

# CIPOS™ Micro IPM 600 V, 4 A

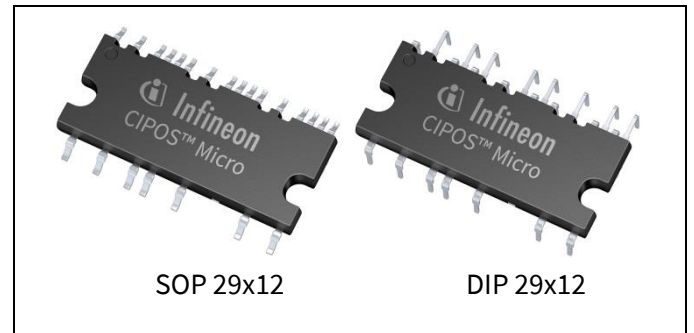
## IM241 Series

### Description

IM241-M6 is a 3-phase Intelligent Power Module (IPM) designed for high-efficiency appliance motor drives such as fans and pumps. This IPM is available in both fast and slow speeds for low loss and low EMI operation respectively.

### Features

- 600V 3-phase inverter including gate drivers & bootstrap function
- Reverse Conducting IGBT Gen 2 (RCD2) optimized for motor drives
- Temperature monitor
- Accurate overcurrent shutdown ( $\pm 5\%$ )
- Fault reporting and programmable fault clear
- Advanced input filter with shoot-through protection
- Optimized  $dV/dt$  for loss and EMI trade offs
- Open-emitter for single and leg-shunt current sensing
- 3.3V logic compatible
- Isolation 2000VRMS, 1min



### Potential Applications

- Fans
- Pumps

### Product validation

Qualified for industrial applications according to the relevant tests of JEDEC47/20/22.

**Table 1** Product Information

Base Part Number	Package Type	Standard Pack	
		Form	Quantity
IM241-M6T2y	DIP 29x12	Tube	240
IM241-M6S1y	SOP 29x12	Tube	240
		Tape & Reel	500

y = B (fast speed for low losses; for x = S, M, L) or J (slow speed for low EMI; for x = S, M)

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**Description1**

**Features 1**

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# 1 Internal Electrical Schematic

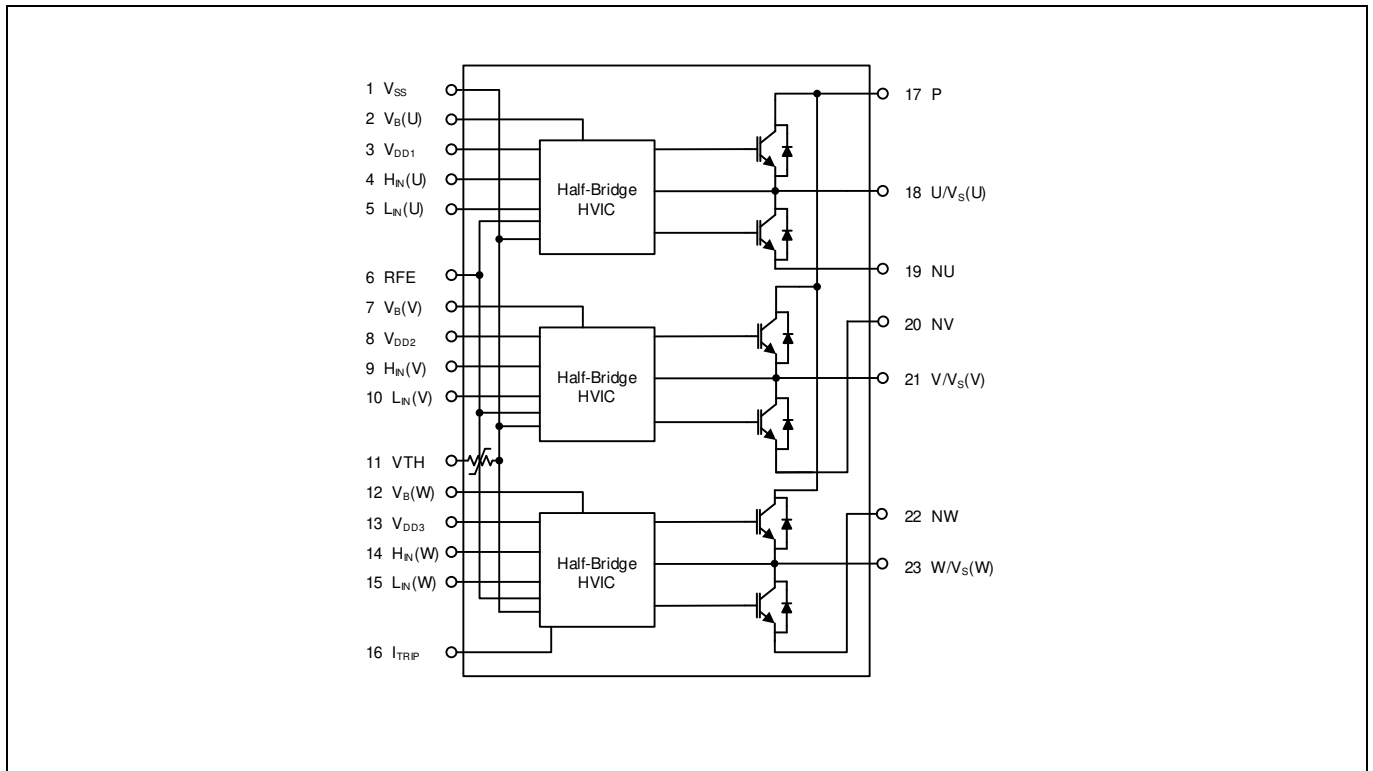


Figure 1 Internal electrical schematic.

## 2 Pin Configuration

### 2.1 Pin Assignment

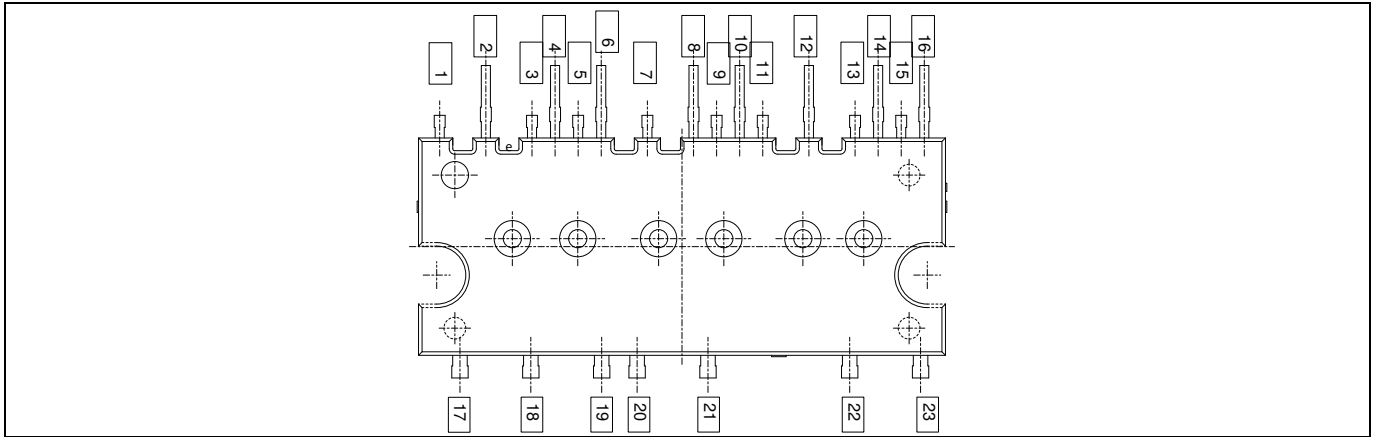


Figure 2 Module pinout

Table 2 Pin Assignment

Pin	Name	Description
1	V <sub>SS</sub>	Logic ground
2	V <sub>B(U)</sub>	U-phase high-side floating IC supply voltage
3	V <sub>DD1</sub>	Low-side control supply 1
4	H <sub>IN(U)</sub>	U-phase high-side gate driver input
5	L <sub>IN(U)</sub>	U-phase low-side gate driver input
6	RFE	RCIN / Fault / Enable
7	V <sub>B(V)</sub>	V-phase high-side floating IC supply voltage
8	V <sub>DD2</sub>	Low-side control supply 2
9	H <sub>IN(V)</sub>	V-phase high-side gate driver input
10	L <sub>IN(V)</sub>	V-phase low-side gate driver input
11	V <sub>TH</sub>	Thermistor output
12	V <sub>B(W)</sub>	W-phase high-side floating IC supply voltage
13	V <sub>DD3</sub>	Low-side control supply 3
14	H <sub>IN(W)</sub>	W-phase high-side gate driver input
15	L <sub>IN(W)</sub>	W-phase low-side gate driver input
16	I <sub>TRIP</sub>	Over-current protection input
17	P	DC bus voltage positive
18	U/V <sub>S(U)</sub>	Motor U-phase output, U-phase high-side floating IC supply offset voltage
19	NU	U-phase low-side emitter
20	NV	U-phase low-side emitter - phase low-side emitter
21	V/V <sub>S(V)</sub>	Motor V-phase output, V-phase high-side floating IC supply offset voltage
22	NW	W-phase low-side emitter
23	W/V <sub>S(W)</sub>	Motor W-phase output, W-phase high-side floating IC supply offset voltage

## 2.2 Pin Descriptions

### $H_{IN}(U,V,W)$ and $L_{IN}(U,V,W)$ (Low side and high side control pins)

These pins are positive logic and they are responsible for the control of the integrated IGBT. The Schmitt-trigger input thresholds of them are such to guarantee LSTTL and CMOS compatibility down to 3.3V controller outputs. Pull-down resistor of about 800kΩ is internally provided to pre-bias inputs during supply start-up and an ESD diode is provided for pin protection purposes. Input Schmitt-trigger and noise filter provide beneficial noise rejection to short input pulses.

