8-Bit MCU with 32K Flash and 1K SRAM

GENERAL DESCRIPTION

IS31CS8969 is a general-purpose microcontroller with 32K embedded flash memory and 1K SRAM. The CPU is based on 1-T 8051 with T0/T1/T2 and additional 16-bit T3/T4, 24-bit T5 and a 30-bit WDT. Embedded in the CPU core are also a full-duplex UART port, an enhanced EUART port with LIN capability, one I ²C master/slave and two I²C pure slave controllers, one SPI mater/slave controller, up to 20 GPIO pins with each GPIO pin configurable as external interrupt and wake up.

The flexibility in clock setting includes an on-chip 16MHz precision oscillator with the accuracy deviation of +/-2%, or a low power internal 32KHz oscillator. The clock selections are combined with flexible power management schemes, including NORMAL, PMM, IDLE, and STOP, and SLEEP modes to balance CPU speed and power consumption.

A Programmable Counter Array (PCA) with 6 channels of Capture/Compare/PWM modules can be used for various purposes controlling external devices. There are additional 2 independent 8-bit PWM and a buzzer waveform generator with frequency range of 128Hz to 2048Hz and programmable duty cycle.

Other digital peripherals include a EUART2 with 16-byte FIFO, which support full LIN protocols, and an I2C slave controller, and a SPI Master/Slave controller with 4-byte FIFO.

Analog peripherals include a high performance 12-bit Analog to Digital Converter (ADC) with 30usec conversion time and a Programmable Gain Amplifier as ADC front-end. There are an on-chip temperature sensor, and a calibrated voltage reference within the ADC block. A 10-bit voltage output Digital to Analog Converter (DAC) is also included.

IS31CS8969 also provides a flexible means of flash programming that supports ISP and IAP. The protection of data loss is implemented in hardware by access restriction of critical storage segments. The code security is reinforced with sophisticated writer commands and ISP commands. The on-chip break point processor also allows easy debugging which can be integrated with ISP. Reliable power-on-reset circuit and low supply voltage detection allows reliable operations under harsh environments.

FEATURES

CPU and Memory

- 1-Cycle 8051 CPU core up to 16MHz
- ◆ 16-bit Timers T0/T1/T2/T3/T4 and 24-bit T5
- ◆ Programmable 30-bit Watch Dog Timer
- \bullet Integrated break point controller and debug port through I²C slave
- External interrupts and wake up shared with GPIO pins
- ◆ Power saving modes Normal, PMM, IDLE, STOP, and SLEEP modes
- \triangleleft Wake-Up noise filter up to 4 msec
- ◆ 256B IRAM and 768B XRAM
	- 32KB Flash Memory and 128B Information Block
		- Code security and data loss protection
		- **Endurance: 100K cycles and Retention: 10** years @85°C

Clock Sources

- ◆ Internal oscillator at 16MHz of +/- 2% accuracy
- ◆ Internal low power 32KHz oscillator

Digital Peripherals

- ◆ 16-bit PCA and 6 channel of CCP modules
	- Capture/Compare/Timer Mode
	- 8/16-bit PWM Mode and 8-bit WPWM Mode
- Two 8-bit PWM Controller
- One buzzer waveform generator
- ◆ One I²C Master
- One I²C pure Slave
	- **Support two consecutive addresses**
- One SPI Master/Slave Controllers
	- 4 bytes transmit and receive FIFO
- One 8051 UART
- One full-duplex LIN-capable EUART2
- **16-bit baud rate register**
	- **16 bytes transmit and receive FIFO**

Analog Peripherals

- ◆ 12-bit monotonic SAR ADC
	- 500KHz, 128usec conversion time (1.8~2.4V)
	- 4Mhz, 16usec conversion time (2.5~5.5V)
	- Programmable Gain Amplifier
	- Up to 16 input shared with GPIO
	- On-chip 1.1V reference
	- On-chip temperature sensor
- 10-bit voltage output DAC
- 4-Channel Analog Comparator
- Capacitance sense touch-key controller scan up to 20 keys through shared GPIO
- ◆ Power on reset (1.5V) and LVD/LVR (1.8V-5.5V)

Miscellaneous

- ◆ Up to 20 GPIO pins
- 2.2V to 5.5V single supply with on-chip 1.8V regulator. Operating down to 1.8V
- ◆ Active current < 300uA/MHz in NORMAL mode
- Low power standby (< 20uA) in SLEEP mode
- ◆ Operating temperature -45°C to 85°C
- ◆ RoHS compliant packages
	- TSSOP 24
	- TSSOP 20
	- $\overline{}$ TSSOP 16
	- QFN 24

IS31CS8969 BLOCK DIAGRAM

IS31CS8969 PIN CONNECTION

Note: The Part Number and Logo is for reference only and do not reflect the actual marking on the package.

IS31CS8969 PIN DESCRIPTIONS

Note: "P" denotes power supply pins

"G" denotes ground pins. All VSS pins are internally shorted resistively. "O", "IO", "A" denotes output only, input/output, and analog types.

"IOCELL2A" denotes IOCELL with two analog inputs

PIN # : TSSOP-24 / TSSOP-20 / TSSOP-16 / QFN-24

MEMORY MAP

There are total 256 bytes internal RAM in IS31CS8969, the same as standard 8052. There are total 1792 bytes auxiliary RAM allocated in the 8051 extended RAM at 0x0100h – 0x07FFh. Programs can use "MOVX" instruction to access the XRAM. The 32KB embedded flash occupies the code address space from 0x0000h – 0x7FFFh. The CPU reset to address 0x7000h. The memory map is shown in the following:

REGISTER MAP SFR(0x80 – 0xFF) and XFR (0xA000 – 0xAFFF)

The SFR address map maintains maximum compatibilities to most commonly used 8051 like MCU. The following table shows the SFR address map. Since SFR can be accessed by direct addressing mode, registers of built-in peripherals that require fast access are mostly located in SFR. XFR is mainly used for on-chip peripheral control and configurations.

IS31CS8969 REGISTER MAP XFR (0xA000 – 0xAFFF)

1. Enhanced 1-Cycle 8051 CPU

The CPU core is an enhanced version of standard 8051 used by series of ISSI MCU products. The CPU core is in RISC architecture and maintains binary instruction set compatible with the industry standard 8051. There is average 10 times performance enhancement in typical applications. The CPU operates at 20-bit addressing space that allows up to 1M bytes of program and data space for expansion. The CPU includes the following enhanced features compared with standard 8051:

- 16-bit LARGE addressing mode and 20-bit FLAT addressing mode control register ACON
- Two data pointers DPTR and DPTR1, and additional DPS, DPX, DPX1, MXAX registers for MOVX instruction
- 8-bit stack pointer for LARGE mode and 16-bit extended stack pointer for FLAT mode control register ESP
- ◆ Hardware Multiplication and Division Unit (MDU) provides 12 times faster performance using MD[5-0] and ARCON
- Programmable wait state for program space for on-chip flash memory using WTST register
- ◆ 256 Bytes of Direct Data Memory
- Enhanced Interrupt Controller allows 15 interrupt sources and 2 priority levels.
- Power Saving modes include IDLE mode, Power Management mode (PMM), and STOP mode. The PMM mode also supports switchback features.
- Access Control of critical registers TA, and TB registers
- Eight break pointers allows integration of common IDE

In addition to standard 8051 peripherals, the CPU core also integrates the following peripherals. These peripherals are in the same CPU clock domain.

- Six 8-Bit I/O ports
- 30-bit Watch Dog Timer. WDT, WDCON, and CKCON registers
- Three 16-bit Timers, T0/T1 and T2. TCON, RLDL, RLDH, TL2, TH2, and T2CON registers
- UART0.
- I2C Master Controller. I2CMSA, I2CMCR, I2CMBUF, and I2CMTP registers.

The following sections describe in detail these enhanced features and peripherals. Assuming readers are familiar with 8051 standard operations and peripherals, the compatible functions is not covered here.

1.1 System Reset

After system reset, all registers resume the default value. The default value is shown in the register description. The reset conditions include power on/off reset, external RSTN pin being pulled low, low supply voltage detection reset, and WDT reset. The block diagram illustrating these reset conditions is shown as follows:

The power on/off reset (POR) is asserted when VDD is less than 1.6V or VDDC is less than 1.5V. The external RSTN pin can also generate reset to the device. In typical applications, the RSTN should have a resister (R1) connected to VDD and a capacitor (C1) to ground. For a system with a hardware reset control, there is usually a button switch connecting RSTN pin to ground. When the switch is pressed, it causes RSTN to short to ground, and the device enters reset state. The RSTN logic has a built-in filter that ignores RSTN duration shorter than 5usec. It is, therefore, recommended that RSTN needs to be actively pulled low for at least 50usec to guarantee a solid reset. The LVD circuits can detect the main supply voltage level VDD and the threshold can be adjusted. LVD reset is disabled by default, yet may be enabled by the software. The LVD output can be enabled to generate LVR (Low Voltage Reset). Both POR and LVR will also forces RSTN low. This ensures a solid and extended reset when the voltage supply to the internal logic and flash memory is lower than the rated level.

The last reset source is from the watchdog counter (WDT). The WDT reset function is enabled whenever a system reset occurs, and WDT timeout is set to maximum. It is recommended that all software should keep WDT enabled to ensure reliable software executions.

The program counter is loaded with 0x07000 after reset. This differs from standard 8051. In typical cases, 0x0F000 starts Calibration and ISP boot codes and then jumps to 0x0000. The clock selection after reset is set to using internal oscillator automatically. The IOSC is disabled only in STOP and SLEEP modes.

1.2 CPU Registers

ACC (0xE0) Accumulator R/W (0x00)

ACC is the CPU accumulator register and is involved in direct operations of many instructions. ACC is bit addressable.

B (0xF0) B Register R/W (0x00)

B register is used in standard 8051 multiply and divide instructions and also used as an auxiliary register for temporary storage. B is also bit addressable.

PSW (0xD0) Program Status Word R/W (0x00)

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SP (0x81) Stack Pointer R/W (0x00)

PUSH will result ACC to be written to SP+1 address. POP will load ACC from IRAM with the address of SP.

ESP (0x9B) Extended Stack Pointer R/W (0x00)

In FLAT address mode, ESP and SP together form a 16-bit address for stack pointer. ESP holds the higher byte of the 16-bit address.

1.3 Addressing Mode and Memory Operations

The LARGE mode, addressing mode is compatible with standard 8051 in 16-bit address. FLAT mode extends the program address to 20-bit and expands the stack space to 16-bit data space. The data space is always 16-bit in either LARGE or FLAT mode.

ACON (0x9D) R/W (0x00) TA

ACON is addressing mode control register.

DPXR (0xDA) R/W (0x00)

DPXR[7-0] is used to replace P2[7-0] for high byte of XRAM address for "MOVX, @R1" or "MOVX, @R0" when DPXREN=1.

The clock speed of an MCU with embedded flash memory is usually limited by the access time of on-chip flash memory. While in modern process technology, the CPU can operate up to 100MHz to 200MHz, but the access time of flash memory is usually around 20 nanoseconds and thus limiting the clock rate to lower than 50MHz. To alleviate this problem, a programmable wait state function is incorporated to allow faster CPU clock rate however slower embedded flash memory. The wait state is controlled by WTST register as shown in the following,

WTST (0x92) R/W (0x07)

WTST is wait state register that controls the program access wait state only.

WTST[3-0] Wait State Control register. WTST sets the wait state in CPU clock period

The default setting of the program wait state register after reset is 0x07 and the software must initialize the setting to change the wait state setting. For typical embedded flash, the read access time is specified as 40 nsec. Therefore the user should set the WTST register according to the SYSCLK frequency. For example, using a SYSCLK of 4MHz, the WTST can be set to minimum because one clock period is 250 nsec which is longer than the embedded flash access time. If SYSCLK is above 16MHz, then WTST should be set higher than 1 to allow enough read access time.

1.4 Dual Data Pointers and MOVX operations

In standard 8051/8052, there is only one data pointers DPH:DPL to perform MOVX. The enhanced CPU provides 2nd data pointer DPH1:DPL1 to speed up the movement, or copying of data block. The active DPTR is selected and operation of DPTR is controlled by setting DPS (Data Pointer Select) register. Through the control DPS, efficient programming can be achieved.

DPS (0x86) R/W (0x00)

ID[1:0] Define the operation of Increment/Decrement functions of selected DPTR for INC DPTR instruction is executed.

TSL Enable toggling selection of DPTR selection. When this bit is set, the selection of DPTR is toggled when DPTR is used in an instruction and executed.

SEL DPTR selection bit. Set to select DPTR1, and clear to select DPTR. SEL is also affected by the state of ID[1:0] and TSL after DPTR is used in an instruction. When read, SEL reflects the current selection of command.

DPL (0x82) Data Pointer Low R/W (0x00)

DPL register holds the low byte of data pointer, DPTR.

DPH (0x83) Data Pointer High R/W (0x00)

DPH register holds the high byte of data pointer, DPTR.

DPL1 (0x84) Extended Data Pointer Low R/W (0x00)

DPL1 register holds the low byte of extended data pointer 1, DPTR1.

DPH1 (0x85) Extended Data Pointer High R/W (0x00)

DPH1 register holds the high byte of extended data pointer 1, DPTR1.

DPX (0x93) Data Pointer Top R/W (0x00)

DPX is used to provide top 8-bit address of DPTR when address above 32KB. The lower 16-bit address is formed by DPH and DPL. Since IS31CS8969 only has on-chip data RAM space, DPX value has no effect.

DPX1 (0x95) Extended Data Pointer Top R/W (0x00)

DPX1 is used to provide top 8-bit address of DPTR when address above 32KB. The lower 16-bit address is formed by DPH and DPL. Since IS31CS8969 only has on-chip data RAM space, DPX value has no effect.

MXAX (0xEA) MOVX Extended Address Register R/W (0x00)

MXAX is used to provide top 8-bit address for an "MOVX @R0" or "MOVX @R1" instruction. The lower 16-bit address is formed by P2 and R0/R1 (if DPXREN=0), or formed by DPXR and R0/R1 (if DPXREN=1).

MCON (0xC6) XRAM Relocation Register R/W (0x00) TA Protected

MCON holds the starting address of XRAM in 4KB steps. For example, if MCON[7-0]=0x01, the starting address is 0x001000h. MCON is not meaningful in IS31CS8969 because it only contains on-chip XRAM and MCON should not be modified from 0x00.

When accessing XRAM using "MOVX, @DPTR" instruction, the address of XRAM access is formed by DPHi:DPLi depending on which data pointer is selected. Another form of MOVX instruction is "MOVX, @Ri". This instruction provides an efficient programming method to move content within a 256-byte data block. In "@RI" instruction, the XRAM address [15-7] can be derived from two sources. If ACON.DPXREN = 0, the high order address $[15-8]$ is from P2 (0xA0), if ACON.DPXREN = 1, the high order address is from DPXR (0xDA) register.

The maximum addressing space of XRAM is up to 16MB thus requiring 24-bit address. For "MOVX, @DPTR", the XRAMADDR [23-16] is from either DPX (0x93) or DPX1 (0x95) depending on which data pointer is selected. For "MOVX, @Ri", the XRAMUADDR [23-16] is from MXAX (0xEA) register.

1.5 Interrupt System

The CPU implements an enhanced Interrupt Control that allows total 15 interrupt sources and each with two programmable priority levels. The interrupts are sampled at rising edge of SYSCLK. If interrupts are present and enabled, the CPU enters interrupt service routine by vectoring to the highest priority interrupt. Of the 15 interrupt sources, 7 of them are from CPU internal integrated peripherals, 6 of them are for on-chip external peripherals, and 2 of them are used for external pin interrupt expansion. When an interrupt is shared, the interrupt service routine must determine which source is requesting the interrupt by examining the corresponding interrupt flags of sharing peripherals.

