

Evaluation Board for CS4362A

Features

- ◆ Demonstrates recommended layout and grounding arrangements
- ◆ CS8416 receives S/PDIF, & EIAJ-340 compatible digital audio
- ◆ Headers for external audio input for either PCM or DSD®
- ◆ Requires only a digital signal source and Power supplies for a complete digital-to-analog converter system

Description

The CDB4362A evaluation board is an excellent means for quickly evaluating the CS4362A 24-bit, 48-pin, 6-channel D/A converter. Evaluation requires an analog signal analyzer, a digital signal source, a PC for controlling the CS4362A (only required for control port mode), and a power supply. Analog line-level outputs are provided via RCA phono jacks.

The CS8416 digital audio receiver IC provides the system timing necessary to operate the digital-to-analog converter and will accept S/PDIF and EIAJ-340-compatible audio data. The evaluation board may also be configured to accept external timing and data signals for operation in a user application during system development.

ORDERING INFORMATION

CDB4362A

Evaluation Board

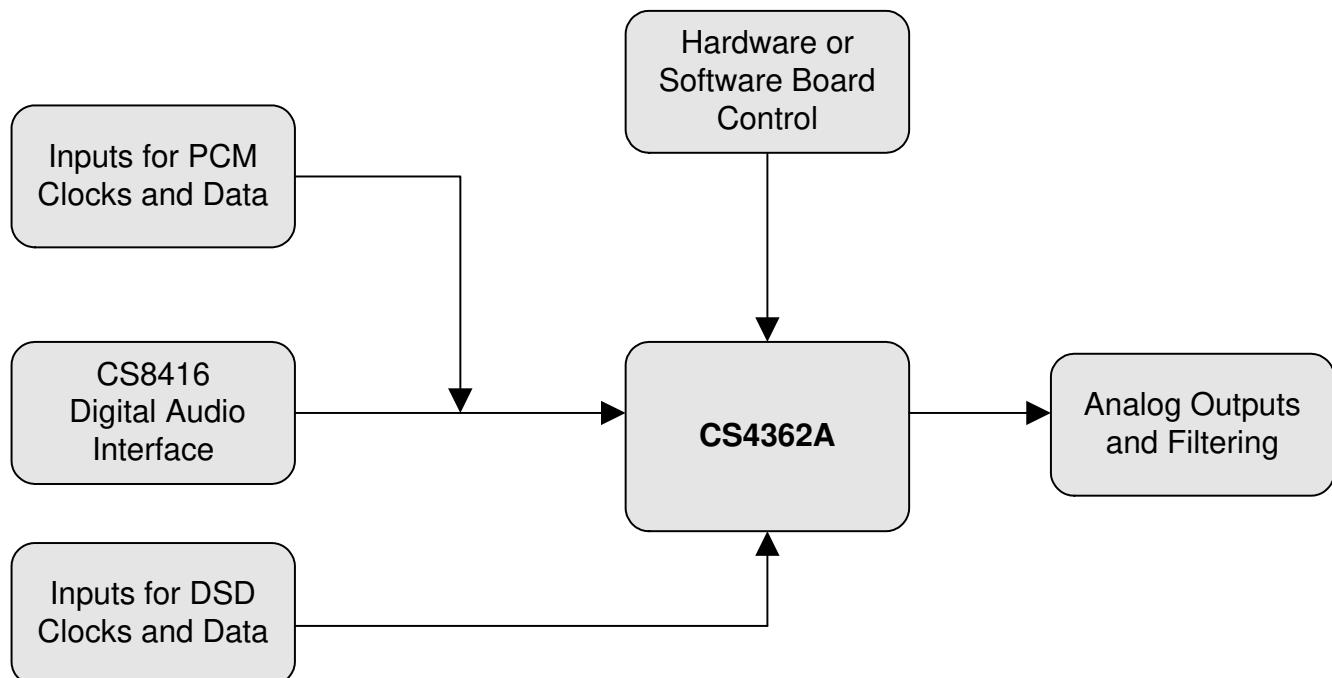


TABLE OF CONTENTS

1. CS4362A DIGITAL-TO-ANALOG CONVERTER	4
2. CS8416 DIGITAL AUDIO RECEIVER	4
3. INPUT FOR CLOCKS AND DATA	4
4. INPUT FOR CONTROL DATA	5
5. POWER SUPPLY CIRCUITRY	5
6. GROUNDING AND POWER SUPPLY DECOUPLING	5
7. ANALOG OUTPUT FILTERING	5
8. PERFORMANCE PLOTS	7
9. CDB4362A SCHEMATICS	17
10. REVISION HISTORY	30

LIST OF FIGURES

Figure 1.FFT (48 kHz, 0 dB)	7
Figure 2.FFT (48 kHz, -60 dB)	7
Figure 3.FFT (48 kHz, No Input)	7
Figure 4.FFT (48 kHz Out-of-Band, No Input)	7
Figure 5.FFT (48 kHz -60 dB Wideband)	8
Figure 6.FFT (IMD 48 kHz)	8
Figure 7.48 kHz, THD+N vs. Input Freq	8
Figure 8.48 kHz, THD+N vs. Level	8
Figure 9.48 kHz, Fade-to-Noise Linearity	8
Figure 10.48 kHz, Frequency Response	8
Figure 11.48 kHz, Crosstalk	9
Figure 12.48 kHz, Impulse Response	9
Figure 13.48 kHz, Impulse Prefilter	9
Figure 14.Dynamic Range 48 kHz	10
Figure 15.FFT (96 kHz, 0 dB)	10
Figure 16.FFT (96 kHz, -60 dB)	10
Figure 17.FFT (96 kHz, No Input)	11
Figure 18.FFT (96 kHz Out-of-Band, No Input)	11
Figure 19.FFT (96 kHz, -60 dB Wideband)	11
Figure 20.FFT (IMD 96 kHz)	11
Figure 21.96 kHz, THD+N vs. Input Freq	11
Figure 22.96 kHz, THD+N vs. Level	11
Figure 23.96 kHz, Fade-to-Noise Linearity	12
Figure 24.96 kHz, Frequency Response	12
Figure 25.96 kHz, Crosstalk	12
Figure 26.96 kHz, Impulse Response	12
Figure 27.96 kHz, Impulse Prefilter	12
Figure 28.Dynamic Range 96 kHz	13
Figure 29.FFT (192 kHz, 0 dB)	13
Figure 30.FFT (192 kHz, -60 dB)	13
Figure 31.FFT (192 kHz, No Input)	14
Figure 32.FFT (192 kHz Out-of-Band, No Input)	14
Figure 33.FFT (192 kHz, -60 dB Wideband)	14
Figure 34.FFT (IMD 192 kHz)	14
Figure 35.192 kHz, THD+N vs. Input Freq	14
Figure 36.192 kHz, THD+N vs. Level	14
Figure 37.192 kHz, Fade-to-Noise Linearity	15
Figure 38.192 kHz, Frequency Response	15
Figure 39.192 kHz, Crosstalk	15
Figure 40.192 kHz, Impulse Response	15

Figure 41.192 kHz, Impulse Prefilter	15
Figure 42.Dynamic Range 192 kHz	16
Figure 43.System Block Diagram and Signal Flow	17
Figure 44.CS4362A	18
Figure 45.Analog Outputs A1 - B1	19
Figure 46.Analog Outputs A2 - B2	20
Figure 47.Analog Outputs A3 - B3	21
Figure 48.CS8416 S/PDIF Input	22
Figure 49.PCM Input Header and Multiplexing	23
Figure 50.DSD Input Header	24
Figure 51.Control Input	25
Figure 52.Power Input	26
Figure 53.Silkscreen Top	27
Figure 54.Top Side	28
Figure 55.Bottom Side	29

LIST OF TABLES

Table 1. System Connections	5
Table 2. CDB4362A Jumper Settings	6

CDB4362A SYSTEM OVERVIEW

The CDB4362A evaluation board is an excellent means of quickly evaluating the CS4362A. The CS8416 digital audio interface receiver provides an easy interface to digital audio signal sources including the majority of digital audio test equipment. The evaluation board also allows the user to supply external PCM or DSD clocks and data through PCB headers for system development.

The CDB4362A uses the CDB4385 as a base PCB board. For this reason, there may be additional circuitry on board that is not populated because it has no function for this device.

The CDB4362A schematic has been partitioned into the nine schematics shown in [Figures 44](#) through [52](#). Each partitioned schematic is represented in the system diagram shown in [Figure 43](#). Notice that the system diagram also includes the interconnections between the partitioned schematics.

1. CS4362A DIGITAL-TO-ANALOG CONVERTER

A description of the CS4362A is included in the CS4362A datasheet.

2. CS8416 DIGITAL AUDIO RECEIVER

The system receives and decodes the standard S/PDIF data format using a CS8416 digital audio receiver ([Figure 48](#)). The outputs of the CS8416 include a serial bit clock, serial data, left-right clock, and a 128/256 Fs master clock. The CS8416 data format is fixed to I²S. The operation of the CS8416 and a discussion of the digital audio interface are included in the CS8416 datasheet.

The evaluation board has been designed such that the input can be either optical or coaxial (See [Figure 48](#)). However, both inputs cannot be driven simultaneously.

Switch position 7 of S1 sets the output MCLK-to-LRCK ratio of the CS8416. This switch should be set to 256 (closed) for inputs $F_s \leq 96$ kHz and 128 (open) for $F_s \geq 64$ kHz. The 8416 must be manually reset using 'HW RST' (S2) or through the software when this switch is changed.

3. INPUT FOR CLOCKS AND DATA

The evaluation board has been designed to allow interfacing to external systems via headers J11 and J7. Header J11 allows the evaluation board to accept externally generated PCM clocks and data. The schematic for the clock/data input is shown in [Figure 49](#). Switch position 6 of S1 selects the source as either CS8416 (open) or header J11 (closed).

Header J7 allows the evaluation board to accept externally generated DSD data and clocks. The schematic for the clock/data input is shown in [Figure 50](#). A synchronous MCLK must still be provided via Header J11. Switch position 8 of S1 selects either PCM (open) or DSD (closed).

Please see the CS4362A datasheet for more information.

4. INPUT FOR CONTROL DATA

The evaluation board can be run in either a stand-alone mode or with a PC. Stand-alone mode uses the CS4362A in hardware mode and the mode pins are configured using switch positions 1 through 5 of S1. PC mode uses software to set up the CS4362A through I²C® using the PC's serial or USB ports. PC mode is automatically selected when the serial or USB port is attached and the CDB4362A software is running.

Header J15 offers the option for external input of RST and SPI™/I²C clocks and data. The board is set up at the factory to use the on-board microcontroller in conjunction with the supplied software. To use an external control source, remove the shunts on J15 and place a ribbon cable so the signal lines are on the center row and the grounds are on the right side. R116 and R119 should be populated with 2-kΩ resistors when using an external I²C source that does not already provide pull-ups.

5. POWER SUPPLY CIRCUITRY

Power is supplied to the evaluation board by four binding posts (GND, +5V, +12V, and -12V), see [Figure 53](#). The '+5V' terminal supplies VA and the rest of the +5-V circuitry on the board. The +3.3-V circuitry is powered from a regulator. The +2.5 volts required for VD is also provided from an on-board regulator. The +5-V supply should be set within the recommended values for VA stated in the CS4362A datasheet.

WARNING: Refer to the CS4362A datasheet for maximum allowable voltage levels. Operation outside this range can cause permanent damage to the device.

6. GROUNDING AND POWER SUPPLY DECOUPLING

As with any high-performance converter, the CS4362A requires careful attention to power supply and grounding arrangements to optimize performance. [Figure 44](#) details the connections to the CS4362A and [Figures 53, 54](#), and [55](#) show the component placement and top and bottom layout. The decoupling capacitors are located as close to the CS4362A as possible. Extensive use of ground-plane fill in the evaluation board yields large reductions in radiated noise.

7. ANALOG OUTPUT FILTERING

The analog output on the CDB4362A has been designed according to the CS4362A datasheet. This output circuit includes an active 2-pole, 50-kHz filter which uses the multiple-feedback topology.

CONNECTOR	INPUT/OUTPUT	SIGNAL PRESENT
+5V	Input	+ 5 V power
GND	Input	Ground connection from power supply
+12V	Input	+12 V positive supply for the on-board filtering
-12V	Input	-12 V negative supply for the on-board filtering
S/PDIF IN - J9	Input	Digital audio interface input via coax
S/PDIF IN - OPT1	Input	Digital audio interface input via optical
PCM INPUT - J11	Input	Input for master, serial, left/right clocks and serial data
DSD INPUT - J7	Input	Input for DSD serial clock and DSD data
OUTA1-B3	Output	RCA line level analog outputs

Table 1. System Connections

JUMPER / SWITCH	PURPOSE	POSITION	FUNCTION SELECTED
J15	Selects source of control data	*shunts on Left shunts removed	*Control from PC and on-board microcontroller External control input using center and right columns
J16	JTAG micro programming	-	Reserved for factory use only
S2	Resets CS8416 and CS4362A		The CS8416 must be reset if switch S1 is changed
S1	CS4362A mode settings M0-M4	1-5	Default: M0, M4 open (HI) M1, M2, M3 closed (LO)
	Sets clock source	6	Sets clock source for CS4362A *open = RX(CS8416), closed = <u>EXT</u> (J11)
	Sets MCLK ratio of CS8416	7	Selects 128x (open) or 256x (*closed) MCLK/LRCK ratio output for CS8416
	Selects PCM or DSD mode	8	For PCM input set to *Open, for DSD set to Closed

*Default Factory Settings

Table 2. CDB4362A Jumper Settings

8. PERFORMANCE PLOTS

The plots in the following section were achieved using an Audio Precision System 2700 and a randomly chosen production CDB4362A. In some cases the performance may be limited by the CDB4362A. All measurements were taken at room temp using the standard AP filter options (20 Hz to 22 kHz) with default board settings and nominal datasheet voltages applied unless otherwise noted.

The impulse response plots were taken both pre-and post filtering as the off-chip filter was degrading the performance at higher sample rates. The pre-filter impulse response plots were taken directly at the output pins of the DAC (with the analog filter still connected) to show the effect of the CDB's analog filtering on the impulse response (as the analog filtering adds its own signature to the impulse response of the DAC, and in the case of the higher sampling rates it was band-limiting it).

