the input voltage lower than the UVLO detector threshold (V_{UVL01}).

During the softstart operation of CH1 and CH2, the timer operates until detecting output voltage of CH2 (VouTN) as the nominal level. Therefore, even if the softstart cannot finish correctly because of the short circuit, the protection timer function will be able to work correctly.

Sequence with S-Wire Control for VOUTN (R1286KxxxX⁽¹⁾)



Default Value Driving

 V_{OUTP} rises up first and secondarily V_{OUTN} goes down. In this time V_{OUTN} is set V_{ONDEF} . Soft-start time (tss) =2.4ms + 2.3 x V_{ONDEF} / -4.9 (Typ.)

⁽¹⁾ X : A to N (Provided, except "B" and "I")

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Adjusted Value Driving

After receiving the adjusted value setting command, V_{OUTN} is changed to the target voltage in multiple steps method. Adjusted value is also selectable with pulse count (Please refer to V_{OUTN} VARIABLE TABLE).

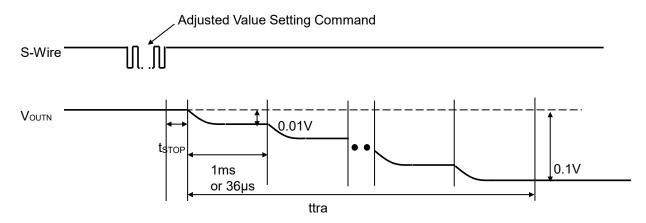
In the case of R1286KxxxA/C/D/E/F/G,

VOUTN change 0.01V step in every 1ms and it takes 10ms per 0.1V that is minimum step for VOUTN setting value.

In the case of R1286KxxxH/J/K/L/M/N,

VOUTN change 0.01V step in every 36us and it takes 360us per 0.1V that is minimum step for VOUTN setting value.

[Multiple steps method (In case of $\Delta V_{OUT} = 0.1V$)]



[•] Multiple step rate : 0.01V / 1ms or 36µs

⁻ Transient time (ttra) for minimum ΔVoutN: 10 ms or 0.36 ms

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VOUTN Variable Table

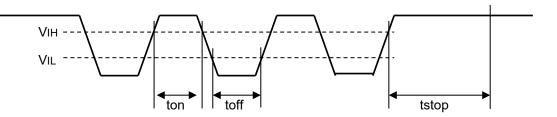
The adjusted value setting command are operated with S-Wire input (pulse count) as the following table.

BIT		D1000KwwwO
(Pulse Count)	R1286KxxxA	R1286KxxxG
0 (Default)	-2.4 to -5.4	-3.0 to -6.0
1	-5.4	-6.0
2	-5.3	-5.9
3	-5.2	-5.8
4	-5.1	-5.7
5	-5.0	-5.6
6	-4.9	-5.5
7	-4.8	-5.4
8	-4.7	-5.3
9	-4.6	-5.2
10	-4.5	-5.1
11	-4.4	-5.0
12	-4.3	-4.9
13	-4.2	-4.8
14	-4.1	-4.7
15	-4.0	-4.6
16	-3.9	-4.5
17	-3.8	-4.4
18	-3.7	-4.3
19	-3.6	-4.2
20	-3.5	-4.1
21	-3.4	-4.0
22	-3.3	-3.9
23	-3.2	-3.8
24	-3.1	-3.7
25	-3.0	-3.6
26	-2.9	-3.5
27	-2.8	-3.4
28	-2.7	-3.3
29	-2.6	-3.2
30	-2.5	-3.1
31	-2.4	-3.0

Voutn Variable Table (31 steps)

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Timing Chart for Commands with S-Wire

