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SN74VMEH22501A-EP 8-Bit Universal Bus Transceiver and Two 1-Bit Bus Transceivers With Split LVTTL Port, Feedback Path, and 3-State Outputs

Technical [Documents](http://www.ti.com/product/SN74VMEH22501A-EP?dcmp=dsproject&hqs=td&#doctype2)

- - One Assembly/Test Site, One Fabrication Site Telecommunications
- Enhanced Diminishing Manufacturing Sources Instrumentation Systems (DMS) Support
- • Enhanced Product-Change Notification **3 Description**
-
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-
-
- Compliant With VME64, 2eVME, and 2eSST Protocols Validated at $T_A = -40^{\circ}$ C to 85°C
- **Device Information[\(3\)](#page-0-0)** Bus Transceiver Split LVTTL Port Provides a **Feedback Path for Control and Diagnostics**
- I/O Interfaces are 5-V Tolerant
-
-
-
-
- 26-Ω Equivalent Series Resistor on 3A Ports and
- Flow-Through Architecture Facilitates Printed
- Distributed V_{CC} and GND Pins Minimize High-
Speed Switching Noise
- Latch-Up Performance Exceeds 100 mA Per JESD 78, Class II
- **ESD Protection Exceeds JESD 22**
	- 2000-V Human-Body Model (A114-A)
	- 200-V Machine Model (A115-A)
	- 1000-V Charged-Device Model (C101)

1 Features 2 Applications

Tools & **[Software](http://www.ti.com/product/SN74VMEH22501A-EP?dcmp=dsproject&hqs=sw&#desKit)**

- ¹ Controlled Baseline Industrial Controls
	-
	-

Qualification Pedigree⁽¹⁾
Meetheen Charles Tree and Meetheen Meetheen as the transceiver has two integral 1-bit three-wire bus Member of the Texas Instruments Widebus™ transceivers and is designed for 3.3-V V_{CC} operation
Family transceivers and is designed for 3.3-V V_{CC} operations and is designed for 3.3-V V_{CC} operations allows Family
UBT™ Transceiver Combines D-Type Latches in the S-V tolerant inputs. The UBT transceiver allows
Transparent, latched, and flip-flop modes of data transparent, latched, and flip-flop modes of data and D-Type Flip-Flops for Operation in transfer, and the separate LVTTL input and outputs Transparent, Latched, or Clocked Modes **on the bus transceivers provide a feedback** path for

control and diagnostics monitoring. This device OECTM Circuitry Improves Signal Integrity and

Reduces Electromagnetic Interference (EMI)

Control of a high-speed interface between cards

operating at LVTTL logic levels and VME64, VME64x,

Compliant With VME64, 2eVME, a

- B-Port Outputs (-48 mA/64 mA) (1) Component qualification in accordance with JEDEC and industry standards to ensure reliable operation over an • Y and A-Port Outputs (–12 mA/12 mA) extended temperature range. This includes, but is not limited l_{off}, Power-Up 3-State, and BIAS V_{CC} Support Live
Insertion intermetallic life, and mold compound
Insertion electromigration, bond intermetallic life, and mold compound life. Such qualification testing should not be viewed as

26-Ω Equivalent Series Resistor on 3A Ports and

26-Ω Equivalent Series Resistor on 3A Ports and performance and environmental limits.
- Y Outputs

Clou Through Architecture Fecilitates Printed (2) VME320 is a patented backplane construction by Arizona

Digital, Inc.
- Circuit Board Layout
Circuit Board Layout the end of the data sheet.

Logic Diagram (Positive Logic)

Pin numbers shown for DGG and DGV

Table of Contents

4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

5 Description (continued)

The SN74VMEH22501A-EP device is pin-for-pin compatible to the SN74VMEH22501 device ([SCES357\)](http://www.ti.com/lit/pdf/sces357), but operates at a wider operating temperature range.

