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ON Semiconductor® FDB035AN06A0

N-Channel PowerTrench[®] MOSFET 60 V, 80 A, 3.5 m Ω

Features

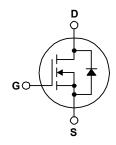
- $R_{DS(on)}$ = 3.2 m Ω (Typ.) @ V_{GS} = 10 V, I_{D} = 80 A
- $Q_{G(tot)}$ = 95 nC (Typ.) @ V_{GS} = 10 V
- · Low Miller Charge
- Low Q_{rr} Body Diode
- UIS Capability (Single Pulse and Repetitive Pulse)

Formerly developmental type 82584

Applications

- Synchronous Rectification for ATX / Server / Telecom PSU
- · Battery Protection Circuit
- Motor drives and Uninterruptible Power Supplies





MOSFET Maximum Ratings T_C = 25°C unless otherwise noted

Symbol	Parameter	FDB035AN06A0	Unit
V _{DSS}	Drain to Source Voltage	60	V
V _{GS}	Gate to Source Voltage	±20	V
	Drain Current		
	Continuous (T _C < 153°C, V _{GS} = 10V)	80	Α
ID	Continuous ($T_{amb} = 25^{\circ}C$, $V_{GS} = 10V$, with $R_{\theta JA} = 43^{\circ}C/W$)	22	Α
	Pulsed	Figure 4	А
E _{AS}	Single Pulse Avalanche Energy (Note 1)	625	mJ
D	Power dissipation	310	W
P_{D}	Derate above 25°C	2.07	W/°C
T _J , T _{STG}	Operating and Storage Temperature	-55 to 175	°C

Thermal Characteristics

$R_{\theta JC}$	Thermal Resistance, Junction to Case, Max.	0.48	°C/W
$R_{\theta JA}$	Thermal Resistance, Junction to Ambient, Max. (Note 2)	62	°C/W
$R_{\theta JA}$	Thermal Resistance, Junction to Ambient, 1in ² copper pad area, Max.	43	°C/W

Package	Marking	and	Ordering	Information
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Device Marking	Device	Package	Reel Size	Tape Width	Quantity
FDB035AN06A0	FDB035AN06A0	D²-PAK	330 mm	24 mm	800 units

Electrical Characteristics $T_C = 25^{\circ}C$ unless otherwise noted

Symbol	Parameter	Test Conditions	Min	Тур	Max	Unit
Off Characteristics						
B _{VDSS}	Drain to Source Breakdown Voltage	$I_D = 250 \mu A, V_{GS} = 0 V$	60	-	-	V
Ince IZero Gate Voltage Drain Current	V _{DS} = 50V	-	-	1	μА	
	$V_{GS} = 0V$ $T_C = 150^{\circ}C$	-	-	250	μΑ	
I _{GSS}	Gate to Source Leakage Current	$V_{GS} = \pm 20V$	-	-	±100	nA

On Characteristics

$V_{GS(TH)}$	Gate to Source Threshold Voltage	$V_{GS} = V_{DS}$, $I_D = 250\mu A$	2	-	4	V
r _{DS(ON)} Drain to Source On Resistance		I _D = 80A, V _{GS} = 10V	-	0.0032	0.0035	
	Drain to Source On Resistance	I _D = 40A, V _{GS} = 6V	-	0.0044	0.0066	0
	$I_D = 80A, V_{GS} = 10V,$ $T_J = 175$ °C	-	0.0065	0.0071	22	

