THS1408-EP 14-BIT, 3/8 MSPS DSP-COMPATIBLE ANALOG-TO-DIGITAL CONVERTER WITH INTERNAL REFERENCE AND PGA

SGLS129B - JULY 2002 - REVISED FEBRUARY 2010

Features

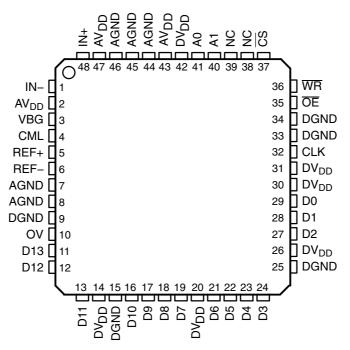
- Controlled Baseline
 - One Assembly/Test Site, One Fabrication Site
- Extended Temperature Performance up to -55°C to 125°C
- Enhanced Diminishing Manufacturing Sources (DMS) Support
- Enhanced Product Change Notification
- Qualification Pedigree[†]
- 14-Bit Resolution
- 3- and 8-MSPS Speed Grade
- Differential Nonlinearity (DNL) ±0.6 LSB Typ
- † Component qualification in accordance with JEDEC and industry standards to ensure reliable operation over an extended temperature range. This includes, but is not limited to, Highly Accelerated Stress Test (HAST) or biased 85/85, temperature cycle, autoclave or unbiased HAST, electromigration, bond intermetallic life, and mold compound life. Such qualification testing should not be viewed as justifying use of this component beyond specified performance and environmental limits.

- Integral Nonlinearity (INL) ±1.5 LSB Typ
- Internal Reference
- Differential Inputs
- Programmable Gain Amplifier
- μP-Compatible Parallel Interface
- Timing Compatible With TMS320C6000 DSP
- 3.3-V Single Supply
- Power-Down Mode
- Monolithic CMOS Design

Applications

- xDSL Front Ends
- Communication
- Industrial Control
- Instrumentation
- Automotive and Selected Military

PHP PACKAGE (TOP VIEW)



NC - No internal connection



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



THS1408-EP 14-BIT, 3/8 MSPS DSP-COMPATIBLE ANALOG-TO-DIGITAL CONVERTER WITH INTERNAL REFERENCE AND PGA

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description

The THS1408 is a 14-bit, 3/8-MSPS single-supply analog-to-digital converter (ADC), with an internal reference, differential inputs, programmable input gain, and an on-chip sample and hold amplifier.

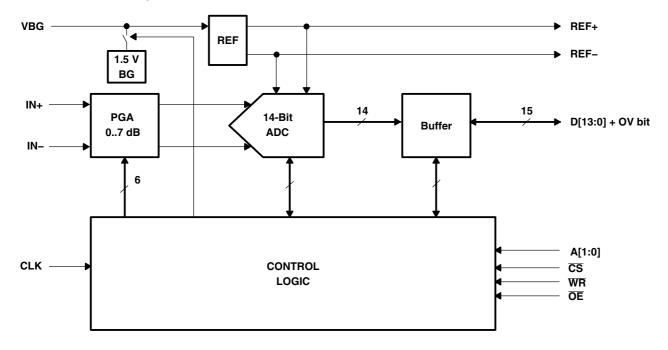
Implemented with a CMOS process, the device has outstanding price/performance and power/speed ratios. The THS1408 is designed for use with 3.3-V systems, and with a high-speed μP-compatible parallel interface, making it the first choice for solutions based on high-performance DSPs like the TI TMS320C6000 series.

ORDERING INFORMATION

T _A	PACKAGE† to 125°C PQFP – PHP	ORDERABLE PART NUMBER	TOP-SIDE MARKING		
–55°C to 125°C	PQFP – PHP	THS1408MPHPEP	THS1408ME		

[†] Package drawings, standard packing quantities, thermal data, symbolization, and PCB design guidelines are available at www.ti.com/sc/package.

functional block diagram





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Terminal Functions

TEF	RMINAL		
NAME	NO.	I/O	DESCRIPTION
A[1:0]	40, 41	I	Address input
AGND	7,8, 44, 45, 46		Analog ground
AV_{DD}	2, 43, 47		Analog power supply
CLK	32	I	Clock input
CML	4		Reference midpoint. This pin requires a 0.1-μF capacitor to AGND.
CS	37	- 1	Chip select input. Active low
DGND	9, 15, 25, 33, 34		Digital ground
DV_DD	14, 20, 26, 30, 31, 42		Digital power supply
D[13:0]	11, 12, 13, 16, 17, 18, 19, 21, 22, 23, 24, 27, 28, 29	I/O	Data inputs/outputs
NC	38, 39		No connection, do not use. Reserved
IN+	48	I	Positive differential analog input
IN-	1	I	Negative differential analog input
ŌĒ	35	I	Output enable. Active low
OV	10	0	Out of range output
REF+	5	0	Positive reference output. This pin requires a 0.1-μF capacitor to AGND.
REF-	6	0	Negative reference output. This pin requires a 0.1-μF capacitor to AGND.
VBG	3	I	Reference input. This pin requires a 1-μF capacitor to AGND.
WR	36	I	Write signal. Active low