High-speed backplane operation is a direct result of the improved OEC circuitry and high drive that has been designed and tested into the VME64x backplane model. The B-port I/Os are optimized for driving large capacitive loads and include pseudo-ETL input thresholds $(\frac{1}{2}$ V_{CC} \pm 50 mV) for increased noise immunity. These specifications support the 2eVME protocols in VME64x (ANSI/VITA 1.1) and 2eSST protocols in VITA 1.5.

With proper design of a 21-slot VME system, a designer can achieve 320-MB transfer rates on linear backplanes and, possibly, 1-GB transfer rates on the VME320 backplane.

All inputs and outputs are 5-V tolerant and are compatible with TTL and 5-V CMOS inputs.

Active bus-hold circuitry holds unused or undriven 3A-port inputs at a valid logic state. Bus-hold circuitry is not provided on 1A or 2A inputs, any B-port input, or any control input. Use of pullup or pulldown resistors with the bus-hold circuitry is not recommended.

This device is fully specified for live-insertion applications using I_{off} , power-up 3-state, and BIAS V_{CC}. The I_{off} circuitry prevents damaging current to backflow through the device when it is powered off/on. The power-up 3 state circuitry places the outputs in the high-impedance state during power up and power down, which prevents driver conflict. The BIAS V_{CC} circuitry precharges and preconditions the B-port input/output connections, preventing disturbance of active data on the backplane during card insertion or removal, and permits true liveinsertion capability.

When V_{CC} is between 0 and 1.5 V, the device is in the high-impedance state during power up or power down. However, to ensure the high-impedance state above 1.5 V, output-enable (OE and OEBY) inputs should be tied to V_{CC} through a pullup resistor and output-enable (OEAB) inputs should be tied to GND through a pulldown resistor; the minimum value of the resistor is determined by the drive capability of the device connected to this input.

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6 Pin Configuration and Functions

Pin Functions

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Pin Functions (continued)

7 Specifications

7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) The input and output negative-voltage ratings may be exceeded if the input and output clamp-current ratings are observed.

7.2 ESD Ratings

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

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7.3 Recommended Operating Conditions

see (1)(2)

(1) All unused control inputs of the device must be held at V_{CC} or GND to ensure proper device operation. Refer to the TI application report, *Implications of Slow or Floating CMOS Inputs*, [SCBA004.](http://www.ti.com/lit/pdf/SCBA004)

(2) Proper connection sequence for use of the B-port I/O precharge feature is GND and BIAS V_{CC} = 3.3 V first, I/O second, and V_{CC} = 3.3 V last, because the BIAS V_{CC} precharge circuitry is disabled when any V_{CC} pin is connected. The control inputs can be connected at any
time, but normally are connected during the I/O stage. If B-port precharge is not req generally, GND is connected first.

7.4 Thermal Information

(1) For more information about traditional and new thermal metrics, see the *Semiconductor and IC Package Thermal Metrics* application report, [SPRA953.](http://www.ti.com/lit/pdf/spra953)

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7.5 Electrical Characteristics

over recommended operating free-air temperature range for A and B ports (unless otherwise noted)

(1) All typical values are at $V_{CC} = 3.3 V$, $T_A = 25^{\circ}C$.

(2) For I/O ports, the parameters I_{OZH} and I_{OZL} include the input leakage current.

(3) The bus-hold circuit can sink at least the minimum low sustaining current at V_{IL} max. I_{BHL} should be measured after lowering V_{IN} to GND, then raising it to V_{IL} max.

(4) The bus-hold circuit can source at least the minimum high sustaining current at V_{IH} min. I_{BHH} should be measured after raising V_{IN} to V_{CC} , then lowering it to V_{IH} min.

(5) An external driver must source at least I_{BHLO} to switch this node from low to high.

(6) An external driver must sink at least I_{BHHO} to switch this node from high to low.
(7) High-impedance state during power up or power down

High-impedance state during power up or power down

Electrical Characteristics (continued)

over recommended operating free-air temperature range for A and B ports (unless otherwise noted)

(8) This is the increase in supply current for each input that is at the specified TTL voltage level, rather than V_{CC} or GND.

