LTC3644
Quad 17V, 1.25A Synchronous Step-Down Regulator with Ultralow Quiescent Current

DESCRIPTION

Demonstration circuit 2383A-A features the LTC®3644: the wide input and output voltage range, high efficiency and power density, quad 1.25A outputs DC/DC synchronous step-down monolithic regulator. The input voltage range of DC2383A-A is 2.7V to 17V. The default demo board setting of V_{OUT1}, V_{OUT2}, V_{OUT3}, V_{OUT4} is 1.2V, 3.3V, 2.5V and 1.8V at 1.25A maximum DC output current per channel. There are two assembly versions. The DC2383A-A features LTC3644 which operates at an internally fixed frequency of 1MHz (Typ), while the DC2383A-B features LTC3644-2 which operates at an internally fixed frequency of 2.25MHz (Typ). Peak current limit is internally fixed at 2.2A typical per channel. Each channel comes with independent run pin control and power good indicators. Phase shift selection of either 0 degree or 180 degrees between switch rising edge of channels 1, 2 and channels 3, 4 is also available.

DC2383A-A provides optional onboard 0Ω jumpers to configure the LTC3644 as 4-phase dual 2.5A/2.5A outputs or 4-phase triple 2.5A/1.25A/1.25A outputs. Optional 0Ω jumpers connecting V_{IN1} to V_{IN2}, V_{IN3}, V_{IN4} are available

for users to operate selected channels of LTC3644 at different input voltages than V_{IN1}.

A user-selectable MODE/SYNC input is provided to allow users to trade off ripple noise for light load efficiency: pulse-skipping mode (PS) or Burst Mode® operation delivers higher efficiency at light load while forced continuous conduction mode (FCM) is preferred for noise sensitive applications. The MODE/SYNC pin can also be used to synchronize the switching frequency to an external clock or set the phase shift between channels 1, 2 and channels 3, 4. Constant frequency, peak current mode control architecture and integrated internal control loop compensation network, allows very fast transient response to line and load changes while maintaining loop stability.
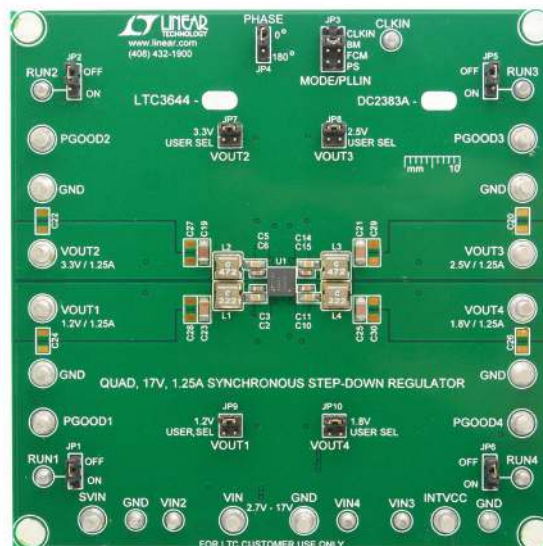
The LTC3644 is available in a thermally enhanced, low profile 36-lead 5mm × 5mm BGA package.

It is recommended to read the data sheet and demo manual of LTC3644 prior to using or making any changes to DC2383A-A.

Design files for this circuit board are available.

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BOARD PHOTO

PERFORMANCE SUMMARY

Specifications are at $T_A = 25°C$

PARAMETER	CONDITIONS	VALUE
Input Voltage Range V_{IN}		2.7V to 17V
Demo Board Default Output Voltages V_{OUT1}, V_{OUT2}, V_{OUT3}, V_{OUT4}	$f_{SW} = 1MHz$ $V_{IN} = 2.7V$ to 17V ($V_{OUT} < V_{IN}$) $I_{LOAD} = 0A$ to 1.25A per Channel	1.2V ±2% 3.3V ±2% 2.5V ±2% 1.8V ±2%
Default Switching Frequency	Internally Fixed Switching Frequency	1MHz ±18%
Maximum Continuous Output Current I_{OUT} per Channel I_{OUT1}, I_{OUT2}, I_{OUT3}, I_{OUT4}	$f_{SW} = 1MHz$ $V_{IN} = 2.7V$ to 17V ($V_{OUT} < V_{IN}$) $V_{OUT} = 1.2V$, 1.8V, 2.5V, 3.3V	1.25A
Efficiency	$V_{IN} = 12V$ $f_{SW} = 1MHz$ $V_{OUT1} = 1.2V$ at $I_{OUT1} = 1.25A$ $V_{OUT2} = 3.3V$ at $I_{OUT2} = 1.25A$ $V_{OUT3} = 2.5V$ at $I_{OUT3} = 1.25A$ $V_{OUT4} = 1.8V$ at $I_{OUT4} = 1.25A$	Channel 1: 79.1% Channel 2: 89.0% Channel 3: 86.7% Channel 4: 84.5% (Figure 4)
Thermal Performance (Peak Temperature)	$V_{IN} = 12V$ $f_{SW} = 1MHz$ $V_{OUT1} = 1.2V$, $V_{OUT2} = 3.3V$, $V_{OUT3} = 2.5V$, $V_{OUT4} = 1.8V$ $I_{OUT} = 1.25A$ per Channel $T_A = 25°C$, No Heatsink, No Forced Airflow	LTC3644: 56.3°C L1: 36.6°C L2: 38.7°C L3: 38.1°C L4: 37.4°C (Figure 6)
Dynamic Load Transient Response $V_{OUT1(P-P)}$ $V_{OUT2(P-P)}$ $V_{OUT3(P-P)}$ $V_{OUT4(P-P)}$	$V_{IN} = 12V$ $f_{SW} = 1MHz$, FCM $V_{OUT1} = 1.2V$, $I_{OUT1\_STEP} = 0.625A$ to 1.25A $V_{OUT2} = 3.3V$, $I_{OUT2\_STEP} = 0.625A$ to 1.25A $V_{OUT3} = 2.5V$, $I_{OUT3\_STEP} = 0.625A$ to 1.25A $V_{OUT4} = 1.8V$, $I_{OUT4\_STEP} = 0.625A$ to 1.25A COUT_ceramic = 1x47µF per Channel	$V_{OUT1(P-P)} = 128mV$ (Figure 7a) $V_{OUT2(P-P)} = 166mV$ (Figure 7d) $V_{OUT3(P-P)} = 141mV$ (Figure 7c) $V_{OUT4(P-P)} = 134mV$ (Figure 7b)