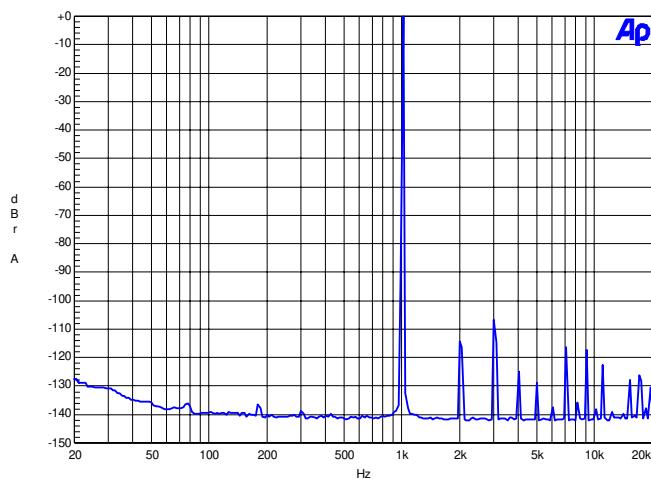


Figure 1. FFT (48 kHz, 0 dB)

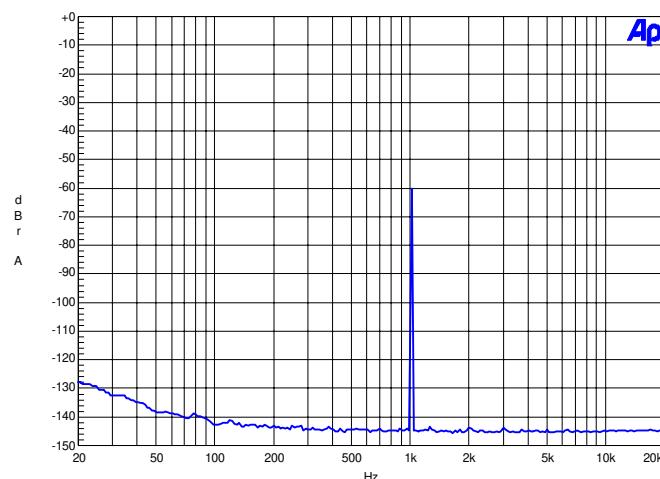


Figure 2. FFT (48 kHz, -60 dB)

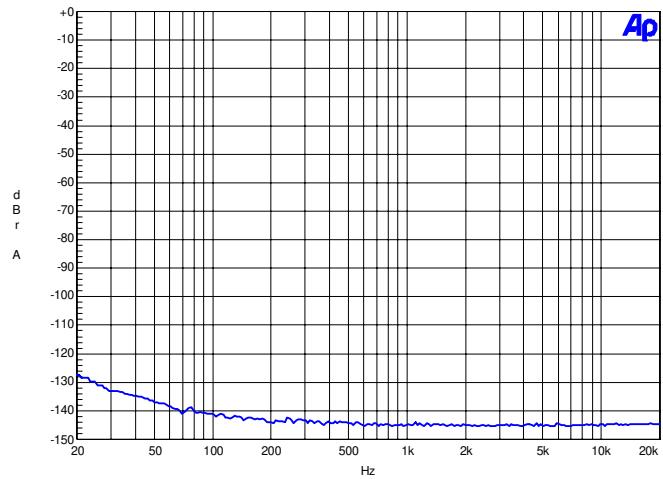


Figure 3. FFT (48 kHz, No Input)

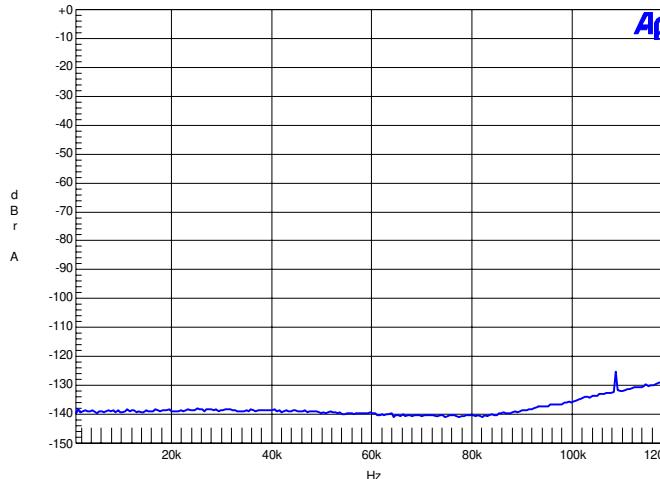


Figure 4. FFT (48 kHz Out-of-Band, No Input)

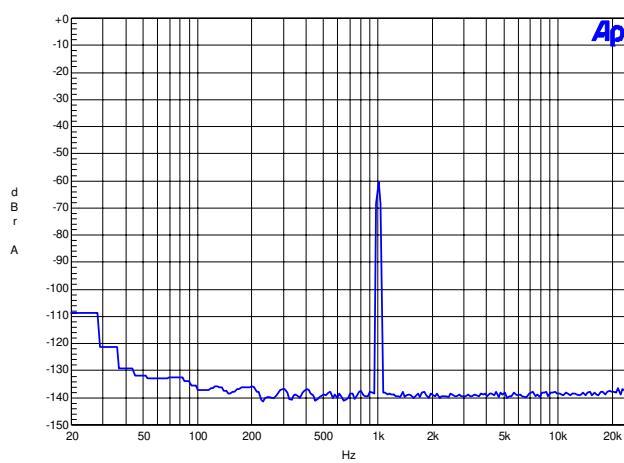


Figure 5. FFT (48 kHz -60 dB Wideband)

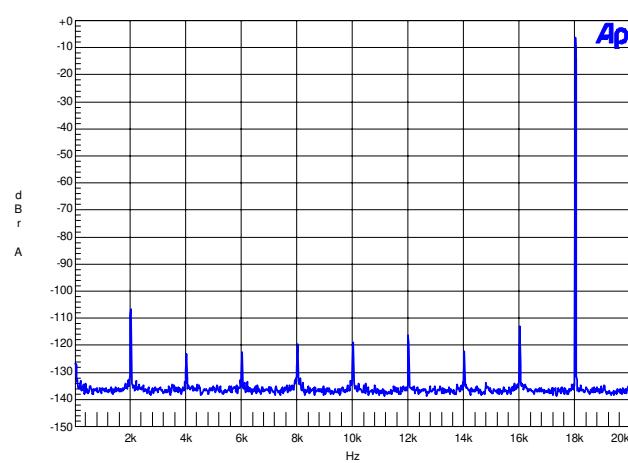


Figure 6. FFT (IMD 48 kHz)

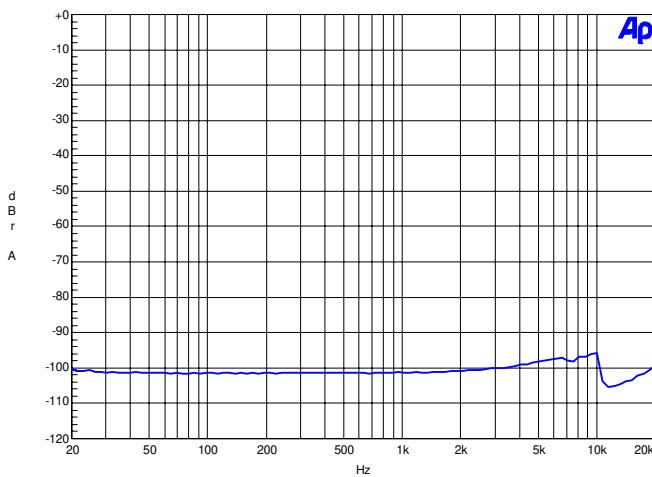


Figure 7. 48 kHz, THD+N vs. Input Freq

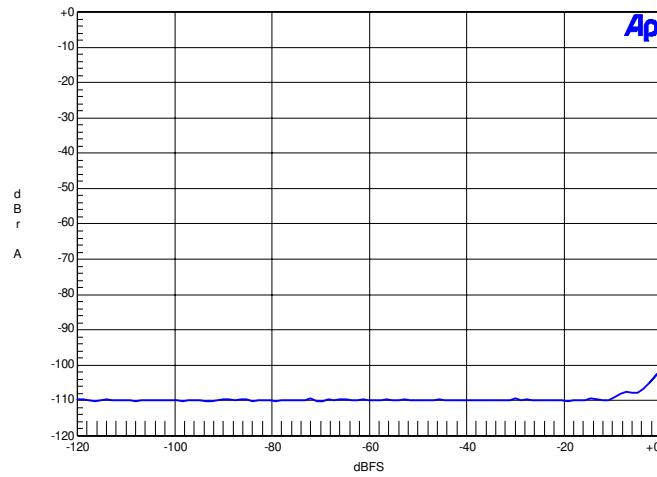


Figure 8. 48 kHz, THD+N vs. Level

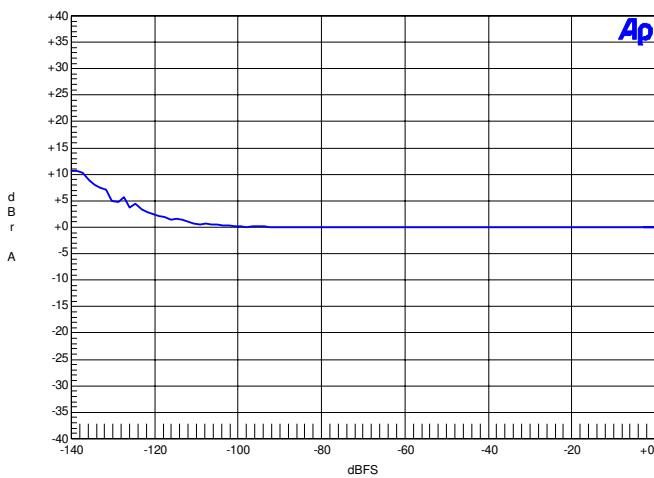


Figure 9. 48 kHz, Fade-to-Noise Linearity

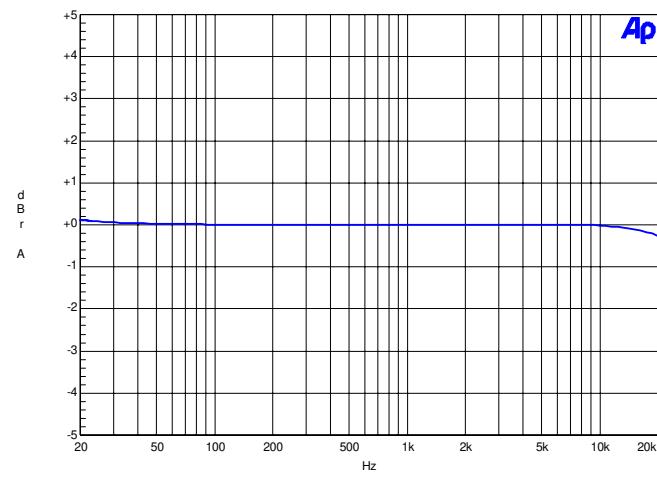


Figure 10. 48 kHz, Frequency Response

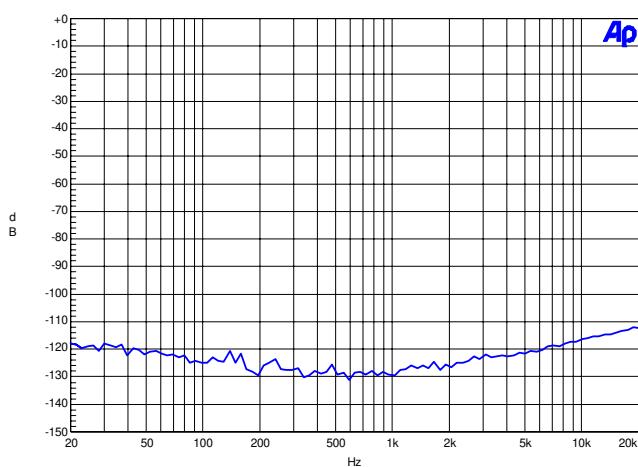


Figure 11. 48 kHz, Crosstalk

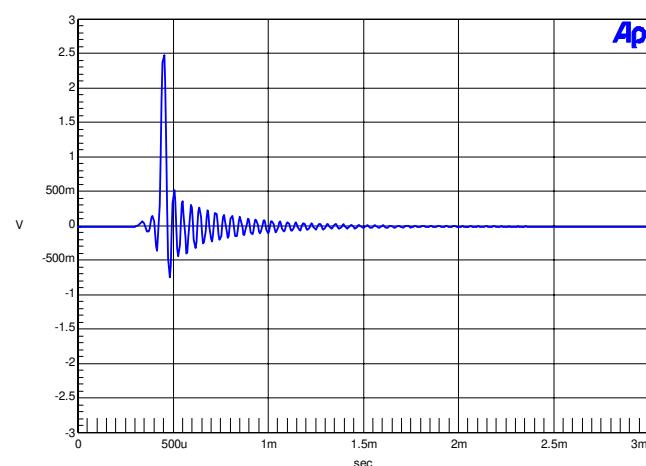


Figure 12. 48 kHz, Impulse Response

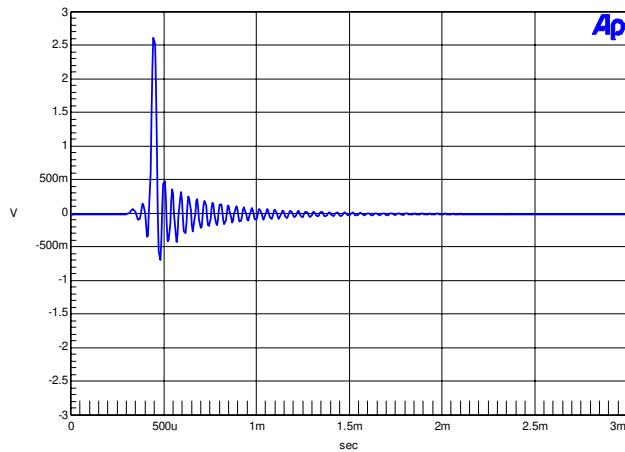


Figure 13. 48 kHz, Impulse Prefilter

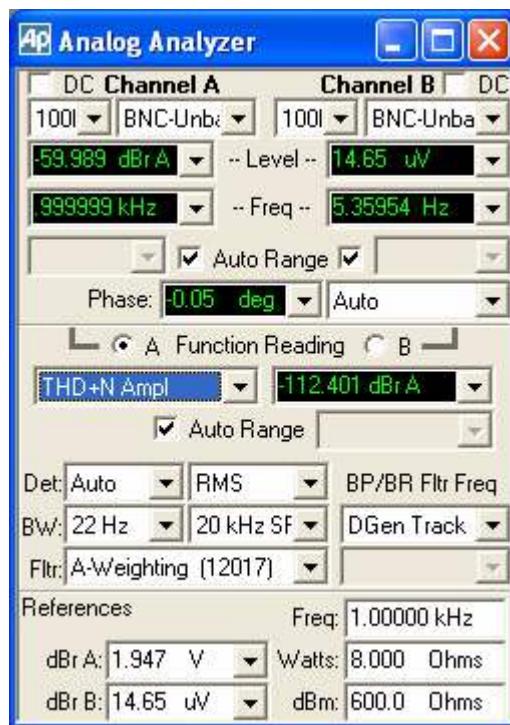


Figure 14. Dynamic Range 48 kHz

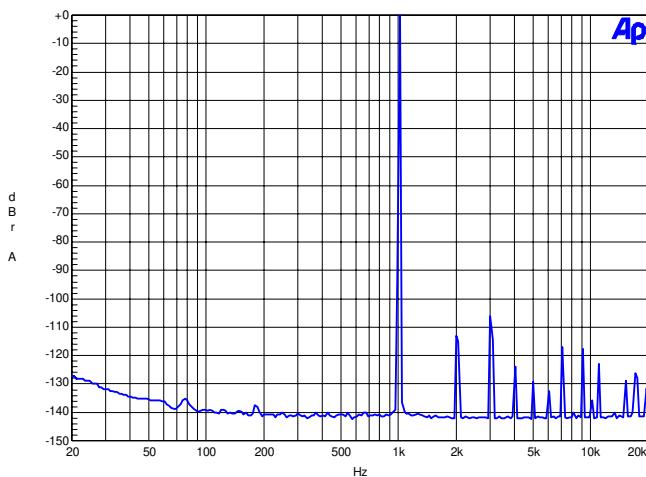


Figure 15. FFT (96 kHz, 0 dB)

