

ISL6112

Dual Slot PCI-Express Hot Plug Controller

FN6456  
Rev 1.00  
August 25, 2011

The ISL6112 targets the PCI-Express add-in card hot plug application. Together with two each of N-Channel and P-Channel MOSFETs, four current sense resistors, and several external passive components, the ISL6112 provides a compliant hot plug power control solution to any combination of two PCI-Express X1, X4, X8 or X16 slots.

The ISL6112 features the ability to program a maximum current regulated level for each of the MAIN outputs for a common programmable duration. With this ability, both fault isolation protection and imperviousness to electrical transients (OC and soft-start protection) are provided to each system supply. For each 12VMAIN supply, the current regulated (CR) level is set by a resistor value dependent on the size of the PCI-Express connector (X1, X4/X8 or X16) to be powered. This resistor is a sub ohm standard value current sense resistor; one each for each of the 3VMAIN and 12VMAIN supplies. The voltage across this resistor is compared to a 50mV reference, providing a nominal CR protection level that would be set above the maximum specified slot limits. The 3.3V supply can use a 15mΩ sense resistor, compared to a 50mV reference, to provide a nominal regulated current limit of 3.3A to all connector sizes. A shutdown without a CR duration delay is invoked if RSENSE voltage is >100mV. VAUX is internally monitored and controlled to provide nominal limiting to 1A of load current.

The ISL6112 is System Management Interface (SMI) capable, with an integrated SMBus link for communication, control, monitoring, and reporting of IC and slot conditions. Information such as UV, OC, STATUS, and power level are available. Additionally, the IC has a minimum of I/O for implementations where Hot-Plug Hardware Interface (HPI) is implemented.

**Features**

- Supports Two Independent PCI-Express Slots
- Highest Available Accuracy External RSENSE Current Monitoring on Main Supplies
- Programmable Current Regulation Protection Function for X1, X4, X8, X16 Connectors
- 12V, 3.3V, and 3.3VAUX Supplies Supported per PCI Express Specification V1.0A
- Voltage Tolerant I/O SMBus Interface for Slot Power Control and Status, Compatible with SMBus 2.0 Systems
- Programmable Current Regulation Duration
- Programmable In-Rush Current Limiting
- Dual-Level Fault Detection for Quick Fault Response Without Nuisance Tripping
- Slot-to-Slot Electrical and Thermal Isolation
- Two General-Purpose Input Pins Suitable for Interface to Logic and Switches
- Pb-Free (RoHS Compliant)

**Applications**

- PCI Express V1.0A Hot-Plug Power Control
- PCI-Express Servers
- Power Supply Distribution and Control

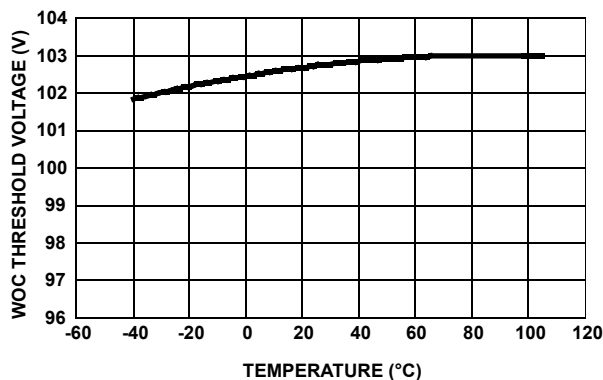


FIGURE 1. FAST TRIP THRESHOLD VOLTAGE vs TEMPERATURE

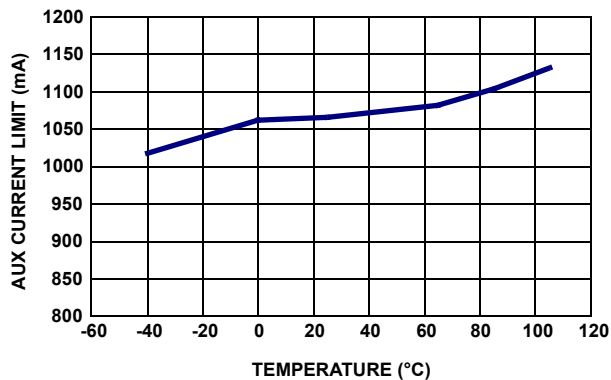
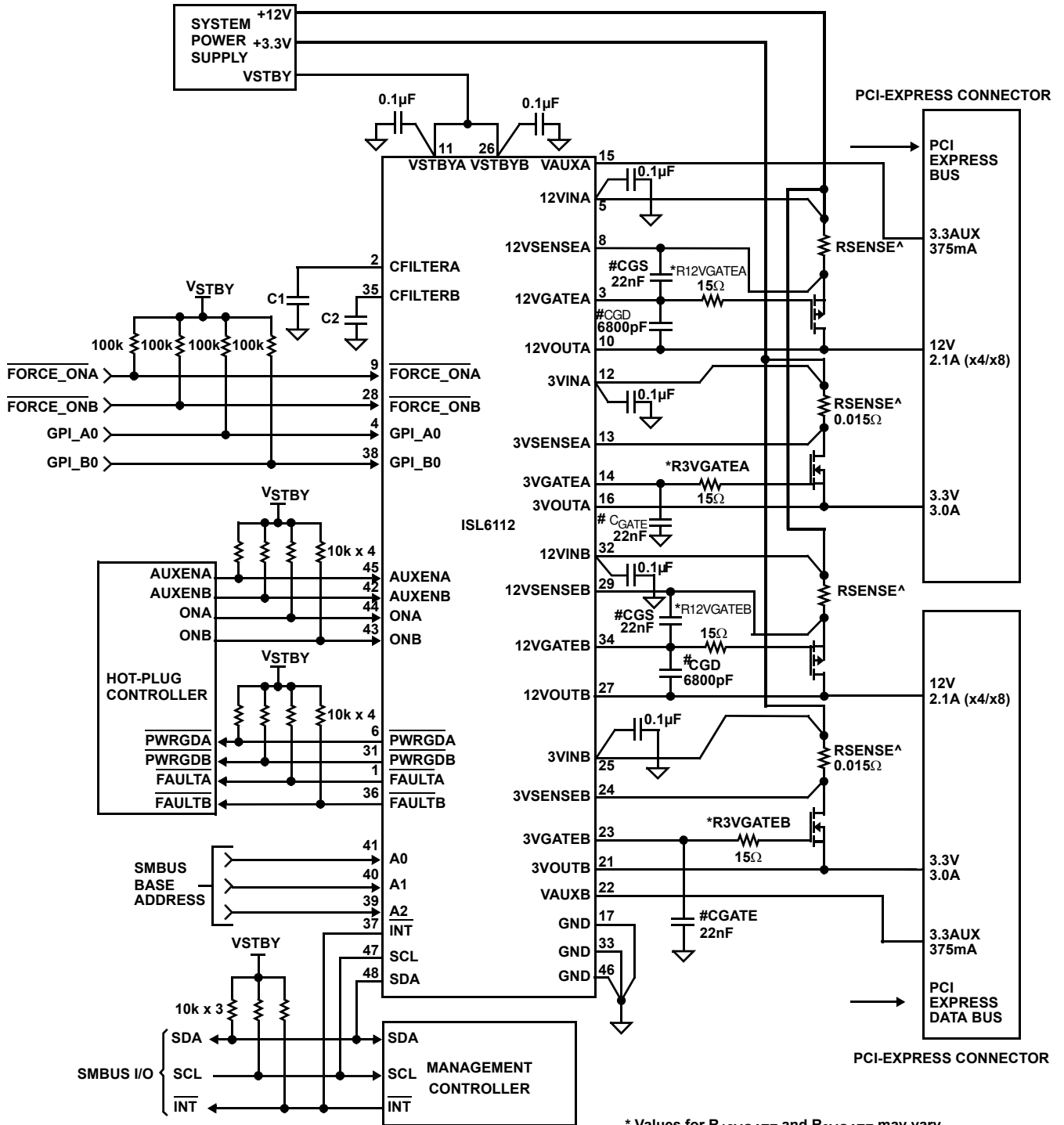


FIGURE 2. AUX CURRENT LIMIT vs TEMPERATURE

# Typical Application Diagram



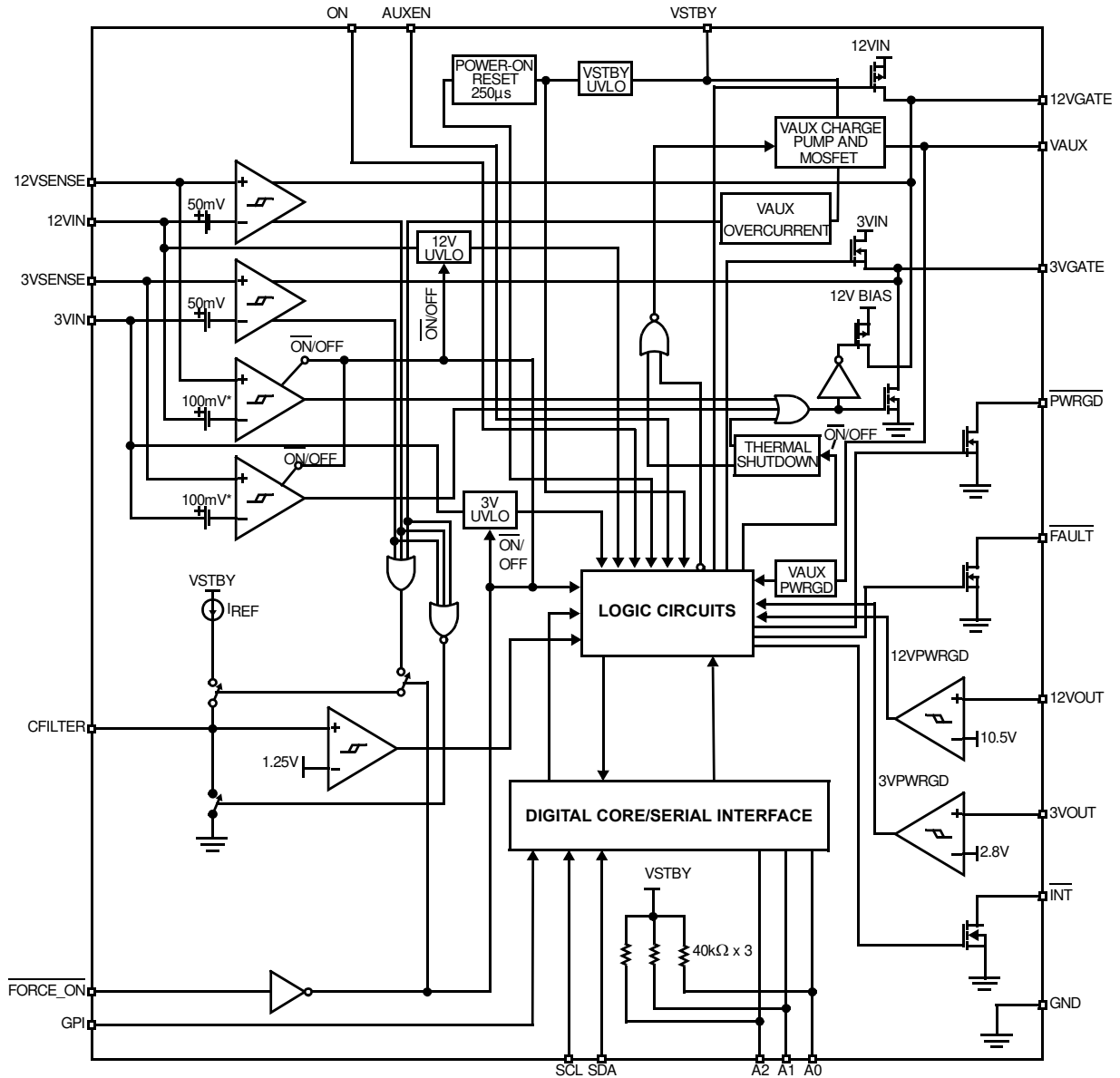
\* Values for R<sub>12VGATE</sub> and R<sub>3VGATE</sub> may vary depending upon the C<sub>GS</sub> of the external MOSFETs.

# These components are not required for ISL6112 operation but can be implemented for GATE output slew rate control (application specific).

• Bold lines indicate high current paths.

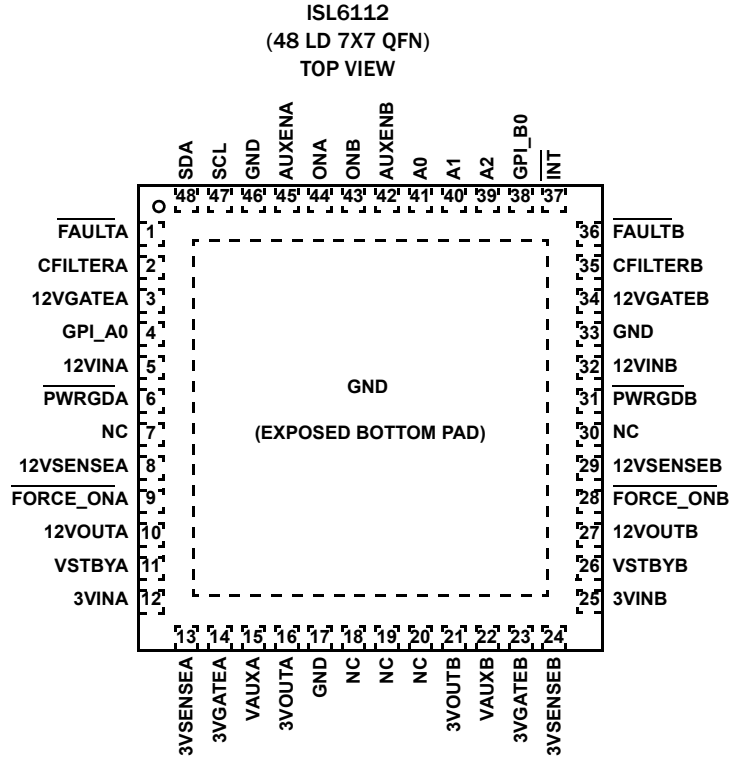
^ R<sub>SENSE</sub> value is application specific.

# Functional Block Diagram (One Channel)



BOTH A AND B SLOTS SHARE THE SCL, SDA, A0, A1, A2, INT PINS.

## Pin Configuration



## Pin Descriptions

PIN NUMBER	PIN NAME	PIN FUNCTION
5, 32	12VINA, 12VINB	Provides 12VMAIN power supply and the high side of the sense resistor inputs. This must be a Kelvin connection between IC and sense resistor. An undervoltage lockout circuit (UVLO) prevents the switches from turning on while this input is less than its lockout threshold.
12, 25	3VINA, 3VINB	Provides 3.3VMAIN power supply and the high side of the sense resistor inputs. This must be a Kelvin connection between IC and sense resistor. An undervoltage lockout circuit (UVLO) prevents the switches from turning on while this input is less than its lockout threshold.
16, 21	3VOUTA, 3VOUTB	3.3VOUT. Connected to 3.3V FET source. These are used to monitor the 3.3V output voltages for Power-Good status.
10, 27	12VOUTA, 12VOUTB	12VOUT. Connected to 12V FET drain. These are used to monitor the 3.3V output voltages for Power-Good status.
8, 29	12VSENSEA, 12VSENSEB	12VMAIN low side of sense resistor connection. When either current limit threshold of the load current across the sense resistor = 50mV is reached, the related 12VGATE pin is modulated to maintain a constant voltage across the sense resistor and thus a constant current into the load. If the 50mV threshold is exceeded for $t_{FLT}$ , the isolation protection is tripped, and the GATE pin for the affected supply's external MOSFET is immediately pulled high. This must be a Kelvin connection between IC and sense resistor.
13, 24	3VSENSEA, 3VSENSEB	3.3VMAIN low side of sense resistor connection. When either current limit threshold of the load current across the sense resistor = 50mV is reached, the related 3V GATE pin is modulated to maintain a constant voltage across the sense resistor and thus a constant current into the load. If the 50mV threshold is exceeded for $t_{FLT}$ , the isolation protection is tripped, and the GATE pin for the affected supply's external MOSFET is immediately pulled low. This must be a Kelvin connection between IC and sense resistor.
3, 34	12VGATEA 12VGATEB	12V gate drive outputs. Each pin connects to the gate of an external P-Channel MOSFET. During power-up, the CGATE and the CGS of the MOSFETs are connected to a 25 $\mu$ A current sink. This controls the value of $dv/dt$ seen at the source of the MOSFETs. During current limit events, the voltage at this pin is adjusted to maintain constant current through the switch for a period of $t_{FLT}$ . Whenever an overcurrent, thermal shutdown, or input undervoltage fault condition occurs, the GATE pin for the affected slot is immediately brought high. These pins are charged by an internal current source during power-down.