QUICK START PROCEDURE

Demonstration circuit 2383A-A is easy to set up to evaluate the performance of the LTC3644. Please refer to Figure 1 for proper measurement equipment setup and follow the test procedures below:

1. With power off, connect the input power supply between V_{IN} (E1) and GND (E7). V_{IN2}, V_{IN3} and V_{IN4} are tied to V_{IN1} (use onboard 0Ω jumpers R31, R29 and R30) by default.

2. Connect the first load between V_{OUT1} (E18) and GND (E20) for channel 1, connect the second load between V_{OUT2} (E14) and GND (E16) for channel 2, connect the third load between V_{OUT3} (E15) and GND (E17), connect the fourth load between V_{OUT4} (E19) and GND (E21). Preset all the loads to 0A.

3. Connect the DMMs between the input test points: V_{IN} (E1) and GND (E7) to monitor the input voltage. Connect DMMs between V_{OUT1} (E18) and GND (E20), V_{OUT2} (E14) and GND (E16), V_{OUT3} (E15) and GND (E17), V_{OUT4} (E19) and GND (E21) to monitor the corresponding DC output voltages of channel 1, channel 2, channel 3 and channel 4.

4. Turn on the power supply at the input. Measure and make sure the input supply voltage is 12V. Place the RUN 1 (JP1), RUN2 (JP2), RUN3 (JP5) and RUN4 (JP6) jumpers to the ON position. The output voltages should be 1.2V, 3.3V, 2.5V and 1.8V ±2% for V_{OUT1}, V_{OUT2}, V_{OUT3} and V_{OUT4}. Four onboard RUN1, RUN2, RUN3 and RUN4 jumpers allow users to enable or disable each channel independently for evaluation purpose. Users need to disable V_{OUT2} (3.3V) and V_{OUT3} (2.5V) when varying the input supply voltage down to 2.7V minimum.

5. Once the input and output voltages are properly established, adjusting the input voltage between 2.7V to 17V and the load current within the operating range of 0A to 1.25A max per channel. Observe the output voltage regulation, output voltage ripples, switching node waveform, load transient response and other parameters. Refer to Figure 2 for proper output voltage ripple measurement. Note 1: To measure the input/output voltage ripples properly, do not use the long ground lead on the oscilloscope probe. See Figure 2 for the proper scope probe technique. Short, stiff leads need to be soldered to the (+) and (–) terminals of an input or output capacitor. The probe's ground ring needs to touch the (–) lead and the probe tip needs to touch the (+) lead.

6. To program other output voltages for channel 1, channel 2, channel 3 and channel 4, put the RUN1 (JP1), RUN2 (JP2), RUN3 (JP5), RUN4 (JP6) jumpers to the OFF positions, move JP9, JP7, JP8, JP10 to the output voltage marking of USER SEL for each channel. Calculate and insert the bottom feedback resistors at R26, R20, R21 and R27 and repeat step 1 to step 5.

7. **(Option) Operation with Different Input Voltages:**

   Channel 2, channel 3 and channel 4 can operate with different input voltages than V_{IN1} and other channels' V_{IN}. DC2383A-A provides onboard 0Ω jumpers connecting the main V_{IN} (or V_{IN1}) to V_{IN2} (R31), V_{IN3} (R29) and V_{IN4} (R30) by default. Each of these 0Ω jumpers can be removed to disconnect V_{IN2}, V_{IN3}, V_{IN4} of selected channel from V_{IN1}. Different input voltages for channel 2, channel 3 and channel 4 should be applied between V_{IN2} (E23) and GND (E22), V_{IN3} (E24) and GND (E26), V_{IN4} (E25) and GND (E26) test points.

   Note: SV_{IN} is the input voltage to power the internal LDO and this pin is tied to V_{IN} (or V_{IN1}) by default. When operating with different channel input voltages, it is important to make sure V_{IN1} is on and $INTV_{CC}$ is present. SV_{IN} can also be disconnected from V_{IN1} (remove R7) and tied to an external voltage source at test point SV_{IN} (E2). SV_{IN} can be a different voltage than V_{IN1}, V_{IN2}, V_{IN3}, V_{IN4} and should be tied to the highest input supply voltage.