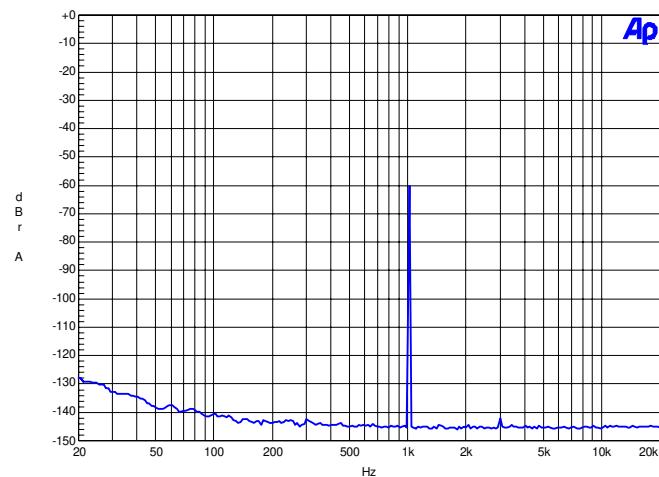


Figure 16. FFT (96 kHz, -60 dB)

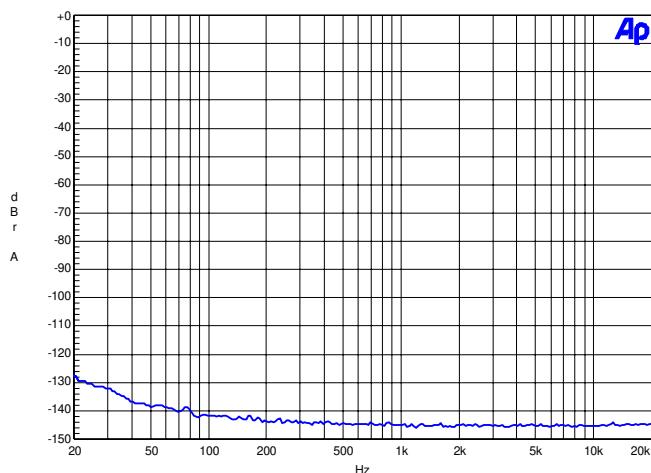


Figure 17. FFT (96 kHz, No Input)

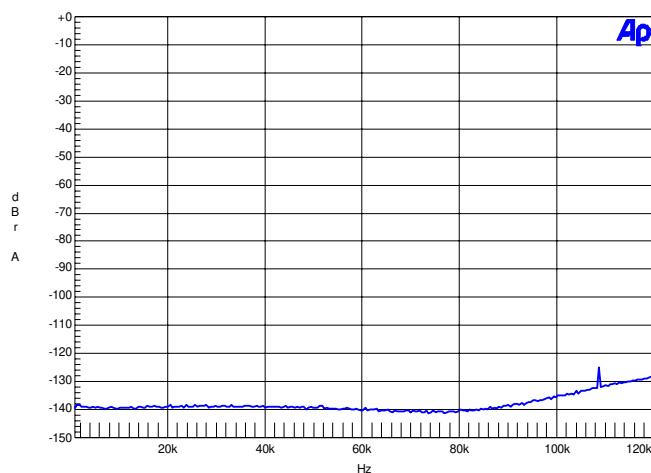


Figure 18. FFT (96 kHz Out-of-Band, No Input)

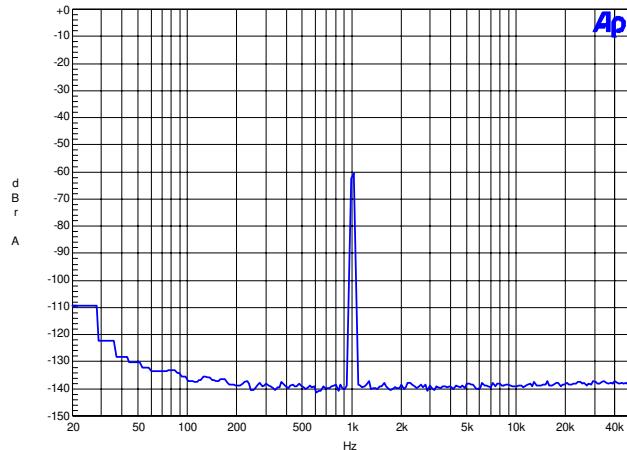


Figure 19. FFT (96 kHz, -60 dB Wideband)

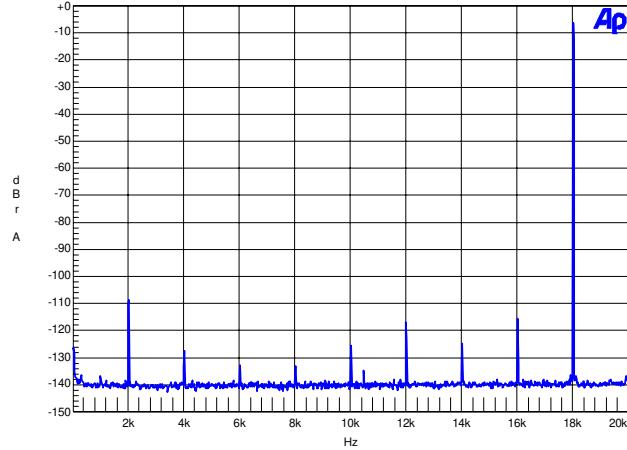


Figure 20. FFT (IMD 96 kHz)

