

BF1218

Dual N-channel dual gate MOSFET

Rev. 01 — 14 April 2010

Product data sheet

1. Product profile

1.1 General description

The BF1218 is a combination of two dual gate MOSFET amplifiers with shared source and gate2 leads and an integrated switch. The integrated switch is operated by the gate1 bias of amplifier B.

The source and substrate are interconnected. Internal bias circuits enable DC stabilization and a very good cross modulation performance during Automatic Gain Control (AGC). Integrated diodes between the gates and source protect against excessive input voltage surges. The transistor has a SOT363 micro-miniature plastic package.

CAUTION



This device is sensitive to ElectroStatic Discharge (ESD). Therefore care should be taken during transport and handling.

1.2 Features and benefits

- Two low noise gain controlled amplifiers in a single package. One with a fully integrated bias and one with a partly integrated bias
- Internal switch to save external components
- Superior cross modulation performance during AGC
- High forward transfer admittance
- High forward transfer admittance to input capacitance ratio

1.3 Applications

- Gain controlled low noise amplifiers for VHF and UHF applications with 5 V supply voltage
 - ◆ digital and analog television tuners
 - ◆ professional communication equipment



1.4 Quick reference data

Table 1. Quick reference data
Per MOSFET unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{DS}	drain-source voltage	DC	-	-	6	V
I_D	drain current	DC	-	-	30	mA
P_{tot}	total power dissipation	$T_{sp} \leq 109\text{ }^\circ\text{C}$	[1]	-	180	mW
$ y_{fs} $	forward transfer admittance	$f = 100\text{ MHz}; T_j = 25\text{ }^\circ\text{C}$				
		amplifier A; $I_D = 19\text{ mA}$	26	31	41	mS
		amplifier B; $I_D = 15\text{ mA}$	25	30	40	mS
$C_{iss(G1)}$	input capacitance at gate1	$f = 100\text{ MHz}$				
		amplifier A	[2]	-	2.1	2.6 pF
		amplifier B	[2]	-	2.1	2.6 pF
C_{rss}	reverse transfer capacitance	$f = 100\text{ MHz}$	[2]	-	20	fF
NF	noise figure	$Y_S = Y_{S(opt)}$				
		amplifier A; $f = 400\text{ MHz}$	-	0.9	1.5	dB
		amplifier B; $f = 800\text{ MHz}$	-	1.4	2.0	dB
Xmod	cross modulation	input level for $k = 1\%$; $f_w = 50\text{ MHz}$; $f_{unw} = 60\text{ MHz}$ at 40 dB AGC				
		amplifier A	[3]	102	105	- dB μ V
		amplifier B	[4]	102	105	- dB μ V
T_j	junction temperature		-	-	150	$^\circ\text{C}$

[1] T_{sp} is the temperature at the soldering point of the source lead.

[2] Calculated from S-parameters.

[3] Measured in [Figure 33](#) test circuit.

[4] Measured in [Figure 34](#) test circuit.

2. Pinning information

Table 2. Discrete pinning

Pin	Description	Simplified outline	Graphic symbol
1	gate1 (AMP A)		
2	gate2		
3	gate1 (AMP B)		
4	drain (AMP B)		
5	source		
6	drain (AMP A)		

3. Ordering information

Table 3. Ordering information

Type number	Package		
	Name	Description	Version
BF1218	-	plastic surface-mounted package; 6 leads	SOT363

4. Marking

Table 4. Marking codes

Type number	Marking code
BF1218	M7

5. Limiting values

Table 5. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions	Min	Max	Unit
Per MOSFET					
V_{DS}	drain-source voltage	DC	-	6	V
I_D	drain current	DC	-	30	mA
I_{G1}	gate1 current		-	±10	mA
I_{G2}	gate2 current		-	±10	mA
P_{tot}	total power dissipation	$T_{sp} \leq 109\text{ °C}$ [1]	-	180	mW
T_{stg}	storage temperature		-65	+150	°C
T_j	junction temperature		-	150	°C

[1] T_{sp} is the temperature at the soldering point of the source lead.

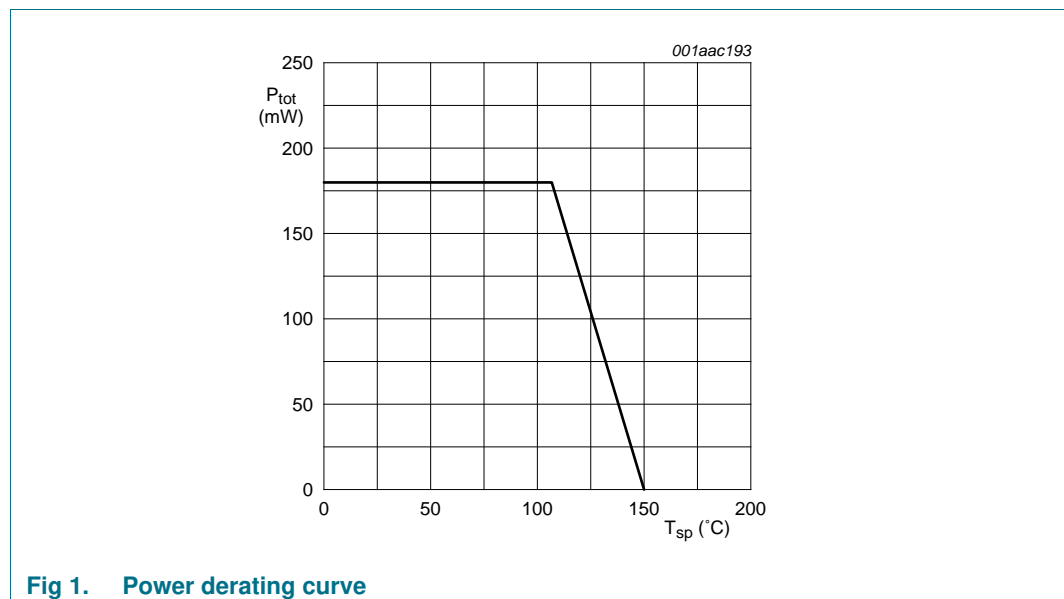


Fig 1. Power derating curve

6. Thermal characteristics

Table 6. Thermal characteristics

Symbol	Parameter	Conditions	Typ	Unit
$R_{th(j-sp)}$	thermal resistance from junction to solder point		225	K/W

7. Static characteristics

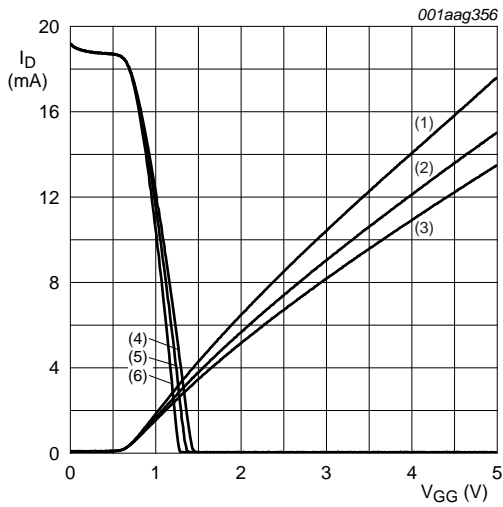
Table 7. Static characteristics

$T_j = 25\text{ }^\circ\text{C}$; unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit	
Per MOSFET; unless otherwise specified							
$V_{(BR)DSS}$	drain-source breakdown voltage	$V_{G1-S} = V_{G2-S} = 0\text{ V}$; $I_D = 10\text{ }\mu\text{A}$					
		amplifier A	6	-	-	V	
		amplifier B	6	-	-	V	
$V_{(BR)G1-SS}$	gate1-source breakdown voltage	$V_{G2-S} = V_{DS} = 0\text{ V}$; $I_{G1-S} = 10\text{ mA}$	6	-	10	V	
$V_{(BR)G2-SS}$	gate2-source breakdown voltage	$V_{G1-S} = V_{DS} = 0\text{ V}$; $I_{G2-S} = 10\text{ mA}$	6	-	10	V	
$V_{F(S-G1)}$	forward source-gate1 voltage	$V_{G2-S} = V_{DS} = 0\text{ V}$; $I_{S-G1} = 10\text{ mA}$	0.5	-	1.5	V	
$V_{F(S-G2)}$	forward source-gate2 voltage	$V_{G1-S} = V_{DS} = 0\text{ V}$; $I_{S-G2} = 10\text{ mA}$	0.5	-	1.5	V	
$V_{G1-S(th)}$	gate1-source threshold voltage	$V_{DS} = 5\text{ V}$; $V_{G2-S} = 4\text{ V}$; $I_D = 100\text{ }\mu\text{A}$	0.3	-	1.0	V	
$V_{G2-S(th)}$	gate2-source threshold voltage	$V_{DS} = 5\text{ V}$; $V_{G1-S} = 5\text{ V}$; $I_D = 100\text{ }\mu\text{A}$	0.4	-	1.0	V	
I_{DS}	drain-source current	$V_{G2-S} = 4\text{ V}$; $V_{DS(B)} = 5\text{ V}$; $R_{G1} = 86\text{ k}\Omega$					
		amplifier A; $V_{DS(A)} = 5\text{ V}$	[1]	14	-	24	mA
		amplifier B	[2]	10	-	20	mA
I_{G1-S}	gate1 cut-off current	$V_{G2-S} = V_{DS(A)} = 0\text{ V}$					
		amplifier A; $V_{G1-S(A)} = 5\text{ V}$; $I_{D(B)} = 0\text{ A}$	-	-	50	nA	
		amplifier B; $V_{G1-S(B)} = 5\text{ V}$; $V_{DS(B)} = 0\text{ V}$	-	-	50	nA	
I_{G2-S}	gate2 cut-off current	$V_{G2-S} = 4\text{ V}$; $V_{G1-S(B)} = 0\text{ V}$; $V_{G1-S(A)} = V_{DS(A)} = V_{DS(B)} = 0\text{ V}$	-	-	20	nA	

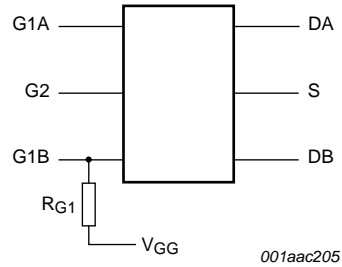
[1] R_{G1} connects gate1 (B) to $V_{GG} = 0\text{ V}$ (see [Figure 3](#)).