## Pin Descriptions (Continued)

PIN NUMBER	PIN NAME	PIN FUNCTION
14, 23	3VGATEA 3VGATEB	3V gate drive outputs. Each pin connects to the gate of an external N-Channel MOSFET. During power-up, the CGATE and the CGS of the MOSFETs are connected to a 25 $\mu$ A current source. This controls the value of dv/dt seen at the source of the MOSFETs, and hence the current flowing into the load capacitance. During current limit events, the voltage at this pin is adjusted to maintain constant current through the switch for a period of $t_{FLT}$ . Whenever an overcurrent, thermal shutdown, or input undervoltage fault condition occurs, the GATE pin for the affected slot is immediately brought low. During power-down, these pins are discharged by an internal current source.
11, 26	VSTBYA, VSTBYB	3.3V standby input voltage. Required to support PCI-Express VAUX output. Additionally, the SMBus logic and internal registers run off of $V_{STBY}$ to ensure that the chip is accessible during standby modes. A UVLO circuit prevents turn-on of this supply until $V_{STBY}$ rises above its UVLO threshold. Both pins must be externally connected together at the ISL6112 controller.
15, 22	VAUXA, VAUXB	3.3VAUX outputs to PCI-Express card slots. These outputs connect the 3.3AUX pin of the PCI-Express connectors to $V_{STBY}$ via internal 400m $\Omega$ MOSFETs. These outputs are 1A current limited and protected against short-circuit faults.
44, 43	ONA, ONB	Enable inputs. Rising-edge triggered. Used to enable or disable the MAINA and MAINB (+3.3V and +12V) outputs. Taking ON low after a fault resets the +12V and/or +3.3V fault latches for the affected slot. Tie these pins to GND if using SMI power control. Also see pin descriptions for $\overline{FAULTA}$ and $\overline{FAULTB}$ .
45, 42	AUXENA, AUXENB	Level sensitive auxiliary enable inputs. Used to enable or disable the VAUX outputs. Taking AUXEN low after a fault resets the respective slot's Aux Output Fault Latch. Tie these pins to GND if using SMI power control. Also see pin descriptions for $\overline{FAULTA}$ and $\overline{FAULTB}$ .
2, 35	CFILTERA, CFILTERB	Overcurrent timers. Capacitors connected between these pins and GND set the duration of $CR_{TIM}$ . $CR_{TIM}$ is the amount of time for which a slot remains in current limit before its isolation protection is invoked.
6, 31	$\overline{PWRGDA}$ $\overline{PWRGDB}$	Power-is-Good outputs. Open-drain, active-low. Asserted when a slot has been commanded to turn on and has successfully begun delivering power to its respective +12V, +3.3V, and VAUX outputs. Each pin requires an external pull-up resistor to $V_{STBY}$ .
1, 36	$\overline{FAULTA}$ , $\overline{FAULTB}$	Fault outputs. Open-drain, active-low. Asserted whenever the isolation protection trips due to a fault condition (overcurrent, input undervoltage, over-temperature). Each pin requires an external pull-up resistor to $V_{STBY}$ . Bringing the slot's ON pin low resets $\overline{FAULT}$ , if $\overline{FAULT}$ was asserted in response to a fault condition on one of the slot's MAIN outputs (+12V or +3.3V). $\overline{FAULT}$ is reset by bringing the slot's AUXEN pin low if $\overline{FAULT}$ was asserted in response to a fault condition on the slot's VAUX output. If a fault condition occurred on both the MAIN and VAUX outputs of the same slot, then both ON and AUXEN must be brought low to de-assert the $\overline{FAULT}$ output.
9, 28	$\overline{FORCE\_ONA}$ $\overline{FORCE\_ONB}$	Enable inputs. Active-low, level-sensitive. Asserting a $\overline{FORCE\_ON}$ input turns on all three of the respective slot's outputs (+12V, +3.3V, and VAUX) while specifically defeating all protections on those supplies. This explicitly includes all overcurrent and short-circuit protections, and on-chip thermal protection for the VAUX supplies. Additionally included are the UVLO protections for the +3.3V and +12VMAIN supplies. The $\overline{FORCE\_ON}$ pins do not disable UVLO protection for the VAUX supplies. These input pins are intended for diagnostic purposes only. Asserting $\overline{FORCE\_ON}$ causes the respective slot's $\overline{PWRGD}$ and $\overline{FAULT}$ pins to enter their open-drain state. Note that the SMBus register set continues to reflect the actual state of each slot's supplies. There is a pair of register bits, accessible via the SMBus, which can be set to disable (unconditionally de-assert) either or both of the $\overline{FORCE\_ON}$ pins; see CNTRL Register Bit D[2].
4, 38	GPI_A0, GPI_B0	General purpose inputs. The states of these two inputs are available by reading the Common Status Register, Bits [4:5]. If not used, connect each pin to GND.
39, 40, 41	A2, A1, A0	SMBus address select pins. Connect to ground or leave open in order to program device SMBus base address. These inputs have internal pull-up resistors to $V_{STBY}$ . Address programmed on rising $V_{STBY}$ .
48	SDA	Bidirectional SMBus data line.
47	SCL	SMBus clock input.
37	$\overline{INT}$	Interrupt output. Open-drain, active-low output. Asserted whenever a power fault is detected if the INTMSK bit (CS Register Bit D[3]) is a logical "0". This output is cleared by performing an "echo reset" to the appropriate fault bits in the STAT or CS registers. This pin requires an external pull-up resistor to $V_{STBY}$ .
17, 33, 46	GND	IC reference pins. Connect together and tie directly to the system's analog GND plane directly at the device.
7, 18, 19, 20, 30	NC	Reserved. Make no external connections to these pins.

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## Ordering Information

PART NUMBER (Note 2)	PART MARKING	TEMP RANGE (°C)	PACKAGE (Pb-Free)	PKG. DWG. #
ISL6112IRZA (Notes 1, 3)	ISL6112 IRZ	-40 to +85	48 Ld 7x7 QFN	L48.7x7
ISL6112EVAL1Z	Evaluation Platform			

**NOTES:**

1. Add "-T\*" suffix for tape and reel. Please refer to [TB347](#) for details on reel specifications.
2. These Intersil Pb-free plastic packaged products employ special Pb-free material sets, molding compounds/die attach materials, and 100% matte tin plate plus anneal (e3 termination finish, which is RoHS compliant and compatible with both SnPb and Pb-free soldering operations). Intersil Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.
3. For Moisture Sensitivity Level (MSL), please see device information page for [ISL6112](#). For more information on MSL please see Tech Brief [TB363](#).

**Absolute Maximum Ratings (Note 4)**

12VIN, 12VSENSE, 12VOUT.....	+14.5V
VSTBY, 3VIN, 3VSENSE, 3VOUT.....	+7V
12VGATE .....	-0.3V to 12VI
3VGATE .....	-0.3V to 12VI
Logic I/O .....	-0.5V to +5.5V
VAUX Output Current .....	Short Circuit Protected
ESD Rating	
Human Body Model .....	2kV
Machine Model .....	200V
Charged Device Model .....	1kV

**Thermal Information**

Thermal Resistance (Typical)	$\theta_{JA}$ (°C/W)	$\theta_{JC}$ (°C/W)
48 Ld 7x7 QFN Package (Notes 5, 6).....	27	3
Maximum Junction Temperature .....	+150°C	
Maximum Storage Temperature Range .....	-65°C to +150°C	
Pb-free reflow profile .....	see link below	
	<a href="http://www.intersil.com/pbfree/Pb-FreeReflow.asp">http://www.intersil.com/pbfree/Pb-FreeReflow.asp</a>	

**Operating Conditions**

12VMAIN Supply Voltage Range.....	+12V ± -10%
3.3VMAIN Supply Voltage Range .....	+3.3V ± -10%
AUXI Supply Voltage Range.....	+3.3V ± -10%
Temperature Range (T <sub>A</sub> ) .....	-40°C to +85°C

CAUTION: Do not operate at or near the maximum ratings listed for extended periods of time. Exposure to such conditions may adversely impact product reliability and result in failures not covered by warranty.

## NOTES:

- All voltages are relative to GND, unless otherwise specified.
- $\theta_{JA}$  is measured in free air with the component mounted on a high effective thermal conductivity test board with “direct attach” features. See Tech Brief [TB379](#).
- For  $\theta_{JC}$ , the “case temp” location is the center of the exposed metal pad on the package underside.

**Electrical Specifications** 12VIN = 12V, 3VIN = 3.3V, VSTBY = 3.3V, T<sub>A</sub> = T<sub>J</sub> = -40°C to +85°C, unless otherwise noted. **Boldface** limits apply over the operating temperature range, -40°C to +85°C.

PARAMETER	SYMBOL	CONDITION	MIN (Note 7)	TYP	MAX (Note 7)	UNITS
<b>POWER CONTROL AND LOGIC SECTIONS</b>						
Supply Current	ICC12	HPI Enabled or SMI enabled with no load		0.9	<b>1.5</b>	mA
	ICC3.3			0.1	<b>0.2</b>	mA
	ICCSTBY			5	<b>6</b>	mA
Undervoltage Lockout Thresholds	VUVLO(12V)	12VIN increasing	<b>8</b>	9	<b>10</b>	V
	VUVLO(3V)	3VIN increasing	<b>2.1</b>	2.5	<b>2.75</b>	V
	VUVLO(STBY)	VSTBY increasing	<b>2.8</b>	2.9	<b>2.96</b>	V
Undervoltage Lockout Hysteresis 12VIN, 3VIN	VHYSUV			180		mV
Undervoltage Lockout Hysteresis VSTBY	VHYSSTBY			50		mV
Power-Good Undervoltage Thresholds	VUVTH(12V)	12VOUT decreasing	<b>10.15</b>	10.5	<b>10.75</b>	V
	VUVTH(3V)	3VOUT decreasing	<b>2.7</b>	2.8	<b>2.9</b>	V
	VUVTH(VAUX)	VAUX decreasing	<b>2.55</b>	2.8	<b>3</b>	V
Power-Good Detect Hysteresis	VHYS PG			30		mV
12VGATE Voltage	VGATE (12V)	Max. Gate Voltage when Enabled	<b>0</b>	0.4	<b>0.55</b>	V
12VGATE Sink Current	IGATE(12VSINK)	Start Cycle	<b>17</b>	25	<b>35</b>	μA
12VGATE Pull-up Current (Fault Off)	IGATE (12VPULL-UP)	Any fault condition (VDD - VGATE) = 2.5V	<b>35</b>	72	-	mA
3VGATE Voltage	VGATE(3V)	Minimum Gate Voltage when Enabled	<b>12VIN - 0.3</b>	12VIN - 0.2	<b>12VIN</b>	V
3VGATE Charge Current	IGATE (3VCHARGE)	Start Cycle	<b>17</b>	25	<b>35</b>	μA
3VGATE Sink Current (Fault Off)	IGATE(3VSINK)	Any fault condition VGATE = 2.5V	<b>80</b>	105		mA

**Electrical Specifications** 12VIN = 12V, 3VIN = 3.3V, VSTBY = 3.3V,  $T_A = T_J = -40^\circ\text{C}$  to  $+85^\circ\text{C}$ , unless otherwise noted. **Boldface limits apply over the operating temperature range,  $-40^\circ\text{C}$  to  $+85^\circ\text{C}$ . (Continued)**

PARAMETER	SYMBOL	CONDITION	MIN (Note 7)	TYP	MAX (Note 7)	UNITS
<b>CFILTER OVERCURRENT DELAY TIME PINS 2 AND 35 FLOATING</b>						
CFILTER Threshold Voltage	VFILTER		<b>1.20</b>	1.25	<b>1.30</b>	V
CFILTER Charging Current Nominal Current Limit Duration = $C_{CFILTER} \times 550k$	IFILTER	VXVIN - VXSENSE > VTHILIMIT	<b>2</b>	2.5	<b>3</b>	$\mu\text{A}$
	tFILTER	CFILTER Open		10		$\mu\text{s}$
Current Limit Threshold Voltages	VTHILIMIT	VXIN - VXSENSE	<b>47.5</b>	50	<b>52.5</b>	mV
Fast-Trip Threshold Voltages	VTHFAST	VXVIN - VXSENSE	<b>85</b>	100	<b>115</b>	mV
XVSENSE Input Current	ISENSE			0.1		$\mu\text{A}$
LOW-Level Input Voltage ON, AUXEN, GPI, FORCE_ON, PRSNT	VIL				<b>0.8</b>	V
Output LOW Voltage $\overline{\text{FAULT}}$ , $\overline{\text{PWRGD}}$	VOL	IOL = 3mA			<b>0.4</b>	V
HIGH-Level Input Voltage ON, AUXEN, GPI, FORCE_ON	VIH		<b>2.1</b>		<b>5</b>	V
Internal Pull-ups to VSTBY	RPULL-UP			40	<b>50</b>	k $\Omega$
12VIN, 3VIN Input Leakage Current	ILKG,OFF XVIN	VSTBY = +3.3V, 12VIN = OFF; 3VIN = OFF		0.5	<b>1</b>	$\mu\text{A}$
Input Leakage Current, ON, AUXEN, $\overline{\text{FORCE\_ON}}$	IIL		<b>-2</b>		<b>2</b>	$\mu\text{A}$
Off-State Leakage Current $\overline{\text{FAULT}}$ , $\overline{\text{PWRGD}}$ , GPI	ILKG(OFF)	GPI ILKG for these two pins measured with VAUX OFF	<b>-2</b>		<b>2</b>	$\mu\text{A}$
Over-temperature Shutdown and Reset Thresholds, with Overcurrent On Slot	TOV	$T_J$ increasing, each slot		140		$^\circ\text{C}$
		$T_J$ decreasing, each slot		130		$^\circ\text{C}$
Over-temperature Shutdown and Reset Thresholds, All Other Conditions (All Outputs Will Latch Off)	TOV	$T_J$ increasing, each slot		160		$^\circ\text{C}$
		$T_J$ decreasing, each slot		150		$^\circ\text{C}$
Output MOSFET Resistance VAUX MOSFET	$r_{DS(AUX)}$	IDS = 375mA			<b>350</b>	m $\Omega$
Off-State Output Offset Voltage VAUX	VOFF(VAUX)	VAUX = Off		25	<b>40</b>	mV
Regulated Current Level	ILIM(AUX)		<b>0.8</b>	<b>1</b>	<b>1.2</b>	A
Output Discharge Resistance	RDIS(12V)	12VOUT = 6.0V		1400	<b>1850</b>	$\Omega$
	RDIS(3V)	3VOUT = 1.65V		140	<b>180</b>	$\Omega$
	RDIS(VAUX)	3VAUX = 1.65V		350	<b>400</b>	$\Omega$
12V Current Limit Response Time (See "Typical Application Diagram" on page 2).	tOFF(12V)	CGATE = 25pF VIN - VSENSE = 140mV		<b>1</b>	<b>2.1</b>	$\mu\text{s}$
3.3V Current Limit Response Time (See "Typical Application Diagram" on page 2).	tOFF(3V)	CGATE = 25pF VIN - VSENSE = 140mV		0.3	<b>1</b>	$\mu\text{s}$
VAUX Current Limit Response Time (See "Typical Application Diagram" on page 2).	tSC	VAUX = 0V, VSTBY = +3.3V		2.5		$\mu\text{s}$
Delay from MAIN Overcurrent to $\overline{\text{FAULT}}$ Output	tPROP (12V FAULT or 3V FAULT)	CFILTER = 0 VIN - VSENSE = 140mV		<b>1</b>		$\mu\text{s}$