The noise filter suppresses control pulses which are below the filter time  $T_{FILIN}$ . The filter acts according to Figure 4.

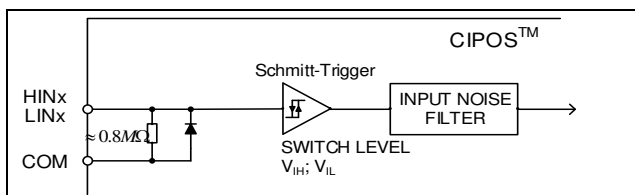


Figure 3 Input pin structure

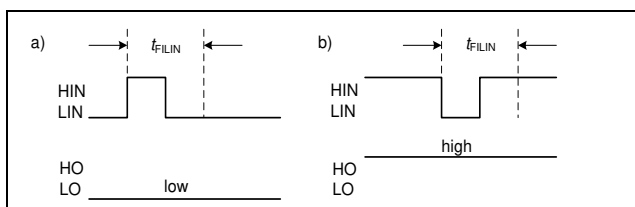


Figure 4 Input filter timing diagram

The integrated gate drive provides additionally a shoot through prevention capability which avoids the simultaneous on-state of the high-side and low-side switch of the same inverter phase. A minimum deadtime insertion of typically 300ns is also provided by driver IC, in order to reduce cross-conduction of the external power switches.

### $V_{DDX}$ , $V_{SS}$ (Low side control supply and reference)

$V_{DD}$  is the control supply and it provides power both to input logic and to the output power stage. Input logic is referenced to  $V_{SS}$  ground.

The under-voltage circuit enables the device to operate at power on when a supply voltage of at least a typical voltage of  $V_{DDUV+} = 11.1V$  is present.

The IC shuts down all the gate drivers power outputs, when the  $V_{DD}$  supply voltage is below  $V_{DDUV-} = 10.9V$ . This prevents the external power switches from critically low gate voltage levels during on-state and therefore from excessive power dissipation.

### $V_B(U,V,W)$ and $V_S(U,V,W)$ (High side supplies)

$V_B$  to  $V_S$  is the high side supply voltage. The high side circuit can float with respect to  $V_{SS}$  following the external high side power device source voltage.

Due to the low power consumption, the floating driver stage is supplied by integrated bootstrap circuit.

The under-voltage detection operates with a rising supply threshold of typical  $V_{BSUV+} = 11.1V$  and a falling threshold of  $V_{BSUV-} = 10.9V$ .

$V_S(U,V,W)$  provide a high robustness against negative voltage in respect of  $V_{SS}$ . This ensures very stable designs even under rough conditions.

### NU, NV, NW (Low side emitters)

The low side emitters are available for current measurements of each phase leg. It is recommended to keep the connection to pin  $V_{SS}$  as short as possible in order to avoid unnecessary inductive voltage drops.

### VTH (Thermistor output)

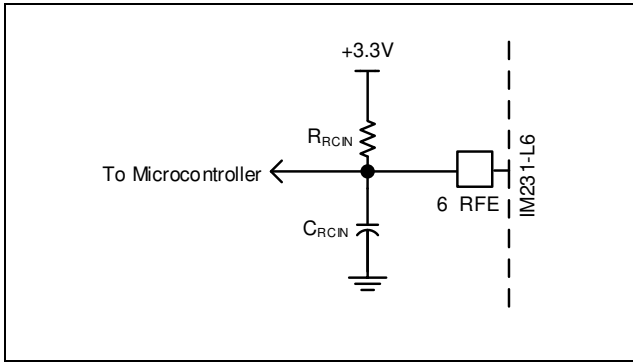
A UL certified NTC resistor is integrated in the module with one terminal of the chip connected to  $V_{SS}$  and the other to VTH. When pulled up to a rail voltage such as  $V_{DD}$  or 3.3V by a resistor, the VTH pin provides an analog voltage signal corresponding to the temperature of the thermistor.

### RFE (RCIN / Fault / Enable)

The RFE pin combines 3 functions in one pin: RCIN or RC-network based programmable fault clear timer, fault output and enable input.

The RFE pin is normally connected to an RC network on the PCB per the schematic in Figure 5. Under normal operating conditions,  $R_{RCIN}$  pulls the RFE pin to 3.3V, thus enabling all the functions in the IPM. The microcontroller can pull this pin low to disable the IPM functionality. This is the Enable function.

Pin Configuration



**Figure 5** Typical PCB circuit connected to the RFE pin

The Fault function allows the IPM to report a Fault condition to the microcontroller by pulling the RFE pin low in one of two situations. The first is an under-voltage condition on V<sub>DD</sub> and the second is when the I<sub>TRIP</sub> pin sees a voltage rising above V<sub>IT,TH+</sub>.

The programmable fault clear timer function provides a means of automatically re-enabling the module operation a preset amount of time (T<sub>FLT-CLR</sub>) after the fault condition has disappeared. Figure 6 shows the RFE-related circuit block diagram inside the IPM.

The length of T<sub>FLT-CLR</sub> can be determined by using the formula below.

$$V_{RFE}(t) = 3.3V * (1 - e^{-t/RC})$$

$$T_{FLT-CLR} = -R_{RCIN} * C_{RCIN} * \ln(1 - V_{RFE+}/3.3V)$$

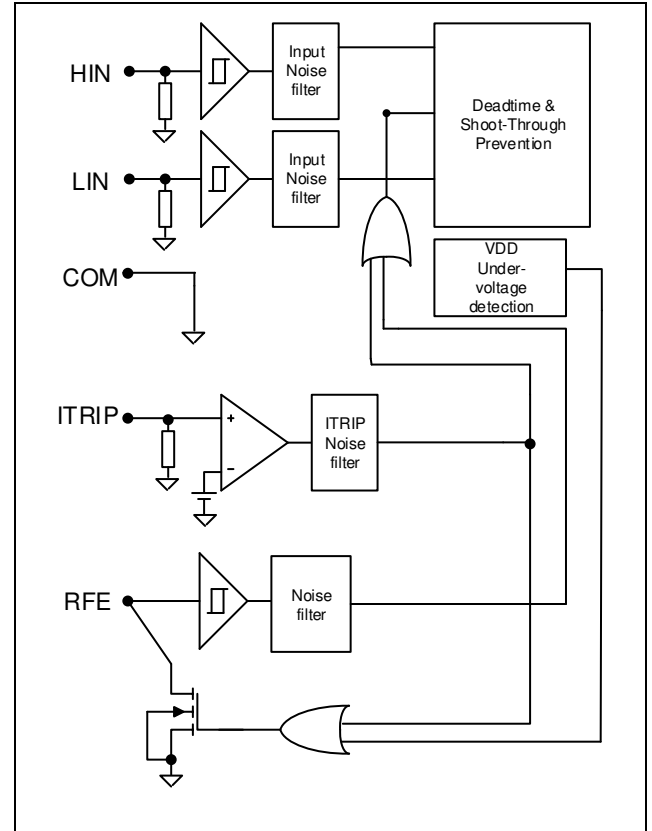
For example, if R<sub>RCIN</sub> is 1.2MΩ and C<sub>RCIN</sub> is 1nF, the T<sub>FLT-CLR</sub> is about 1.7ms with V<sub>RFE+</sub> of 2.2V. It is also important to note that C<sub>RCIN</sub> needs to be minimized in order to make sure it is fully discharged in case of over current event.

Since the I<sub>TRIP</sub> pin has a 500ns input filter, it is appropriate to ensure that C<sub>RCIN</sub> will be discharged below V<sub>RFE-</sub> by the open-drain MOSFET, after 350ns. Therefore, the max C<sub>RCIN</sub> can be calculated as:

$$V_{RFE}(t) = 3.3V * e^{-t/RC} < V_{RFE-}$$

$$C_{RCIN} < 350ns / (- \ln(V_{RFE-} / 3.3V)) * R_{RFE\_ON}$$

Consider V<sub>RFE-</sub> of 0.8V and R<sub>RFE\_ON</sub> of 50ohm, C<sub>RCIN</sub> should be less than 5nF. It is also suggested to use a R<sub>RCIN</sub> of between 0.5MΩ and 2MΩ.



**Figure 6** RFE internal circuit structure

**U/V<sub>s</sub>(U) , V/V<sub>s</sub>(V), W/V<sub>s</sub>(W) (High side emitter and low side collector)**

These pins are connected to motor U, V, W input pins.

**P (Positive bus input voltage)**

The high side IGBTs are connected to the bus voltage. It is noted that the bus voltage should not exceed 450V.

