

# **High Performance 8-Bit Microcontrollers**

# **Z8 Encore! XP<sup>®</sup> 64K Series Flash Microcontrollers**

**Product Specification** 

PS019919-1207





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## **Revision History**

Each instance in the Revision History reflects a change to this document from its previous revision. For more details, refer to the corresponding pages or appropriate links given in the table below.

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December 2007	19	Updated Zilog logo, Disclaimer section, and implemented style guide. Updated Table 112. Changed Z8 Encore! 64K Series to Z8 Encore! XP 64K Series Flash Microcontrollers throughout the document.	All
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# **Manual Objectives**

This Product Specification provides detailed operating information for the Flash devices within Zilog's Z8 Encore! XP<sup>®</sup> 64K Series Flash Microcontrollers Microcontroller (MCU) products. Within this document, the Z8F642x, Z8F482x, Z8F322x, Z8F242x, and Z8F162x devices are referred to collectively as the Z8 Encore! XP<sup>®</sup> 64K Series Flash Microcontrollers unless specifically stated otherwise.

#### **About This Manual**

Zilog<sup>®</sup> recommends that you read and understand everything in this manual before setting up and using the product. However, we recognize that there are different styles of learning. Therefore, we have designed this Product Specification to be used either as a *how to* procedural manual or a reference guide to important data.

#### **Intended Audience**

This document is written for Zilog customers who are experienced at working with micro-controllers, integrated circuits, or printed circuit assemblies.

#### **Manual Conventions**

The following assumptions and conventions are adopted to provide clarity and ease of use:

#### **Courier Typeface**

Commands, code lines and fragments, bits, equations, hexadecimal addresses, and various executable items are distinguished from general text by the use of the Courier typeface. Where the use of the font is not indicated, as in the Index, the name of the entity is presented in upper case.

• Example: FLAGS[1] is smrf.

#### **Hexadecimal Values**

Hexadecimal values are designated by uppercase H suffix and appear in the Courier typeface.

• Example: R1 is set to F8H.

#### **Brackets**

The square brackets, [], indicate a register or bus.

• Example: For the register R1[7:0], R1 is an 8-bit register, R1[7] is the most significant bit, and R1[0] is the least significant bit.

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#### **Braces**

The curly braces, { }, indicate a single register or bus created by concatenating some combination of smaller registers, buses, or individual bits.

• Example: The 12-bit register address {0H, RP[7:4], R1[3:0]} is composed of a 4-bit hexadecimal value (0H) and two 4-bit register values taken from the Register Pointer (RP) and Working Register R1. 0H is the most-significant nibble (4-bit value) of the 12-bit register, and R1[3:0] is the least significant nibble of the 12-bit register.

#### **Parentheses**

The parentheses, (), indicate an indirect register address lookup.

• Example: (R1) is the memory location referenced by the address contained in the Working Register R1.

#### Parentheses/Bracket Combinations

The parentheses, (), indicate an indirect register address lookup and the square brackets, [], indicate a register or bus.

• Example: Assume PC[15:0] contains the value 1234h. (PC[15:0]) then refers to the contents of the memory location at address 1234h.

#### Use of the Words Set, Reset and Clear

The word *set* implies that a register bit or a condition contains a logical 1. The words re*set* or *clear* imply that a register bit or a condition contains a logical 0. When either of these terms is followed by a number, the word *logical* may not be included; however, it is implied.

### **Notation for Bits and Similar Registers**

A field of bits within a register is designated as: Register [n:n].

• Example: ADDR[15:0] refers to bits 15 through bit 0 of the Address.

#### Use of the Terms LSB, MSB, Isb, and msb

In this document, the terms *LSB* and *MSB*, when appearing in upper case, mean *least* significant byte and most significant byte, respectively. The lowercase forms, *lsb* and *msb*, mean *least* significant bit and most significant bit, respectively.

## **Use of Initial Uppercase Letters**

Initial uppercase letters designate settings and conditions in general text.

- Example 1: The receiver forces the SCL line to Low.
- Example 2: The Master can generate a Stop condition to abort the transfer.

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## **Use of All Uppercase Letters**

The use of all uppercase letters designates the names of states, modes, and commands.

- Example 1: The bus is considered BUSY after the Start condition.
- Example 2: A START command triggers the processing of the initialization sequence.
- Example 3: STOP mode.

## **Bit Numbering**

Bits are numbered from 0 to n-1 where n indicates the total number of bits. For example, the 8 bits of a register are numbered from 0 to 7.

## **Safeguards**

It is important that you understand the following safety terms, which are defined here.



Caution:

Indicates a procedure or file may become corrupted if you do not follow directions.

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## Introduction

Zilog's Z8 Encore! XP MCU family of products are a line of Zilog<sup>®</sup> microcontroller products based upon the 8-bit eZ8 CPU. The Z8 Encore! XP<sup>®</sup> 64K Series Flash Microcontrollers, hereafter referred to collectively as the Z8 Encore! XP or the 64K Series adds Flash memory to Zilog's extensive line of 8-bit microcontrollers. The Flash in-circuit programming capability allows for faster development time and program changes in the field. The new eZ8<sup>TM</sup> CPU is upward compatible with existing Z8<sup>®</sup> instructions. The rich-peripheral set of the Z8 Encore! XP makes it suitable for a variety of applications including motor control, security systems, home appliances, personal electronic devices, and sensors.

#### **Features**

The features of Z8 Encore! XP 64K Series Flash Microcontrollers include:

- 20 MHz eZ8 CPU
- Up to 64 KB Flash with in-circuit programming capability
- Up to 4 KB register RAM
- 12-channel, 10-bit Analog-to-Digital Converter (ADC)
- Two full-duplex 9-bit UARTs with bus transceiver Driver Enable control
- Inter-integrated circuit (I<sup>2</sup>C)
- Serial Peripheral Interface (SPI)
- Two Infrared Data Association (IrDA)-compliant infrared encoder/decoders
- Up to four 16-bit timers with capture, compare, and PWM capability
- Watchdog Timer (WDT) with internal RC oscillator
- Three-channel DMA
- Up to 60 input/output (I/O) pins
- 24 interrupts with configurable priority
- On-Chip Debugger
- Voltage Brownout (VBO) Protection
- Power-On Reset (POR)
- Operating voltage of 3.0 V to 3.6 V with 5 V-tolerant inputs
- 0 °C to +70 °C, –40 °C to +105 °C, and –40 °C to +125 °C operating temperature ranges

## **Part Selection Guide**

Table 1 identifies the basic features and package styles available for each device within the Z8 Encore! XP product line.

Table 1. Z8 Encore! XP 64K Series Flash Microcontrollers Part Selection Guide

Part Number	Flash (KB)	RAM (KB)	I/O	16-bit Timers with PWM	ADC Inputs	UARTs with IrDA	I <sup>2</sup> C	SPI		64/68-pin packages	80-pin package
Z8F1621	16	2	31	3	8	2	1	1	Х		
Z8F1622	16	2	46	4	12	2	1	1		Х	
Z8F2421	24	2	31	3	8	2	1	1	Х		
Z8F2422	24	2	46	4	12	2	1	1		Х	
Z8F3221	32	2	31	3	8	2	1	1	Х		
Z8F3222	32	2	46	4	12	2	1	1		Х	
Z8F4821	48	4	31	3	8	2	1	1	Χ		
Z8F4822	48	4	46	4	12	2	1	1		Х	
Z8F4823	48	4	60	4	12	2	1	1			Х
Z8F6421	64	4	31	3	8	2	1	1	Х		
Z8F6422	64	4	46	4	12	2	1	1		Х	
Z8F6423	64	4	60	4	12	2	1	1			Х
Die Form Sales	Contact Zilog <sup>®</sup>										

## **Block Diagram**

Figure 1 displays the block diagram of the architecture of the Z8 Encore! XP 64K Series Flash Microcontrollers.

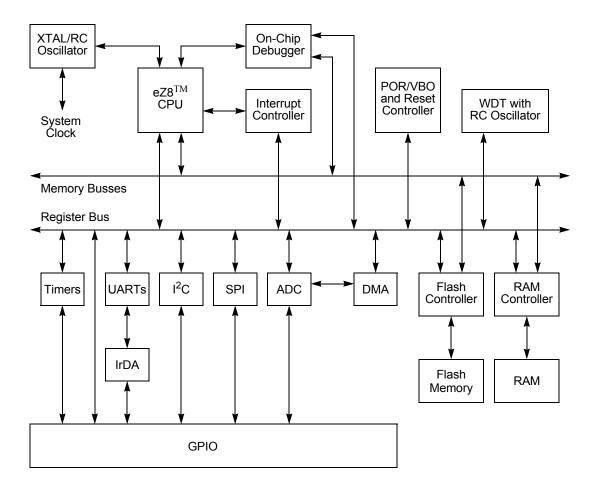


Figure 1. Z8 Encore! XP 64K Series Flash Microcontrollers Block Diagram

## **CPU and Peripheral Overview**

## eZ8<sup>™</sup> CPU Features

The latest 8-bit eZ8 CPU meets the continuing demand for faster and more code-efficient microcontrollers. The eZ8 CPU executes a superset of the original Z8<sup>®</sup> instruction set.

The eZ8 CPU features include:

- Direct register-to-register architecture allows each register to function as an accumulator, improving execution time and decreasing the required Program Memory
- Software stack allows much greater depth in subroutine calls and interrupts than hardware stacks
- Compatible with existing Z8 code
- Expanded internal Register File allows access of up to 4 KB
- New instructions improve execution efficiency for code developed using higher-level programming languages, including C
- Pipelined instruction fetch and execution
- New instructions for improved performance including BIT, BSWAP, BTJ, CPC, LDC, LDCI, LEA, MULT, and SRL
- New instructions support 12-bit linear addressing of the Register File
- Up to 10 MIPS operation
- C-Compiler friendly
- 2 to 9 clock cycles per instruction

For more information on the eZ8 CPU, refer to  $eZ8^{TM}$  CPU Core User Manual (UM0128) available for download at <a href="https://www.zilog.com">www.zilog.com</a>.

## **General-Purpose Input/Output**

The 64K Series features seven 8-bit ports (Ports A-G) and one 4-bit port (Port H) for general-purpose input/output (GPIO). Each pin is individually programmable. All ports (except B and H) support 5 V-tolerant inputs.

### Flash Controller

The Flash Controller programs and erases the Flash memory.

## 10-Bit Analog-to-Digital Converter

The Analog-to-Digital Converter converts an analog input signal to a 10-bit binary number. The ADC accepts inputs from up to 12 different analog input sources.

#### **UARTs**

Each UART is full-duplex and capable of handling asynchronous data transfers. The UARTs support 8- and 9-bit data modes, selectable parity, and an efficient bus transceiver Driver Enable signal for controlling a multi-transceiver bus, such as RS-485.

## $I^2C$

The I<sup>2</sup>C controller makes the Z8 Encore! XP compatible with the I<sup>2</sup>C protocol. The I<sup>2</sup>C controller consists of two bidirectional bus lines, a serial data (SDA) line and a serial clock (SCL) line.

## **Serial Peripheral Interface**

The serial peripheral interface allows the Z8 Encore! XP to exchange data between other peripheral devices such as EEPROMs, A/D converters and ISDN devices. The SPI is a full-duplex, synchronous, character-oriented channel that supports a four-wire interface.

#### Timers

Up to four 16-bit reloadable timers can be used for timing/counting events or for motor control operations. These timers provide a 16-bit programmable reload counter and operate in One-Shot, Continuous, Gated, Capture, Compare, Capture and Compare, and PWM modes. Only 3 timers (Timers 0-2) are available in the 44-pin packages.

## **Interrupt Controller**

The 64K Series products support up to 24 interrupts. These interrupts consist of 12 internal and 12 GPIO pins. The interrupts have 3 levels of programmable interrupt priority.

#### **Reset Controller**

The Z8 Encore! can be reset using the RESET pin, Power-On Reset, Watchdog Timer, STOP mode exit, or Voltage Brownout (VBO) warning signal.

## On-Chip Debugger

The Z8 Encore! XP features an integrated On-Chip Debugger. The OCD provides a rich set of debugging capabilities, such as reading and writing registers, programming the Flash, setting breakpoints and executing code. A single-pin interface provides communication to the OCD.

#### **DMA Controller**

The 64K Series features three channels of DMA. Two of the channels are for register RAM to and from I/O operations. The third channel automatically controls the transfer of data from the ADC to the memory.

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# **Signal and Pin Descriptions**

## **Overview**

The Z8 Encore! XP 64K Series Flash Microcontrollers product are available in a variety of packages styles and pin configurations. This chapter describes the signals and available pin configurations for each of the package styles. For information on physical package specifications, see Packaging on page 265.