The following table shows the interrupt sources and corresponding interrupt vectors. The Flag Reset column shows whether the corresponding interrupt flag is cleared by hardware (self-cleared) or software. Please note the software can only clear the interrupt flag but not set the interrupt flag. The Natural Priority column shows the inherent priority if more than one interrupts are assigned to the same priority level. Please note that the interrupts assigned with higher priority levels always get serviced first compared with interrupts assigned with lower priority levels regardless of the natural priority sequence.

 In addition to the 15 peripheral interrupts, there are two highest priority interrupts associated with debugging and break point. DBG interrupt is generated when ¹²C slave is configured as a debug port and a debug request from

the host matches the debug ID. BKP interrupt is generated when break point match condition occurs. DBG has higher priority than BKP. The BKP and DBG interrupts are not affected by global interrupt enable, EA bit, IE reigster (0xA8).

The following diagram shows the interrupt sources and the expanded pin interrupts

The interrupt related registers are listed in the following. Each interrupt can be individually enabled or disabled by setting or clearing corresponding bits in IE, EXIE and integrated peripherals' control registers.

IE (0xA8) R/W (0x00)

EA Global Interrupt Enable bit. ES2 LIN-capable 16550-like UART2 Interrupt Enable bit. ET2 Timer 2 Interrupt Enable bit. ES0 UART0 Interrupt Enable bit. ET1 Timer 1 Interrupt Enable bit. PINT1EN Pin PINT1.x Interrupt Enable bit.

ET0 Timer 0 Interrupt Enable bit.

PINT0EN Pin PINT0.x Interrupt Enable bit.

EXIE (0xE8) R/W (0x00)

EINT8 Timer 3/4/5 Interrupt Enable bit.

EINT7 SPI and I²C Slave Interrupt Enable bit.

EINT6 PCA and PWM Interrupt Enable bit.

EWD1 Watchdog Timer Interrupt Enable bit.

EINT4 ADC Interrupt Enable bit.

EINT3 Analog Comparator, Buzzer, Touch Key and Remote filter Interrupt Enable bit.

EINT2 Low Voltage Detection Interrupt Enable bit.

EI2CM I²C Master Interrupt Enable bit.

Each interrupt can be individually assigned to either high or low. When the corresponding bit is set to 1, it indicates it is of high priority.

IP (0xB8) R/W (0x00)

PS2 LIN-capable 16550-like UART2 Priority bit.

PT2 Timer 2 Priority bit.

PS0 UART 0 Priority bit.

PT1 Timer 1 Priority bit.

PX1 Pin Interrupt INT1 Priority bit.

PT0 Timer 0 Priority bit.

PX0 Pin Interrupt INT0 Priority bit.

EXIP (0xF8) R/W (0x00)

EINT6 **INT6 PCA and PWM Priority bit.**

EWDI Watchdog Priority bit.

EINT4 **INT4 ADC Priority bit.**

EINT3 INT3 Analog Comparator, Buzzer, Touch Key and Remote filter Priority bit.

EINT2 **INT2 Low Voltage Detection Priority bit.**

EI2CM I²C Master Priority bit.

EXIF (0x91) R/W (0x00)

Writing to INT2F to INT8F has no effect.

The interrupt flag of internal peripherals are stored in the corresponding flag registers in the peripheral and EXIF registers. These peripherals include T0, T1, T2, and WDT. Therefore to clear the interrupt flags the software needs to clear the corresponding flags located in the peripherals (for T0, T1, and T2, and WDT). For I2CM, the interrupt flag is located in the EXIF register bit I2CMIF. This needs to be cleared by software.

INT2 to INT8 are used to connect to the external peripherals. INT2F to INT8F are direct equivalents of the interrupt flags from the corresponding peripherals. These peripherals include RTC, I²Cs, PCA, ADC, etc.

PINT0 and PINT1 are used for external GPIO pin Interrupts. All GPIO pin can be enabled to generate the PINT0 or PINT1 depending on its MFCFG register setting. Each GPIO pin also contains the rising/falling edge detections and either or both edges can be used for interrupt triggering. The same signaling can be used for generating wake-up.

TCON (0x88) R/W (0x00)

1.6 Register Access Control

One important aspect of the embedded MCU is its reliable operations under a harsh environment. Many system failures results from the accidental loss of data or changes of critical registers that may lead to catastrophic effects. The CPU provides several protection mechanisms, which are described in this section.

TA (0xC7) Time Access A Control Register2 WO xxxxxxx0

TA access control emulates a ticket that must be purchased before modifying a critical register. To modify or write into a TA protected register, TA must be accessed in a predefined sequence to obtain the ticket. The ticket is used when an intended modification operation is done to the TA protected register. To obtain the next access a new ticket must be obtained again by performing the same predefined sequence on TA. TA does not limit the read access of the TA protected registers. The TA protected register includes WDCON (0xD8), MCON (0xC6), and ACON (0x9D) registers. The following predefined sequence is required to modify the content of MCON.

MOV TA, #0xAA;

MOV TA, #0x55;

MOV MCON, #0x01;

Once the access is granted, there is no time limitation of the access. The access is voided if any operation is performed in TA address. When read, the bit of TA indicates whether TA is locked or not (1 indicates "unlock" and 0 indicates "lock").

TB (0xC9) Time Access B Control Register2 R/W (0x00)

TB access control functions are similar to TA control, except the ticket is for multiple uses with a time limit. Once access is granted, the access is open for 256 clock periods and then expires. The software can also read TB address to obtain the current TB status. The TB protected registers include two SFR registers, CKSEL (0x8F) and WKMASK (0x9F), and twelve XFR registers lodging REGTRM (0xA000), IOSCITRM (0xA001), IOSCVTRM (0xA002), XOSCCFG (0xA007), LVDCFG (0xA010), LVDTHD (0xA011), LVDHYS (0xA016), CNTPCTL (0xA025), CNTPCTH (0xA026), INTPCT1 (0xA013), INTPCT2 (0xA014), BPINTE (0xA0E1), and SI2C_DebugID (0xA0EF). To modify registers with TB protection, the following procedure must be performed.

MOV TB, #0xAA

MOV TB, #0x55

This action creates a timed window of 256 SYSCLK periods to allow write access of these TB protected registers. If any before-mentioned sequences are repeated before the 128 cycles expires, a new 128 cycles is extended. The current 256 cycles can be terminated immediately by writing #0x00 to TB registers, such as

MOV TB, #0x00

It is recommended to terminate the TB access window once the user program finishes the modifications of TB protected registers.

Because TA and TB are critical reassurance of the reliable operation of the MCU that prevents accidental hazardous uncontrollable modifications of critical registers, the operation of these two registers should bear extreme cautions. It is strongly advised that these two registers should be turned on only when needed.. Both registers use synchronous CPU clock, therefore it is imperative that any running tasks of TA and TB should be terminated before entering IDLE mode or STOP mode. Both modes turn off the CPU clock and if TA and TB are enabled, they stay enabled until the CPU clock resumes thus may create vulnerabilities for critical registers.

Another reliability concern of embedded Flash MCU is that the important content on the Flash can be accidentally erased. This concern is addressed by the content protection in the Flash controller.

1.7 Clock Control and Power Management Modes

This section describes the clock control and power saving modes of the CPU and its integrated peripherals. The settings are controlled by PCON (0x87) and PMR (0xC4) registers. The register description is defined as following.

PCON (0x87) R/W (0x00)

PMR (0xC4) R/W (010xxxxx)

STATUS (0xC5) RO (0x0C)

STATUS register can be accessed by program to determine the status of critical events occurring in the integrated peripherals. The program should check status conditions before entering SLEEP, STOP, IDLE, or PMM modes to prevent loss of intended functions from delayed entry until these events are finished.

CKSEL (0x8F) R/W (0x0C) System Clock Selection Register TB Protected

The SIOSC should be selected as the system clock before change the IOSC Pre-divider to avoid any clock glitch. IOSCDIV[3-0] IOSC Pre-Divider

REGRDY[1-0] Wake up delay time for main regulator stable time from reset or from sleep mode wake up

CLKSEL[1-0] Clock Source Selection

These two bits define the clock source of the system clock SYSCLK. The selections are shown in the following table. The default setting after reset is IOSC.

WKMASK (0x9F) Wake Up Mask Register R/W (0xFF) TB Protected

WEINT7 Set this bit to allow INT7 to trigger the wake up of CPU from STOP modes.

WKMASK register defines the wake up control of the interrupt signals from the STOP mode. The wake-up is performed by these interrupts and if enabled the internal oscillator is turned on and SYSCLK resumes. The interrupt can be set as a level trigger or an edge trigger and the wake-up always runs in accordance with the edge. Please note the wake-up control is wired separately from the interrupt logic, therefore, after waking up, the CPU does not necessarily enter the interrupt service routine if the corresponding interrupt is not enabled. In this case, the CPU continues onto the next instruction, which initiates the STOP mode. Extra attention should be exerted as designing the exit and re-entry of modes to ensure proper operation.

Please note that all clocks are stopped in STOP mode, therefore peripherals require clock such as I²C slave, UARTx, ADC, LVD, and T3 cannot perform wake-up function. Only external pins and peripherals that do not require a clock can be used for wake up purposes. Such peripherals in IS31CS8969 are an analog comparator

1.7.1 PMM mode

PMM mode is enabled by setting CD[1:0] bits in PMR register to both 1. In PMM mode, the CPU and its integrated peripheral such as WDT, UART0, LIN-capable 16550-like UART2, T0/T1/T2, T3, and I²C Master operate at 257 times slower than SYSCLK. All other external peripherals such as PCA, ADC etc. are still operating under normal clock. The PMM mode saves power because the CPU, internal Flash memory and SRAM by operating at much slower frequency. The program continues to run while the CPU is operating at a reduced rate. To further save power, the unused external peripherals can be turned off or disabled. Normal mode operation can be recovered from PMM mode by program itself that set CD[1:0] = 01. Another way of recovery is to enable the SWITCHBACK function by setting SWB bit to high in PMR register. When switchback is enabled, the following conditions trigger the CPU to exit PMM mode and resume normal operations.

External Interrupt INT0/1/2/3/4/6/7/8 and any external peripherals interrupt OR-ED with these interrupts.

UART0 receive Start bit detection

UART0 transmit buffer loaded

When an external interrupt is intended to perform switchback, the corresponding interrupt must be enabled and not blocked by higher priority interrupts. In the case of UART-triggered switchback, the triggering is not generated by the UART-associated interrupt. This is because UART operating under PMM mode may not operate correctly to receive or transmit data. The switchback is thus initiated by the reception of the falling edge of the Start bit. The UART receive switchback is enabled only if the associated receive bit (SCON0.4 or SCON1.4) is set. The UART transmit initiated switchback is triggered when UART transmit buffer is loaded. Thus CPU operating under PMM mode recovers to normal mode automatically when it writes in the transmit buffer. Once it recovers, UART operates under normal frequency to correctly transmit the data.

The return of PMM mode after switchback must be activated manually with software. The exit of PMM mode occurs when WDT or external RSTN resets.

Since the purpose of the PMM mode is to save power consumption, the internal oscillator clock IOSC is recommended to be used as the system clock as IOSC consumes significantly less power than the crystal oscillator.

1.7.2 IDLE Mode

IDLE mode provides a further power saving than PMM mode by stopping the clock for CPU and its integrated peripherals while keeping the external peripherals at normal operating conditions. The external peripherals still function normally thus can generate interrupts that wake up the CPU from IDLE mode. The IDLE mode is introduced by setting Idle bit to 1.

When the CPU is idle mode, no processing is possible. All integrated internal peripherals such as T0/T1/T2, UART0, LIN-capable 16550-like UART2, and I²C Master are inaccessible during idling. The IDLE mode can be excited by hardware reset through RSTN pin or by external interrupts as well as the interrupts from external peripherals that are OR-ed with the external interrupts. The triggering external interrupts need be enabled properly. Upon exiting from IDLE mode, the CPU resumes operation as the clock is being turned on. CPU immediately vectors to the interrupt service routine of the corresponding interrupt sources that wake up the CPU. When the interrupt service routine completes, RETI returns to the program and immediately follows the one that invokes the IDLE mode. Upon returning from IDLE mode to normal mode, Idle bit in PCON is automatically cleared. As the purpose of the IDLE mode is to save power, the use of IOSC clock is strongly recommended in place of SYSCLK before entering IDLE mode since it consumes significantly less power than the crystal oscillator or other clock sources.

1.7.3 STOP Mode

STOP mode provides the lowest power consumption by stopping clocks to all components in the system. STOP mode is entered by setting STOP=1. To achieve minimum power consumption, before entering STOP mode, it is essential to turn off all peripherals and the current operating clock oscillators such as crystal oscillator and PLL. It is also important that the software switches to the IOSC clock and disables all other clock generators such as crystal oscillator or PLL clock generator before entering STOP mode. This is critical to ensure a smooth transition when resuming its normal operations. Selecting other clock sources, such as XTAL oscillator or PLL clock as CPU system clock may burden the system as the clock sources may take a significant amount of time to stabilize during the wakeup. Upon entering STOP mode, the system uses the last edge of IOSC clock to shut down the IOSC clock generator. The minimum power consumption state is achieved through this mechanism.

Hardware reset through RSTN pin or by interrupts generated via external pins (INT0 and INT1) or INT2 to INT8 brings the system out of STOP mode. Since all clocks are inactive, none of the peripherals like UART, Timers, I ²C master and slave, ADC, or LVD contribute to the exit of STOP mode. Peripherals like Analog comparator and RTC interrupt; however, can be used to trigger the exit of STOP mode as they are implemented asynchronously or their own clock sources.

The triggering interrupt source must be enabled and its Wake-up bit is set in the WKMASK register. External pins require LOW-level triggers; however the INT flags of on-chip external peripherals require HIGH-level triggers. The IOSC circuit is activated by triggering event and the CPU is woken up at the first IOSC clock edge. Please note that the IOSC is activated as soon as STOP mode exits. As CPU resumes the normal operation using the IOSC clock when an interrupt presents, the CPU immediately vectors to the interrupting service routine of the corresponding interrupt source. When the interrupt service routine completes, RETI returns to the program immediately to execute the instruction that invokes the STOP mode. The Stop bit in PCON is automatically cleared by hardware reset during the waking up.

Please note the wake-up control WKMASK register and interrupt enable registers IE and EXIE which are specifically responsible for the wake-up and interrupt. Extra attention should be taken while programming for coherent application design. In STOP mode, clocks of CPU and peripherals are disabled (except RTC). Therefore only external pins and peripherals such as analog comparator and RTC that do not require clock can be used to initiate the wake-up process. Peripherals such as UART, Timers, ¹²C master and slave, ADC, or LVD can not generate wake-up interrupt in this mode.

1.7.4 SLEEP Mode

In STOP mode, the main regulator providing 1.8V (VDDC) to internal logic, memory and flash circuits are still active. The regulator and its internal Bandgap reference circuits consumes approximately about 200uA. SLEEP mode is used to further reduce the standby power through turning off the regulator and reference circuits. The logic behavior of SLEEP mode is the same as STOP mode and is entered by setting both STOP and SLEEP bits to 1 in PCON register. In SLEEP mode, a very low-power back-up regulator is used to provide supply voltage to the internal logic, memory and flash circuits. The back-up regulator consumes about 10uA to 20uA, and can supply up to 1mA of load. The output voltage of the back-up regulator is lower than the main regulator, and typically is around 1.45V.

The exit of SLEEP mode is the same as exit of STOP mode by wake-up events, and exits directly back to normal operation and the main regulator is turned on. Note the enabling time of the main regulator is about 10usec, therefore, after wake-up from SLEEP mode, the software should be kept at NOP for at least 20usec before resuming. It is also recommended that if SLEEP mode is used, the decoupling capacitor on VDDC should contains at least 10uF.

1.7.5 Clock Control

The clock selection is defined by CKSEL register (0x8F). An IOSC is a critical component in MCU although not integrated in the CPU core. It is enabled except in STOP mode. An IOSC also handles critical timing conformance for flash programming and the default manufactured calibrated IOSC is set at 16MHz. Although users can manually reset the IOSC frequency but reset value should not deviate more than 50% from its typical setting to avoid flash performance problems.