7.6 Live-Insertion Specifications

over recommended operating free-air temperature range for B port

(1) All typical values are at $V_{\text{CC}} = 3.3$ V, $T_A = 25^{\circ}\text{C}$

(2) V_{CC} – 0.5 V < BIAS V_{CC}

7.7 Timing Requirements for UBT Transceiver (I Version)

over recommended operating conditions (unless otherwise noted) (see [Figure 7](#page-20-0) and [Figure 8](#page-21-0)); $T_A = -40^{\circ}$ C to 85°C

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7.8 Switching Characteristics for Bus Transceiver Function (I Version)

over recommended operating conditions (unless otherwise noted) (see [Figure 7](#page-20-0) and [Figure 8](#page-21-0)); $T_A = -40^{\circ}$ C to 85°C

7.9 Switching Characteristics for Bus Transceiver Function (M Version)

over recommended operating conditions (unless otherwise noted) (see [Figure 7](#page-20-0) and [Figure 8](#page-21-0)); $T_A = -55^{\circ}$ C to 125°C

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7.10 Switching Characteristics for UBT Transceiver (I Version)

over recommended operating conditions (unless otherwise noted) (see [Figure 7](#page-20-0) and [Figure 8](#page-21-0)); $T_A = -40^{\circ}$ C to 85°C

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7.11 Switching Characteristics for UBT Transceiver (M Version)

over recommended operating conditions (unless otherwise noted) (see [Figure 7](#page-20-0) and [Figure 8](#page-21-0)); $T_A = -55^{\circ}$ C to 125°C

7.12 Switching Characteristics for Bus Transceiver Function (I Version)

driver in slot 11, with receiver cards in all other slots (full load); over recommended operating conditions (unless otherwise noted) (see [Figure 6\)](#page-19-2); $T_A = -40^{\circ}C$ to 85°C

(1) All typical values are at V_{CC} = 3.3 V, T_A = 25°C. All values are derived from TI-SPICE models.

(2) All t_r and t_f times are taken at the first receiver.

7.13 Switching Characteristics for UBT (I Version)

driver in slot 11, with receiver cards in all other slots (full load); over recommended operating conditions (unless otherwise noted) (see [Figure 6\)](#page-19-2); $T_A = -40^{\circ}$ C to 85°C

(1) All typical values are at $V_{CG} = 3.3 V$, $T_A = 25°C$. All values are derived from TI-SPICE models.

(2) All t_r and t_f times are taken at the first receiver.

7.14 Switching Characteristics for Bus Transceiver Function (I Version)

driver in slot 1, with one receiver in slot 21 (minimum load); over recommended operating conditions (unless otherwise noted) (see [Figure 6](#page-19-2)); $T_A = -40^{\circ}$ C to 85°C

(1) All typical values are at $V_{CC} = 3.3 V$, $T_A = 25^{\circ}C$. All values are derived from TI-SPICE models.

(2) All t_r and t_f times are taken at the first receiver.

7.15 Switching Characteristics for UBT (I Version)

driver in slot 1, with one receiver in slot 21 (minimum load); over recommended operating conditions (unless otherwise noted) (see [Figure 6](#page-19-2)); $T_A = -40^{\circ}$ C to 85°C

(1) All typical values are at $V_{CC} = 3.3 V$, $T_A = 25°C$. All values are derived from TI-SPICE models.

(2) All t_r and t_f times are taken at the first receiver.

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7.16 Skew Characteristics for Bus Transceiver (I Version)

for specific worst-case $\rm V_{CC}$ and temperature within the recommended ranges of supply voltage and operating free-air temperature (see [Figure 7](#page-20-0) and [Figure 8](#page-21-0)); T_A = –40°C to 85°C

(1) t_{sk(t)} – Output-to-output skew is defined as the absolute value of the difference between the actual propagation delay for all outputs of the same packaged device. The specifications are given for specific worst-case V $_{\rm CC}$ and temperature and apply to any outputs switching in opposite directions, both low to high (LH) and high to low (HL) $[t_{\sf sk(t)}].$