Dynamic Characteristics

C _{ISS}	Input Capacitance	V 05V V 0V	-	6400	-	pF	
C _{OSS}	Output Capacitance	v _{DS} = 25v, v _{GS} : f = 1MHz	$V_{DS} = 25V, V_{GS} = 0V,$ f = 1MHz		1123	-	pF
C _{RSS}	Reverse Transfer Capacitance	1 - 1101112		-	367	-	pF
$Q_{g(TOT)}$	Total Gate Charge at 10V	V_{GS} = 0V to 10V			95	124	nC
$Q_{g(TH)}$	Threshold Gate Charge	$V_{GS} = 0V \text{ to } 2V$ $V_{DD} = 30V$	-	12	15	nC	
Q_{gs}	Gate to Source Gate Charge		I _D = 80A	-	30	-	nC
Q _{gs2}	Gate Charge Threshold to Plateau	$I_g = 1.0 \text{mA}$	$I_g = 1.0 \text{mA}$	-	18	-	nC
Q_{gd}	Gate to Drain "Miller" Charge			-	24	-	nC

Switching Characteristics $(V_{GS} = 10V)$

t _{ON}	Turn-On Time		-	-	163	ns
t _{d(ON)}	Turn-On Delay Time		-	15	-	ns
t _r	Rise Time	V _{DD} = 30V, I _D = 80A	-	93	-	ns
t _{d(OFF)}	Turn-Off Delay Time	$V_{GS} = 10V, R_{GS} = 2.4\Omega$	-	38	-	ns
t _f	Fall Time		-	13	-	ns
t _{OFF}	Turn-Off Time		-	-	75	ns

Drain-Source Diode Characteristics

V _{SD}	ISource to Drain Diode Voltage	I _{SD} = 80A	-	-	1.25	V
		I _{SD} = 40A	-	-	1.0	V
t _{rr}	Reverse Recovery Time	$I_{SD} = 75A$, $dI_{SD}/dt = 100A/\mu s$	-	-	38	ns
Q _{RR}	Reverse Recovered Charge	$I_{SD} = 75A$, $dI_{SD}/dt = 100A/\mu s$	-	-	39	nC

Notes:1: Starting $T_J = 25^{\circ}C$, L = 0.255mH, $I_{AS} = 70$ A.
2: Pulse Width = 100s

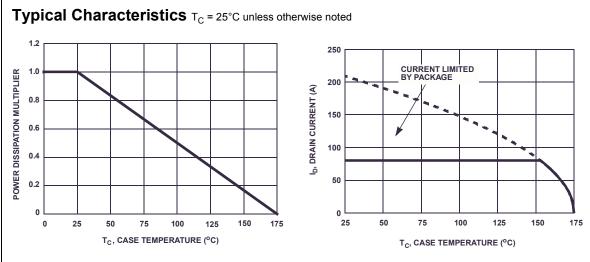


Figure 1. Normalized Power Dissipation vs Ambient Temperature

Figure 2. Maximum Continuous Drain Current vs Case Temperature

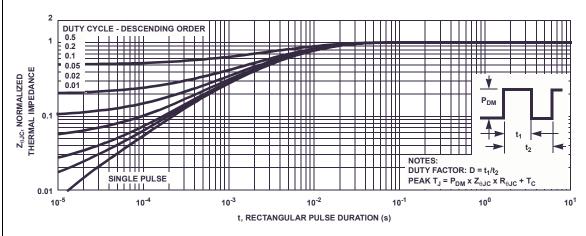


Figure 3. Normalized Maximum Transient Thermal Impedance

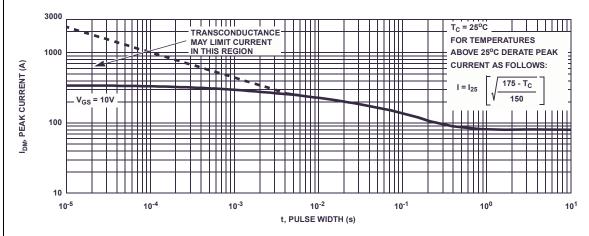


Figure 4. Peak Current Capability

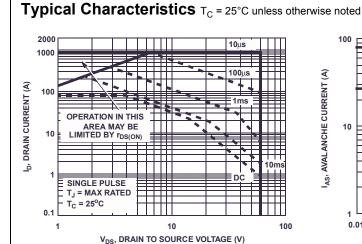
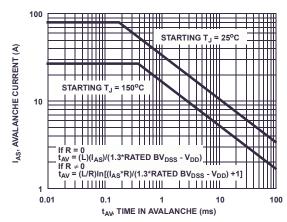