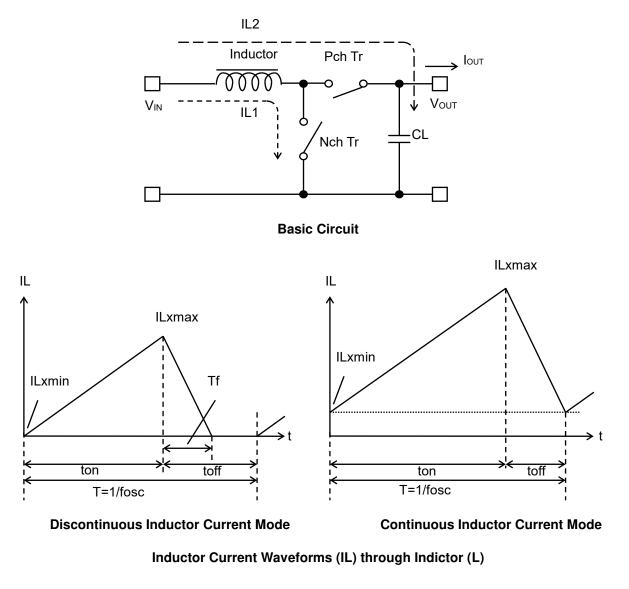


Timing specification

Item	Symbol	Min.	Тур.	Max.	Unit
Soft-start time	tss		tssp + tssn		ms
V _{OUTN} Transient time (1 step)	ttra		10 (R1286KxxxA/C/D/E/F/G) 0.36 (R1286KxxxH/J/K/L/M/N)		ms
Turn-off delay time	toff_dly	70	90	110	μs
Vout discharge time	tvo_off		2.0		ms
CE pin input voltage, high	VIH	1.2			V
CE pin input voltage, low	VIL			0.4	V
S-Wire time, high	ton	2	10	20	μs
S-Wire time, low	toff	2	10	20	μs
S-Wire command stop time	tstop	70	90	110	μS

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Operation of Set-up DC/DC Converter (CH1) and Output Current



The PWM control type of CH1 has two operation modes characterized by the continuity of inductor current: discontinuous inductor current mode and continuous inductor current mode. When a NMOS Tr. is in On-state, the voltage to be applied to the inductor (L) is described as V_{IN} . An increase in the inductor current (IL1) can be written as follows:

IL1 = V_{IN} x ton / L·····Equation 1

In the CH1 circuit, the energy accumulated during the On-state is transferred into the capacitor even in the Offstate. A decrease in the inductor current (IL2) can be written as follows: $IL2 = (V_{OUT} - V_{IN}) \times tf / L$ Equation 2

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In the PWM control, IL1 and IL2 become continuous when tf = toff, which is called continuous inductor current mode. When the device is in continuous inductor current mode and operates in steady-state conditions, the variations of IL1 and IL2 are same:

$V_{IN} x \text{ ton } / L = (V_{OUT} - V_{IN}) x \text{ toff } / L \cdots Equation 3$
Therefore, the duty cycle in continuous inductor current mode is:
Duty = ton / (ton + toff) = (V _{OUT} - V _{IN}) / V _{OUT} ·····Equation 4
If the input voltage (V _{IN}) is equal to V _{OUT} , the output current (I _{OUT}) is:
$I_{OUT} = V_{IN}^2 x \text{ ton } / (2 x L x V_{OUT}) \cdots Equation 5$
If I_{OUT} is larger than Equation 5, the device switches to continuous inductor current mode. The L _x peak current flowing through L (ILxmax) is:
ILxmax = I _{OUT} x V _{OUT} / V _{IN} + V _{IN} x ton / (2 x L) ·····Equation 6

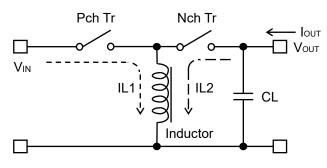
ILxmax = Iout x Vout / VIN + VIN x T x (Vout - VIN) / (2 x L x Vout) ······Equation 7

The L_X peak current limit circuit operates in both modes if the ILxmax becomes more than the L_X peak current limit. When considering the input and output conditions or selecting the external components, please pay attention to ILxmax.

Notes: The above calculations are based on the ideal operation of the device. They do not include the losses caused by the external components or L_X switch. The actual maximum output current will be 70% to 90% of the above calculation results. Especially, if IL is large or V_{IN} is low, it may cause the switching losses.