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absolute maximum ratings over operating free-air temperature (unless otherwise noted)†

Sup	ply voltage, (AV _{DD} to AGND)		 4 V
	ply voltage, (DV _{DD} to DGND)		
	erence input voltage range, VBG		
Ana	log input voltage range		 √ to AV _{DD} + 0.3 V
Digi	tal input voltage range		 / to DV _{DD} + 0.3 V
Ope	rating free-air temperature range, T _A :	Q suffix	 40°C to 125°C
		M suffix	 55°C to 125°C
Stor	age temperature range, T _{stq}		 65°C to 150°C
Lead	d temperature 1.6 mm (1/1ĕ inch) from	case for 10 seconds	 260°C

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

THERMAL CHARACTERISTICS

	TYP	UNIT
Thermal resistance, junction-to-ambient, Θ_{JA}	28.8	°C/W
Thermal resistance, junction-to-case, $\Theta_{\mbox{\scriptsize JC}}$	0.79	°C/W

[†] Thermal resistance is modeled data, is not production tested, and is given for informational purposes only.

recommended operating conditions

			MIN	NOM	MAX	UNIT
Supply voltage, AV _{DD} , DV _{DD}			3	3.3	3.6	V
High level digital input, V _{IH}			2	3.3		V
Low level digital input, V _{IL}				0	0.8	V
Load capacitance, C _L				5	15	pF
Clock frequency, f _{CLK}			0.1	8	8	MHz
Clock duty cycle			45%	50%	55%	
On susting free sintenangueting	Q s	uffix	-40	25	125	°C
Operating free-air temperature	Ms	uffix	-55	25	125	°C



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electrical characteristics over recommended operating free-air temperature range, $AV_{DD} = DV_{DD} = 3.3 \text{ V}$ (unless otherwise noted)

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT		
Power S	Supply								
I _{DDA}	Analog supply current		AV _{DD} = 3.6 V		81	90	mA		
I _{DDD}	Digital supply current		DV _{DD} = 3.6 V		5	10	mA		
	Power		$AV_{DD} = DV_{DD} = 3.6 \text{ V}$		270	360	mW		
	Power down current				20		μА		
DC Cha	racteristics								
	Resolution				14		Bits		
DNL	Differential nonlinearity				±0.6	±1	LSB		
		THS1408 at 3 MSPS	Best fit		±2	±4			
INL	Integral nonlinearity	THS1408 at 8 MSPS	Best fit		±3.5	±10.5	LSB		
	Offset error		IN+ = IN-, PGA = 0 dB			0.3	%FSR		
	Gain error		PGA = 0 dB			1.75	%FSR		
AC Cha	racteristics								
ENOB	Effective number of bits			11.2	11.5		Bits		
			f _i = 100 kHz		-81				
THD	Total harmonic distortion		f _i = 1 MHz		-78		dB		
			f _i = 4 MHz		-77				
			f _i = 100 kHz		72				
SNR	Signal-to-noise ratio		f _i = 1 MHz	70	72		dB		
			f _i = 4 MHz		71				
			f _i = 100 kHz		70				
SINAD	Signal-to-noise ratio + distortion		f _i = 1 MHz	69	70		dB		
			f _i = 4 MHz		70				
			f _i = 100 kHz		80				
SFDR	Spurious free dynamic range		f _i = 1 MHz	71	80	_	dB		
		f _i = 4 MHz		80	_				
	Analog input bandwidth				140		MHz		

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electrical characteristics over recommended operating free-air temperature range, $AV_{DD} = DV_{DD} = 3.3 \text{ V}$ (unless otherwise noted) (continued)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Refere	nce Voltage	•				
	Bandgap voltage, internal mode		1.425	1.5	1.575	V
VBG	Input impedance			40		kΩ
	Positive reference voltage, REF+			2.5		V
	Negative reference voltage, REF-			0.5		V
	Reference difference, ∆REF, REF+ – REF–			2		V
	Accuracy, internal reference			5%		
	Temperature coefficient			40		ppm/°C
	Voltage coefficient			200		ppm/V
Analog	Inputs	•				
	Positive analog input, IN+		0		AV_{DD}	V
	Negative analog input, IN-		0		AV_{DD}	V
	Analog input voltage difference	ΔAin = IN+ – IN-, V _{ref} = REF+ – REF-	-V _{ref}		V _{ref}	V
	Input impedance			25		kΩ
	PGA range		0		7	dB
	PGA step size			1		dB
	PGA gain error				±0.25	dB
Digital	Inputs	·	•			
V_{IH}	High-level digital input		2			V
V_{IL}	Low-level digital input				0.8	V
	Input capacitance			5		pF
	Input current	0		±1	μΑ	
Digital	Outputs	·	•			
V_{OH}	High-level digital output	I _{OH} = 50 μA	2.6			V
V_{OL}	Low-level digital output	I _{OL} = 50 μA			0.4	V
l _{OZ}	Output current, high impedance				±10	μΑ
Clock	Timing (CS low)		•			•
f _{CLK}	Clock frequency		0.1†	8	8	MHz
t _d	Output delay time				25	ns
	Latency			9.5		Cycles