An IOSC is recommended that for the transition of clock-source-switching to ensure a smooth and glitch-free transition. This is also true for switching among different power saving modes. Please note that when waking up from STOP mode, the clock selection is switched automatically to IOSC. If other clock sources are preferred, optional configurations are available through software set-up.

When switching clock sources, it is also important to note the crystal oscillator, real time clock and the phase lock loop take a significant amount of time to stabilize. The software needs to be designed to turn on the corresponding clock source first and wait for the stabilization time before CKSEL settings take place.

The typical power dissipation relationship to the CPU frequency is shown in the following graph.

The values of performance frequency in IDLE and PMM modes are close therefore the lines appear overlapped in the graph.

The IDD result does not include the power dissipation of the clock oscillator. The graph shows that during normal operation, the power dissipation increases approximately at ~0.36mA/MHz; in idle mode it increases at about ~0.2mA/MHz (the power dissipation still increases as the frequency increases due to operation of peripheral clock).

WARNING: If an uninstalled clock source is being selected, it may cause the system to hang. There is NO hardware protection against this peril. Therefore extreme precautions must be exerted during programming.

1.8 Break Point Controller

The CPU core also includes a Break Point Controller for software debugging purposes and handling exceptions. Program Counter break point triggers at PC address matching, and there are seven PC matching settings available. Single Step break point triggers at interaction return from an interrupt routine.

Upon the matching of break point conditions, the Break Point Controller issues BKP Interrupt for handling the break points. The BKP Interrupt vector is located at 0x7B. Upon entering the BKP ISR (Break Point Interrupt Service Routine), all interrupts and counters (WDT, T0, T1, and T2) are disabled To allow further interrupts and continuing counting, the BKP ISR must be enabled. At the exiting, the BKP ISR setting must be restored to resume normal operations.

This register is for reading the Break Points interrupt flags.

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STEP IF This bit is set when the Break Point conditions set by a new instruction fetching from an interrupt routine. This bit must be cleared by software.

PC7IF – PC1IF These bits are set when Break Point conditions are set by PC7 – PC1 address. These bits must be cleared by software.

BPINTE (A0E1h) Break Point Interrupt Enable Register R/W (0x00) TB Protected

This register controls the enabling of individual Break Points interrupt.

STEP IE Set this bit to enable Single Step event break point interrupt.

PC7IE – PC1IE Set these bits to enable PC7 to PC1 address match break point interrupts.

BPINTC (A0E2h) Break Point Interrupt Control Register R/W (0x00)

This register is reserved for other applications.

BPCTRL (A0E3h) DBG and BKP ISR Control and Status Register R/W (b'11111100)

When entering the DBG or BKP ISR (Interrupt Service Routine), all interrupts and timers are disabled. The enabled bits are cleared by hardware reset in this register. As the interrupts and timers are disabled, the ISR can process debugging requirement in a suspended state. If a specific timer should be kept active, it must be enabled by ISR after ISR entry. Before exit of DBG and BKP ISR, the control bits should be enabled to allow the timers to resume operating. This register should be modified only in Debug ISR.

- DBGINTEN Set this bit to enable all interrupts (except WDT interrupt). This bit is cleared automatically at the entry of DBG and BKP ISR. Set this bit to allow ISR to be further interrupted by other interrupts. This is sometimes necessary if DBG or BKP ISR needs to use UART or I²C, for example.
- DBGWDEN Set this bit to allow WDT counting during the DBG and BKP ISR. This bit should always be set before exiting the ISR.
- DBGT2EN Set this bit to allow T2 counting during the DBG and BKP ISR. This bit should always be set before exiting the ISR. This bit only controls the counting but not T2 interrupt.
- DBGT1EN Set this bit to allow T1 counting during the DBG and BKP ISR. This bit should always be set before exiting the ISR. This bit only controls the counting but not T1 interrupt.
- DBGT0EN Set this bit to allow T0 counting during the DBG and BKP ISR. This bit should always be set before exiting the ISR. This bit only controls the counting but not T0 interrupt.
- DBGST This bit indicates the DBG and BKP ISR status. This bit is set to 1 when entering DBG and BKP ISR. This signal should be cleared when exiting the DBG and BKP ISR. Checking this bit allows other interrupt routine to determine whether it is a sub-service of the DBG and BKP ISR.

PC1AL (A0F0h) Program Counter Break Point 1 Low Address Register R/W (b'00000000)

This register defines the PC low address for PC match break point 1.

PC1AH (A0F1h) Program Counter Break Point 1 High Address Register R/W (b'00000000)

This register defines the PC high address for PC match break point 1.

PC1AT (A0F2h) Program Counter Break Point 1 Top Address Register R/W (b'00000000)

This register defines the PC top address for PC match break point 1. PC1AT:PC1HT:PC1LT together form a 24 bit compare value of break point 1 for Program Counter.

PC2AL (A0F4h) Program Counter Break Point 2 Low Address Register R/W (b'00000000)

This register defines the PC low address for PC match break point 2.

PC2AH (A0F5h) Program Counter Break Point 2 High Address Register R/W (b'00000000)

This register defines the PC high address for PC match break point 2.

PC2AT (A0F6h) Program Counter Break Point 2 Top Address Register R/W (b'00000000)

This register defines the PC top address for PC match break point 2 PC2AT:PC2HT:PC2LT together form a 24-bit compare value of PC break point 2 for Program Counter.

PC3AL (A0F8h) Program Counter Break Point 3 Low Address Register R/W (b'00000000)

This register defines the PC low address for PC match break point 3.

PC3AH (A0F9h) Program Counter Break Point 3 High Address Register R/W (b'00000000)

This register defines the PC high address for PC match break point 3.

PC3AT (A0FAh) Program Counter Break Point 3 Top Address Register R/W (b'00000000)

This register defines the PC top address for PC match break point 3. PC3AT:PC3HT:PC3LT together form a 24-bit compare value of break point 3 for Program Counter.

PC4AL (A0FCh) Program Counter Break Point 4 Low Address Register R/W (b'00000000)

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This register defines the PC low address for PC match break point 4.

PC4AH (A0FDh) Program Counter Break Point 4 High Address Register R/W (b'00000000)

This register defines the PC high address for PC match break point 4.

PC4AT (A0FEh) Program Counter Break Point 4 Top Address Register R/W (b'00000000)

This register defines the PC top address for PC match break point 4. PC4AT:PC4HT:PC4LT together form a 24-bit compare value of break point 4 for Program Counter.

STEPCTR (A0FFh) Single Step Break Point Interrupt Control Register R/W (0x00)

The value "0x96" must be programed into this register to enable the single step break point interrupt

PC5AL (A0E4h) Program Counter Break Point 5 Low Address Register R/W (b'00000000)

This register defines the PC low address for PC match break point 5.

PC5AH (A0E5h) Program Counter Break Point 5 High Address Register R/W (b'00000000)

This register defines the PC high address for PC match break point 5.

PC5AT (A0E6h) Program Counter Break Point 5 Top Address Register R/W (b'00000000)

This register defines the PC top address for PC match break point 5. PC5AT:PC5HT:PC5LT together form a 24-bit compare value of break point 5 for Program Counter.

PC6AL (A0E8h) Program Counter Break Point 6 Low Address Register R/W (b'00000000)

This register defines the PC low address for PC match break point 6.

PC6AH (A0E9h) Program Counter Break Point 6 High Address Register R/W (b'00000000)

This register defines the PC high address for PC match break point 6.
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PC6AT (A0EAh) Program Counter Break Point 6 Top Address Register R/W (b'00000000)

This register defines the PC top address for PC match break point 6 PC6AT:PC6HT:PC6LT together form a 24-bit compare value of PC break point 6 for Program Counter.

PC7AL (A0ECh) Program Counter Break Point 7 Low Address Register R/W (b'00000000)

This register defines the PC low address for PC match break point 7.

PC7AH (A0EDh) Program Counter Break Point 7 High Address Register R/W (b'00000000)

This register defines the PC high address for PC match break point 7.

PC7AT (A0EEh) Program Counter Break Point 7 Top Address Register R/W (b'00000000)

This register defines the PC top address for PC match break point 7. PC7AT:PC7HT:PC7LT together form a 24-bit compare value of break point 7 for Program Counter.

Host or program can obtain the status of the break point controller through the current break point address and next PC address register. DBPCID[23-0] contains the PC address of just executed instruction when the break point occurs. DBNXPC[23-0] contains the next PC address to be executed when the break point occurs, therefore, it is usually exactly the same value of the break pointer setting.

DBPCIDL (A098h) Debug Program Counter Address Low Register RO (b'00000000)

DBPCIDH (A099h) Debug Program Counter Address High Register RO (b'00000000)

DBPCIDT (A09Ah) Debug Program Counter Address Top Register RO (b'00000000)

DBPCNXL (A09Bh) Debug Program Counter Next Address Low Register RO (b'00000000)

DBPCNXH (A09Ch) Debug Program Counter Next Address High Register RO (b'00000000)

DBPCNXT (A09Dh) Debug Program Counter Next Address Top Register RO (b'00000000)

1.9 Debug and ISP

The I²C Slave 2 (I2CS2) can be configured as the debug and ISP port for IS31CS8969. This is achieved by assigning a predefined debug ID for the I²C Slave address. When a host issues an I²C access to this special address, a DBG interrupt is generated. DBG Interrupt has the highest priority. The DBG interrupt vector is located at 0x83. DBG ISR is used to communicate with the host and is usually closely associated with BKP ISR.

SI2CDBGID (A0EFh) Slave I²C Debug ID Register R/W (b'00110110) TB Protected

DBGSI2C2EN DBGSI2C2EN=1 enables I2CS2 as debug port. When I2CS2 receives an access of I²C address matching SI2CDBGID[6:0], a debug interrupt is generated.

SI2CDBGID[6:0] Slave ²C ID address for debug function.

1.10 Watchdog Timer

The Watchdog Timer is a 30-bit timer that can be used by a system supervisor or as an event timer. The Watchdog timer can be used to generate an interrupt or to issue a system reset depending on the control settings. This section describes the register related to the operation of Watchdog Timer and its functions. The following diagram shows the structure of the Watchdog Timer. Note WDT shares the same clock with the CPU, thus WDT is disabled in IDLE mode or STOP mode however it runs at a reduced rate in PMM mode.

WDCON (0xD8) R/W (0x00)

WDIF WDT Interrupt Flag bit. This bit is set when the session expires regardless of a WDT interrupt is enabled or not. Note the WDT interrupt enable control is located in EIE (0xE8).4 EWDI bit. It must be cleared by software

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must be unlocked then and then followed by writing RWT bit to 1. If TA is still locked, the program can write 1 into RWT bit, but it does not reset the Watchdog timer.

CKCON (0x8E) R/W (0xC7)

WD[2:0] This register controls the time out value of WDT as the following table. The time out value is shown as follows and the default is set to maximum:

1.11 System Timers – T0 and T1

The CPU contains three 16-bit timers/counters, Timer 0, Timer 1 and Timer 2. In timer mode, Timer 0, Timer 1 registers are incremented every 12 SYSCLK period when the appropriate timer is enabled. In the timer mode, Timer 2 registers are incremented every 12 or 2 SYSCLK period (depending on the operating mode). In the counter mode, the timer registers are incremented every falling edge on their corresponding inputs: T0, T1, and T2. These inputs are read every SYSCLK period.

Timer 0 and Timer 1 are fully compatible with the standard 8051. Timer 0 and 1 are controlled by TCON (0x88) and TMOD (0x89) registers while each timer consists of two 8-bit registers TH0 (0x8C), TL0 (0x8A), TH1 (0x8D), TL1 (0x8B).

TF1 can also be cleared by software.

TR0 Timer 0 Run Control bit. Set to enable Timer 0, and clear to disable Timer 0. PINT1F/PINT1EG/PINT0F/PINT0EG

> These bits are related to configurations of the expanded interrupt PINT1 and PINT0. These are described in the Interrupt System section.

TMOD (0x89h) Timer 0 and 1 Mode Control Register R/W (0x00)

1.11.1 Mode 0

In this mode, TL serves as a 5-bit prescaler and TH functions as an 8-bit counter/timer, together working as a 13-bit counter/timer. The Mode 0 operation is shown in the following diagram.

1.11.2 Mode 1

Mode 1 operates the same way Mode 0 does, except TL is configured as 8-bit and thus forming a 16-bit counter/timer. This is shown as the following diagram.

1.11.3 Mode 2

Mode 2 configures the timer as an 8-bit reloadable counter. The counter is TL while TH stores the reload data. The reload occurs when TL overflows. The operation is shown in the following diagram:

1.11.4 Mode 3

Mode 3 is a special mode for Timer 0 only. In this mode, Timer 0 is configured as two separate 8-bit counters. TL0 uses control and interrupt flags of Timer 0,whereas TH0 uses control and interrupt flag of Timer 1. Since Timer 1's control and flag are occupied, Timer 2 can only be used for counting purposes such as Baud rate generating while Timer 0 is in Mode 3. The operation flow of Mode 3 is shown in the following diagram.

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1.12 System Tir

1.12 System Timer – T2

Timer 2 is fully compatible with the standard 8052 timer 2. Timer 2 can be used as the reloadable counter, capture timer, or baud rate generator. Timer 2 uses five SFRs as counter registers, capture registers and a control register.

Timer 2 can be configured in three modes of operations – Auto-reload Counter, Capture Timer, or Baud Rate Generator. These modes are defined by RCLK, TCLK, CPRL2 and TR2 bits of T2CON registers. The definition is illustrated in the following table:

The block diagram of the Timer 2 operating in Auto-reload Counter and Capture Timer modes are shown in the following diagram:

The block diagram of the Timer 2 operating in Baud Rate Generator is shown in the following diagram:

1.13 System Timer – T3 and T4

Both Timer 3 and Timer 4 are simple 16-Bit reload timers or free-run counters and are clocked by the system clock. The block diagram is shown as below.

T34CON (0xCFh) Timer 3 and Timer 4 Control and Status Register R/W (0x00)

Timer 3 Run Control bit. Set to enable Timer 3, and clear to stop Timer 3. Timer 3 Interrupt Enable bit. T3IEN=0 disable the Timer 3 overflow interrupt T3IEN=1 enable the Timer 3 overflow interrupt

TL3 (0xAEh) Timer 3 Low Byte Register 0 R/W (0X00)

TH3 (0xAFh) Timer 3 High Byte Register 0 R/W (0X00)

TL4 (0xACh) Timer 4 Low Byte Register 0 R/W (0X00)

TH4 (0xADh) Timer 4 High Byte Register 0 R/W (0X00)

T3[15-0] and T4[15-0] function differently when been read or written. When written in auto-reload mode, its reload value register is written, and in free-run mode, the counter value is written immediately. When been read, the return value is always the present counter value. There is no snapshot buffer in the read operation, so software should always read the high byte then the low byte.

1.14 System Timer – T5

T5 is a 24-Bit simple timer. It can select four different clock sources and can be used for extended sleep mode wake up. The clock sources include IOSC and SIOSC. T5 can be configured either as free-run mode or auto-reload mode. Timer 5 does not depend on the SYSCLK, therefore it continues to count under STOP or SLEEP mode if the clock source is present. The following diagram shows the block diagram of Timer 5.

T5CON (0xA003h) Timer 5 Control and Status Register R/W (0x00)

T5SEL[1-0] Timer 5 Clock Selection bits.

 $T5SEL[1-0] = 00$, IOSC $T5SEL[1-0] = 01$, IOSC

TL5 (0xA004) Timer 5 Low Byte Register 0 R/W (0X00)

 $T50511401 - 40, 01000$

TH5 (0xA005) Timer 5 Medium Byte Register 0 R/W (0X00)

TT5 (0xA006) Timer 5 High Byte Register 0 R/W (0X00)

T5[23-0] functions differently when been read or written. When written in auto-reload mode, its reload value register is written, and in free-run mode, the counter value is written immediately. When been read, the return value is always the present counter value. There is no snapshot buffer in the read operation, so software should always read the high byte then the low byte.