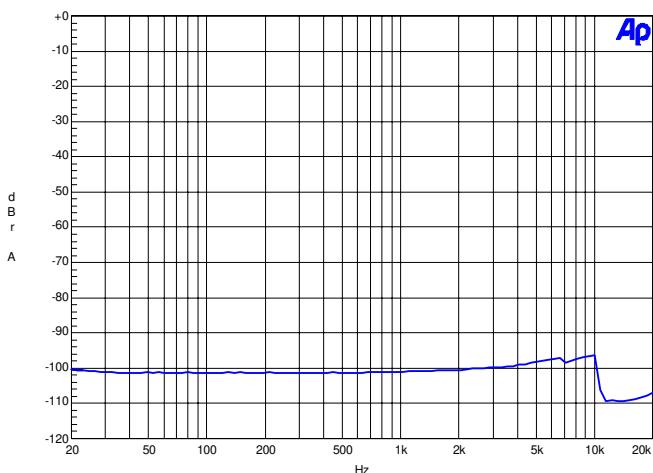


Figure 21. 96 kHz, THD+N vs. Input Freq

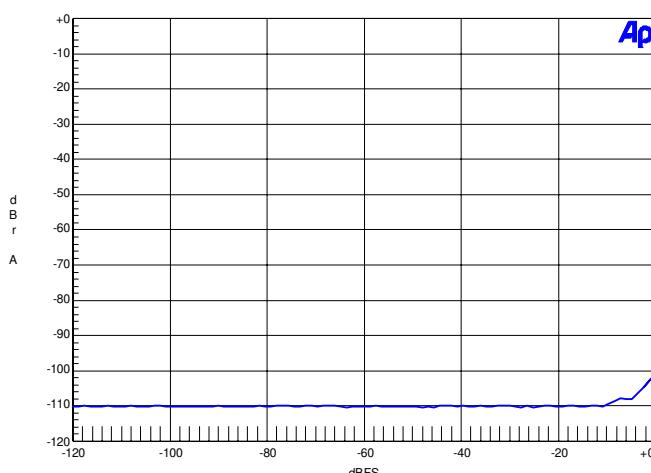


Figure 22. 96 kHz, THD+N vs. Level

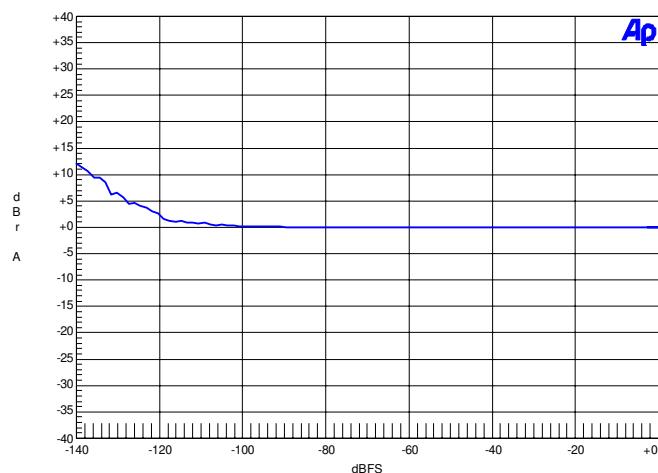


Figure 23. 96 kHz, Fade-to-Noise Linearity

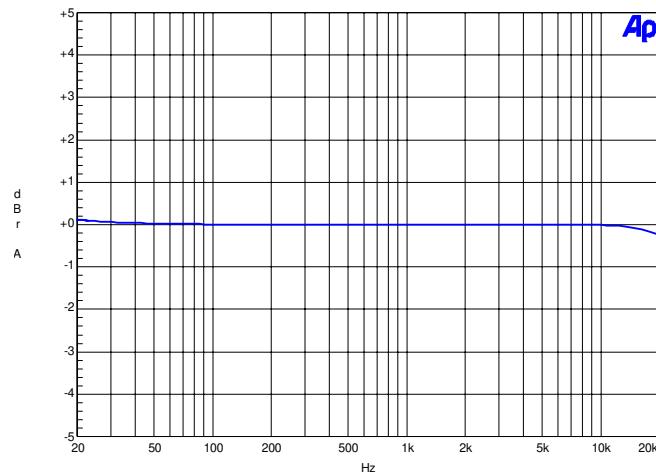


Figure 24. 96 kHz, Frequency Response

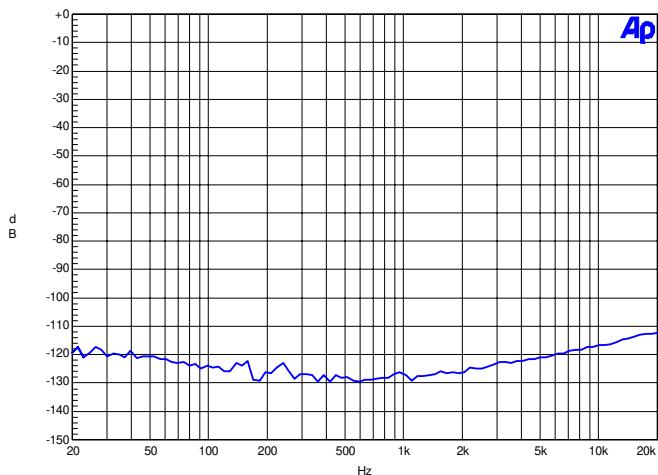


Figure 25. 96 kHz, Crosstalk

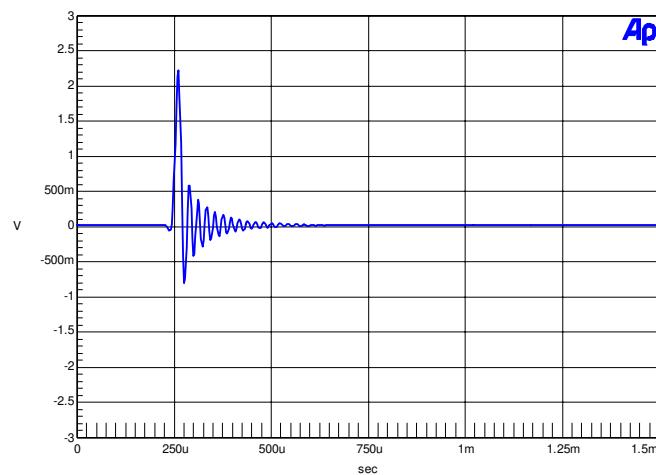


Figure 26. 96 kHz, Impulse Response

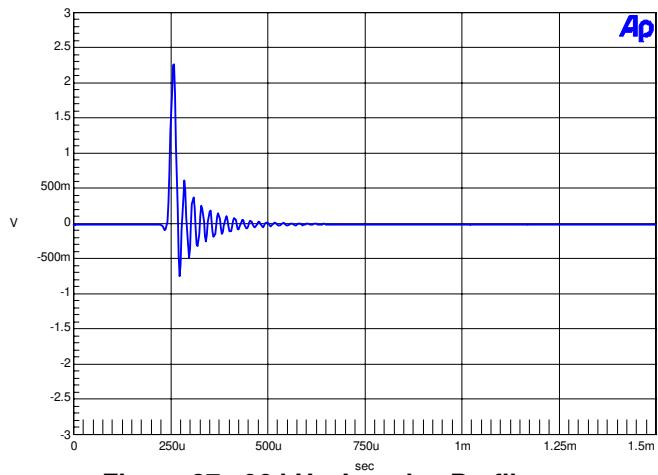


Figure 27. 96 kHz, Impulse Prefilter

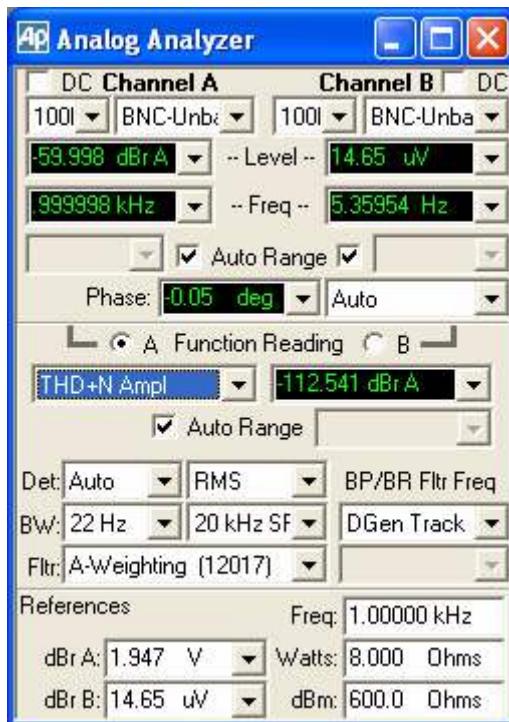


Figure 28. Dynamic Range 96 kHz

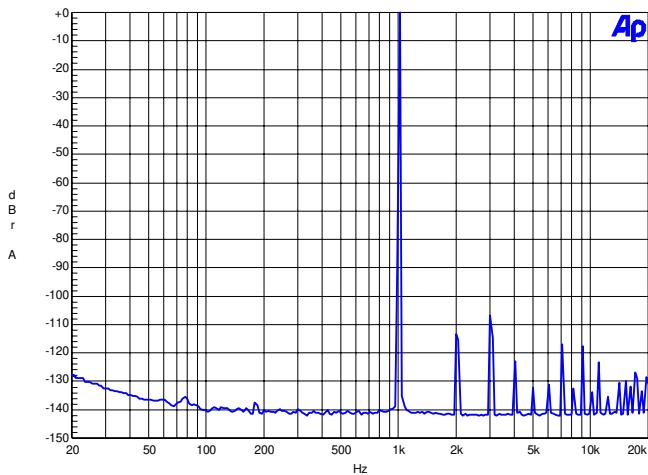


Figure 29. FFT (192 kHz, 0 dB)

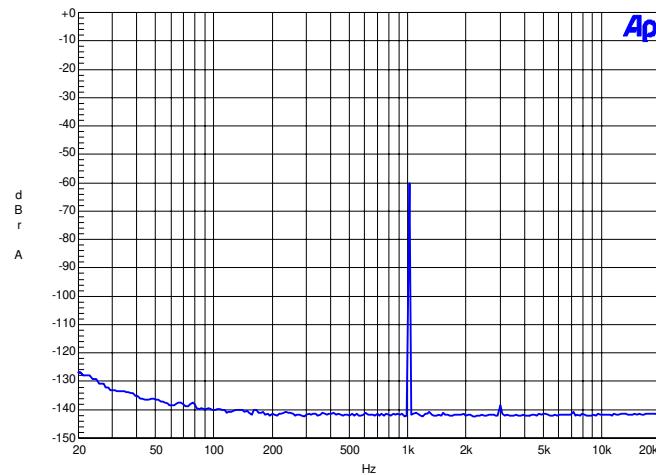


Figure 30. FFT (192 kHz, -60 dB)

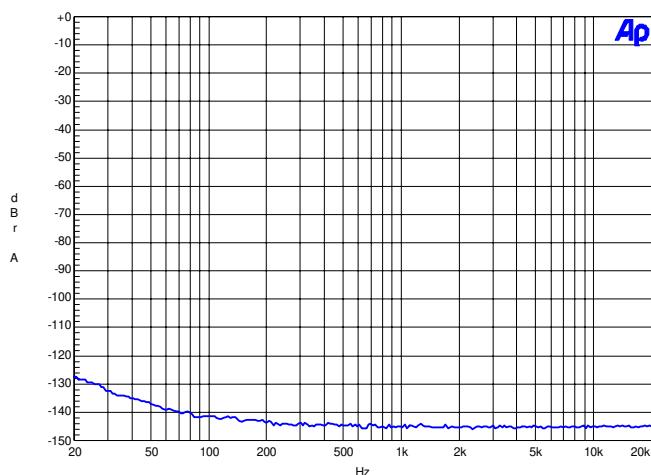


Figure 31. FFT (192 kHz, No Input)

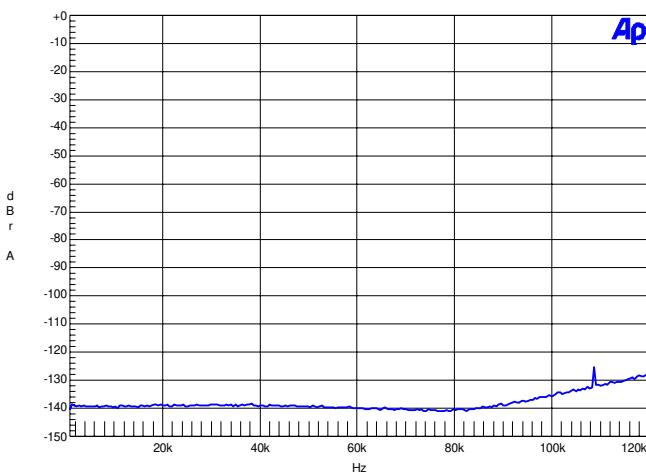


Figure 32. FFT (192 kHz Out-of-Band, No Input)

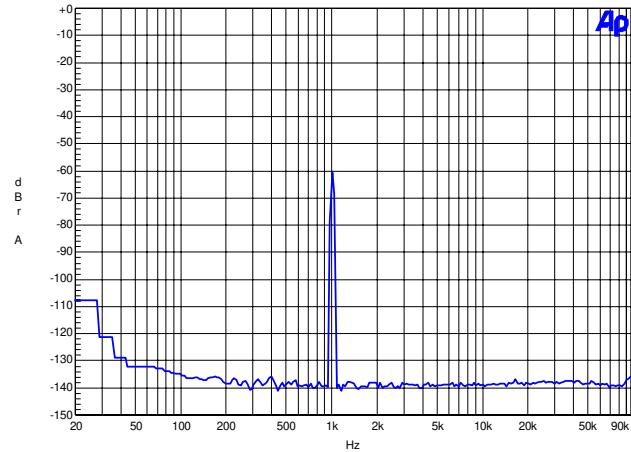


Figure 33. FFT (192 kHz, -60 dB Wideband)

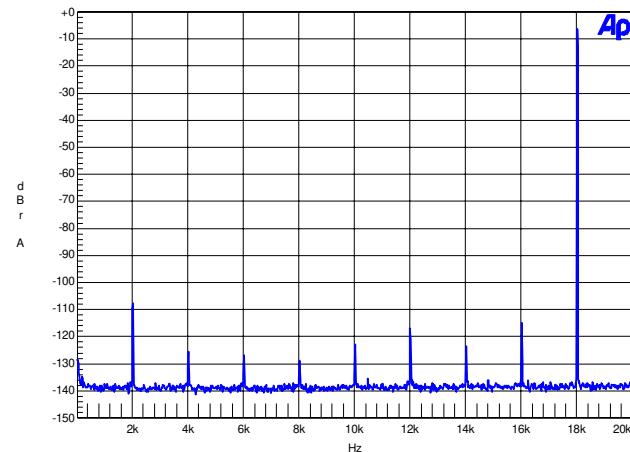


Figure 34. FFT (IMD 192 kHz)