[†] This parameter is not production tested.



PARAMETER MEASUREMENT INFORMATION

sample timing

The core is based on a pipeline architecture with a latency of 9.5 samples. The conversion results appear on the digital output 9.5 clock cycles after the input signal was sampled.

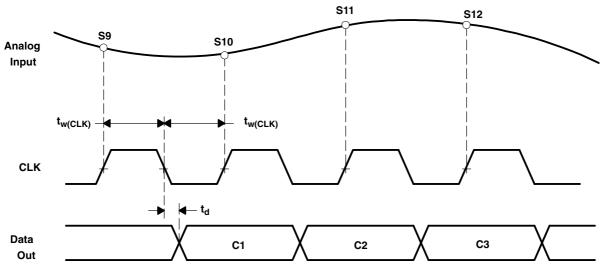


Figure 1. Sample Timing

The parallel interface of the ADC features 3-state buffers, making it possible to directly connect it to a data bus. The output buffers are enabled by driving the OE input low.

Besides the sample results, it is also possible to read back the values of the control register, the PGA register, and the offset register. Which register is read is determined by the address inputs A[1,0]. The ADC results are available at address 0.

The timing of the control signals is described in the following sections.

PARAMETER MEASUREMENT INFORMATION

read timing (15-pF load)

	PARAMETER	MIN	TYP	MAX	UNIT
t _{su(OE-ACS)}	Address and chip select setup time	4			ns
t _{en}	Output enable			15	ns
t _{dis}	Output disable		10		ns
t _{h(A)}	Address hold time	1			ns
t _{h(CS)}	Chip select hold time	0			ns

NOTE: All timing parameters refer to a 50% level.

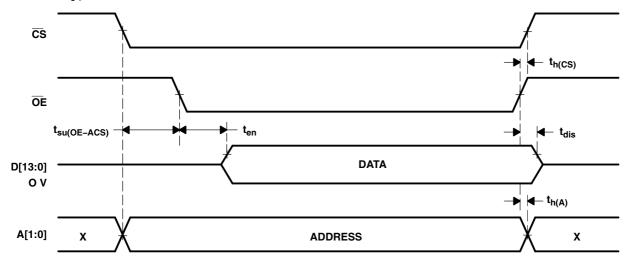


Figure 2. Read Timing



PARAMETER MEASUREMENT INFORMATION

write timing (15-pF load)

t _{su(WE-CS)}	Chip select setup time	4			ns				
t _{su(DA)}	Data and address setup time	29			ns				
t _{h(DA)}	Data and address hold time	0			ns				
t _{h(CS)}	Chip select hold time	0			ns				
t _{wH(WE)}	Write pulse duration high	15			ns				

NOTE: All timing parameters refer to a 50% level.

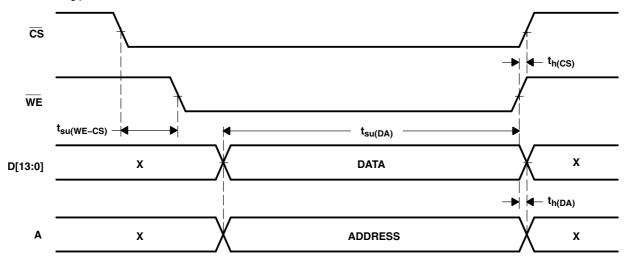
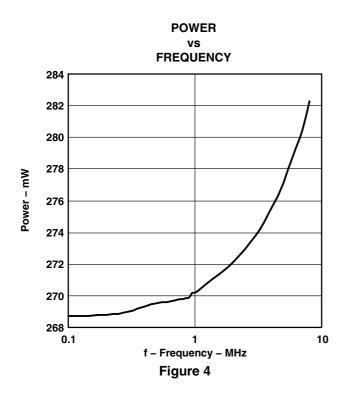
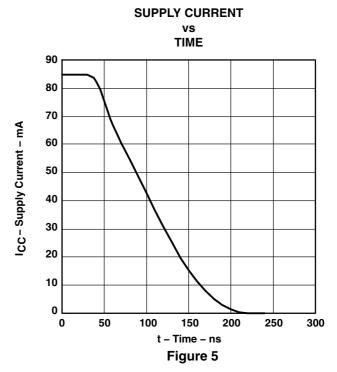