7.17 Skew Characteristics for Bus Transceiver (M Version)

for specific worst-case V_{CC} and temperature within the recommended ranges of supply voltage and operating free-air temperature (see [Figure 7](#page-20-0) and [Figure 8](#page-21-0)); $T_A = -55^{\circ}$ C to 125 $^{\circ}$ C

 (1) $t_{sk(t)}$ – Output-to-output skew is defined as the absolute value of the difference between the actual propagation delay for all outputs of the same packaged device. The specifications are given for specific worst-case V_{CC} and temperature and apply to any outputs switching in opposite directions, both low to high (LH) and high to low (HL) $[t_{\text{sk}(t)}].$

7.18 Skew Characteristics for UBT (I Version)

for specific worst-case V_{CC} and temperature within the recommended ranges of supply voltage and operating free-air temperature (see [Figure 7](#page-20-0) and [Figure 8](#page-21-0)); $T_A = -40^{\circ}$ C to 85°C

 (1) $t_{sk(t)}$ – Output-to-output skew is defined as the absolute value of the difference between the actual propagation delay for all outputs of the same packaged device. The specifications are given for specific worst-case V_{CC} and temperature and apply to any outputs switching in opposite directions, both low to high (LH) and high to low (HL) $[t_{sk(t)}]$.

7.19 Skew Characteristics for UBT (M Version)

for specific worst-case V_{CC} and temperature within the recommended ranges of supply voltage and operating free-air temperature (see [Figure 7](#page-20-0) and [Figure 8](#page-21-0)); $T_A = -55^{\circ}$ C to 125°C

Skew Characteristics for UBT (M Version) (continued)

for specific worst-case V_{CC} and temperature within the recommended ranges of supply voltage and operating free-air temperature (see [Figure 7](#page-20-0) and [Figure 8](#page-21-0)); $T_A = -55^{\circ}$ C to 125°C

(1) $t_{\rm sk(t)}$ – Output-to-output skew is defined as the absolute value of the difference between the actual propagation delay for all outputs of the same packaged device. The specifications are given for specific worst-case V_{CC} and temperature and apply to any outputs switching in opposite directions, both low to high (LH) and high to low (HL) $[t_{\sf sk(t)}].$

7.20 Skew Characteristics for Bus Transceiver (I Version)

driver in slot 11, with receiver cards in all other slots (full load); for specific worst-case V_{CC} and temperature within the recommended ranges of supply voltage and operating free-air temperature (see [Figure 6\)](#page-19-2); $T_A = -40^{\circ}C$ to 85°C

(1) All typical values are at $V_{CC} = 3.3$ V, $T_A = 25^{\circ}$ C. All values are derived from TI-SPICE models.
(2) $t_{\text{other}} -$ Output-to-output skew is defined as the absolute value of the difference between the ac-

 $t_{sk(t)}$ – Output-to-output skew is defined as the absolute value of the difference between the actual propagation delay for all outputs of the same packaged device. The specifications are given for specific worst-case V_{CC} and temperature and apply to any outputs switching in opposite directions, both low to high (LH) and high to low (HL) $[t_{\rm sk(t)}]$.

7.21 Skew Characteristics for UBT (I Version)

driver in slot 11, with receiver cards in all other slots (full load); for specific worst-case $V_{\rm CC}$ and temperature within the recommended ranges of supply voltage and operating free-air temperature (see [Figure 6\)](#page-19-2); $T_A = -40^{\circ}$ C to 85°C

(1) All typical values are at $V_{CC} = 3.3 V$, $T_A = 25^{\circ}C$. All values are derived from TI-SPICE models.
(2) $t_{\text{c}}(t) = 0$ utput-to-output skew is defined as the absolute value of the difference between the ac

 $t_{s(k(t))}$ – Output-to-output skew is defined as the absolute value of the difference between the actual propagation delay for all outputs of the same packaged device. The specifications are given for specific worst-case V_{CC} and temperature and apply to any outputs switching in opposite directions, both low to high (LH) and high to low (HL) $[t_{sk(t)}].$

7.22 Skew Characteristics for Bus Transceiver (I Version)

driver in slot 1, with one receiver in slot 21 (minimum load); for specific worst-case V_{CC} and temperature within the recommended ranges of supply voltage and operating free-air temperature (see [Figure 6\)](#page-19-2); $T_A = -40^{\circ}$ C to 85°C

(1) All typical values are at $V_{CC} = 3.3$ V, $T_A = 25^{\circ}$ C. All values are derived from TI-SPICE models.