Absolute Maximum **Rating**

### 3 Absolute Maximum Rating

#### 3.1 Module

**Table 3**

Parameter	Symbol	Condition		Units
Storage temperature	$T_{STG}$		-40 ~ 150	°C
Operating case temperature	$T_C$		-40 ~ 125	°C
Operating junction temperature	$T_J$		-40 ~ 150	°C
Isolation test voltage	$V_{ISO}$	1min, RMS, f = 60Hz	2000	V

#### 3.2 Inverter

**Table 4 IM241-M6**

Parameter	Symbol	Condition		Units
Max. blocking voltage	$V_{CES}/V_{RRM}$		600	V
Output current	$I_O$	$T_C = 25^\circ\text{C}$	4	A
Peak output current	$I_{OP}$	$T_C = 25^\circ\text{C}, t_p < 1\text{ms}$	6	A
Peak power dissipation per IGBT	$P_{tot}$	$T_C = 25^\circ\text{C}$	13	W
Short circuit withstand time	$T_{SC}$	$V_{DD}=15\text{V}, V_{DC} \leq 400\text{V}, T_J=150^\circ\text{C}$ Allowed number of short circuits: <1000, time between short circuit: $\geq 1\text{s}$	3	$\mu\text{s}$

#### 3.3 Control

**Table 5**

Parameter	Symbol	Condition		Units
Low side control supply voltage	$V_{DD}$		-0.3 ~ 20	V
Input voltage	$V_{IN}$	LIN, HIN, $I_{TRIP}$ , RFE	-0.3 ~ $V_{DD}$	V
High side floating supply voltage ( $V_B$ reference to $V_S$ )	$V_{BS}$		-0.3 ~ 20	V

## 4 Thermal Characteristics

**Table 6** IM241-M6

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Units
Single IGBT thermal resistance, junction-case	$R_{TH(J-C)}$	Low side V-phase IGBT	-	9.84	11.8	K/W
Single diode thermal resistance, junction-case	$R_{TH(J-C)}$	Low side V-phase Diode	-	10.5	12.6	K/W



## 5 Recommended Operating Conditions

**Table 7**

Parameter	Symbol	Min.	Typ.	Max.	Units
Positive DC bus input voltage	P	-	-	450	V
Low side control supply voltage	V <sub>DD</sub>	13.5	-	16.5	V
High side floating supply voltage	V <sub>BS</sub>	12.5	-	17.5	V
Input voltage (L <sub>IN</sub> ,H <sub>IN</sub> ,I <sub>TRIP</sub> ,RFE)	V <sub>IN</sub>	0	-	5	V
PWM carrier frequency	F <sub>PWM</sub>	-	20	-	kHz
External dead time between H <sub>IN</sub> & L <sub>IN</sub>	DT	1	-	-	μs
Voltage between V <sub>SS</sub> and N(U,V,W)	V <sub>COMP</sub>	-5	-	5	V
Minimum input pulse width	PW <sub>IN(ON)</sub> , PW <sub>IN(OFF)</sub>	1	-	-	μs

## 6 Static Parameters

### 6.1 Inverter

$(V_{DD}-V_{SS}) = (V_B - V_S) = 15\text{ V}$ .  $T_C = 25^\circ\text{C}$  unless otherwise specified.

**Table 8 IM241-M6 (B and J type)**

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Units
Collector-to-emitter saturation voltage	$V_{CE(sat)}$	$I_C = 0.5\text{A}$	-	1.13	-	V
		$I_C = 1\text{A}$	-	1.37	1.58	V
		$I_C = 1\text{A}, T_J = 150^\circ\text{C}$	-	1.36	-	V
Collector emitter leakage current of high side IGBT	$I_{LKH}$	$V_{IN} = 0\text{V}, V_+ = 600\text{V}$	-	-	80	$\mu\text{A}$
		$V_{IN} = 0\text{V}, V_+ = 600\text{V}, T_J = 150^\circ\text{C}$	-	9.8	-	$\mu\text{A}$
Diode forward voltage	$V_F$	$I_C = 0.5\text{A}$	-	1.18	-	V
		$I_C = 1\text{A}$	-	1.38	1.60	V
		$I_C = 1\text{A}, T_J = 150^\circ\text{C}$	-	1.34	-	V

### 6.2 Control

$(V_{DD}-V_{SS}) = (V_B - V_S) = 15\text{ V}$ .  $T_C = 25^\circ\text{C}$  unless otherwise specified. The  $V_{IN}$  and  $I_{IN}$  parameters are referenced to  $V_{SS}$  and are applicable to all six channels. The  $V_{DDUV}$  parameters are referenced to  $V_{SS}$ . The  $V_{BSUV}$  parameters are referenced to  $V_S$ .

**Table 9**

Parameter	Symbol	Min.	Typ.	Max.	Units
Logic "1" input voltage (LIN, HIN)	$V_{IN,TH+}$	2.2	-	-	V
Logic "0" input voltage (LIN, HIN)	$V_{IN,TH-}$	-	-	0.8	V
$V_{DD}/V_{BS}$ supply undervoltage, positive going threshold	$V_{DD,UV+}, V_{BS,UV+}$	10.6	11.1	11.6	V
$V_{DD}/V_{BS}$ supply undervoltage, negative going threshold	$V_{DD,UV-}, V_{BS,UV-}$	10.4	10.9	11.4	V
$V_{DD}/V_{BS}$ supply undervoltage lock-out hysteresis	$V_{DDUVH}, V_{BSUVH}$	-	0.2	-	V
RFE positive going threshold	$V_{RFE+}$	-	1.9	2.2	V
RFE negative going threshold	$V_{RFE-}$	0.8	1.1	-	V
ITRIP positive going threshold	$V_{IT,TH+}$	0.475	0.500	0.525	V
ITRIP negative going threshold	$V_{IT,TH-}$	-	0.430	-	V
ITRIP input hysteresis	$V_{IT,HYS}$	-	0.07	-	V
Quiescent $V_{BS}$ supply current	$I_{QBS}$	-	-	70	$\mu\text{A}$
Quiescent $V_{DD}$ supply current per channel	$I_{QDD}$	-	-	3	mA
Input bias current $V_{IN}=5\text{V}$ for LIN, HIN	$I_{IN+}$	-	6.25	12.5	$\mu\text{A}$
Input bias current $V_{IN}=5\text{V}$ for RFE	$I_{IN,RFE+}$	-	-	1	$\mu\text{A}$

Static Parameters

<b>Parameter</b>	<b>Symbol</b>	<b>Min.</b>	<b>Typ.</b>	<b>Max.</b>	<b>Units</b>
Input bias current $V_{IN}=5V$ for ITRIP	$I_{ITRIP+}$	-	5	20	$\mu A$
Bootstrap resistance	$R_{BS}$	-	200	-	$\Omega$
RFE low on resistance	$R_{RFE}$	-	34	60	$\Omega$

## 7 Dynamic Parameters

### 7.1 Inverter

$(V_{DD} - V_{SS}) = (V_B - V_S) = 15\text{ V}$ .  $T_C = 25^\circ\text{C}$  unless otherwise specified.

**Table 10 IM241-M6, B Type**

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Units
Input to output turn-on propagation delay	$T_{ON}$	$I_C = 1\text{ A}$ , $V_+ = 300\text{ V}$	-	508	-	ns
Turn-on rise time	$T_R$		-	7.44	-	ns
Turn-on switching time	$T_{C(on)}$		-	52.6	-	ns
Input to output turn-off propagation delay	$T_{OFF}$	$I_C = 1\text{ A}$ , $V_+ = 300\text{ V}$	-	737	-	ns
Turn-off fall time	$T_F$		-	116	-	ns
Turn-off switching time	$T_{C(off)}$		-	101	-	ns
RFE low to six switch turn-off propagation delay	$T_{EN}$	$V_{IN}=0$ or $V_{IN}=5\text{ V}$ , $V_{RFE}=5\text{ V}$	-	420	-	ns
ITRIP to six switch turn-off propagation delay	$T_{ITRIP}$	$V_+ = 300\text{ V}$ , no cap on RFE	-	1.3	-	$\mu\text{s}$
Turn-on slew rate	$dV/dt$	$I_C = 1\text{ A}$ , $V_+ = 300\text{ V}$ , $V_{DD} = 15\text{ V}$ , $L = 6\text{ mH}$	-	6.91	-	V/ns
Turn-on switching energy	$E_{ON}$	$I_C = 1\text{ A}$ , $V_+ = 300\text{ V}$ , $V_{DD} = 15\text{ V}$ , $L = 6\text{ mH}$ , mean of high side and low side	-	19.9	-	$\mu\text{J}$
Turn-off switching energy	$E_{OFF}$		-	13.6	-	
Diode reverse recovery energy	$E_{REC}$		-	16.0	-	
Diode reverse recovery time	$T_{RR}$		-	84.2	-	ns
Turn-on switching energy	$E_{ON}$	$I_C = 1\text{ A}$ , $V_+ = 300\text{ V}$ , $V_{DD} = 15\text{ V}$ , $L = 6\text{ mH}$ , $T_J = 150^\circ\text{C}$ , mean of high side and low side	-	42.4	-	$\mu\text{J}$
Turn-off switching energy	$E_{OFF}$		-	21.2	-	
Diode reverse recovery energy	$E_{REC}$		-	29.8	-	
Diode reverse recovery time	$T_{RR}$		-	129	-	ns