8. **(Option) Frequency Synchronization/Phase Selection:**

   The MODE/SYNC pin can be used to synchronize the internal oscillator clock frequency to the external clock signal. Place JP3 (MODE/PLLIN) at CLKIN position, apply an external clock signal at CLKIN test point (E10) to vary the switching frequency within ±50% of the internal programmed frequency.

Rev. 0

QUICK START PROCEDURE

The MODE/SYNC pin can also be used to set the phase shift between channels 1, 2 and channels 3, 4 while keeping the PHASE pin tied to INTV$_{CC}$. The phase shift can be set by modulating the duty cycle of an external clock on the MODE/SYNC pin. In this case, the phase shift will be determined by the applied external clock rising and falling edges. The switch rising edge of channels 1, 2 is synced to the rising edge of the external clock and switch rising edge of channels 3, 4 synced to the falling edge of the external clock. Crosstalk between channels can be avoided by adjusting the phase shift between channels such that the SW edges do not coincide.

9a. (Option) 4-Phase Dual 2.5A/2.5A Output Circuit Configuration:

DC2383A-A can be configured as dual 2.5A/2.5A outputs.

Channel 1 and channel 4 are master channels, channel 2 and channel 3 are slaves

The following simple modification is required:

1. Tie V$_{IN1}$, V$_{IN2}$, V$_{IN3}$, V$_{IN4}$ together or tie V$_{IN1}$ and V$_{IN2}$, V$_{IN3}$ and V$_{IN4}$ together if operating channel 1 and channel 2 at different input voltage than that of channel 3 and channel 4. Make sure SV$_{IN}$ is tied to the highest input supply voltage.

2. Tie SW1 and SW2, SW3 and SW4 together. Since SW1 and SW2, SW3 and SW4 are tied together, there is only one inductor needed for each output voltage rail. Calculate and insert the inductors needed for L1 and L4, remove L2 and L3.

3. Tie FB2 and FB3 to INTV$_{CC}$.

4. Float (do not use) PGOOD2 and PGOOD3. Only PGOOD1 and PGOOD4 are active.

5. Tie RUN1 and RUN2, RUN3 and RUN4 together. Note: Make sure to float all the unused onboard RUN pin jumpers to avoid accidently shorting V$_{IN}$ to GND.

6. Tie PHASE pin to INTV$_{CC}$.

   Refer to the demo board DC2383A-A schematic for more details.

9b. (Option) 4-Phase Triple 2.5A/1.25A/1.25A Output Circuit Configuration:

DC2383A-A can be configured as triple 2.5A/1.25A/1.25A outputs.

Channel 1 is master channel, channel 2 is slave. Channel 3 and channel 4 are independent channels.

The following simple modification is required:

1. Tie V$_{IN1}$, V$_{IN2}$, V$_{IN3}$, V$_{IN4}$ together or tie V$_{IN1}$ and V$_{IN2}$ together, V$_{IN3}$ and V$_{IN4}$ can be at different input voltages than V$_{IN1}$ and V$_{IN2}$. Make sure SV$_{IN}$ is tied to the highest input supply voltage.

2. Tie SW1 and SW2 together. There is only one inductor needed for this output voltage rail. Calculate and insert the inductor needed for L1 and remove L2.

3. Tie FB2 to INTV$_{CC}$.

4. Float (do not use) PGOOD2. Only PGOOD1, PGOOD3 and PGOOD4 are active.

5. Tie RUN1 and RUN2 together. Note: Make sure to float all the unused onboard RUN pin jumpers to avoid accidently shorting V$_{IN}$ to GND.