**Electrical Specifications** 12VIN = 12V, 3VIN = 3.3V, VSTBY = 3.3V,  $T_A = T_J = -40^\circ\text{C}$  to  $+85^\circ\text{C}$ , unless otherwise noted. **Boldface limits apply over the operating temperature range,  $-40^\circ\text{C}$  to  $+85^\circ\text{C}$ . (Continued)**

PARAMETER	SYMBOL	CONDITION	MIN (Note 7)	TYP	MAX (Note 7)	UNITS
Delay from VAUX Overcurrent to $\overline{\text{FAULT}}$ Output	tPROP (VAUXFAULT)	$I_{\text{LIM(AUX)}}$ to $\overline{\text{FAULT}}$ output CFILTER = 0 VAUX Output Grounded		1		$\mu\text{s}$
ON, AUXEN, $\overline{\text{PRSNT}}$ Minimum Pulse Width	tW			100		ns
Power-On Reset Time after VSTBY Becomes Valid	tPOR			250		$\mu\text{s}$
<b>SMBUS TIMING</b>						
SCL (clock) period	t1		<b>2.5</b>			$\mu\text{s}$
Data In setup time to SCL HIGH	t2		<b>100</b>			ns
Data Out stable after SCL LOW	t3		<b>300</b>			ns
Data LOW setup time to SCL LOW	t4		<b>100</b>			ns
Data HIGH hold time after SCL HIGH	t5		<b>100</b>			ns

**NOTE:**

7. Parameters with MIN and/or MAX limits are 100% tested at  $+25^\circ\text{C}$ , unless otherwise specified. Temperature limits established by characterization and are not production tested.

# Typical Performance Curves

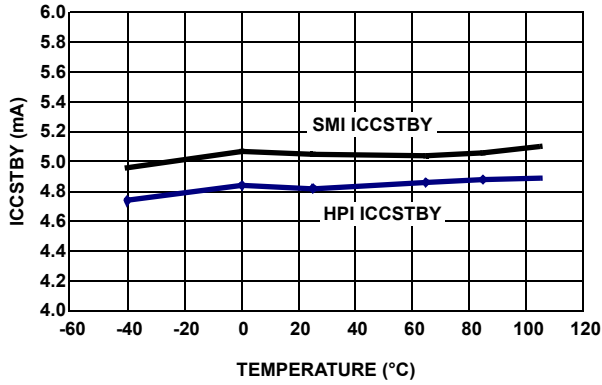


FIGURE 3. ICCSTBY CURRENT vs TEMPERATURE

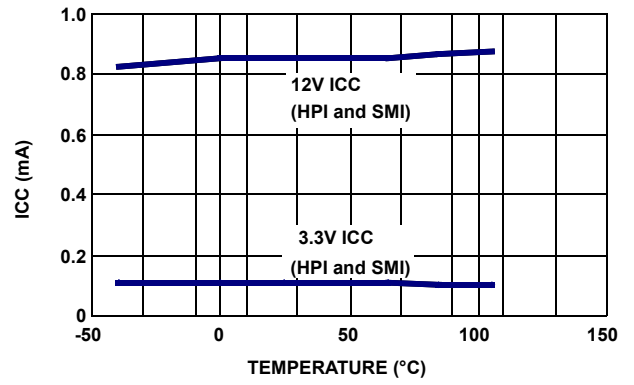


FIGURE 4. ICC CURRENT vs TEMPERATURE

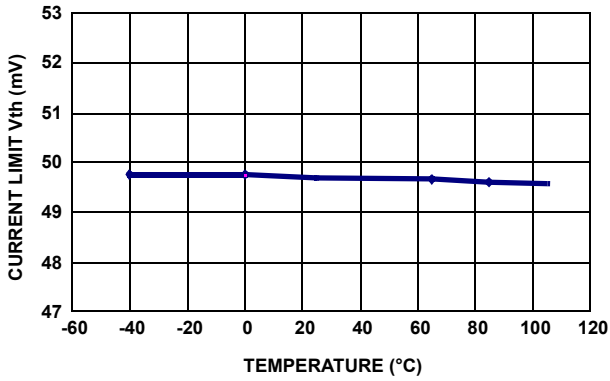


FIGURE 5. CURRENT LIMIT THRESHOLD VOLTAGE vs TEMPERATURE

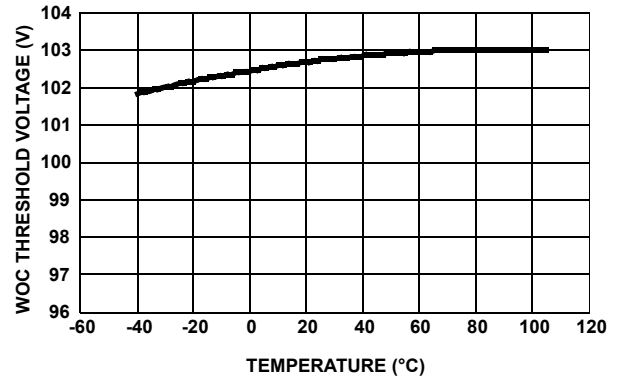


FIGURE 6. FAST TRIP THRESHOLD VOLTAGE vs TEMPERATURE

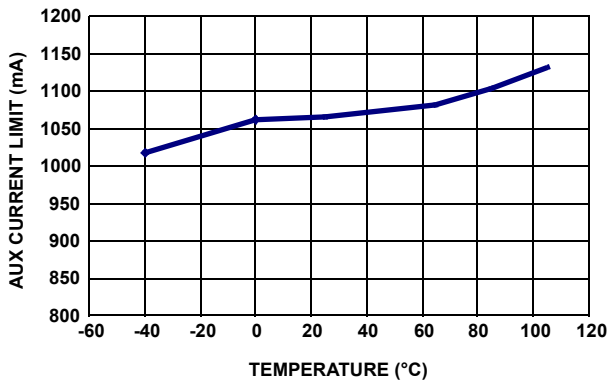


FIGURE 7. AUX CURRENT LIMIT vs TEMPERATURE

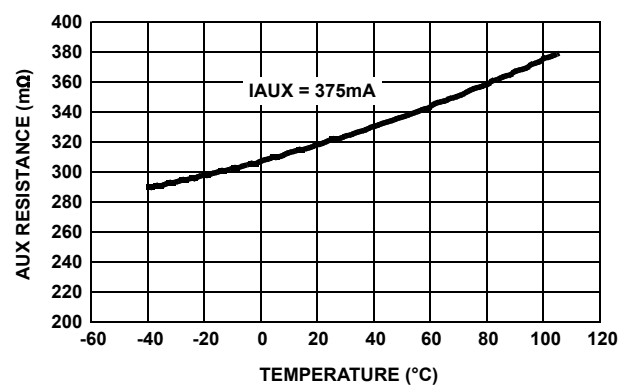


FIGURE 8. AUX  $r_{DS(ON)}$  vs TEMPERATURE

## Typical Performance Curves (Continued)

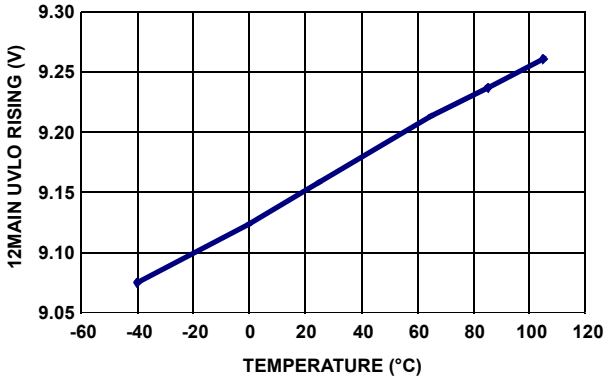


FIGURE 9. 12MAIN RISING UVLO THRESHOLD VOLTAGE vs TEMPERATURE

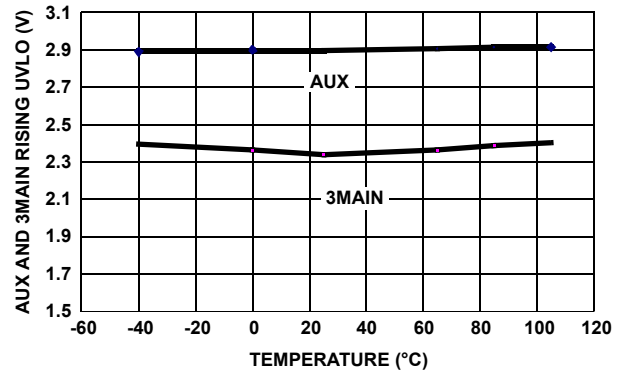


FIGURE 10. AUX AND 3.3MAIN RISING UVLO THRESHOLD VOLTAGE vs TEMPERATURE

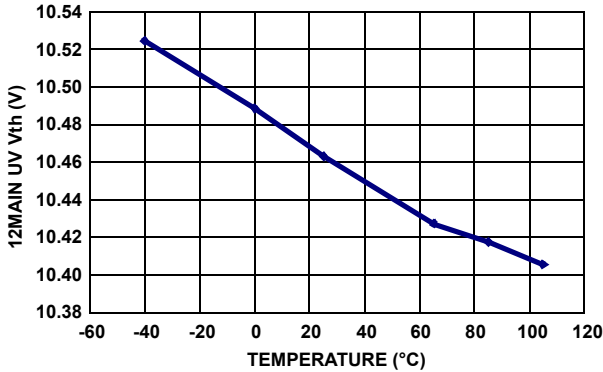


FIGURE 11. 12MAIN POWER GOOD THRESHOLD VOLTAGE vs TEMPERATURE

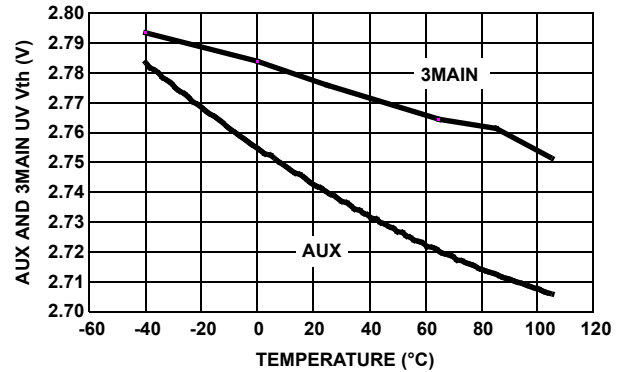


FIGURE 12. AUX AND 3MAIN POWER GOOD THRESHOLD VOLTAGE vs TEMPERATURE

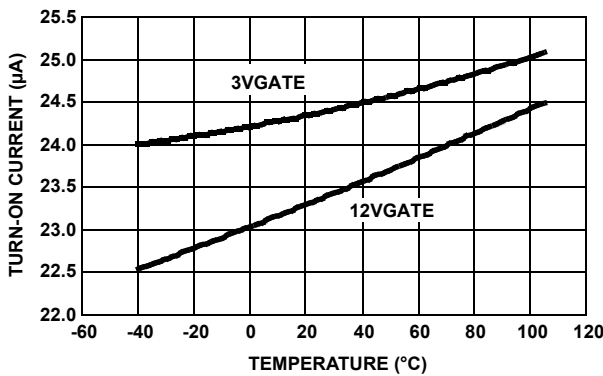


FIGURE 13. ISL6112 GATE TURN-ON CURRENT (ABS) vs TEMPERATURE

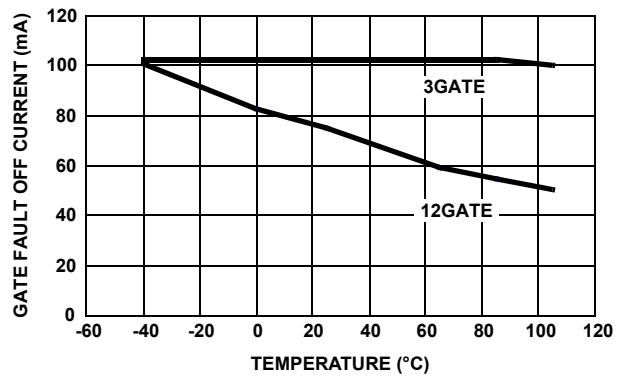


FIGURE 14. GATE FAULT OFF CURRENT (ABS) vs TEMPERATURE

## Typical Performance Curves (Continued)

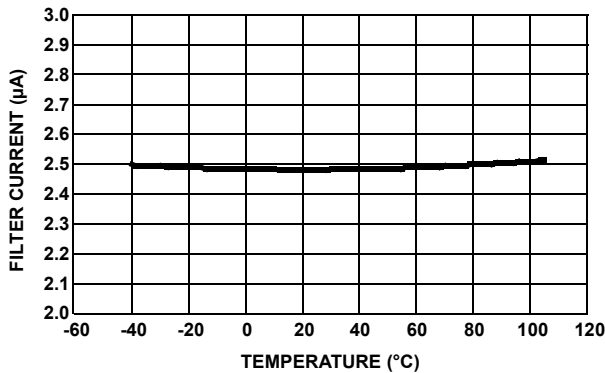


FIGURE 15. FILTER CHARGE CURRENT vs TEMPERATURE

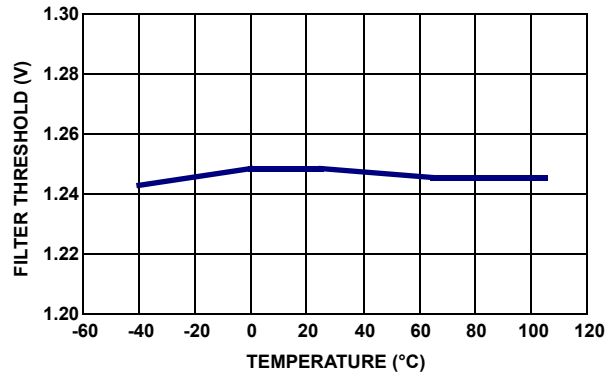


FIGURE 16. FILTER THRESHOLD VOLTAGE vs TEMPERATURE

## Functional Description

The ISL6112 protects the power supplies in PCI-Express systems that utilize hot-pluggable add-in cards. This IC, together with two each of N-Channel and P-Channel MOSFETs, four current sense resistors, and a few external passive components, provides a compliant hot plug power control solution to any combination of two PCI-Express X1, X4, X8 or X16 slots.

The ISL6112 primarily features start-up in-rush current protection, maximum current regulated (CR) levels for each of the MAIN and AUX outputs, and programmable CR duration so that both fault isolation protection and imperviousness to electrical transients are provided. The ISL6112 also offers input and output voltage supervisory functions and two operational system interfaces for implementation flexibility.