 (2) $t_{sk(t)}$ – Output-to-output skew is defined as the absolute value of the difference between the actual propagation delay for all outputs of the same packaged device. The specifications are given for specific worst-case V_{CC} and temperature and apply to any outputs switching in opposite directions, both low to high (LH) and high to low (HL) $[t_{sk(t)}].$

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7.23 Skew Characteristics for UBT (I Version)

driver in slot 1, with one receiver in slot 21 (minimum load); for specific worst-case V_{CC} and temperature within the recommended ranges of supply voltage and operating free-air temperature (see [Figure 6\)](#page-19-2); $T_A = -40^{\circ}$ C to 85°C

(1) All typical values are at $V_{CC} = 3.3 V$, $T_A = 25^{\circ}C$. All values are derived from TI-SPICE models.
(2) t_{ek(t)} – Output-to-output skew is defined as the absolute value of the difference between the ac

 $t_{sk(t)}$ – Output-to-output skew is defined as the absolute value of the difference between the actual propagation delay for all outputs of the same packaged device. The specifications are given for specific worst-case V_{CC} and temperature and apply to any outputs switching in opposite directions, both low to high (LH) and high to low (HL) $[t_{sk(t)}]$.

7.24 Maximum Data Transfer Rates

(1) See data sheet for absolute maximum and minimum recommended operating conditions.

(2) Silicon operating life design goal is 10 years at 105°C junction temperature (does not include package interconnect life).

(3) Enhanced plastic product disclaimer applies.

Figure 1. SN74VMEH22501A-EP Derating Chart

7.25 Typical Characteristics

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8 Parameter Measurement Information

8.1 Distributed-Load Backplane Switching Characteristics

The switching characteristics tables show the switching characteristics of the device into the lumped load shown in this section (see [Figure 7](#page-20-0) and [Figure 8\)](#page-21-0). All logic devices currently are tested into this type of load. However, the designer's backplane application probably is a distributed load. For this reason, this device has been designed for optimum performance in the VME64x backplane as shown in [Figure 6.](#page-19-2)

1. Unloaded backplane trace natural impedance (Z_O) is 45 Ω. 45 Ω to 60 Ω is allowed, with 50 Ω being ideal.

2. Card stub natural impedance (Z_O) is 60 Ω.

Figure 6. VME64x Backplane

The following switching characteristics tables derived from TI-SPICE models show the switching characteristics of the device into the backplane under full and minimum loading conditions, to help the designer better understand the performance of the VME device in this typical backplane.

Distributed-Load Backplane Switching Characteristics (continued)

- A. C_L includes probe and jig capacitance.
- B. Waveform 1 is for an output with internal conditions such that the output is low, except when disabled by the output control.

Waveform 2 is for an output with internal conditions such that the output is high, except when disabled by the output control.

- C. All input pulses are supplied by generators having the following characteristics: PRR ≈ 10 MHz, $Z_0 = 50 \Omega$, $t_r \approx 2$ ns, t_f ≈ 2 ns.
- D. The outputs are measured one at a time, with one transition per measurement.

Figure 7. A Port Load Circuit and Voltage Waveforms

Distributed-Load Backplane Switching Characteristics (continued)

-
- B. Waveform 1 is for an output with internal conditions such that the output is low, except when disabled by the output control.

Waveform 2 is for an output with internal conditions such that the output is high, except when disabled by the output control.

- C. All input pulses are supplied by generators having the following characteristics: PRR ≈ 10 MHz, $Z_0 = 50 \Omega$, $t_r \approx 2$ ns, t_f ≈ 2 ns.
- D. The outputs are measured one at a time, with one transition per measurement.