## **Available Packages**

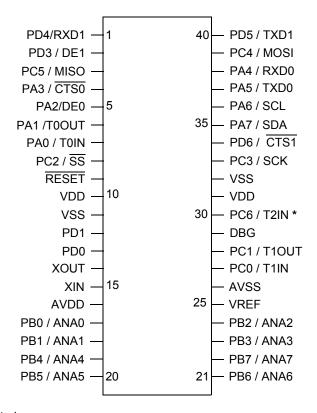
Table 2 identifies the package styles that are available for each device within the Z8 Encore! XP 64K Series Flash Microcontrollers product line.

Table 2. Z8 Encore! XP 64K Series Flash Microcontrollers Package Options

Part Number	40-Pin PDIP	44-pin LQFP	44-pin PLCC	64-pin LQFP	68-pin PLCC	80-pin QFP
Z8F1621	Х	Х	Х			
Z8F1622				Χ	Χ	
Z8F2421	X	Х	X			
Z8F2422				Х	Χ	
Z8F3221	Χ	Х	Х			
Z8F3222				Х	Х	
Z8F4821	Χ	Х	X			
Z8F4822				Х	Х	
Z8F4823						Х
Z8F6421	Χ	Χ	Χ			
Z8F6422				Х	Χ	
Z8F6423						X

## **Pin Configurations**

Figure 2 through Figure 7 on page 13 display the pin configurations for all of the packages available in the Z8 Encore! XP 64K Series Flash Microcontrollers. For description of the signals, see Table 3 on page 14. Timer 3 is not available in the 40-pin and 44-pin packages.



Note: Timer 3 is not supported.

Figure 2. Z8 Encore! XP 64K Series Flash Microcontrollers in 40-Pin Dual Inline Package (PDIP)

<sup>\*</sup> T2OUT is not supported.

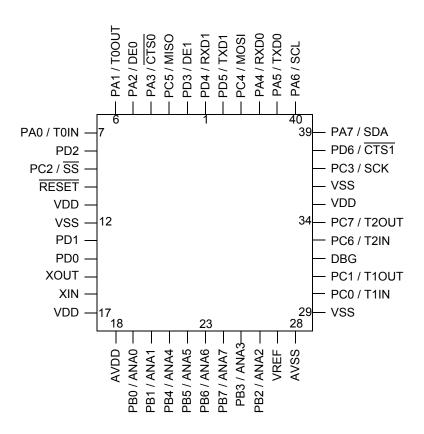


Figure 3. Z8 Encore! XP 64K Series Flash Microcontrollers in 44-Pin Plastic Leaded Chip Carrier (PLCC)

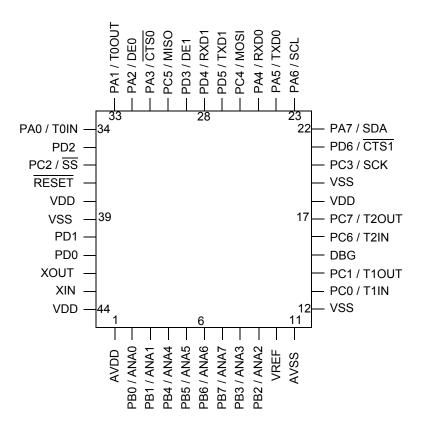


Figure 4. Z8 Encore! XP 64K Series Flash Microcontrollers in 44-Pin Low-Profile Quad Flat Package (LQFP)

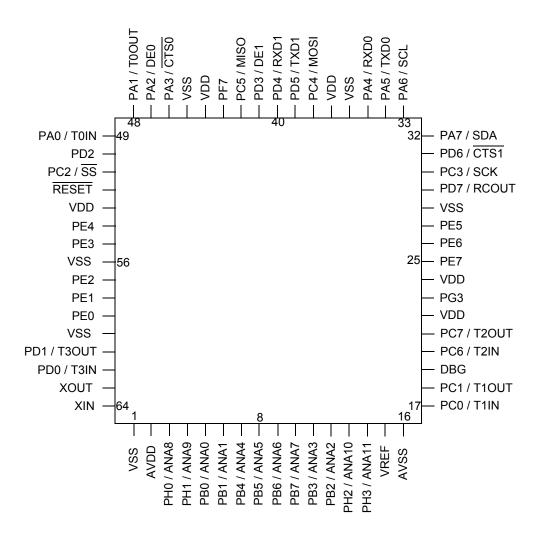


Figure 5. Z8 Encore! XP 64K Series Flash Microcontrollers in 64-Pin Low-Profile Quad Flat Package (LQFP)

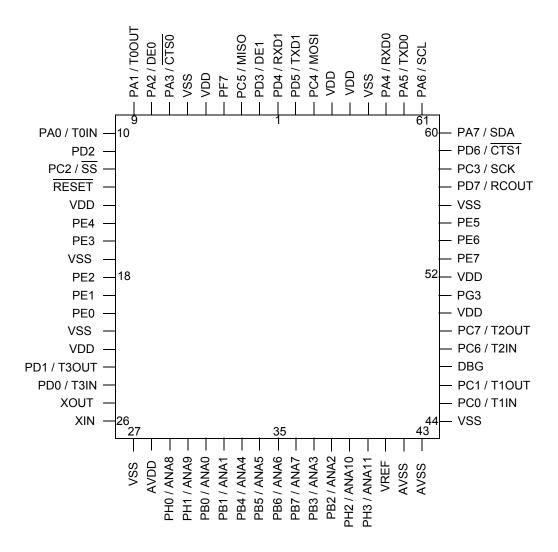


Figure 6. Z8 Encore! XP 64K Series Flash Microcontrollers in 68-Pin Plastic Leaded Chip Carrier (PLCC)

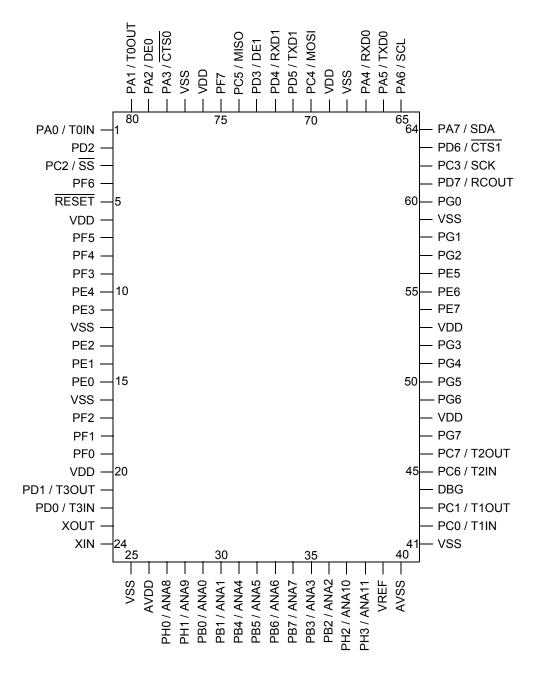


Figure 7. Z8 Encore! XP 64K Series Flash Microcontrollers in 80-Pin Quad Flat Package (QFP)

## **Signal Descriptions**

Table 3 describes the Z8 Encore! XP signals. To determine the signals available for the specific package styles, see Pin Configurations on page 8.

**Table 3. Signal Descriptions** 

Signal Mnemonic	I/O	Description	
General-Purp	ose I/O Poi	rts A-H	
PA[7:0]	I/O	Port A[7:0]. These pins are used for general-purpose I/O and support 5 V-tolerant inputs.	
PB[7:0]	I/O	Port B[7:0]. These pins are used for general-purpose I/O.	
PC[7:0]	I/O	Port C[7:0]. These pins are used for general-purpose I/O. These pins are used for general-purpose I/O and support 5 V-tolerant inputs	
PD[7:0]	I/O	Port D[7:0]. These pins are used for general-purpose I/O. These pins are used for general-purpose I/O and support 5 V-tolerant inputs	
PE[7:0]	I/O	Port E[7:0]. These pins are used for general-purpose I/O. These pins are used for general-purpose I/O and support 5 V-tolerant inputs.	
PF[7:0]	I/O	Port F[7:0]. These pins are used for general-purpose I/O. These pins are used for general-purpose I/O and support 5 V-tolerant inputs.	
PG[7:0]	I/O	Port G[7:0]. These pins are used for general-purpose I/O. These pins are used for general-purpose I/O and support 5 V-tolerant inputs.	
PH[3:0]	I/O	Port H[3:0]. These pins are used for general-purpose I/O.	
I <sup>2</sup> C Controller	•		
SCL	0	Serial Clock. This is the output clock for the $I^2$ C. This pin is multiplexed with a general-purpose I/O pin. When the general-purpose I/O pin is configured for alternate function to enable the SCL function, this pin is open-drain.	
SDA	I/O	Serial Data. This open-drain pin transfers data between the I <sup>2</sup> C and a slave. This pin is multiplexed with a general-purpose I/O pin. When the general-purpose I/O pin is configured for alternate function to enable the SDA function, this pin is open-drain.	
SPI Controlle	r		
SS	I/O	Slave Select. This signal can be an output or an input. If the Z8 Encore! XP 64K Series Flash Microcontrollers is the SPI master, this pin may be configured as the Slave Select output. If the Z8 Encore! XP 64K Series Flash Microcontrollers is the SPI slave, this pin is the input slave select. It is multiplexed with a general-purpose I/O pin.	

**Table 3. Signal Descriptions (Continued)** 

Signal Mnemonic	I/O	Description	
SCK	I/O	SPI Serial Clock. The SPI master supplies this pin. If the Z8 Encore! XP 64K Series Flash Microcontrollers is the SPI master, this pin is an output. If the Z8 Encore! XP 64K Series Flash Microcontrollers is the SPI slave, this pin is an input. It is multiplexed with a general-purpose I/O pin.	
MOSI	I/O	Master-Out/Slave-In. This signal is the data output from the SPI master device and the data input to the SPI slave device. It is multiplexed with a general-purpose I/O pin.	
MISO	I/O	Master-In/Slave-Out. This pin is the data input to the SPI master device and the data output from the SPI slave device. It is multiplexed with a general-purpose I/O pin.	
<b>UART Controll</b>	ers		
TXD0 / TXD1	О	Transmit Data. These signals are the transmit outputs from the UARTs. The TXD signals are multiplexed with general-purpose I/O pins.	
RXD0 / RXD1	I	Receive Data. These signals are the receiver inputs for the UARTs and IrDAs. The RXD signals are multiplexed with general-purpose I/O pins.	
CTS0 / CTS1	I	Clear To Send. These signals are control inputs for the UARTs. The CTS signals are multiplexed with general-purpose I/O pins.	
DE0 / DE1	0	Driver Enable. This signal allows automatic control of external RS-485 drivers. This signal is approximately the inverse of the Transmit Empty (TXE) bit in the UART Status 0 register. The DE signal may be used to ensure an external RS-485 driver is enabled when data is transmitted by the UART.	
Timers			
T0OUT/T1OUT/ T2OUT/T3OUT	<b>О</b>	Timer Output 0-3. These signals are output pins from the timers. The Timer Output signals are multiplexed with general-purpose I/O pins. T3OUT is not available in 44-pin package devices.	
T0IN/T1IN/ T2IN/T3IN	I	Timer Input 0-3. These signals are used as the capture, gating and counter inputs. The Timer Input signals are multiplexed with general-purpose I/O pins. T3IN is not available in 44-pin package devices.	
Analog			
ANA[11:0]	I	Analog Input. These signals are inputs to the ADC. The ADC analog inputs are multiplexed with general-purpose I/O pins.	
VREF	I	Analog-to-Digital converter reference voltage input. The VREF pin must be left unconnected (or capacitively coupled to analog ground) if the internal voltage reference is selected as the ADC reference voltage.	
Oscillators			

**Table 3. Signal Descriptions (Continued)** 

Signal Mnemon	ic	I/O	Description	
XIN		I	External Crystal Input. This is the input pin to the crystal oscillator. A crystal can be connected between it and the <b>XOUT</b> pin to form the oscillator. This signal is usable with external RC networks and an external clock driver.	
XOUT		О	External Crystal Output. This pin is the output of the crystal oscillator. A crystal can be connected between it and the XIN pin to form the oscillator. When the system clock is referred to in this manual, it refers to the frequency of the signal at this pin. This pin must be left unconnected when not using a crystal.	
RCOUT		0	RC Oscillator Output. This signal is the output of the RC oscillator. It is multiplexed with a general-purpose I/O pin. This signal must be left unconnected when not using a crystal.	
On-Chip	Debug	ger		
DBG		I/O	Debug. This pin is the control and data input and output to and from the On-Chip Debugger. This pin is open-drain.	
	<u>^</u>	Caution:	For operation of the On-Chip Debugger, all power pins ( $V_{DD}$ and $AV_{DD}$ ) must be supplied with power and all ground pins ( $V_{SS}$ and $AV_{SS}$ ) must be properly grounded.	
			The DBG pin is open-drain and must have an external pull-up resistor to ensure proper operation.	
Reset				
RESET		I	RESET. Generates a Reset when asserted (driven Low).	
Power Su	upply			
VDD		I	Power Supply.	
AVDD		I	Analog Power Supply.	
VSS		Ι	Ground.	
AVSS		I	Analog Ground.	