Figure 3. Write Timing

TYPICAL CHARACTERISTICS





TYPICAL CHARACTERISTICS

FAST FOURIER TRANSFORM

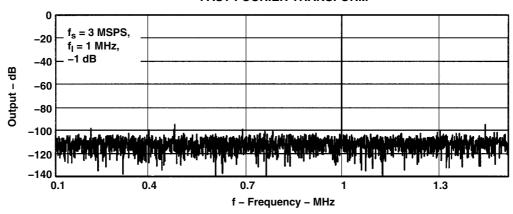


Figure 6

FAST FOURIER TRANSFORM

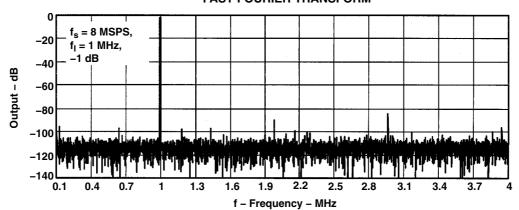


Figure 7

TYPICAL CHARACTERISTICS

INTEGRAL NONLINEARITY 2 INL - Integral Nonlinearity - LSB $f_s = 3 MSPS$ 1.5 0.5 -0.5 -1 -1.5 2048 4096 8192 0 6144 10240 12288 14336 16384 Samples

Figure 8

INTEGRAL NONLINEARITY INL - Integral Nonlinearity - LSB f_s = 8 MSPS 0 -2 -3 0 2048 4096 6144 8192 10240 12288 14336 16384 Samples

Figure 9



TYPICAL CHARACTERISTICS

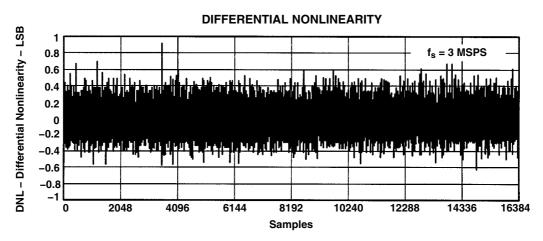


Figure 10

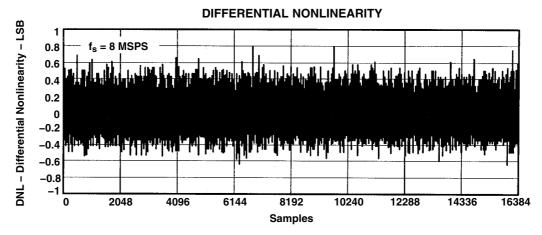
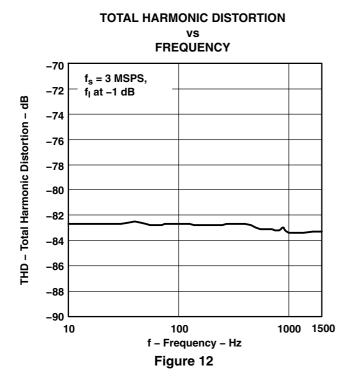
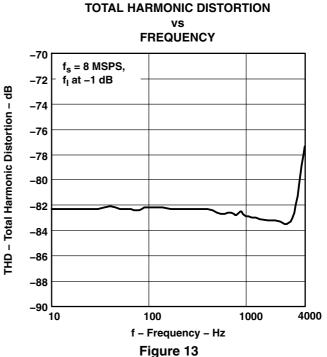
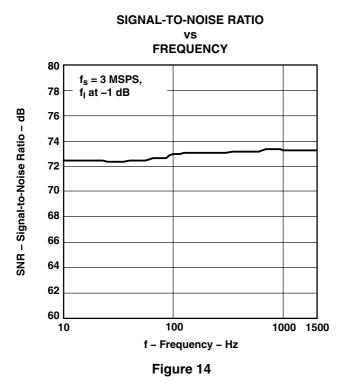


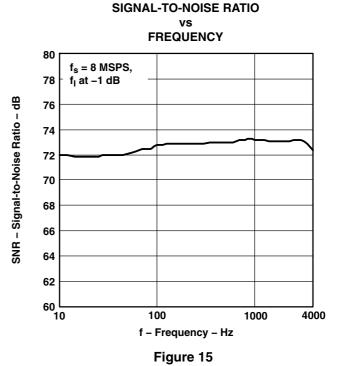
Figure 11

TYPICAL CHARACTERISTICS









PRINCIPLES OF OPERATION

registers

The device contains several registers. The A register is selected by the values of bits A1 and A0:

A1	A0	Register				
0	0	Conversion result				
0	1	PGA				
1	0	Offset				
1	1	Control				

Tables 1 and 2 describe how to read the conversion results and how to configure the data converter. The default values (where applicable) show the state after a power-on reset.