Figure 5. Forward Bias Safe Operating Area



NOTE: Refer to ON Semiconductor Application Notes AN7514 and AN7515

Figure 6. Unclamped Inductive Switching Capability

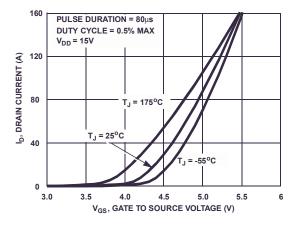


Figure 7. Transfer Characteristics

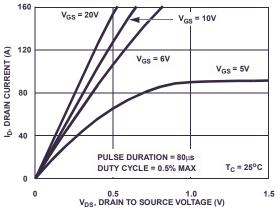


Figure 8. Saturation Characteristics

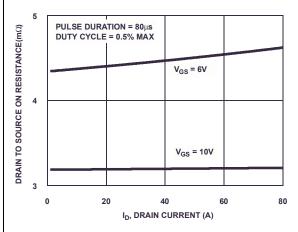


Figure 9. Drain to Source On Resistance vs Drain Current

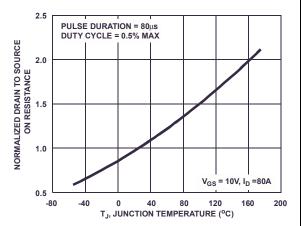


Figure 10. Normalized Drain to Source On Resistance vs Junction Temperature

Typical Characteristics T_C = 25°C unless otherwise noted

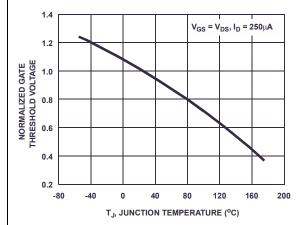


Figure 11. Normalized Gate Threshold Voltage vs Junction Temperature

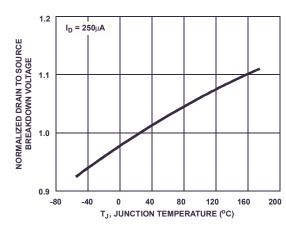


Figure 12. Normalized Drain to Source Breakdown Voltage vs Junction Temperature

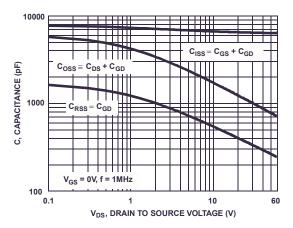


Figure 13. Capacitance vs Drain to Source Voltage

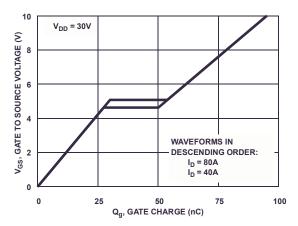


Figure 14. Gate Charge Waveforms for Constant Gate Current

Test Circuits and Waveforms

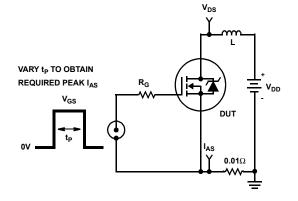


Figure 15. Unclamped Energy Test Circuit

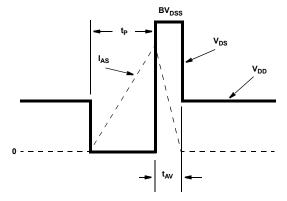


Figure 16. Unclamped Energy Waveforms

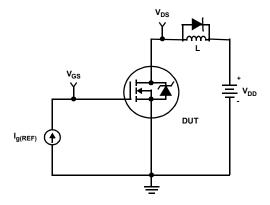


Figure 17. Gate Charge Test Circuit

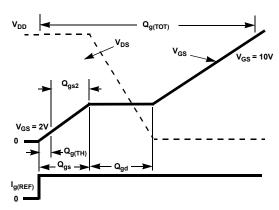


Figure 18. Gate Charge Waveforms

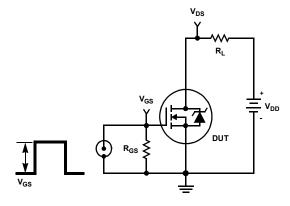


Figure 19. Switching Time Test Circuit

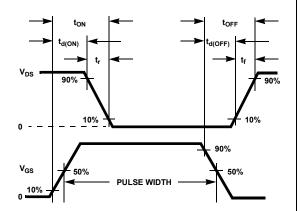


Figure 20. Switching Time Waveforms

Thermal Resistance vs. Mounting Pad Area

The maximum rated junction temperature, T_{JM} , and the thermal resistance of the heat dissipating path determines the maximum allowable device power dissipation, P_{DM} , in an application. Therefore the application's ambient temperature, T_A (°C), and thermal resistance $R_{\theta JA}$ (°C/W) must be reviewed to ensure that T_{JM} is never exceeded. Equation 1 mathematically represents the relationship and serves as the basis for establishing the rating of the part.