6. Tie PHASE pin to INTV$_{CC}$.

7. Channel 3 and channel 4 are left unchanged since these two channels operate as independent channels.

   Refer to the demo board DC2383A-A schematic for more details.
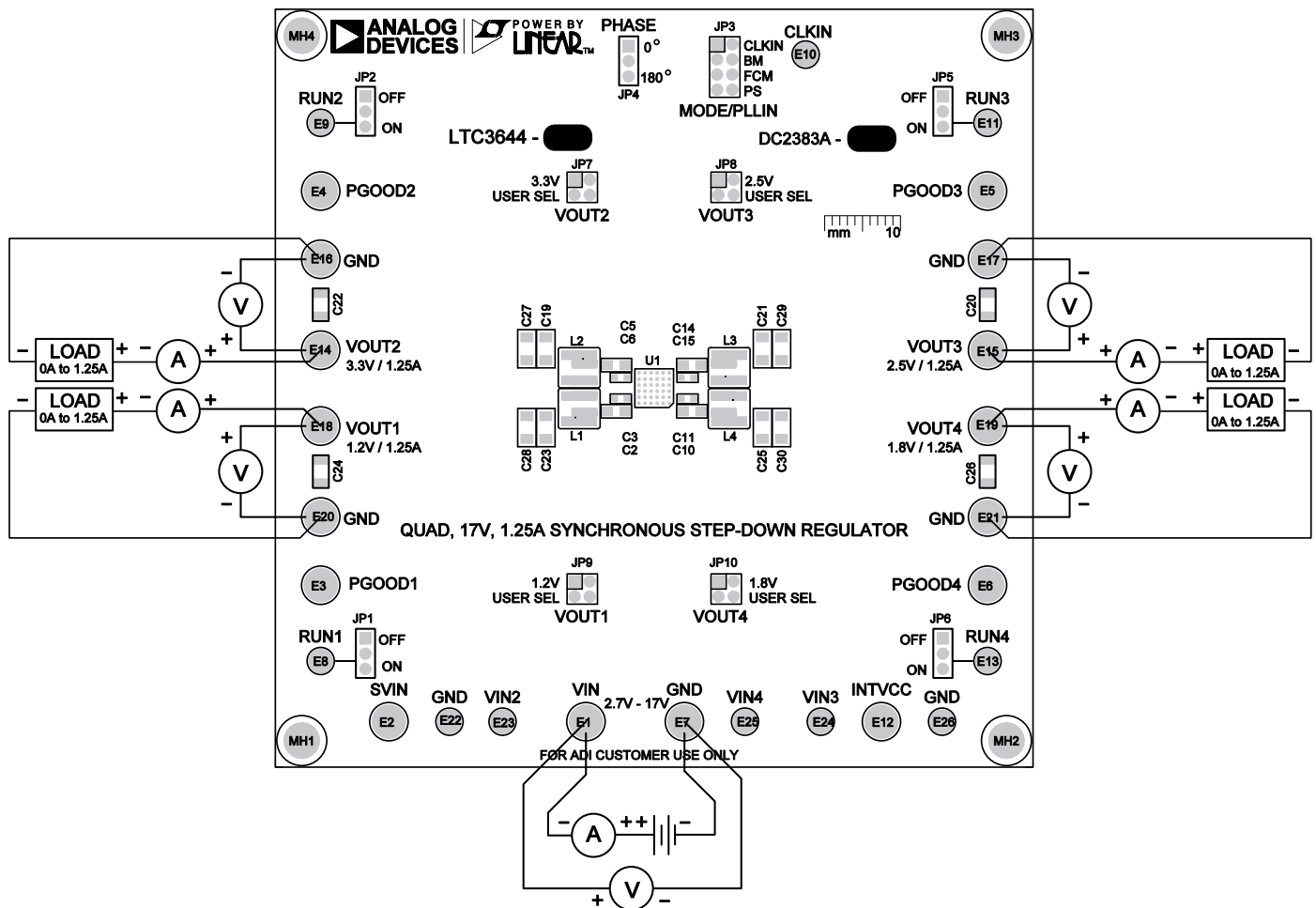
QUICK START PROCEDURE

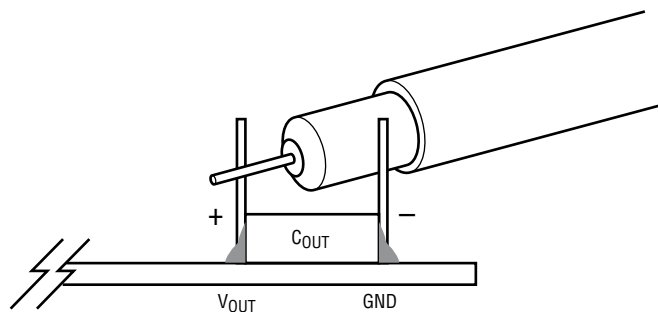**Figure 1. Proper Measurement Equipment Setup**

Figure 2. Scope Probe Placement for Measuring Input or Output Voltage Ripples
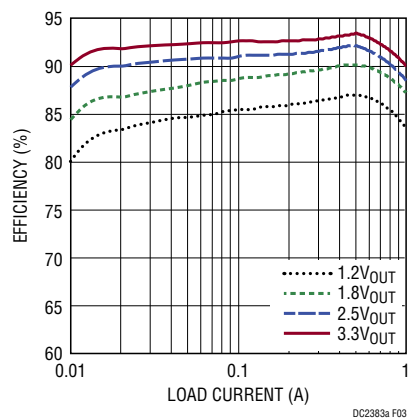
QUICK START PROCEDURE