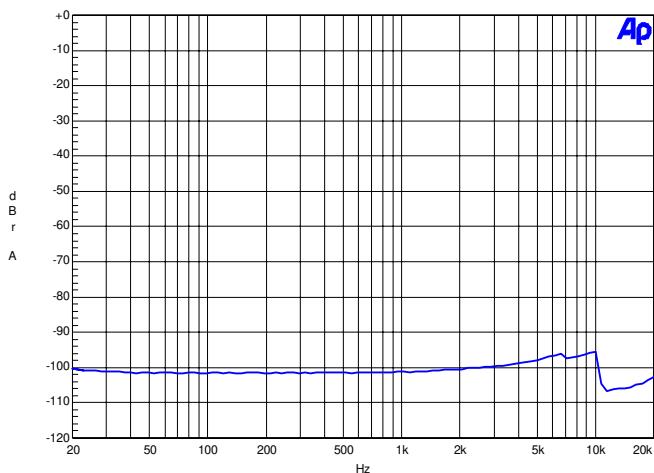


Figure 35. 192 kHz, THD+N vs. Input Freq

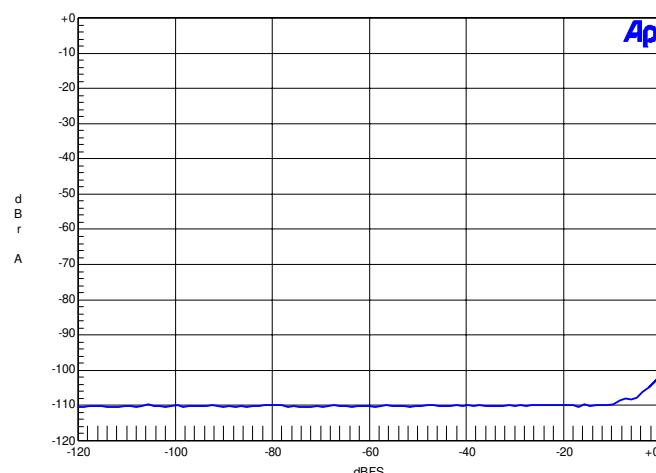


Figure 36. 192 kHz, THD+N vs. Level

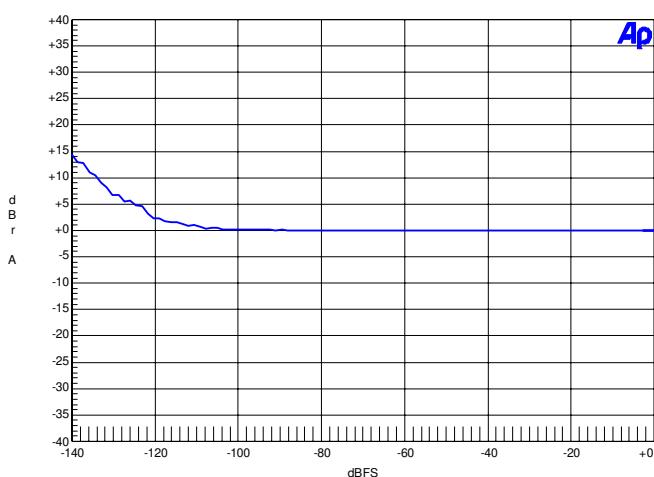


Figure 37. 192 kHz, Fade-to-Noise Linearity

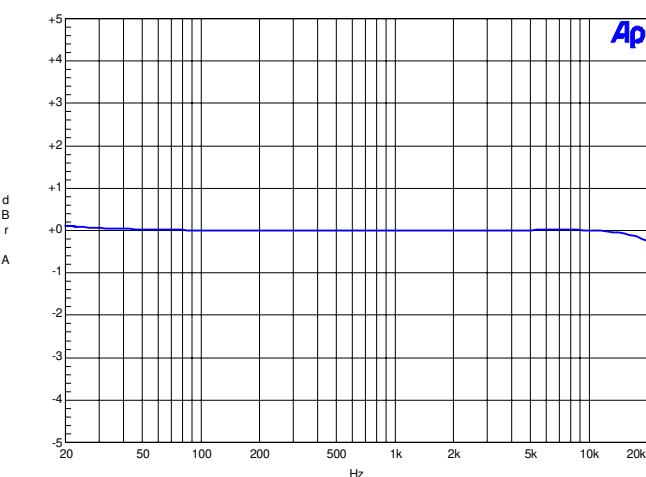


Figure 38. 192 kHz, Frequency Response

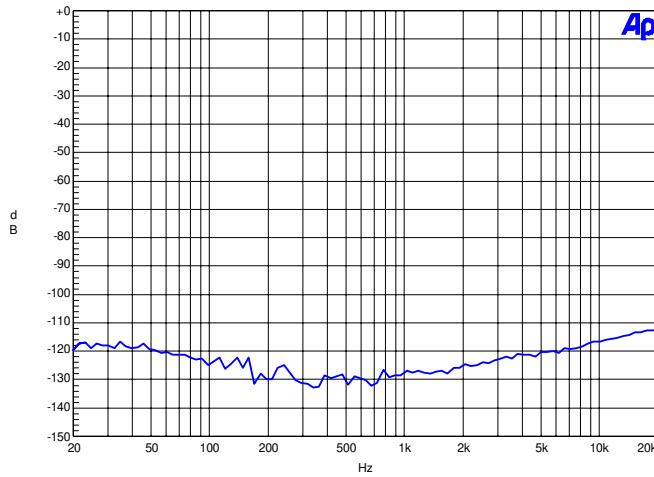


Figure 39. 192 kHz, Crosstalk

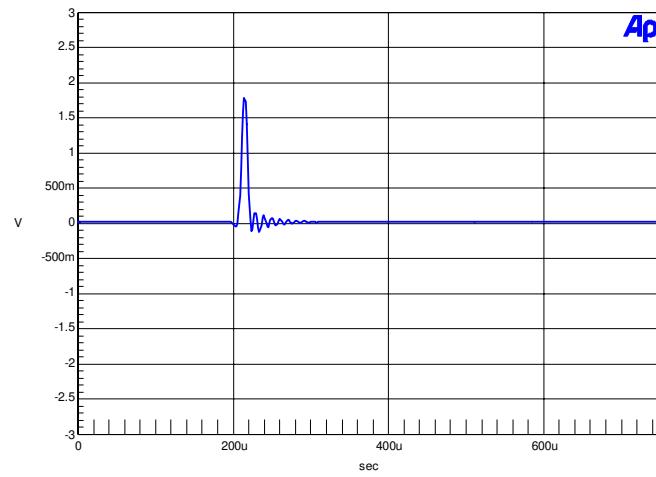


Figure 40. 192 kHz, Impulse Response

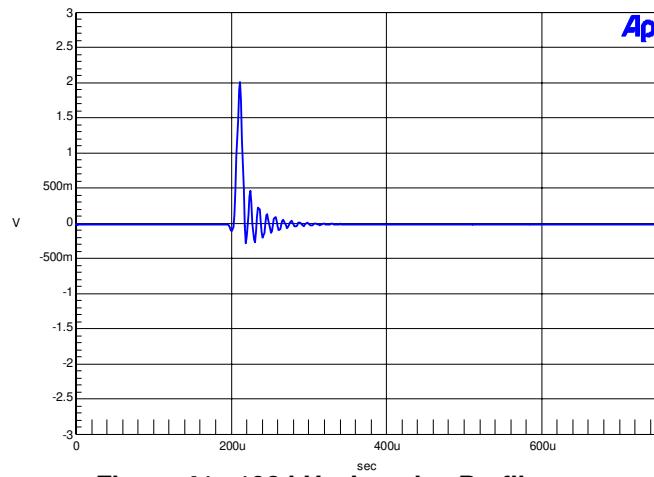


Figure 41. 192 kHz, Impulse Prefilter

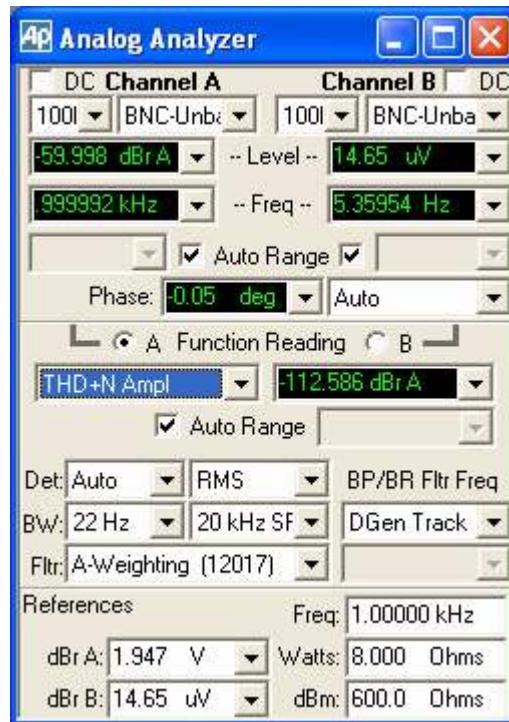


Figure 42. Dynamic Range 192 kHz

Figure 43. System Block Diagram and Signal Flow

9. CDB4362A SCHEMATICS

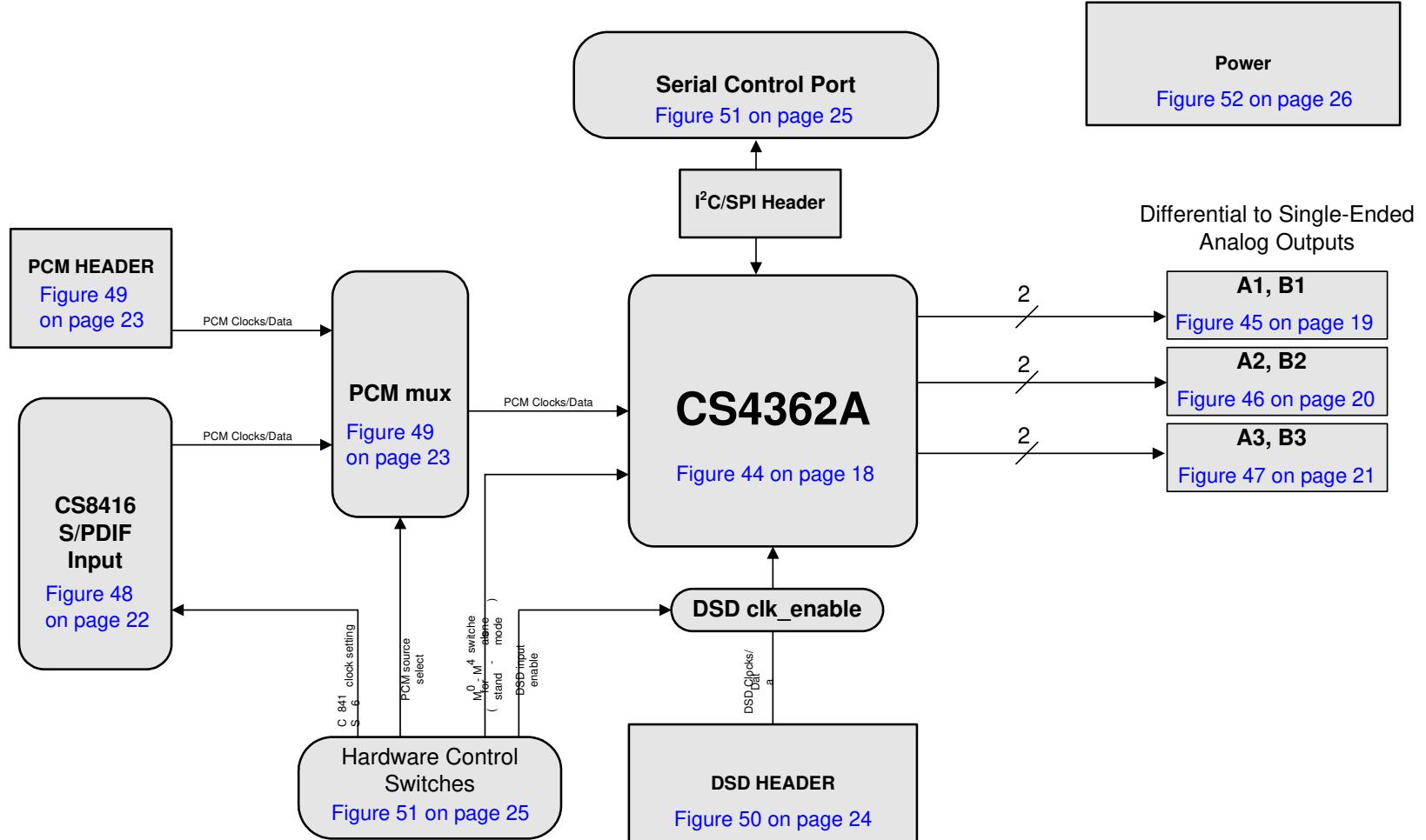
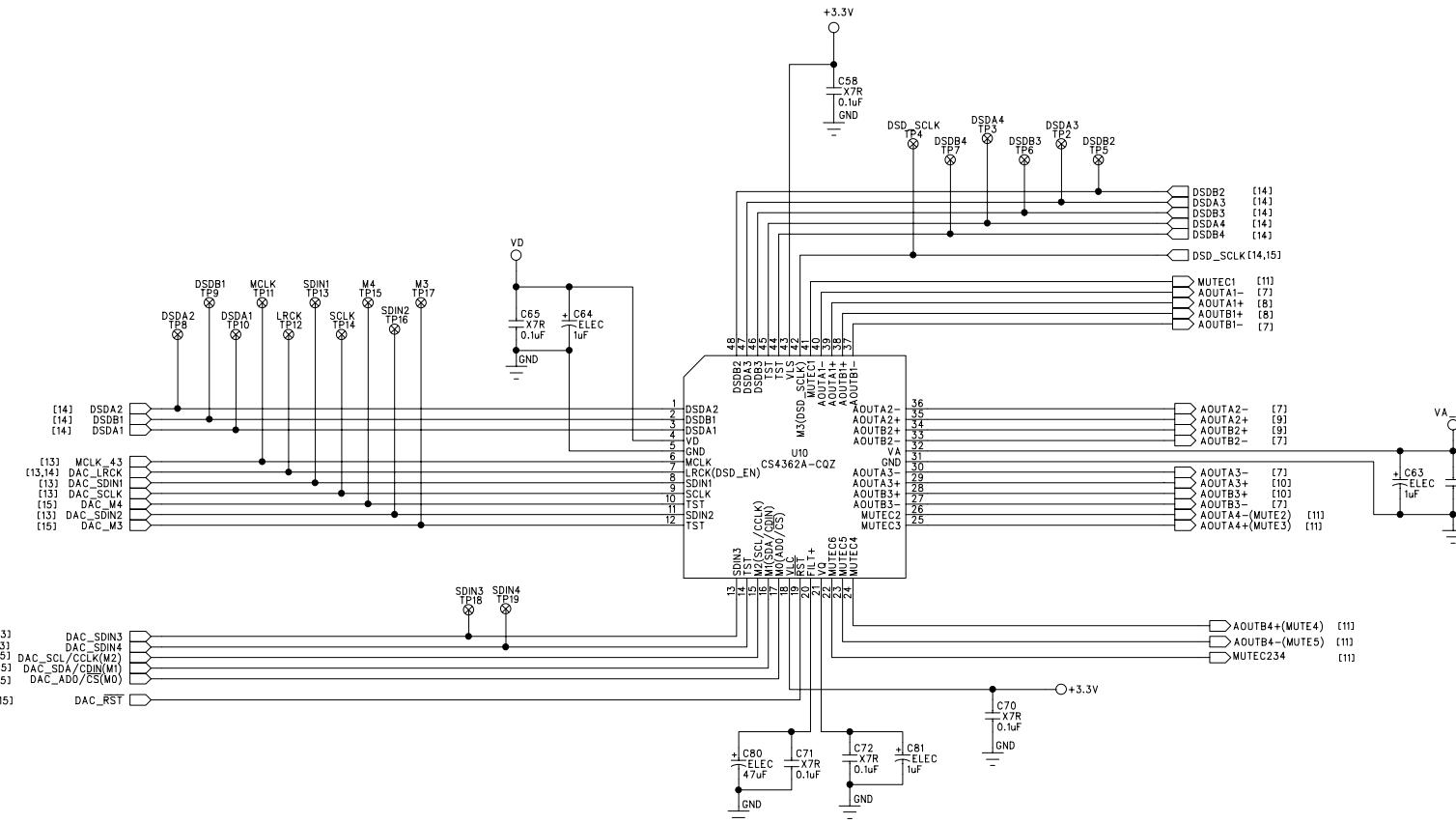
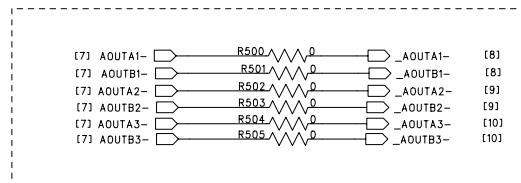


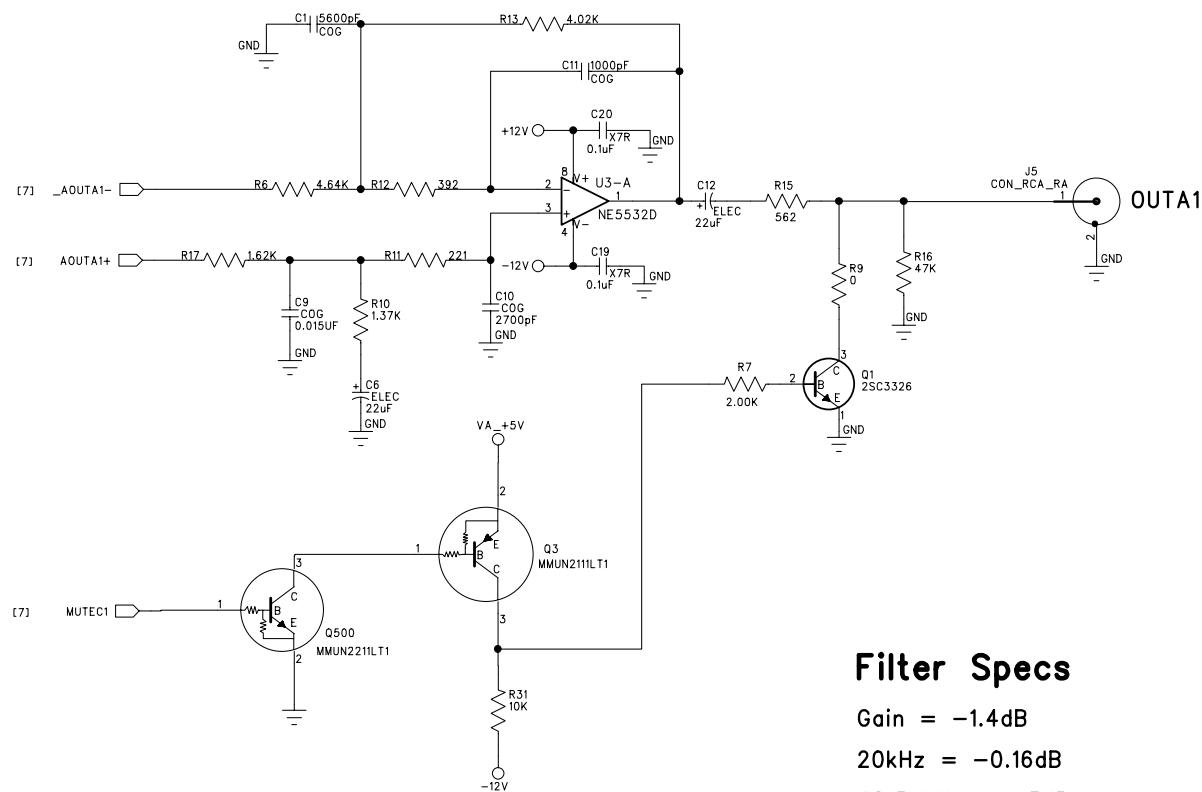


Figure 44. CS4362A



NOTE: ALL RESISTOR VALUES ARE IN OHMS
UNLESS OTHERWISE SPECIFIED.