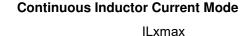
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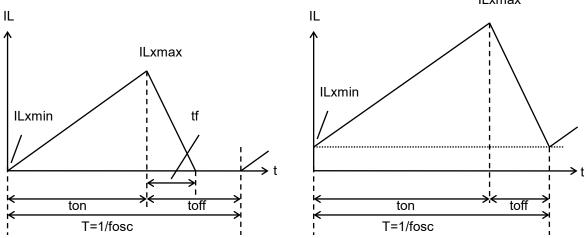
Operation of Inverting DC/DC Converter (CH2) and Output Current



Basic Circuit

Discontinuous Inductor Current Mode





Inductor Current Waveforms (IL) through Indictor (L)

The PWM control type of CH2 has two operation modes characterized by the continuity of inductor current: discontinuous inductor current mode and continuous inductor current mode.

When a PMOS Tr. is in ON-state, the voltage to be applied to the inductor (L) is described as $V_{\mathbb{N}}$. An increase in the inductor current (IL1) can be written as follows:

IL1 = V_{IN} x ton / L·····Equation 8

In the CH2 circuit, the energy accumulated during the On-state is transferred into the capacitor even in the Offstate. A decrease in the inductor current (IL2) can be written as follows:

IL2 = |V_{OUT}| x tf / L·····Equation 9

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In the PWM control type, when tf = toff, the inductor current will be continuous and the operation of CH2 will be continuous inductor current mode. When the device is in continuous inductor current mode and operates in steady-state conditions, the variation of IL1 and IL2 are same: $V_{IN} x \text{ ton } / L = |V_{OUT}| x \text{ toff } / L \dots$ Equation 10 Therefore, the duty cycle in continuous inductor current mode is:

 $Duty = ton / (ton + toff) = |V_{OUT}| / (|V_{OUT}| + V_{IN}) \cdots Equation 11$

If the input voltage (V_{IN}) equal to V_{OUT}, the output current (I_{OUT}) is:

 $I_{OUT} = V_{IN}^2 x \text{ ton } / (2 x L x |V_{OUT}|) \cdots$ Equation 12

If I_{OUT} is larger than Equation 12, the device switches to continuous inductor current mode. The L_X peak current flowing through L (ILxmax) is:

 $ILxmax = I_{OUT} x (|V_{OUT}| + V_{IN}) / V_{IN} + V_{IN} x \text{ ton } / (2 x L) \cdots Equation 13$

ILxmax = Iout x (|Vout| + VIN) / VIN + VIN x |Vout| x T / { 2 x L x (|Vout| + VIN) } Equation 14

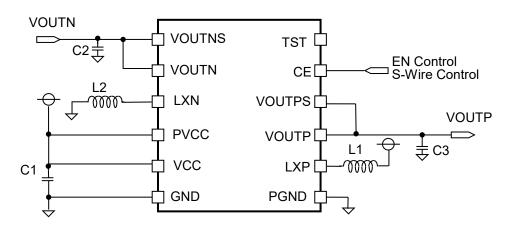
The L_x peak current limit circuit operates in both modes if the ILxmax becomes more than the L_x peak current limit. When considering the input and output conditions or selecting the external components, please pay attention to ILxmax.

Notes: The above calculations are based on the ideal operation of the device. They do not include the losses caused by the external components or L_X switch. The actual maximum output current will be 70% to 90% of the above calculation results. Especially, if IL is large or V_{IN} is low, it may cause the switching losses.

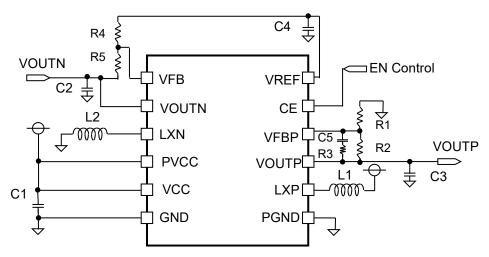
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APPLICATION INFORMATION

Typical Application Circuits



R1286KxxxX (Fixed Output Voltage Type) Typical Application Circuit



R1286K001B (Adjustable Output Voltage Type) Typical application Circuit

Symbol	Description
L1	VLF302510M-4R7M (TDK)、VLF3010S-4R7M (TDK)
L2	VLF4012S-4R7M (TDK)、NR4012T4R7M (TAIYOYUDEN)
C1(CIN), C2(COUTN), C3(COUTP)	4.7μF、 2012size X5R T=0.85max
C4 (C _{REF}) ⁽¹⁾	0.1µF、0603size