$$P_{DM} = \frac{(T_{JM} - T_A)}{R_{\theta JA}} \tag{EQ. 1}$$

In using surface mount devices such as the TO-263 package, the environment in which it is applied will have a significant influence on the part's current and maximum power dissipation ratings. Precise determination of P_{DM} is complex and influenced by many factors:

- Mounting pad area onto which the device is attached and whether there is copper on one side or both sides of the board
- The number of copper layers and the thickness of the board.
- 3. The use of external heat sinks.
- 4. The use of thermal vias.
- 5. Air flow and board orientation.
- 6. For non steady state applications, the pulse width, the duty cycle and the transient thermal response of the part, the board and the environment they are in.

ON Semiconductor provides thermal information to assist the designer's preliminary application evaluation. Figure 21

defines the $R_{\theta JA}$ for the device as a function of the top copper (component side) area. This is for a horizontally positioned FR-4 board with 1oz copper after 1000 seconds of steady state power with no air flow. This graph provides the necessary information for calculation of the steady state junction temperature or power dissipation. Pulse applications can be evaluated using the ON Semiconductor device Spice thermal model or manually utilizing the normalized maximum transient thermal impedance curve.

Thermal resistances corresponding to other copper areas can be obtained from Figure 21 or by calculation using Equation 2 or 3. Equation 2 is used for copper area defined in inches square and equation 3 is for area in centimeters square. The area, in square inches or square centimeters is the top copper area including the gate and source pads.

$$R_{\theta JA} = 26.51 + \frac{19.84}{(0.262 + Area)}$$
 (EQ. 2)

Area in Inches Squared

$$R_{\theta JA} = 26.51 + \frac{128}{(1.69 + Area)}$$
 (EQ. 3)