Table 1. Conversion Result Register, Address 0, Read

BIT	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
Function	MSB													LSB

The output can be configured for two's complement or straight binary format (see D11/control register).

The output code is given by:

2s complement: Straight binary:

-8192 at $\Delta IN = -\Delta REF$ 0 at $\Delta IN = -\Delta REF$

0 at $\Delta IN = 0$ 8192 at $\Delta IN = 0$

8191 $\triangle IN = -\triangle REF - 1 LSB$ 16383 at $\triangle IN = -\triangle REF - 1 LSB$

 $1 LSB = \frac{2\Delta REF}{16384}$

Table 2. PGA Gain Register, Address 1, Read/Write

BIT	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
Function	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	G2	G1	G0
Default	0	0	0	0	0	0	0	0	0	0	0	0	0	0

The PGA gain is determined by writing to G2-0.

Gain (dB) = $1dB \times G2-0$. max = 7dB. The range of G2-0 is 0 to 7.

Table 3. Offset Register, Address 2, Read/Write

BIT	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
Function	Х	Х	Х	Х	Х	Х	MSB							LSB
Default	0	0	0	0	0	0	0	0	0	0	0	0	0	0

The offset correction range is from -128 to 127 LSB. This value is added to the conversion results from the ADC.



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PRINCIPLES OF OPERATION

Table 4. Control Register, Address 3, Read

BIT	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
Function	PWD	REF	FOR	TM2	TM1	TM0	OFF	RES						

Table 5. Control Register, Address 3, Write

BIT	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
Function	PWD	REF	FOR	TM2	TM1	TMO	OFF	RES						
Default	0	0	0	0	0	0	0	0	0	0	0	0	0	0

PWD: Power down 0 = normal operation 1 = power down

REF: Reference select 0 = internal reference 1 = external referenceFOR: Output format 0 = straight binary 1 = 2s complement

TM2-0: Test mode 000 = normal operation

001 = both inputs = REF-

010 = IN+ at $V_{ref}/2$, IN- at REF-011 = IN+ at REF+, IN- at REF-

100 = normal operation101 = both inputs = REF+

110 = IN+ at REF-, IN- at $V_{ref}/2$ 111 = IN+ at REF-, IN- at REF+

OF: Offset correction 0 = enable 1 = disable

RES Reserved Must be set to 0.

APPLICATION INFORMATION

driving the analog input

The ADC has a fully differential input. A differential input is advantageous with respect to SNR, SFDR, and THD performance because the signal peak-to-peak level is 50% of a comparable single-ended input.

There are three basic input configurations:

- Fully differential
- Transformer coupled single-ended to differential
- Single-ended

fully differential configuration

In this configuration, the ADC converts the difference (Δ IN) of the two input signals on IN+ and IN-.

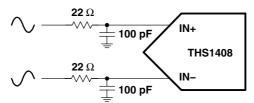


Figure 16. Differential Input

The resistors and capacitors on the inputs decouple the driving source output from the ADC input and also serve as first order low pass filters to attenuate out of band noise.

The input range on both inputs is 0 V to AV_{DD}. The full-scale value is determined by the voltage reference. The positive full-scale output is reached, if Δ IN equals Δ REF, the negative full-scale output is reached, if Δ IN equals $-\Delta$ REF.

∆IN [V]	OUTPUT
–∆REF	– full scale
0	0
∆REF	+ full scale

APPLICATION INFORMATION

transformer coupled single-ended to differential configuration

If the application requires the best SNR, SFDR, and THD performance, the input should be transformer coupled.

The signal amplitude on both inputs of the ADC is one half as high as in a single-ended configuration thus increasing the ADC ac performance.

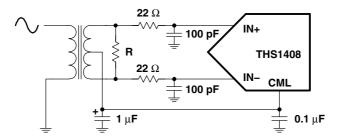


Figure 17. Transformer Coupled

The following table shows the input voltages for negative full-scale output, zero output, and positive full-scale output:

IN [V _{PEAK}]	OUTPUT [PEAK]
–∆REF	 full scale[†]
0	0
ΔREF	+ full scale [†]

† n = 1 (winding ratio)

The resistor R of the transformer coupled input configuration must be set to match the signal source impedance R = n^2 Rs, where Rs is the source impedance and n is the transformer winding ratio.



APPLICATION INFORMATION

single-ended configuration

In this configuration, the input signal is level shifted by $\Delta REF/2$.