[2] R_{G1} connects gate1 (B) to $V_{GG} = 5\text{ V}$ (see [Figure 3](#)).



- (1) $I_{D(B)}$; $R_{G1} = 68 \text{ k}\Omega$.
- (2) $I_{D(B)}$; $R_{G1} = 86 \text{ k}\Omega$.
- (3) $I_{D(B)}$; $R_{G1} = 100 \text{ k}\Omega$.
- (4) $I_{D(A)}$; $R_{G1} = 100 \text{ k}\Omega$.
- (5) $I_{D(A)}$; $R_{G1} = 86 \text{ k}\Omega$.
- (6) $I_{D(A)}$; $R_{G1} = 68 \text{ k}\Omega$.

Fig 2. Drain currents of MOSFET A and B as a function of V_{GG}



$V_{GG} = 5 \text{ V}$: amplifier A is off; amplifier B is on.
 $V_{GG} = 0 \text{ V}$: amplifier A is on; amplifier B is off.

Fig 3. Functional diagram

8. Dynamic characteristics

8.1 Dynamic characteristics for amplifier A

Table 8. Dynamic characteristics for amplifier A^[1]

Common source; $T_{amb} = 25 \text{ }^\circ\text{C}$; $V_{G2-S} = 4 \text{ V}$; $V_{DS} = 5 \text{ V}$; $I_D = 19 \text{ mA}$; unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$ y_{fs} $	forward transfer admittance	$f = 100 \text{ MHz}$; $T_j = 25 \text{ }^\circ\text{C}$	26	31	41	mS
$C_{iss(G1)}$	input capacitance at gate1	$f = 100 \text{ MHz}$	[2] -	2.1	2.6	pF
$C_{iss(G2)}$	input capacitance at gate2	$f = 100 \text{ MHz}$	[2] -	3.4	-	pF
C_{oss}	output capacitance	$f = 100 \text{ MHz}$	[2] -	0.8	-	pF
C_{rss}	reverse transfer capacitance	$f = 100 \text{ MHz}$	[2] -	20	-	fF
G_{tr}	transducer power gain	$B_S = B_{S(opt)}$; $B_L = B_{L(opt)}$				
		$f = 200 \text{ MHz}$; $G_S = 2 \text{ mS}$; $G_L = 0.5 \text{ mS}$	32	36	40	dB
		$f = 400 \text{ MHz}$; $G_S = 2 \text{ mS}$; $G_L = 1 \text{ mS}$	28	32	36	dB
		$f = 800 \text{ MHz}$; $G_S = 3.3 \text{ mS}$; $G_L = 1 \text{ mS}$	24	28	33	dB
NF	noise figure	$f = 11 \text{ MHz}$; $G_S = 20 \text{ mS}$; $B_S = 0 \text{ S}$	-	3.0	-	dB
		$f = 400 \text{ MHz}$; $Y_S = Y_{S(opt)}$	-	0.9	1.5	dB
		$f = 800 \text{ MHz}$; $Y_S = Y_{S(opt)}$	-	1.1	1.7	dB

Table 8. Dynamic characteristics for amplifier A[1] ...continued

Common source; $T_{amb} = 25\text{ }^{\circ}\text{C}$; $V_{G2-S} = 4\text{ V}$; $V_{DS} = 5\text{ V}$; $I_D = 19\text{ mA}$; unless otherwise specified.

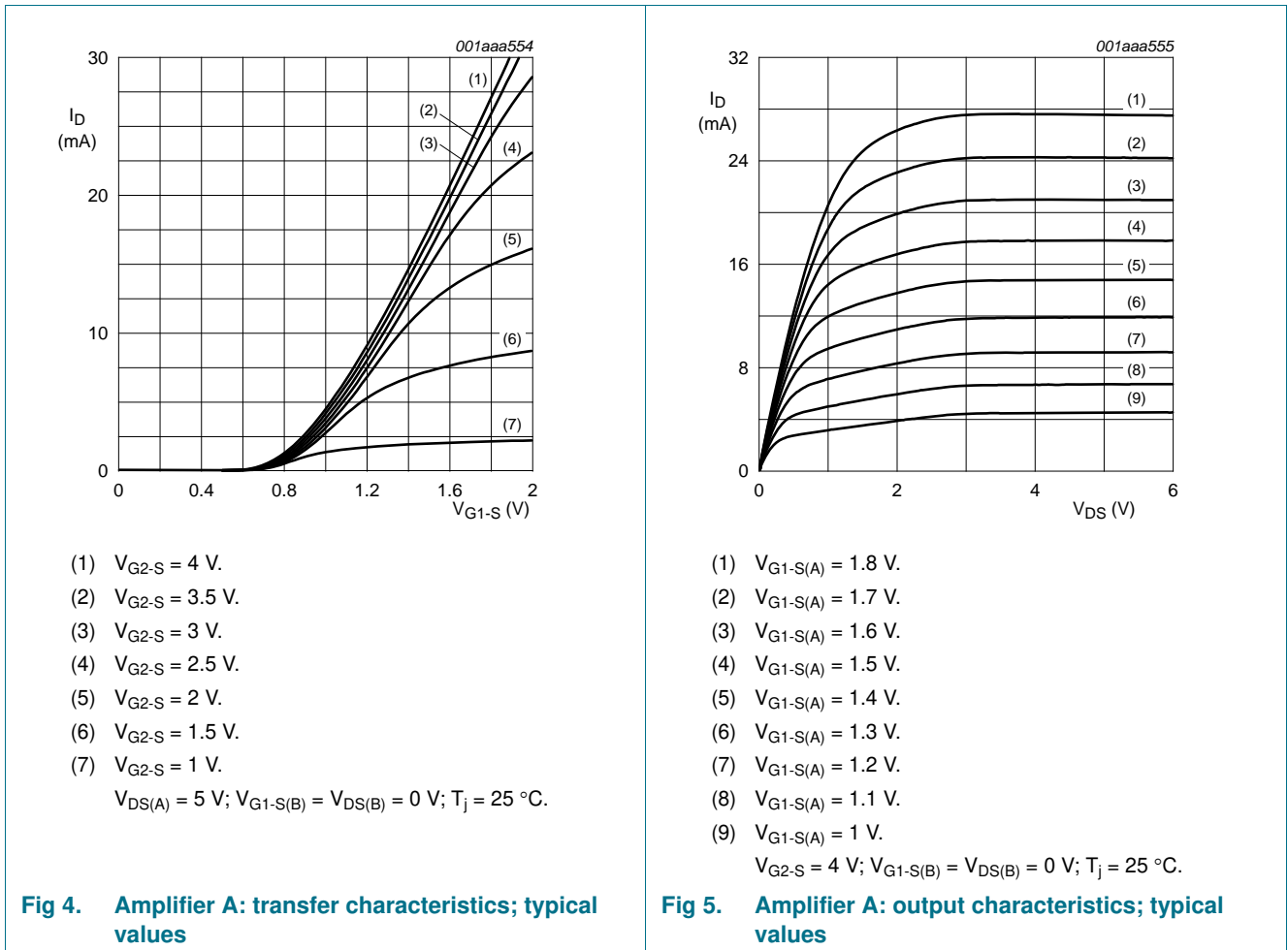
Symbol	Parameter	Conditions	Min	Typ	Max	Unit
Xmod	cross modulation	input level for $k = 1\%$; $f_w = 50\text{ MHz}$; $f_{unw} = 60\text{ MHz}$	[3]			
		at 0 dB AGC	90	-	-	dB μ V
		at 10 dB AGC	-	90	-	dB μ V
		at 20 dB AGC	-	99	-	dB μ V
		at 40 dB AGC	102	105	-	dB μ V

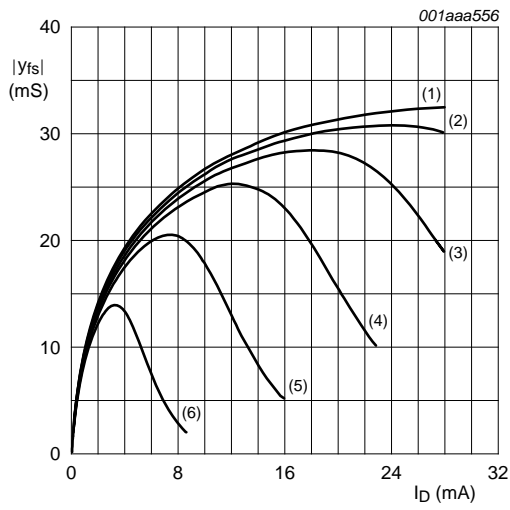
[1] For the MOSFET not in use: $V_{G1-S(B)} = 0\text{ V}$; $V_{DS(B)} = 0\text{ V}$.