Figure 3. Measured Efficiency at V_{IN} = 5V, f_{SW} = 1MHz, Burst Mode Operation
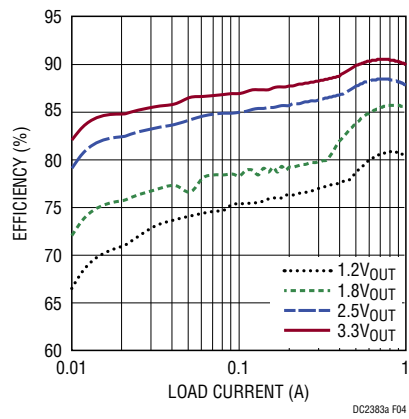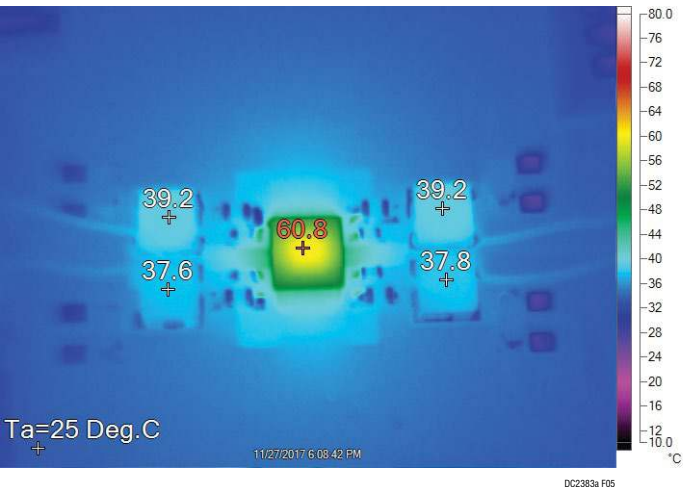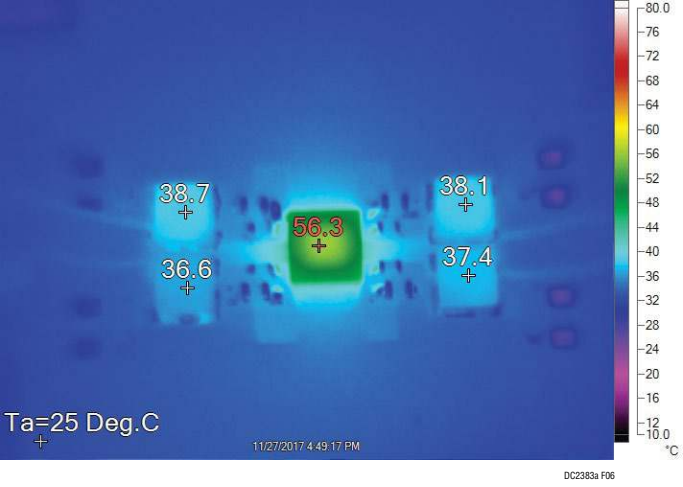

Figure 4. Measured Efficiency at V_{IN} = 12V, f_{SW} = 1MHz, Burst Mode Operation

V_{OUT1} = 1.2V, V_{OUT2} = 3.3V, V_{OUT3} = 2.5V, V_{OUT4} = 1.8V

I_{LOAD} = 1.25A PER CHANNEL
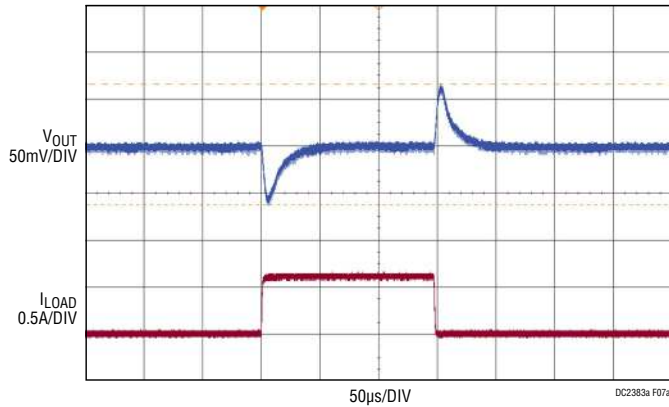
T_A = 25°C, NO HEAT SINK, NO FORCED AIRFLOW

Figure 5. Thermal Performance at V_{IN} = 5V, f_{SW} = 1MHz
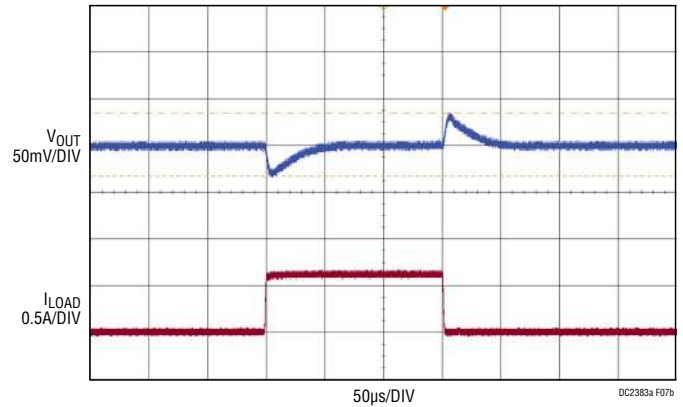