1.15 Multiplication and Division Unit (MDU)

MDU provides acceleration on unsigned integer operations of 16-bit multiplications, 32-bit division, and shifting and normalizing operations. The following table shows the execution characteristics of these operations. The MDU does not contain the operation completion status flag. Therefore the most efficient utilization of MDU uses NOP delay for the required clock time of the MDU operation types. The number of the clock cycles required for each operation is shown in the following table and it is counted from the last write of the writing sequence.

The MDU is accessed through MD0 to MD5 that contains the operands and the results, and the operation is controlled by ARCON register.

ARCON (0xFF) MDU Control R/W 00000000

SC4-0 Shift Count Control and Result bit. If SC0-4 is written with 00000, the normalization operation performed by MDU. When the normalization is completed, SC4-0 contains the number of shift performed in the normalization. If SC4-0 is written with a non-zero value, then the shift operation is performed by MDU with the number of shift specified by SC4-0 value.

MD0 (0xF9) MDU Data Register 0 R/W (0X00)

MDU operation consists of three phases.

Loading MD0 to MD5 data registers in an appropriate order depending on the operation.

Execution of the operations.

Reading result from MD0 to MD5 registers.

The following list shows the MDU read and write sequences. Each operation has its unique writing sequence and reading sequence of MD0 to MD5 registers therefore a precise access sequence is required.

1.15.1 Division – 32-bit divide by 16-bit or 16-bit divide by 16-bit

WR ND5[7-0]

Follow the following write-sequence. The first write of MD0 resets the MDU and initiates the MDU error flag mechanism. The last write incites calculation of MDU.

Write MD0 with Dividend LSB byte

Write MD1 with Dividend LSB+1 byte

Write MD2 with Dividend LSB+2 byte (ignore this step for 16-bit divide by 16-bit)

Write MD3 with Dividend MSB byte (ignore this step for 16-bit divide by 16-bit)

Write MD4 with Divisor LSB byte

Write MD5 with Divisor MSB byte

Then follow the following read-sequence. The last read prompts MDU for the next operations.

Read MD0 with Quotient LSB byte Read MD1 with Quotient LSB+1 byte Read MD2 with Quotient LSB+2 byte (ignore this step for 16-bit divide by 16-bit) Read MD3 with Quotient MSB byte (ignore this step for 16-bit divide by 16-bit) Read MD4 with Remainder LSB byte Read MD5 with Remainder MSB byte Read ARCON to determine error or overflow condition

Please note if the sequence is violated, the calculation may be interrupted and result in errors.

1.15.2 Multiplication – 16-bit multiply by 16-bit

Follow the following write sequence.

Write MD0 with Multiplicand LSB byte

Write MD4 with Multiplier LSB byte

Write MD1 with Multiplicand MSB byte

Write MD5 with Multiplier MSB byte

Then follow the following read sequence.

Read MD0 with Product LSB byte

Read MD1 with Product LSB+1 byte

Read MD2 with Product LSB+2 byte

Read MD3 with Product MSB byte

Read ARCON to determine error or overflow condition

1.15.3 Normalization – 32-bit

 Normalization is obtained with integer variables stored in MD0 to MD3. After normalization, all leading zeroes are removed by shift left operations. To start the normalization operation, SC4-0 in ARCON is first written with 00000. After completion of the normalization, SC4-0 is updated with the number of leading zeroes and the normalized result is restored on MD0 to MD3. The number of the shift of the normalization can be used as exponents. The following write sequence should be followed. The last write to ARCON initiates the normalization operations by MDU.

Write MD0 with Operand LSB byte Write MD1 with Operand LSB+1 byte Write MD2 with Operand LSB+2 byte Write MD3 with Operand MSB byte Write ARCON with SC4-0 = 00000

Then follow the following read sequence.

Read MD0 with Result LSB byte Read MD1 with Result LSB+1 byte

Read MD2 with Result LSB+2 byte

Read MD3 with Result MSB byte

Read SC[4-0] from ARCON for normalization count or error flag

1.15.4 Shift – 32-bit

Shift is done with integer variables stored in MD0 to MD3. To start the shift operation, SC4-0 in ARCON is first written with shift count and SLR with shift direction. After completion of the Shift, the result is stored back to MD0 to MD3. The following write sequence should be followed. The last write to ARCON initiates the normalization operations by MDU.

Write MD0 with Operand LSB byte

Write MD1 with Operand LSB+1 byte

Write MD2 with Operand LSB+2 byte

Write MD3 with Operand MSB byte

Write ARCON with SC4-0 = Shift count and SLR with shift direction

Then follow the following read sequence.

Read MD0 with Result LSB byte

Read MD1 with Result LSB+1 byte Read MD2 with Result LSB+2 byte

Read MD3 with Result MSB byte Read ARCON's for error flag

1.15.5 MDU Flag

The error flag (MDEF) of MDU indicates improperly performed operations. The error mechanism starts at the first MD0 write and finishes with the last read of MD result register. MDEF is set if current operation is interrupted or restarted by improper write of MD register before the operation completes. MDEF is cleared if the operations and proper write/read sequences successfully complete. The overflow flag (MDOV) of MDU indicates an error of operations. MDOV is set if

the divisor is zero

Multiplication overflows

Normalization operation is performed on already normalized variables (MD3.7 = 1)

1.16 Serial Port – UART0

UART0 is full duplex and fully compatible with the standard 8052 UART. The receive path of the UART0 is double-buffered that can commence reception of second byte before previously received byte is read from the receive register. Writing to SBUF0 loads the transmit register while reading SBUF0, reads a physically separate receive register. The UART0 can operate in four modes: one synchronous (Mode 0) and three asynchronous modes (Mode 1, 2, and 3). Mode 2 and Mode 3 share a special provision for multi-processor communications. This feature is enabled by setting SM2 bit in SCON0 register. The master processor first sends out an address byte, which identifies the slave. An address byte differs from a data byte in the $9th$ bit: 1 defines an address byte whereas 0 defines a data byte. When SM2 is set to 1, no slave can be interrupted by a data byte. An address byte can interrupt slaves. The addressed slave clears its SM2 bit and prepares to receive the following incoming data bytes. The slaves that are not addressed leave their SM2 set and ignore the incoming data. The UART0-related registers are SBUF0, SCON0, PCON, IE, and IP.

SCON0 (0x98h) UART0 Configuration Register R/W (0x00)

SBUF0 (0x99h) UART0 Data Buffer Register R/W (0x00)

SBUF0 is used for both transmission and reception. Writing a data byte into SBUF0 puts this data in UART0's transmit buffer and starts a transmission. Reading a byte from SBUF means data being read from the UART0's receive buffer.

1.16.1 Mode 0

Mode 0 is a simple synchronous shift register mode. TXD0 outputs the shift clock, which is fixed at CPUCLK/12. RXD0 is a bidirectional I/O port that serves as a data-shifting port. To utilize this mode, TXD0 pin must be enabled as an output pin, while RXD0 needs to be configured as an open-drain type of I/O port. The shift data changes at the rising edge of the shift clock and is valid at the falling edge of the shift clock. The transmission starts when a new byte is written in SBUF0 as TI is cleared to 0. When the byte is transmitted, TI is set and the UART0 waits for the next byte to be transmitted. The reception is initiated by setting REN=1 and RI cleared to 0. When a byte is received, RI is set by UART0.

1.16.2 Mode 1

8-bit UART mode. RXD0 is the serial input and TXD0 is the serial output. To utilize this mode, the corresponding RXD0 and TXD0 pin configuration should also be set correctly. 10-bit data (including a Start bit, 8 data bit, and a Stop bit) are transferred. For UART0, the baud rate is set by Timer 1 or Timer 2 overflow rate. The control is determined by SMOD0.PCON, and RCLK.T2CON, TCLK.T2CON. When SMOD0.PCON is 1, Timer 1 overflow is selected, and SMOD0.PCON is 0, Timer 1 overflow rate divided by 2 is selected. And if RCLK.T2CON, or TCLK.T2CON is set, the Timer 2 overflow rate is selected and overwrites the SMOD0 setting.

1.16.3 Mode 2

9-bit UART mode. RXD0 is the serial input and TXD0 is the serial output. To utilize this mode, the corresponding RXD0 and TXD0 pins should be configured correctly. 11-bit data including a Start bit (always 0), 8 data bits, a programmable 9th bit, and a Stop bit (always 1) are transferred. The 9th bit can be configured as a parity bit configured by software through TB8 in SCON0. The received 9th bit can be read from TB8. The software determines the correctness of the parity check. The baud rate in Mode 2 is fixed at 1/32 or 1/64 of CPU clock. This is controlled by SMOD0 in PCON register.

1.16.4 Mode 3

Similar to Mode 2 (9-bit UART mode). RXD0 is the serial input and TXD0 is the serial output. To utilize this mode, the corresponding RXD0 and TXD0 pins should also be configured properly. 11-bit data including a Start bit (always 0), 8 data bits, a programmable 9th bit, and a Stop bit (always 1) are transferred. The 9th bit can serve as a parity bit configured by software through TB8 in SCON0. The received 9th bit can be read from TB8. The software determines the correctness of the parity check. The mechanism of the baud rate control in Mode 3 is similar to which in Mode 1. that is determined by Timer 1 or Timer 2 overflow and is set by SMOD0, and T2CON.

1.17 I²C Master

The I²C master controller provides the interface to I²C slave devices. It can be programmed to operate with arbitration and clock synchronization to allow it to operate in multi-master configurations. The master uses SCL and SDA pins. The controller contains a built-in 8-bit timer to allow various I²C bus speed. The maximum I2C bus speed is limited to SYSCLK/12.

I2CMTP (0xF7h) I²C Master Time Period R/W (0X00)

This register set the period time of I²C bus clock – SCL. The SCL period time is set according to SCLPERIOD = $8 * (1 + I2CMTP) * CPUCLK$ PERIORD when the I2CMTP[7-0] is equal to or larger than 0x01. If I2CMTP[7-0] = 0x00, the maximum I2C bus speed is limited to SYSCLK/12.

I2CMSA (0xF4) I²C Master Slave Address R/W (0X00)

RD	SA[6-0]								
WR	SA[6-0]							RS	
	Slave Address $SA6-01$ defines the slave address the ^{12}C master uses to communicate $SAI6-01$								

 $\mathsf{SA}[\mathsf{6}\text{-}\mathsf{0}] \hspace{25pt} \mathsf{S}_\mathsf{A}[\mathsf{6}\text{-}\mathsf{0}]$ defines the slave address the I C_C master uses to communication RS Receive/Send Bit. RS determines if the following operation is to RECEIVE (RS=1) or SEND $(RS=0)$.

I2CMBUF (0xF6) I²C Master Data Buffer Register R/W (0X00)

I2CMBUF functions as a transmit-data register when written and as a receive-data register when read. When written, TD is sent to the bus by the next SEND or BURST SEND operations. TD[7] is sent first. When read, RD contains the 8-bit data received from the bus upon the last RECEIVE or BURST RECEIVE operation.

I2CMCR (0xF5) I²C Master Control and Status Register R/W (0X00)

The I2CMCR register is used for setting control when it is written, and as a status signal when read.

I2CMRST Writing 1 to this bit forces the I2CM to perform reset and clear its internal state machine. At the end of the initialization, all SFRs will return to the default value. This bit is cleared automatically by hardware.

INFILEN Input Noise Filter Enable. When IFILEN is set, pulses shorter than 50 nsec on inputs of SDA and SCL are filtered out.

IDLE This bit indicates that ¹²C master is in the IDLE mode.

BUSY This bit indicates that I²C master is receiving or transmitting data, and other status bits are not valid.

BUSBUSY This bit indicates that the external I²C bus is busy and access to the bus is not possible. This bit is set/reset by START and STOP conditions.

- ERROR This bit indicates that an error occurs in the last operation. The errors include slave address was not acknowledged, or transmitted data is not acknowledged, or the master controller loses arbitration.
- ADDRERR This bit is automatically set when the last operation slave address transmitted is not acknowledged.
- DATAERR This bit is automatically set when the last operation transmitted data is not acknowledged. ARBLOST This bit is automatically set when the last operation I^2C master controller loses the bus arbitration.

START, STOP, RUN and HS, RS, ACK bits are used to drive I²C Master to initiate and terminate a transaction. The Start bit generates START, or REPEAT START protocol. The Stop bit determines if the cycle stops at the end of the data cycle or continues to a burst. To generate a single read cycle, the designated address is written in SA, RS is set to 1, ACK=0, STOP=1, START=1, RUN=1 are set in I2CMCR to perform the operation and then STOP. When the operation is completed (or aborted due to errors), I²C master generates an interrupt. The ACK bit must be set to 1. This causes the controller to send an ACK automatically after each byte transaction. The ACK bit must be reset when set to 0 when the master operates in receive mode and not to receive further data from the slave devices.

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The following table lists the permitted control bits combinations in master TRANSMITTER mode.

The following table lists the permitted control bits combinations in master RECEIVER mode.

All other control-bit combinations not included in three tables above are NOP. In Master RECEIVER mode, STOP should be generated only after data negative ACK executed by Master or address negative ACK executed by slave. Negative ACK means SDA is pulled low when the acknowledge clock pulse is generated.

2. Flash Controller

The flash controller connects the CPU to the on-chip embedded FLASH memory. The FLASH memory functions as the program storage as well as non-volatile data storage. The program access of the FLASH does not require any special attention. When the FLASH is used as data storage, the software sends commands to the FLASH controller through the XFR registers. And when the FLASH controller processes these commands, CPU is held idle until the command is executed. The embedded Flash memory contains two blocks – Main Memory and Information Block (IFB). The Main Memory is 32K x 8 with uniform 1024 Byte page (sector) size. The Information Block is 256 Byte and sits in a separate sector.

The commands performed by a Flash Controller are defined in FLSHCMD registers. The defined operations allow the user program to use on-chip flash as a program memory, and a non-volatile data memory in In-System-Programming as well as In-Application-Programming. The maximum flexibility of the on-chip flash memory can be achieved through user program. The manufacturer provides a default ISP boot program located on the top sectors of the flash. The preset ISP boot program can be used or modified or replaced based on application requirements.

FLSHCMD (A020h) Flash Controller Command Register R/W 10000000 TB Protected

For all commands, the address of the flash is composed from FLSHADM:FLSHADH:FLSHADL and the data is referred at FLSHDAT registers. The erase command operation is sector-based, the address of the sector is determined from the high order address bits. For example, to point to the sector of 0x0C000-0xCFFF, the upper 8 bits "0C" are used. And the erase command erases the whole addressed sector contents. For Erase and Write command, the Flash Controller also checks if the destination address falls within the protection zone defined by CNTPCTL and

CNTPCTH registers. If it is protected, the Flash Controller does not execute the command and return with FAIL result bit. For IFB Byte-write, the Flash Controller does not execute the command and return with Fail result bit if the byte address falls into manufacturer data range. Please also note the Fuse block is used for manufacturer to store manufacturing related and calibration data and thus can only be read and not writable or erasable. Fuse block can only be erased or written under writer mode.

ISPCLKF (A024h) Flash Command Clock Scaler R/W 00100101 TB Protected

ISPCLKF[7-0] configures the clock time base for generation of Flash erase and write timing. ISPCLK = SYSCLK * (ISPCLKF[7-0]+1)/256. For correct timing, ISPCLK should be set to approximately at 2MHz.

