

Clock Generator for Intel®Calistoga Chipset

Features

- **Compliant to Intel® CK410M**
- **Selectable CPU frequencies**
- **Differential CPU clock pairs**
- **100-MHz differential SRC clocks**
- **96-MHz differential dot clock**
- **27-MHz Spread and Non-spread video clock**
- **48-MHz USB clock**
- **SRC clocks independently stoppable through CLKREQ#[1:9]**
- **96/100-MHz spreadable differential video clock**
- **33-MHz PCI clocks**
- **Buffered Reference Clock 14.318MHz**
- **Low-voltage frequency select inputs**
- **I2C support with readback capabilities**
- **Ideal Lexmark Spread Spectrum profile for maximum electromagnetic interference (EMI) reduction**
- **3.3V power supply**
- **72-pin QFN package**

Pin Description

Pin Description (continued)

Frequency Select Pins (FSA, FSB, and FSC)

Host clock frequency selection is achieved by applying the appropriate logic levels to FSA, FSB, FSC inputs prior to VTT_PWRGD# assertion (as seen by the clock synthesizer). Upon VTT_PWRGD# being sampled LOW by the clock chip (indicating processor VTT voltage is stable), the clock chip samples the FSA, FSB, and FSC input values. For all logic levels of FSA, FSB, and FSC, VTT_PWRGD# employs a one-shot functionality in that once a valid LOW on VTT_PWRGD# has been sampled, all further VTT_PWRGD#, FSA, FSB, and FSC transitions will be ignored, except in test mode.

Serial Data Interface

To enhance the flexibility and function of the clock synthesizer, a two-signal serial interface is provided. Through the Serial Data Interface, various device functions, such as individual clock output buffers, can be individually enabled or disabled. The registers associated with the Serial Data Interface

Table 1. Frequency Select Table FSA, FSB, and FSC[\[1](#page-2-1)]

initialize to their default setting upon power-up, and therefore use of this interface is optional. Clock device register changes are normally made upon system initialization, if any are required. The interface cannot be used during system operation for power management functions.

Data Protocol

The clock driver serial protocol accepts byte write, byte read, block write, and block read operations from the controller. For block write/read operation, the bytes must be accessed in sequential order from lowest to highest byte (most significant bit first) with the ability to stop after any complete byte has been transferred. For byte write and byte read operations, the system controller can access individually indexed bytes. The offset of the indexed byte is encoded in the command code, as described in *[Table 2](#page-2-0)*.

The block write and block read protocol is outlined in *[Table 3](#page-3-0)* while *[Table 4](#page-3-1)* outlines the corresponding byte write and byte read protocol. The slave receiver address is 11010010 (D2h)

. **Table 2. Command Code Definition**

Note:

1. 27-MHz and 96-MHz can not be output at the same time.

Table 3. Block Read and Block Write Protocol

Table 4. Byte Read and Byte Write Protocol

Control Registers

Byte 0: Control Register 0

Byte 1: Control Register 1

Byte 2: Control Register 2

Byte 3: Control Register 3

Byte 4: Control Register 4

Byte 5: Control Register 5

Byte 5: Control Register 5 (continued)

Byte 6: Control Register 6

Byte 7: Vendor ID

Byte 8: Control Register 8

Byte 9: Control Register 9

Byte 10: Control Register 10

Byte 11: Control Register 11

Byte 12: Control Register 12

Byte 13: Control Register 13

Table 5. Crystal Recommendations

The CY28447 requires a Parallel Resonance Crystal. Substituting a series resonance crystal will cause the CY28447 to operate at the wrong frequency and violate the ppm specification. For most applications there is a 300-ppm frequency shift between series and parallel crystals due to incorrect loading.

Crystal Loading

Crystal loading plays a critical role in achieving low ppm performance. To realize low ppm performance, the total capacitance

the crystal will see must be considered to calculate the appropriate capacitive loading (CL).