## **Pin Characteristics**

Table 4 on page 17 provides detailed information on the characteristics for each pin available on the 64K Series products and the data is sorted alphabetically by the pin symbol mnemonic.

Table 4. Pin Characteristics of the Z8 Encore! XP 64K Series Flash Microcontrollers

Symbol Mnemonic	Direction	Reset Direction	Active Low or Active High	Tri-State Output	Internal Pull-up or Pull-down	Schmitt- Trigger Input	Open Drain Output
AVSS	N/A	N/A	N/A	N/A	No	No	N/A
AVDD	N/A	N/A	N/A	N/A	No	No	N/A
DBG	I/O	I	N/A	Yes	No	Yes	Yes
VSS	N/A	N/A	N/A	N/A	No	No	N/A
PA[7:0]	I/O	I	N/A	Yes	No	Yes	Yes, Programmable
PB[7:0]	I/O	I	N/A	Yes	No	Yes	Yes, Programmable
PC[7:0]	I/O	I	N/A	Yes	No	Yes	Yes, Programmable
PD[7:0]	I/O	I	N/A	Yes	No	Yes	Yes, Programmable
PE7:0]	I/O	I	N/A	Yes	No	Yes	Yes, Programmable
PF[7:0]	I/O	I	N/A	Yes	No	Yes	Yes, Programmable
PG[7:0]	I/O	I	N/A	Yes	No	Yes	Yes, Programmable
PH[3:0]	I/O	I	N/A	Yes	No	Yes	Yes, Programmable
RESET	I	I	Low	N/A	Pull-up	Yes	N/A
VDD	N/A	N/A	N/A	N/A	No	No	N/A
XIN	I	I	N/A	N/A	No	No	N/A
XOUT	0	0	N/A	Yes, in STOP mode	No	No	No

**Note:** *x* represents integer 0, 1,... to indicate multiple pins with symbol mnemonics that differ only by the integer.

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# **Address Space**

#### Overview

The eZ8<sup>TM</sup> CPU can access three distinct address spaces:

- The Register File contains addresses for the general-purpose registers and the eZ8 CPU, peripheral, and general-purpose I/O port control registers.
- The Program Memory contains addresses for all memory locations having executable code and/or data.
- The Data Memory consists of the addresses for all memory locations that hold only

These three address spaces are covered briefly in the following subsections. For more information on eZ8 CPU and its address space, refer to  $eZ8^{TM}$  CPU Core User Manual (UM0128) available for download at <a href="https://www.zilog.com"><u>www.zilog.com</u></a>.

## Register File

The Register File address space in the 64K Series is 4 KB (4096 bytes). The Register File is composed of two sections—control registers and general-purpose registers. When instructions are executed, registers are read from when defined as sources and written to when defined as destinations. The architecture of the eZ8 CPU allows all general-purpose registers to function as accumulators, address pointers, index registers, stack areas, or scratch pad memory.

The upper 256 bytes of the 4 KB Register File address space are reserved for control of the eZ8 CPU, the on-chip peripherals, and the I/O ports. These registers are located at addresses from F00H to FFFH. Some of the addresses within the 256-byte control register section are reserved (unavailable). Reading from an reserved Register File addresses returns an undefined value. Writing to reserved Register File addresses is not recommended and can produce unpredictable results.

The on-chip RAM always begins at address 000H in the Register File address space. The 64K Series provide 2 KB to 4 KB of on-chip RAM depending upon the device. Reading from Register File addresses outside the available RAM addresses (and not within the control register address space) returns an undefined value. Writing to these Register File addresses produces no effect. To determine the amount of RAM available for the specific 64K Series device, see Part Selection Guide on page 2.

## **Program Memory**

The eZ8<sup>™</sup> CPU supports 64 KB of Program Memory address space. The Z8 Encore! XP 64K Series Flash Microcontrollers contains 16 KB to 64 KB of on-chip Flash in the Program Memory address space, depending upon the device. Reading from Program Memory addresses outside the available Flash memory addresses returns FFH. Writing to these unimplemented Program Memory addresses produces no effect. Table 5 describes the Program Memory maps for the 64K Series products.

Table 5. Z8 Encore! XP 64K Series Flash Microcontrollers Program Memory Maps

Program Memory Address (Hex)	Function
Z8F162x Products	
0000-0001	Option Bits
0002-0003	Reset Vector
0004-0005	WDT Interrupt Vector
0006-0007	Illegal Instruction Trap
0008-0037	Interrupt Vectors*
0038-3FFF	Program Memory
Z8F242x Products	
0000-0001	Option Bits
0002-0003	Reset Vector
0004-0005	WDT Interrupt Vector
0006-0007	Illegal Instruction Trap
0008-0037	Interrupt Vectors*
0038-5FFF	Program Memory
Z8F322x Products	
0000-0001	Option Bits
0002-0003	Reset Vector
0004-0005	WDT Interrupt Vector
0006-0007	Illegal Instruction Trap
0008-0037	Interrupt Vectors*
0038-7FFF	Program Memory
Z8F482x Products	

Table 5. Z8 Encore! XP 64K Series Flash Microcontrollers Program Memory Maps (Continued)

Program Memory Address (Hex)	Function		
0000-0001	Option Bits		
0002-0003	Reset Vector		
0004-0005	WDT Interrupt Vector		
0006-0007	Illegal Instruction Trap		
0008-0037	Interrupt Vectors*		
0038-BFFF	Program Memory		
Z8F642x Products			
0000-0001	Option Bits		
0002-0003	Reset Vector		
0004-0005	WDT Interrupt Vector		
0006-0007	Illegal Instruction Trap		
0008-0037	Interrupt Vectors*		
0038-FFFF	Program Memory		
*See Table 23 on page 68 for a list of the interrupt vectors.			

## **Data Memory**

The Z8 Encore! XP 64K Series Flash Microcontrollers does not use the eZ8 CPU's 64 KB Data Memory address space.

## **Information Area**

Table 6 on page 22 describes the Z8 Encore! XP 64K Series Flash Microcontrollers Information Area. This 512 byte Information Area is accessed by setting bit 7 of the Page Select Register to 1. When access is enabled, the Information Area is mapped into the Program Memory and overlays the 512 bytes at addresses FE00H to FFFFH. When the Information Area access is enabled, execution of LDC and LDCI instruction from these Program Memory addresses return the Information Area data rather than the Program Memory data. Reads of these addresses through the On-Chip Debugger also returns the Information Area data. Execution of code from these addresses continues to correctly use the Program Memory. Access to the Information Area is read-only.

Table 6. Z8 Encore! XP 64K Series Flash Microcontrollers Information Area Map

Program Memory	
Address (Hex)	Function
FE00H-FE3FH	Reserved
FE40H-FE53H	Part Number 20-character ASCII alphanumeric code Left justified and filled with zeros (ASCII Null character)
FE54H-FFFFH	Reserved

# Register File Address Map

Table 7 provides the address map for the Register File of the 64K Series products. Not all devices and package styles in the 64K Series support Timer 3 and all of the GPIO Ports. Consider registers for unimplemented peripherals as Reserved.

Table 7. Z8 Encore! XP 64K Series Flash Microcontrollers Register File Address Map

Address (Hex)	Register Description	Mnemonic	Reset (Hex)	Page No
General-Purpo	se RAM			
000-EFF	General-Purpose Register File RAM	_	XX	
Timer 0				
F00	Timer 0 High Byte	T0H	00	90
F01	Timer 0 Low Byte	T0L	01	90
F02	Timer 0 Reload High Byte	T0RH	FF	91
F03	Timer 0 Reload Low Byte	T0RL	FF	91
F04	Timer 0 PWM High Byte	T0PWMH	00	92
F05	Timer 0 PWM Low Byte	T0PWML	00	92
F06	Timer 0 Control 0	T0CTL0	00	93
F07	Timer 0 Control 1	T0CTL1	00	94
Timer 1				
F08	Timer 1 High Byte	T1H	00	90
F09	Timer 1 Low Byte	T1L	01	90
F0A	Timer 1 Reload High Byte	T1RH	FF	91
F0B	Timer 1 Reload Low Byte	T1RL	FF	91
F0C	Timer 1 PWM High Byte	T1PWMH	00	92
F0D	Timer 1 PWM Low Byte	T1PWML	00	92
F0E	Timer 1 Control 0	T1CTL0	00	93
F0F	Timer 1 Control 1	T1CTL1	00	94
Timer 2				
F10	Timer 2 High Byte	T2H	00	90
F11	Timer 2 Low Byte	T2L	01	90
F12	Timer 2 Reload High Byte	T2RH	FF	91
F13	Timer 2 Reload Low Byte	T2RL	FF	91
F14	Timer 2 PWM High Byte	T2PWMH	00	92
F15	Timer 2 PWM Low Byte	T2PWML	00	92
F16	Timer 2 Control 0	T2CTL0	00	93
F17	Timer 2 Control 1	T2CTL1	00	94

Table 7. Z8 Encore! XP 64K Series Flash Microcontrollers Register File Address Map (Continued)

Address (Hex)	Register Description	Mnemonic	Reset (Hex)	Page No
	nilable in the 44-pin packages)			
F18	Timer 3 High Byte	T3H	00	90
F19	Timer 3 Low Byte	T3L	01	90
F1A	Timer 3 Reload High Byte	T3RH	FF	91
F1B	Timer 3 Reload Low Byte	T3RL	FF	91
F1C	Timer 3 PWM High Byte	T3PWMH	00	92
F1D	Timer 3 PWM Low Byte	T3PWML	00	92
F1E	Timer 3 Control 0	T3CTL0	00	93
F1F	Timer 3 Control 1	T3CTL1	00	94
20-3F	Reserved	_	XX	
UART 0				
F40	UART0 Transmit Data	U0TXD	XX	114
	UART0 Receive Data	U0RXD	XX	115
F41	UART0 Status 0	U0STAT0	0000011Xb	115
F42	UART0 Control 0	U0CTL0	00	117
F43	UART0 Control 1	U0CTL1	00	117
F44	UART0 Status 1	U0STAT1	00	115
F45	UART0 Address Compare Register	U0ADDR	00	120
F46	UART0 Baud Rate High Byte	U0BRH	FF	120
F47	UART0 Baud Rate Low Byte	U0BRL	FF	120
UART 1				
F48	UART1 Transmit Data	U1TXD	XX	114
	UART1 Receive Data	U1RXD	XX	115
F49	UART1 Status 0	U1STAT0	0000011Xb	115
F4A	UART1 Control 0	U1CTL0	00	117
F4B	UART1 Control 1	U1CTL1	00	117
F4C	UART1 Status 1	U1STAT1	00	115
F4D	UART1 Address Compare Register	U1ADDR	00	120
F4E	UART1 Baud Rate High Byte	U1BRH	FF	120
F4F	UART1 Baud Rate Low Byte	U1BRL	FF	120
I <sup>2</sup> C				
F50	I <sup>2</sup> C Data	I2CDATA	00	156
F51	I <sup>2</sup> C Status	I2CSTAT	80	157
F52	I <sup>2</sup> C Control	I2CCTL	00	158
F53	I <sup>2</sup> C Baud Rate High Byte	I2CBRH	FF	160
F54	I <sup>2</sup> C Baud Rate Low Byte	I2CBRL	FF	160
F55	I <sup>2</sup> C Diagnostic State	I2CDST	C0	161
F56	I <sup>2</sup> C Diagnostic Control	I2CDIAG	00	163
F57-F5F	Reserved	_	XX	
Serial Peripher	ral Interface (SPI)			
F60	SPI Data	SPIDATA	XX	137

Table 7. Z8 Encore! XP 64K Series Flash Microcontrollers Register File Address Map (Continued)