FLSHDAT (A021h) Flash Controller Data Register R/W 00000000 TB Protected

FLSHADL (A023h) Flash Controller Low Address Data Register R/W 00000000 TB Protected

FLSHADH (A022h) Flash Controller High Address Data Register R/W 00000000 TB Protected

FLSHADM (A012h) Flash Controller MSB Address Data Register R/W 00000000 TB Protected

A very common problem of embedded flash memory is when being used as both data and program storage, which leads to content loss due to software or other problems caused by program flow or noise. It induces executions of modifying stored contents. The design of Flash controller takes into considerations of these events and provides further protection to avoid accidental erasure or modifications of critical information or software codes. When a command is sent to the Flash Controller through FLSHCMD register, the controller checks whether the destination of the command falls in the content protection zones. If it falls within the protection zones, the flash control aborts its operations and returns with command failure message. Two protections zones are defined by 00000 to CNTPCTL, and CNTPCTH to 1FFFF.

CNTPCTL (A025h) Flash Content Protection Low Zone Register R/W 11111111 TB Protected

This register defines the high bound address from 00000h of the flash which is protected against erasure or modifications. The data is processed in 256 Byte increments. The protected region is defined from 00 to (CNTPCTL1 -1) Note that CNTPCTL defaults to FFh which protects the whole 32KB of flash memory. User program needs to write the appropriate data into CNTPCTL to enable erase and write access.

CNTPCTH (A026h) Flash Content Protection High Zone Register R/W 00000000 TB Protected

This register defines the low bound address from 0xFFFF of the flash to be protected against erasure or modifications. The data is processed in the increments of 256 Byte. The protected region is greater than (CNTPCTH+1) and less than or equal to FF. Note that CNTPCTH defaults to 00h which means the protection of 64KB of flash memory is on.

User program needs to write the appropriate data into CNTPCTH by reading IFB-protected information to protect boot code and expand the protection zone under application considerations.

There is an additional content protection against internal program. This protects sensitive data from unauthorized access. The protection range is from 0x1000 to 0xFFFF of embedded flash memory. The protection is achieved by two special registers, INTPCT1 (0xA013) and INTPCT2 (0xA014). After any reset condition such as power-up, RSTN, LVR, or WDT reset, INTPCT1 and INTPCT2 are initialized to 0x00. The bits in INTPCT1 and INTPCT2 can only be written to "1". When the embedded flash memory has been protected, this means accessing this protected range returns with 0x00 either by program instruction such as "MOVC" or by Flash Main Memory Byte Read access. The internal protection is by default not turned on after reset because both INTPCT1 and INTPCT2 are 0x00. Both registers are protected by TB. To turn on the internal protection, INTPCEN must be enabled by writing a "0x80" into enable the Internal Protection function. When INTPCEN is set, user can define the protection range by program INTPCT1 or INTPCT2. The INTPCEN will be cleared and stuck-on zero after setting any protection range. In other words, the protection range is single time programmable. Once the protection is turned on, it can't be turned off or modified because INTPCEN is stuck-on zero. To restore unprotected state, the chip must go through a reset. The internal protection should be enabled with extreme cautiousness. It is important that once it is turned on, program execution should not reach the protected zone, otherwise unpredicted program errors may occur.

INTPCT1 (A013h) Internal Protection Enable Register R/W 00000000 TB Protected

This register can be written to "1" only. Writing "0" into any bit of this register does not alter the content. This register is cleared to 0x00 after reset. And the value of this register can be cleared only by a reset.

INTPCT2 (A014h) Internal Protection Enable Register R/W X000000 TB Protected

This register can be written to "1" only. Writing "0" into any bit of this register does not alter the content. This register is set to 0x00 after a reset. And the value of this register can be set only by a reset.

 INTPCEN INTPCT1/INTPCT2 updates enable. This bit is single time programmable. After setting any protection range, this bit will be stuck-on zero and prohibit another protection updating. Because the INTPCEN is disabled in the initiation, a "0x80" must be programmed into INTPCT2 to enable INTPCEN. When INTPCEN is set, the expected program protection can be defined by setting INTPC1 or INTPC2[6-0].

The following table summarizes the internal program protection with different INCTPCT1 and INTPCT2 settings and protected range from program read access. Because IS31CS8969 only has 32KB Flash Memory, the setting for the Protected Region from 0x08000 to 0x0FFFF is reserved.

3. Essential Analog Blocks

3.1 On-Chip 1.8V Regulator

An on-chip regulator is used to provide supply for core logic and internal E-Flash memory (VDDC). The regulator is partitioned into a back-up regulator and a main regulator. The main regulator is enabled only in normal, and STOP modes, and disabled in SLEEP mode when the back up regulator is turned on. The main regulator consumes about 100uA itself, while the back-up regulator consumes about 5uA. The back-up regulator can supply up to 500uA of current therefore it is important all peripheral circuits should be kept off during SLEEP mode. After reset, the main regulator defaults to on state. The regulator requires an external capacitor, which should be connected to VDDC pin. A minimum of 1uF plus a 0.1uF in parallel is required for the stability of the regulator. The main regulator outputs about can be adjusted by REGTRM (with default at its maximum) and the back-up regulator outputs about 1.60V. A manufacturer calibrated value of REGTRM for 1.8V is stored in IFB.

REGTRM (A000h) Regulator Trim Register R/W 11111111 TB Protected

			∼			-					
RD	REGTRM[7-0]										
WR	REGTRM[7-0]										

REGTRM[7-0] Trim Register for main 1.8V regulator.

REGTRM[7-0]=FF corresponds to maximum output level. REGTRM[7-0]=00 corresponds to minimum output level. The in-between value in general is linear to the output level. Typically the maximum is around 2.0V while the minimum is around 1.6V

3.2 1.0V Reference (BGREF)

This reference is derived from on-chip band-gap reference and has very low temperature coefficient and supply voltage dependency. Typical value of BGREF is 1.0V. This reference is connected to ADC input and can be used for calibration.

3.3 Precision Internal 16MHz Oscillator (IOSC)

The internal oscillator is a very important peripheral as it provides the default clock source after reset and other critical timing. The internal oscillator has the salient features that it behaves well during the enable and disable transient. No clock glitches or extra clock edge is generated during the on/off transition, and the oscillator can reach to stable oscillations within very short time typically within 10 cycles. The IOSC consumes around 300uA when enabled. The IOSC is always enabled except entering into STOP mode. And in STOP mode when it is disabled, IOSC only consumes less than 1uA standby current.

Similar to the on-chip regulator, IOSC also exhibits chip-to-chip variations. A calibrated value that set IOSC at 16MHz +/- 2% is stored in IFB. The user program can set this value to IOSC trimming register, IOSCITRM (A001h) and IOSCVTRM (A002h). The IOSC frequency has very little variations over the operation range (-40o – 85oC and VDD = $2.5V - 5.5V$). The variation is typically less than $+/-2%$ over the operation conditions. It is possible that user program to set a different frequency other than 16MHz as long as user program provide a calibration method to set IOSC frequency at the desired value at typical operation condition, and it will be stable and accurate over the entire operation range. Please note that the trimming register will be set to its default value after resets, the user program must reinitialize to its calibrated value. The total trim range of the IOSC is roughly from 7MHz up to 24MHz.

The IOSC is also equipped with Spread Spectrum capability for EMI sensitive applications. The SS deviation can be controlled to fit various requirements. However, once SS is enabled, the accuracy of IOSC cannot be maintained.

IOSCITRM (A001h) IOSC Coarse Trim Register R/W 00000010 TB Protected

$SSA[1-0] = 01, +/-8$ $SSA[1-0] = 00, +/- 4$

ITRM[1-0] ITRM[1-0] is the coarse trimming of the IOSC.

IOSCVTRM (A002h) IOSC Fine Trim Register R/W 10001110 TB Protected

This register provides fine trimming of the IOSC frequency. The higher the value of IOSCVTRM, the lower the frequency is.

The manufacturer trim value is stored in IFB and is trimmed to 16MHz. The user program provides the freedom to set the IOSC at a preferred frequency as long as the program is able to calibrate the frequency. Once set, the IOSC frequency has accuracy deviation within +/- 2% over the operation conditions. The following lists the range of the typical IOSC frequency for each trimming setting.

ITRM[1-0]=00, F_IOSC= 14.0MHz - 9.5MHz - 7.0MHz (VTRM[7-0]= 00 - 80 - FF) ITRM[1-0]=01, F_IOSC= 18.0MHz - 12.5MHz - 9.3MHz (VTRM[7-0]= 00 - 80 - FF) ITRM[1-0]=10, F_IOSC= 22.0MHz - 15.5MHz - 11.5MHz (VTRM[7-0]= 00 - 80 - FF)

ITRM[1-0]=11, F_IOSC= 25.5MHz - 18.5MHz - 13.5MHz (VTRM[7-0]= 00 - 80 - FF)

The trimming of the IOSC should use the following procedure to obtain the default setting for 16MHz.

Set $ITRM = 01$, and

Set VTRM = 00, measure frequency

Set VTRM = FF, measure frequency

Set VTRM = 7F, measure frequency

Use binary search to obtain the closest setting for 16MHz

Note: The frequency versus VTRM setting is monotonic. When VTRM = 00, the frequency is highest, and when VTRM = FF, the frequency is lowest.

A hardware Spread Spectrum can be enabled for the IOSC. This is controlled by SSC[3-0]. When SSC[3-0] = 0, the spread spectrum is disabled, and IOSC functions normally as a fixed frequency oscillator. If SSC[3-0] is not 0, then Spread Spectrum is enabled and IOSC frequency is swept according to the setting of SSC[3-0] and SSA[1-0]. The spread is achieved by varying the actual VTRM output to the oscillator circuit thus effectively changes the oscillation frequency. The effect of SSC[3-0] and SSA[1-0] is shown in the following graph.

When Spread Spectrum is enabled, the actual controlling output to IOSC is VTRM[7-0] +/- SSA. This is shown in the graph above as the bold curve. The above example shows $SSA[1:0] = 01$, and the deviation is $+/$ - 8. SSC[3-0] defines the update time in IOSC cycles. Then we can calculate the period of a complete sweep is 4*SSC*(2SSA+1) IOSC cycles, and we can obtain the sweep frequency from this period. When SS is enabled, the frequency of IOSC varies according to time and setting, therefore, the accuracy of IOSC frequency cannot be guaranteed. Please also note that VTRMOUT is VTRM[7-0] +/- SSA but is bounded by 0 and 255. Therefore for a linear non-clipped sweep, VTRM[7-0] needs to be within the range of SSA – 256-SSA, for example, SSA[10] = 01, then SSA is 8. VTRM[7-0] should be in the range from 8 to 248 to prevent the sweep from being clipped. As Spread Spectrum suggests, the total EMI energy is not reduced, but the energy is spread over wider frequency. It is recommended that SS usage should be carefully evaluated and the setting of spread amplitude and the sweep frequency should be chosen carefully for reducing EMI effect.

3.4 Slow Internal Oscillator (SIOSC)

The SIOSC is a 32 KHz low power internal R/C oscillator. The oscillator consumes about 2uA and is always enabled. SIOSC can be used as system clock or as T5 clocking source to provide extended long period timing for wake up purpose. The accuracy of SIOSC is not guaranteed and typically lies within 20KHz to 50KHz, and variations temperature is about +/- 30%.

3.5 Supply Low Voltage Detection (LVD)

The supply Low Voltage Detection (LVD) circuit detects VDD < VTH condition and can be used to generates an interrupt or reset condition. LVD defaults to disabled state to save power. An enabled LVD circuit consumes about 100uA to 200uA. The LVDTHD[6-0] sets the compare threshold according to the following equation when LVDTHV is the detection voltage.

LVDTHV = BGREF/(0.168 + 0.55 *(128-LVDTH[6-0])/128))

Where typical value of BGREF is 1.0V. The following graph shows the plot of the equations and four device characteristics.

The manufacturing process also stores the LVDTHD value for detection of 4.0V and 3.0V in IFB. The user program can either use the equation to calculate the desired detection value or uses these two values.

LVDCFG (A010h) Supply Low Voltage Detection Configuration Register R/W 10000000 TB Protected

LVTIF is set by hardware when LVD detection occurs and must be cleared by software.

LVDTHD (A011h) Supply Low Voltage Detection Threshold Register R/W X1111111 TB Protected

LVDTHD = 0x00 will set the detection threshold at its minimum, and LVDTHD = 0x7F will set the detection threshold at its maximum.

LVDHYS (A016h) Supply Low Voltage Detection Threshold Hysteresis Register R/W 00000000 TB Protected

To ensure a solid Low Voltage detect, a digital controlled hysteresis is used. If LVDHYEN=1, when LVD is asserted a new threshold defined by LVDHYS[6-0] replaces LVDTHD[6-0]. In typical applications, LVDHYS[6-0] should be set to be greater than LVDTHD[6-0].

4. I²C Slave Controller 2 (I2CS2)

The I²C Slave Controller 2 has dual functions – as a debug port for communication with host or as a regular ¹²C slave port. Please note both functions can coexist. Also the I²C Slave controller support sthe clock stretching functions¹

The debug accessed by the host is through ¹²C slave address defined by SI2CSDBGID register and enabled by DBGSI2C2EN=1. When I2CS2 received this address match, a DBG interrupt is generated. This is described in the Debug and ISP sections. If DBGSI2C2EN=0, then I2CS2 functions as a regular I²C slave. The address of the slave is set by I2CSADR2 register. The MSB in I2CSADDR2 is the enable bit for the I²C slave controller and I2CSADR2[6-0] specifies the actual slave address.

In receive mode, the controller detects a valid matching address and issues an ADDRMI interrupt. At the same time, the data bit on SDA line is shifted into receive buffer. The RCBI interrupt is generated whenever a complete byte is received and is ready to be read from I2CSDAT. If for any reason, the software does not respond to RCBI interrupt in time (i.e. RCBI is not cleared), and a new byte is received, the controller either forces an NACK response on I²C (if CLKSTREN bit is not set) or by pulling and holding SDA low (if CLKSTREN bit is set) to stretch the SCL low duration to force the master into a wait state. In clock stretching mode, SCL is released when the software responds to RCBI interrupt and cleared RCBI flag.

In transmit mode, the controller detects a valid matching address and issue an ADDRMI interrupt. At the same time, the data preloaded in the transmit data register through I2CSDAT is transferred to the transmit shift register and is serially shifted out onto SDA line. When this occurs, the controller generates a TXBI interrupt to inform the software that a new byte can be written into I2CSDAT. When the shift register is empty and ready for the next transmit, the slave controller checks if the new byte is written to the I2CSDAT. If TXBI is not cleared, it indicates lack of new data and the slave controller holds SCL line low to stretch the current clock cycle if CLKSTREN is set. If the clock stretching is not enabled, the slave controller takes the old byte into the shift register and replies with NACK, thus causing data corruption. On the other hand, if the master returns the NACK after the byte transfer, this indicates the end of data to the I²C slave. In this case, the I²C slave releases the data line to allow the master to generate a STOP or REPEAT START.

The I²C slave controller also implements the input noise spike filter. The INFILEN bit in the I2CSCON register enables this. The filter is implemented using digital circuit. When INFILEN is set, the spikes less than 1/2 EPPCLK period on the input of SDA and SCL lines are filtered out. If INFILEN is low, no input filtering is done. The following registers are related to I²C Slave Controller.

I2CSCON2 (0xDB) I2CS2 Configuration Regiter R/W (0x00)

I2CSST2 (0xDC) I2CS2 Status Register R/W (0x00)

I2CSADR1 (0xDD) I2CS2 Slave Address1 Register R/W (0x00)

I2CSADR2 (0xDF) I2CS2 Slave Address2 Register WO (0x00)

5. EUART2 with LIN Controller (EUART2)

LIN-capable 16550-like EUART2 is an enhanced UART controller (EUART) with separate transmit and receive FIFO. Both transmit and receive FIFO are 15-bytes deep and can be parameterized for interrupt triggering. The addition of FIFO significantly reduces the CPU load to handle high-speed serial interface. Transmit FIFO and receive FIFO have respective interrupt trigger levels that can be set based on optimal CPU performance adjustment. The EUART2 also has dedicated 16-bit Baud Rate generator and thus provides accurate baud rate under wide range of system clock frequency. The EUART2 also provides LIN extensions that incorporate message handling and baudrate synchronization. The block diagram of EUART2 is shown in the following.