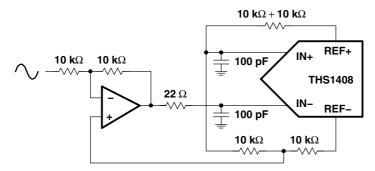


Figure 18. Single-Ended With Level Shift

The following table shows the input voltages for negative full-scale output, zero output, and positive full-scale output:

∆IN+ [V]	OUTPUT
–∆REF	- full scale
0	0
ΔREF	+ full scale

Note that the resistors of the op-amp and the op-amp all introduce gain and offset errors. Those errors can be trimmed by varying the values of the resistors.

Because of the added offset, the op-amp does not necessarily operate in the best region of its transfer curve (best linearity around zero) and, therefore, may introduce unacceptable distortion. For ac signals, an alternative is described in the following section.



APPLICATION INFORMATION

AC-coupled single-ended configuration

If the application does not require the signal bandwidth to include dc, the level shift shown in Figure 4 is not necessary.

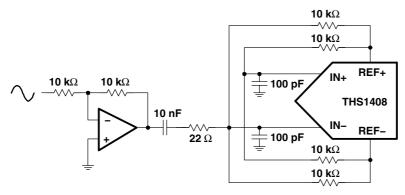


Figure 19. Single-Ended With Level Shift

Because the signal swing on the op-amp is centered around ground, it is more likely that the signal stays within the linear region of the op-amp transfer function, thus increasing the overall ac performance.

IN [V _{PEAK}]	OUTPUT [PEAK]
–∆REF	- full scale
0	0
ΔREF	+ full scale

Compared to the transformer-coupled configuration, the swing on IN- is twice as big, which can decrease the ac performance (SNR, SFD, and THD).



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APPLICATION INFORMATION

internal/external reference operation

The THS1408 ADC can either be operated using the built-in band gap reference or using an external precision reference in case very high dc accuracy is needed.

The REF+ and REF+ outputs are given by:

REF + = VBG
$$\left(1 + \frac{2}{3}\right)$$
 and REF- = VBG $\left(1 - \frac{2}{3}\right)$

If the built-in reference is used, VBG equals 1.5 V, which results in REF+ = 2.5 V, REF- = 0.5 V and Δ REF = 2 V.

The internal reference can be disabled by writing 1 to D12 (REF) in the control register (address 3). The band gap reference is then disconnected and can be substituted by a voltage on the VBG pin.

programmable gain amplifier

The on-chip programmable gain amplifier (PGA) has eight gain settings. The gain can be changed by writing to the PGA gain register (address 1). The range is 0 to 7 dB in steps of one dB.

out of range indication

The OV output of the ADC indicates an out of range condition. Every time the difference on the analog inputs exceeds the differential reference, this signal is asserted. This signal is updated the same way as the digital data outputs and therefore subject to the same pipeline delay.

offset compensation

With the offset register it is possible to automatically compensate system offset errors, including errors caused by additional signal conditioning circuitry. If the offset compensation is enabled (D7 (OFF) in the control register), the value in the offset register (address 2) is automatically subtracted from the output of the ADC.

In order to set the correct value of the offset compensation register, the ADC result when the input signal is 0 must be read by the host processor and written to the offset register (address 2).

test modes

The ADC core operation can be tested by selecting one of the available test modes (see control register description). The test modes apply various voltages to the differential input depending on the setting in the control register.

digital I/O

The digital inputs and outputs of the THS1408 ADC are 3-V CMOS compatible. In order to avoid current feed back errors, the capacitive load on the digital outputs should be as low as possible (50 pF max). Series resistors (100 Ω) on the digital outputs can improve the performance by limiting the current during output transitions.

The parallel interface of the THS1408 ADC features 3-state buffers, making it possible to directly connect it to a data bus. The output buffers are enabled by driving the \overline{OE} input low.

Refer to the read and write timing diagrams in the parameter measurement information section for information on read and write access.





PACKAGE OPTION ADDENDUM

10-Dec-2020

PACKAGING INFORMATION

www.ti.com

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
THS1408MPHPEP	ACTIVE	HTQFP	PHP	48	250	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-55 to 125	THS1408ME	Samples
V62/03608-03XE	ACTIVE	HTQFP	PHP	48	250	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-55 to 125	THS1408ME	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.



PACKAGE OPTION ADDENDUM

10-Dec-2020

OTHER QUALIFIED VERSIONS OF THS1408-EP:

• Military: THS1408M

NOTE: Qualified Version Definitions:

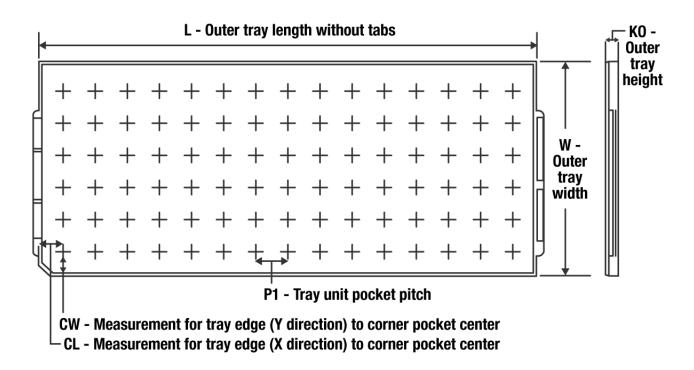
• Catalog - TI's standard catalog product

• Military - QML certified for Military and Defense Applications



www.ti.com 5-Jan-2022

TRAY



Chamfer on Tray corner indicates Pin 1 orientation of packed units.