Figure 8. B Port Load Circuit and Voltage Waveforms

9 Detailed Description

9.1 Overview

The SN74VMEH22501A-EP device is a high-drive (–48/64 mA), 8-bit UBT transceiver containing D-type latches and D-type flip-flops for data-path operation in transparent, latched, or flip-flop modes. Data transmission is true logic. The SN74VMEH22501A-EP device is uniquely partitioned as 8-bit UBT transceivers with two integrated 1 bit three-wire bus transceivers.

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9.2 Functional Block Diagram

Pin numbers shown are for the DGG and DGV packages.

Figure 9. Logic Diagram (Positive Logic)

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9.3 Feature Description

9.3.1 Functional Description for Two 1-Bit Bus Transceivers

The OEAB inputs control the activity of the 1B or 2B port. When OEAB is high, the B-port outputs are active. When OEAB is low, the B-port outputs are disabled.

Separate 1A and 2A inputs and 1Y and 2Y outputs provide a feedback path for control and diagnostics monitoring. The OEBY inputs control the 1Y or 2Y outputs. When OEBY is low, the Y outputs are active. When OEBY is high, the Y outputs are disabled.

The OEBY and OEAB inputs can be tied together to form a simple direction control where an input high yields A data to B bus and an input low yields B data to Y bus.

Table 1. 1-Bit Bus Transceiver Function Table

9.3.2 Functional Description for 8-Bit UBT Transceiver

The 3A and 3B data flow in each direction is controlled by the \overline{OE} and direction-control (DIR) inputs. When \overline{OE} is low, all 3A- or 3B-port outputs are active. When \overline{OE} is high, all 3A- or 3B-port outputs are in the high-impedance state.

Table 2. Function Table

The UBT transceiver functions are controlled by latch-enable (LE) and clock (CLKAB and CLKBA) inputs. For 3A-to-3B data flow, the UBT operates in the transparent mode when LE is high. When LE is low, the 3A data is latched if CLKAB is held at a high or low logic level. If LE is low, the 3A data is stored in the latch/flip-flop on the low-to-high transition of CLKAB.

The UBT transceiver data flow for 3B to 3A is similar to that of 3A to 3B, but uses CLKBA.

(1) 3A-to-3B data flow is shown; 3B-to-3A data flow is similar, but uses CLKBA.

(2) Output level before the indicated steady-state input conditions were established, provided that CLKAB was high before LE went low.

(3) Output level before the indicated steady-state input conditions were established.

The UBT transceiver can replace any of the functions as shown in [Table 4](#page-25-0).

Table 4. SN74VMEH22501A-EP UBT Transceiver Replacement Functions(1)

(1) SN74VMEH22501A-EP UBT transceiver replaces all above functions

9.3.3 VMEbus Summary

In 1981, the VMEbus was introduced as a backplane bus architecture for industrial and commercial applications. The data-transfer protocols used to define the VMEbus came from the Motorola® VERSA bus architecture that owed its heritage to the then recently introduced Motorola 68000 microprocessor. The VMEbus, when introduced, defined two basic data-transfer operations: single-cycle transfers consisting of an address and a data transfer, and a block transfer (BLT) consisting of an address and a sequence of data transfers. These transfers were asynchronous, using a master-slave handshake. The master puts address and data on the bus and waits for an acknowledgment. The selected slave either reads or writes data to or from the bus, then provides a dataacknowledge (DTACK*) signal. The VMEbus system data throughput was 40 MBps. Previous to the VMEbus, it was not uncommon for the backplane buses to require elaborate calculations to determine loading and drive current for interface design. This approach made designs difficult and caused compatibility problems among manufacturers. To make interface design easier and to ensure compatibility, the developers of the VMEbus architecture defined specific delays based on a 21-slot terminated backplane and mandated the use of certain high-current TTL drivers, receivers, and transceivers.