The following registers are used for configurations of and interface with EUART2.

SCON2 (0xC2) UART2 Configuration Register R/W 0x00

SFIFO2 (0xA5) UART2 FIFO Status/Control Register R/W 0x00

Receive and transmit FIFO can be reset by clear FIFO operation. This is done by setting BR[11-0]=0 and EUARTEN=0. This also clears RFO, RFU and TFO interrupt flags without writing the interrupt register. The LIN counter LCNTR is also cleared.

SBUF2 (0xA6) UART2 Data Buffer Register R/W 0x00

This register is the virtual data buffer register for both receive and transmit FIFO. When being read, it reads out the top byte of the RX FIFO; when being written, it writes into the top byte of the TX FIFO.

EUART2 can be configured to add LIN capability. The major enhancement of LIN includes master/slave configurations, auto baud-rate synchronization, and frame based protocol with header. Under LIN extension mode, all EUART2 registers and functions are still effective and operational. LIN is a single-wire bus and it requires external components to combine RX and TX signals externally. LIN is frame based and consists of message protocols with

master/slave configurations. The following diagram shows the basic composition of a header message sent by the master. It starts with BREAK, the SYNC byte, ID bytes, DATA bytes, and CRC bytes.

A LIN frame structure is shown and the frame time matches the number of bits sent and has a fixed timing.

LIN bus protocol is based on frame. Each frame is partitioned into several parts as shown above. For master to initiate a frame, the software follows the following procedure.

Initiate a SBK command. (SW needs to check if the bus is in idle state, and there is no pending transmit data). Write "55" into TFIFO.

Write "PID" into TFIFO.

Wait for SBK to complete interrupts and then write the following transmit data if applicable. (This is optional). The following diagram shows Finite State Machine (FSM) of the LIN extension and is followed by registers within EUART2.

LINCTRL (0xA090) LIN Status/Control Register R/W 0x00

TUMKO A Division of **IKE** $MASUI$ is meaningful only if $ASII=0$ $MASII=1$ will enable the auto sync update on the next

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LINTMRH (0xA091) LIN Timer Register High R/W 0xFF

LINTMRL (0xA092) LIN Time Register Low R/W 0xFF

LCNTR[15-0] is read only and is an internal 16-bit counter clocked by the baud rate clock. LINTMR[15-0] is write only and is the timer limit for LCNTR[15-0]. If MASEN=1 as LIN master mode, this timer is used to generate Frame time base. The internal counter LCNTR[15-0] is cleared whenever a "SEND BREAK" command is executed, and when the counter reaches LINTMR [15-0] (LCNTR[15-0] >= LINTMR[15-0]), a LCNTRO interrupt is generated. Thus the software can write a Frame Time value into LINTMR and use interrupts to initiate frames. If MASEN=0 as LIN slave mode, this timer is used for determining the accumulated bus idle time. The internal counter is cleared whenever a RX transition occurs. When the internal counter reaches LINTMR[15-0], an LCNTRO interrupt is generated. The software can use this interrupt to enter sleep mode by writing the required bus idling time into LINTMR[15-0].

LINSBRH (0xA093) EUART/LIN Baud Rate Register High byte R/W 0x00

LINSBRL (0xA094) EUART/LIN Baud Rate Register Low byte R/W 0x00

When a slave receives a BREAK followed by a valid SYNC field, an RSI interrupt is generated and the acquired baud rate from SYNC field is stored in SBR[15-0]. The acquired baud rate is BAUD RATE = SYSCLK/SBR[15-0]. The software can just update this acquired value into BR[15-0] to achieve synchronization with the master. If Auto-Sync Update (ASU) register bit is enabled under LIN slave mode, LIN controller will automatically perform the update of BR[15-0] with SBR[15-0] and issue another ASUI interrupt when received a valid SYNC field.

LININTF (0xA095) LIN Interrupt Flag Register R/W 0x00

LSTAT LIN Bus Status bit (1: Recessive / 0: Dominant), Read only.

LSTAT $=$ 1 indicates that the LIN bus (RX pin) is in recessive state.

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LININTEN (0xA096) LIN Interrupt Enable Register R/W 0x00

LCNTRIE LIN Counter Overflow Interrupt Enable (1: Enable / 0: Disable)

6. Serial Peripheral Interface (SPI)

The Serial Peripheral Interface (SPI) is an enhanced synchronous serial hardware which is compatible with Motorola's SPI specifications. The SPI Controller includes 4-bytes FIFO for both transmit and receive. SPI Interface uses Master-Out-Slave-In (MOSI), Master-In-Slave-Out (MISO), Serial Clock (SCK) and Slave Select (SSN) for interface. SSN is low active and only meaningful in slave mode.

SPICR (0xA1) SPI Configuration Register R/W (0b001000xx)

SPIMR (0xA2) SPI Mode Control Register R/W (0x00)

SPIST (0xA3) SPI Status Register R/W (0x00)

SPIDATA (0xA4) SPI Data Register R/W (0xXX)

6.1 SPI Master Timing Illustration

6.1.1 CPOL=0 CPHA=0

6.1.2 CPOL=0 CPHA=1

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6.1.3 CPOL=1 CP

6.1.3 CPOL=1 CPHA=0

6.1.4 CPOL=1 CPHA=1

6.2 SPI Slave Timing Illustration

6.2.1 CPOL=0 CPHA=0

6.2.2 CPOL=0 CPHA=1

7. Programmable Counter Array (PCA) and Compare/Capture/PWM (CCP)

The PCA provides enhanced timing functions with less CPU intervention than the standard 8051 timers T0, T1, and T2. The PCA is partitioned in three parts. The main PCA Counter consists of CH and CL. There are 6 channels of Compare/Capture/PWM modules.

The MAIN COUNTER (CH and CL) is configured and controlled by two registers, CMOD and PCACON. The counter value is accessed by CH and CL registers. The counter can be configured as either FREE-RUN or AUTO-RELOADED mode. The counter values of CH and CL can be captured in CHSR and CLSR triggered by software or hardware. There is also a counter compare register CHIR and CHLR. An interrupt can be enabled at CH:CL == CHIR:CLIR. This allows the PCA to easily synchronize with the software control. CHIR and CLIR are double-buffered.

PCACPS (0xA0A5) PCA Counter Clock Scaling Register R/W (0x00)

PCACPS sets the clock input to the PCA at SYSCLK/(PCACS[7-0]+1).

PCAMOD (0xD1) PCA Mode Control Register R/W (0x00)

PCACON (0xE1) PCA Configuration Register R/W (0x00)

CF is set to 1 by hardware when overflow condition occurs. The overflow condition occurs at either of 0xFFFF to 0x0000 (OVF8EN=0) or 0xXXFF to 0xXX00 (OVF8EN=1). This bit must be cleared by software. CIF Count Compare Flag bit CIF is set by hardware when $CH:CL = CHIR:CLIR$. This bit must be cleared by software. CCF5 - CCF0 Module Interrupt Flag 5-0 This is set by hardware as its corresponding module generates an interrupt. These bits must be cleared by software.

CH (0xE9) PCA Main Counter High Byte R/W (0x00)

CH holds the upper 8-bit of the main counter value.

CL (0xD9) PCA Main Counter Low Byte R/W (0x00)

CL holds the lower 8-bit of the main counter value. Reading CL triggers a snapshot action to copy CH:CL to CHSR:CLSR.

CHRLD (0xA0A7) PCA Counter CH Reload Value Register R/W (0x00)

This register holds the reload value for CH in AUTO-RELOAD mode.

CLRLD (0xA0A6) PCA Counter CL Reload Value Register R/W (0x00)

This register holds the reload value for CL in AUTO-RELOAD mode.

CHSR (0xF3) PCA Snapshot Register of CH RW (0x00)

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CHIR[7-0] CH Counter Compare Interrupt . The compare value is double- buffered.

CLSR (0xF2) PCA Snapshot Register of CL RO (0x00)

CLSR[7-0] CL Snapshot Register. It is read-only.
CLIR[7-0] CL Counter Compare Interrupt The co

CL Counter Compare Interrupt The compare value is double-buffered.

CCAPMn CCP Module Configuration Register (0xB2, 0xB3, 0xB4, 0xB5, 0xB6, 0xB7) R/W (0x00)

The Compare/Capture modules receive the 16-bit count value from the main counter as the time base. Each module is configured by its mode register CCAPMn and contains two 8-bit registers used for comparing value holder or capturing value in storage. There are several basic modes of operation for CCP modules and each CCP module can be configured in the same or different modes.

CCAPnL register holds the compare value or capture value. It is used as PWM value register.

CCAPnH CCP Compare Value High Register (0xD3, 0xD5, 0xD7, 0xE3, 0xE5, 0xE7) R/W (0x00)

CCAPnH register holds the compare value or capture value. It is used as PWM value register.

7.1 16-Bit Capture Mode

The capture mode is used to measure the elapse time of an external event between the edges of the enabled external signal when either CAPP or CAPN is set. The external CEX is sampled for transition. When a valid capture edge occurs in CEX, the current CH/CL count value is loaded into CCAPnH and CCAPnL registers. At the same time CCFn in PCACON register is set, and interrupt is generated if enabled. The block diagram of the configuration is shown below:

7.2 16-Bit COMPARE TIMER MODE

The COMPARE TIMER mode can be used as a software timer or to generate a PWM output. This mode is enabled when ECOM is set and CAPP CAPN are set to low. To allow the compare result to be used, MAT/ECCF also needs to be set. The CCAPnH and CCAPnL hold the 16-bit Timer value and are compared against the incrementing value CH and CL from the main counter. The compare value is double-buffered and is updated when the main counter overflows. This prevents any unexpected comparator output during updating a new value to CCAPnH and CCAPnL. When a match occurs, CCF is set and an interrupt is generated. The block diagram of this mode is shown as following.

The match result can also be used to generate CEX output change. Depending on the CCAPM's OF and TOG setting, CEX output is changed at the compare-match instant. However, the triggering of the change of CEX does not require MAT qualifier. Using CEX, waveform of precision duty cycle waveform or frequency modulation can be generated. The effect of OF/TOG on CEX is described in CCAPM register. To avoid unwanted glitches or a match condition when updating the CCAPnH and CCAPnL registers, when ECOM is set and the writing to CCAPnL causes ECOM to clear. Writing to CCAPnH sets ECOM to start the comparator. Therefore user program should update CCAPnL first and then CCAPnH. Of course, ECOM bit can be controlled directly through CCAPMn register.

7.3 8-Bit Pulse Width Modulator Mode

This mode is used to generate 8-bit precision PWM output on CEX. The time base of the PWM is provided by CL of the main counter. CCAPnL is used for compare value. When $CL \le CCAPnL$, the output is 0 and when $CL >$ CCAPnL, the output is 1. The compare value is double-buffered and is updated when CL overflows from FF to 00. The PWM mode is enabled when ECOM and PWM bits are both set, and CAPP, CAPN are both low. Note that under the above compare method, the maximum CEX duty cycle is 255/256. If TOG is set to 1 in this mode, CEX is forced high to provide 256/256 with full high duty cycle. If ECCF bit is set, then when CCAPnL=CL (i.e. the output change), CCF is also set to 1 by hardware and triggers a PCA interrupt. The following block diagram shows the PWM mode operation.

7.4 16-Bit Pulse Width Modulator Mode

This mode is similar to the 8-bit PWM mode except it uses the 16-bit CH:CL count value for the time base of the PWM. The compare value is composed of CCAPnH:CCAPnL and is double-buffered. When CH:CL ≦ CCAPnHL:CCAPnL, the output is 0; when CH:CL > CCAPnHL:CCAPnL, the output is 1. The output can be forced to 1 by setting TOG=1. The PWM mode is enabled when both ECOM and PWM bits and CAPP are set while CAPN is low. An interrupt is enabled by ECCF and triggered when CH:CL==CCAPnH:CCAnL.

7.5 8-BIT Windowed Pulse Width Modulator (WPWM) Mode

This mode is used to generate 8-bit PWM output on CEX. The difference from regular PWM mode is that the CEX becomes high during a window of CL count. CEX becomes high when CL is greater than CCAPnL, CEX is reset to low when CL is greater than CCAPnH. The compare values are double-buffered. Therefore the value in CCAPnH must be larger than CCAPnL to prevent abnormal operations. The output of CEX can be inverted by setting TOG to 1. An interrupt can be enabled by ECCF, if MAT=0, then CL=CCAPnL generates an interrupt. Setting MAT to 1 and CL=CCAPnH also generates an interrupt.

7.6 CCP Function Summary

ECOM cannot be set to 1 by hardware (when writing to CCAPnH) if all bits (OF, ECOM, CAPP, CAPN, MAT, TOG, PWM) in CCAPM are set to 0 (NO OPERATION mode).

In 16-bit compare mode, ECOM can be set to 1 by hardware (when writing to CCAnPH) or software, and can be cleared to 0 by hardware (when writing to CCAnPL) or software. When ECOM is cleared to 0 in this mode, the CCP function enters NO OPERATION mode. The compare value is CCAPnH:CCAPnL and is double-buffered.

8. PWM0/PWM1 Controller

PWM0/1 controller provides an 8-bit programmable duty cycle output. The counting clock of PWM0/1 is programmable and the base frequency of the PWM0/1 is just the counting clock divided by 256. The duty cycle setting is always double buffered. PWM0 output is connected through P04, P06, P12, P17 or P23, PWM1 output is connected through P05, P13, P15, P16 or P22.

PWM0CFG (0xA088) PWM0 Clock Scaling Setting Register R/W (0x00)

The counting clock is SYSCLK/CS[7-5]/(CS[4-0]+1) and one cycle has 256 counting clocks. Assuming SYSCLK is 16MHz, and we want the PWM base frequency of 120Hz. First we get 16MHz/120Hz/256 = 520.8. Where 520.8 needs to be separated as two multiplicands of CS[7-5] and CS[4-0]. Then by trial and error we can select CS[7-5] = 100, and CS[4-0] = 0x0F. This gives $16MHz/256/32/(15+1) = 122Hz$.

PWM0DTY (0xA089) PWM0 Duty Register R/W (0x00)

PWM0DTY registers define the PWM0 duty cycle. The maximum duty cycle is 255/256. PWM0DTY is always double buffered and is loaded to duty cycle comparator when the current counting cycle is completed.

PWM1CFG (0xA08A) PWM1 Clock Scaling Setting Register R/W (0x00)

PWM1DTY (0xA089) PWM1 Duty Register R/W (0x00)

PWM1DTY registers define the PWM1 duty cycle. The maximum duty cycle is 255/256. PWM1DTY is always double buffered and is loaded to duty cycle comparator when the current counting cycle is completed.

PWMFG (0xA08C) PWM0/1 Interrupt Enable and Flag Register R/W (0x00)

9. Buzzer controller

This buzzer controller is a simple sound generator. The waveform generated is shown in the following. BZPRD is determined from the buzzer frequency BZFRQ. The basic element of the waveform is a FRAME where each FRAME is a combination of a number of pulses defined by FMPCNT and a number of pulse duration of silence defined by FMSCNT. This combination allows simple two-tone generations, one defined by BZFRQ, and one subharmonic defined by BZFRQ/(FMPCNT+FMSCNT). The SESSION defines the number of FRAMES, which essentially determines the duration of the sound. The output can be configured as software start, which triggers a session output when START is set, or as continuous if CONT is set. In either configuration, session end interrupt can be enabled to inform software for intervention.

BZFRQ (0xA128) Buzzer frequency controlled R/W (0x10)

BZFRQ[7-0] Buzzer Frequency Setting

The buzzer frequency is defined by BZCLK/BZFRQ[8-0], BZFRQ[8] is located at BZCFG[4].

BZFMCFG (0xA129) Buzzer Frame Configuration Register R/W (0xF0)

FMPCNT[3-0] Pulse Count for Frame On

This number of pulse during Frame on duration is FMPCNT[3-0]

FMSCNT[3-0] Pulse Count for Frame Silence

This number of pulse during Frame silence duration is FMPCNT[3-0]

FMPCNT[3-0] + FMSCNT[3-0] should be greater than 1 for normal operations.