V_{OUT1} = 1.2V, V_{OUT2} = 3.3V, V_{OUT3} = 2.5V, V_{OUT4} = 1.8V

I_{LOAD} = 1.25A PER CHANNEL

T_A = 25°C, NO HEAT SINK, NO FORCED AIRFLOW

Figure 6. Thermal Performance at V_{IN} = 12V, f_{SW} = 1MHz

QUICK START PROCEDURE

(a) Load Transient Response: V_{OUT} = 1.2V

(b) Load Transient Response: V_{OUT} = 1.8V
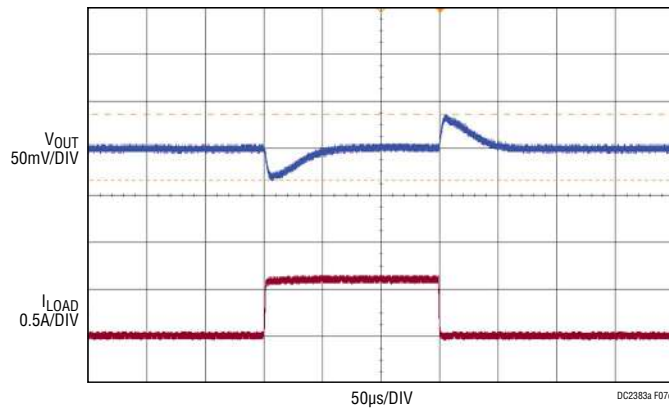
(c) Load Transient Response: V_{OUT} = 2.5V
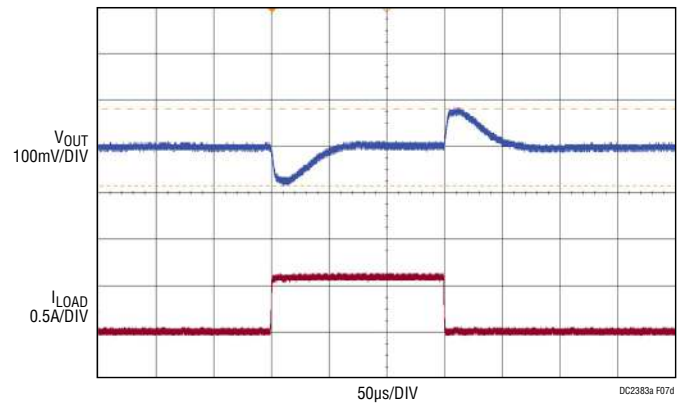
(d) Load Transient Response: V_{OUT} = 3.3V

Figure 7. Load Transient Responses

Load Transient Response Test Conditions:
V_{IN} = 12V, f_{SW} = 1MHz Typical
V_{OUT1} = 1.2V, V_{OUT2} = 3.3V
V_{OUT3} = 2.5V, V_{OUT4} = 1.8V
L1 = L4 = 2.2µH
L2 = L3 = 4.7µH
Load Step = 0.625A to 1.25A at di/dt = 0.625A/µs
COUT_ceramic = 1x47µF/16V/X5R/1206 (per Channel)
Feedforward Capacitor: CFF = 22pF (per Channel)