Area in Centimeters Squared

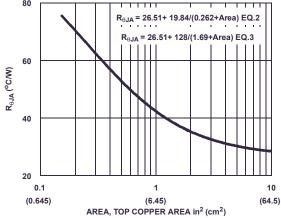


Figure 21. Thermal Resistance vs Mounting
Pad Area

PSPICE Electrical Model .SUBCKT FDB035AN06A0 2 1 3; rev July 04, 2002 Ca 12 8 1.5e-9 Cb 15 14 1.5e-9 I DRAIN DPLCAP DRAIN Cin 6 8 6.1e-9 10 Dbody 7 5 DbodyMOD RLDRAIN €RSLC1 Dbreak 5 11 DbreakMOD DBREAK 3 Dplcap 10 5 DplcapMOD RSLC2 € 5 51 FSI C Ebreak 11 7 17 18 69.3 Eds 14 8 5 8 1 50 Egs 13 8 6 8 1 RDRAIN ▲ DBODY 6 8 EBREAK FSG Esa 6 10 6 8 1 **EVTHRES** Evthres 6 21 19 8 1 21 Evtemp 20 6 18 22 1 **MWEAK** I GATE **EVTEMP** RGATE GATE (18 22 **★**MMED It 8 17 1 20 9 MSTR RLGATE Lgate 1 9 4.81e-9 LSOURCE CIN SOURCE Ldrain 2 5 1.0e-9 Lsource 3 7 4.63e-9 RSOURCE RLSOURCE RLgate 1 9 48.1 S1A **RBREAK** RLdrain 2 5 10 17 RLsource 3 7 46.3 **≨**RVTEMP oS2B Mmed 16 6 8 8 MmedMOD CB 19 CA Mstro 16 6 8 8 MstroMOD ΙT **(**†) Mweak 16 21 8 8 MweakMOD VBAT 6 5 EGS **EDS** Rbreak 17 18 RbreakMOD 1 Rdrain 50 16 Rdrain MOD 1e-4 **RVTHRES** Rgate 9 20 1.36 RSLC1 5 51 RSLCMOD 1e-6 RSLC2 5 50 1e3 Rsource 8 7 RsourceMOD 2.5e-3 Rythres 22 8 RythresMOD 1 Rvtemp 18 19 RvtempMOD 1 S1a 6 12 13 8 S1AMOD S1b 13 12 13 8 S1BMOD S2a 6 15 14 13 S2AMOD S2b 13 15 14 13 S2BMOD Vbat 22 19 DC 1 ESLC 51 50 VALUE={(V(5,51)/ABS(V(5,51)))*(PWR(V(5,51)/(1e-6*250),10))} .MODEL DbodyMOD D (IS=2.4E-11 N=1.04 RS=1.65e-3 TRS1=2.7e-3 TRS2=2e-7 + CJO=4.35e-9 M=5.4e-1 TT=1e-9 XTI=3.9) .MODEL DbreakMOD D (RS=1.5e-1 TRS1=1e-3 TRS2=-8.9e-6) .MODEL DplcapMOD D (CJO=1.7e-9 IS=1e-30 N=10 M=0.47) .MODEL MmedMOD NMOS (VTO=3.3 KP=9 IS=1e-30 N=10 TOX=1 L=1u W=1u RG=1.36 T_abs=25) .MODEL MstroMOD NMOS (VTO=4.00 KP=275 IS=1e-30 N=10 TOX=1 L=1u W=1u T_abs=25) .MODEL MweakMOD NMOS (VTO=2.72 KP=0.03 IS=1e-30 N=10 TOX=1 L=1u W=1u RG=13.6 RS=0.1 T_abs=25) .MODEL RbreakMOD RES (TC1=9e-4 TC2=-9e-7) .MODEL RdrainMOD RES (TC1=4e-2 TC2=1.75e-4) .MODEL RSLCMOD RES (TC1=1e-3 TC2=1e-5) .MODEL RsourceMOD RES (TC1=5e-3 TC2=1e-6) .MODEL RvthresMOD RES (TC1=-6.7e-3 TC2=-1.5e-5) .MODEL RvtempMOD RES (TC1=-2.5e-3 TC2=1e-6) .MODEL S1AMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-4 VOFF=-1.5) .MODEL S1BMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-1.5 VOFF=-4) MODEL S2AMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-1 VOFF=0.5) .MODEL S2BMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=0.5 VOFF=-1). **FNDS** Note: For further discussion of the PSPICE model, consult A New PSPICE Sub-Circuit for the Power MOSFET Featuring Global Temperature Options; IEEE Power Electronics Specialist Conference Records, 1991, written by William J. Hepp and C. Frank