BZSNFMCNT (0xA12A) Buzzer Session Frame Count Register R/W (0x80)

SNFMCNT[7-0] Frame Repeat Number of a Session

SNFMCNT[7-0] defines the number of frame in a session.

BZCFG (0xA12B) Buzzer Configure Register R/W (0x00)

BZEN Buzzer Control Enable

BZEN=1 enables the buzzer controller

BZEN=0 disables the buzzer controller

BZSNMUTE (0xA12C) Buzzer Session MUTE Frame Register R/W (0xff)

BZSNMUTE [7-0] start mute Frame number of a Session

Mute buzzer when frame count greater than BZSNMUTE in a session.

10. Touch Key Controller

The Touch Key Controller utilizes capacitance based touch key sensing. The touch is detected by detecting the change of an external sensing capacitance. When a key is touched, the capacitance of the key goes up. The capacitance detection is based on a CMOS relaxation oscillator as shown, which uses the key capacitance as the oscillation parameters. The Touch Key Controller in operation typically consume at 25uA (when slow comparator is selected) and at 250uA (when fast comparator is selected). The inputs of the Touch Key Controller are multiplexed with multi-function GPIO's ANIO pins. To use these ports as Touch Key inputs, the ANEN must be enabled and other drivers to be in high-impedance state.

The touch detection algorithm is done in software and is assisted with the following hardware. The hardware consists of one 24-bit Timer and a 24-bit Counter. The clock into the timer and the counter can be interchanged by MODE selection.

When MODE=0 (MODE0), the counter uses internal clock and timer uses sensing clock. This typically is used when the capacitance oscillation is slower. When MODE=1 (MODE1), the clock relationship is reversed and this is typically used when sensing oscillation is faster. Typical waveforms are shown in the following timing diagram.

The START command clears the counter and load the timer, and then starts the timer and the counter. When the timer (TKCTM) expires, the counter value (TKCVA) is captured and an interrupt is triggered. Two hardware moving average filters, one with slow time constant (TKMAS) and one with fast time constant (TKMAF), are also included in the Touch Key Controller. The software should write the previous moving average into TKMAF and TKMAS before issuing START. When a capture is completed, the moving average filters are also updated. The software can read out TKCVA as well as TKMAF and TKMAS for algorithm use.

The Touch Key Controller also allows auto detection. The auto detection is primarily used fore wake up purpose. The TKCTM and TKCVA functions the same as the normal mode, and the capture cycles continues without interrupt. However, in AUTO mode, TKCTM and TKCVA, as well as TKMAF and TKMAS are limited in 16-bit width. The hardware monitors the TKMAF and TKMAS, and when the fast moving-average is greater than the slow movingaverage by a predefined threshold, a wake-up interrupt is triggered. The following diagram illustrates the progressive count curves.

TKCCFG (0xA130h) Touch Key Controller Configuration Register R/W (0x00)

MODE=0 selects Capacitor Clock as timer and internal clock as counter MODE=1 selects internal clock as timer and capacitor clock as counter

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TKCAFG (0xA131h) Touch Key Controller Analog Configuration Register R/W (0x00)

TKCFFG (0xA132h) Touch Key Controller Filter Configuration Register R/W (0x00)

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TKCMVF0 (0xA148h) Touch Key Fast Moving Average Register 0 RO (0x00)

TKCMVF1 (0xA149h) Touch Key Fast Moving Average Register 1 RO (0x00)

TKCMVF2 (0xA14Ah) Touch Key Fast Moving Average Register 2 RO (0x00)

TKCMVS0 (0xA14Ch) Touch Key Slow Moving Average Register 0 RO (0x00)

TKCMAS1 (0xA14Dh) Touch Key Slow Moving Average Register 1 RO (0x00)

TKCMAS2 (0xA14Eh) Touch Key Slow Moving Average Register 2 RO (0x00)

11. 12-Bit SAR ADC (ADC)

The on-chip ADC is a 12-bit SAR based ADC with maximum ADC clock rate of 4MHz (2.5V – 5V) or 500KHz (1.8V – 2.4V) and each conversion takes 64 cycles to complete. The conversion is partitioned into 4-sub conversion (each 16 cycles) for cancellation of internal offsets. Typical ADC accuracy is about 10-Bit to 11-Bit at 5V reference. The ADC clock is programmable and set by the ADC clock scaler. The ADC full range reference can be selected using VDD or VDDC (1.8V). The full range can be calibrated using on-chip 1.1V reference as the input. When enabled, the ADC consumes about 2mA of current. The ADC also includes hardware to perform result average. Average can be set to 1 to 8 times.

In IS31CS8969, The inputs of the ADC are multiplexed with multi-function GPIO's ANIO1 pins. To use these ports as ADC inputs, the ANEN1 must be enabled and other drivers to be in high-impedance state. The block diagram of ADC is shown in the following.

To enhance the resolution of ADC, a Programmable Gain Amplifier (PGA) can be inserted in the input path. The gain setting can be up to 20. The PGA is an inverting type using OPAMP, and its block diagram is shown as following. R1 is approximately 20K Ohm. VX is a voltage reference of VDDH-0.4V or VDDC-0.4V. GND controls the PGA input to 0V. CHOP is used to control OPAMP offset. For precision measurement, software should always first set GND=0 and obtain an ADC result (ADCR1). Then a second result is obtained by connecting to PGAIN (ADCR2). The measurement is obtained by the difference between ADCR1 and ADCR2. In this procedure, offset of the PGA as well as VX is canceled automatically. Because PGA input is connected to the GPIO ANIO bus, when switching the IOCELL's ANIO1, the ADC must also wait after PGA to settle. PGA settling time is approximately 10usec.

ADCCTL (0xCEh) ADC Control Register R/W 00000000

AVG[2-0] AVG[2-0] controls the hardware averaging logic of ADC readout. It is recommended the setting is changed only when ADC is stopped. If multiple channels are enabled, then each channel is averaged in sequence. The default is 000.

CHSEL[2-0] ADC Channel Select

ADCIF ADC Conversion Completion Interrupt Flag bit

ADCIF is set by hardware when the conversion is completed and new result is written to ADCL and ADCH result registers. If ADC interrupt is enabled, this also generates an interrupt. This bit is cleared when ADCL is read. When this flag is set, no new conversion result is updated.

CSTART Software Start Conversion bit Set this CSTART=1 to trigger an ADC conversion on selected channels. This bit is selfcleared when the conversion is done.

ADCPGA (0xB9h) ADC PGA Control Register WO 000XX000

ADCH and ADCL are the high and low byte result registers respectively, and are read-only. Reading low byte result also clears its corresponding interrupt flag. If the flag is not cleared, no new result is updated. The software should always read the low byte last. The format of the high byte and low byte depends on ADCFM setting.

If ADCFM = 1, the valid ADC Result is located on ADCAH[7:0] and ADCAL[3:0]. If ADCFM = 0, the valid ADC Result is located on ADCAH[3:0] and ADCAL[7:0].

ADCAL (0xBAh) Channel A Result Register Low Byte RO XXXXXXXX

ADCAH (0xBBh) Channel A Result Register High Byte RO XXXXXXXX

11.1 On-Chip Temperature Sensor

The on-chip temperature sensor is derived from a positive temperature coefficient current from the internal Bandgap circuit. The hardware block diagram is shown in the following:

The current is applied to an on-chip low temperature coefficient resistor and develops to a voltage that is in terms proportional to the temperature. The voltage can be written in as an equation of $V = K^*$ Temperature, where K is a constant. The following graph shows the temperature sensor curve with the ADC Result. The vertical axis is the ADC Result in decimal and the horizontal axis is the chip temperature in ℃.

In IS31CS8969, the temperature coefficient (TC) is a positive constant. This calibration value can be found in the Flash Information Block (IFB) address 0x2D. The byte contains the calibrated temperature coefficient for 1℃ in LSB of ADC. This is a binary number in 4:4 formats, i.e. the radix point is between bit 4 and bit 3. For example, 0b0111.1100 refers to 7.75 LSB. If calibration of temperature sensor is not done, then this byte would be written as 0x00. In IFB 0x2B and 0x2C, these two bytes contain ADC measurement of internal temperature sensor (Aoffset) at

calibration temperature (Toffset). The ADC should use VDDC (1.8V) as the full-scale reference. IFB 0x2B is LSB and 0x2C is MSB. The upper 4-bit of the MSB is the offset of the calibration temperature from 20℃. For example, if the

calibration is done at 25℃, then the upper 4-bit should be 0x05. If calibration of temperature sensor is not done, then these two bytes would be written as 0x00.

The detection temperature can be represented in the following equation. Where the Adec is the ADC measurement result of ADC at the temperature Tdec.

Tdec (℃) = (Adec – Aoffset) / TC + 20 + Toffset

For example, in IFB TC = 7.75, Toffset = 5, and Aoffset = 2250, if the ADC measurement result Adec = 2870, then the measurement temperature $Tdec = (2870-2250) / 7.75 + 20 + 5 = 105$ °C.

12. 10-Bit Voltage Output DAC (VDAC)

A 10-bit voltage output DAC is included. The DAC is composed of LSB 7-bit R2R and MSB 3-bit linear DAC. The output is buffered by unity configured OPAMP. The output range of the DAC is from 0V to VDD. Due to the circuit structure of the OPAMP, the output accuracy will suffer some loss from ¾ VDD to VDD. The output impedance of the buffer is less than 1K Ohm and should not drive high capacitance load. Please also note that the linearity and accuracy of the DAC will suffer when the output is close to rail or 0V because of the OPAMP. The update of DAC must start with low-byte first and then high-byte because the low-byte is double-buffered. The DAC output is connected to P1.1 IOCELL's ANIO1. To use this port as DAC output, P1.1 IOCELL's ANEN1 must be enabled and other drivers to be in high-impedance state.

DACH (0xA037h) DAC High Register R/W 0x00

DAC[9-8] DAC[9-8] Data bits

These two bits are MSB of 10-bit DAC data. Writing to this register updates the DAC output.

DACL (0xA036h) DAC Low Register R/W 0x00

This register is double-buffered and the output is not updated until DACH is written.

13. Analog Comparators (ACMP)

IS31CS8969 has four analog comparators as its on-chip external peripherals. When enabled, each comparator consumes about 250uA. The input signal range is from 0 to VDD. There are two 8-bit R-2R DAC associated with the comparators to generate the compare threshold. The R-2R DAC uses the internal 1.8V supply as the full-scale range thus limiting the comparator threshold from 0V to 1.8V in 256 steps. Channel B/C/D can select a common external threshold. The inputs of the comparators are multiplexed with multi-function GPIO's ANIO1 pins. To use these ports as comparator inputs, the ANEN1 must be enabled and other drivers to be in high-impedance state. P1.4 is used for CMPA and CMPD thus can be configured to detect two thresholds simultaneously. P1.5 and P1.6 are used for CMPB and CMPC. P1.7 is used for comparator external threshold.

The CPU can read the real-time outputs of the comparator directly through register access. The output is also sent to an edge-detector and any edge transition can be used to trigger an interrupt. The stabilization time from off state to enabled state of the comparator block is about 20usec. The block diagram of the analog comparator is shown in the following diagram.

CMPCFGAB (0xA030h) Analog Comparator A/B Configuration Register R/W 0x00

CMPCFGCD (0xA031h) Analog Comparator C/D Configuration Register R/W 0X00

CMPVTH0 (0xA032h) Analog Comparator Threshold Control Register R/W 0X00

CMPVTH0 register controls the comparator threshold VTH0 through 8-bit DAC. When set to 0x00h, the threshold is 0V. When set to 0xFFh, the threshold is at 1.8V. When not used, it should be set to 0x00 to save power consumption.

CMPVTH1 (0xA033h) Analog Comparator Threshold Control Register R/W 0X00

CMPVTH1 register controls the comparator threshold VTH1 through 8-bit DAC. When set to 0x00h, the threshold is 0V. When set to 0xFFh, the threshold is at 1.8V. When not used, it should be set to 0x00 to save power consumption.

CMPST (0x94h) Analog Comparator Status Register R/W 0X00

IS31CS8969 14. GPIO Port Function and Pin Configurations

This section describes the pin functions and configurations. Almost all signal pins are multi-functional with default setting as a GPIO port pin. Therefore each signal pin requires two registers to configure the I/O capability and the function selection. The following describes the control and contents of these registers and the register names and pin names are referenced by their default GPIO port name; IS31CS8969 employs a configurable I/O buffer design. The standardized I/O design allows flexible configuration of the digital I/O function such as open-drain, open-source, pull-up, pull-down, bus-holder capabilities. In addition to digital I/O function, the standardized I/O also provides analog I/O capability that can be selected when the GPIO pin is shared with analog peripheral purposes such as analog OPAMP, ADC input or DAC output.

14.1 IO Cell Configurations

The supply voltage of the I/O buffer uses VDD (2.5V to 5.5V). The input and output level is referenced to VDD and 0V. Since the design is standardized, the I/O design offers a uniform ESD performance. The functional block diagram of the standard I/O buffer is shown in the following diagram.

IOCELL2A is IOCELL with two analog multiplexer switches, ANIO and ANIO1. The control of IOCELL2A uses IOCFG's ANEN1 to replace LATEN.

From the diagram, there are 7 control bits for the IOCFGPx.y register, and these registers are located at XFR 0xA040 – 0xA047 for P0.0 to P0.7, 0xA048 – 0xA04F for P1.0 to P1.7, 0xA060 – 0xA063 for P2.0 to P2.3, and the definitions of IOCFGPx.y are described in the following table.

		6	5	4	3	2		0		
RD	INEN	ANEN1	PUEN	PDEN	ANEN	PDRVEN	NDRVEN	IPOL		
WR	INEN	ANEN1	PUEN	PDEN	ANEN	PDRVEN	NDRVEN	IPOL		
Input buffer control. Set this bit to enable the GPIO input buffer. If the input buffer is not INEN used, it should be disabled to prevent leakage current when pin is floating. DISABLE is the default value.										
	Analog MUX 1 enables control. Set this bit to connect the pin to the internal analog ANEN1 peripheral. DISABLE is the default value.									
	PUEN	Pull up resistor enable control. Set this bit to enable pull-up resistor connection to the pin. The pull-up resistor is approximately 100K Ohm. DISABLE is the default value.								
PDEN		Pull down resistor enable control. Set this bit to enable pull-down resistor connection to the pin. The pull-down resistor is approximately 100K Ohm. DISABLE is the default value.								
Analog MUX enable control. Set this bit to connect the pin to the internal analog peripheral. ANEN DISABLE is the default value.										

IOCFGP (0xA040h – 0xA04Fh, 0xA060 – 0xA06F) IO Buffer Configuration Registers R/W (0x00)

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PDRVEN Output PMOS driver enable. Set this bit to enable the PMOS of the output driver. DISABLE is the default value. NDRVEN Output NMOS driver enable. Set this bit to enable the NMOS of the output driver. DISABLE

is the default value. IPOL Input logic polarity. IPOL=0 for normal polarity and IPOL=1 for reverse polarity.

The following table shows various configurations of the I/O buffer.

Please note the following exceptions exist for IOCFG registers.

14.2 GPIO Port Multifunction

Because each signal pin is a multi-functional and the function is shared with GPIO port, therefore each pin requires MFCFGP register to control, which function is in effect and which peripherals are connected to the signal pins. These selection and definitions are pin specific and product specific. The following description describes the selection and control for IS31CS8969 signal pins.