**Table 11 IM241-M6, J Type**

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Units
Input to output turn-on propagation delay	$T_{ON}$	$I_C = 1\text{ A}$ , $V_+ = 300\text{ V}$	-	588	-	ns
Turn-on rise time	$T_R$		-	16.4	-	ns
Turn-on switching time	$T_{C(on)}$		-	188	-	ns
Input to output turn-off propagation delay	$T_{OFF}$	$I_C = 1\text{ A}$ , $V_+ = 300\text{ V}$	-	1220	-	ns
Turn-off fall time	$T_F$		-	163	-	ns
Turn-off switching time	$T_{C(off)}$		-	157	-	ns

Dynamic Parameters

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Units
RFE low to six switch turn-off propagation delay	$T_{EN}$	$V_{IN}=0$ or $V_{IN}=5V$ , $V_{RFE}=5V$	-	490	-	ns
ITRIP to six switch turn-off propagation delay	$T_{ITRIP}$	$V_+ = 300V$ , no cap on RFE	-	1.3	-	$\mu s$
Turn-on slew rate	$dV / dt$	$I_C = 1A$ , $V_+ = 300V$ , $V_{DD} = 15V$ , $L = 6mH$ , mean of high side and low side	-	1.6	-	V/ns
Turn-on switching energy	$E_{ON}$	$I_C = 1A$ , $V_+ = 300V$ , $V_{DD} = 15V$ , $L = 6mH$ , mean of high side and low side	-	59.6	-	$\mu J$
Turn-off switching energy	$E_{OFF}$		-	21.3	-	
Diode reverse recovery energy	$E_{REC}$		-	21.2	-	
Diode reverse recovery time	$T_{RR}$		-	184	-	ns
Turn-on switching energy	$E_{ON}$	$I_C = 1A$ , $V_+ = 300V$ , $V_{DD} = 15V$ , $L = 6mH$ $T_J = 150^\circ C$ , mean of high side and low side	-	94.1	-	$\mu J$
Turn-off switching energy	$E_{OFF}$		-	24.1	-	
Diode reverse recovery energy	$E_{REC}$		-	24.4	-	
Diode reverse recovery time	$T_{RR}$		-	261	-	ns

## 7.2 Control

$(V_{DD} - V_{SS}) = (V_B - V_S) = 15V$ .  $T_C = 25^\circ C$  unless otherwise specified.

**Table 12**

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Units
Input filter time (HIN, LIN)	$T_{FIL,IN}$	$V_{IN} = 0$ or $V_{IN} = 5V$	-	300	-	ns
Input filter time (ITRIP)	$T_{FIL,ITRIP}$	$V_{IN}=0$ or $V_{IN}=5V$	-	500	-	ns
Internal dead time	$DT_{IC}$	$V_{IN} = 0$ or $V_{IN} = 5V$	-	300	-	ns
Matching propagation delay time (on and off) for same phase high-side and low-side	$M_T$	External dead time > 500ns	-	-	50	ns

## 8 Thermistor Characteristics

Table 13

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Units
Resistance	$R_{25}$	$T_C = 25^\circ\text{C}$ , $\pm 5\%$ tolerance	44.65	47	49.35	$\text{k}\Omega$
Resistance	$R_{125}$	$T_C = 125^\circ\text{C}$	1.27	1.39	1.51	$\text{k}\Omega$
B-constant (25/100)	B	$\pm 1\%$ tolerance	-	4006	-	K
Temperature Range			-20	-	150	$^\circ\text{C}$

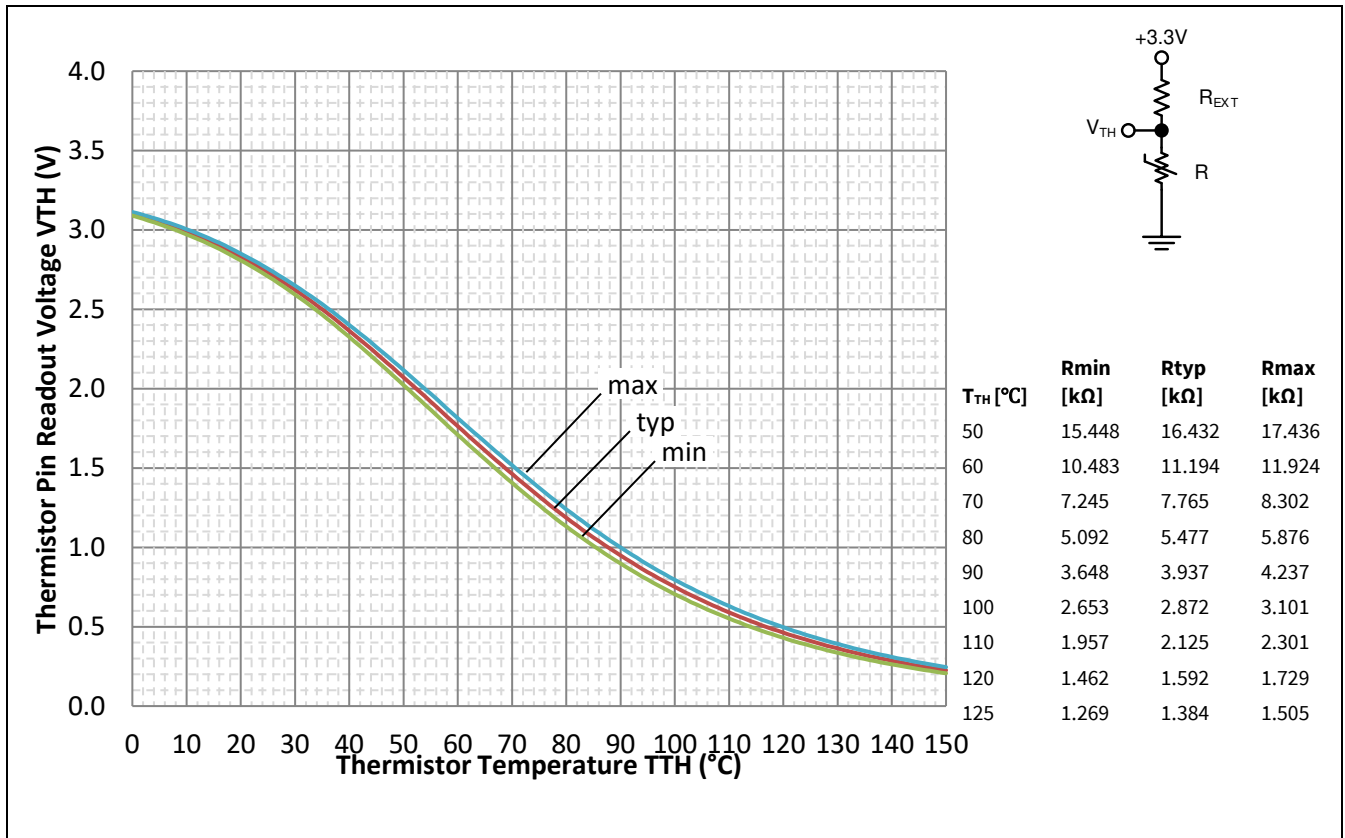


Figure 7 Thermistor resistance - temperature curve, for  $R_{EXT}=9.76\text{k}\Omega$ , and thermistor resistance variation with temperature.

## 9 Mechanical Characteristics and Ratings

**Table 14**

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Units
Comparative Tracking Index	CTI		550	-	-	V
Curvature of module backside	BC	See Figure 9	-50	-	50	μm
Mounting Torque	T	M3 screw & washer, thermal grease	0.4	0.8	1.2	Nm
		M3 screw & washer, SIL-PAD 1500ST	-	0.6	1.0	Nm
Weight	W		-	3	-	g

## 10 Qualification Information

**Table 15**

<b>UL Certification</b>	UL-US-L252584-15-22508102-2	
<b>Moisture sensitivity level (SOP 29 x 12 only)</b>	MSL3	
<b>RoHS Compliant</b>	Yes	
<b>ESD</b>	Human body model	1C
	Charge discharge model	C3