(1) R1286K001B Only

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Precautions for Selecting External Components

- Place a ceramic capacitor of 4.7µF or more (C1) between VCC pin/PVCC pin and GND pin/ PGND pin.
- Place a ceramic capacitor of 4.7µF or more (C2, C3) between VOUTP pin / VOUTN pin and GND.
- Place a ceramic capacitor of 0.1µF to 2.2µF (C4) between VREF pin and GND. [R1286K001B]
- Step-up DC/DC Converter Output Voltage Setting [R1286K001B] The output voltage Voute of the step-up DC/DC converter is controlled with maintaining the VFBP as 1.0V.
 Voute can be set with adjusting the values of R1 and R2 as in the next formula.

 $V_{OUTP} = V_{FBP} \times (R1 + R2) / R1$

VOUTP can be set from 4.6V to 5.8V. The appropriate value range of R1 is from $20k\Omega$ to 60k Ω .

• Inverting DC/DC Converter Output Voltage Setting [R1286K001B]

The output voltage V_{OUTN} of the inverting DC/DC converter is controlled with maintaining the V_{FBN} as 0V. V_{OUTN} can be set with adjusting the values of R1 and R2 as in the next formula.

Voutn = Vfbn - (Vref - Vfbn) \times R5 / R4

VOUTN can be set from -2.0V to -6.0V. The appropriate value range of R4 is from $2.5k\Omega$ to $60k\Omega$.

• Phase Compensation of Step-up DC/DC Converter [R1286K001B]

DC/DC converter's phase may lose 180 degree by external components of L and C and load current. Because of this, the phase margin of the system will be less and the stability will be worse. Therefore, the phase must be gained.

Zero will be formed with R1 and C5. C5 [pF] = 300 / R1 [k Ω]

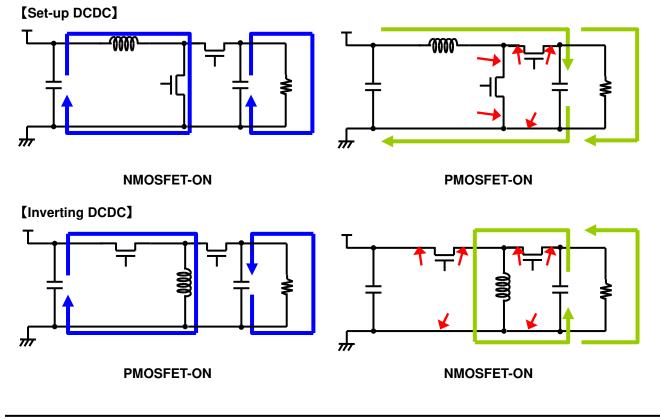
If the noise of the system is large, the output noise affects the feedback and the operation may be unstable. In that case, another resistor R3 will be set. The appropriate value range is from 10Ω to $1k\Omega$.

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TECHNICAL NOTES

The performance of a power source circuit using this device is highly dependent on a peripheral circuit. A peripheral component or the device mounted on PCB should not exceed a rated voltage, a rated current or a rated power. When designing a peripheral circuit, please be fully aware of the following points.