In 1989, multiplexing block transfer (MBLT) effectively increased the number of bits from 32 to 64, thereby doubling the transfer rate. In 1995, the number of handshake edges was reduced from four to two in the doubleedge transfer (2eVME) protocol, doubling the data rate again. In 1997, the VMEbus International Trade Association (VITA) established a task group to specify a synchronous protocol to increase data-transfer rates to 320 MBps, or more. The unreleased specification, VITA 1.5 [double-edge source synchronous transfer (2eSST)], is based on the asynchronous 2eVME protocol. It does not wait for acknowledgment of the data by the receiver and requires incident-wave switching. Sustained data rates of 1 GBps, more than ten times faster than traditional VME64 backplanes, are possible by taking advantage of 2eSST and the 21-slot VME320 star-configuration backplane. The VME320 backplane approximates a lumped load, allowing substantially higher-frequency operation over the VME64x distributed-load backplane. Traditional VME64 backplanes with no changes theoretically can sustain 320 MBps.

From BLT to 2eSST – A Look at the Evolution of VMEbus Protocols by John Rynearson, Technical Director, VITA, provides additional information on VMEbus and can be obtained at www.vita.com.

9.4 Device Functional Modes

9.4.1 True Driver With Feedback Path Mode (1-Bit Transceiver)

When OEAB is high and OEABY is s low, the 1-bit transceiver will act as true driver with a feedback path through the Y port for control and diagnostic monitoring.

9.4.2 Direction Control Model (1-Bit Transceiver)

The two 1 bit transceiver can act as a true driver when OEBY and OEAB are tied together.

- 1. Input high:data move from A port to B bus
- 2. Input low: data move from B port to Y bus

9.4.3 Direction Control for 8 Bit UBT

The UBT data flow is controlled by DIR pin. DIR set as high, it will be 3A-3B data flow and if DIR set as low, it will be 3B-3A dataflow. When LE is high, the UBT is in transparent mode and all inputs will be translated to the output.

9.4.4 Latch Storage and Clock Storage

When LE is low and CLK at high or low level, data is latched. During latch state, the output level is per the previous state. When the CLK transitions from low to high, the latched data will be output.

10 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

10.1 Application Information

Target applications for VME backplanes include industrial controls, telecommunications, simulation, high-energy physics, office automation, and instrumentation systems.

10.2 Typical Application

Figure 10. Application Schematic

10.2.1 Design Requirements

The SN74VMEH22501-EP is a combination of 8-bit universal bus transceivers (UBT) and two-bit transceivers, with split LVTTL ports for control and diagnostic monitoring purposes. For the UBTs, 3B1 to 3B8 are the VMEside I/O ports and 3A1 to 3A8 are the LVTTL-side I/O ports. For the two split LVTTL-port transceivers, 1A, 2A are the LVTTL-side input ports, 1Y, 2Y are the LVTTL-side output ports, and 1B, 2B are the VME-side I/O ports (see Figure 5). The UBTs allow transparent, latched, and flip-flop modes of data transfer. It operates at 3.3-V VCC, but can accept 5.5-V input signals at both VME and LVTTL ports. The LVTTL 3A ports and Y outputs have 26-Ω series resistors to reduce the line mismatch on the daughter-card LVTTL side. With the help of Ioff, powerup 3-state, and precharge (BIAS VCC) features, the SN74VMEH22501-EP supports live insertion.

Typical Application (continued)

The VME-side input port has tightly controlled input-switching thresholds of 1/2 VCC ±50 mV for increased noise immunity. In the VMEbus, this input threshold is a clear advantage over the normal TTL or LVTTL type inputs, where V_{IH(min)} is 2.0 V and V_{IL(max)} is 0.8 V. Because the input threshold follows the VCC, data transfer is more immune to the fluctuation of supply voltage, as opposed the ABTE family, where the input threshold is fixed at 1.5 V ±100 mV. To optimize performance, the SN74VMEH22501-EP has been designed into a distributed VME backplane. The OEC™ circuitry, for output edge-rate control, helps reduce reflections as well as electromagnetic interference. The OEC circuitry and high ac drive strength are instrumental in achieving the goal of incident-wave switching. The VME port can source and sink very-high transient currents, which effectively helps to overdrive the reflection on the backplane during transition.