## 11 Diagrams & Tables

### 11.1 T<sub>c</sub> Measurement Point

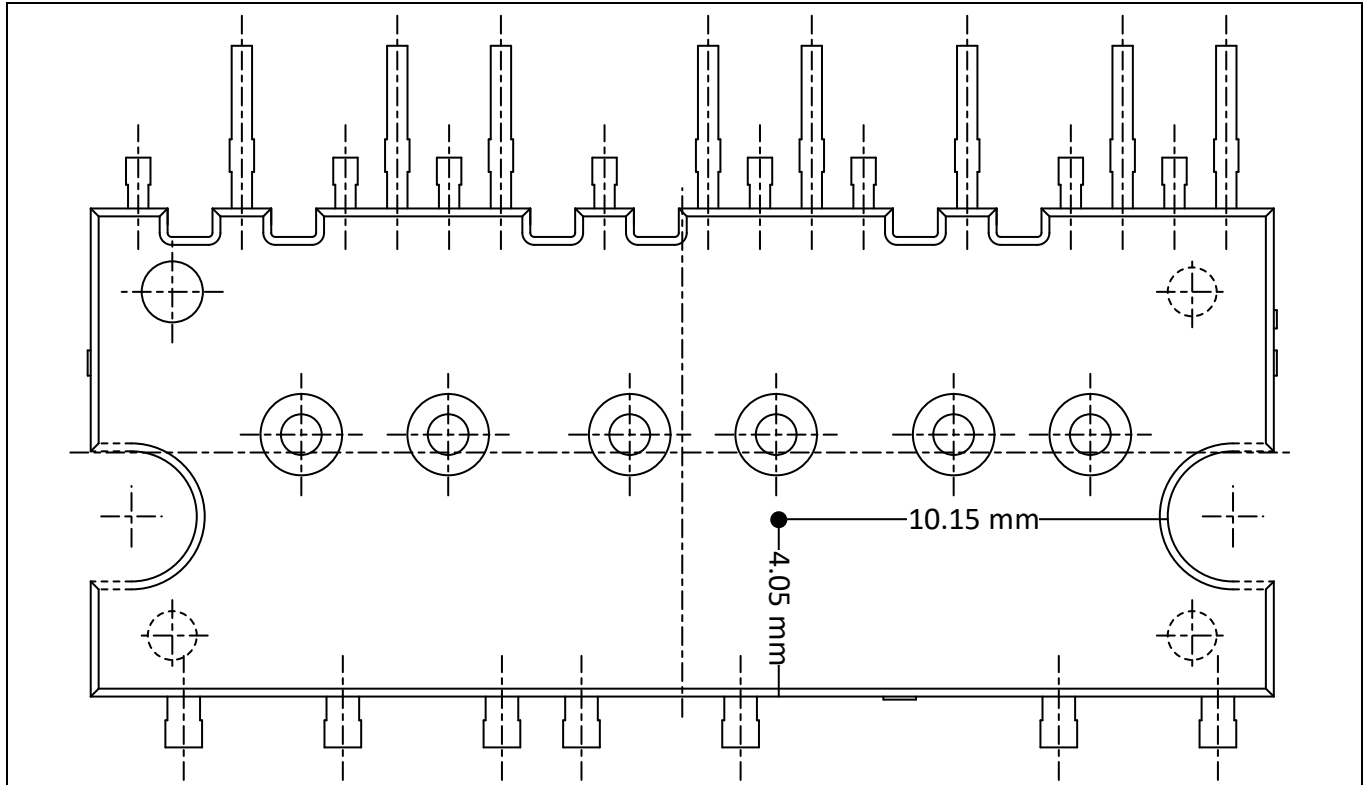


Figure 8 T<sub>c</sub> measurement point

### 11.2 Backside Curvature Measurement Points

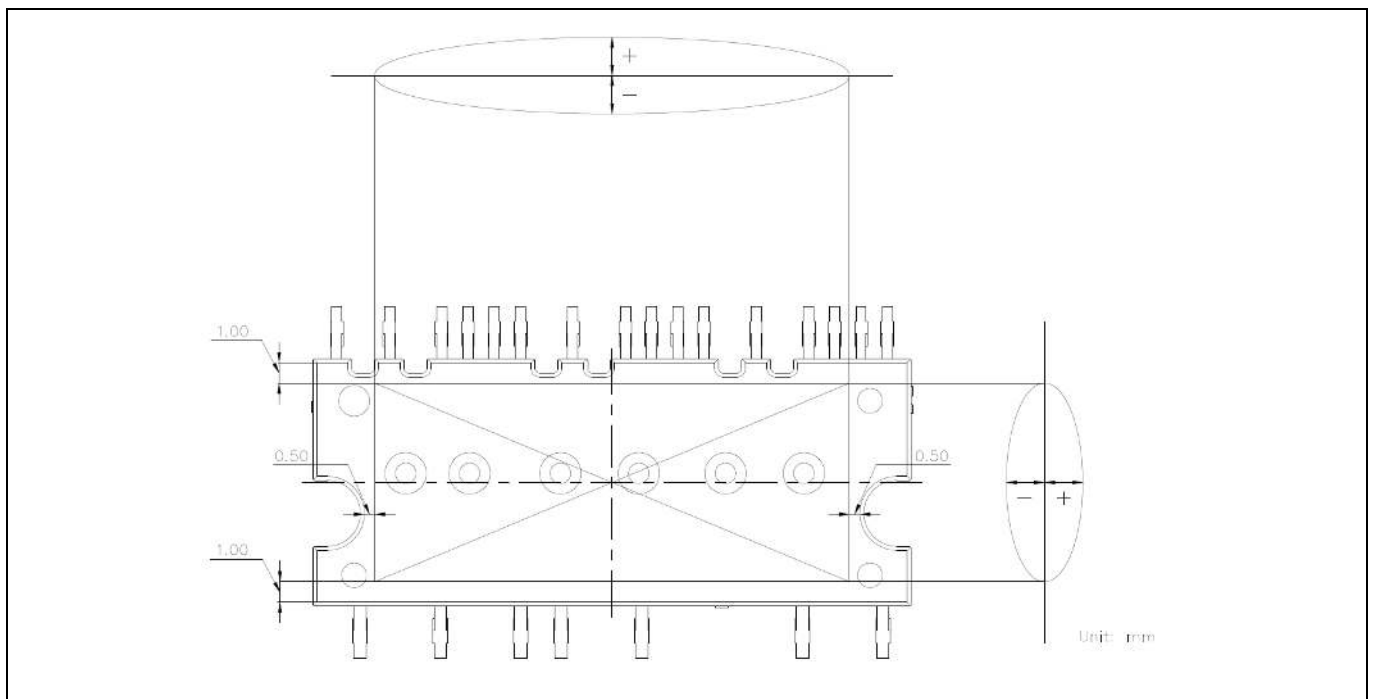


Figure 9 Curvature measurement points

### 11.3 Input-Output Logic Table

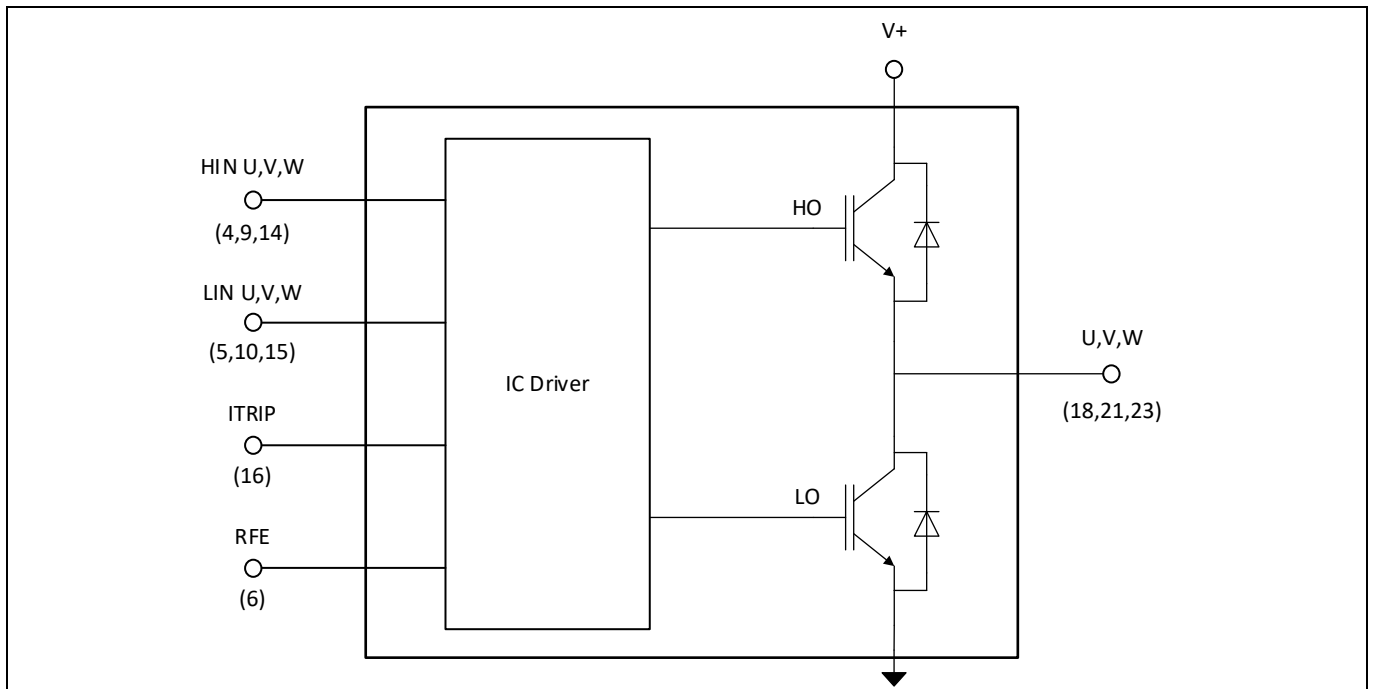


Figure 10 Module block diagram

Table 16

RFE	ITRIP	HIN U,V,W	LIN U,V,W	U,V,W
1	0	1	0	V+