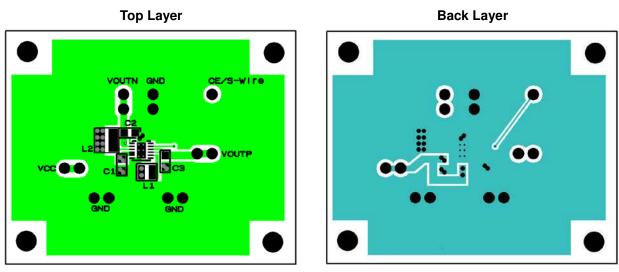
- Wire the bypass capacitor (C1) between the VCC pin, the GND pin, or the PVCC pin as short as possible. The GND pin should be connected to the GND plane of the PCB.
- Wire the GND of the output capacitors (C2, C3) to the GND pin of the device as short as possible.
- The wiring among each GND line of C1, C2, and C3 and the GND pin of the device must be short as possible via the device.
- The wiring between L_{XP} pin, L_{XN} pin and inductor each should be as short as possible and mount output capacitors (C2 and C3) as close as possible to the V_{OUTP}, V_{OUTN} each.
- Input impedance of VOUTPS pin, VOUTNS pin, VFBP pin, and VFBN pin is high, therefore, the external noise may affect the performance. The coupling capacitance between these nodes and switching lines must be as short as possible.
- As shown in the diagrams of the current paths of boost DC/DC converter and the current path of inverting DC/DC converter, the parasitic impedance, inductance, and the capacitance in the parts pointed with red arrows have an influence against the stability of the DC/DC converters and become a cause of the noise. Therefore, such parasitic elements must be made as small as possible. Wiring of the current paths must be short and thick.



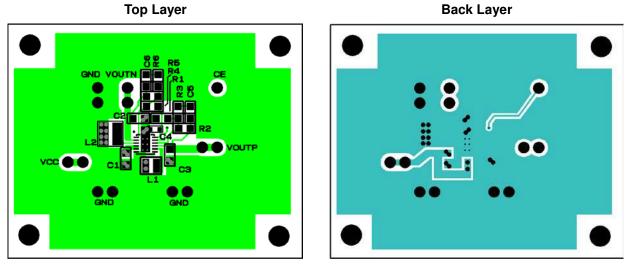
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PCB Layout

R1286K Board Layout [PKG: DNF (PL) 2730-12]



R1286KxxxX⁽¹⁾ (Fixed Output Voltage Type) Board Layout



R1286K001B (Adjustable Output Voltage Type) Board Layout

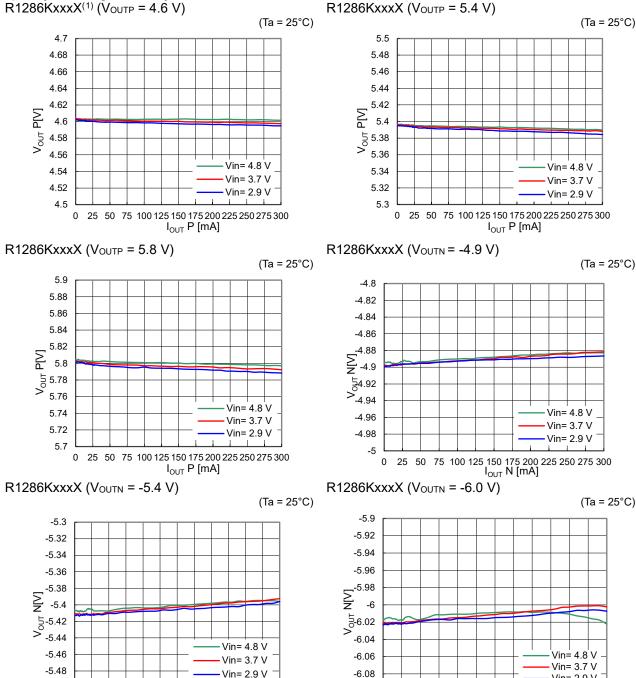
⁽¹⁾ X : A to N (Provided, except "B" and "I")

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TYPICAL CHARACTERISTICS

Typical Characteristics are intended to be used as reference data, they are not guaranteed.

1) Output Voltage vs. Output Current



⁽¹⁾X : A to N (Provided, except "B" and "I")

25 50 75 100 125 150 175 200 225 250 275 300

I_{OUT} N [mA]

-5.5

0

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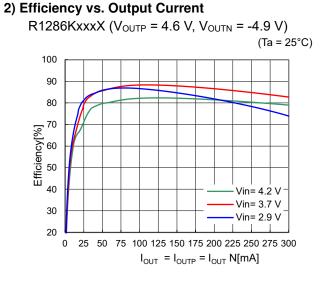
-6.1