Filter Specs

Gain = -1.4dB

20kHz = -0.16dB

48.5 kHz = -3dB

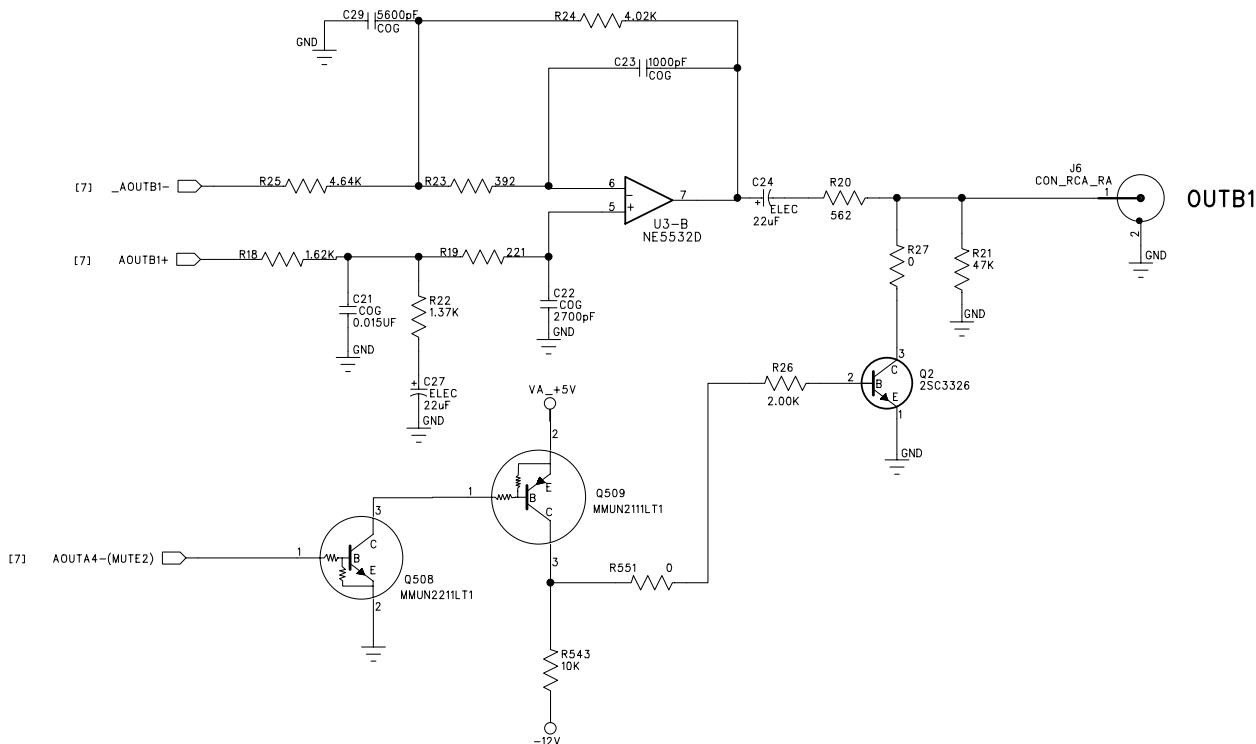
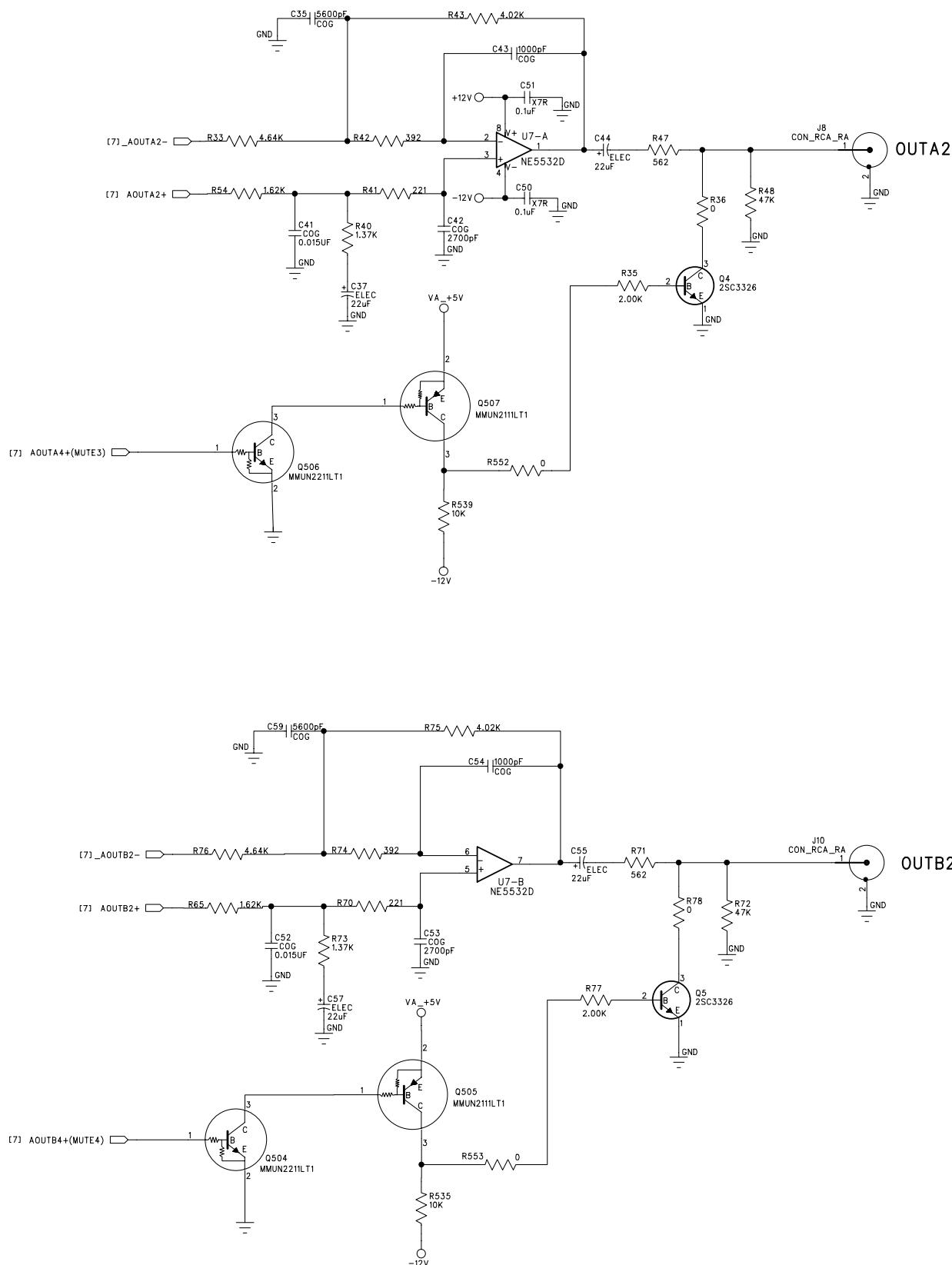


Figure 45. Analog Outputs A1 - B1


Figure 46. Analog Outputs A2 - B2

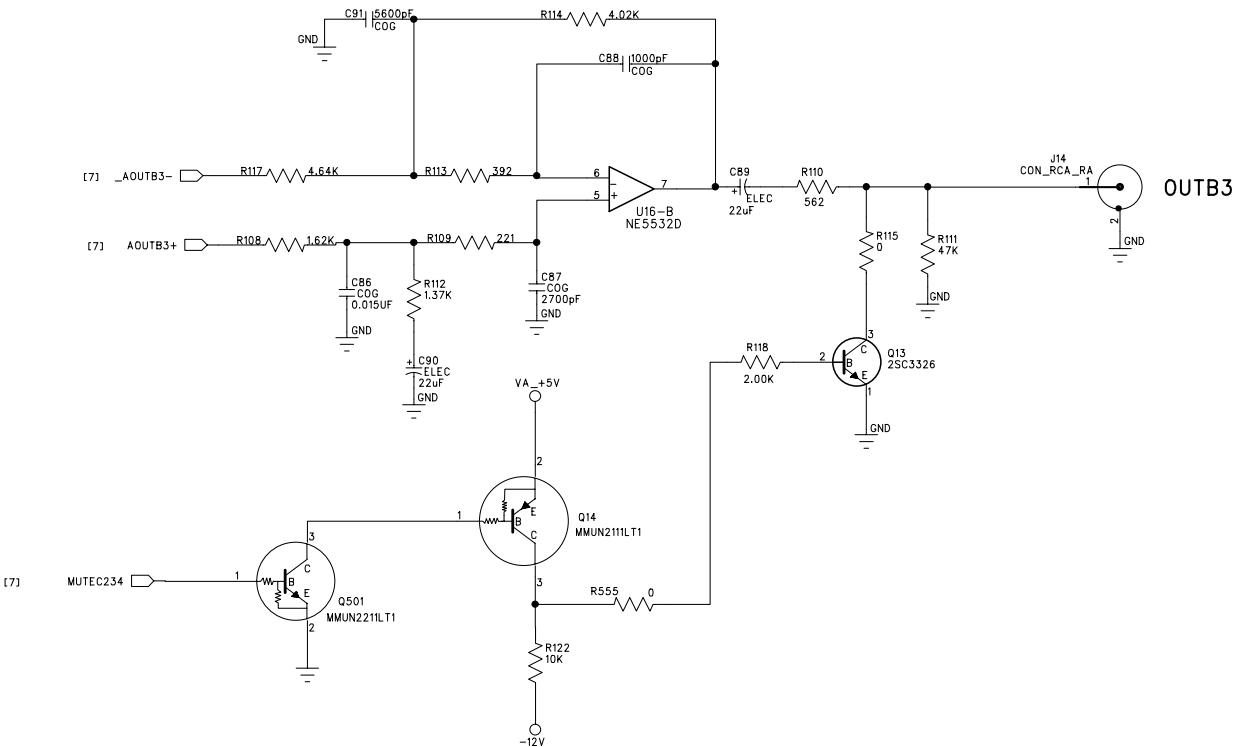
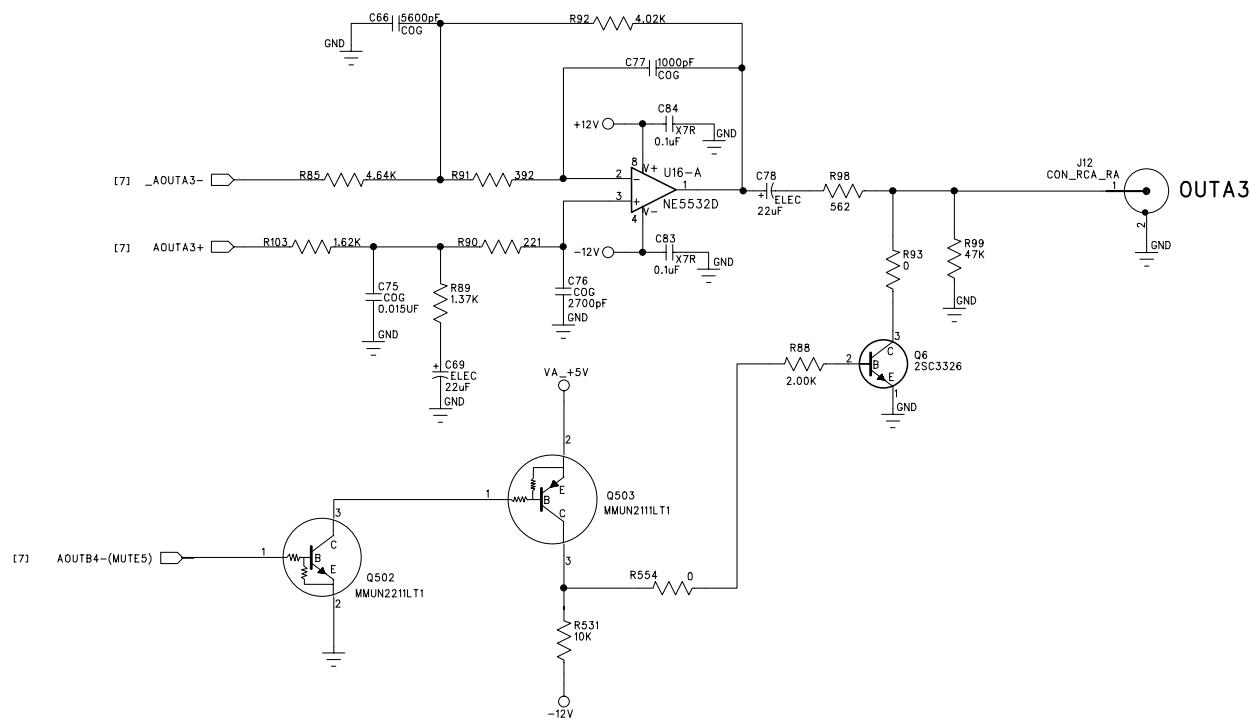
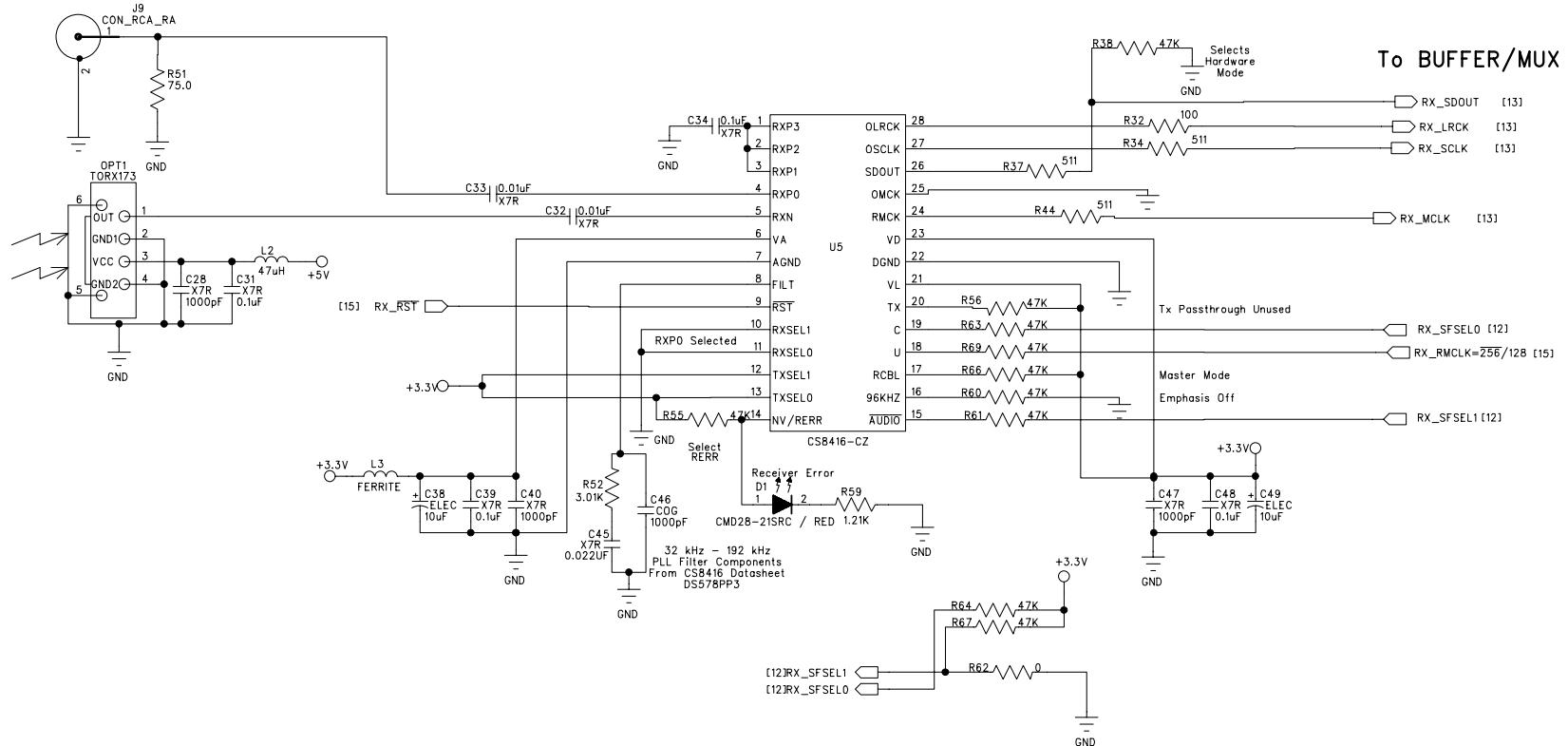