[2] Calculated from S-parameters.

[3] Measured in [Figure 33](#) test circuit.

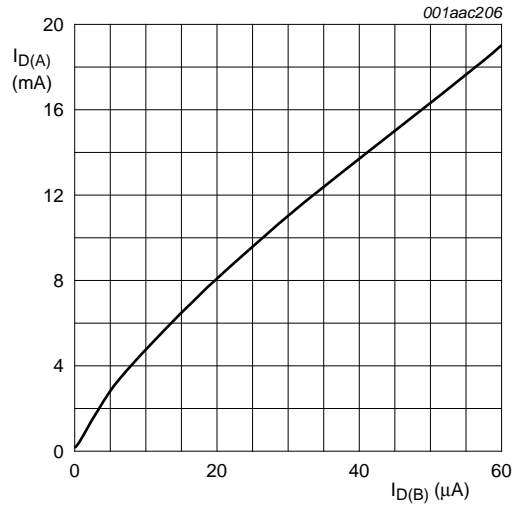
8.1.1 Graphics for amplifier A





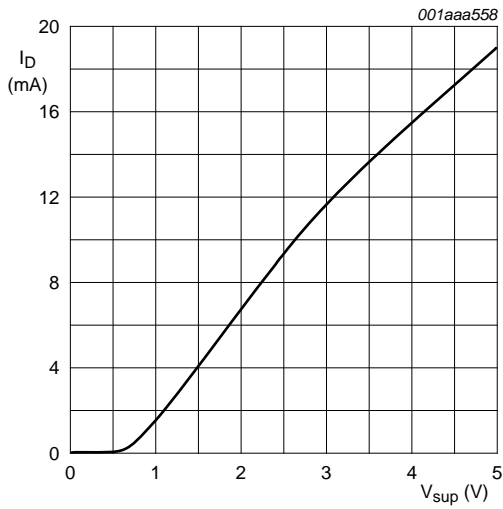
- (1) $V_{G2-S} = 4 \text{ V}$.
 - (2) $V_{G2-S} = 3.5 \text{ V}$.
 - (3) $V_{G2-S} = 3 \text{ V}$.
 - (4) $V_{G2-S} = 2.5 \text{ V}$.
 - (5) $V_{G2-S} = 2 \text{ V}$.
 - (6) $V_{G2-S} = 1.5 \text{ V}$.
- $V_{DS(A)} = 5 \text{ V}; V_{G1-S(B)} = V_{DS(B)} = 0 \text{ V}; T_j = 25 \text{ }^\circ\text{C}$.

Fig 6. Amplifier A: forward transfer admittance as a function of drain current; typical values



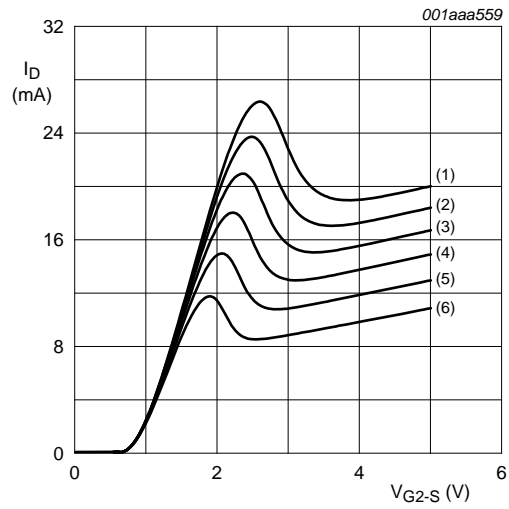
- $V_{DS(A)} = 5 \text{ V}; V_{G2-S} = 4 \text{ V}; V_{DS(B)} = 5 \text{ V}; V_{G1-S(B)} = 0 \text{ V}; T_j = 25 \text{ }^\circ\text{C}$.
- $I_{D(B)}$ = internal gate1 current = current in pin drain (AMP B) if MOSFET (B) is switched off.

Fig 7. Amplifier A: drain current as a function of internal gate1 current; typical values



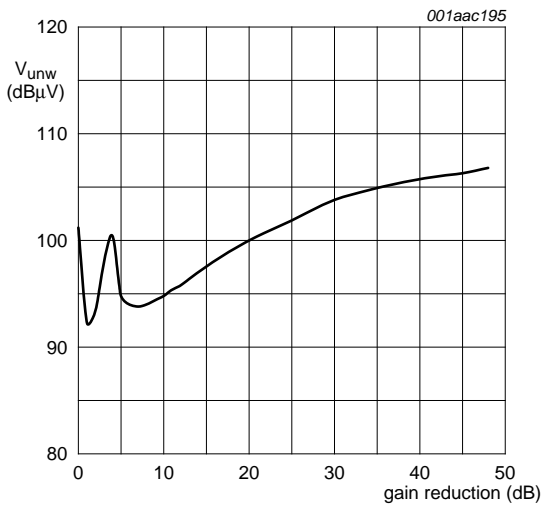
$V_{DS(A)} = V_{DS(B)} = V_{sup}$; $V_{G2-S} = 4\text{ V}$; $T_j = 25\text{ }^\circ\text{C}$;
 $R_{G1} = 86\text{ k}\Omega$ (connected to ground); see [Figure 3](#).

Fig 8. Amplifier A: drain current of amplifier A as a function of supply voltage of A and B amplifier; typical values



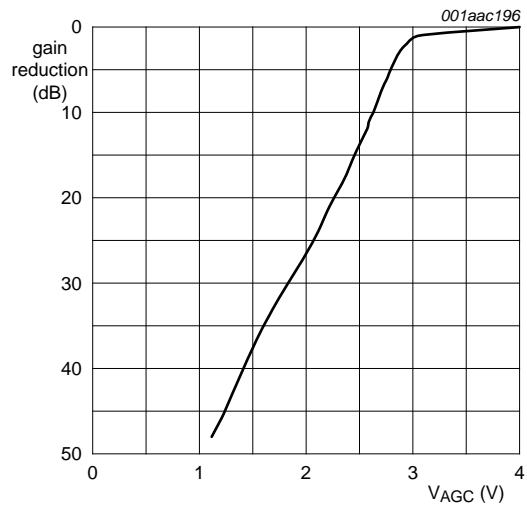
(1) $V_{DS(B)} = 5\text{ V}$.
 (2) $V_{DS(B)} = 4.5\text{ V}$.
 (3) $V_{DS(B)} = 4\text{ V}$.
 (4) $V_{DS(B)} = 3.5\text{ V}$.
 (5) $V_{DS(B)} = 3\text{ V}$.
 (6) $V_{DS(B)} = 2.5\text{ V}$.
 $V_{DS(A)} = 5\text{ V}$; $V_{G1-S(B)} = 0\text{ V}$; gate1 (AMP A) is open;
 $T_j = 25\text{ }^\circ\text{C}$.

Fig 9. Amplifier A: drain current as a function of gate2 voltage; typical values



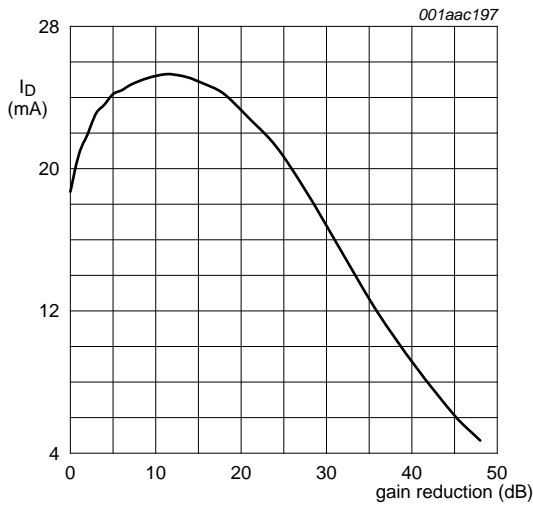
$V_{DS(A)} = V_{DS(B)} = 5\text{ V}$; $V_{G1-S(B)} = 0\text{ V}$; $f_w = 50\text{ MHz}$;
 $f_{unw} = 60\text{ MHz}$; $T_{amb} = 25\text{ }^\circ\text{C}$; see [Figure 33](#).