### In-Rush Current Protection

When any electronic circuitry is powered up, there is an in-rush of current due to the charging of bulk capacitance that resides across the circuit board supply pins. This transient in-rush current may cause the system supply voltages to temporarily droop out of regulation, causing data loss or system lock-up. The ISL6112 addresses these issues by limiting the in-rush currents to the PCI-Express add-in cards, thereby controlling the rate at which the load circuits turn on. See Figures 17, 18, 19, 20, 21 and 22 for AUX and MAIN turn-on examples that illustrate the current limiting capabilities across a variety of compensation component values.

### MAIN Supply Overcurrent Protection

For each of the 3VMAIN and 12VMAIN supplies, the current regulated (CR) levels are set by a sub ohm value sense resistor. The value for 12VMAIN is dependant on the size of the PCI-Express connector (X1, X4/X8 or X16) to be powered. The voltage across this resistor is compared to a 50mV internal reference, providing a nominal CR protection level that would be set above the maximum specified slot limits. The 3.3VMAIN supply can use a 15mΩ sense resistor compared to a 50mV reference to provide a nominal regulated current limit of 3.3A, as this supply has a common 3A maximum across all slot sizes. For both MAIN supplies, there is a Way Overcurrent (WOC) shutdown

protocol that is without a CR duration. WOC is invoked if the load current causes the RSENSE voltage to be >100mV (see Figures 23 and 24).

### VAUX Supply Overcurrent Protection

The VAUX load current is internally monitored and controlled via an internal power FET. This FET has a typical  $r_{DS(ON)}$  of 320mΩ at a VAUX current of 375mA to minimize distribution losses to typically <100mV through the IC. Using active monitoring and control, the ISL6112 provides nominal limiting to ~1000mA of load current across the temperature range and for various loading conditions. See Figures 17, 25 and 26 for examples of this performance.

### Current Regulation (CR) Duration

The CR duration for each slot is set by an external capacitor between the associated CFILTER pin and ground. This feature masks current transients and overcurrents prior to supply turn-off. Once the CR duration has expired, the IC quickly turns off the associated MAIN outputs via its external FETs or the failed AUX output, unloading the faulted load card from the supply voltage rails.

### UVLO, Power-Good, and FAULT

The ISL6112 incorporates undervoltage lock-out (UVLO) protections on each of the four MAIN VIN and two VSTBY supplies to prevent operation during a brown-out condition. Likewise, on the outputs are minimum voltage compliances that must be satisfied for the Power-Good output, PWRGD, to be asserted. There is some hysteresis on the UVLO levels as the voltage on VIN decreases to ensure IC operation below the minimum operating supply standards. The  $\overline{\text{FAULT}}$  output is asserted (low) whenever there is an OC, OT or UV condition. The  $\overline{\text{FAULT}}$  is cleared once the appropriate enable is deasserted.

### Operational System Interfaces

The ISL6112 employs two system interfaces: the hardware Hot-Plug Interface (HPI) and the System Management Interface (SMI). The HPI I/O includes ON,  $\overline{\text{AUXEN}}$ ,  $\overline{\text{FAULT}}$  and PWRGD. The SMI I/O consists of SDA, SCL, and  $\overline{\text{INT}}$ , the signals of which conform to the levels and timing of the SMBus specification (see

“SMI-only Control Applications” on page 18). The ISL6112 can be operated exclusively from either the SMI or HPI, or can employ the HPI for power control while continuing to use the SMI for access to all but the power control registers.

In addition to the basic power control features of the ISL6112 accessible by the HPI, the SMI gives the host access to the following information from the part:

- Fault conditions occurring on each supply. These faults include overcurrent, over-temperature and undervoltage.
- GPI pin status when using the system.

When using the SMI for power control, the HPI must not be used. Conversely, when using the HPI for power control, power control commands must not be executed over the SMI bus. All other register accesses via the SMI bus remain permissible while in HPI control mode. When using SMI exclusively, the HPI input pins (ON, AUXEN, and FORCE\_ON) should be configured as shown in Figure 27 (disabling HPI when SMI control is used). This configuration safeguards the power slots if the SMBus communication link is disconnected for any reason.

When using HPI exclusively, the SMBus (or SMI) will be inactive if the input pins (SDA, SCL, A0, A1, and A2) are configured as shown in Figure 27.

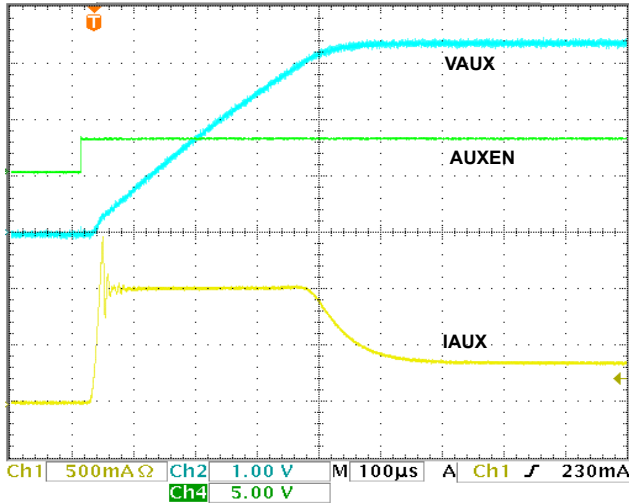


FIGURE 17. VAUX TURN-ON  $R_{LOAD} = 10\Omega$ ,  $C_{LOAD} = 100\mu F$

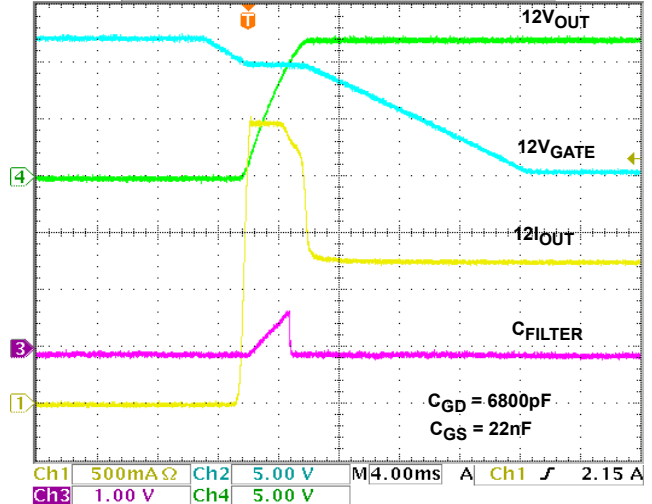


FIGURE 18. 12VMAIN START-UP  $R_{LOAD} = 10\Omega$ ,  $C_{LOAD} = 470\mu F$

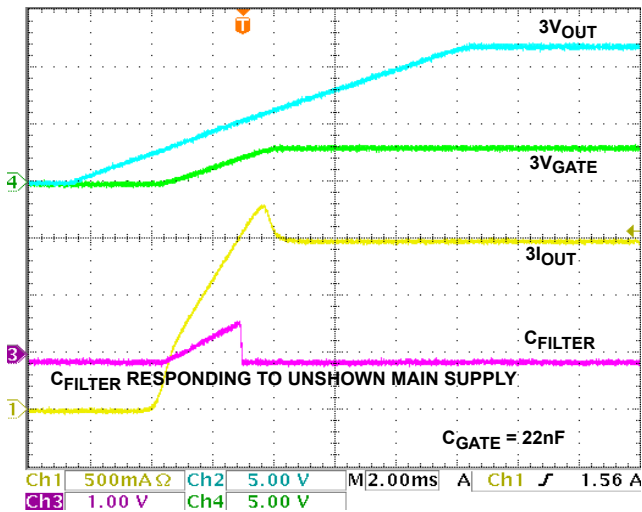


FIGURE 19. 3VMAIN START-UP  $R_{LOAD} = 2\Omega$ ,  $C_{LOAD} = 470\mu F$

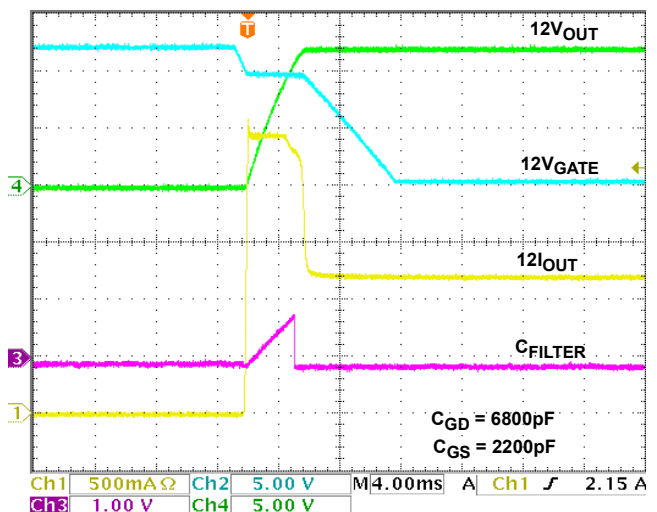


FIGURE 20. 12VMAIN START-UP  $R_{LOAD} = 10\Omega$ ,  $C_{LOAD} = 470\mu F$

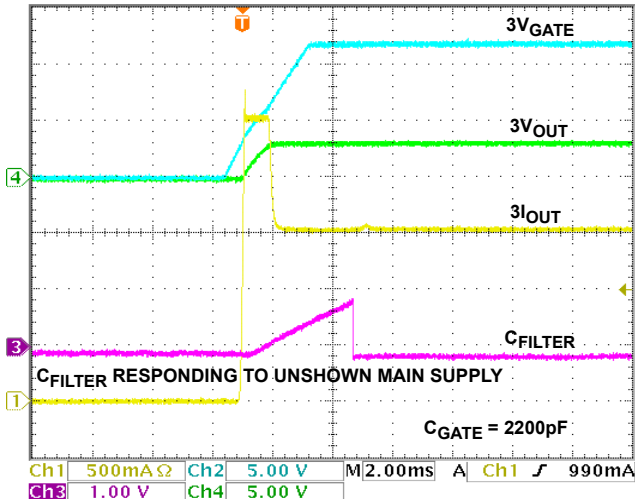


FIGURE 21. 3VMAIN START-UP  $R_{LOAD} = 2\Omega$ ,  $C_{LOAD} = 470\mu F$

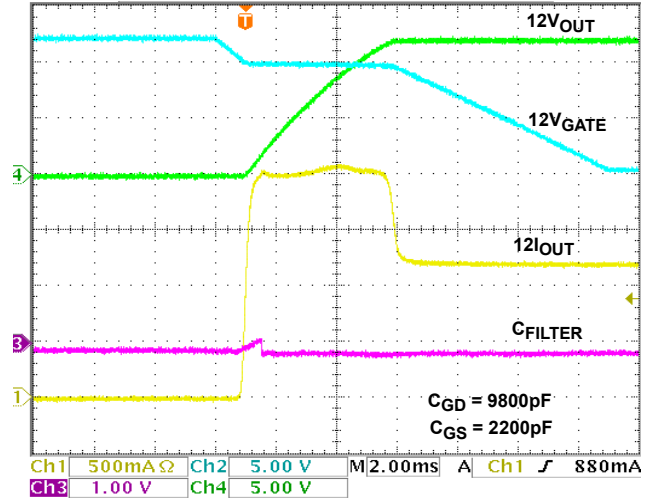


FIGURE 22. 12VMAIN START-UP  $R_{LOAD} = 10\Omega$ ,  $C_{LOAD} = 470\mu F$

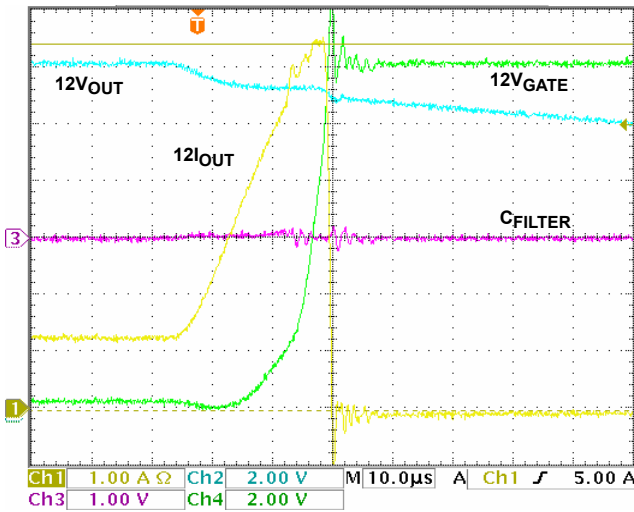


FIGURE 23. 12VMAIN WOC SHUT-DOWN

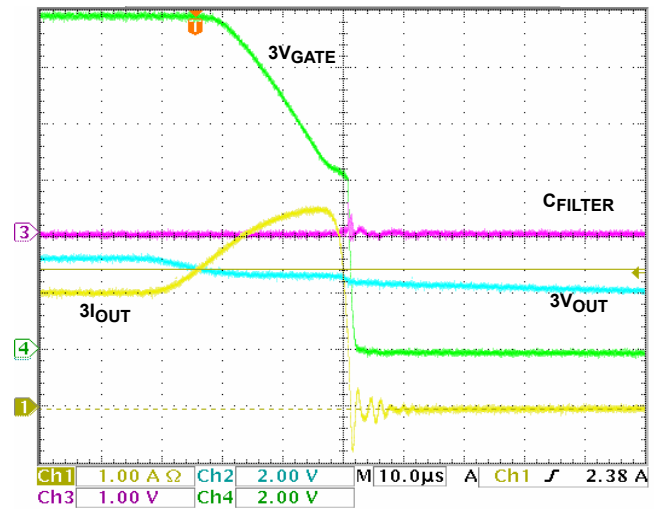


FIGURE 24. 3VMAIN WOC SHUT-DOWN

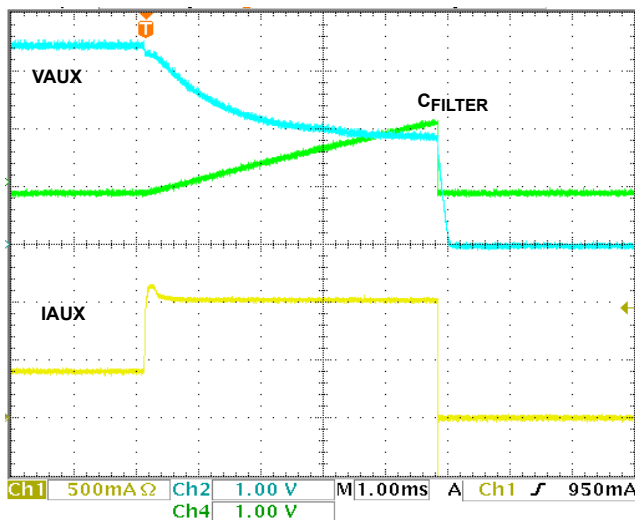


FIGURE 25. VAUX OC REGULATION AND SHUT-DOWN

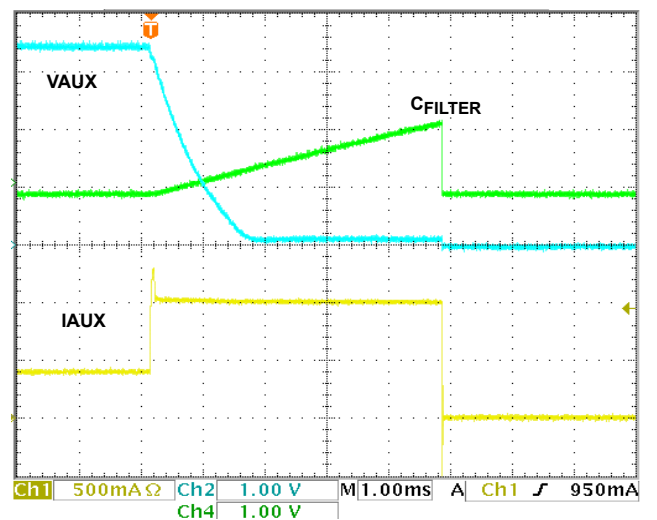
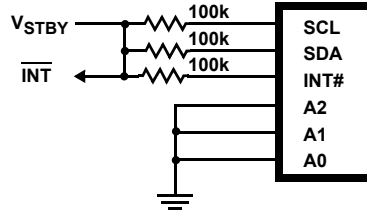
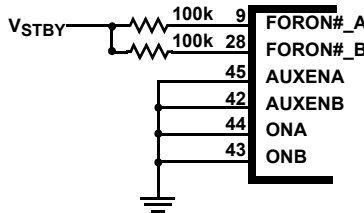