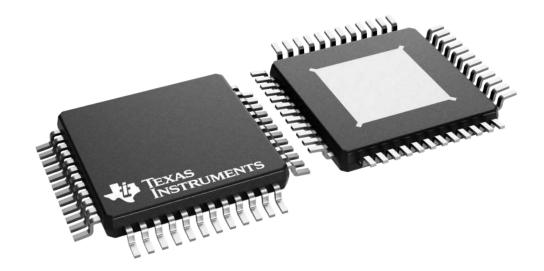
*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	Unit array matrix	Max temperature (°C)	L (mm)	W (mm)	Κ0 (μm)	P1 (mm)	CL (mm)	CW (mm)
THS1408MPHPEP	PHP	HTQFP	48	250	10 x 25	150	315	135.9	7620	12.2	11.1	11.25
V62/03608-03XE	PHP	HTQFP	48	250	10 x 25	150	315	135.9	7620	12.2	11.1	11.25

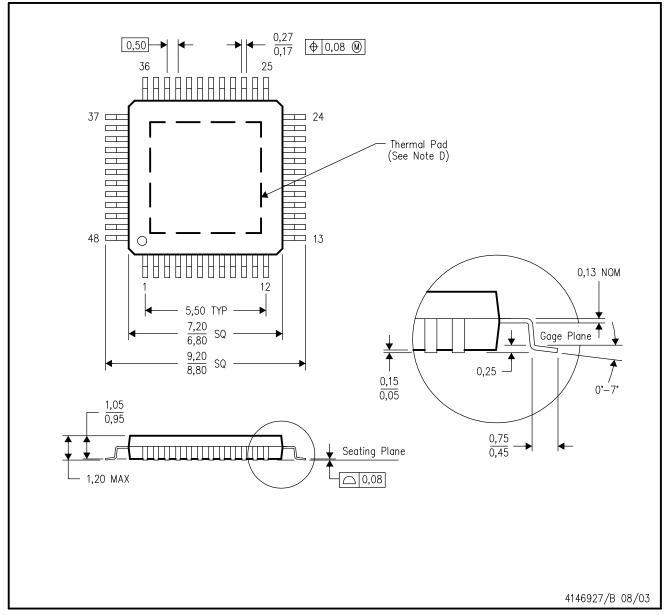
7 x 7, 0.5 mm pitch

QUAD FLATPACK

This image is a representation of the package family, actual package may vary. Refer to the product data sheet for package details.



PowerPAD™ PLASTIC QUAD FLATPACK



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at www.ti.com www.ti.com.
- E. Falls within JEDEC MS-026

PowerPAD is a trademark of Texas Instruments.



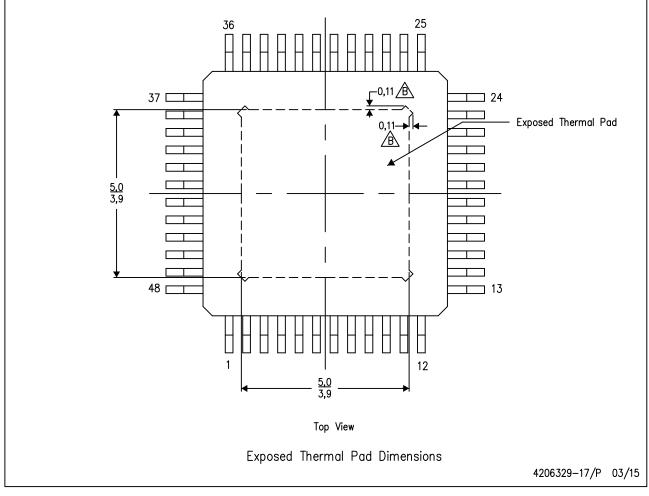
PowerPAD™ PLASTIC QUAD FLATPACK

THERMAL INFORMATION

This PowerPAD $^{\mathbf{m}}$ package incorporates an exposed thermal pad that is designed to be attached to a printed circuit board (PCB). The thermal pad must be soldered directly to the PCB. After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For additional information on the PowerPAD package and how to take advantage of its heat dissipating abilities, refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 and Application Brief, PowerPAD Made Easy, Texas Instruments Literature No. SLMA004. Both documents are available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