Fig 10. Amplifier A: unwanted voltage for 1 % cross modulation as a function of gain reduction; typical values



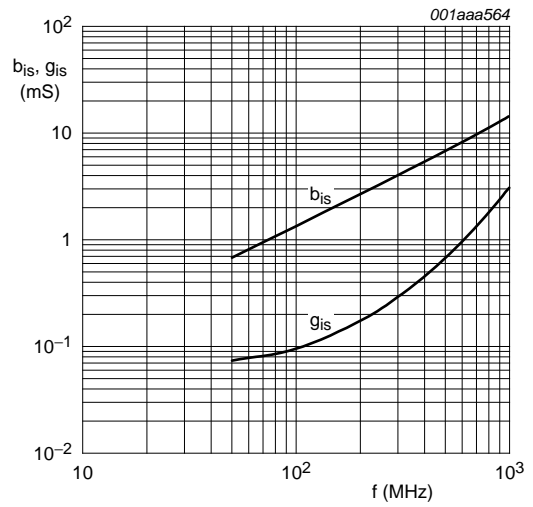
$V_{DS(A)} = V_{DS(B)} = 5\text{ V}$; $V_{G1-S(B)} = 0\text{ V}$; $f = 50\text{ MHz}$; see [Figure 33](#).

Fig 11. Amplifier A: gain reduction as a function of AGC voltage; typical values



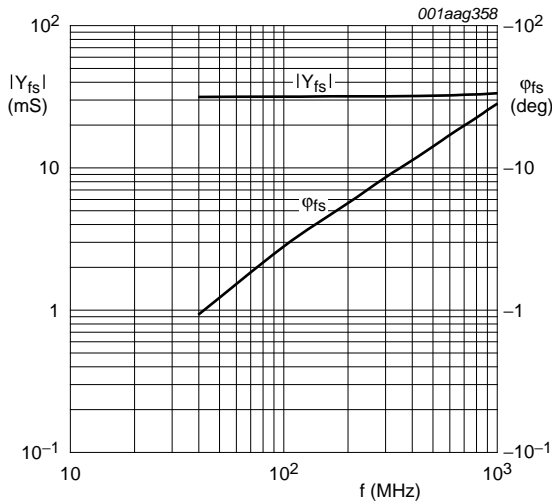
$V_{DS(A)} = V_{DS(B)} = 5\text{ V}$; $V_{G1-S(B)} = 0\text{ V}$; $f = 50\text{ MHz}$;
 $T_{amb} = 25\text{ }^{\circ}\text{C}$; see [Figure 33](#).

Fig 12. Amplifier A: drain current as a function of gain reduction; typical values



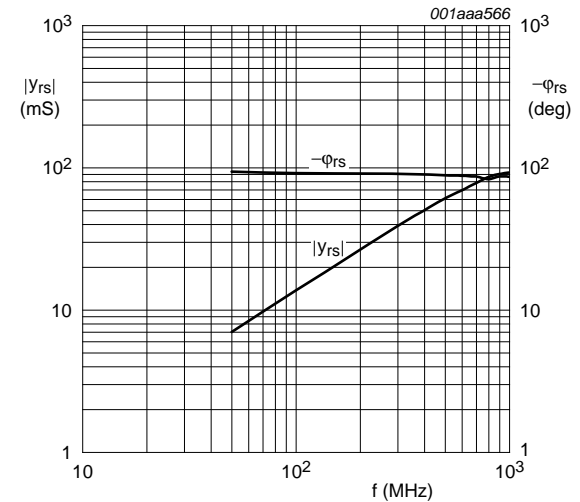
$V_{DS(A)} = 5\text{ V}$; $V_{G2-S} = 4\text{ V}$; $V_{DS(B)} = V_{G1-S(B)} = 0\text{ V}$;
 $I_{D(A)} = 19\text{ mA}$

Fig 13. Amplifier A: input admittance as a function of frequency; typical values



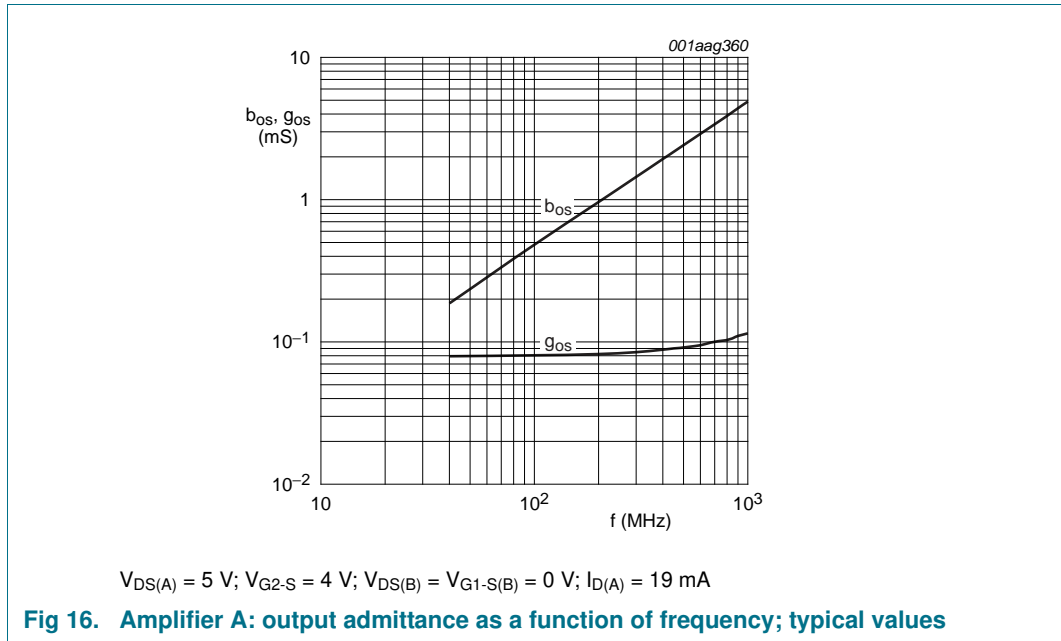
$V_{DS(A)} = 5\text{ V}$; $V_{G2-S} = 4\text{ V}$; $V_{DS(B)} = V_{G1-S(B)} = 0\text{ V}$;
 $I_{D(A)} = 19\text{ mA}$

Fig 14. Amplifier A: forward transfer admittance and phase as a function of frequency; typical values



$V_{DS(A)} = 5\text{ V}$; $V_{G2-S} = 4\text{ V}$; $V_{DS(B)} = V_{G1-S(B)} = 0\text{ V}$;
 $I_{D(A)} = 19\text{ mA}$

Fig 15. Amplifier A: reverse transfer admittance and phase as a function of frequency; typical values



8.1.2 Scattering parameters for amplifier A

Table 9. Scattering parameters for amplifier A

$V_{DS(A)} = 5\text{ V}; V_{G2-S} = 4\text{ V}; I_{D(A)} = 19\text{ mA}; V_{DS(B)} = 0\text{ V}; V_{G1-S(B)} = 0\text{ V}; T_{amb} = 25\text{ }^{\circ}\text{C};$ typical values.

F (MHz)	S ₁₁		S ₂₁		S ₁₂		S ₂₂	
	Magnitude (ratio)	Angle (degree)	Magnitude (ratio)	Angle (degree)	Magnitude (ratio)	Angle (degree)	Magnitude (ratio)	Angle (degree)
40	0.9927	-4.10	3.1833	175.69	0.0006	92.99	0.9927	-1.24
100	0.9897	-7.68	3.1743	171.77	0.0011	81.72	0.9923	-2.54
200	0.9852	-15.36	3.1494	163.56	0.0023	79.23	0.9912	-5.09
300	0.9758	-22.84	3.1146	155.46	0.0033	74.65	0.9904	-7.60
400	0.9655	-30.19	3.0718	147.53	0.0042	70.46	0.9890	-10.10
500	0.9513	-37.55	3.0156	139.61	0.0049	66.38	0.9874	-12.60
600	0.9341	-44.85	2.9482	131.74	0.0056	62.22	0.9853	-15.11
700	0.9160	-51.99	2.8755	124.04	0.0061	58.44	0.9832	-17.61
800	0.8964	-58.99	2.8003	116.41	0.0064	54.48	0.9806	-20.12
900	0.8737	-65.84	2.7206	108.93	0.0066	50.78	0.9793	-22.57
1000	0.8499	-72.51	2.6352	101.62	0.0067	46.49	0.9776	-25.05

8.1.3 Noise data for amplifier A

Table 10. Noise data for amplifier A

$V_{DS(A)} = 5\text{ V}; V_{G2-S} = 4\text{ V}; I_{D(A)} = 19\text{ mA}; V_{DS(B)} = 0\text{ V}; V_{G1-S(B)} = 0\text{ V}; T_{amb} = 25\text{ }^{\circ}\text{C};$ typical values; unless otherwise specified.

f (MHz)	NF _{min} (dB)	Γ _{opt}		r _n (ratio)
		(ratio)	(degree)	
400	0.9	0.77	22.7	0.65
800	1.1	0.73	45.75	0.62

8.2 Dynamic characteristics for amplifier B

Table 11. Dynamic characteristics for amplifier B^[1]

Common source; $T_{amb} = 25\text{ °C}$; $V_{G2-S} = 4\text{ V}$; $V_{DS} = 5\text{ V}$; $I_D = 15\text{ mA}$; unless otherwise specified.