0

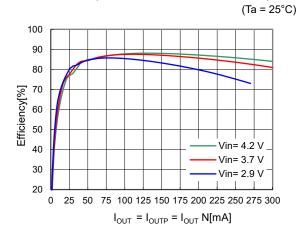
Vin= 2.9 V

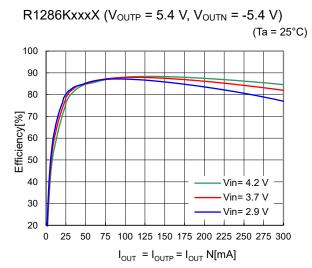
25 50 75 100 125 150 175 200 225 250 275 300 $I_{\rm OUT}^{}$ N [mA]

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R1286KxxxX (V_{OUTP} = 5.8 V, V_{OUTN} = -6.0 V)

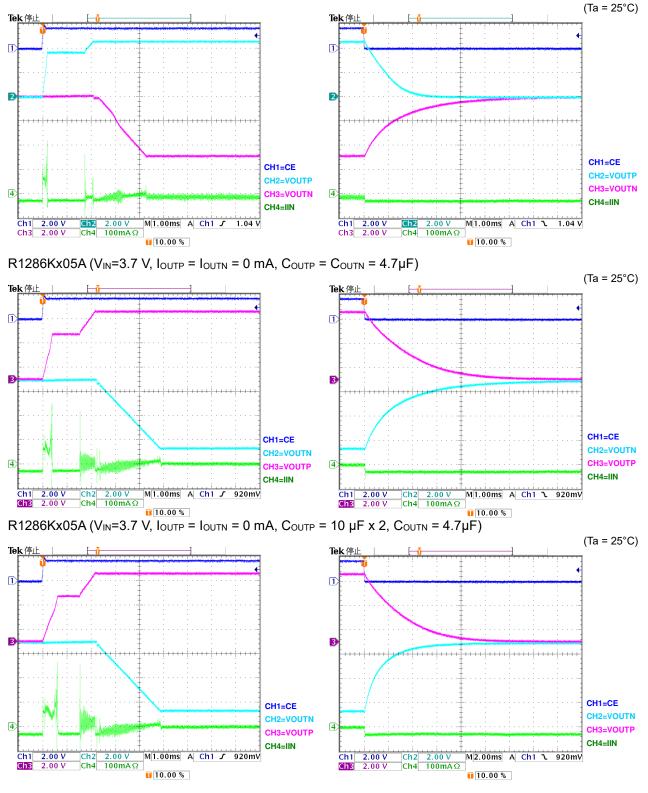




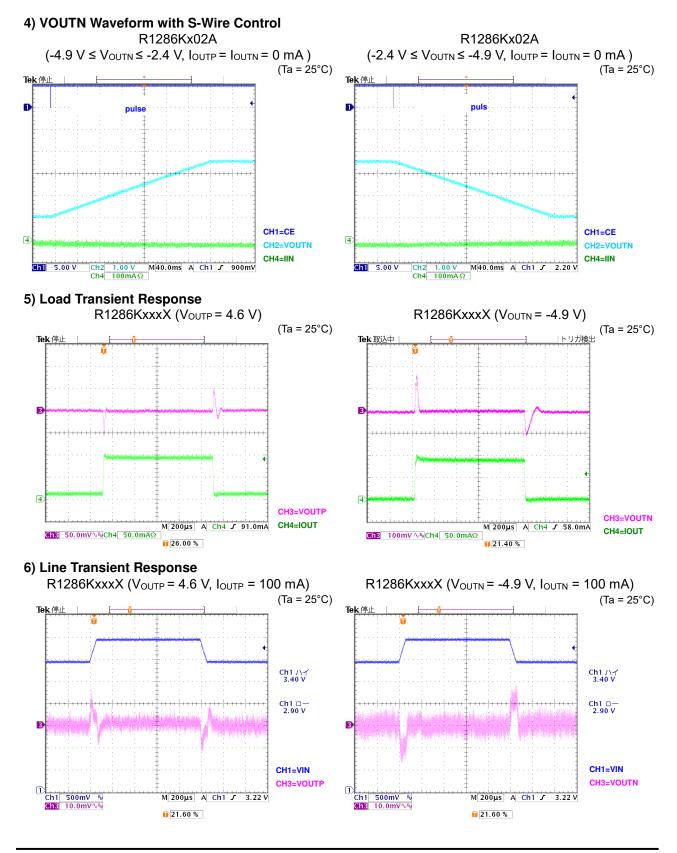
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3) Turn-on/Turn-off Waveform by CE