FIGURE 26. VAUX WOC REGULATION AND SHUT-DOWN



DISABLING SMI WHEN HPI CONTROL IS USED



DISABLING HPI WHEN SMI CONTROL IS USED

FIGURE 27. I/O CONFIGURATION FOR DISABLING HPI/SMI CONTROL

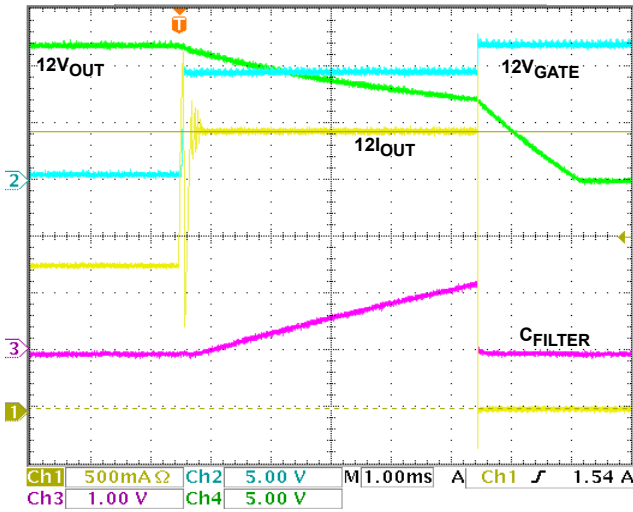


FIGURE 28. 12VMAIN CR AND SHUT-DOWN

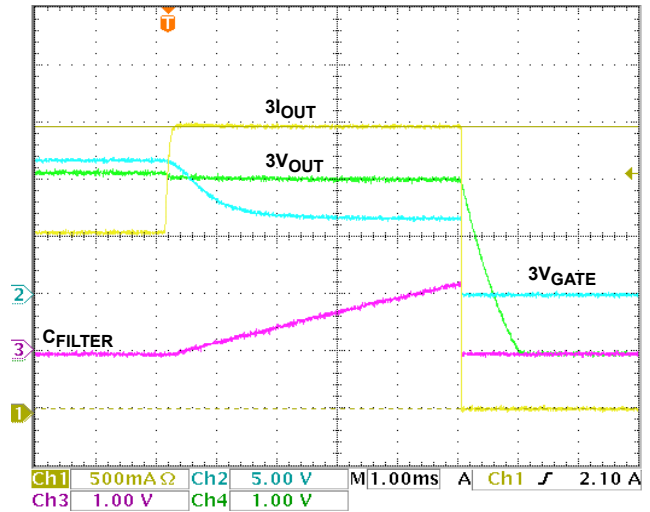


FIGURE 29. 3VMAIN CR AND SHUT-DOWN

### ISL6112 Bias, Power-On Reset, and Power Cycling

The ISL6112 uses VSTBY as the only supply source. VSTBY is required for proper operation of the ISL6112 SMBus and registers and must be applied at all times. A Power-On-Reset (POR) cycle is initiated after VSTBY rises above its UVLO threshold and remains satisfied for 250µs. All internal registers are cleared after POR. If VSTBY is recycled, the ISL6112 enters a new POR cycle. VSTBY must be the first supply voltage applied, followed by the MAIN supply inputs of 12VIN and 3VIN. The SMBus is ready for access at the end of the POR cycle (250µs after VSTBY is valid). During t<sub>POR</sub>, all outputs remain off.

### Enabling the VAUX Outputs

Upon asserting an AUXEN input, the related internal power switch turns on, connecting the nominally 3.3V VSTBY supply to its VAUX load. The turn-on is slew rate limited and invokes the current

regulation feature so as not to droop the supply due to in-rush current load. Figure 17 illustrates the VAUX turn-on performance into a 10Ω, 100µF load.

### Standby Mode

Standby mode is entered when one or more of the MAIN supply inputs (12VIN and/or 3VIN) is either absent, below its respective UVLO threshold, or OFF. The ISL6112 also has 3.3V auxiliary outputs (VAUXA and VAUXB), satisfying an optional PCI-Express requirement. These outputs are fed from the VSTBY input pins. They are independent of the MAIN outputs and are controlled by the AUXEN input pins or via their respective bits in the control registers. Should the MAIN supply inputs fall below their respective UVLO thresholds, VAUX still functions, as long as VSTBY is compliant. Prior to standby mode, ONA and ONB (or the control register MAINA and MAINB bits) inputs must be deasserted, or the ISL6112 asserts its FAULT outputs. If an



undervoltage condition on either of the MAIN supply inputs is detected,  $\overline{\text{INT}}$  also asserts, if interrupts are enabled.

## Enabling the MAIN GATE Outputs

When a slot's MAIN supplies are off, the 12VGATE pin is held high with an internal pull-up to the 12VIN voltage. Similarly, the 3VGATE pin is internally held low to GND. When the MAIN supplies of the ISL6112 are enabled by asserting ON, the related 3VGATE and 12VGATE pins are each connected to a constant current supply. For the 3VGATE pin, this supply is nominally a 25 $\mu$ A current source. For the 12VGATE pin, the supply is nominally a -25 $\mu$ A current sink. The 3VGATE is charged up to the 12VIN voltage, while the 12GATE is pulled down to GND, for maximum enhancement of the N-Channel and P-Channel FETs, respectively.

## Estimating In-Rush Current and V<sub>OUT</sub> Slew Rate at Start-Up

The expected in-rush current can be estimated by using Equation 1:

$$I_{\text{IN-RUSH}} \text{ NOMINALLY} = 25\mu\text{A} \left( \frac{C_{\text{LOAD}}}{C_{\text{GATE}}} \right) \quad (\text{EQ. 1})$$

where 25 $\mu$ A is the GATE pin charge current, and  $C_{\text{LOAD}}$  is the load capacitance.  $C_{\text{GATE}}$  is the total GATE capacitance, including  $C_{\text{ISS}}$  of the external MOSFET and any external capacitance connected from the GATE output pin to the GATE reference, GND, or source.

An estimate for the output slew rate of 3.3V outputs and 12V outputs, where there is little or no external 12VGATE output capacitors, can be determined from Equation 2:

$$dv/dt \text{ VOUT NOMINALLY} = \frac{I_{\text{LIM}}}{C_{\text{LOAD}}} \quad (\text{EQ. 2})$$

where  $I_{\text{LIM}} = 50\text{mV}/R_{\text{SENSE}}$  and  $C_{\text{LOAD}}$  is the load capacitance. As a consequence, the CR duration,  $t_{\text{FILTER}}$ , must be programmed to exceed the time it takes to fully charge the output load to the input rail voltage level.

## MAIN Outputs (Start-up Delay and Slew-Rate Control)

The 3.3V outputs act as source followers. In this mode of operation,  $V_{\text{SOURCE}} = [V_{\text{GATE}} - V_{\text{TH(ON)}}]$  until the associated output reaches 3.3V. The voltage on the gate of the MOSFET continues to rise until it reaches 12V, which ensures minimum  $r_{\text{DS(ON)}}$ . For the 12V outputs, the MOSFET can be optionally configured as a Miller integrator by adding a  $C_{\text{GD}}$  capacitor connected between the MOSFET's gate and drain to adjust the VOUT ramp time. In this configuration, the feedback action from drain to gate of the MOSFET causes the voltage at the drain of the MOSFET to slew in a linear fashion at a rate estimated by Equation 3:

$$dv/dt \text{ VOUT NOMINALLY} = \frac{25\mu\text{A}}{C_{\text{GD}}} \quad (\text{EQ. 3})$$

Table 1 approximates the output slew-rate for various values of  $C_{\text{GATE}}$  when start-up is dominated by GATE capacitance (external

$C_{\text{GATE}}$  from GATE pin to ground, plus  $C_{\text{GS}}$  of the external MOSFET for the 3.3V rail and  $C_{\text{GD}}$  for the 12V rail).

TABLE 1. 3.3V AND 12V OUTPUT SLEW-RATE SELECTION FOR GATE CAPACITANCE DOMINATED START-UP

IGATE   = 25 $\mu$ A	
$C_{\text{GATE}}$ or $C_{\text{GD}}$	dv/dt (load)
0.01 $\mu$ F (Note)	2.5V/ms
0.022 $\mu$ F (Note)	1.136V/ms
0.047 $\mu$ F	0.532 V/ms
0.1 $\mu$ F	0.250V/ms

NOTE: Values in this range are affected by the internal parasitic capacitances of the MOSFETs used, and should be verified experimentally.

Note that all of these performance estimates are useful only for first-order time and loading expectations, because they do not look at other significant loading factors. Figures 18, 19, 20, 21 and 22 illustrate empirically the discussed turn-on performance with the noted loading and compensation conditions.

Notice the degree of control over the in-rush current and the GATE ramp rate as the values are changed, which provides for highly customizable turn-on characteristics.

In some scope shots, although the  $C_{\text{FILTER}}$  shows ramping in the absence of excessive displayed loading current,  $C_{\text{FILTER}}$  is responding to the other MAIN supply current that is not displayed.

All scope shots were taken from the ISL6112EVAL1Z, with any component changes noted.

## Current Regulation (CR) Function

The ISL6112 provides a current regulation and limiting function that protects the input voltage supplies against excessive loads, including short circuits. When the current from any slot's MAIN outputs exceeds the current limit threshold ( $I_{\text{LIM}} = 50\text{mV}/R_{\text{SENSE}}$ ) for a duration greater than  $t_{\text{FILTER}}$ , the isolation protection is tripped, and both related MAIN supplies are shut off, as shown in Figures 28 and 29. Should the load current cause a MAIN output  $V_{\text{SENSE}}$  to exceed  $V_{\text{THFAST}}$ , the output is immediately shut off, with no  $t_{\text{FILTER}}$  delay, as shown in Figures 28 and 29.

The VAUX outputs have a different isolation protection function. The VAUX isolation circuit does not incorporate a fast-trip detector; instead, they regulate the output current into a fault to avoid exceeding their operating current limit. The protection circuit trips due to an overcurrent on VAUX when the programmable CR duration timer,  $t_{\text{FILTER}}$ , expires. This use of the  $t_{\text{FILTER}}$  timer prevents the circuit from tripping prematurely due to brief current transients. See Figures 25 and 26 for illustrations of the VAUX protection performance into a slight OC and more severe OC condition, respectively. The ISL6112 AUX current control responds proportionally to the severity of the OC condition, resulting in faster VAUX pull-down and current regulation until  $t_{\text{FILTER}}$  has expired.

Following a fault condition, the outputs can be turned on again (1) via the ON inputs if the fault occurred on one of the MAIN outputs, (2) via the AUXEN inputs if the fault occurred on the AUX outputs, or (3) by cycling both ON and AUXEN if faults occurred on



both the MAIN and AUX outputs. A fault condition can alternatively be cleared under SMI control of the ENABLE bits in the CNTRL registers (see “Control Register Bits D[1:0]” on page 21). When the circuit protection trips,  $\overline{\text{FAULT}}$  is asserted if the outputs were enabled through the HPI inputs. If SMI is enabled,  $\overline{\text{INT}}$  is asserted (unless interrupts are masked). Note that  $\overline{\text{INT}}$  is deasserted by writing a Logic 1 back into the respective fault bit positions in the STAT register or the “Common Status Register (CS) 8-Bits, Read/Write” on page 24.

The ISL6112 current regulation duration ( $t_{\text{FILTER}}$ ) is individually set for each slot by an external capacitor at the CFILTER pin to GND. Once the CR mode is entered, the external cap is charged with a 2.5 $\mu\text{A}$  current source to 1.25V. Once this threshold has been reached, the IC turns off only the related faulted outputs [either both MAIN (external FETs) outputs or AUX (internal FET)] and sets the  $\overline{\text{FAULT}}$  output low. For a desired  $t_{\text{FILTER}}$ , the value for  $C_{\text{FILTER}}$  is given by Equation 4:

$$C_{\text{FILTER}} = \frac{\text{nominal } t_{\text{FILTER}}}{500\text{k}\Omega} \quad (\text{EQ. 4})$$

where 500k $\Omega$  is the nominal  $V_{\text{FILTER}}/\text{nominal } I_{\text{FILTER}}$  and where  $t_{\text{FILTER}}$  is the desired response time, with the values for  $I_{\text{FILTER}}$  and  $V_{\text{FILTER}}$  being found in the “Electrical Specifications Table” on page 7. See Table 2 for nominal  $t_{\text{FILTER}}$  times for given  $C_{\text{FILTER}}$  cap values.

**TABLE 2. NOMINAL  $t_{\text{FILTER}}$  DURATION**

C <sub>FILTER</sub> CAPACITANCE ( $\mu\text{F}$ )	TIME (ms)
Open	0.01
0.01	5
0.022	11
0.047	24
0.1	50

NOTE: Nominal CR\_DUR =  $C_{\text{FILTER}}$  cap ( $\mu\text{F}$ ) \* 500k $\Omega$ .

Because the ISL6112 has its CR feature invoked as it turns on the FETs into the load,  $t_{\text{FILTER}}$  consideration is minimal. A maximum bulk capacitance is specified for each supported power level, and it must be charged at the CR limit. In-rush current time must be considered when determining  $t_{\text{FILTER}}$  duration.

## Power-Down Cycle

When a slot is turned off under either HPI or SMI control, internal discharge FETs connected to the output load provide a discharge path for load capacitance connected to the part's outputs. These FETs ensure the outputs are pulled to GND. This is a compliance requirement if a replacement add-in card is inserted into the slot.

## Thermal Shutdown

The internal VAUX MOSFETs are protected against damage not only by current limiting, but by a dual-mode over-temperature protection scheme as well. Each slot controller on the ISL6112 is thermally isolated from the other. Should an overcurrent condition raise the junction temperature of one slot's controller and pass elements to +140°C, all outputs for that slot (including VAUX) are shut off, and the slot's  $\overline{\text{FAULT}}$  output is asserted. The other slot's operating condition remains unaffected. However, should the ISL6112 die temperature exceed +160°C, both slots

(all outputs, including VAUXA and VAUXB) are shut off, whether or not a current limit condition exists. A +160°C over-temperature condition additionally sets the over-temperature bit (OT\_INT) in the Common Status Register (see “Common Status Register (CS) 8-Bits, Read/Write” on page 24).

## PWRGD Outputs

The ISL6112 has two  $\overline{\text{PWRGD}}$  outputs; one for each slot. These open-drain, active-low outputs require an external pull-up resistor to  $V_{\text{STBY}}$ . Each output is asserted when a slot has been enabled and has successfully begun delivering power to its respective +12V, +3.3V, and VAUX outputs. An equivalent logic diagram for  $\overline{\text{PWRGD}}$  is shown in Figure 30.