[Figure 1](#page-9-0) shows a typical crystal configuration using the two trim capacitors. An important clarification for the following discussion is that the trim capacitors are in series with the crystal not parallel. It's a common misconception that load capacitors are in parallel with the crystal and should be approximately equal to the load capacitance of the crystal. This is not true.

Figure 1. Crystal Capacitive Clarification

Calculating Load Capacitors

In addition to the standard external trim capacitors, trace capacitance and pin capacitance must also be considered to correctly calculate crystal loading. As mentioned previously, the capacitance on each side of the crystal is in series with the crystal. This means the total capacitance on each side of the crystal must be twice the specified crystal load capacitance (CL). While the capacitance on each side of the crystal is in series with the crystal, trim capacitors (Ce1,Ce2) should be calculated to provide equal capacitive loading on both sides.

Figure 2. Crystal Loading Example

As mentioned previously, the capacitance on each side of the crystal is in series with the crystal. This means the total capacitance on each side of the crystal must be twice the specified load capacitance (CL). While the capacitance on each side of the crystal is in series with the crystal, trim capacitors

(Ce1,Ce2) should be calculated to provide equal capacitance loading on both sides.

Use the following formulas to calculate the trim capacitor values for Ce1 and Ce2.

Load Capacitance (each side)

$$
Ce = 2 \cdot CL - (Cs + Ci)
$$

Total Capacitance (as seen by the crystal)

$$
CLe = \frac{1}{\left(\frac{1}{Ce1 + Cs1 + Ci1} + \frac{1}{Ce2 + Cs2 + Ci2}\right)}
$$

CLK_REQ# Description

The CLKREQ# signals are active LOW inputs used for clean enabling and disabling selected SRC outputs. The outputs controlled by CLKREQ# are determined by the settings in register byte 8. The CLKREQ# signal is a de-bounced signal in that it's state must remain unchanged during two consecutive rising edges of SRCC to be recognized as a valid assertion or deassertion. (The assertion and deassertion of this signal is absolutely asynchronous.)

CLK_REQ[1:9]# Assertion (CLKREQ# -> LOW)

All differential outputs that were stopped are to resume normal operation in a glitch-free manner. The maximum latency from the assertion to active outputs is between 2 and 6 SRC clock periods (2 clocks are shown) with all SRC outputs resuming simultaneously. All stopped SRC outputs must be driven HIGH within 10 ns of CLKREQ# deassertion to a voltage greater than 200 mV.

CLK_REQ[1:9]# Deassertion (CLKREQ# -> HIGH)

The impact of deasserting the CLKREQ# pins is that all SRC outputs that are set in the control registers to stoppable via deassertion of CLKREQ# are to be stopped after their next transition. The final state of all stopped DIF signals is LOW, both SRCT clock and SRCC clock outputs will not be driven.

PD (Power-down) Clarification

The VTT PWRGD# /PD pin is a dual-function pin. During initial power-up, the pin functions as VTT_PWRGD#. Once VTT_PWRGD# has been sampled LOW by the clock chip, the pin assumes PD functionality. The PD pin is an asynchronous active HIGH input used to shut off all clocks cleanly prior to shutting off power to the device. This signal is synchronized internal to the device prior to powering down the clock synthesizer. PD is also an asynchronous input for powering up the system. When PD is asserted HIGH, all clocks need to be driven to a LOW value and held prior to turning off the VCOs and the crystal oscillator.