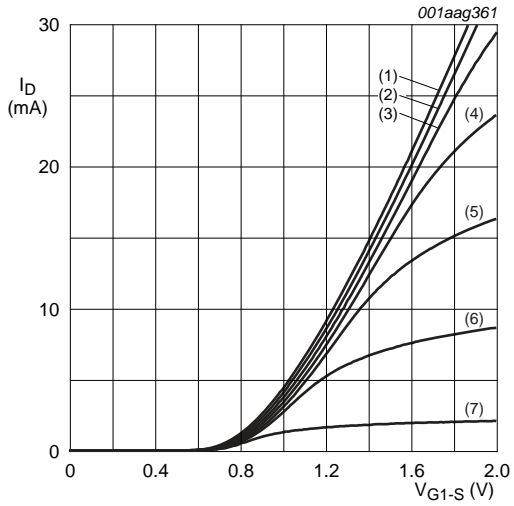
Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$ y_{fs} $	forward transfer admittance	$f = 100\text{ MHz}$; $T_j = 25\text{ °C}$	25	30	40	mS
$C_{iss(G1)}$	input capacitance at gate1	$f = 100\text{ MHz}$	[2]	-	2.1	2.6 pF
$C_{iss(G2)}$	input capacitance at gate2	$f = 100\text{ MHz}$	[2]	-	3.4	- pF
C_{oss}	output capacitance	$f = 100\text{ MHz}$	[2]	-	0.85	- pF
C_{rss}	reverse transfer capacitance	$f = 100\text{ MHz}$	[2]	-	20	- fF
G_{tr}	transducer power gain	$B_S = B_{S(opt)}$; $B_L = B_{L(opt)}$				
		$f = 200\text{ MHz}$; $G_S = 2\text{ mS}$; $G_L = 0.5\text{ mS}$	31	35	39	dB
		$f = 400\text{ MHz}$; $G_S = 2\text{ mS}$; $G_L = 1\text{ mS}$	28	32	36	dB
		$f = 800\text{ MHz}$; $G_S = 3.3\text{ mS}$; $G_L = 1\text{ mS}$	26	30	34	dB
NF	noise figure	$f = 11\text{ MHz}$; $G_S = 20\text{ mS}$; $B_S = 0\text{ S}$	-	3	-	dB
		$f = 400\text{ MHz}$; $Y_S = Y_{S(opt)}$	-	1.1	1.7	dB
		$f = 800\text{ MHz}$; $Y_S = Y_{S(opt)}$	-	1.4	2.0	dB
Xmod	cross modulation	input level for $k = 1\%$; $f_w = 50\text{ MHz}$; $f_{unw} = 60\text{ MHz}$	[3]			
		at 0 dB AGC	90	-	-	dB μ V
		at 10 dB AGC	-	90	-	dB μ V
		at 20 dB AGC	-	98	-	dB μ V
		at 40 dB AGC	102	105	-	dB μ V

[1] For the MOSFET not in use: $V_{G1-S(A)} = 0\text{ V}$; $V_{DS(A)} = 0\text{ V}$.

[2] Calculated from S-parameters.

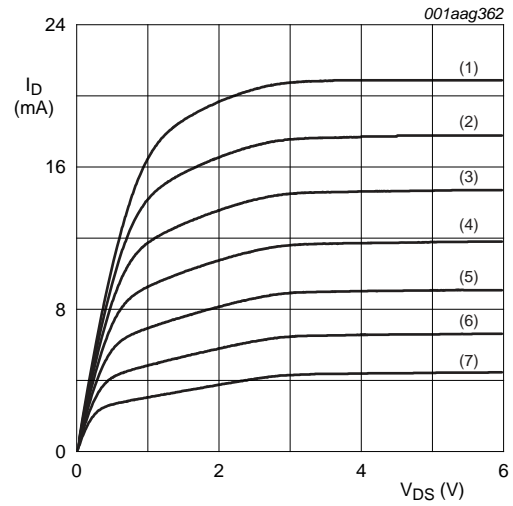
[3] Measured in [Figure 34](#) test circuit.

8.2.1 Graphics for amplifier B



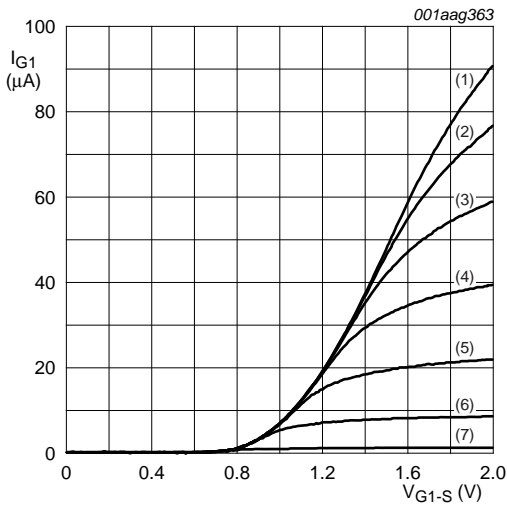
- (1) $V_{G2-S} = 4 \text{ V}$.
 - (2) $V_{G2-S} = 3.5 \text{ V}$.
 - (3) $V_{G2-S} = 3 \text{ V}$.
 - (4) $V_{G2-S} = 2.5 \text{ V}$.
 - (5) $V_{G2-S} = 2 \text{ V}$.
 - (6) $V_{G2-S} = 1.5 \text{ V}$.
 - (7) $V_{G2-S} = 1 \text{ V}$.
- $V_{DS(B)} = 5 \text{ V}$; $V_{DS(A)} = V_{G1-S(A)} = 0 \text{ V}$; $T_j = 25 \text{ }^\circ\text{C}$.

Fig 17. Amplifier B: transfer characteristics; typical values



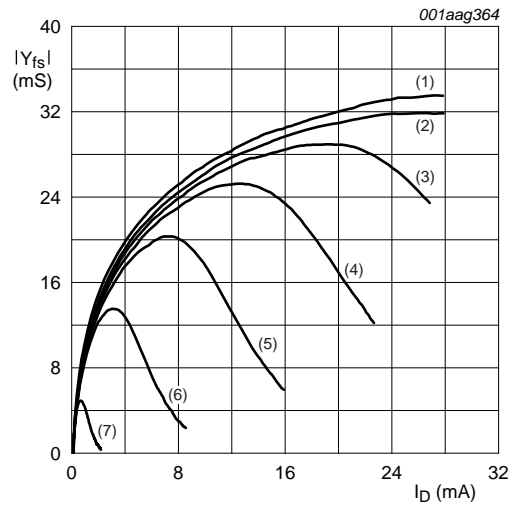
- (1) $V_{G1-S(B)} = 1.6 \text{ V}$.
 - (2) $V_{G1-S(B)} = 1.5 \text{ V}$.
 - (3) $V_{G1-S(B)} = 1.4 \text{ V}$.
 - (4) $V_{G1-S(B)} = 1.3 \text{ V}$.
 - (5) $V_{G1-S(B)} = 1.2 \text{ V}$.
 - (6) $V_{G1-S(B)} = 1.1 \text{ V}$.
 - (7) $V_{G1-S(B)} = 1 \text{ V}$.
- $V_{G2-S} = 4 \text{ V}$; $V_{DS(A)} = V_{G1-S(A)} = 0 \text{ V}$; $T_j = 25 \text{ }^\circ\text{C}$.

Fig 18. Amplifier B: output characteristics; typical values



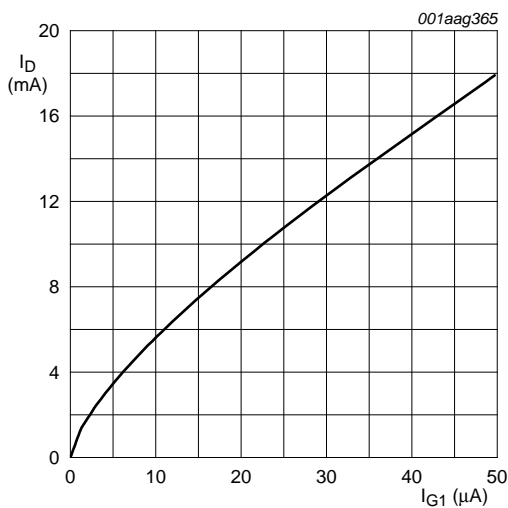
(1) $V_{G2-S} = 4 \text{ V}$.
 (2) $V_{G2-S} = 3.5 \text{ V}$.
 (3) $V_{G2-S} = 3 \text{ V}$.
 (4) $V_{G2-S} = 2.5 \text{ V}$.
 (5) $V_{G2-S} = 2 \text{ V}$.
 (6) $V_{G2-S} = 1.5 \text{ V}$.
 (7) $V_{G2-S} = 1 \text{ V}$.
 $V_{DS(B)} = 5 \text{ V}$; $V_{DS(A)} = V_{G1-S(A)} = 0 \text{ V}$; $T_j = 25 \text{ }^\circ\text{C}$.