	Register Description	Mnemonic	Reset (Hex)	Page No
F61	SPI Control	SPICTL	00	137
F62	SPI Status	SPISTAT	01	139
F63	SPI Mode	SPIMODE	00	140
F64	SPI Diagnostic State	SPIDST	00	141
F65	Reserved	_	XX	
F66	SPI Baud Rate High Byte	SPIBRH	FF	142
F67	SPI Baud Rate Low Byte	SPIBRL	FF	142
F68-F6F	Reserved	_	XX	
Analog-to-Digit	tal Converter			
F70	ADC Control	ADCCTL	20	179
F71	Reserved	_	XX	
F72	ADC Data High Byte	ADCD_H	XX	180
F73	ADC Data Low Bits	ADCD_L	XX	180
F74-FAF	Reserved		XX	
DMA 0				
FB0	DMA0 Control	DMA0CTL	00	167
FB1	DMA0 I/O Address	DMA0IO	XX	169
FB2	DMA0 End/Start Address High Nibble	DMA0H	XX	169
FB3	DMA0 Start Address Low Byte	DMA0START	XX	170
FB4	DMA0 End Address Low Byte	DMA0END	XX	170
DMA 1	•			
FB8	DMA1 Control	DMA1CTL	00	167
FB9	DMA1 I/O Address	DMA1IO	XX	169
FBA	DMA1 End/Start Address High Nibble	DMA1H	XX	169
FBB	DMA1 Start Address Low Byte	DMA1START	XX	170
FBC	DMA1 End Address Low Byte	DMA1END	XX	170
DMA ADC				
FBD	DMA ADC Address	DMAA ADDR	XX	171
FBE	DMA ADC Control	DMAACTL	00	172
FBF	DMA ADC Status	DMAASTAT	00	173
Interrupt Contr	<b>_</b>			
FC0	Interrupt Request 0	IRQ0	00	71
FC1	IRQ0 Enable High Bit	IRQ0ENH	00	74
FC2	IRQ0 Enable Low Bit	IRQ0ENL	00	74
FC3	Interrupt Request 1	IRQ1	00	72
FC4	IRQ1 Enable High Bit	IRQ1ENH	00	75
FC5	IRQ1 Enable Low Bit	IRQ1ENL	00	75
FC6	Interrupt Request 2	IRQ2	00	73
FC7	IRQ2 Enable High Bit	IRQ2ENH	00	76
FC8	IRQ2 Enable Low Bit	IRQ2ENL	00	76
-	· · · · · · · · · · · · · · · · · · ·	-,	-	-

Table 7. Z8 Encore! XP 64K Series Flash Microcontrollers Register File Address Map (Continued)

Address (Hex)	<u> </u>	Mnemonic	Reset (Hex)	Page No
FCD	Interrupt Edge Select	IRQES	00	78
FCE	Interrupt Port Select	IRQPS	00	78
FCF	Interrupt Control	IRQCTL	00	79
GPIO Port A				
FD0	Port A Address	PAADDR	00	61
FD1	Port A Control	PACTL	00	62
FD2	Port A Input Data	PAIN	XX	66
FD3	Port A Output Data	PAOUT	00	66
GPIO Port B				
FD4	Port B Address	PBADDR	00	61
FD5	Port B Control	PBCTL	00	62
FD6	Port B Input Data	PBIN	XX	66
FD7	Port B Output Data	PBOUT	00	66
GPIO Port C				
FD8	Port C Address	PCADDR	00	61
FD9	Port C Control	PCCTL	00	62
FDA	Port C Input Data	PCIN	XX	66
FDB	Port C Output Data	PCOUT	00	66
GPIO Port D				
FDC	Port D Address	PDADDR	00	61
FDD	Port D Control	PDCTL	00	62
FDE	Port D Input Data	PDIN	XX	66
FDF	Port D Output Data	PDOUT	00	66
GPIO Port E				
FE0	Port E Address	PEADDR	00	61
FE1	Port E Control	PECTL	00	62
FE2	Port E Input Data	PEIN	XX	66
FE3	Port E Output Data	PEOUT	00	66
GPIO Port F				
FE4	Port F Address	PFADDR	00	61
FE5	Port F Control	PFCTL	00	62
FE6	Port F Input Data	PFIN	XX	66
FE7	Port F Output Data	PFOUT	00	66
GPIO Port G	-			
FE8	Port G Address	PGADDR	00	61
FE9	Port G Control	PGCTL	00	62
FEA	Port G Input Data	PGIN	XX	66
FEB	Port G Output Data	PGOUT	00	66
GPIO Port H	·			
FEC	Port H Address	PHADDR	00	61
FED	Port H Control	PHCTL	00	62
	Port H Input Data	PHIN	XX	66

Table 7. Z8 Encore! XP 64K Series Flash Microcontrollers Register File Address Map (Continued)

Address (Hex)	Register Description	Mnemonic	Reset (Hex)	Page No
FEF	Port H Output Data	PHOUT	00	66
Watchdog Time	er			
FF0	Watchdog Timer Control	WDTCTL	XXX00000b	100
FF1	Watchdog Timer Reload Upper Byte	WDTU	FF	101
FF2	Watchdog Timer Reload High Byte	WDTH	FF	101
FF3	Watchdog Timer Reload Low Byte	WDTL	FF	101
FF4-FF7	Reserved	_	XX	
Flash Memory	Controller			
FF8	Flash Control	FCTL	00	190
FF8	Flash Status	FSTAT	00	190
FF9	Page Select	FPS	00	191
FF9 (if enabled)	Flash Sector Protect	FPROT	00	192
FFA	Flash Programming Frequency High Byte	FFREQH	00	192
FFB	Flash Programming Frequency Low Byte	FFREQL	00	192
FF4-FF8	Reserved	_	XX	
Read-Only Mer	nory Controller			
FF9	Page Select	RPS	00	
FFA-FFB	Reserved	_	XX	
eZ8 CPU				
FFC	Flags	_	XX	Refer to eZ8 <sup>TM</sup>
FFD	Register Pointer	RP	XX	CPU Core
FFE	Stack Pointer High Byte	SPH	XX	User Manual
FFF	Stack Pointer Low Byte	SPL	XX	(UM0128)
Note: XX=Undefin	ned			

# Control Register Summary

**Timer 0 High Byte** T0H (F00H - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0 Timer 0 current count value [15:8] **Timer 0 Low Byte** T0L (F01H - Read/Write) D7|D6|D5|D4|D3|D2|D1|D0| Timer 0 current count value [7:0] Timer 0 Reload High Byte T0RH (F02H - Read/Write) D7|D6|D5|D4|D3|D2|D1|D0| Timer 0 reload value [15:8] **Timer 0 Reload Low Byte** TORL (HF03 - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0 Timer 0 reload value [7:0] **Timer 0 PWM High Byte** T0PWMH (F04H - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0 Timer 0 PWM value [15:8] Timer 0 Control 0 T0CTL0 (F06H - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0 Reserved Cascade Timer 0 = Timer 0 Input signal is GPIO pin 1 = Timer 0 Input signal is Timer 3 Reserved

Timer 0 Control 1
T0CTL1 (F07H - Read/Write)

| D7|D6|D5|D4|D3|D2|D1|D0|
| Timer Mode

Timer Mode 000 = One-Shot mode 001 = CONTINUOUS mode 010 = COUNTER mode 011 = PWM mode 100 = CAPTURE mode 101 = COMPARE mode 110 = GATED mode 111 = Capture/COMPARE mode Prescale Value 000 = Divide by 1 001 = Divide by 2 010 = Divide by 4 011 = Divide by 8 100 = Divide by 0 100 = Divide by 16 101 = Divide by 32 110 = Divide by 64 111 = Divide by 128 Timer Input/Output Polarity Operation of this bit is a function of the current operating mode of the timer Timer Enable 0 = Timer is disabled 1 = Timer is enabled

Timer 1 High Byte T1H (F08H - Read/Write) D7|D6|D5|D4|D3|D2|D1|D0

— Timer 1 current count value [15:8]

Timer 1 Low Byte
T1L (F09H - Read/Write)

D7|D6|D5|D4|D3|D2|D1|D0

Timer 1 current count value [7:0]

Timer 1 PWM High Byte T1PWMH (F0CH - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0	Timer 2 High Byte T2H (F10H - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0
Timer 1 PWM value [15:8]	Timer 2 current count value [15:8]
Timer 1 PWM Low Byte T1PWML (F0DH - Read/Write)	Timer 2 Low Byte T2L (F11H - Read/Write)
D7 D6 D5 D4 D3 D2 D1 D0	D7 D6 D5 D4 D3 D2 D1 D0
Timer 1 PWM value [7:0]	Timer 2 current count value [7:0]
<b>Timer 1 Control 0</b> T1CTL0(F0EH - Read/Write)	<b>Timer 2 Reload High Byte</b> T2RH (F12H - Read/Write)
D7 D6 D5 D4 D3 D2 D1 D0	D7 D6 D5 D4 D3 D2 D1 D0
Reserved	Timer 2 reload value [15:8]
Cascade Timer 0 = Timer 1 Input signal is GPIO pin 1 = Timer 1 Input signal is Timer 0	
out	Timer 2 Reload Low Byte T2RL (F13H- Read/Write)
Reserved	D7 D6 D5 D4 D3 D2 D1 D0
	Timer 2 reload value [7:0]
Timer 1 Control 1	
T1CTL1 (F0FH - Read/Write)	Timer 2 PWM High Byte
D7 D6 D5 D4 D3 D2 D1 D0	T2PWMH (F14H - Read/Write)
Timer Mode	D7 D6 D5 D4 D3 D2 D1 D0
000 = One-Shot mode 001 = CONTINUOUS mode	Timer 2 PWM value [15:8]
010 = COUNTER mode 011 = PWM mode	
100 = CAPTURE mode	
101 = COMPARE mode 110 = GATED mode	Timer 2 PWM Low Byte
111 = Capture/COMPARE mode	T2PWML (F15H - Read/Write)
Prescale Value	D7 D6 D5 D4 D3 D2 D1 D0
000 = Divide by 1 001 = Divide by 2	Timer 2 PWM value [7:0]
010 = Divide by 4	
011 = Divide by 8 100 = Divide by 16	Times 2 Control 0
101 = Divide by 32	Timer 2 Control 0 T2CTL0 (F16H - Read/Write)
110 = Divide by 64 111 = Divide by 128	D7 D6 D5 D4 D3 D2 D1 D0
Timer Input/Output Polarity Operation of this bit is a function of	Reserved
the current operating mode of the	Cascade Timer
timer	0 = Timer 2 Input signal is GPIO pin 1 = Timer 2 Input signal is Timer 1
Timer Enable	out
0 = Timer is disabled 1 = Timer is enabled	Reserved

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## **Timer 2 Control 1** T2CTL1 (F17H - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0 Timer Mode 000 = One-Shot mode 001 = CONTINUOUS mode 010 = COUNTER mode 011 = PWM mode 100 = CAPTURE mode 101 = COMPARE mode 110 = GATED mode 111 = CAPTURE/COMPARE mode Prescale Value 000 = Divide by 1 001 = Divide by 2 010 = Divide by 4 011 = Divide by 8 100 = Divide by 16 101 = Divide by 32 110 = Divide by 64 111 = Divide by 128 Timer Input/Output Polarity Operation of this bit is a function of the current operating mode of the timer Timer Enable 0 = Timer is disabled 1 = Timer is enabled **Timer 3 High Byte** T3H (F18H - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0 Timer 3 current count value [15:8] **Timer 3 Low Byte** T3L (F19H - Read/Write) D7|D6|D5|D4|D3|D2|D1|D0| Timer 3 current count value [7:0] **Timer 3 Reload High Byte** T3RH (F1AH - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0 Timer 3 reload value [15:8]

**Timer 3 PWM High Byte** T3PWMH (F1CH - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0 - Timer 3 PWM value [15:8] **Timer 3 PWM Low Byte** T3PWML (F1DH - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0 Timer 3 PWM value [7:0] Timer 3 Control 0 T3CTL0 (F1EH - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0 Reserved Cascade Timer 0 = Timer 3 Input signal is GPIO pin 1 = Timer 3 Input signal is Timer 2 out Reserved

Timer 3 Control 1 T3CTL1 (F1FH - Read/Write)

D7 D6 D5 D4 D3 D2 D1 D0 Timer Mode 000 = One-Shot mode 001 = CONTINUOUS mode 010 = COUNTER mode 011 = PWM mode 100 = CAPTURE mode 101 = COMPARE mode 110 = GATED mode 111 = Capture/COMPARE mode Prescale Value 000 = Divide by 1001 = Divide by 2 010 = Divide by 4 011 = Divide by 8 100 = Divide by 16 101 = Divide by 32 110 = Divide by 64 111 = Divide by 128 Timer Input/Output Polarity Operation of this bit is a function of the current operating mode of the timer Timer Enable 0 = Timer is disabled 1 = Timer is enabled