Figure 47. Analog Outputs A3 - B3


Figure 48. CS8416 S/PDIF Input

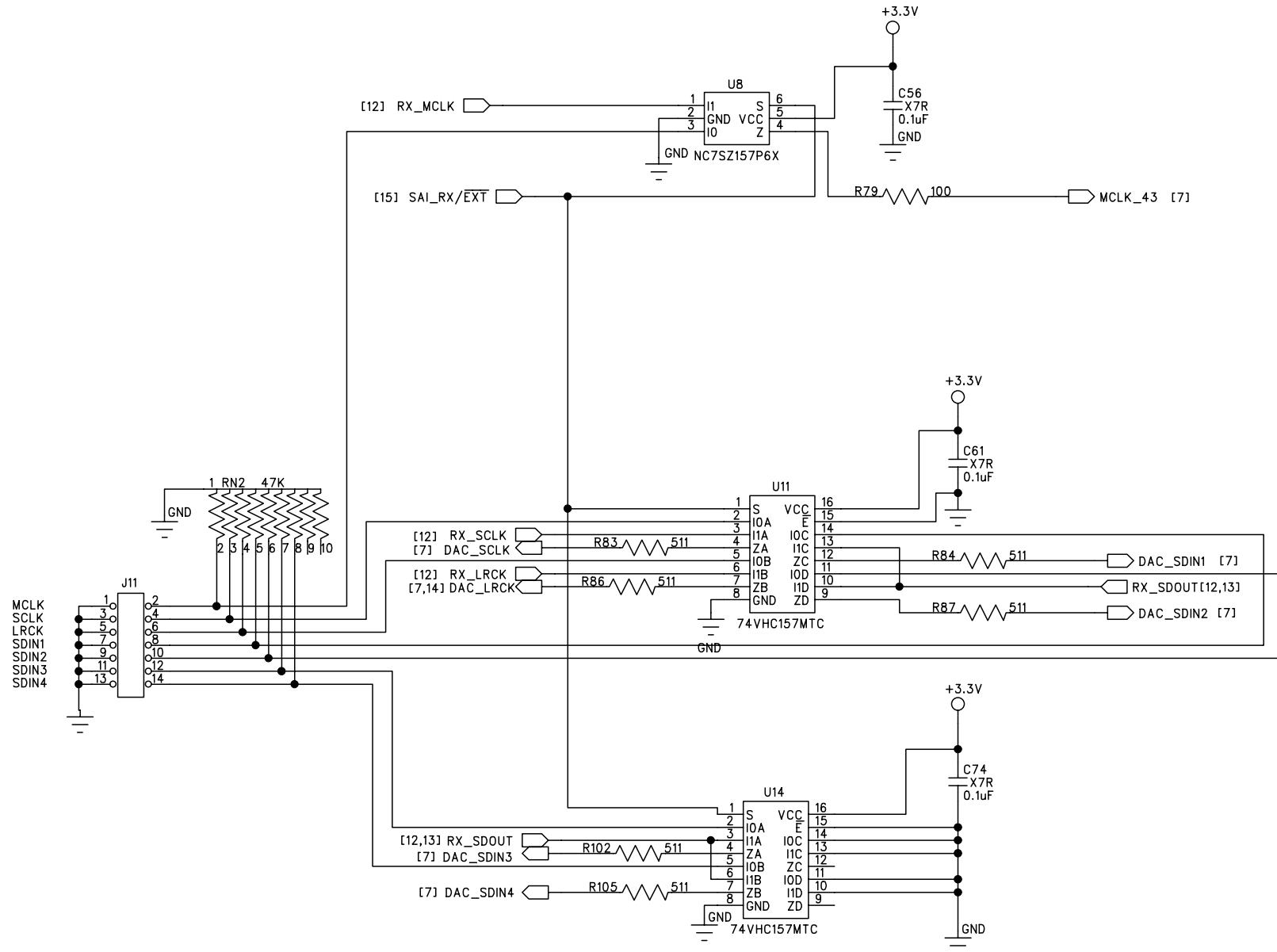


Figure 49. PCM Input Header and Multiplexing

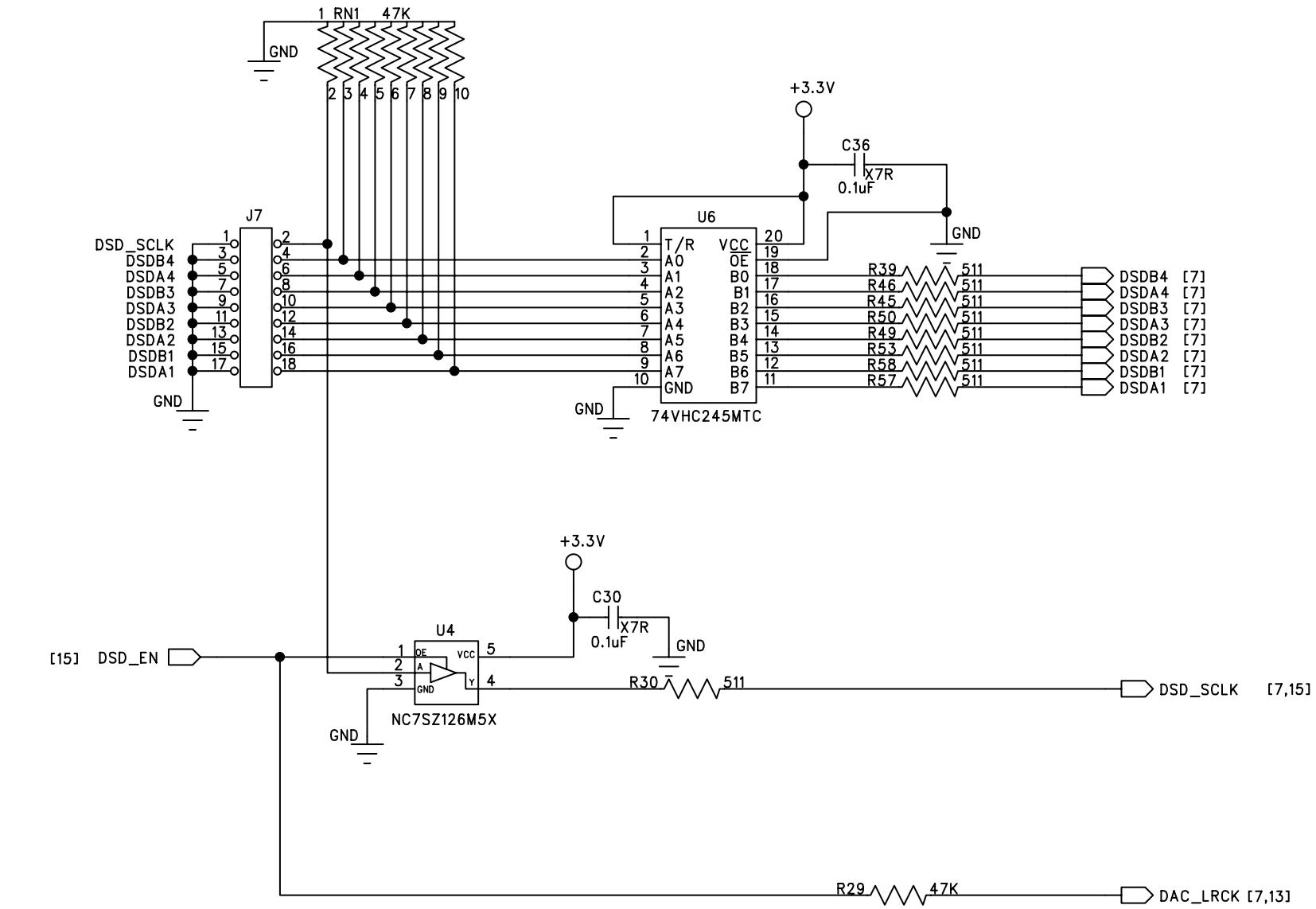


Figure 50. DSD Input Header

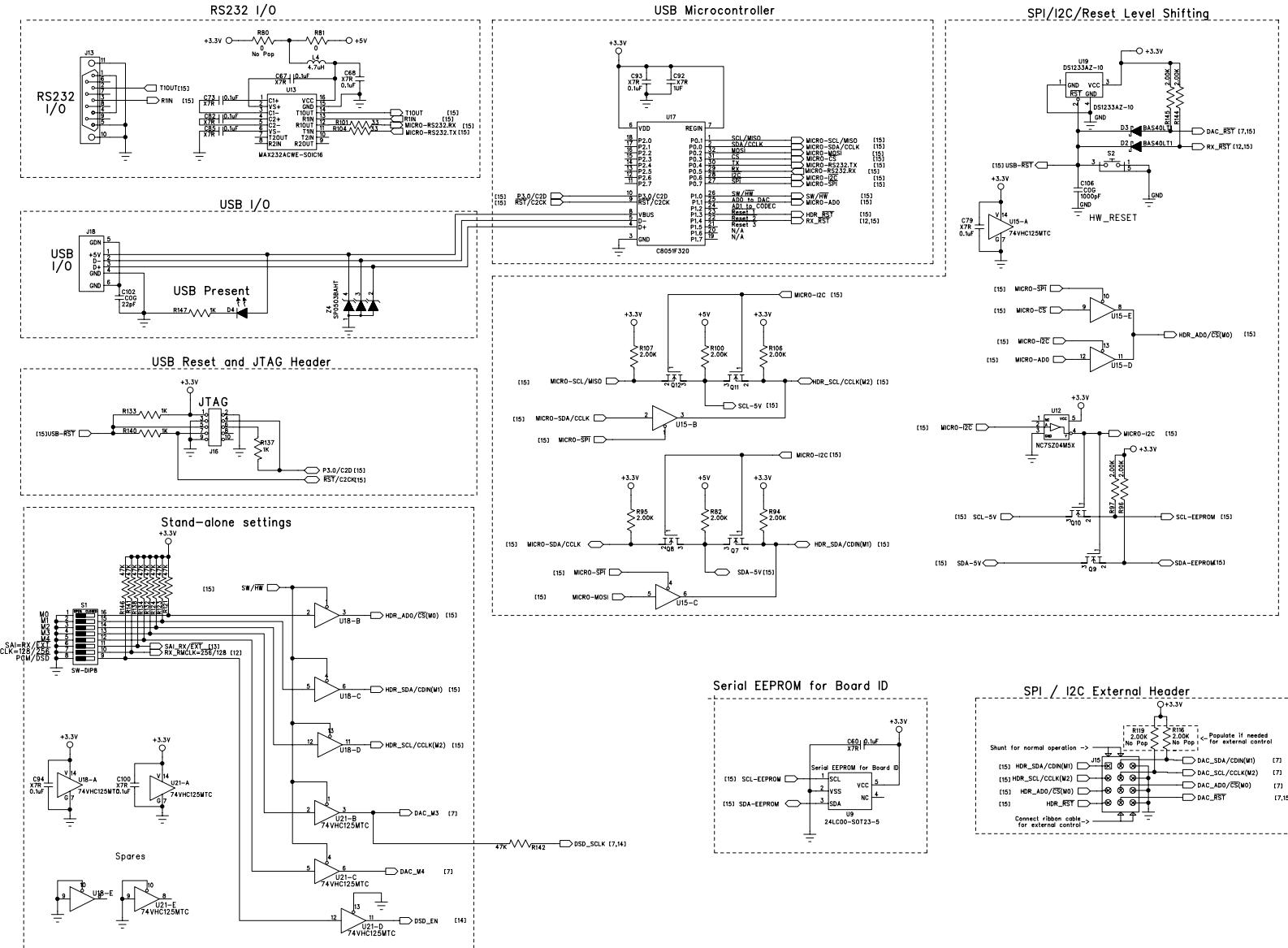


Figure 51. Control Input

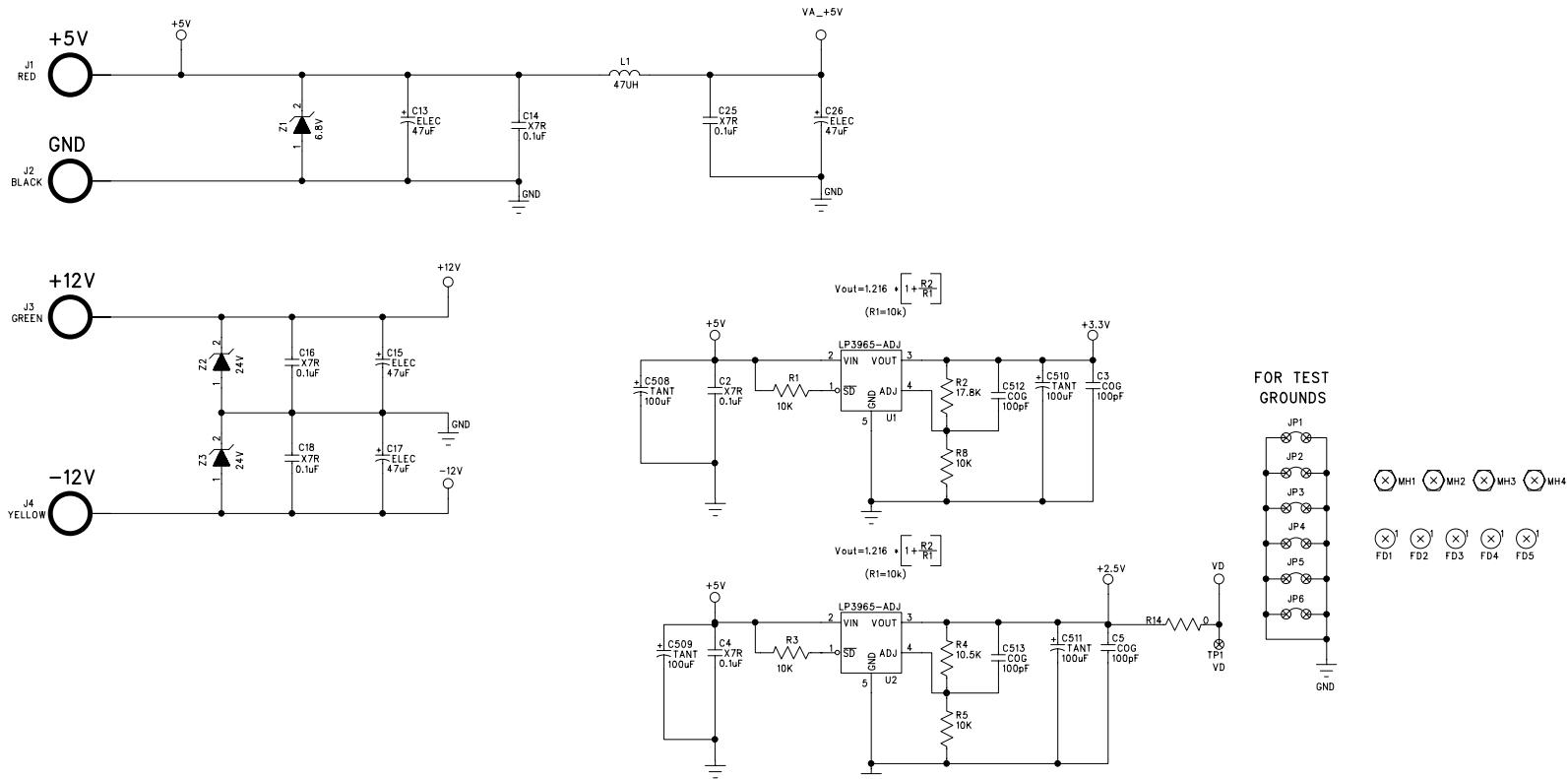
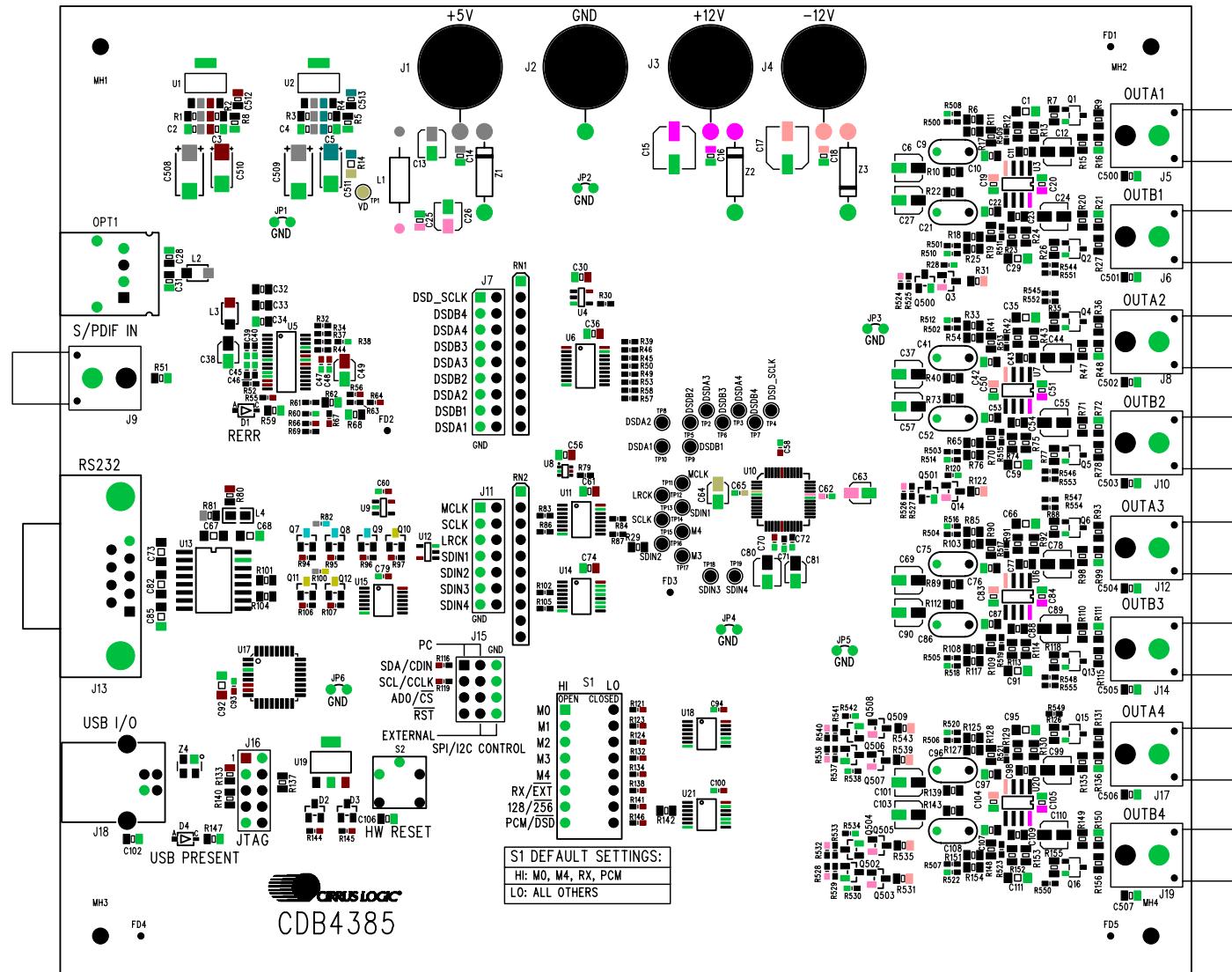


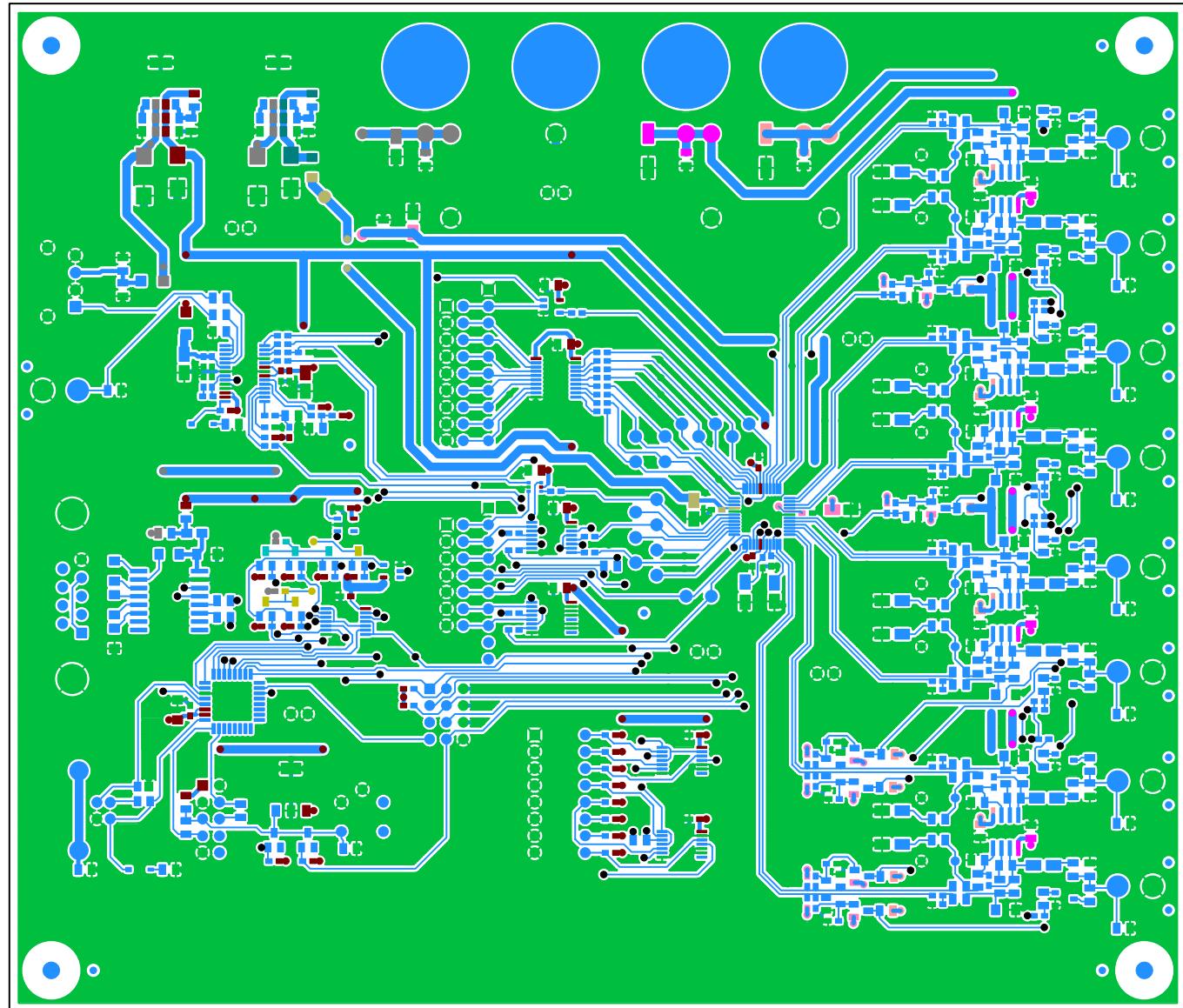
Figure 52. Power Input



CIRRUS LOGIC CDB4385 PCB 240-00123-01 Rev C

SILKSCREEN TOP

Figure 53. Silkscreen Top