PARTS LIST

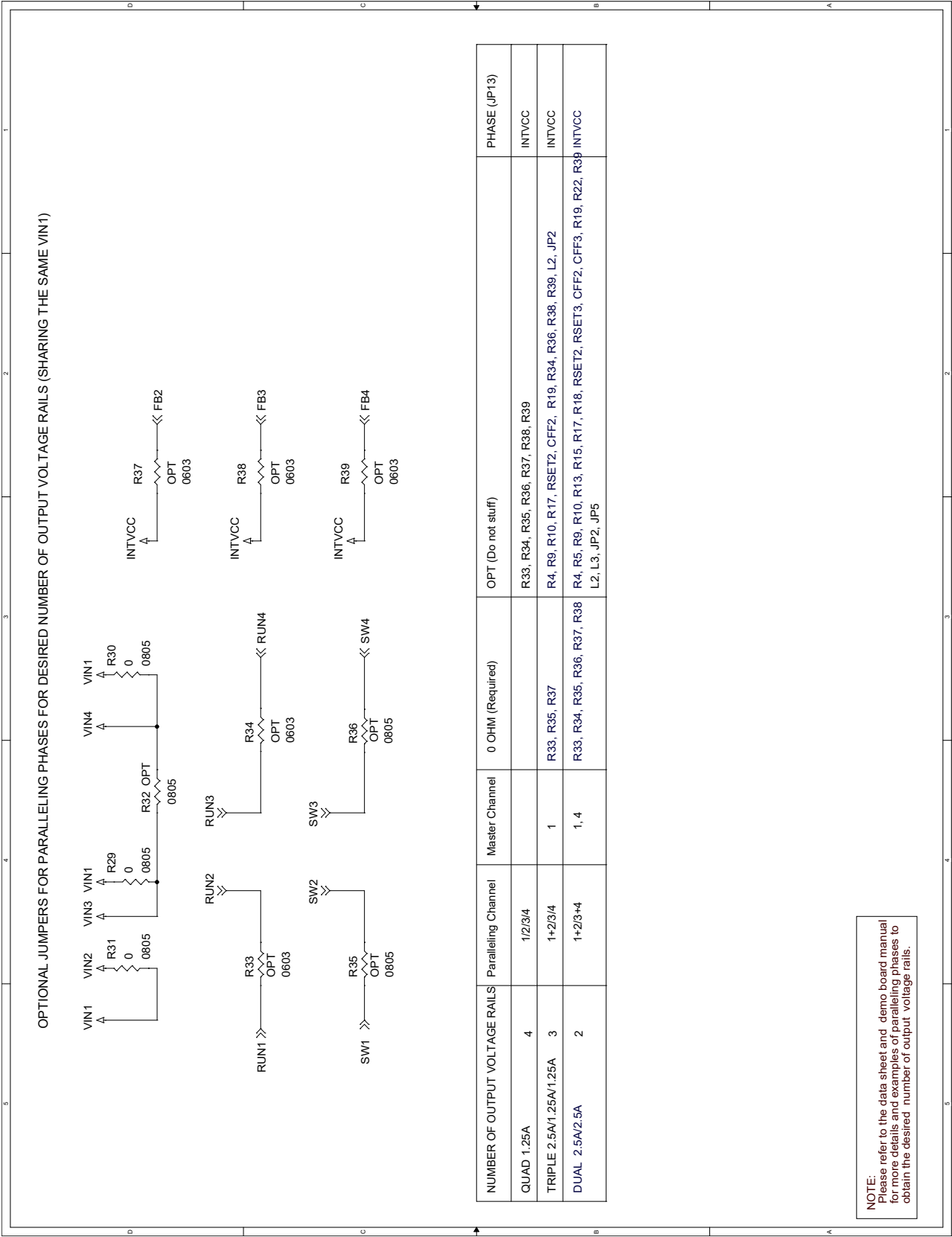| ITEM	QTY	REFERENCE	PART DESCRIPTION	MANUFACTURER/PART NUMBER
Required Circuit Components				
1	4	CFF1, CFF2, CFF3, CFF4	CAP, 0402 22pF 5% 50V COG	MURATA GRM1555C1H220JA01D
2	1	C1	CAP, 2917 22µF 20% 35V TANT	AVX TPSE226M035R0125
3	7	C2, C4, C5, C7, C9, C10, C14	CAP, 0805 10µF 10% 25V X5R	MURATA GRM21BR61E106KA73L
4	5	C3, C6, C11, C12, C15	CAP, 0603 4.7µF 20% 25V X5R	MURATA GRM188R61E475ME11D
5	1	C16	CAP, 0603 1µF 10% 25V X7R	MURATA GRM188R71E105KA12D
6	4	C19, C21, C23, C25	CAP, 1206 47µF 10% 16V X5R	MURATA GRM31CR61C476ME44L
7	4	C20, C22, C24, C26	CAP, 1206 10µF 10% 16V X7R	TDK C3216X7R1C106K160AC
8	2	L1, L4	IND, 2.2µH	COILCRAFT XFL4020-222ME
9	2	L2, L3	IND, 4.7µH	COILCRAFT XFL4020-472ME
10	5	RSET1, RSET2, RSET3, RSET4, R25	RES, 0402 619kΩ 1% 1/16W	VISHAY CRCW0402619KFKED
11	9	R1, R2, R9, R13, R14, R17, R18, R23, R24	RES, 0402 0Ω JUMPER	VISHAY CRCW04020000Z0ED
12	4	R3, R4, R5, R6	RES, 0402 100kΩ 1% 1/16W	VISHAY CRCW0402100KFKED
13	1	R7	RES, 0402 2.2Ω 1% 1/16W	VISHAY CRCW04022R20FNED
14	6	R8, R10, R11, R12, R15, R16	RES, 0402 10MΩ 1% 1/16W	VISHAY CRCW040210M0FKED
15	1	R19	RES, 0402 137kΩ 1% 1/16W	VISHAY CRCW0402137KFKED
16	1	R22	RES, 0402 196kΩ 1% 1/16W	VISHAY CRCW0402196KFKED
17	1	R28	RES, 0402 309kΩ 1% 1/16W	VISHAY CRCW0402309KFKED
18	3	R29, R30, R31	RES, 0805 0Ω JUMPER	VISHAY CRCW08050000Z0EA
19	1	U1	IC, QUAD 17V, 1A SYNCHRONOUS STEP-DOWN REGULATOR	ANALOG DEVICES LTC3644BGA
Additional Demo Board Circuit Components				
1	0	C8, C13, C17, C18	CAP, 0402 OPTION	OPTION
2	0	C27, C28, C29, C30	CAP, 1206 OPTION	OPTION
3	0	R20, R21, R26, R27	RES, 0402 OPTION	OPTION
4	0	R33, R34, R37, R38, R39	RES, 0603 OPTION	OPTION
5	0	R32, R35, R36	RES, 0805 OPTION	OPTION
Hardware: For Demo Board Only				
1	16	E1 TO E7, E12, E14 TO E21	TURRET	MILL-MAX 2501-2-00-80-00-00-07-0
2	10	E8 TO E11, E13, E22 TO E26	TURRET	MILL-MAX 2308-2-00-80-00-00-07-0
3	5	JP1, JP2, JP4, JP5, JP6	HEADER, 3PIN, 2mm	WURTH ELEKTRONIK 62000311121
4	1	JP3	HEADER, 2x4PINS, 2mm	WURTH ELEKTRONIK 62000821121
5	4	JP7 TO JP10	HEADER, 2PIN, DBL ROW 2mm	WURTH ELEKTRONIK 62000421121
6	4	MH1 TO MH4	STANDOFF, SNAP ON 12.7mm	WURTH ELEKTRONIK 702935000
7	10	XJP1 TO XJP10	SHUNT, 2mm	WURTH ELEKTRONIK 60800213421

SCHEMATIC DIAGRAM

OPTIONAL JUMPERS FOR PARALLELING PHASES FOR DESIRED NUMBER OF OUTPUT VOLTAGE RAILS (SHARING THE SAME VIN1)