	7	6	5	4	3	\mathcal{P}		0		
RD	KEY0EN	MSDAEN	SSDA2EN	CEX0EN	MISOEN	PINT ₁ EN	PINTOEN	GPIOEN		
WR	KEY0EN	MSDAEN	SSDA2EN	CEX0EN	MISOEN	PINT ₁ EN	PINTOEN	GPIOEN		
	KEY0EN MSDAEN SSDA2EN CEX0EN MISOEN PINT1EN	Touch Key 0 Enable bit. IOCFGP0.0 ANEN must be enabled too. MSDAEN=1 enables this pin as I2CM SDA I/O. This must be configured as OD output. SSDA2EN=1 enables this pin as I2CS2 SDA I/O. This must be configured as OD output. CEX0EN=1 enable this pin as CEX I/O for CCP0. MISOEN=1 uses this pin as SPI MISO I/O. Pin Interrupt Enable Control Bit.								
	PINTOEN GPIOEN	PINT1EN=1 configures this pin as an input condition to PINT1 interrupt. More than one pin can be assigned to PINT1. And one pin can be assigned to both PINT0 both PINT1. Pin Interrupt Enable Control Bit. PINTOEN=1 configures this pin as an input condition to PINTO interrupt. More than one pin can be assigned to PINT0. And one pin can be assigned to both PINT0 both PINT1. GPIO Function Enable Bit. Set this bit to enable GPIO function. DISABLE is the default value.								

MFCFGP0.0 (0xA050) GPIO P0.0 Function Configuration Register R/W (0x00)

MFCFGP0.1 (0xA051h) GPIO P0.1 Function Configuration Register R/W (0x00)

MFCFGP0.2 (0xA052h) GPIO P0.2 Function Configuration Register R/W (0x00)

MFCFGP0.3 (0xA053h) GPIO P0.3 Function Configuration Register R/W (0x00)

MFCFGP0.4 (0xA054h) GPIO P0.4 Function Configuration Register R/W (0x00)

configures this pin as an input condition to PINT0 interrupt. More than one pin can be assigned to PINT0. And one pin can be assigned to both PINT0 both PINT1. GPIOEN GPIO Function Enable Bit. Set this bit to enable GPIO function. DISABLE is the default value.

MFCFGP0.5 (0xA055h) GPIO P0.5 Function Configuration Register R/W (0x00)

MFCFGP0.6 (0xA056h) GPIO P0.6 Function Configuration Register R/W (0x00)

MFCFGP0.7 (0xA057h) GPIO P0.7 Function Configuration Register R/W (0x00)

PINT0EN=1 configures this pin as an input condition to PINT0 interrupt. More than one pin can be assigned to PINT0. And one pin can be assigned to both PINT0 both PINT1. GPIOEN GPIO Function Enable Bit. Set this bit to enable GPIO function. DISABLE is the default value.

MFCFGP1.0 (0xA058h) GPIO P1.0 Function Configuration Register R/W (0x00)

MFCFGP1.1 (0xA059h) GPIO P1.1 Function Configuration Register R/W (0x00)

MFCFGP1.2 (0xA05Ah) GPIO P1.2 Function Configuration Register R/W (0x00)

GPIOEN GPIO Function Enable Bit. Set this bit to enable GPIO function. DISABLE is the default value.

MFCFGP1.3 (0xA05Bh) GPIO P1.3 Function Configuration Register R/W (0x00)

MFCFGP1.4 (0xA05Ch) GPIO P1.4 Function Configuration Register R/W (0x00)

MFCFGP1.5 (0xA05Dh) GPIO P1.5 Function Configuration Register R/W (0x00)

GPIOEN GPIO Function Enable Bit. Set this bit to enable GPIO function. DISABLE is the default value.

MFCFGP1.6 (0xA05Eh) GPIO P1.6 Function Configuration Register R/W (0x00)

MFCFGP1.7 (0xA05Fh) GPIO P1.7 Function Configuration Register R/W (0x00)

MFCFGP2.0 (0xA070h) GPIO P2.0 Function Configuration Register R/W (0x00)

MFCFGP2.1 (0xA071h) GPIO P2.1 Function Configuration Register R/W (0x00)

MFCFGP2.2 (0xA072h) GPIO P2.2 Function Configuration Register R/W (0x00)

MFCFGP2.3 (0xA073h) GPIO P2.3 Function Configuration Register R/W (0x00)

14.3 GPIO Edge Interrupt

The GPIO pins can be configured as external pin interrupt input or for wake up purpose. Each port has edge detection logic and latch for rising and falling edge detections.

PIOEDGR0 (0xA028h) Port0 IO Input Rising Edge Detection Register R/W (0x00)

PREN0[i]=1 enables the rising edge detection.

PIOEDGF0 (0xA038h) Port0 IO Input Falling Edge Detection Register R/W (0x00)

PIOEDGR1 (0xA029h) Port1 IO Input Rising Edge Detection Register R/W (0x00)

PIOEDGF1 (0xA039h) Port1 IO Input Falling Edge Detection Register R/W (0x00)

14.4 GPIO Noise filtered Wake Up

The CPU in IS31CS8969 can enter sleep mode to save standby power consumption. An external pin status change can trigger the wake up of CPU. Each port pin in GPIO Port 0 and Port 1 can be configured as GPIO Edge interrupts as described in 14.3. Any enabled edge interrupt can also serve as wake up event, i.e. the event performs wake up of CPU and at the same time triggers an interrupt.

In additional to the edge interrupt wake-up mechanism, a separate GPIO wake up scheme is included which provide noise filter on the triggering signals. The triggering conditions can be defined as either or both of the input rising and falling edges. A noise filter is included following the edge detection as shown in the following diagram. Any noise pulses less than FLTTM are filtered out. The detection sources are GPIO port P0.0 to P0.7 and P1.0 to P1.7, and only one of this can be selected at one time.

GPWKCFG (0xA02Eh) GPIO Wakeup Configuration Register R/W (0x00)

15. IFB Block and Writer Mode and Boot Code/ISP

15.1 IFB Block

The main flash memory is 32Kx8 and also contains a separate 256B Information Block (IFB). The IFB is partitioned into two parts. 00 to 4F range contains critical manufacture and calibration information. And 50 to FF range contain user data, which can be programmed one time. The IFB cannot be erased but programmable through Flash Controller Command but it can be erased and written through Writer mode. The user data portion can only serve as One-Time-Programmable storage by the user. The following table shows the IFB contents.

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Note 1: M data cannot be modified and can only be written in writer mode when the entire Flash is erased. Note 2: U data reads out as FF after the Flash is erased. It can only be programmed once after the Flash is mass erased.

****** The erasure of IFB or modifications of manufacture information in IFB void any manufacture warranty.

**** This table is for reference only. Please refer to most updated AP note and boot code documents.

15.2 Test Mode and Writer Mode Entry from RSTN

RSTN pin serve as external reset input. With proper input waveform, RSTN is also used to enter test and WRITER mode. The waveform starts with RSTN=0, and uses duty cycle to encode bit stream. Each bit time is divided into 16-T and is encoded into 4 valid states as shown in the following.

The entry is allowed in the following exact sequence. And if RSTN=1 for over 2 msec, then the test mode is forced to exit.

The sequence is initiated for $RSTN=0 > 4$ msec (BREAK field), then followed by three SYNC bit, and two bytes sequence and a T-S bit. RSTN then must return to 0 to maintain the test mode. If the sequence is not exactly matched, a BREAK field must be inserted to start over the sequence. The hardware block diagram is shown in the following and key parameters are also illustrated.

15.3 Writer Mode

Writer Mode (WM) is used by the manufacturer or by users to program the flash (including IFB) through a dedicated hardware (Writer or Gang Writer). Under this set up, only WM related pins should be connected and all other unused pins left floating. Writer mode follows a proprietary protocol and is not released to general users. Users must obtain it through a formal written request to the manufacturer and must sign a strict Non-Disclosure-Agreement.

The Writer Mode provides the following commands.

ERASE Main Memory ERASE Main Memory and IFB READ AND VERIFY Main Memory (8-Byte) WRITE BYTE Main Memory READ BYTE IFB WRITE BYTE IFB Fast Continuous WRITE Fast Continuous READ

The writer mode is protected against code piracy. The power-on state of the device deactivates the writer mode. Only ERASEMM and ERASEMMIFB, and READVERIFYMM commands can be executed. It is activated by READVERIFY the range of 0xFFF8~0xFFFF where a security key can be placed by the user program. The probability of guessing the key is 1 in 2^64 = 1.8E19. Since each trial of READVERIFY takes 10usec, it takes about 6E6 years to exhaust the combinations. If the key is unknown, a user can choose to issue the ERASEMM command then fully erase the entire contents (including the key). Once fully erased, all data in the flash is 0xFF, and it can be successfully unlocked by READVERIFYMM with 8-bytes of 0xFF.

The users must not erase the information in IFB. And the user should not modify the manufacturer data. Any violation of this results in the void of manufacturer warranty.

15.3.1 Writer Mode Pins

15.4 Boot Code and In-System-Programming

After production testing of the packaged devices, the manufacture writes the manufacturer information and calibration data in the IFB. At the last stage, it writes a fixed boot-code in the main memory residing from 0x07000 to 0x77FF. The boot code is executed after resets. The boot code first reads 0x077F0 to 0x77FF, and if any bytes of these is not 0xFF, it skips the remaining of the boot code and jumps to 0x0000 as a normal 8051 reset. If all bytes are in 0x077F0 to 0x77FF are 0xFF, the boot code scans the ¹²C slave 0 and 1, as well as UART0 for any In-System-Programming request. This scanning takes about 10msec. If any valid request is valid during the scan, the boot-code proceeds to follow the request and performs the programming from the host. The default ISP commands available are

UNLOCK DEVICE NAME BOOTC VERSION READ AND VERIFY Main Memory (8-Byte) ERASE Main Memory exclude Boot Code ERASE SECTOR Main Memory WRITE BYTE Main Memory SET ADDRESS CONTINUOUSE WRITE CONTINUOUS READ READ BYTE IFB WRITE BYTE IFB

Similar to writer mode, ISP is in default locked state. No command is accepted under locked state. To unlock the ISP, an 8-byte READVERIFY of 0x07FF8 to 0x07FFF must be successfully executed. Thus default ISP boot program provides similar code security as the Writer mode.

16. Electrical Characteristics

16.1 Absolute Maximum Ratings

16.2 Recommended Operating Condition

16.3 DC Electrical Characteristics (VDDHIO=VDDHA=3.0V to 5.5V TA= -40C to 85C)

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- **PACKAGE OUTLINE** $17.$
- 24-pin TSSOP 17.1

Notes:

- *1. JEDEC OUTLINE:*
	- *MO-153 AD REV.F*
- *2. DIMENSION "D" DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.15 PER SIDE.*
- *3. DIMENSION "E1" DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION. INTERLEAD SHALL NOT EXCEED 0.25 PER SIDE.*
- *4. DIMENSION "b" DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.08 MM. TOTAL IN EXCESS OF THE "b" DIMENSION AT MAXIMUM MATERIAL CONDITION. DAMBAR CANNOT BE LOCATED ON THE LOWER RADIUS OF THE FOOT. MINIMUM SPACE BETWEEN PROTRUSION AND ADJACENT LEAD IS 0.07 MM.*
- 5. *DIMENSIONS "D" AND "E1" TO BE DETERMINED AT DATUM PLANE H.*

17.2 20-pin TSSOP

Notes:

- *1. JEDEC OUTLINE:*
	- *MO-153 AC REV.F*
- *2. DIMENSION "D" DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.15 PER SIDE.*
- *3. DIMENSION "E1" DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION. INTERLEAD SHALL NOT EXCEED 0.25 PER SIDE.*
- *4. DIMENSION "b" DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.08 MM. TOTAL IN EXCESS OF THE "b" DIMENSION AT MAXIMUM MATERIAL CONDITION. DAMBAR* CANNOT BE LOCATED ON THE LOWER RADIUS OF THE FOOT. MINIMUM SPACE BETWEEN *PROTRUSION AND ADJACENT LEAD IS 0.07 MM.*
- *5. DIMENSIONS "D" AND "E1" TO BE DETERMINED AT DATUM PLANE H.*

17.3 16-pin TSSOP

VARIATIONS (ALL DIMENSIONS SHOWN IN MM)

Notes:

- *1. JEDEC OUTLINE:*
	- *STANDARD: MO-153 AB REV.F*
- *2. DIMENSION "D" DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.15 PER SIDE.*
- *3. DIMENSION "E1" DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION. INTERLEAD FLASH OR PROTRUSION SHALL NOT EXCEED 0.25 PER SIDE.*
- *4. DIMENSION "b" DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.08 MM. TOTAL IN EXCESS OF THE "b" DIMENSION AT MAXIMUM MATERIAL CONDITION. DAMBAR CANNOT BE LOCATED ON THE LOWER RADIUS OF THE FOOT. MINIMUM SPACE BETWEEN PROTRUSION AND ADJACENT LEAD IS 0.07 MM.*
- *5. DIMENSIONS "D" AND "E1" TO BE DETERMINED AT DATUM PLANE H.*

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24-pin QFN 17.4

UNIT : mm

UNIT: mm

NOTES :

- 1. JEDEC OUTLINE : N/A.
- 2. DIMENSION b APPLIES TO METALLIZED TERMINAL AND IS MEASURED BETWEEN 0.15mm AND 0.30mm FROM THE TERMINAL TIP. IF THE TERMINAL HAS THE OPTIONAL RADIUS ON THE OTHER END OF THE TERMINAL, THE DIMENSION b SHOULD NOT BE MEASURED IN THAT RADIUS AREA.
- 3. THE MINIMUM "K" VALUE OF 0.20mm APPLIES.
- 4. BILATERAL COPLANARITY ZONE APPLIES TO THE EXPOSED HEAT SINK SLUG AS WELL AS THE TERMINALS.

IS31CS8969 18. ORDERING INFORMATION

Operating temperature -40°C to 85°C

Lumissil Microsystems does not recommend the use of any of its products in life support applications where the failure or malfunction of the product can reasonably be expected to cause failure of the life support system or to significantly affect its safety or effectiveness. Products are not authorized for use in such applications unless Lumissil Microsystems receives written assurance to its satisfaction, that:

a.) the risk of injury or damage has been minimized;

b.) the user assume all such risks; and

c.) potential liability of Lumissil Microsystems is adequately protected under the circumstances

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19. Revision History

Revision A

Formal mass production release version

V1.25

Change to ISSI naming rule and ordering information

V1.24

Correct ADCIF and ADCINTE in Interrupt Diagram at page 25 ADCIF is a read-only bit ADCPGA (0xB9h) is write-only SFR I2CSST2[0] (NACK) cannot be cleared by S/W PWMFG is located at 0XA08C

V1.23

Correct the typos in FEATURES, Flash Controller CYC[2-0] table, and Electrical Characteristics. Revise TCON.

V1.22

Revise ADCCTL[0] with CSTART Update the LVDTH formula Revise I2C Master SFR. I2CMCR[7] is I2CMRST and I2CMCR[6] is INFILEN

V1.21

Correct CLKSEL default value Remove SPI SSN function from P1.7 Remove the STATUS (0xC5) description from section 1.2 and merge it into section 1.7

V1.2

Update 3.2 1.0V Reference (BGREF) Revise 3.5 Supply Low Voltage Detection (LVD), LVD range = 1.8V ~ 5.5V Add 11.1 On-Chip Temperature Sensor

V1.1 for B0

Add IS31CS8969W with QFN-24 package

V1.0 for B0

Revise the PIN definition included P1.7, P2.0, P2.1, P2.2, and P2.3 Revise Register Map SFR and Register Map XFR Add LVDHYS at XFR 0xA016 and I2CSADR2 Add EUART OPL at LININTEN[6] Update IO Configuration and Multifunction Configuration Correct some errors in painting.

V0.90

Modify IFB Block description For Engineering Sample Release

V0.13

WDCON default value is 0x00 not 0x02 Revise IOSCDIV table Timer0 and Timer2 don't support external input T0 and T2EX function. MFCFGP1.4[7] is T1EN not KEY12EN , MFCFGP1.5[7] is T2EN not KEY13EN

V0.12

Add VDAC functions.

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IS31CS8969

V0.11

Modified the pin connections and pin descriptions

V0.10

Inherit from CS5523 V0.23. Add ADDR0DC control in I2CS2.