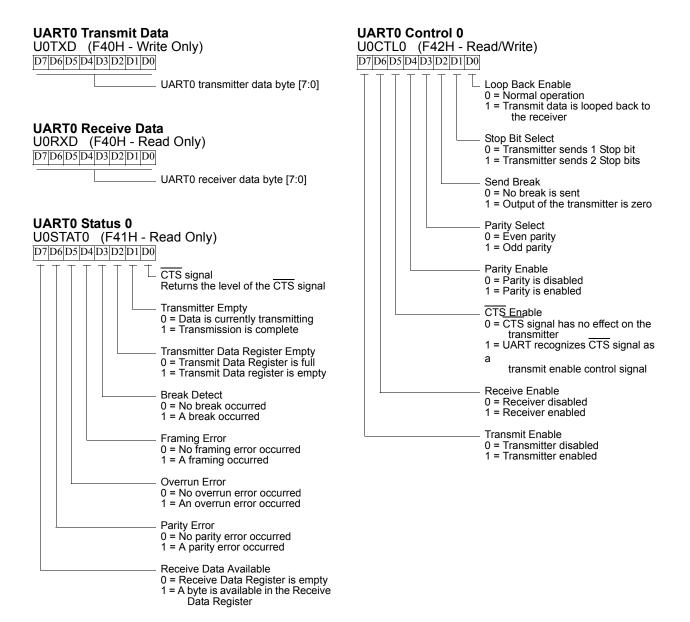
**Timer 3 Reload Low Byte** 

D7|D6|D5|D4|D3|D2|D1|D0

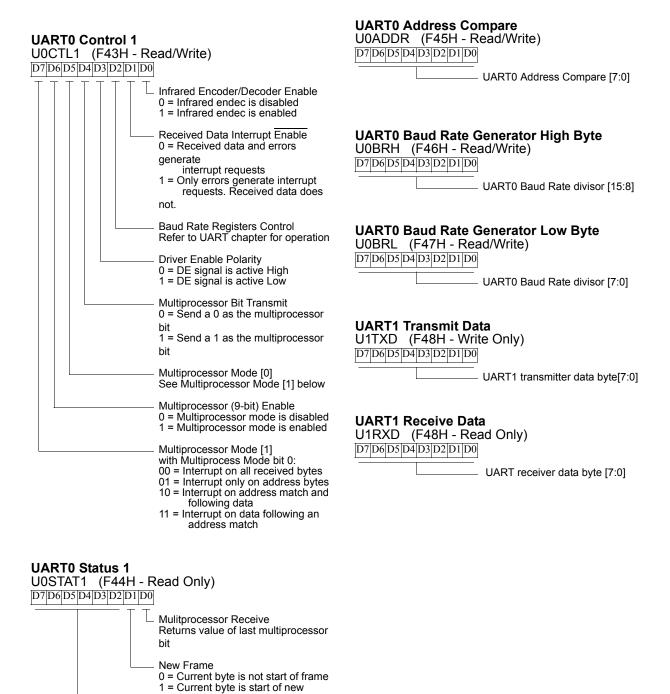
T3RL (F1BH - Read/Write)

Timer 3 reload value [7:0]

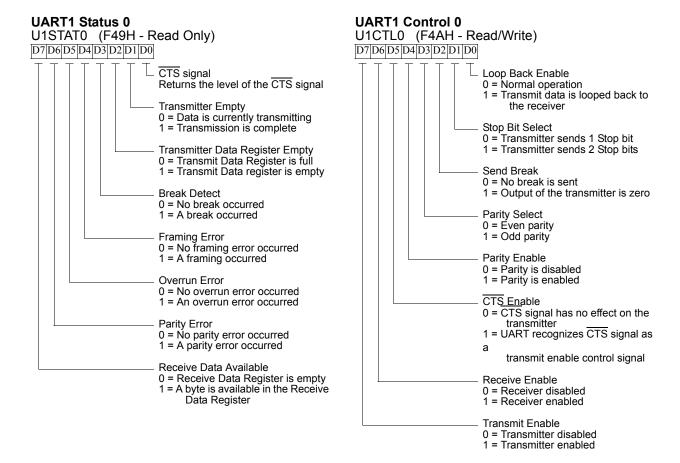


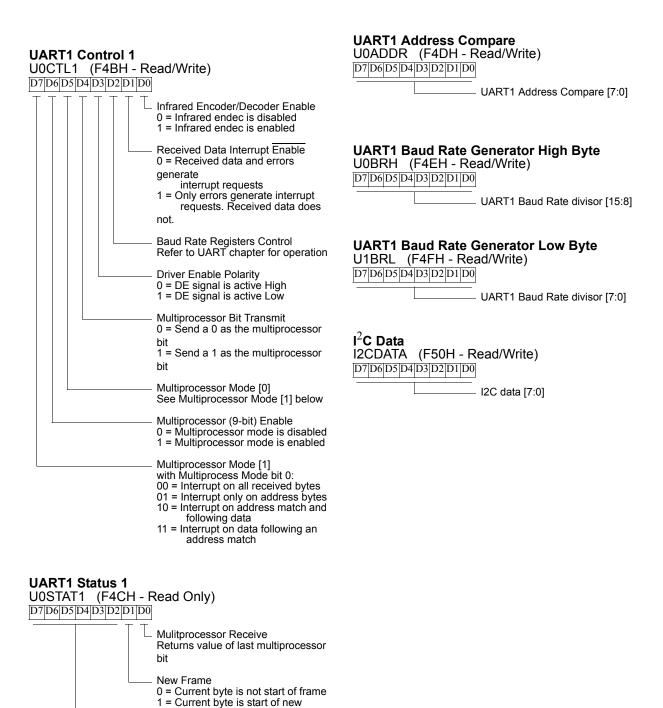






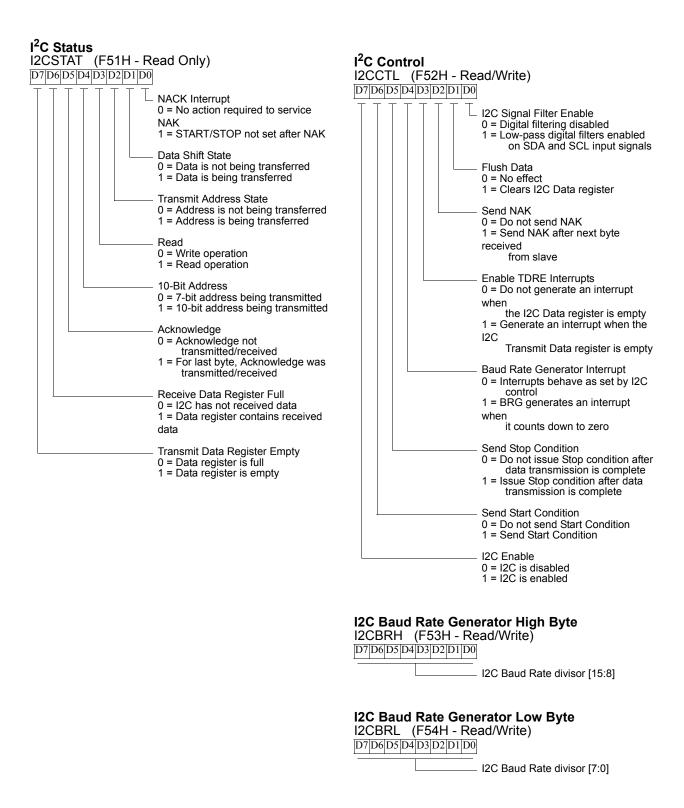
frame Reserved



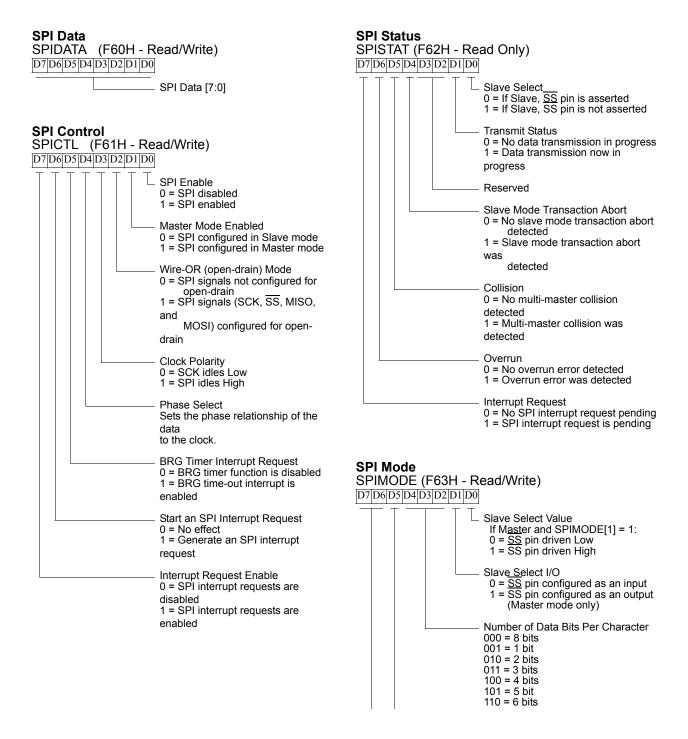


frame Reserved

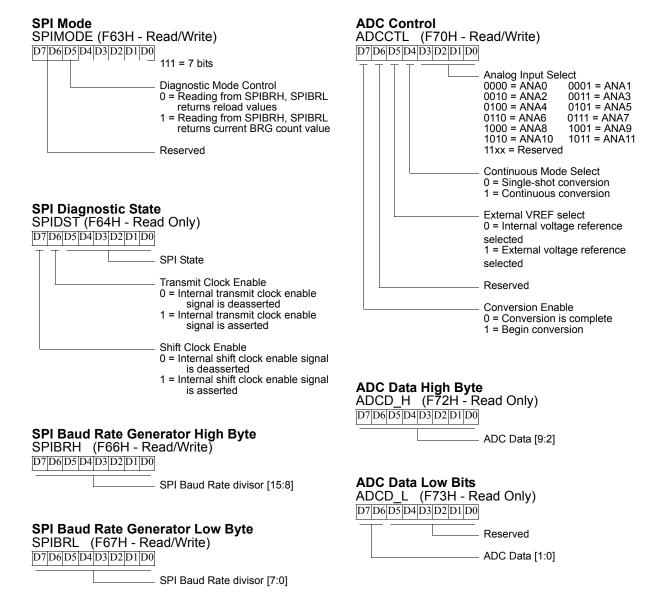




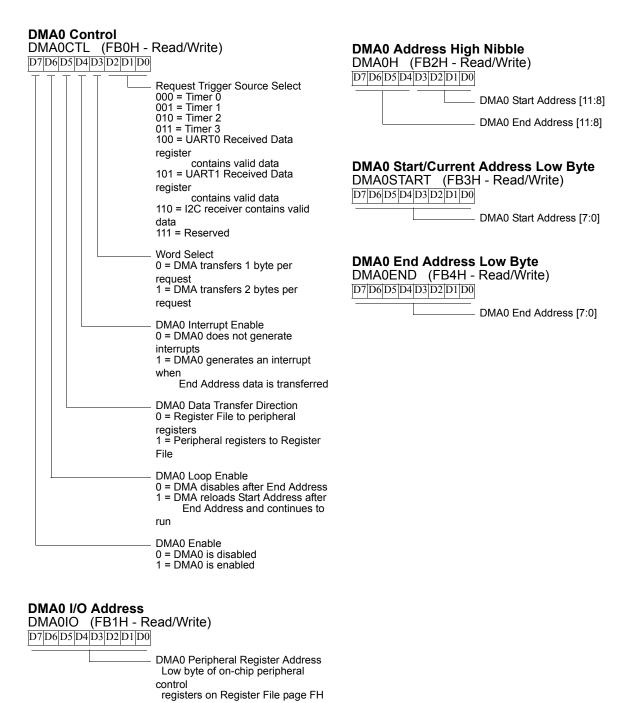




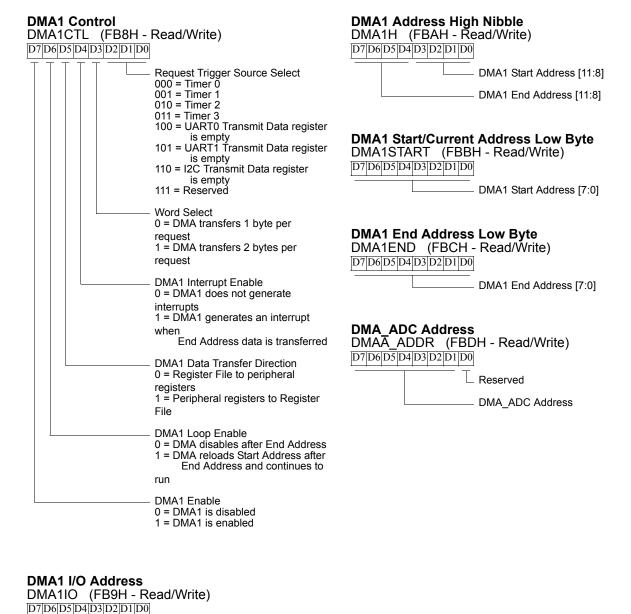












DMA1 Peripheral Register Address Low byte of on-chip peripheral control registers on Register File page FH