PD (Power-down) Assertion

When PD is sampled HIGH by two consecutive rising edges of CPUC, all single-ended outputs will be held LOW on their next HIGH-to-LOW transition and differential clocks must be held HIGH or tri-stated (depending on the state of the control register drive mode bit) on the next diff clock# HIGH-to-LOW transition within 4 clock periods. When the SMBus PD drive mode bit corresponding to the differential (CPU, SRC, and DOT) clock output of interest is programmed to '0', the clock outputs are held with "Diff clock" pin driven HIGH at 2 x Iref, and "Diff clock#" tri-state. If the control register PD drive mode

bit corresponding to the output of interest is programmed to "1", then both the "Diff clock" and the "Diff clock#" are tri-state. Note that *[Figure 4](#page-10-0)* shows CPUT = 133 MHz and PD drive mode = '1' for all differential outputs. This diagram and description is applicable to valid CPU frequencies 100, 133, 166, and 200 MHz. In the event that PD mode is desired as the initial power-on state, PD must be asserted HIGH in less than 10 μ s after asserting Vtt_PwrGd#. It should be noted that 96 100 SSC will follow the DOT waveform is selected for 96 MHz and the SRC waveform when in 100-MHz mode.

PD Deassertion

The power-up latency is less than 1.8 ms. This is the time from the deassertion of the PD pin or the ramping of the power supply until the time that stable clocks are output from the clock chip. All differential outputs stopped in a three-state condition resulting from power down will be driven high in less than 300 μ s of PD deassertion to a voltage greater than 200 mV. After the clock chip's internal PLL is powered up and locked, all outputs will be enabled within a few clock cycles of each other. *[Figure 5](#page-10-1)* is an example showing the relationship of clocks coming up. It should be noted that 96_100_SSC will follow the DOT waveform is selected for 96 MHz and the SRC waveform when in 100-MHz mode.

CPU_STP# Assertion

The CPU STP# signal is an active LOW input used for synchronous stopping and starting the CPU output clocks while the rest of the clock generator continues to function. When the CPU STP# pin is asserted, all CPU outputs that are set with the SMBus configuration to be stoppable via assertion of CPU_STP# will be stopped within two–six CPU clock periods after being sampled by two rising edges of the internal CPUC clock. The final states of the stopped CPU signals are $CPUT = HIGH$ and $CPUC = LOW$. There is no change to the output drive current values during the stopped state. The

CPUT is driven HIGH with a current value equal to 6 x (Iref), and the CPUC signal will be tri-stated.

CPU_STP# Deassertion

The deassertion of the CPU_STP# signal will cause all CPU outputs that were stopped to resume normal operation in a synchronous manner, synchronous manner meaning that no short or stretched clock pulses will be produce when the clock resumes. The maximum latency from the deassertion to active outputs is no more than two CPU clock cycles.

PCI_STP# Assertion

The PCI_STP# signal is an active LOW input used for synchronous stopping and starting the PCI outputs while the rest of the clock generator continues to function. The set-up time for capturing PCI_STP# going LOW is 10 ns (t_{SU}) . (See *[Figure 10](#page-12-0)*.) The PCIF clocks will not be affected by this pin if their corresponding control bit in the SMBus register is set to allow them to be free running.

Figure 10. PCI_STP# Assertion Waveform

PCI_STP# Deassertion

The deassertion of the PCI_STP# signal will cause all PCI and stoppable PCIF clocks to resume running in a synchronous manner within two PCI clock periods after PCI_STP# transitions to a HIGH level.

Figure 13. Single-ended Load Configuration

 $VTT_PWRGD# = toggle$

Absolute Maximum Conditions

Multiple Supplies: The Voltage on any input or I/O pin cannot exceed the power pin during power-up. Power supply sequencing is NOT required.

DC Electrical Specifications

AC Electrical Specifications

Note: 2. Measured with one REF on.

AC Electrical Specifications (continued)

AC Electrical Specifications (continued)

AC Electrical Specifications (continued)

Test and Measurement Set-up

For Single-ended Signals and Reference

The following diagram shows test load configurations for the single-ended PCI, USB, and REF output signals.

Figure 15.Single-ended Load Configuration Low Drive Option

Figure 16. Single-ended Load Configuration High Drive Option

The following diagram shows the test load configuration for the differential CPU and SRC outputs.

Ordering Information

Package Diagram

72-Lead QFN 10 x 10 mm (Punch Version) LF72A

Document History Page

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