NUMBER OF OUTPUT VOLTAGE RAILS	Paralleling Channel	Master Channel	0 OHM (Required)	OPT (Do not stuff)	PHASE (JP13)
QUAD 1.25A	4	1/2/3/4			R33, R34, R35, R36, R37, R38, R39	
TRIPLE 2.5A/1.25A/1.25A	3	1+2/3/4	1	R33, R35, R37	R4, R9, R10, R17, RSET2, CFF2, R19, R34, R36, R38, R39, L2, JP2	INTVCC
DUAL 2.5A/2.5A	2	1+2/3+4	1, 4	R33, R34, R35, R36, R37, R38	R4, R5, R9, R10, R13, R15, R17, R18, RSET2, RSET3, CFF2, CFF3, R19, R22, R39, L2, L3, JP2, JP5	INTVCC

NOTE:
Please refer to the data sheet and demo board manual for more details and examples of paralleling phases to obtain the desired number of output voltage rails.

SCHEMATIC DIAGRAM

SCHEMATIC DIAGRAM

USED TO MANUFACTURE PCB

MH1 MH2 MH3 MH4

REPRESENTS STAND OFFS

VOUT4 XL4 SW4 XL8 [3] [3]

VOUT3 XL2 SW3 XL7 [3] [3]

VOUT2 XL1 SW2 XL6 [3] [3]

VOUT1 XL3 SW1 XL5 [3] [3]
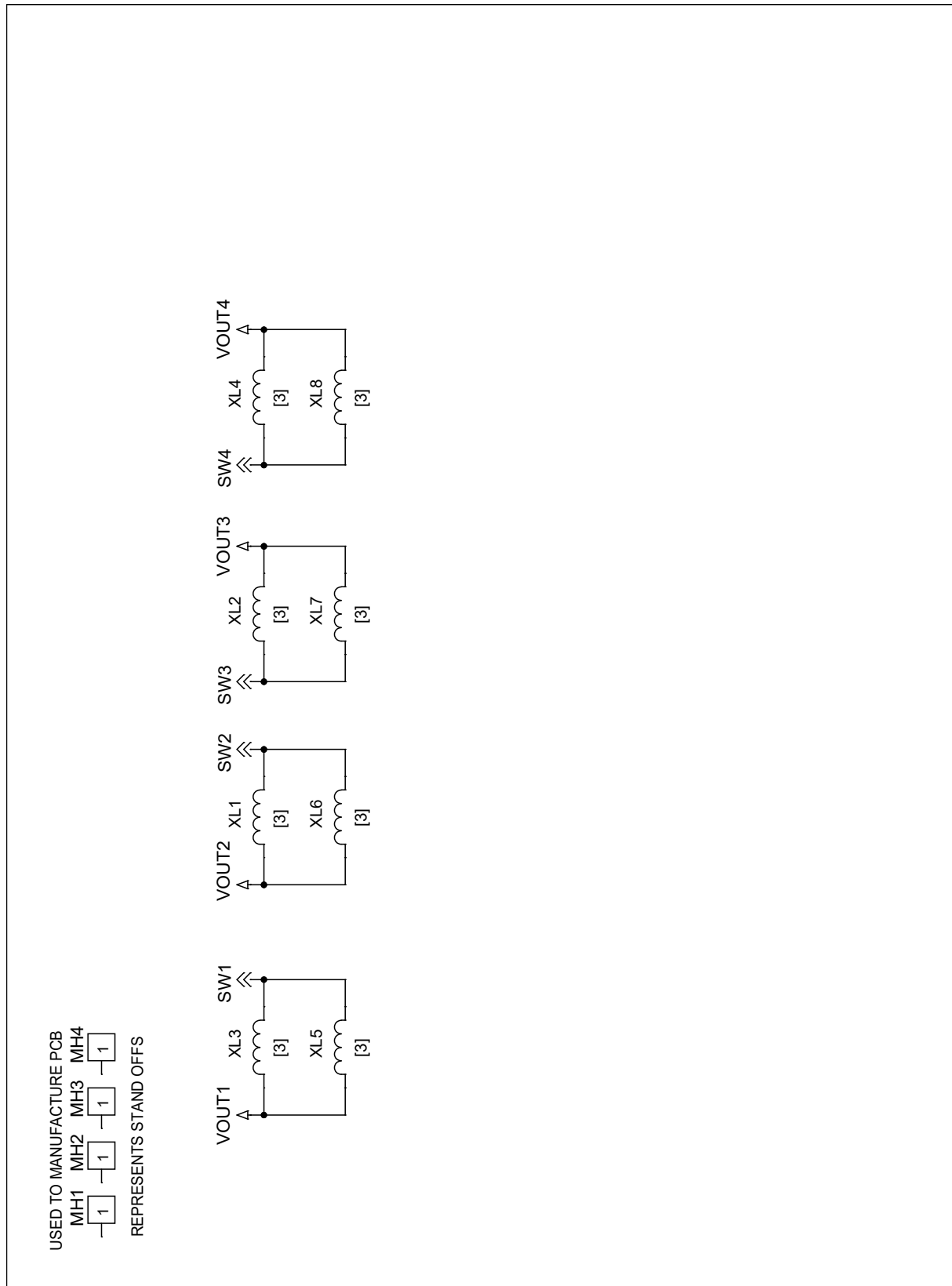
ESD Caution

ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

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