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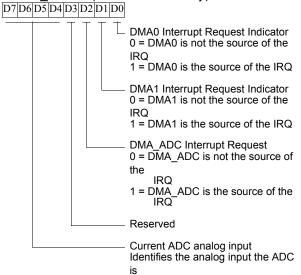
**DMA\_ADC Control**DMAACTL (FBEH - Read/Write)

D7 D6 D5 D4 D3 D2 D1 D0 **ADC Analog Input Number** 0000 = Analog input 0 updated 0001 = Analog input 0-1 updated 0010 = Analog input 0-2 updated 0011 = Analog input 0-3 updated 0100 = Analog input 0-4 updated 0101 = Analog input 0-5 updated 0100 = Analog input 0-6 updated 0101 = Analog input 0-7 updated 1000 = Analog input 0-8 updated 1001 = Analog input 0-9 updated 1010 = Analog input 0-10 updated 1011 = Analog inputs 0-11 updated 11xx = Reserved Reserved Interrupt request enable 0 = DMA\_ADC does not generate interrupt requests 1 = DMA\_ADC generates interrupt requests after last analog input

DMA ADC Enable 0 = DMA\_ADC is disabled 1 = DMA\_ADC is enabled

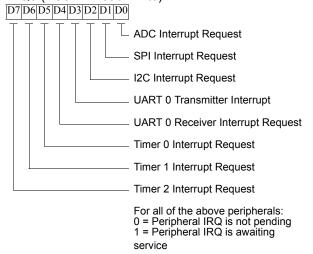
#### **DMA Status**

DMAA STAT (FBFH - Read Only)



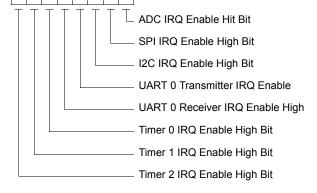
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#### **Interrupt Request 0** IRQ0 (FC0H - Read/Write)



#### IRQ0 Enable High Bit IRQ0ENH (FC1H - Read/Write)

D7 D6 D5 D4 D3 D2 D1 D0



**Interrupt Request 2 IRQ0 Enable Low Bit** IRQ2 (FC6H - Read/Write) IRQ0ENL (FC2H - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0 D7 D6 D5 D4 D3 D2 D1 D0 Port C Pin Interrupt Request ADC IRQ Enable Hit Bit 0 = IRQ from corresponding pin [3:0] is not pending SPI IRQ Enable Low Bit 1 = IRQ from corresponding pin [3:0] is awaiting service I2C IRQ Enable Low Bit **DMA Interrupt Request UART 0 Transmitter IRQ Enable UART 1 Transmitter Interrupt** UART 0 Receiver IRQ Enable Low **UART 1 Receiver Interrupt Request** Timer 0 IRQ Enable Low Bit Timer 3 Interrupt Request Timer 1 IRQ Enable Low Bit For all of the above peripherals: Timer 2 IRQ Enable Low Bit 0 = Peripheral IRQ is not pending 1 = Peripheral IRQ is awaiting service **Interrupt Request 1** IRQ1 (FC3H - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0 IRQ2 Enable High Bit IRQ2ENH (FC7H - Read/Write) Port A or D Pin Interrupt Request D7 D6 D5 D4 D3 D2 D1 D0 0 = IRQ from corresponding pin [7:0] is not pending 1 = IRQ from corresponding pin [7:0] Port C Pin IRQ Enable High Bit is awaiting service DMA IRQ Enable High Bit **UART 1 Transmitter IRQ Enable** IRQ1 Enable High Bit UART 1 Receiver IRQ Enable High IRQ1ENH (FC4H - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0 Timer 3 IRQ Enable High Bit Port A or D Pin IRQ Enable High Bit **IRQ2 Enable Low Bit** IRQ2ENL (FC8H - Read/Write) **IRQ1 Enable Low Bit** D7 D6 D5 D4 D3 D2 D1 D0 IRQ1ENL (FC5H - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0 Port C Pin IRQ Enable Low Bit Port A or D Pin IRQ Enable Low Bit DMA IRQ Enable Low Bit **UART 1 Transmitter IRQ Enable** UART 1 Receiver IRQ Enable Low Timer 3 IRQ Enable Low Bit

> Interrupt Edge Select IRQES (FCDH - Read/Write)

D7 D6 D5 D4 D3 D2 D1 D0

Port A or D Interrupt Edge Select

0 = Falling edge 1 = Rising edge

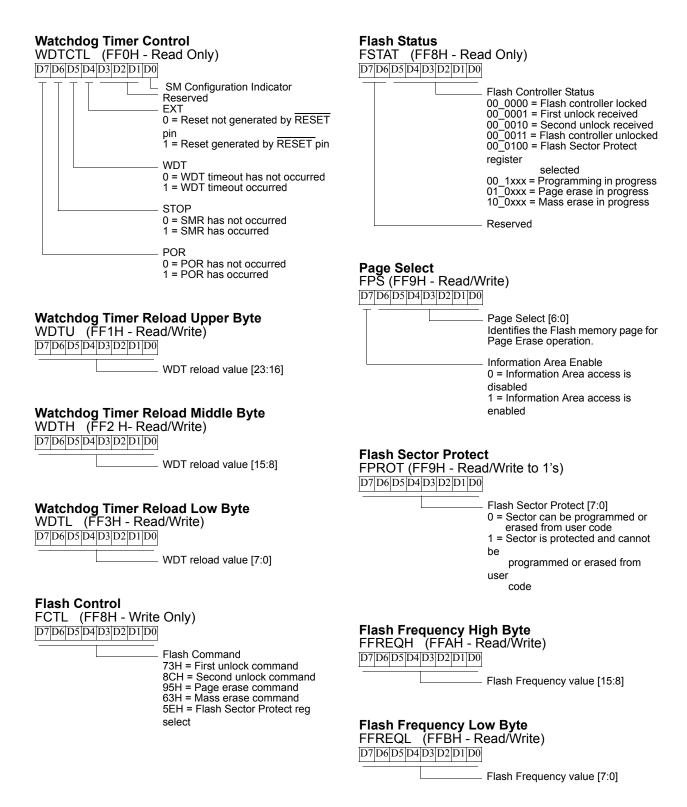
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interrupt Port Select IRQPS (FCEH - Read		PBADDR (FD4H - R	Pead/Mrite)
D7 D6 D5 D4 D3 D2 D1 D0	arvine)	D7 D6 D5 D4 D3 D2 D1 D0	(ead/write)
	Port A or D Port Pin Select [7:0] 0 = Port A pin is the interrupt source 1 = Port D pin is the interrupt source		Port B Address[7:0] Selects Port Sub-Registers: 00H = No function 01H = Data direction 02H = Alternate function 03H = Output control (open-drain) 04H = High drive enable 05H = Stop Mode Recovery enable 06H-FFH = No function
Port A Address	Reserved Interrupt Request Enable 0 = Interrupts are disabled 1 = Interrupts are enabled	Port B Control PBCTL (FD5H - Rea D7 D6 D5 D4 D3 D2 D1 D0	ad/Write)  Port B Control[7:0]  Provides Access to Port Sub- Registers
PAADDR (FD0H - R D7 D6 D5 D4 D3 D2 D1 D0	ead/white)		
Port A Control PACTL (FD1H - Rea	Port A Control[7:0]	Port B Input Data PBIN (FD6H - Read D7 D6 D5 D4 D3 D2 D1 D0  Port B Output Data PBOUT (FD7H - Re D7 D6 D5 D4 D3 D2 D1 D0  Port C Address	Port B Input Data [7:0]  ad/Write)  Port B Output Data [7:0]
Port A Input Data PAIN (FD2H - Read D7 D6 D5 D4 D3 D2 D1 D0	Provides Access to Port Sub-Registers  Only)  Port A Input Data [7:0]	PCADDR (FD8H - F D7 D6 D5 D4 D3 D2 D1 D0	Port C Address[7:0] Selects Port Sub-Registers: 00H = No function 01H = Data direction 02H = Alternate function 03H = Output control (open-drain) 04H = High drive enable 05H = Stop Mode Recovery enable
Port A Output Data PAOUT (FD3H - Rea D7 D6 D5 D4 D3 D2 D1 D0	ad/Write) Port A Output Data [7:0]		06H-FFH = No function

Port C Control PCCTL (FD9H - Read/Write)  D7 D6 D5 D4 D3 D2 D1 D0 Port C Control[7:0] Provides Access to Port Sub- Registers	Port D Output Data PDOUT (FDFH - Read/Write)  D7 D6 D5 D4 D3 D2 D1 D0  Port D Output Data [7:0]
Port C Input Data PCIN (FDAH - Read Only)  D7 D6 D5 D4 D3 D2 D1 D0  Port C Input Data [7:0]  Port C Output Data PCOUTPUT DATA PC	Port E Address PEADDR (FE0H - Read/Write)  D7 D6 D5 D4 D3 D2 D1 D0  Port E Address[7:0] Selects Port Sub-Registers: 00H = No function 01H = Data direction 02H = Alternate function 03H = Output control (open-drain) 04H = High drive enable 05H = Stop Mode Recovery enable 06H-FFH = No function
Port D Address PDADDR (FDCH - Read/Write)  D7 D6 D5 D4 D3 D2 D1 D0	Port E Control PECTL (FE1H - Read/Write)  D7 D6 D5 D4 D3 D2 D1 D0  Port E Control[7:0] Provides Access to Port Sub- Registers
Port D Address[7:0] Selects Port Sub-Registers: 00H = No function 01H = Data direction 02H = Alternate function 03H = Output control (open-drain) 04H = High drive enable 05H = Stop Mode Recovery enable 06H-FFH = No function	Port E Input Data PEIN (FE2H - Read Only)  D7 D6 D5 D4 D3 D2 D1 D0  Port E Input Data [7:0]
Port D Control PDCTL (FDDH - Read/Write)  D7 D6 D5 D4 D3 D2 D1 D0  Port D Control[7:0] Provides Access to Port Sub- Registers	Port E Output Data PEOUT (FE3H - Read/Write)  D7 D6 D5 D4 D3 D2 D1 D0  Port E Output Data [7:0]
Port D Input Data PDIN (FDE H- Read Only)  D7 D6 D5 D4 D3 D2 D1 D0  Port D Input Data [7:0]	Port F Address PFADDR (FE4H - Read/Write)  D7 D6 D5 D4 D3 D2 D1 D0  Port F Address[7:0] Selects Port Sub-Registers: 00H = No function 01H = Data direction 02H = Alternate function 03H = Output control (open-drain) 04H = High drive enable 05H = Stop Mode Recovery enable 06H-FFH = No function

	Port F Control[7:0] Provides Access to Port Sub- Registers	Port G Output Data PGOUT (FEBH - Re D7 D6 D5 D4 D3 D2 D1 D0	ead/Write) Port G Output Data [7:0]
Port F Input Data PFIN (FE6H - Read D7 D6 D5 D4 D3 D2 D1 D0  Port F Output Data PFOUT (FE7H - Rea D7 D6 D5 D4 D3 D2 D1 D0	Port F Input Data [7:0]	Port H Address PHADDR (FECH - F D7 D6 D5 D4 D3 D2 D1 D0	Port H Address[7:0] Selects Port Sub-Registers: 00H = No function 01H = Data direction 02H = Alternate function 03H = Output control (open-drain) 04H = High drive enable 05H = Stop Mode Recovery enable 06H-FFH = No function
	Port F Output Data [7:0] Read/Write)	Port H Control PHCTL (FEDH - Rea D7 D6 D5 D4 D3 D2 D1 D0	ad/Write)  Port H Control [3:0]  Provides Access to Port Sub-
	Port G Address[7:0] Selects Port Sub-Registers: 00H = No function 01H = Data direction 02H = Alternate function 03H = Output control (open-drain) 04H = High drive enable 05H = Stop Mode Recovery enable 06H-FFH = No function	Port H Input Data PHIN (FEEH - Read	Registers Reserved Only)
Port G Control PGCTL (FE9H - Rea D7 D6 D5 D4 D3 D2 D1 D0	ad/Write)		Port H Input Data [3:0] Reserved
	Port G Control[7:0] Provides Access to Port Sub- Registers	Port H Output Data PHOUT (FEFH - Re D7 D6 D5 D4 D3 D2 D1 D0	ad/Write) Port H Output Data [3:0]
Port G Input Data PGIN (FEAH - Read D7 D6 D5 D4 D3 D2 D1 D0	Only) Port G Input Data [7:0]		Reserved