1	0	0	1	0
1	0	0	0	‡
1	0	1	1	‡
1	1	x	x	‡
0	x	x	x	‡

‡ Voltage depends on direction of phase current

### 11.4 Switching Time Definitions

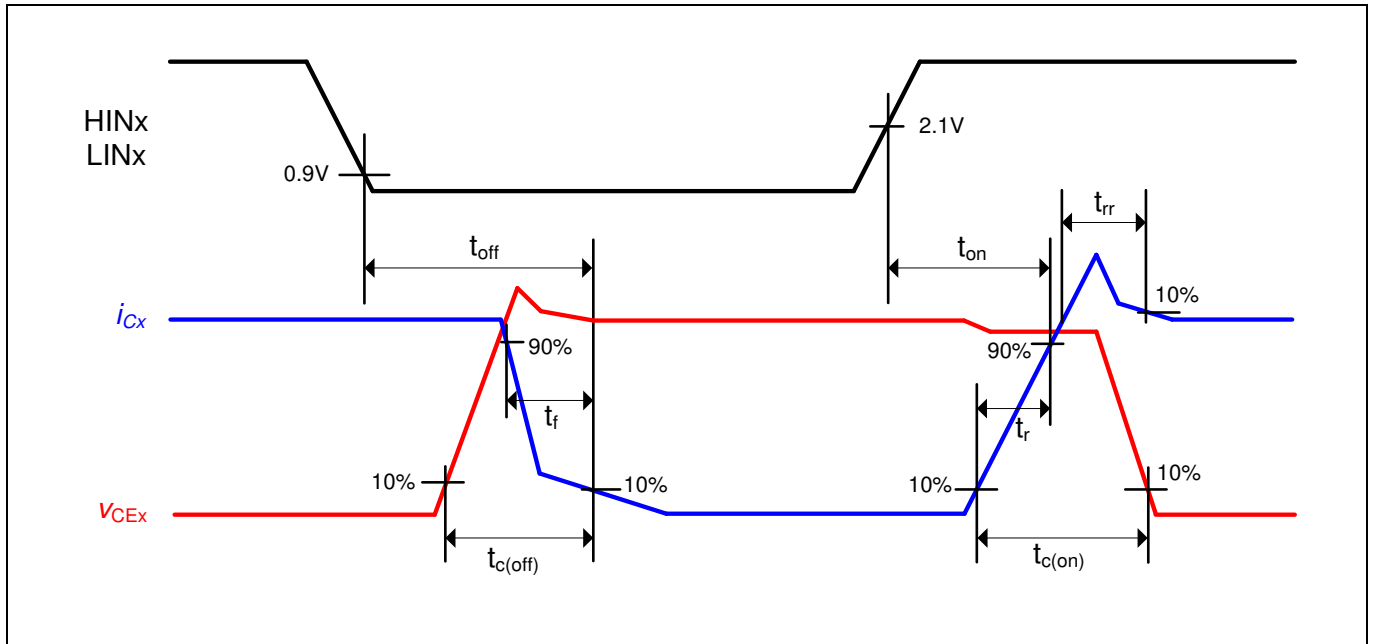


Figure 11 Switching times definition

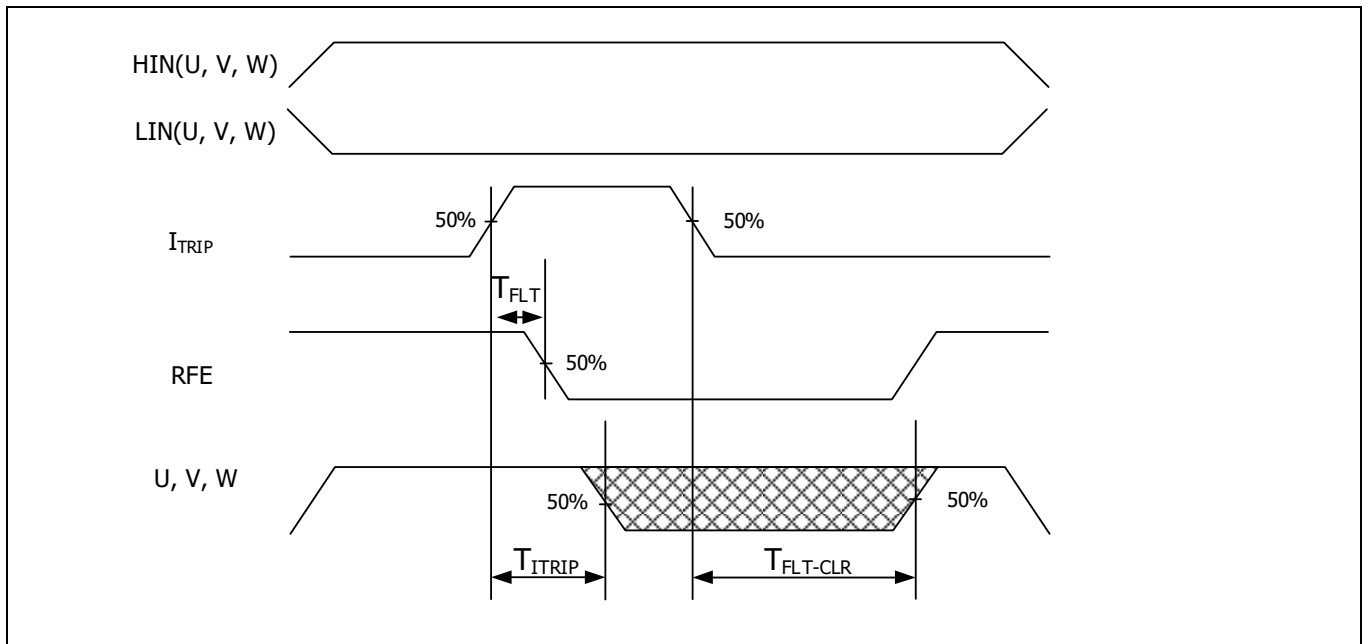


Figure 12 ITRIP time waveform

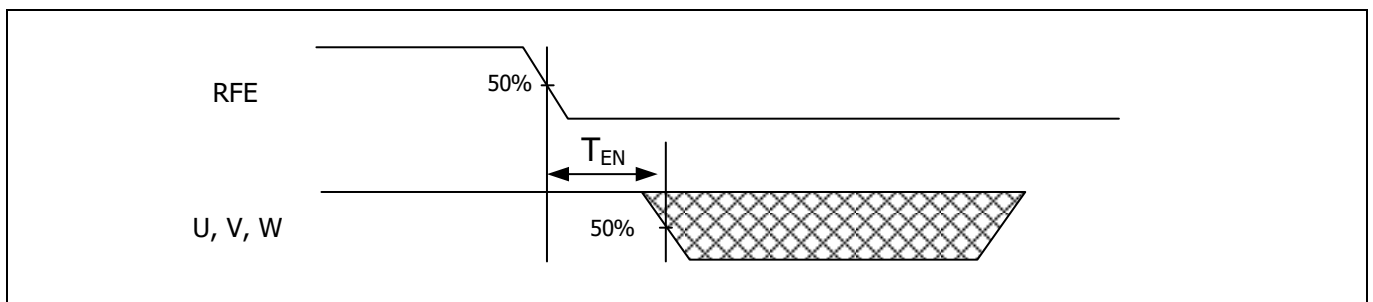


Figure 13 Output disable timing diagram

## 12 Application Guide

### 12.1 Typical Application Schematic

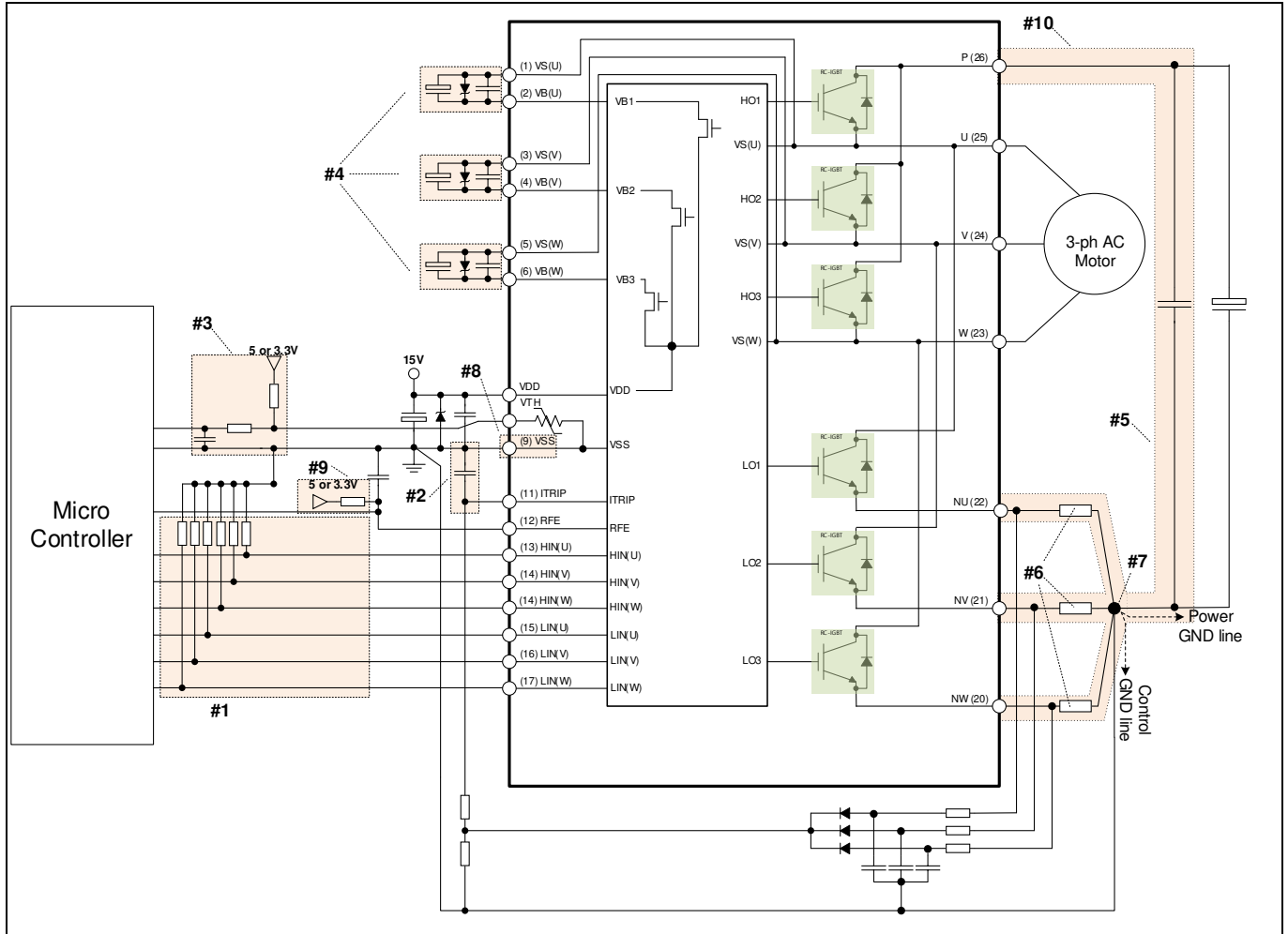