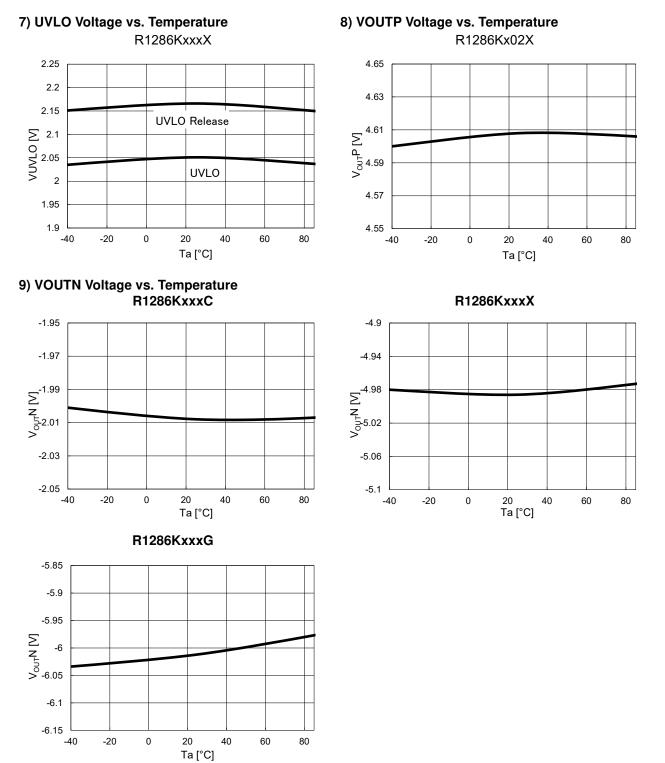
R1286Kx02A (V_{IN} =3.7 V, I_{OUTP} = I_{OUTN} = 0 mA)



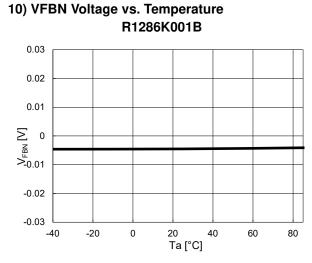
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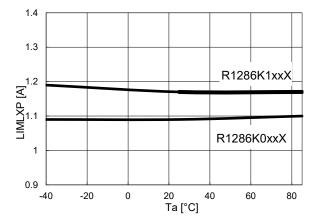
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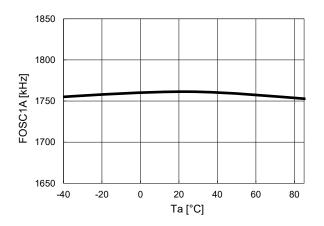
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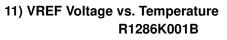


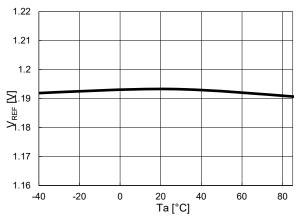




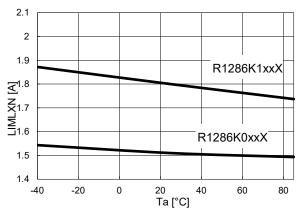
14) Oscillator Frequency vs. Temperature R1286KxxxX