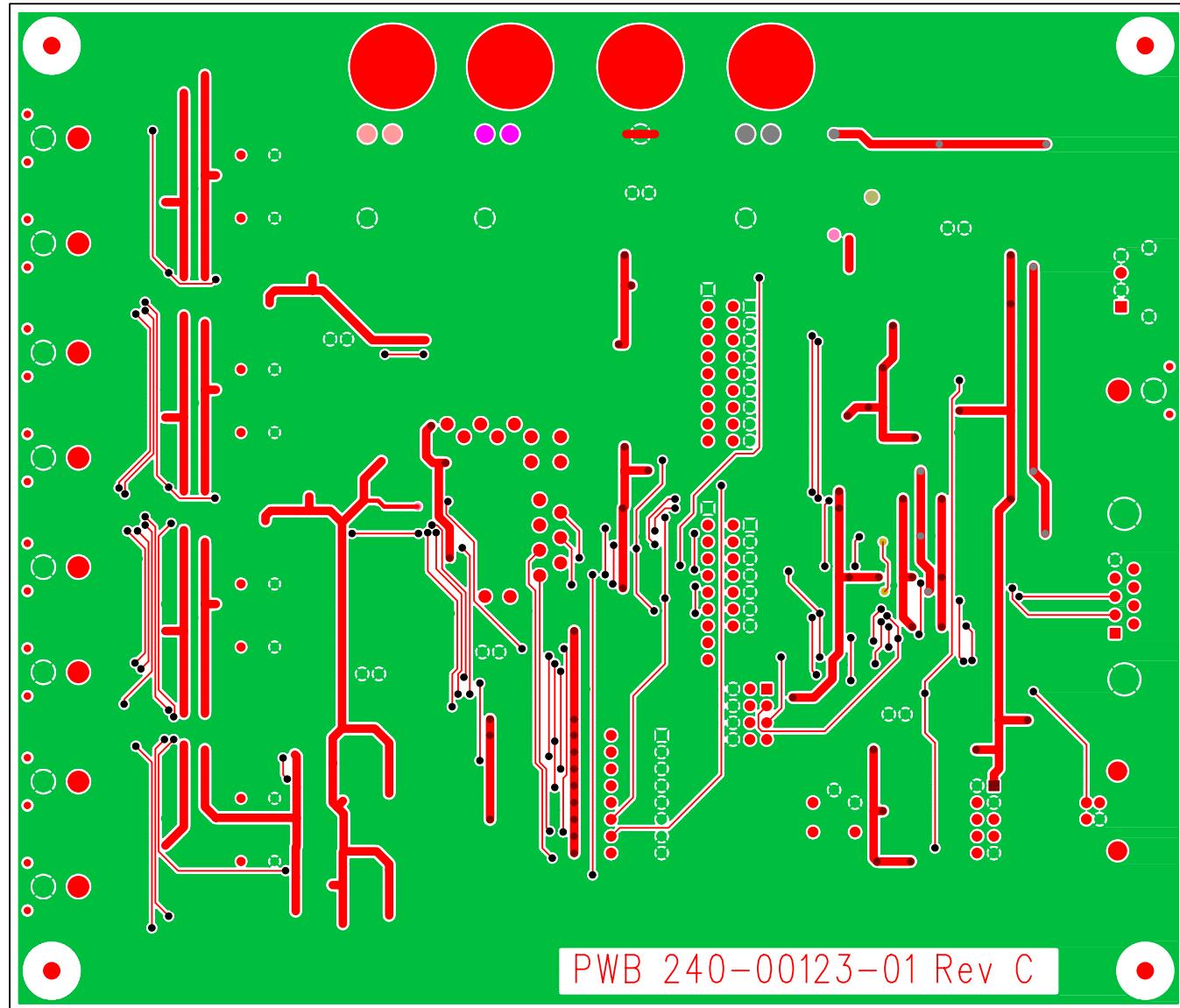
CIRRUS LOGIC

CDB4385

PCB 240-00123-01 Rev C

TOP SIDE

Figure 54. Top Side



CIRRUS LOGIC PCB J40-00123-01 Rev C

Figure 55. Bottom Side

10.REVISION HISTORY

Release	Changes
DB1	Initial Release
DB2	Added Performance Plots
DB3	Added USB support to Section 4. Input for Control Data

Contacting Cirrus Logic Support

For all product questions and inquiries, contact a Cirrus Logic Sales Representative.

To find the one nearest to you, go to www.cirrus.com.

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