Figure 14 Application schematic

12.2 T<sub>J</sub> vs T<sub>TH</sub>

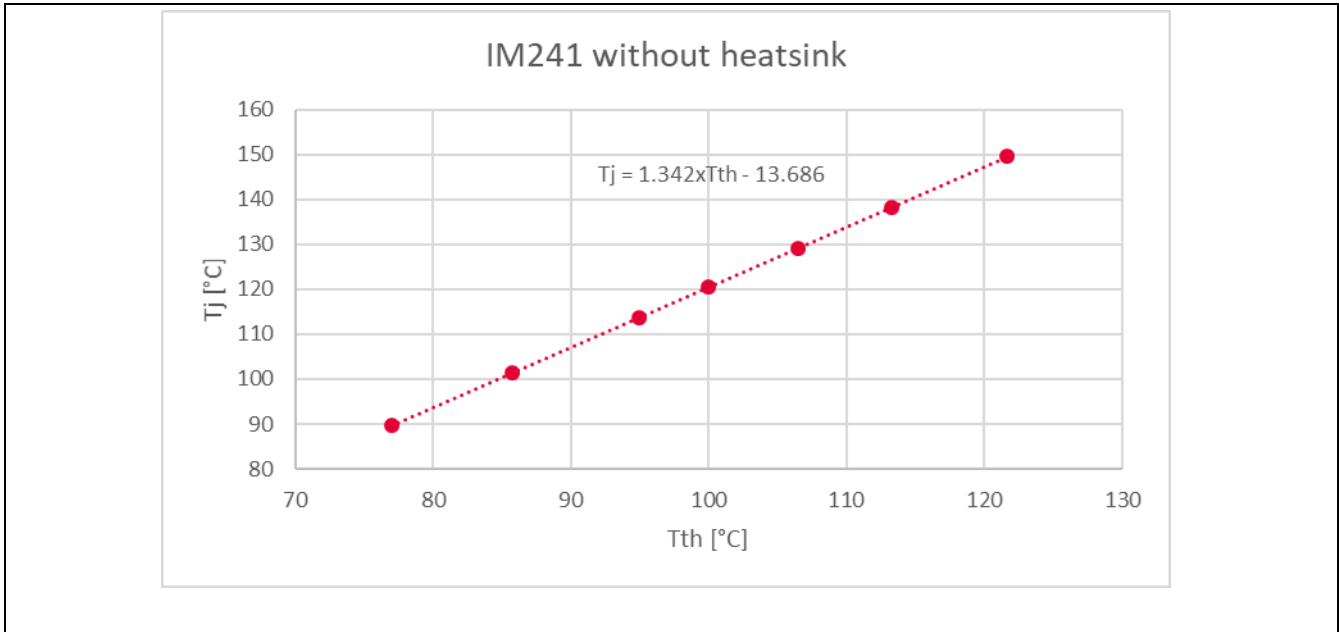


Figure 15 Typical T<sub>J</sub> vs T<sub>TH</sub> correlation without heatsink (AN-2021-08 for reference)

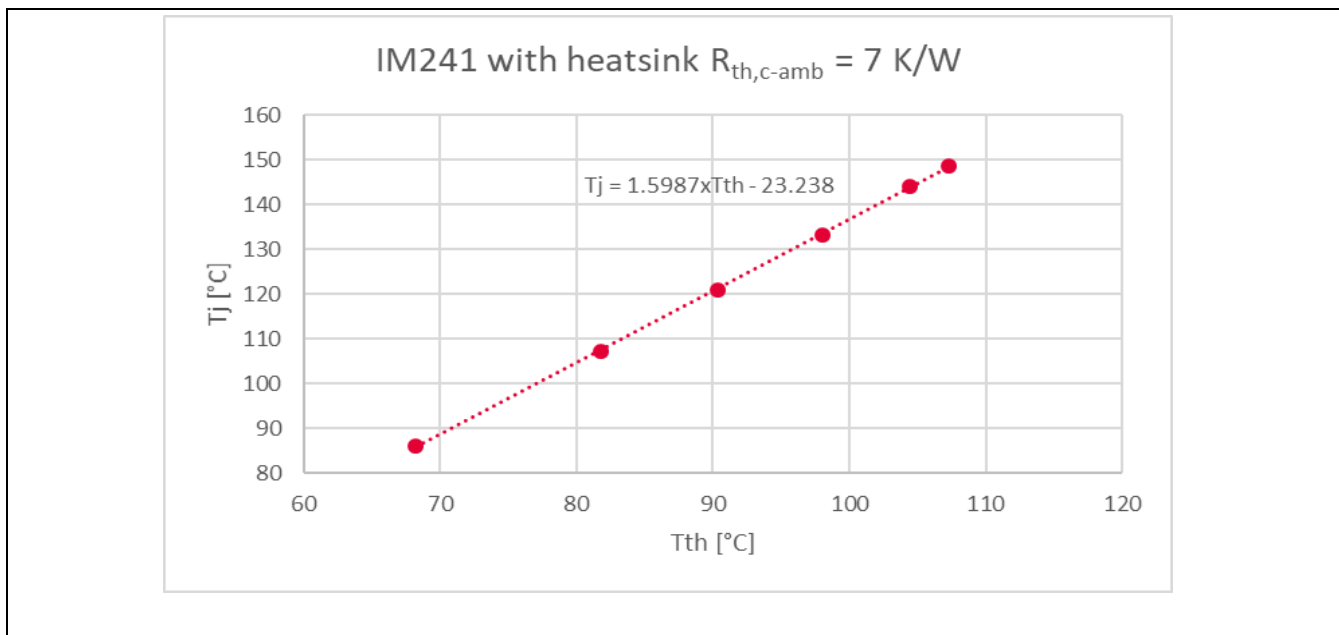


Figure 16 Typical T<sub>J</sub> vs T<sub>TH</sub> correlation with heatsink and  $R_{th,c-amb} = 7 \text{ K/W}$  (AN-2021-08 for reference)