Wheatley

SABER Electrical Model rev July 4, 2002 template FDB035AN06A0 n2,n1,n3 = m_temp electrical n2,n1,n3 number m_temp=25 var i iscl dp..model dbodymod = (isl=2.4e-11,nl=1.04,rs=1.65e-3,trs1=2.7e-3,trs2=2e-7,cjo=4.35e-9,m=5.4e-1,tt=1e-9,xti=3.9) dp..model dbreakmod = (rs=1.5e-1,trs1=1e-3,trs2=-8.9e-6) dp..model dplcapmod = (cjo=1.7e-9,isl=10e-30,nl=10,m=0.47) m..model mmedmod = (type=_n,vto=3.3,kp=9,is=1e-30, tox=1) m..model mstrongmod = $(type=_n, vto=4.00, kp=275, is=1e-30, tox=1)$ LDRAIN m..model mweakmod = (type=_n,vto=2.72,kp=0.03,is=1e-30, tox=1,rs=0.1) DRAIN sw_vcsp..model s1amod = (ron=1e-5,roff=0.1,von=-4,voff=-1.5) sw_vcsp..model s1bmod = (ron=1e-5,roff=0.1,von=-1.5,voff=-4) RLDRAIN sw_vcsp..model s2amod = (ron=1e-5,roff=0.1,von=-1,voff=0.5) RSLC1 sw_vcsp..model s2bmod = (ron=1e-5,roff=0.1,von=0.5,voff=-1) RSLC2 € c.ca n12 n8 = 1.5e-9ISCL c.cb n15 n14 = 1.5e-9c.cin n6 n8 = 6.1e-9DBREAK 3 RDRAIN dp.dbodv n7 n5 = model=dbodvmod ESG DBODY dp.dbreak n5 n11 = model=dbreakmod **EVTHRES** dp.dplcap n10 n5 = model=dplcapmod (<u>19</u>) MWEAK **LGATE EVTEMP** RGATE spe.ebreak n11 n7 n17 n18 = 69.3 **EBREAK** MMFD 20 $\frac{1}{100}$ spe.eds n14 n8 n5 n8 = 1 **RLGATE** spe.egs n13 n8 n6 n8 = 1 LSOURCE spe.esg n6 n10 n6 n8 = 1 CIN SOURCE spe.evthres n6 n21 n19 n8 = 1 spe.evtemp n20 n6 n18 n22 = 1 **RSOURCE** RLSOURCE i.it n8 n17 = 1RBREAK 17 18 I.lgate n1 n9 = 4.81e-9**₹**RVTEMP I.ldrain n2 n5 = 1.0e-9СВ 19 I.lsource n3 n7 = 4.63e-9CA IT (14 **VBAT** res.rlgate n1 n9 = 48.1 EGS res.rldrain n2 n5 = 10 res.rlsource n3 n7 = 46.3 **RVTHRES** m.mmed n16 n6 n8 n8 = model=mmedmod, temp=m_temp, l=1u, w=1u m.mstrong n16 n6 n8 n8 = model=mstrongmod, temp=m_temp, l=1u, w=1u m.mweak n16 n21 n8 n8 = model=mweakmod, temp=m_temp, l=1u, w=1u res.rbreak n17 n18 = 1, tc1=9e-4,tc2=-9e-7 res.rdrain n50 n16 = 1e-4, tc1=4e-2,tc2=1.75e-4 res.rgate n9 n20 = 1.36 res.rslc1 n5 n51 = 1e-6, tc1=1e-3,tc2=1e-5 res.rslc2 n5 n50 = 1e3res.rsource n8 n7 = 2.5e-3, tc1=5e-3,tc2=1e-6 res.rvthres n22 n8 = 1, tc1=-6.7e-3,tc2=-1.5e-5 res.rvtemp n18 n19 = 1, tc1=-2.5e-3,tc2=1e-6 sw_vcsp.s1a n6 n12 n13 n8 = model=s1amod sw_vcsp.s1b n13 n12 n13 n8 = model=s1bmod sw_vcsp.s2a n6 n15 n14 n13 = model=s2amod sw_vcsp.s2b n13 n15 n14 n13 = model=s2bmod v.vbat n22 n19 = dc=1 equations { i (n51->n50) +=iscl iscl: v(n51,n50) = ((v(n5,n51)/(1e-9+abs(v(n5,n51))))*((abs(v(n5,n51)*1e6/250))**10))