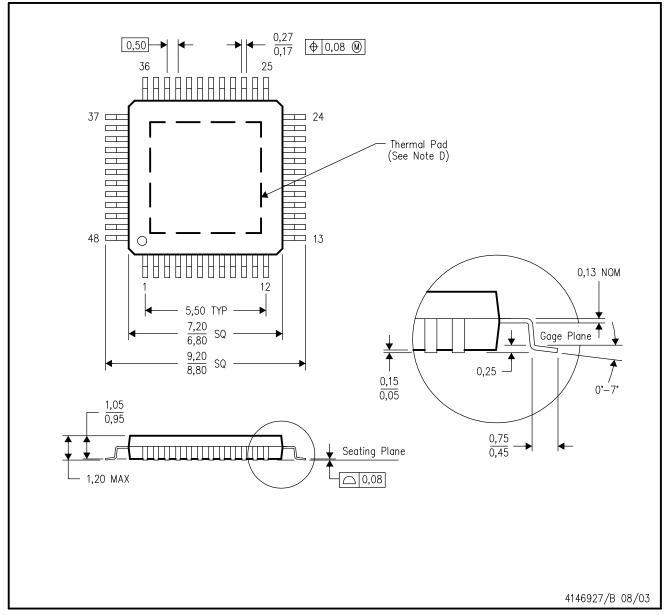
NOTE: A. All linear dimensions are in millimeters

\(\hat{\text{R}} \) Tie strap features may not be present.

Developed to the developed of Terror leaf records



PowerPAD™ PLASTIC QUAD FLATPACK



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at www.ti.com www.ti.com.
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PowerPAD is a trademark of Texas Instruments.



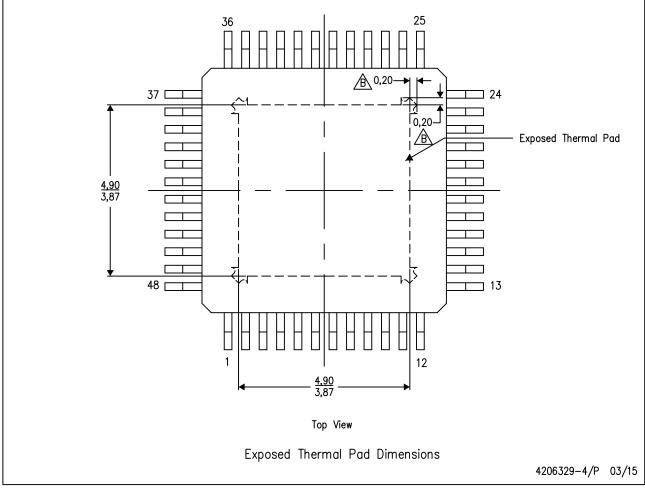
PowerPAD™ PLASTIC QUAD FLATPACK

THERMAL INFORMATION

This PowerPAD $^{\mathbf{M}}$ package incorporates an exposed thermal pad that is designed to be attached to a printed circuit board (PCB). The thermal pad must be soldered directly to the PCB. After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

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The exposed thermal pad dimensions for this package are shown in the following illustration.



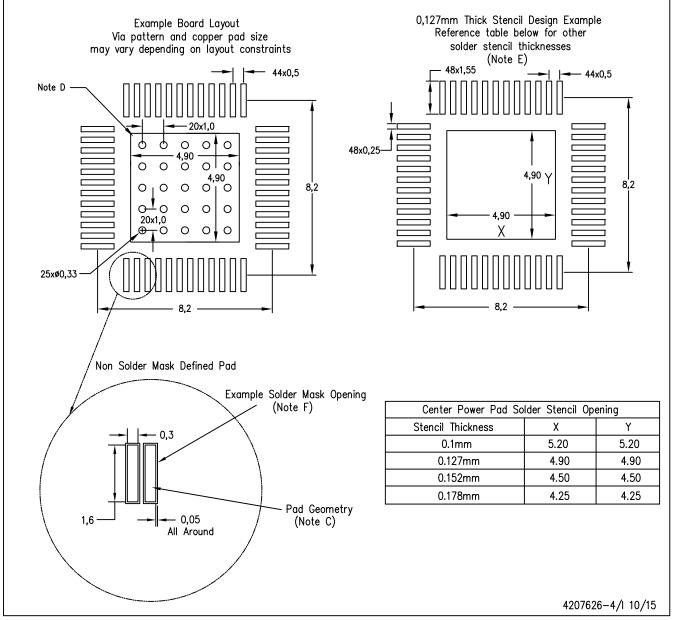
NOTE: A. All linear dimensions are in millimeters

B Tie strap features may not be present.

PowerPAD is a trademark of Texas Instruments



PowerPAD™ PLASTIC QUAD FLATPACK



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002, SLMA004, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com http://www.ti.com.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- F. Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting options for vias placed in the thermal pad.

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