Fig 19. Amplifier B: gate1 current as a function of gate1 voltage; typical values



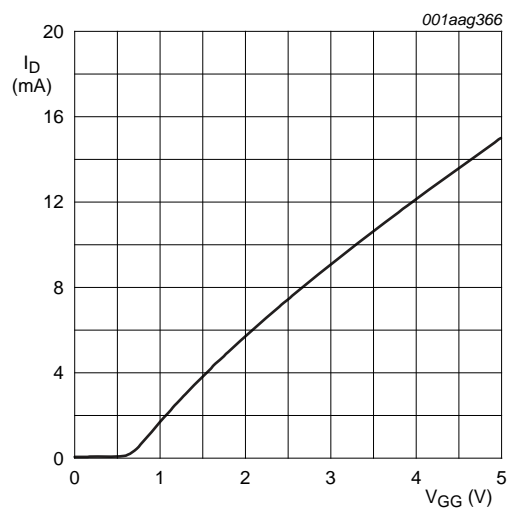
(1) $V_{G2-S} = 4 \text{ V}$.
 (2) $V_{G2-S} = 3.5 \text{ V}$.
 (3) $V_{G2-S} = 3 \text{ V}$.
 (4) $V_{G2-S} = 2.5 \text{ V}$.
 (5) $V_{G2-S} = 2 \text{ V}$.
 (6) $V_{G2-S} = 1.5 \text{ V}$.
 (7) $V_{G2-S} = 1 \text{ V}$.
 $V_{DS(B)} = 5 \text{ V}$; $V_{DS(A)} = V_{G1-S(A)} = 0 \text{ V}$; $T_j = 25 \text{ }^\circ\text{C}$.

Fig 20. Amplifier B: forward transfer admittance as a function of drain current; typical values



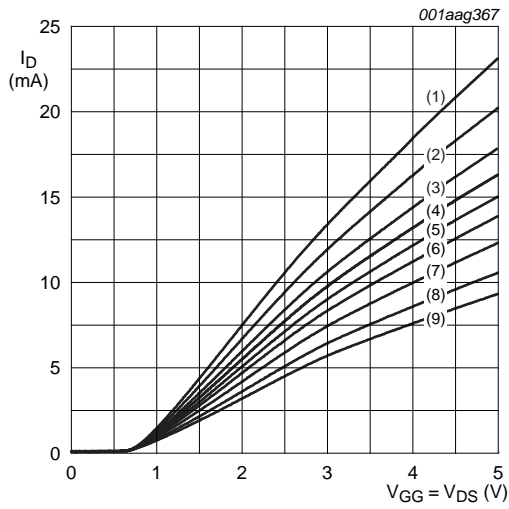
$V_{DS(B)} = 5 \text{ V}$; $V_{G2-S} = 4 \text{ V}$; $V_{DS(A)} = V_{G1-S(A)} = 0 \text{ V}$;
 $T_j = 25 \text{ }^\circ\text{C}$.

Fig 21. Amplifier B: drain current as a function of gate1 current; typical values



$V_{DS(B)} = 5 \text{ V}$; $V_{G2-S} = 4 \text{ V}$; $V_{DS(A)} = V_{G1-S(A)} = 0 \text{ V}$;
 $T_j = 25 \text{ }^\circ\text{C}$; $R_{G1} = 86 \text{ k}\Omega$ (connected to V_{GG}); see [Figure 3](#).

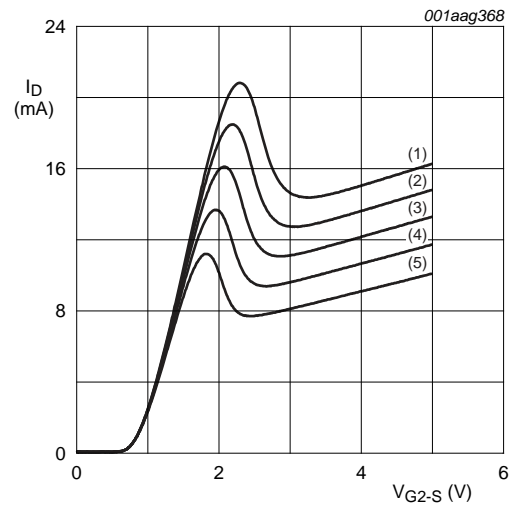
Fig 22. Amplifier B: drain current as a function of gate1 supply voltage; typical values



- (1) $R_{G1} = 47 \text{ k}\Omega$.
- (2) $R_{G1} = 56 \text{ k}\Omega$.
- (3) $R_{G1} = 68 \text{ k}\Omega$.
- (4) $R_{G1} = 82 \text{ k}\Omega$.
- (5) $R_{G1} = 86 \text{ k}\Omega$.
- (6) $R_{G1} = 100 \text{ k}\Omega$.
- (7) $R_{G1} = 120 \text{ k}\Omega$.
- (8) $R_{G1} = 150 \text{ k}\Omega$.
- (9) $R_{G1} = 180 \text{ k}\Omega$.

$V_{G2-S} = 4 \text{ V}$; $V_{DS(A)} = V_{G1-S(A)} = 0 \text{ V}$; $T_j = 25 \text{ }^\circ\text{C}$; R_{G1} is connected to V_{GG} ; see [Figure 3](#).

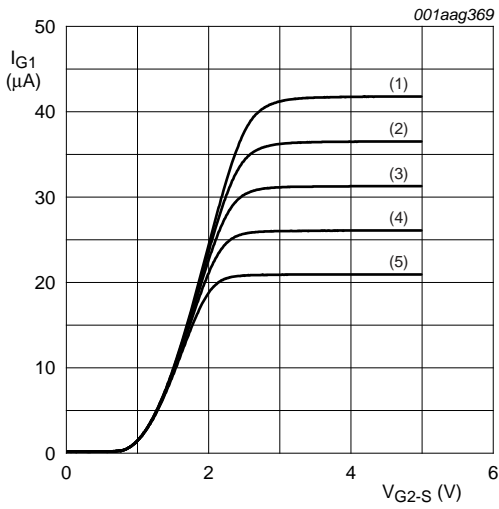
Fig 23. Amplifier B: drain current as a function of gate1 supply voltage and drain supply voltage; typical values



- (1) $V_{GG} = 5.0 \text{ V}$.
- (2) $V_{GG} = 4.5 \text{ V}$.
- (3) $V_{GG} = 4.0 \text{ V}$.
- (4) $V_{GG} = 3.5 \text{ V}$.
- (5) $V_{GG} = 3.0 \text{ V}$.

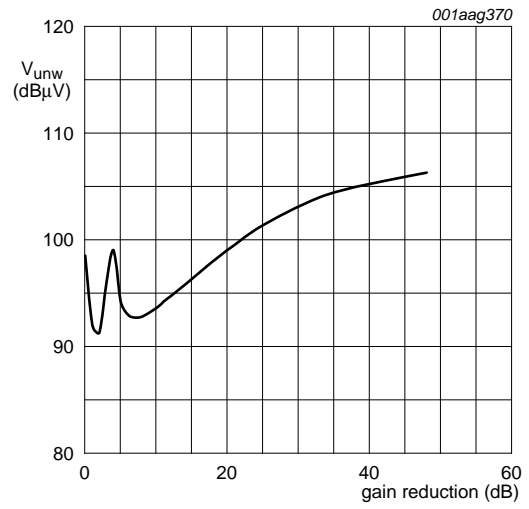
$V_{DS(B)} = 5 \text{ V}$; $V_{DS(A)} = V_{G1-S(A)} = 0 \text{ V}$; $T_j = 25 \text{ }^\circ\text{C}$; $R_{G1} = 86 \text{ k}\Omega$ (connected to V_{GG}); see [Figure 3](#).

Fig 24. Amplifier B: drain current as a function of gate2 voltage; typical values



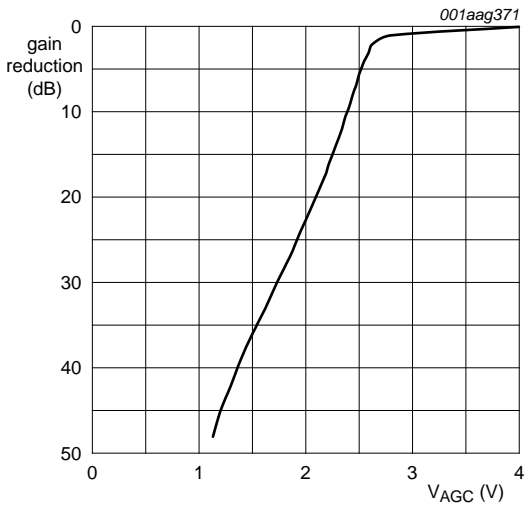
(1) $V_{GG} = 5.0\text{ V}$.
 (2) $V_{GG} = 4.5\text{ V}$.
 (3) $V_{GG} = 4.0\text{ V}$.
 (4) $V_{GG} = 3.5\text{ V}$.
 (5) $V_{GG} = 3.0\text{ V}$.
 $V_{DS(B)} = 5\text{ V}$; $V_{DS(A)} = V_{G1-S(A)} = 0\text{ V}$; $T_j = 25\text{ }^\circ\text{C}$;
 $R_{G1} = 86\text{ k}\Omega$ (connected to V_{GG}); see [Figure 3](#).