## FORCE\_ON Inputs

The level-sensitive, active-low  $\overline{\text{FORCE\_ON}}$  inputs are provided to facilitate design or debugging of systems using the ISL6112. Asserting  $\overline{\text{FORCE\_ON}}$  turns on all three of the respective slot's outputs (12MAIN, 3MAIN, and VAUX), while specifically defeating all overcurrent and short circuit protections and on-chip thermal protection for the VAUX supplies. Additionally, asserting  $\overline{\text{FORCE\_ON}}$  disables all input and output UVLO protections, with the exception of the VSTBY input, UVLO.

Asserting  $\overline{\text{FORCE\_ON}}$  causes the slot  $\overline{\text{PWRGD}}$  and  $\overline{\text{FAULT}}$  outputs to enter the open-drain state. Additionally, there are two SMBus accessible register bits (see Control Register Bit D[2] in Tables 5 and 7) that can be set to disable the corresponding slot's  $\overline{\text{FORCE\_ON}}$  pins. This allows system software to prevent these hardware overrides from being inadvertently activated during normal use. When not used, each  $\overline{\text{FORCE\_ON}}$  pin can be connected to  $V_{\text{STBY}}$  by using an external pull-up resistor, or it can simply be shorted to VSTBY.

## General Purpose Input (GPI) Pins

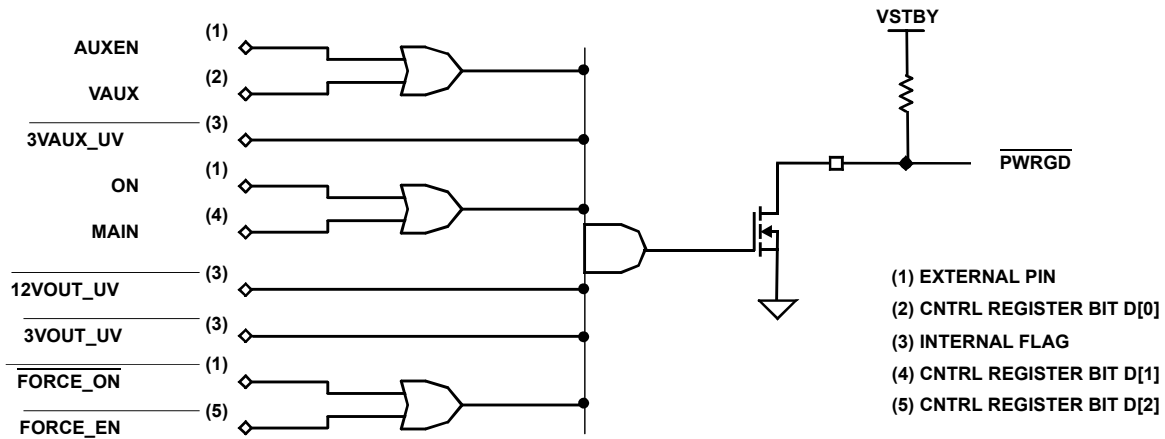
Two pins on the ISL6112 are available for use as GPI pins. The logic state of each of these pins can be determined by polling Bits [4:5] of the Common Status Register. Both of these inputs are compliant to 3.3V. If GPI is not used, each input must be connected to GND.

## Hot-Plug Interface (HPI)

After the input supplies are above their respective UVLO thresholds, the Hot-Plug Interface can be used for power control by enabling the control input pins (AUXEN and ON) for each slot. For the ISL6112 to turn on the VAUX supply for either slot, the AUXEN control must be enabled after the power-on-reset delay,  $t_{\text{POR}}$  (typically, 250 $\mu\text{s}$ ), has elapsed.

## System Management Interface (SMI)

The ISL6112 System Management Interface (SMI) uses the Read\_Byte and Write\_Byte subsets of the SMBus protocols to communicate with its host via the SMI bus. The  $\overline{\text{INT}}$  output signals the controlling processor that one or more events need attention, if an interrupt-driven architecture is used. Note that the ISL6112 does not participate in the SMBus Alert Response Address (ARA) portion of the SMBus protocol.

FIGURE 30.  $\overline{\text{PWRGD}}$  LOGIC DIAGRAM

## Fault Reporting and Interrupt Generation

### SMI-ONLY CONTROL APPLICATIONS

For applications in which the ISL6112 is controlled only by SMI, ON and AUXEN are connected to GND, and the  $\overline{\text{FORCE\_ON}}$  pins are either shorted or are connected to  $V_{\text{STBY}}$  as shown in Figure 27. In these cases, the ISL6112 FAULT outputs and STATUS Register Bit D[7] (FAULT) are not activated, because fault status is determined by polling STATUS Register Bits D[4], D[2], D[0] and CS (Common Status) Register Bits D[2:1]. Individual fault bits in the STATUS and CS registers are asserted after POR, when:

- Either or both CNTRL Register Bits D[1:0] are asserted, AND
- 12VIN, 3VIN, or VSTBY input voltage is lower than its respective ULVO threshold, OR
- The fast OC circuit isolation protection has tripped, OR
- The slow OC circuit isolation protection has tripped AND its filter time-out has expired, OR
- The slow OC circuit isolation protection has tripped AND Slot die temperature > +140°C, OR
- The ISL6112 global die temperature > +160°C

Once asserted, to clear any one or all STATUS Register Bits D[4], D[2], D[0] or CS Register Bits D[2], D[1], a software subroutine can perform an “echo reset” in which a Logical “1” is written back to those register bit locations that have indicated a fault. This method of “echo reset” allows data to be retained in the STATUS and/or CS registers until such time as the system is prepared to operate on that data.

The ISL6112 can operate in interrupt mode or polled mode. For interrupt-mode operation, the open-drain, active-LOW  $\overline{\text{INT}}$  output signal is activated after POR if the INTMSK bit (CS Register Bit D[3]) has been reset to Logical “0”. Once activated, the  $\overline{\text{INT}}$  output is asserted by any one of the fault conditions previously listed. It is deasserted when one or all STATUS Register Bits D[4], D[2], D[0] or CS Register Bits D[2], D[1] are reset upon the execution of an SMBus “echo reset” WRITE\_BYTE cycle. For polled-mode operation, the INTMSK bit should be set to Logical “1,” thereby inhibiting  $\overline{\text{INT}}$  output pin operation.

For SMI control applications in which the  $\overline{\text{FORCE\_ON}}$  inputs are needed for diagnostic purposes, the  $\overline{\text{FORCE\_ON}}$  inputs must be enabled; that is, CNTRL Register Bit D[2] should read Logical “0.” Once  $\overline{\text{FORCE\_ON}}$  inputs are asserted, all output voltages are present with all circuit protection features disabled, including over-temperature protection on VAUX outputs. To inhibit  $\overline{\text{FORCE\_ON}}$  operation, a Logical “1” is written to the CNTRL Register Bit D[2] locations.

### HPI-ONLY CONTROL APPLICATIONS

For applications in which the ISL6112 is controlled only by HPI, SMBus signals SCL, SDA, and  $\overline{\text{INT}}$  are connected to  $V_{\text{STBY}}$  as shown in Figure 27. In this configuration, the ISL6112 FAULT outputs are activated after POR and become asserted when:

Either or both external ON and AUXEN input signals are asserted, AND

- 12VIN, 3VIN, or VSTBY input voltage is lower than its respective ULVO threshold, OR
- The fast OC circuit isolation protection has tripped, OR
- The slow OC circuit isolation protection has tripped AND its filter time-out has expired, OR
- The slow OC circuit isolation protection has tripped AND slot die temperature > +140°C, OR
- The ISL6112 global die temperature > +160°C.

To clear FAULT outputs, once asserted, either or both ON and AUXEN input signals must be deasserted. (See FAULT pin in “Pin Descriptions” table on page 4 for additional information.) If the  $\overline{\text{FORCE\_ON}}$  inputs are used for diagnostic purposes, both the FAULT and  $\overline{\text{PWRGD}}$  outputs are deasserted after the  $\overline{\text{FORCE\_ON}}$  inputs are asserted.

### Serial Port Operation

The ISL6112 uses standard SMBus Write\_Byte and Read\_Byte operations for communication with its host. The SMBus Write\_Byte operation involves sending the device’s target address, with the R/W bit (LSB) set to the low (write) state, followed by a command byte and a data byte. The SMBus Read\_Byte operation is similar, but it is a composite write and read operation. The host first sends the device’s target address, followed by the command byte, as in a write operation. A new

“Start” bit must then be sent to the ISL6112, followed by a repeat of the device address, with the R/W bit set to the high (read) state. The data to be read from the part may then be clocked out. The exception to this rule is that, if the location latched in the pointer register from the last write operation is known to be correct (i.e., points to the desired register within the ISL6112), the “Receive\_Byte” procedure may be used. To perform a Receive\_Byte operation, the host sends an address byte to select the target ISL6112, with the R/W bit set to the high (read) state, and then retrieves the data byte. Figures 33, 34 and 35 show the formats for these data read and data write procedures.

The Command Register is eight bits (one byte) wide. This byte carries the address of the ISL6112 register to be operated upon. Command byte values corresponding to the ISL6112 register addresses are shown in Table 4. Command byte values other than 0000 0XXX<sub>b</sub> = 00<sub>h</sub> – 07<sub>h</sub> are reserved and should not be used.

### ISL6112 SMBus Address Configuration

The ISL6112 responds to its own unique SMBus address, which is assigned using A2, A1, and A0. These represent the three Least Significant Bits (LSB) of its 7-bit address, as shown in Table 3. These address bits are assigned only during power-up of the VSTBY supply input. These address bits allow up to eight ISL6112 devices in a single system. These pins are either grounded or left unconnected to specify a logical 0 or logical 1, respectively. A pin designated as a logical 1 may also be pulled up to VSTBY.

TABLE 3. ISL6112 SMBUS ADDRESSING

INPUTS			ISL6112 DEVICE ADDRESS	
A2	A1	A0	BINARY	HEX
0	0	0	1000 000X*b	80h
0	0	1	1000 001Xb	82h
0	1	0	1000 010Xb	84h
0	1	1	1000 011Xb	86h
1	0	0	1000 100Xb	88h
1	0	1	1000 101Xb	8Ah
1	1	0	1000 110Xb	8Ch
1	1	1	1000 111Xb	8Eh

\* Where X = “1” for READ and “0” for WRITE

# Timing Diagrams

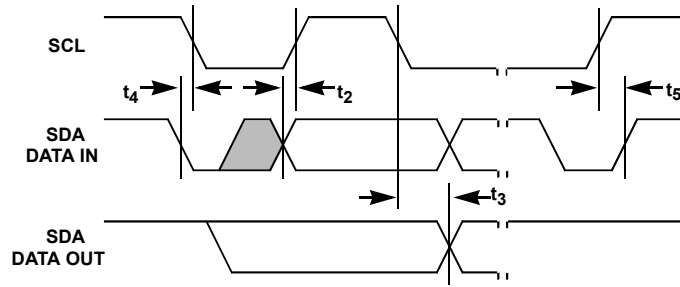


FIGURE 31. SMBUS TIMING

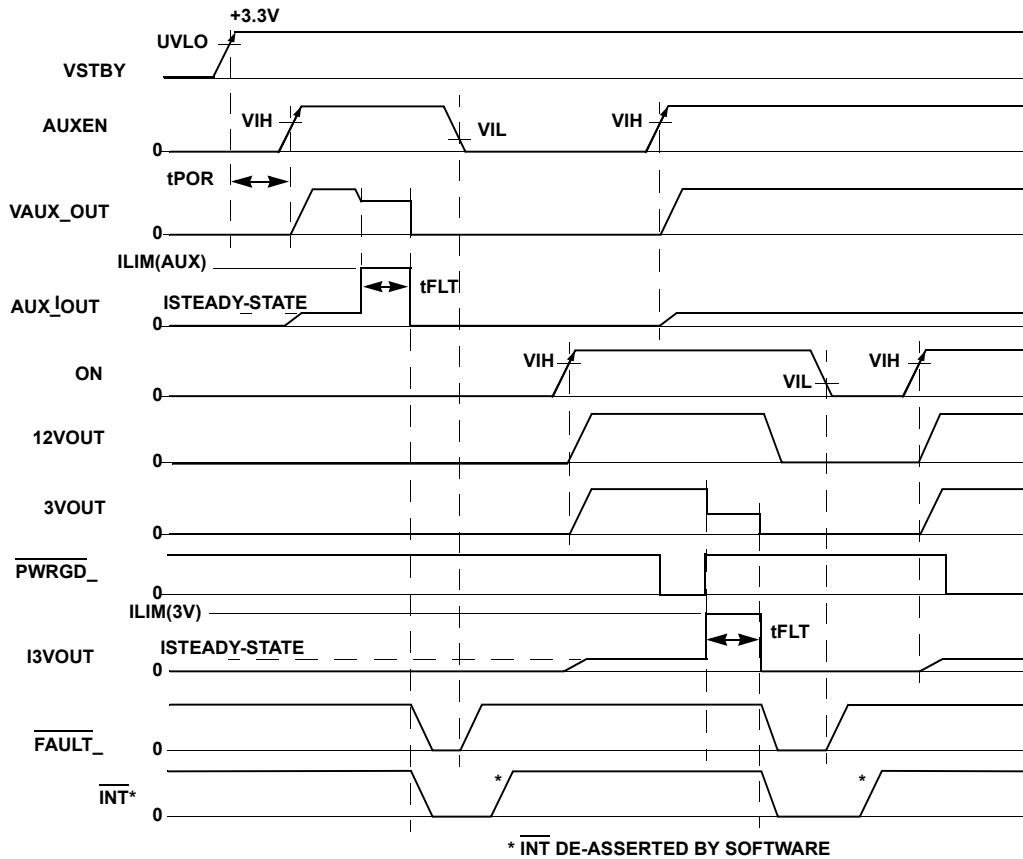


FIGURE 32. HOT-PLUG INTERFACE OPERATION

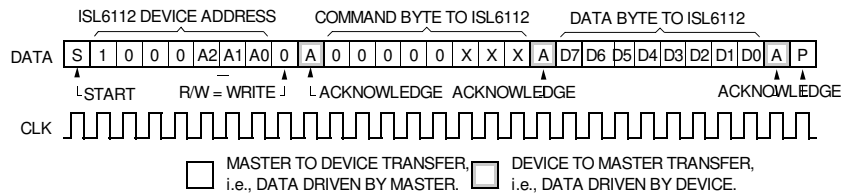


FIGURE 33. WRITE\_BYTE PROTOCOL

## Timing Diagrams (Continued)

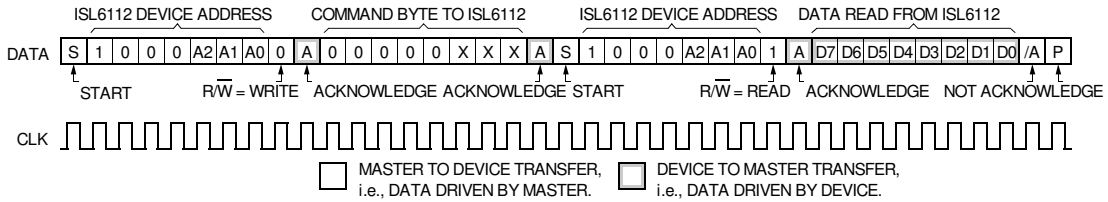


FIGURE 34. READ\_BYTE PROTOCOL

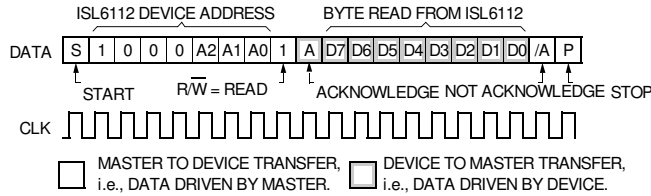


FIGURE 35. RECEIVE\_BYTE PROTOCOL

## ISL6112 Register Set and Programmer’s Model

TABLE 4. ISL6112 REGISTER ADDRESSES

TARGET REGISTER		COMMAND BYTE VALUE		POWER-ON DEFAULT
LABEL	DESCRIPTION	READ	WRITE	
CNTRLA	Control Register Slot A	02h	02h	00h
CNTRLB	Control Register Slot B	03h	03h	00h
STATA	Slot A Status	04h	04h	00h
STATB	Slot B Status	05h	05h	00h
CS	Common Status Register	06h	06h	xxxx 0000b
Reserved	Reserved/Do Not Use	07h - FFh	07h - FFh	Undefined