### 12.3 -V<sub>s</sub> Immunity

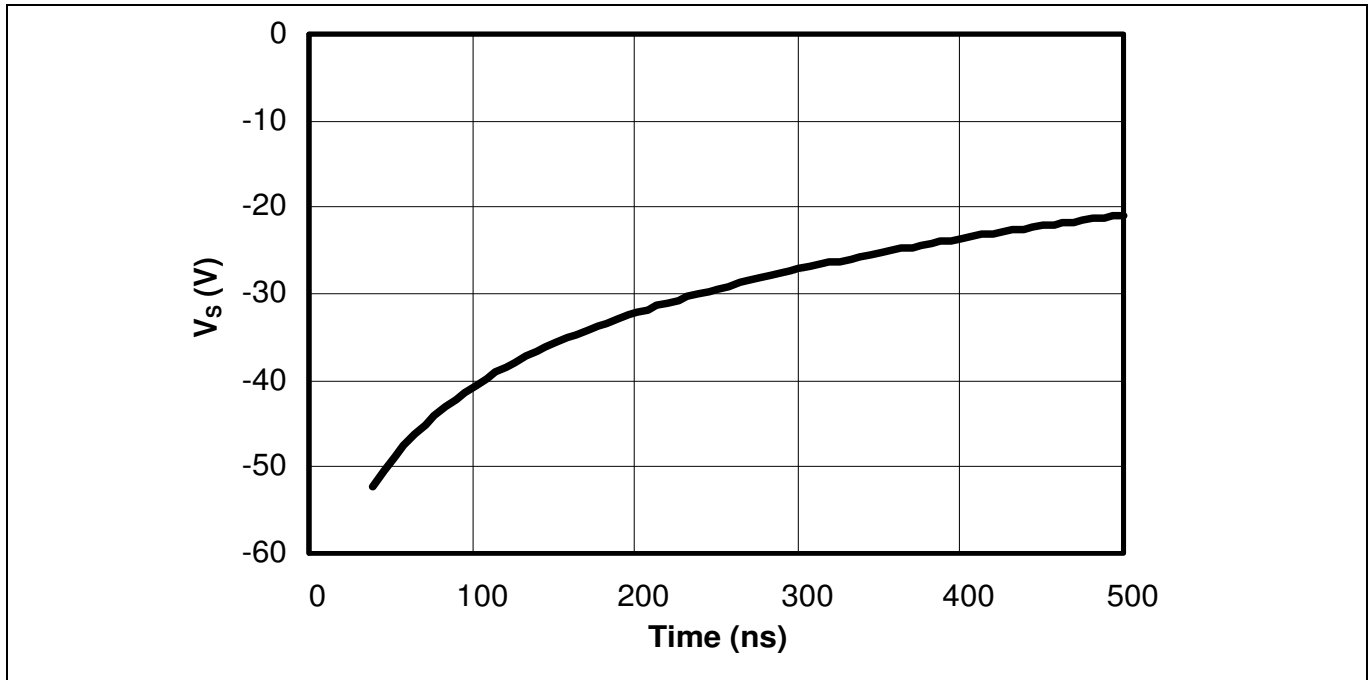
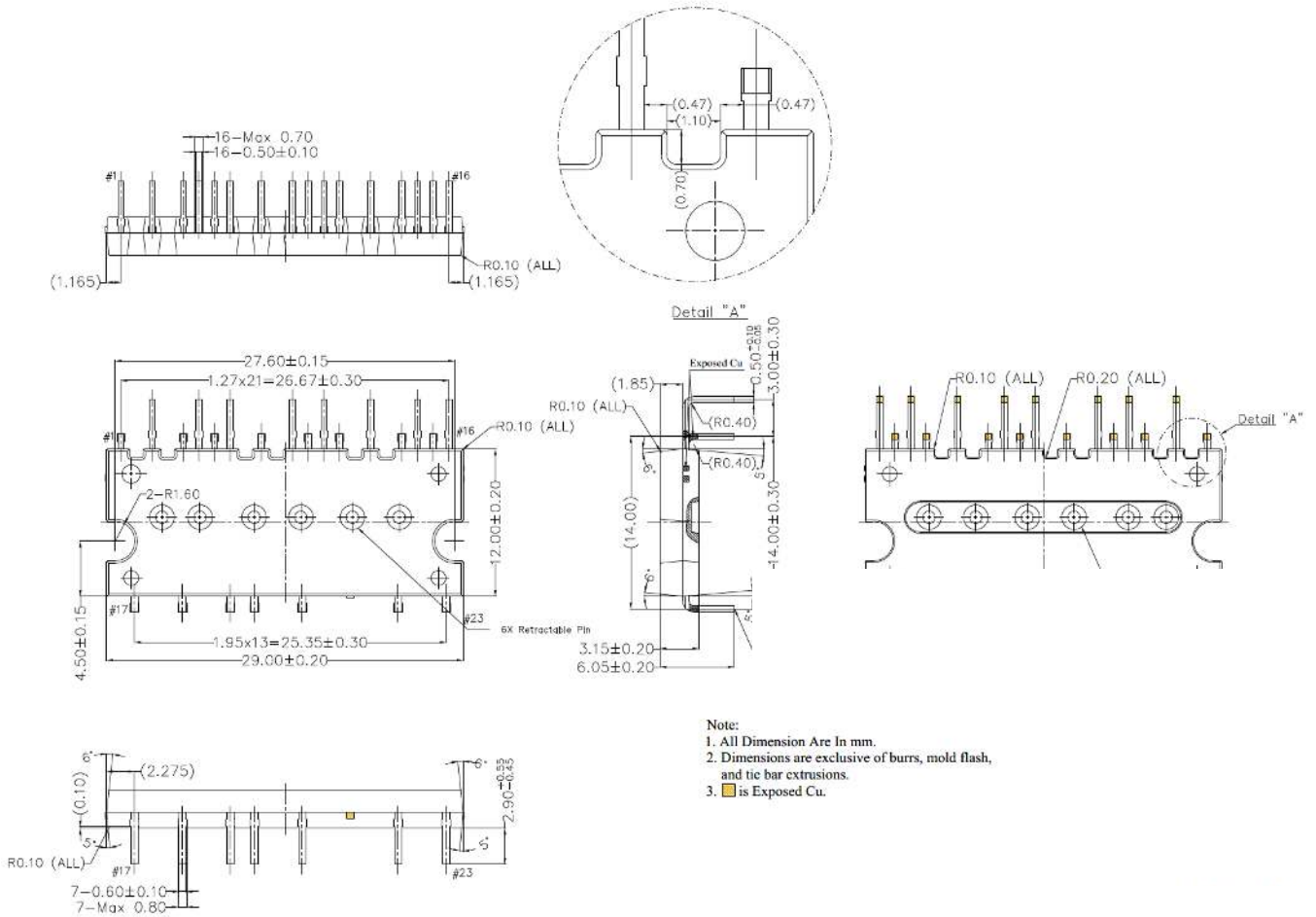


Figure 17 Negative transient V<sub>s</sub> SOA for integrated gate driver

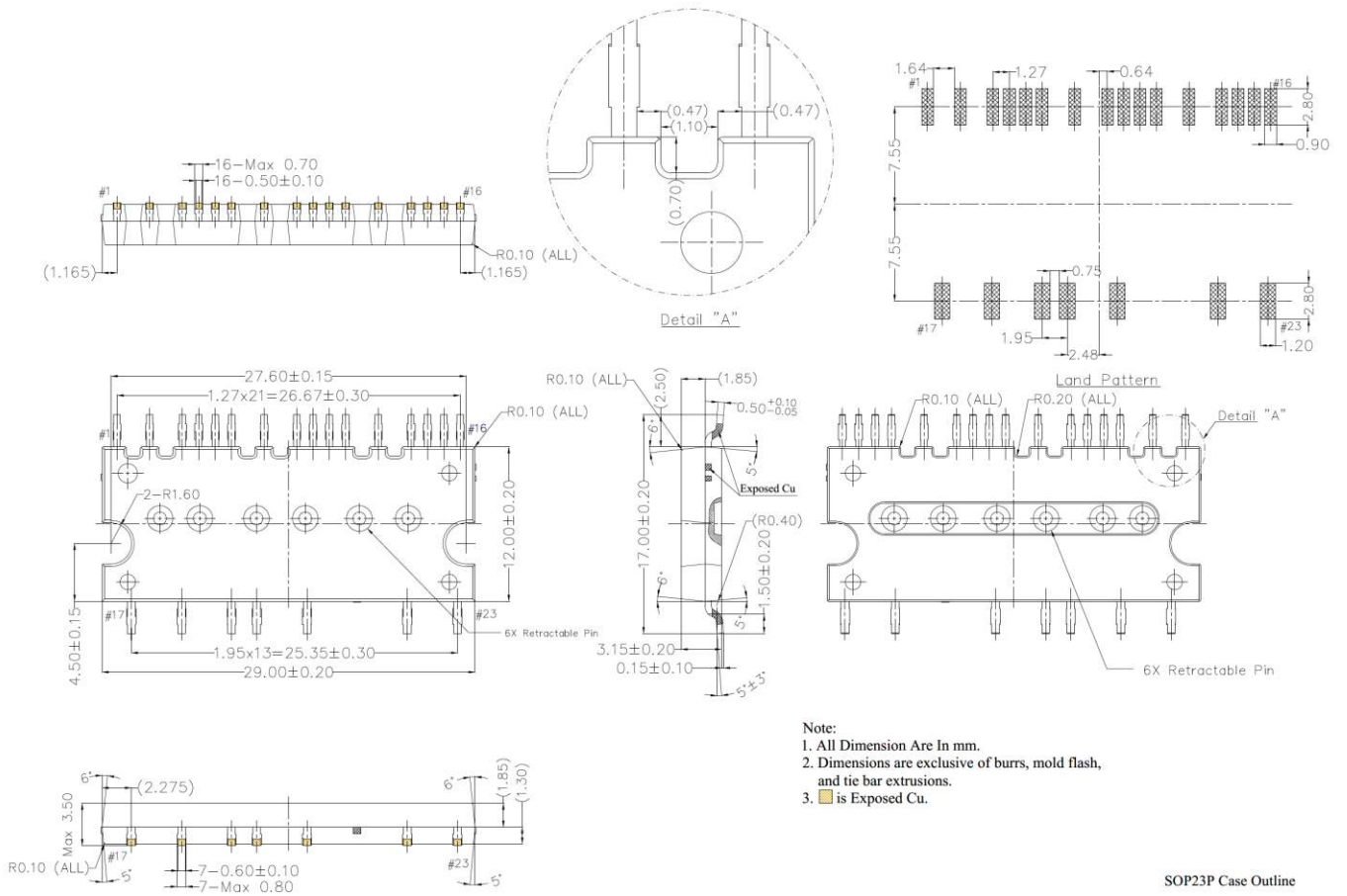
# 13 Package Outline

## 13.1 DIP 29x12



Dimensions in mm

**13.2 SOP 29x12**



**Note:**  
 1. All Dimension Are In mm.  
 2. Dimensions are exclusive of burrs, mold flash, and tie bar extrusions.  
 3. is Exposed Cu.

SOP23P Case Outline

Dimensions in mm



## **14 Revision History**

### **Major changes since the last revision**

<b>Page or Reference</b>	<b>Description of change</b>

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