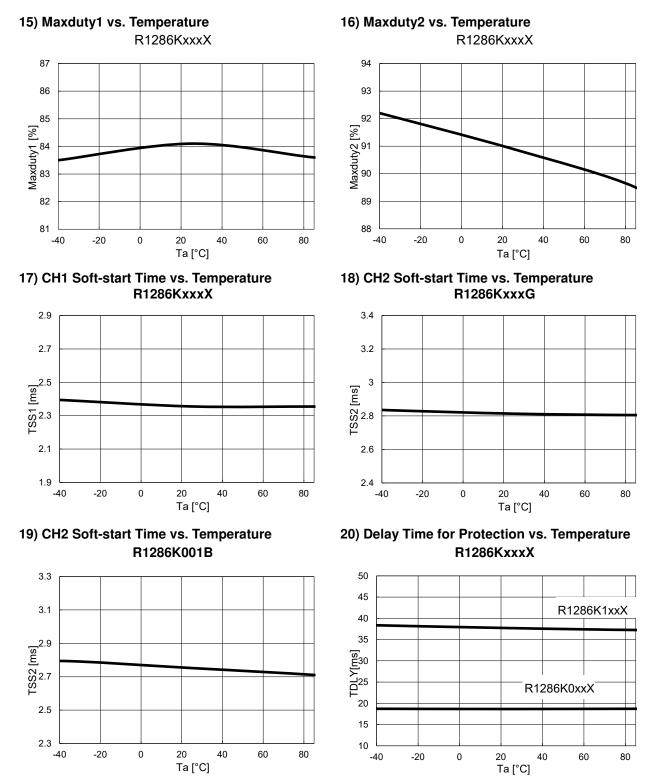




13) LXN Limit Current vs. Temperature R1286KxxxX



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POWER DISSIPATION

DFN(PL)2730-12

Ver. A

The power dissipation of the package is dependent on PCB material, layout, and environmental conditions. The following measurement conditions are based on JEDEC STD. 51-7.

Measurement Conditions

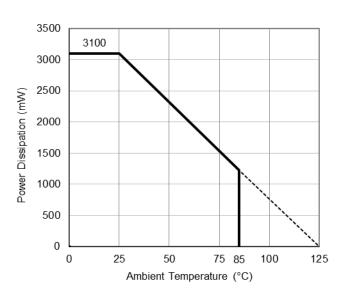
Item	Measurement Conditions
Environment	Mounting on Board (Wind Velocity = 0 m/s)
Board Material	Glass Cloth Epoxy Plastic (Four-Layer Board)
Board Dimensions	76.2 mm × 114.3 mm × 0.8 mm
Copper Ratio	Outer Layer (First Layer): Less than 95% of 50 mm Square Inner Layers (Second and Third Layers): Approx. 100% of 50 mm Square Outer Layer (Fourth Layer): Approx. 100% of 50 mm Square
Through-holes	φ 0.3 mm × 23 pcs

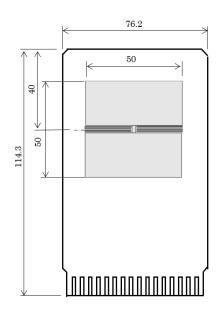
Measurement Result

(Ta = 25°C, Tjmax = 125°C) Item **Measurement Result Power Dissipation** 3100 mW Thermal Resistance (θja) θ ja = 32°C/W Thermal Characterization Parameter (wjt) $\psi jt = 8^{\circ}C/W$

θja: Junction-to-Ambient Thermal Resistance

wit: Junction-to-Top Thermal Characterization Parameter





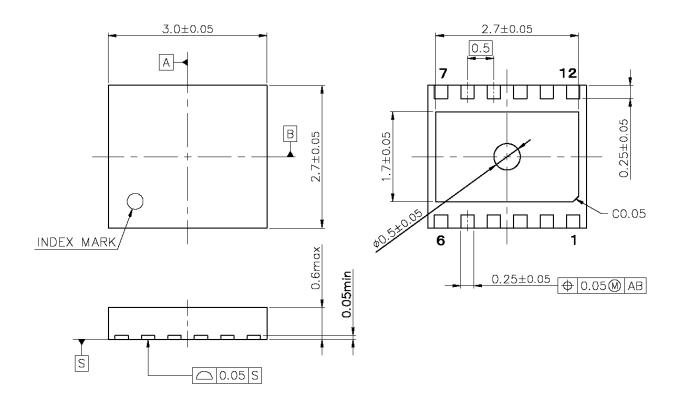
Power Dissipation vs. Ambient Temperature

Measurement Board Pattern

PACKAGE DIMENSIONS

DFN(PL)2730-12

DM-DFN(PL)2730-12-JE-B



DFN(PL)2730-12 Package Dimensions (Unit: mm)

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