Fig 25. Amplifier B: gate1 current as a function of gate2 voltage; typical values



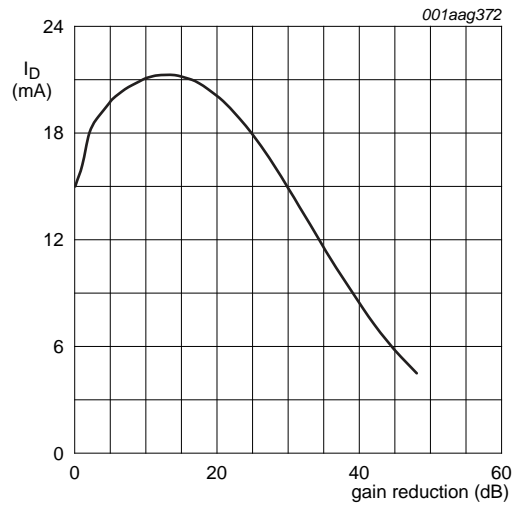
$V_{DS(B)} = 5\text{ V}$; $V_{GG} = 5\text{ V}$; $V_{DS(A)} = V_{G1-S(A)} = 0\text{ V}$;
 $R_{G1} = 86\text{ k}\Omega$ (connected to V_{GG}); $f_w = 50\text{ MHz}$;
 $f_{unw} = 60\text{ MHz}$; $T_{amb} = 25\text{ }^\circ\text{C}$; see [Figure 34](#).

Fig 26. Amplifier B: unwanted voltage for 1 % cross modulation as a function of gain reduction; typical values



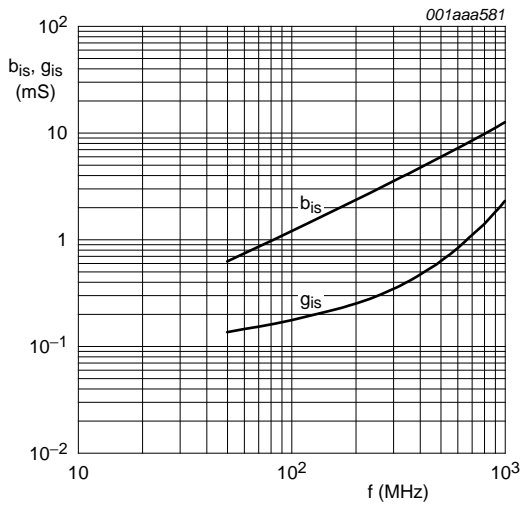
$V_{DS(B)} = 5\text{ V}$; $V_{GG} = 5\text{ V}$; $V_{DS(A)} = V_{G1-S(A)} = 0\text{ V}$;
 $R_{G1} = 86\text{ k}\Omega$ (connected to V_{GG}); $f = 50\text{ MHz}$;
 $T_{amb} = 25\text{ }^\circ\text{C}$; see [Figure 34](#).

Fig 27. Amplifier B: gain reduction as a function of AGC voltage; typical values



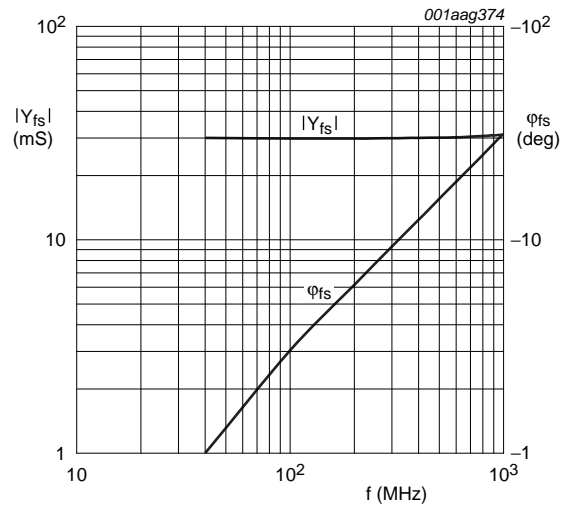
$V_{DS(B)} = 5\text{ V}$; $V_{GG} = 5\text{ V}$; $V_{DS(A)} = V_{G1-S(A)} = 0\text{ V}$;
 $R_{G1} = 86\text{ k}\Omega$ (connected to V_{GG}); $f = 50\text{ MHz}$;
 $T_{amb} = 25\text{ }^\circ\text{C}$; see [Figure 34](#).

Fig 28. Amplifier B: drain current as a function of gain reduction; typical values



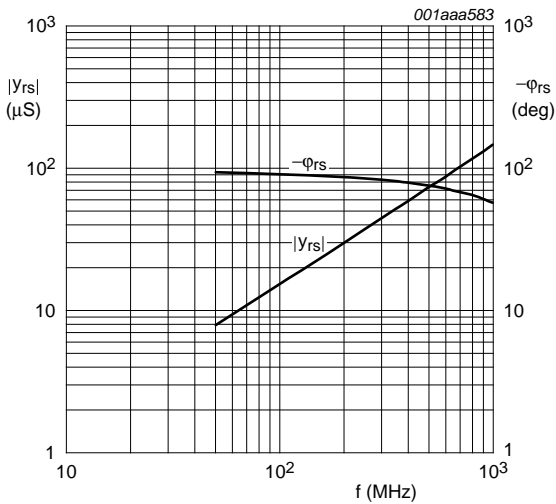
$V_{DS(B)} = 5\text{ V}; V_{G2-S} = 4\text{ V}; V_{DS(A)} = V_{G1-S(A)} = 0\text{ V}; I_{D(B)} = 15\text{ mA}$

Fig 29. Amplifier B: input admittance as a function of frequency; typical values



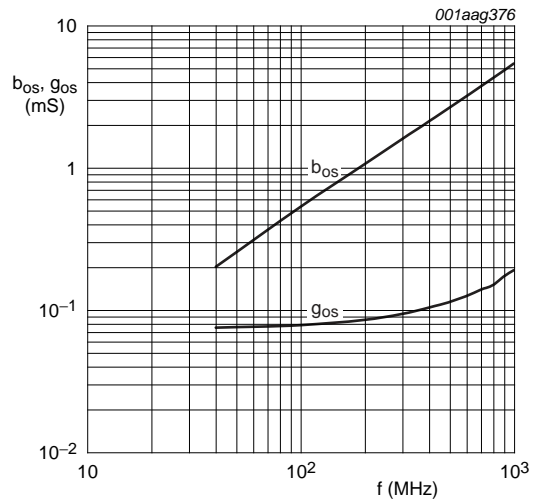
$V_{DS(B)} = 5\text{ V}; V_{G2-S} = 4\text{ V}; V_{DS(A)} = V_{G1-S(A)} = 0\text{ V}; I_{D(B)} = 15\text{ mA}$

Fig 30. Amplifier B: forward transfer admittance and phase as a function of frequency; typical values



$V_{DS(B)} = 5\text{ V}; V_{G2-S} = 4\text{ V}; V_{DS(A)} = V_{G1-S(A)} = 0\text{ V}; I_{D(B)} = 15\text{ mA}$

Fig 31. Amplifier B: reverse transfer admittance and phase as a function of frequency; typical values



$V_{DS(B)} = 5\text{ V}; V_{G2-S} = 4\text{ V}; V_{DS(A)} = V_{G1-S(A)} = 0\text{ V}; I_{D(B)} = 15\text{ mA}$

Fig 32. Amplifier B: output admittance as a function of frequency; typical values

8.2.2 Scattering parameters for amplifier B

Table 12. Scattering parameters for amplifier B

$V_{DS(B)} = 5\text{ V}$; $V_{G2-S} = 4\text{ V}$; $I_{D(B)} = 15\text{ mA}$; $V_{DS(A)} = 0\text{ V}$; $V_{G1-S(A)} = 0\text{ V}$; $T_{amb} = 25\text{ °C}$; typical values.

f (MHz)	S ₁₁		S ₂₁		S ₁₂		S ₂₂	
	Magnitude (ratio)	Angle (degree)	Magnitude (ratio)	Angle (degree)	Magnitude (ratio)	Angle (degree)	Magnitude (ratio)	Angle (degree)
40	0.9841	-4.20	2.9772	175.44	0.0005	106.03	0.9923	-1.40
100	0.9799	-7.68	2.9694	171.40	0.0011	88.52	0.9927	-2.88
200	0.9775	-15.24	2.9472	162.86	0.0023	87.60	0.9914	-5.77
300	0.9706	-22.70	2.9147	154.41	0.0034	85.98	0.9902	-8.61
400	0.9632	-30.08	2.8754	146.10	0.0046	85.09	0.9888	-11.43
500	0.9515	-37.46	2.8213	137.77	0.0056	84.03	0.9870	-14.26
600	0.9377	-44.80	2.7560	129.44	0.0065	83.30	0.9839	-17.16
700	0.9229	-52.10	2.6865	121.24	0.0075	82.99	0.9810	-20.05
800	0.9062	-59.33	2.6119	113.09	0.0084	82.08	0.9777	-22.93
900	0.8864	-66.35	2.5318	105.04	0.0091	81.36	0.9754	-25.77
1000	0.8650	-73.21	2.4437	97.11	0.0098	80.34	0.9714	-28.64

8.2.3 Noise data for amplifier B

Table 13. Noise data for amplifier B

$V_{DS(B)} = 5\text{ V}$; $V_{G2-S} = 4\text{ V}$; $I_{D(B)} = 15\text{ mA}$; $V_{DS(A)} = 0\text{ V}$; $V_{G1-S(A)} = 0\text{ V}$; $T_{amb} = 25\text{ °C}$; typical values; unless otherwise specified.

f (MHz)	NF _{min} (dB)	Γ _{opt}		r _n (Ω)
		(ratio)	(degree)	
400	1.1	0.72	22.83	0.66
800	1.4	0.68	46.42	0.64