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**Flags** FLAGS (FFC - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0 F1 - User Flag 1 F2 - User Flag 2 H - Half Carry D - Decimal Adjust V - Overflow Flag S - Sign Flag Z - Zero Flag C - Carry Flag Register Pointer
RP (FFDH - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0 Working Register Page Address Working Register Group Address Stack Pointer High Byte SPH (FFEH - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0 Stack Pointer [15:8] Stack Pointer Low Byte SPL (FFFH - Read/Write) D7 D6 D5 D4 D3 D2 D1 D0

Stack Pointer [7:0]

# Reset and Stop Mode Recovery

#### Overview

The Reset Controller within the Z8 Encore! XP 64K Series Flash Microcontrollers controls Reset and Stop Mode Recovery operation. In typical operation, the following events cause a Reset to occur:

- Power-On Reset
- Voltage Brownout
- Watchdog Timer time-out (when configured via the WDT RES Option Bit to initiate a Reset)
- External RESET pin assertion
- On-Chip Debugger initiated Reset (OCDCTL[0] set to 1)

When the 64K Series devices are in STOP mode, a Stop Mode Recovery is initiated by either of the following:

- Watchdog Timer time-out
- GPIO Port input pin transition on an enabled Stop Mode Recovery source
- DBG pin driven Low

### **Reset Types**

The 64K Series provides two different types of reset operation (system reset and Stop Mode Recovery). The type of Reset is a function of both the current operating mode of the 64K Series devices and the source of the Reset. Table 8 lists the types of Reset and their operating characteristics.

Table 8. Reset and Stop Mode Recovery Characteristics and Latency

	Reset Characteristics and Latency		
Reset Type	Control Registers	eZ8 <sup>TM</sup> CPU	Reset Latency (Delay)
System reset	Reset (as applicable)	Reset	66 WDT Oscillator cycles + 16 System Clock cycles
Stop Mode Recovery	Unaffected, except WDT_CTL register	Reset	66 WDT Oscillator cycles + 16 System Clock cycles

#### **System Reset**

During a system reset, the 64K Series devices are held in Reset for 66 cycles of the Watchdog Timer oscillator followed by 16 cycles of the system clock. At the beginning of Reset, all GPIO pins are configured as inputs.

During Reset, the eZ8 CPU and on-chip peripherals are idle; however, the on-chip crystal oscillator and Watchdog Timer oscillator continue to run. The system clock begins operating following the Watchdog Timer oscillator cycle count. The eZ8 CPU and on-chip peripherals remain idle through the 16 cycles of the system clock.

Upon Reset, control registers within the Register File that have a defined Reset value are loaded with their reset values. Other control registers (including the Stack Pointer, Register Pointer, and Flags) and general-purpose RAM are undefined following Reset. The eZ8 CPU fetches the Reset vector at Program Memory addresses 0002H and 0003H and loads that value into the Program Counter. Program execution begins at the Reset vector address.

#### **Reset Sources**

Table 9 lists the reset sources as a function of the operating mode. The text following provides more detailed information on the individual Reset sources. A Power-On Reset/Voltage Brownout event always takes priority over all other possible reset sources to ensure a full system reset occurs.

Table 9. Reset Sources and Resulting Reset Type

Operating Mode	Reset Source	Reset Type
NORMAL or HALT modes	Power-On Reset/Voltage Brownout	system reset
	Watchdog Timer time-out when configured for Reset	system reset
	RESET pin assertion	system reset
	On-Chip Debugger initiated Reset (OCDCTL[0] set to 1)	system reset except the On-Chip Debugger is unaffected by the reset
STOP mode	Power-On Reset/Voltage Brownout	system reset
	RESET pin assertion	system reset
	DBG pin driven Low	system reset

#### **Power-On Reset**

Each device in the 64K Series contains an internal Power-On Reset circuit. The POR circuit monitors the supply voltage and holds the device in the Reset state until the supply voltage reaches a safe operating level. After the supply voltage exceeds the POR voltage threshold ( $V_{POR}$ ), the POR Counter is enabled and counts 66 cycles of the Watchdog Timer oscillator. After the POR counter times out, the XTAL Counter is enabled to count a total of 16 system clock pulses. The devices are held in the Reset state until both the POR Counter and XTAL counter have timed out. After the 64K Series devices exit the Power-On Reset state, the eZ8 CPU fetches the Reset vector. Following Power-On Reset, the POR status bit in the Watchdog Timer Control (WDTCTL) register is set to 1.

Figure 8 displays Power-On Reset operation. For the POR threshold voltage (V<sub>POR</sub>), see Electrical Characteristics on page 215.

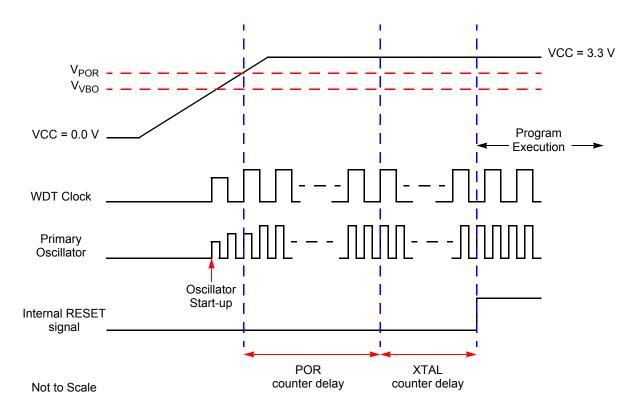


Figure 8. Power-On Reset Operation

#### **Voltage Brownout Reset**

The devices in the 64K Series provide low Voltage Brownout protection. The VBO circuit senses when the supply voltage drops to an unsafe level (below the VBO threshold voltage) and forces the device into the Reset state. While the supply voltage remains below the Power-On Reset voltage threshold ( $V_{POR}$ ), the VBO block holds the device in the Reset state.

After the supply voltage again exceeds the Power-On Reset voltage threshold, the devices progress through a full system reset sequence, as described in the Power-On Reset section. Following Power-On Reset, the POR status bit in the Watchdog Timer Control (WDTCTL) register is set to 1. Figure 9 displays Voltage Brownout operation. For the VBO and POR threshold voltages ( $V_{VBO}$  and  $V_{POR}$ ), see Electrical Characteristics on page 215.

The Voltage Brownout circuit can be either enabled or disabled during STOP mode. Operation during STOP mode is set by the VBO\_AO Option Bit. For information on configuring VBO\_AO, see Option Bits page 195.

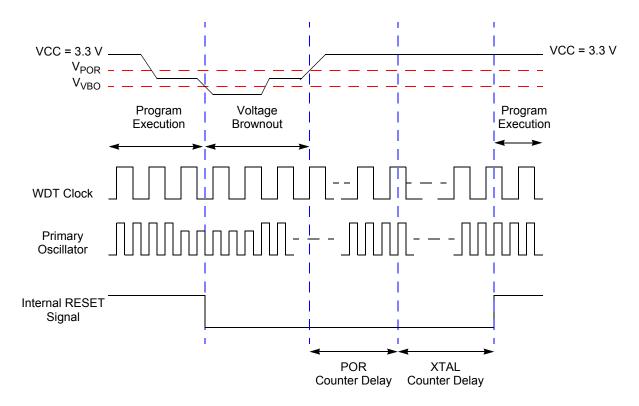


Figure 9. Voltage Brownout Reset Operation

#### **Watchdog Timer Reset**

If the device is in normal or HALT mode, the Watchdog Timer can initiate a system reset at time-out if the WDT\_RES Option Bit is set to 1. This capability is the default (unprogrammed) setting of the WDT\_RES Option Bit. The WDT status bit in the WDT Control register is set to signify that the reset was initiated by the Watchdog Timer.

#### **External Pin Reset**

The RESET pin has a Schmitt-triggered input, an internal pull-up, an analog filter and a digital filter to reject noise. Once the RESET pin is asserted for at least 4 system clock cycles, the devices progress through the system reset sequence. While the RESET input pin is asserted Low, the 64K Series devices continue to be held in the Reset state. If the RESET pin is held Low beyond the system reset time-out, the devices exit the Reset state immediately following RESET pin deassertion. Following a system reset initiated by the external RESET pin, the EXT status bit in the Watchdog Timer Control (WDTCTL) register is set to 1.

#### **On-Chip Debugger Initiated Reset**

A Power-On Reset can be initiated using the On-Chip Debugger by setting the RST bit in the OCD Control register. The On-Chip Debugger block is not reset but the rest of the chip goes through a normal system reset. The RST bit automatically clears during the system reset. Following the system reset the POR bit in the WDT Control register is set.

## **Stop Mode Recovery**

STOP mode is entered by the eZ8 executing a STOP instruction. For detailed STOP mode information, see Low-Power Modes on page 47. During Stop Mode Recovery, the devices are held in reset for 66 cycles of the Watchdog Timer oscillator followed by 16 cycles of the system clock. Stop Mode Recovery only affects the contents of the Watchdog Timer Control register. Stop Mode Recovery does not affect any other values in the Register File, including the Stack Pointer, Register Pointer, Flags, peripheral control registers, and general-purpose RAM.

The eZ8<sup>™</sup> CPU fetches the Reset vector at Program Memory addresses 0002H and 0003H and loads that value into the Program Counter. Program execution begins at the Reset vector address. Following Stop Mode Recovery, the STOP bit in the Watchdog Timer Control Register is set to 1. Table 10 lists the Stop Mode Recovery sources and resulting actions.

**Table 10. Stop Mode Recovery Sources and Resulting Action** 

Operating Mode	Stop Mode Recovery Source	Action
STOP mode Watchdog Timer time-out Stop Mode when configured for Reset		Stop Mode Recovery
	Watchdog Timer time-out when configured for interrupt	Stop Mode Recovery followed by interrupt (if interrupts are enabled)
	Data transition on any GPIO Port pin enabled as a Stop Mode Recovery source	Stop Mode Recovery

## **Stop Mode Recovery Using Watchdog Timer Time-Out**

If the Watchdog Timer times out during STOP mode, the device undergoes a Stop Mode Recovery sequence. In the Watchdog Timer Control register, the WDT and STOP bits are set to 1. If the Watchdog Timer is configured to generate an interrupt upon time-out and the 64K Series devices are configured to respond to interrupts, the eZ8 CPU services the Watchdog Timer interrupt request following the normal Stop Mode Recovery sequence.

## Stop Mode Recovery Using a GPIO Port Pin Transition HALT

Each of the GPIO Port pins may be configured as a Stop Mode Recovery input source. On any GPIO pin enabled as a Stop Mode Recovery source, a change in the input pin value (from High to Low or from Low to High) initiates Stop Mode Recovery. The GPIO Stop Mode Recovery signals are filtered to reject pulses less than 10 ns (typical) in duration. In the Watchdog Timer Control register, the STOP bit is set to 1.



**Caution:** *In STOP mode, the GPIO Port Input Data registers (PxIN) are disabled. The* Port Input Data registers record the Port transition only if the signal stays on the Port pin through the end of the Stop Mode Recovery delay. Thus, short pulses on the Port pin can initiate Stop Mode Recovery without being written to the Port Input Data register or without initiating an interrupt (if enabled for that pin).

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## **Low-Power Modes**

#### **Overview**

The 64K Series products contain power-saving features. The highest level of power reduction is provided by STOP mode. The next level of power reduction is provided by the HALT mode.

#### **STOP Mode**

Execution of the eZ8<sup>TM</sup> CPU's STOP instruction places the device into STOP mode. In STOP mode, the operating characteristics are:

- Primary crystal oscillator is stopped; the XIN pin is driven High and the XOUT pin is driven Low.
- System clock is stopped.
- eZ8 CPU is stopped.
- Program counter (PC) stops incrementing.
- The Watchdog Timer and its internal RC oscillator continue to operate, if enabled for operation during STOP mode.
- The Voltage Brownout protection circuit continues to operate, if enabled for operation in STOP mode using the associated Option Bit.
- All other on-chip peripherals are idle.