### Detailed Register Descriptions

#### Control Register, Slot A (CNTRLA) 8-Bits, Read/Write

TABLE 5. CONTROL REGISTER, SLOT A (CNTRLA)

D[7]	D[6]	D[5]	D[4]	D[3]	D[2]	D[1]	D[0]
read-only	read-only	read-only	read-only	read-only	read/write	read/write	read/write
AUXAPG	MAINAPG	Reserved	Reserved	Reserved	FORCE_A ENABLE	MAINA	VAUXA

BIT(s)	FUNCTION	OPERATION
AUXAPG	AUX output power-good status, Slot A	1 = Power-is-Good (VAUXA Output is above its UVLO threshold)
MAINAPG	MAIN output power-good status, Slot A	1 = Power-is-Good (MAINA Outputs are above their UVLO thresholds)
D[5]	Reserved	Always read as zero
D[4]	Reserved	Always read as zero

BIT(s)	FUNCTION	OPERATION
D[3]	Reserved	Always read as zero
$\overline{\text{FORCE\_A}}$ ENABLE	Allows or inhibits the operation of the $\overline{\text{FORCE\_ONA}}$ input pin	0 = $\overline{\text{FORCE\_ONA}}$ is enabled 1 = $\overline{\text{FORCE\_ONA}}$ is disabled
MAINA	MAIN enable control, Slot A	0 = OFF, 1 = ON
VAUXA	VAUX enable control, Slot A	0 = OFF, 1 = ON

Power-Up Default Value: 0000 0000<sub>b</sub> = 00<sub>h</sub>  
 Read Command\_Byte Value (R/W): 0000 0010<sub>b</sub> = 02<sub>h</sub>

**The power-up default value is 00<sub>h</sub>. Slot is disabled upon power-up; i.e., all supply outputs are off.**

**NOTES:**

- The state of the  $\overline{\text{PWRGDA}}$  pin is the logical AND of the values of the AUXAPG and the MAINAPG bits, except when  $\overline{\text{FORCE\_ONA}}$  is asserted. If  $\overline{\text{FORCE\_ONA}}$  is asserted (the pin is pulled low), and  $\overline{\text{FORCE\_AENABLE}}$  is set to a logic zero, the  $\overline{\text{PWRGDA}}$  pin will be unconditionally forced to its open-drain ("Power Not Good") state.
- The values of the MAINAPG and AUXAPG register bits are not affected by  $\overline{\text{FORCE\_ONA}}$ , but will instead continue to read as high if power is "Good," and as low if the conditions that indicate power is good are not met.

## Status Register Slot A (STATUSA) 8-Bits, Read-Only

TABLE 6. STATUS REGISTER, SLOT A (STATA)

D[7]	D[6]	D[5]	D[4]	D[3]	D[2]	D[1]	D[0]
read-only	read-only	read-only	read/write	read-only	read/write	read-only	read/write
FAULTA	MAINA	VAUXA	VAUXAF	Reserved	12VAF	Reserved	3VAF

BIT(s)	FUNCTION	OPERATION
FAULTA	FAULT Status - Slot A	1 = Fault pin asserted ( $\overline{\text{FAULTA}}$ pin is LOW); 0 = Fault pin deasserted ( $\overline{\text{FAULTA}}$ pin is HIGH). See Notes 10, 11, and 12.
MAINA	MAIN Enable Status - Slot A	Represents the actual state (on/off) of the two Main Power outputs for Slot A (+12V and +3.3V). 1 = Main Power ON; 0 = Main Power OFF
VAUXA	VAUX Enable Status - Slot A	Represents the actual state (on/off) of the Auxiliary Power output for Slot A. 1 = AUX Power ON; 0 = AUX Power OFF
VAUXAF	Overcurrent Fault: VAUXA supply	1 = Fault; 0 = No fault
D[3]	Reserved	Always read as zero
12VAF	Overcurrent Fault: +12V supply	1 = Fault; 0 = No fault
D[1]	Reserved	Always read as zero
3VAF	Overcurrent Fault: 3.3V supply	1 = Fault; 0 = No fault

BIT(s)	FUNCTION	OPERATION
Power-Up Default Value:	0000 0000 <sub>b</sub> = 00 <sub>h</sub>	
Command_Byte Value (R/W):	0000 0100 <sub>b</sub> = 04 <sub>h</sub>	

The power-up default value is 00<sub>h</sub>. Both slots are disabled upon power-up; i.e., all supply outputs are off. In response to an overcurrent fault condition, writing a Logical 1 back into the active (or set) bit position will clear the bit and deassert INT. The status of the FAULTA pin is not affected by reading the Status Register or by clearing active status bits.

NOTES:

- If FAULTA has been set by an overcurrent condition on one or more of the MAIN outputs, the ONA input must go LOW to reset FAULTA. If FAULTA has been set by a VAUXA overcurrent event, the AUXENA input must go LOW to reset FAULTA. If an overcurrent has occurred on both a MAIN output and the VAUX output of slot A, both ONA and AUXENA of the slot must go low to reset FAULTA.
- Neither the FAULTA bits nor the FAULTA pins are active when the ISL6112 power paths are controlled by SMI. When using SMI power path control, AUXENA and ONA pins for that slot must be tied to GND.
- If FORCE\_ONA is asserted (low), the FAULTA pin will be unconditionally forced to its open-drain state. Note that the value in the FAULTA register bit is not affected by FORCE\_ONA. It will instead continue to read as a high if no faults are present on Slot A, and as a low if any fault conditions exist that would disable Slot A if FORCE\_ONA was not asserted.

## Control Register, Slot B (CNTRLB) 8-Bits, Read/Write

TABLE 7. CONTROL REGISTER, SLOT B (CNTRLB)

D[7]	D[6]	D[5]	D[4]	D[3]	D[2]	D[1]	D[0]
read-only	read-only	read-only	read-only	read-only	read/write	read/write	read/write
AUXBPG	MAINBPG	Reserved	Reserved	Reserved	FORCE_B ENABLE	MAINB	VAUXB

BIT(s)	FUNCTION	OPERATION
AUXBPG	AUX output power-good status, Slot B	1 = Power-is-Good (VAUXB Output is above its UVLO threshold)
MAINBPG	MAIN output power-good status, Slot B	1 = Power-is-Good (MAINB Outputs are above their UVLO thresholds)
D[5]	Reserved	Always read as zero
D[4]	Reserved	Always read as zero
D[3]	Reserved	Always read as zero
FORCE_B ENABLE	Allows or inhibits the operation of the FORCE_ONB input pin	0 = FORCE_ONB is enabled 1 = FORCE_ONB is disabled
MAINB	MAIN enable control, Slot B	0 = OFF, 1 = ON
VAUXB	VAUX enable control, Slot B	0 = OFF, 1 = ON

Power-Up Default Value: 0000 0000<sub>b</sub> = 00<sub>h</sub>  
 Command\_Byte Value (R/W): 0000 0011<sub>b</sub> = 03<sub>h</sub>

The power-up default value is 00<sub>h</sub>. Slot is disabled upon power-up; i.e., all supply outputs are off.

NOTES:

- The state of the PWRGDB pin is the logical AND of the values of the AUXBPG and MAINBPG bits, except when FORCE\_ONB is asserted. If FORCE\_ONB is asserted (the pin is pulled low), and FORCE\_BENABLE is set to a logic zero, the PWRGDB pin will be unconditionally forced to its open-drain ("Power Not Good") state.
- The values of the MAINBPG and AUXBPG register bits are not affected by FORCE\_ONB, but will instead continue to read as high if power is "Good," and as low if the conditions, which indicate that power is good, are not met.

## Status Register Slot B (STATB) 8-Bits, Read-Only

TABLE 8. STATUS REGISTER, SLOT B (STATB)

D[7]	D[6]	D[5]	D[4]	D[3]	D[2]	D[1]	D[0]
read-only	read-only	read-only	read/write	read-only	read/write	read-only	read/write
FAULTB	MAINB	VAUXB	VAUXBF	Reserved	12VBF	Reserved	3VBF

BIT(s)	FUNCTION	OPERATION
FAULTB	FAULT Pin Status - Slot B	1 = Fault pin asserted ( $\overline{\text{FAULTB}}$ pin is LOW); 0 = Fault pin deasserted ( $\overline{\text{FAULTB}}$ pin is HIGH). See Notes 15, 16, and 17.
MAINB	MAIN Enable Status - Slot B	Represents the actual state (on/off) of the four Main Power outputs for Slot B (+12V and +3.3V): 1 = MAIN Power ON 0 = MAIN Power OFF
VAUXB	VAUX Enable Status - Slot B	Represents the actual state (on/off) of the Auxiliary Power output for Slot B: 1 = AUX Power ON 0 = AUX Power OFF
VAUXBF	Overcurrent Fault: VAUXB supply	1 = Fault; 0 = No fault
D[3]	Reserved	Always read as zero
12VBF	Overcurrent Fault: +12V supply	1 = Fault; 0 = No fault
D[1]	Reserved	Always read as zero
3VBF	Over current Fault: 3.3V supply	1 = Fault; 0 = No fault

Power-Up Default Value: 0000 0000h = 00h

Command\_Byte Value (R/W): 0000 0101<sub>b</sub> = 05<sub>h</sub>

**The power-up default value is 00<sub>h</sub>. Both slots are disabled upon power-up; i.e., all supply outputs are off. In response to an overcurrent fault condition, writing a logical 1 back into the active (or set) bit position will clear the bit and deassert INT. The status of the  $\overline{\text{FAULTB}}$  pin is not affected by reading the Status Register or by clearing active status bits.**

### NOTES:

- If FAULTB has been set by an overcurrent condition on one or more of the MAIN outputs, the ONB input must go LOW to reset FAULTB. If FAULTB has been set by a VAUXB overcurrent event, the AUXENB input must go LOW to reset FAULTB. If an overcurrent has occurred on both a MAIN output and the VAUX output of Slot B, both ONB and AUXENB of the slot must go low to reset FAULTB.
- Neither the FAULTB bits nor the  $\overline{\text{FAULTB}}$  pins are active when the ISL6112 power paths are controlled by SMI. When using SMI power path control, the AUXENB and ONB pins for that slot must be tied to GND.
- If  $\overline{\text{FORCE\_ONB}}$  is asserted (low), the  $\overline{\text{FAULTB}}$  pin will be unconditionally forced to its open-drain state. Note that the value in the FAULTB register bit is not affected by  $\overline{\text{FORCE\_ONB}}$ , but will instead continue to read as a high if no faults are present on Slot B, and as a low if any fault conditions exist that would disable Slot B if  $\overline{\text{FORCE\_ONB}}$  was not asserted.

## Common Status Register (CS) 8-Bits, Read/Write

TABLE 9. COMMON STATUS REGISTER (CS)

D[7]	D[6]	D[5]	D[4]	D[3]	D[2]	D[1]	D[0]
read-write	read-write	read-only	read-only	read-write	read-write	read-write	read-only
Reserved	Reserved	GPI_B0	GPI_A0	INTMSK	UV_INT	OT_INT	Reserved

BIT(s)	FUNCTION	OPERATION
D[7]	Reserved	Always read as zero
D[6]	Reserved	Always read as zero



BIT(s)	FUNCTION	OPERATION
GPI_B0	General Purpose Input 0, Slot B	State of GPI_B0 pin
GPI_A0	General Purpose Input 0, Slot A	State of GPI_A0 pin
INTMSK	Interrupt Mask	0 = $\overline{\text{INT}}$ generation is enabled 1 = $\overline{\text{INT}}$ generation is disabled The ISL6112 does not participate in the SMBus Alert Response Address (ARA) protocol.
UV_INT	Undervoltage Interrupt	0 = No UVLO fault 1 = UVLO fault Set whenever a circuit isolation protection fault condition occurs as a result of an undervoltage lockout condition on one of the main supply inputs. This bit is only set if a UVLO condition occurs while the ON pin is asserted or the MAIN control bits are set.
OT_INT	Over-temperature Interrupt	0 = Die Temp < +160 °C. 1 = Fault: Die Temp > +160 °C. Set if a fault occurs as a result of the ISL6112 die temperature exceeding +160 °C.
D[0]	Reserved	Undefined

Power-Up Default Value: 00000000<sub>b</sub> = 00<sub>h</sub>  
Command\_Byte Value (R/W): 00000110<sub>b</sub> = 06<sub>h</sub>

NOTE: To reset the OT\_INT and UV\_INT fault bits, a logical 1 must be written back to these bits.

## PCI-Express Application Recommendations

For each of the 3.3VMAIN and +12VMAIN supplies, the CR level is set by an external sense resistor value. This value depends on the maximum power specified for the PCI-Express connector and the application (X1, 10W or 25W; X4, X8, 25W; X16, 25W or 75W; and X16 Graphics -ATX, 150W). The power rating is a combination of the main and the optional auxiliary supplies. The sense resistor is a low ohmic, standard value current sense resistor (one for each slot). The voltage across this resistor is compared to a 50mV reference. The 3.3VMAIN supply is rated for 3A maximum across all slots, regardless of size and power. The use of a 15mΩ sense resistor compared to the 50mV reference provides a nominal CR of 3.3A, or 11% higher than the 3A maximum specification.

On the 12VMAIN, for a 10W connector, a 75mΩ sense resistor provides a nominal CR level of 0.66A, 32% above the 0.5A maximum specification. For a 25W connector, a 20mΩ sense resistor provides a nominal CR level of 2.5A, which is 19% above the 2.1A maximum specification. For a 75W connector, an 8mΩ sense resistor provides a nominal CR level of 6.25A, or 14% above the 5.5A maximum specification. The X16 Graphics-ATX 150W card is a special case. The 150W is provided by two slots, each providing up to a maximum of 75W from the 12VMAIN, as this type of card does not consume 3.3VMAIN or AUX supply power. For each of the slots, a 7mΩ sense resistor provides a nominal CR level of 7.1A, which is 14% above the 6.25A maximum specification.

The ISL6112 provides a best-in-class ±5% current regulation threshold specification over temperature for the MAIN supplies. This is the highest accuracy and lowest variability for this critical

parameter. Table 10 shows recommended 12VMAIN sense resistor values for particular power levels.

TABLE 10. NOMINAL CURRENT REGULATION LEVEL

12VMAIN R <sub>SENSE</sub> (mΩ)	12VMAIN CR (A)	PCI-E ADD-IN BOARD POWER LEVEL SUPPORTED (W)
75	0.7	10
20	2.5	25
8	6.2	75
7	7	150

NOTE: CR Level =  $V_{THLIMIT}/R_{SENSE}$ .

Providing a nominal CR protection level above the maximum specified limits of the card ensures that the card is able to draw its maximum specified loads. It also ensures enough headroom before a regulated current limiter is invoked to protect against transients and other events. This headroom margin can be adjusted up or down by using different sense resistor values.