SPICE Thermal Model

REV 23 July 4, 2002

FDB035AN06A0T

CTHERM1 TH 6 6.45e-3 CTHERM2 6 5 3e-2 CTHERM3 5 4 1.4e-2 CTHERM4 4 3 1.65e-2 CTHERM5 3 2 4.85e-2 CTHERM6 2 TL 1e-1

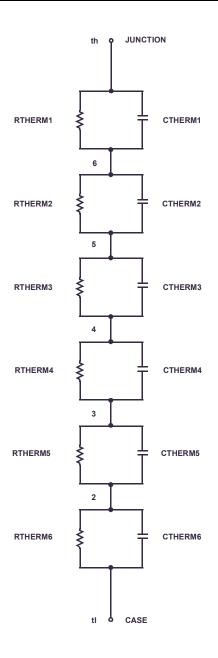
RTHERM1 TH 6 3.24e-3 RTHERM2 6 5 8.08e-3 RTHERM3 5 4 2.28e-2 RTHERM4 4 3 1e-1 RTHERM5 3 2 1.1e-1 RTHERM6 2 TL 1.4e-1

SABER Thermal Model

SABER thermal model FDB035AN06A0T template thermal_model th tI thermal_c th, tI {
ctherm.ctherm1 th 6 = 6.45e-3 ctherm.ctherm2 6 5 = 3e-2 ctherm.ctherm3 5 4 - 1 4e-2

ctherm.ctherm2 6 5 = 3e-2 ctherm.ctherm3 5 4 = 1.4e-2 ctherm.ctherm4 4 3 = 1.65e-2 ctherm.ctherm5 2 = 4.85e-2 ctherm.ctherm6 2 tl = 1e-1

rtherm.rtherm1 th 6 = 3.24e-3 rtherm.rtherm2 6 5 = 8.08e-3 rtherm.rtherm3 5 4 = 2.28e-2 rtherm.rtherm4 3 = 1e-1 rtherm.rtherm5 3 2 = 1.1e-1 rtherm.rtherm6 2 tl=1.4e-1



Mechanical Dimensions

TO-263 2L (D²PAK)

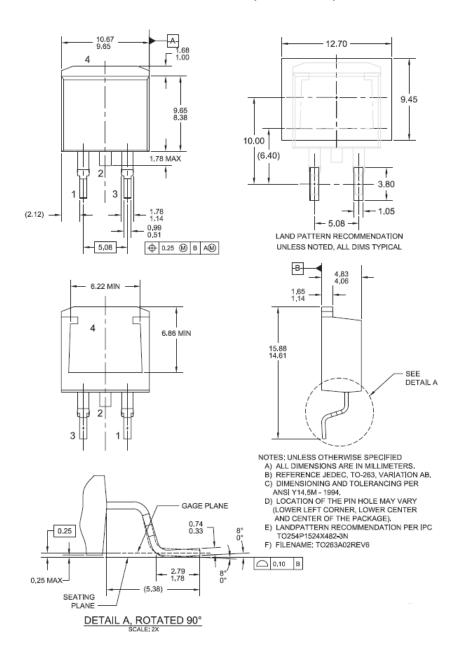


Figure 22. 2LD, TO263, Surface Mount

Package drawings are provided as a service to customers considering ON Semiconductor components. Drawings may change in any manner without notice. Please note the revision and/or date on the drawing and contact a ON Semiconductor representative to verify or obtain the most recent revision. Package specifications do not expand the terms of ON Semiconductor's worldwide terms and conditions, specif-ically the warranty therein, which covers ON Semiconductor products.

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