9. Test information

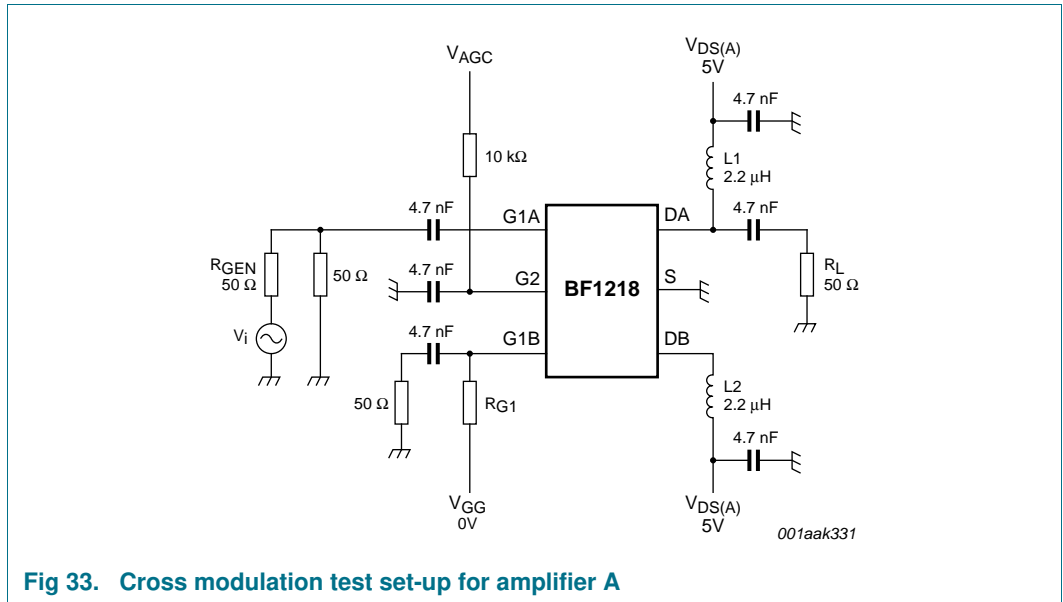


Fig 33. Cross modulation test set-up for amplifier A

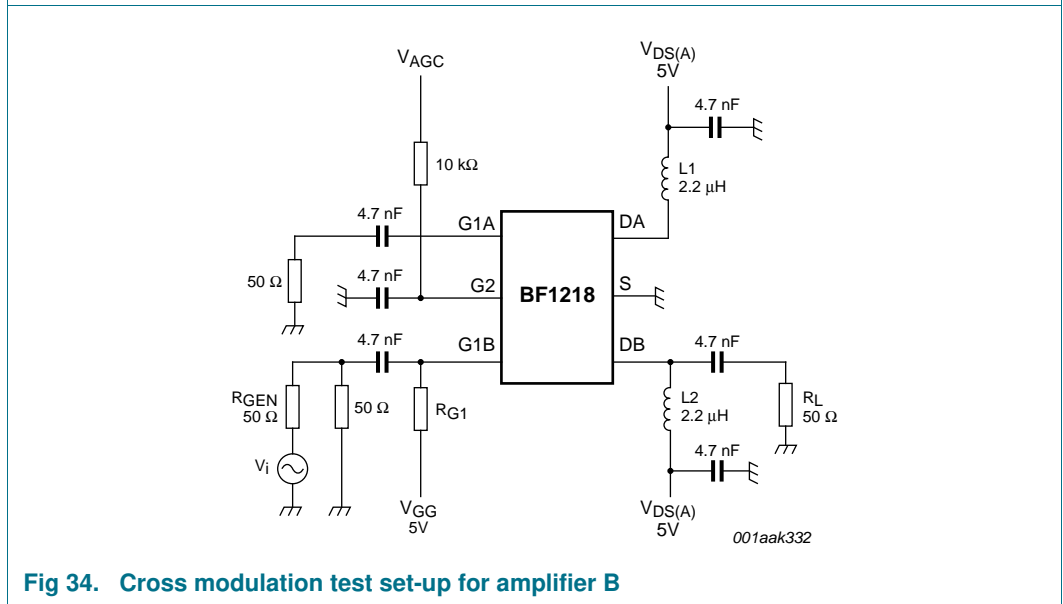


Fig 34. Cross modulation test set-up for amplifier B

10. Package outline

Plastic surface-mounted package; 6 leads

SOT363

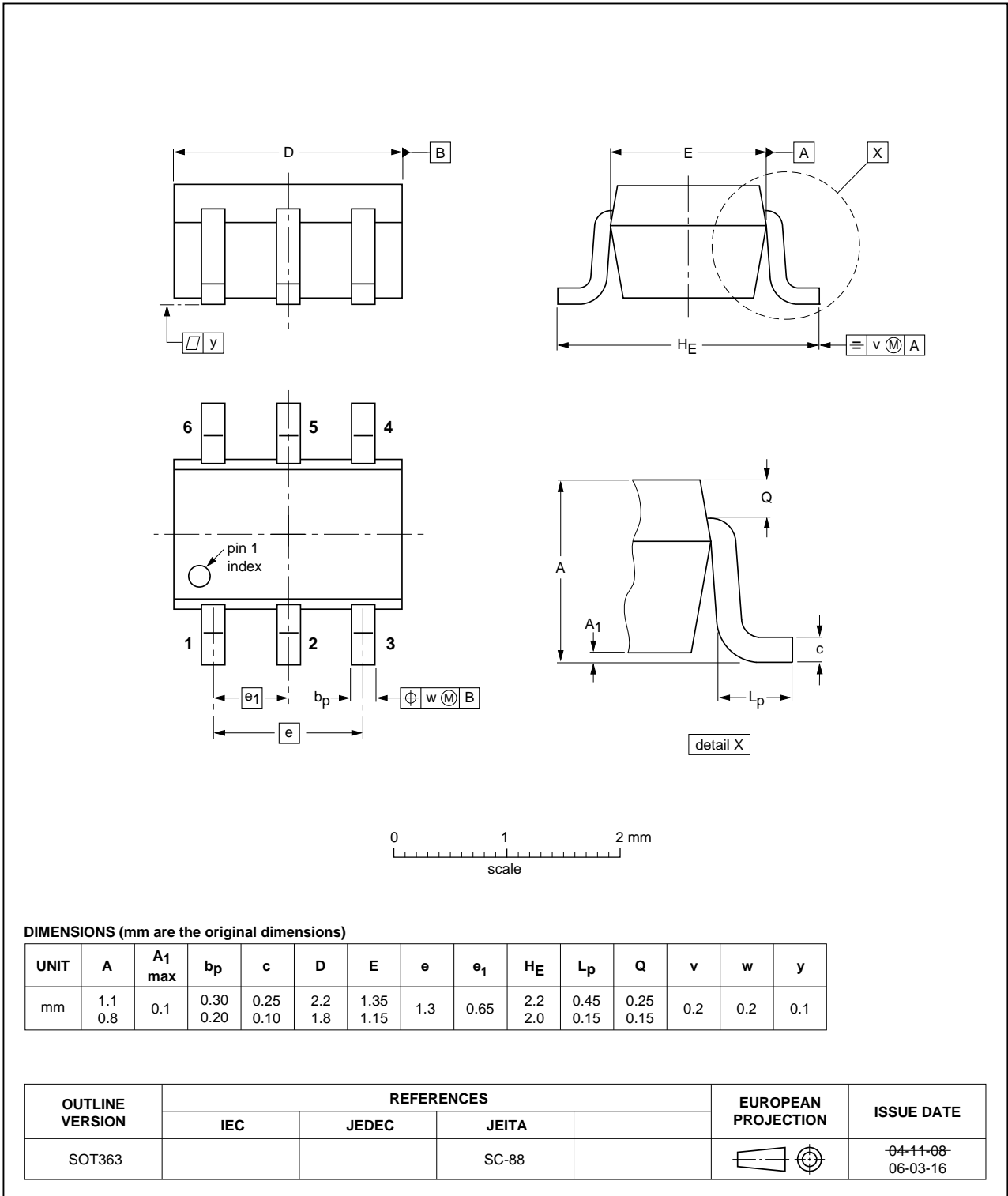


Fig 35. Package outline SOT363

11. Abbreviations

Table 14. Abbreviations

Acronym	Description
AGC	Automatic Gain Control
DC	Direct Current
MOSFET	Metal-Oxide Semiconductor Field-Effect Transistor
UHF	Ultra High Frequency
VHF	Very High Frequency

12. Revision history

Table 15. Revision history

Document ID	Release date	Data sheet status	Change notice	Supersedes
BF1218_1	20100414	Product data sheet	-	-

13. Legal information

13.1 Data sheet status

Document status ^{[1][2]}	Product status ^[3]	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
Preliminary [short] data sheet	Qualification	This document contains data from the preliminary specification.
Product [short] data sheet	Production	This document contains the product specification.

[1] Please consult the most recently issued document before initiating or completing a design.

[2] The term 'short data sheet' is explained in section "Definitions".

[3] The product status of device(s) described in this document may have changed since this document was published and may differ in case of multiple devices. The latest product status information is available on the Internet at URL <http://www.nxp.com>.

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