10.2.2 Detailed Design Procedure

By simulating the performance of the device using the VME64x backplane (see [Figure 6\)](#page-19-2), the maximum peak current in or out of the B-port output, as the devices switch from one logic state to another, was found to be equivalent to driving the lumped load shown in [Figure 11](#page-28-0).

Figure 11. Equivalent AC Peak Output-Current Lumped Load

In general, the rise- and fall-time distribution is shown in [Figure 12](#page-28-1). Because VME devices were designed for use into distributed loads like the VME64x backplane (B/P), there are significant differences between low-to-high (LH) and high-to-low (HL) values in the lumped load shown in the PMI (see [Figure 7](#page-20-0) and [Figure 8\)](#page-21-0).

Characterization-laboratory data in [Figure 13](#page-29-3) and [Figure 14](#page-29-3) show the absolute ac peak output current, with different supply voltages, as the devices change output logic state. A typical nominal process is shown to demonstrate the devices' peak ac output drive capability.

11 Power Supply Recommendations

Place 0.1-μF bypass capacitors close to the power supply pins to reduce errors coupling in from noisy or high impedance power supplies.

12 Layout

12.1 Layout Guidelines

The stub length from the VMEH22501 to the connector should be as short as possible. To reduce system skew, stub lengths should be matched for all the data and control bits. Populating both sides of the daughter card may help optimize the stub lengths.

The 5-row connector and the 3-row connector specifications correspond completely. All the data and control lines have the same pin positions in these two connectors. This allows easy migration from a 3-row connector to a 5 row connector. If a 5-row connector is used instead of a 3-row connector, some bypass capacitors between the supply pins and GND of the external rows (at the back of the connector) will help reduce some ground-bounce noise.

TI recommends to use multiple bypass capacitors to stabilize the supply line. To reduce high-frequency noise, TI recommends a small capacitor $(0.1 \mu F)$, or less) for every two VCC pins on the VME side of the VMEH22501. The capacitors should be as close as possible to the VCC pins. An additional large capacitor close to the chip helps maintain the dc level of the power-supply line.

If live insertion is required, TI recommends a specific power-up sequence to use the full live-insertion capability of the VMEH22501. The power-up sequence should be GND, BIAS VCC, OE pin, I/O ports, then VCC.

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Typical Application (continued)

10.2.3 Application Curves

12.2 Layout Example

Figure 15. Layout Recommendation

EXAS **NSTRUMENTS**

13 Device and Documentation Support

13.1 Device Support

13.1.1 Third-Party Products Disclaimer

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13.2 Related Links

The table below lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

Table 5. Related Links

13.3 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of](http://www.ti.com/corp/docs/legal/termsofuse.shtml) [Use.](http://www.ti.com/corp/docs/legal/termsofuse.shtml)

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13.5 Electrostatic Discharge Caution

These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

13.6 Glossary

[SLYZ022](http://www.ti.com/lit/pdf/SLYZ022) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

14 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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PACKAGING INFORMATION

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

⁽⁶⁾ Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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OTHER QUALIFIED VERSIONS OF SN74VMEH22501A-EP :

- Catalog: [SN74VMEH22501A](http://focus.ti.com/docs/prod/folders/print/sn74vmeh22501a.html)
- NOTE: Qualified Version Definitions:
	- Catalog TI's standard catalog product

TEXAS

TAPE AND REEL INFORMATION

ISTRUMENTS

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

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PACKAGE MATERIALS INFORMATION

www.ti.com 21-Jan-2023

*All dimensions are nominal

PACKAGE OUTLINE

DGG0048A TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE

NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. Reference JEDEC registration MO-153.

EXAMPLE BOARD LAYOUT

DGG0048A TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE

NOTES: (continued)

5. Publication IPC-7351 may have alternate designs.

6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

DGG0048A TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE

7. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

8. Board assembly site may have different recommendations for stencil design.

MECHANICAL DATA

MTSS003D – JANUARY 1995 – REVISED JANUARY 1998

DGG (R-PDSO-G) PLASTIC SMALL-OUTLINE PACKAGE**

48 PINS SHOWN

NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold protrusion not to exceed 0,15.
- D. Falls within JEDEC MO-153

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