## Using the ISL6112EVAL1Z Platform

The primary ISL6112 evaluation platform is shown photographically in Figure 36 and schematically on page 28. This evaluation board highlights a PCB layout that confines all necessary active and passive components in an area measuring 12mmx55mm. This width is smaller than the specified PCI-Express socket-to-socket spacing, allowing for intimate co-location of the load power control and the load itself.

Around the central highlighted layout are numerous labeled test points and configuration jumpers where there are node names such as A0(L/R). The pin name outside the parentheses relates to the ISL6112. The ISL6113 and ISL6114 also use this evaluation platform, as all three parts have a common pinout for the common pin functions. The pin names in parentheses are for the ISL6113 and ISL6114. The specific evaluation board ordered and received will reflect the part number in the area below the Intersil logo, either by label or silk-screened lettering. For pins that are not common across the ISL6112, ISL6113, and ISL6114, there is a matrix detailing the differences in the bottom left corner.

The ISL6112EVAL1Z is provided in HPI mode, with the clock shorted to ground.

The evaluation platform is to be biased through the six banana jacks: turn on the VSTBY supply first, and then the other MAIN supplies, in any order. With appropriate signaling to the AUXEN and ON inputs, the user should see turn-on waveforms. External current loading must be added to demonstrate OC and WOC response performance.

The SCL and SDA inputs in the top right quadrant of the evaluation board can be used to demonstrate SMI operation. The board's default address is configured as '000' via three jumpers located on the right side of the board and labeled A0, A1, and A2. The HPI inputs must be disabled as shown in Figure 27. If additional software is needed to configure and control, there is a LabView based program available from Intersil for demonstration of the ISL6112 functionality. User lab test hardware and instrument support is not available.

**CAUTION:** The ISL6112EVAL1Z gets very hot to the touch after operating for a few minutes. The hottest areas are marked on the evaluation board.

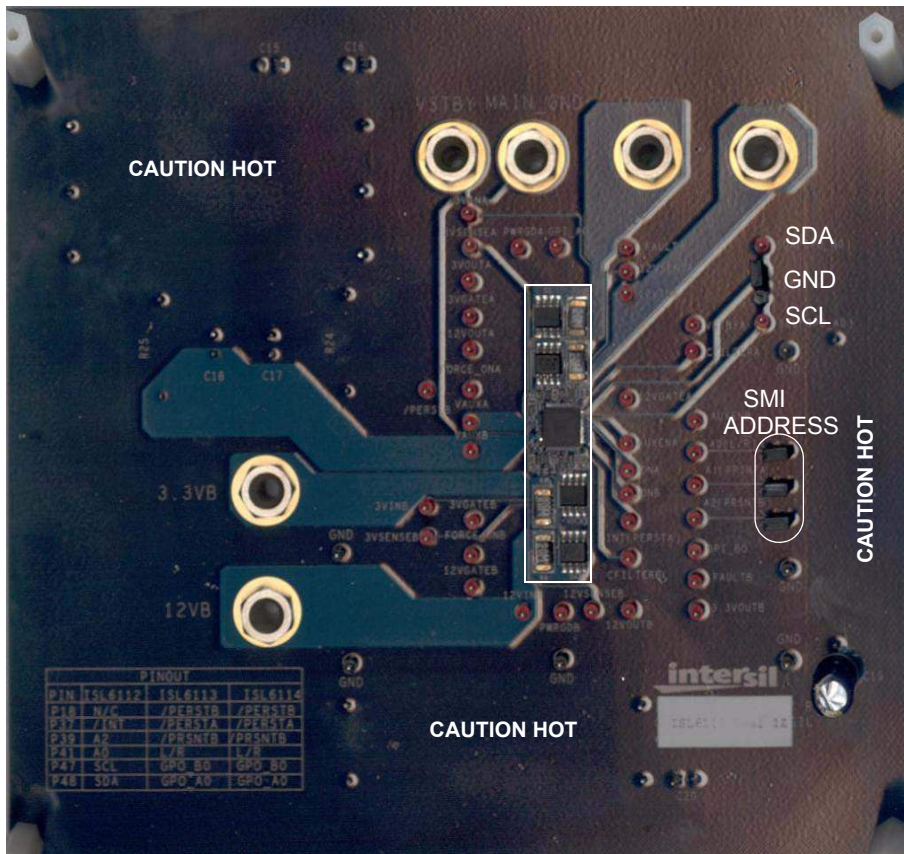
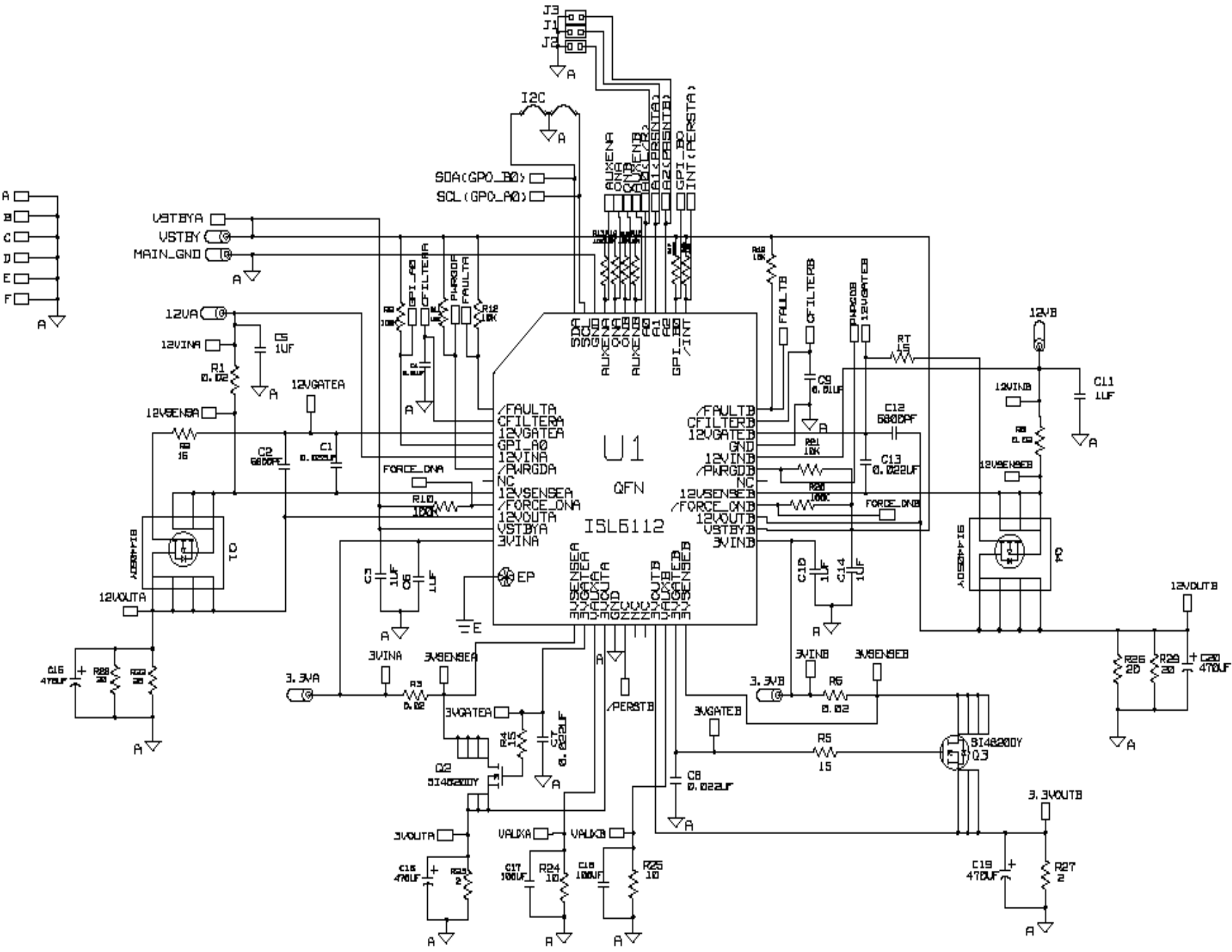


FIGURE 36. ISL6112EVAL1Z BOARD PHOTO

**TABLE 11. ISL6112EVAL1Z BOARD COMPONENTS LISTING**

COMPONENT DESIGNATOR	COMPONENT FUNCTION	COMPONENT DESCRIPTION
U1	ISL6112	PCI-Express Dual Slot Hot Plug Controller
Q1, Q4	Voltage Rail Switches	SI4405DY or equivalent, P-Channel MOSFET
Q2, Q3	Voltage Rail Switches	SI4820DY or equivalent, N-Channel MOSFET
R1, R3, R6, R8	Current Sense Resistor	0.020Ω 1%, 2512
R9, R10, R17, R20	Pull-up Resistors on FORCEON and GPI Inputs	100kΩ, 0201
R11, R12, R13, 14, R15, R16, R18, 19, R21	I/O Pull-up resistors	10kΩ, 0201
R2, R4, R5, R7	FET Gate Series Resistance	15Ω, 0201
C1, C7, C8, C13	FET Gate Capacitance	22nF 10%, 16V, 0402
C3, C5, C6, C10, C11, C14	MAIN and VSTBY Decoupling Capacitance	1μF 10%, 6.3V, 0402
C2, C12	P-FET Gate To Drain Capacitance	6.8nF 10%, 6.3V, 0201
C4, C9	CFILTER Capacitance (5ms)	0.01μF 10%, 6.3V, 0201
R24, R25	AUX Load Resistance	10W 20%, 3W
C17, C18	AUX Load Capacitance	100μF 20%, 25V, Radial Electrolytic
R22, R26, R28, 29	12MAIN Load Resistance	20W 20%, 10W
R23, R27	3MAIN Load Resistance	2W 20%, 10W
C15, C16, C19, C20	12MAIN and 3MAIN Load Capacitance	470μF 20%, 16V, Radial Electrolytic



## Revision History

The revision history provided is for informational purposes only and is believed to be accurate, but not warranted. Please go to web to make sure you have the latest revision.

DATE	REVISION	CHANGE
7/12/2011	FN6456.1	<ul style="list-style-type: none"> <li>Changed title from "Dual Slot PCI-Express Power Controller" to "Dual Slot PCI-Express Hot Plug Controller".</li> <li>Removed retired parts ISL6112INZA and ISL6112INZA-T* from Ordering Information. Removed package outline drawing, pin configuration diagram, and thermal information for TQFP package used for retired parts.</li> </ul>
9/30/2007	FN6456.0	Initial Release

## Products

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\*For a complete listing of Applications, Related Documentation and Related Parts, please see the respective device information page on intersil.com: [ISL21400](http://www.intersil.com/ISL21400)

To report errors or suggestions for this datasheet, please go to: [www.intersil.com/askourstaff](http://www.intersil.com/askourstaff)

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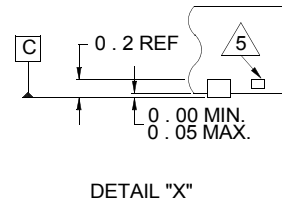
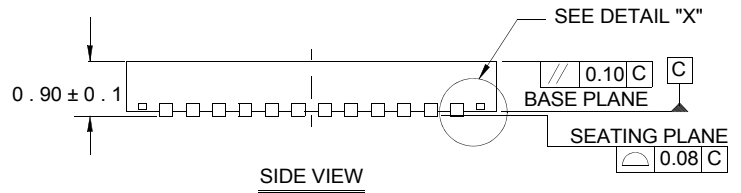
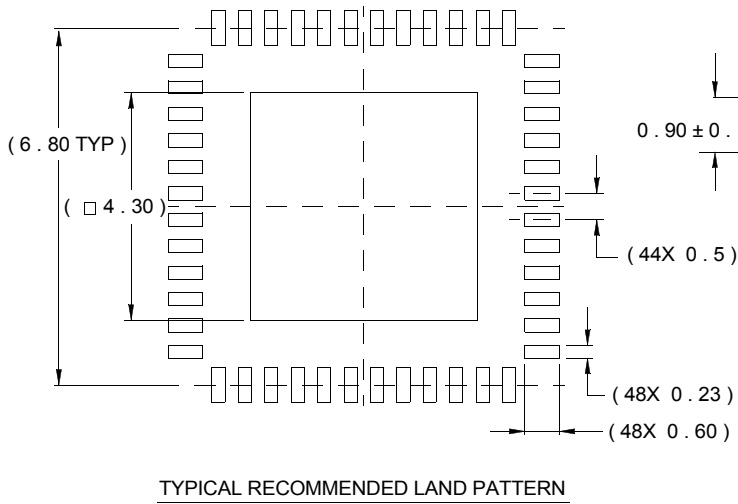
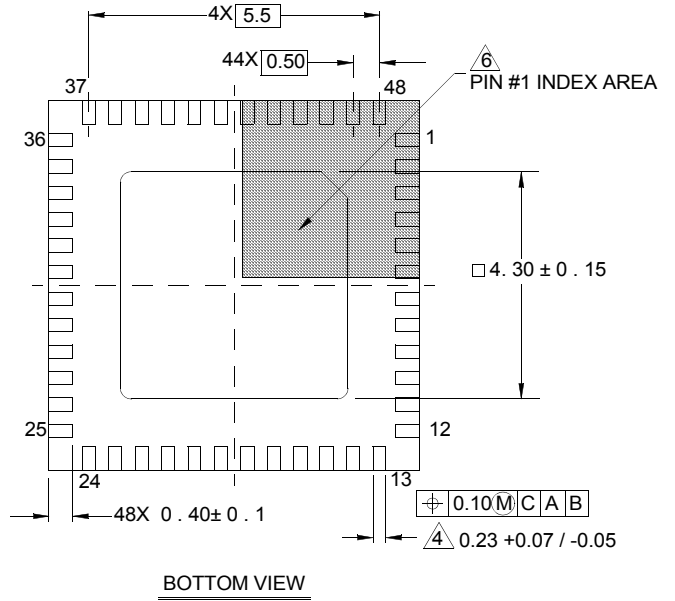
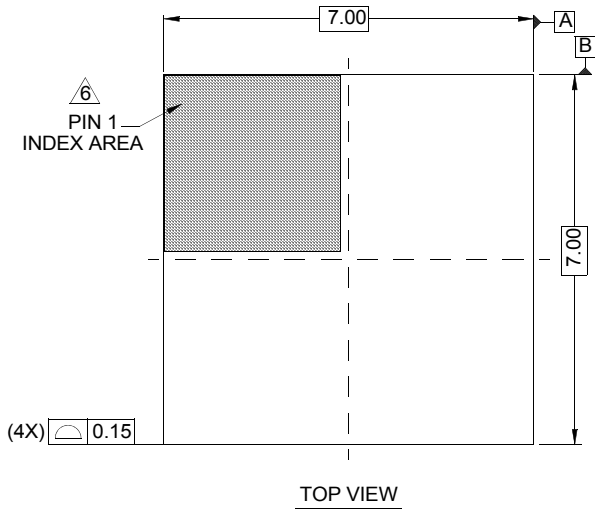
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# Package Outline Drawing

## L48.7x7

48 LEAD QUAD FLAT NO-LEAD PLASTIC PACKAGE

Rev 5, 4/10



NOTES:

1. Dimensions are in millimeters.  
Dimensions in ( ) for Reference Only.
2. Dimensioning and tolerancing conform to AMSE Y14.5m-1994.
3. Unless otherwise specified, tolerance : Decimal ± 0.05
4. Dimension applies to the metallized terminal and is measured between 0.15mm and 0.30mm from the terminal tip.
5. Tiebar shown (if present) is a non-functional feature.
6. The configuration of the pin #1 identifier is optional, but must be located within the zone indicated. The pin #1 identifier may be either a mold or mark feature.