To minimize current in STOP mode, all GPIO pins that are configured as digital inputs must be driven to one of the supply rails ( $V_{CC}$  or GND), the Voltage Brownout protection must be disabled, and the Watchdog Timer must be disabled. The devices can be brought out of STOP mode using Stop Mode Recovery. For more information on Stop Mode Recovery, see Reset and Stop Mode Recovery on page 47.

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**Caution:** STOP mode must not be used when driving the 64K Series devices with an external clock driver source.

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#### **HALT Mode**

Execution of the eZ8 CPU's HALT instruction places the device into HALT mode. In HALT mode, the operating characteristics are:

- Primary crystal oscillator is enabled and continues to operate.
- System clock is enabled and continues to operate.
- eZ8 CPU is stopped.
- Program Counter stops incrementing.
- Watchdog Timer's internal RC oscillator continues to operate.
- The Watchdog Timer continues to operate, if enabled.
- All other on-chip peripherals continue to operate.

The eZ8 CPU can be brought out of HALT mode by any of the following operations:

- Interrupt
- Watchdog Timer time-out (interrupt or reset)
- Power-On Reset
- Voltage Brownout Reset
- External RESET pin assertion

To minimize current in HALT mode, all GPIO pins which are configured as inputs must be driven to one of the supply rails ( $V_{CC}$  or GND).

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# **General-Purpose I/O**

#### **Overview**

The 64K Series products support a maximum of seven 8-bit ports (Ports A-G) and one 4-bit port (Port H) for general-purpose input/output (GPIO) operations. Each port consists of control and data registers. The GPIO control registers are used to determine data direction, open-drain, output drive current and alternate pin functions. Each port pin is individually programmable. All ports (except B and H) support 5 V-tolerant inputs.

## **GPIO Port Availability By Device**

Table 11 lists the port pins available with each device and package type.

Table 11. Port Availability by Device and Package Type

Device	Packages	Port A	Port B	Port C	Port D	Port E	Port F	Port G	Port H
Z8X1621	40-pin	[7:0]	[7:0]	[6:0]	[6:3, 1:0]	-	-	-	-
Z8X1621	44-pin	[7:0]	[7:0]	[7:0]	[6:0]	-	-	-	-
Z8X1622	64- and 68-pin	[7:0]	[7:0]	[7:0]	[7:0]	[7:0]	[7]	[3]	[3:0]
Z8X2421	40-pin	[7:0]	[7:0]	[6:0]	[6:3, 1:0]	-	-	-	-
Z8X2421	44-pin	[7:0]	[7:0]	[7:0]	[6:0]	-	-	-	-
Z8X2422	64- and 68-pin	[7:0]	[7:0]	[7:0]	[7:0]	[7:0]	[7]	[3]	[3:0]
Z8X3221	40-pin	[7:0]	[7:0]	[6:0]	[6:3, 1:0]	-	-	-	-
Z8X3221	44-pin	[7:0]	[7:0]	[7:0]	[6:0]	-	-	-	-
Z8X3222	64- and 68-pin	[7:0]	[7:0]	[7:0]	[7:0]	[7:0]	[7]	[3]	[3:0]
Z8X4821	40-pin	[7:0]	[7:0]	[6:0]	[6:3, 1:0]	-	-	-	-
Z8X4821	44-pin	[7:0]	[7:0]	[7:0]	[6:0]	-	-	-	-
Z8X4822	64- and 68-pin	[7:0]	[7:0]	[7:0]	[7:0]	[7:0]	[7]	[3]	[3:0]

General-Purpose I/O PS019919-1207



Table 11. Port Availability by Device and Package Type (Continued)

Device	Packages	Port A	Port B	Port C	Port D	Port E	Port F	Port G	Port H
Z8X4823	80-pin	[7:0]	[7:0]	[7:0]	[7:0]	[7:0]	[7:0]	[7:0]	[3:0]
Z8X6421	40-pin	[7:0]	[7:0]	[6:0]	[6:3, 1:0]	-	-	-	-
Z8X6421	44-pin	[7:0]	[7:0]	[7:0]	[6:0]	-	-	-	-
Z8X6422	64- and 68-pin	[7:0]	[7:0]	[7:0]	[7:0]	[7:0]	[7]	[3]	[3:0]
Z8X6423	80-pin	[7:0]	[7:0]	[7:0]	[7:0]	[7:0]	[7:0]	[7:0]	[3:0]

#### **Architecture**

Figure 10 displays a simplified block diagram of a GPIO port pin. In Figure 10, the ability to accommodate alternate functions and variable port current drive strength are not illustrated.

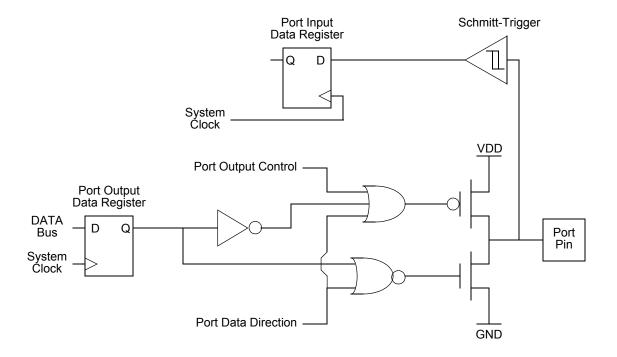


Figure 10. GPIO Port Pin Block Diagram

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### **GPIO Alternate Functions**

Many of the GPIO port pins can be used as both general-purpose I/O and to provide access to on-chip peripheral functions such as the timers and serial communication devices. The Port A–H Alternate Function sub-registers configure these pins for either general-purpose I/O or alternate function operation. When a pin is configured for alternate function, control of the port pin direction (input/output) is passed from the Port A–H Data Direction registers to the alternate function assigned to this pin. Table 12 lists the alternate functions associated with each port pin.

**Table 12. Port Alternate Function Mapping** 

Port	Pin	Mnemonic	Alternate Function Description				
Port A	PA0	T0IN	Timer 0 Input				
	PA1	T0OUT	Timer 0 Output				
	PA2	DE0	UART 0 Driver Enable				
	PA3	CTS0	UART 0 Clear to Send				
	PA4	RXD0/IRRX0	UART 0/IrDA 0 Receive Data				
	PA5	TXD0/IRTX0	UART 0/IrDA 0 Transmit Data				
	PA6	SCL	I <sup>2</sup> C Clock (automatically open-drain)				
	PA7	SDA	I <sup>2</sup> C Data (automatically open-drain)				
Port B	PB0	ANA0	ADC Analog Input 0				
	PB1	ANA1	ADC Analog Input 1				
	PB2	ANA2	ADC Analog Input 2				
	PB3	ANA3	ADC Analog Input 3				
	PB4	ANA4	ADC Analog Input 4				
	PB5	ANA5	ADC Analog Input 5				
	PB6	ANA6	ADC Analog Input 6				
	PB7	ANA7	ADC Analog Input 7				

**Table 12. Port Alternate Function Mapping (Continued)** 

Port	Pin	Mnemonic	Alternate Function Description					
Port C	PC0	T1IN	Timer 1 Input					
	PC1	T10UT	Timer 1 Output					
	PC2	SS	SPI Slave Select					
	PC3	SCK	SPI Serial Clock					
	PC4	MOSI	SPI Master Out/Slave In					
	PC5	MISO	SPI Master In/Slave Out					
	PC6	T2IN	Timer 2 In					
	PC7	T2OUT	Timer 2 Out					
Port D	PD0	T3IN	Timer 3 In (unavailable in 44-pin packages)					
	PD1	T3OUT	Timer 3 Out (unavailable in 44-pin packages)					
	PD2	N/A	No alternate function					
	PD3	DE1	UART 1 Driver Enable					
	PD4	RXD1/IRRX1	UART 1/IrDA 1 Receive Data					
	PD5	TXD1/IRTX1	UART 1/IrDA 1 Transmit Data					
	PD6	CTS1	UART 1 Clear to Send					
	PD7	RCOUT	Watchdog Timer RC Oscillator Output					
Port E	PE[7:0]	N/A	No alternate functions					
Port F	PF[7:0]	N/A	No alternate functions					
Port G	PG[7:0]	N/A	No alternate functions					
Port H	PH0	ANA8	ADC Analog Input 8					
	PH1	ANA9	ADC Analog Input 9					
	PH2	ANA10	ADC Analog Input 10					
	PH3	ANA11	ADC Analog Input 11					

# **GPIO Interrupts**

Many of the GPIO port pins can be used as interrupt sources. Some port pins may be configured to generate an interrupt request on either the rising edge or falling edge of the pin input signal. Other port pin interrupts generate an interrupt when any edge occurs (both rising and falling). For more information on interrupts using the GPIO pins, see Interrupt Controller on page 67.

# **GPIO Control Register Definitions**

Four registers for each Port provide access to GPIO control, input data, and output data. Table 13 lists these Port registers. Use the Port A–H Address and Control registers together to provide access to sub-registers for Port configuration and control.

Table 13. GPIO Port Registers and Sub-Registers

Port Register Mnemonic	Port Register Name
PxADDR	Port A–H Address Register (Selects sub-registers)
PxCTL	Port A–H Control Register (Provides access to sub-registers)
PxIN	Port A–H Input Data Register
PxOUT	Port A–H Output Data Register
Port Sub-Register Mnemonic	Port Register Name
$\frac{ \text{Port Sub-Register Mnemonic}}{ \overline{\text{PxDD}}}$	Port Register Name  Data Direction
PxDD	Data Direction
PxDD PxAF	Data Direction  Alternate Function

# Port A-H Address Registers

The Port A–H Address registers select the GPIO Port functionality accessible through the Port A–H Control registers. The Port A–H Address and Control registers combine to provide access to all GPIO Port control (Table 14).

Table 14. Port A–H GPIO Address Registers (PxADDR)

BITS	7	6	5	4	3	2	1	0			
FIELD		PADDR[7:0]									
RESET		00Н									
R/W		R/W									
ADDR		FD0	H, FD4H, F	D8H, FDCH	, FE0H, FE4	H, FE8H, FI	ECH				



# PADDR[7:0]—Port Address

The Port Address selects one of the sub-registers accessible through the Port Control register.

PADDR[7:0]	Port Control sub-register accessible using the Port A–H Control Registers
00H	No function. Provides some protection against accidental Port reconfiguration
01H	Data Direction
02H	Alternate Function
03H	Output Control (Open-Drain)
04H	High Drive Enable
05H	Stop Mode Recovery Source Enable
06H-FFH	No function

# Port A-H Control Registers

The Port A–H Control registers set the GPIO port operation. The value in the corresponding Port A–H Address register determines the control sub-registers accessible using the Port A–H Control register (Table 15).

Table 15. Port A-H Control Registers (PxCTL)

BITS	7	6	5	4	3	2	1	0			
FIELD		PCTL									
RESET		00H									
R/W		R/W									
ADDR		FD1	H, FD5H, F	D9H, FDDH	FE1H, FE5	SH, FE9H, FI	EDH				

PCTL[7:0]—Port Control

The Port Control register provides access to all sub-registers that configure the GPIO Port operation.



# Port A-H Data Direction Sub-Registers

The Port A–H Data Direction sub-register is accessed through the Port A–H Control register by writing 01H to the Port A–H Address register (Table 16).

Table 16. Port A-H Data Direction Sub-Registers

вітѕ	7	6	5	4	3	2	1	0			
FIELD	DD7	DD6	DD5	DD4	DD3	DD2	DD1	DD0			
RESET		1									
R/W		R/W									
ADDR	If 01F	l in Port A-l	H Address R	egister, acce	essible throu	gh Port A-F	l Control Re	gister			

DD[7:0]—Data Direction

These bits control the direction of the associated port pin. Port Alternate Function operation overrides the Data Direction register setting.

- 0 = Output. Data in the Port A–H Output Data register is driven onto the port pin.
- 1 = Input. The port pin is sampled and the value written into the Port A–H Input Data Register. The output driver is tri-stated.

# Port A-H Alternate Function Sub-Registers

The Port A–H Alternate Function sub-register (Table 17) is accessed through the Port A-H Control register by writing 02H to the Port A-H Address register. The Port A-H Alternate Function sub-registers select the alternate functions for the selected pins. To determine the alternate function associated with each port pin, see GPIO Alternate Functions on page 59.



**Caution:** Do not enable alternate function for GPIO port pins which do not have an associated alternate function. Failure to follow this guideline may result in unpredictable operation.

Table 17. Port A-H Alternate Function Sub-Registers

BITS	7	6	5	4	3	2	1	0			
FIELD	AF7	AF7         AF6         AF5         AF4         AF3         AF2         AF1         A									
RESET		0									
R/W		R/W									
ADDR	If 02F	l in Port A-l	H Address R	egister, acce	essible throu	gh Port A–F	l Control Re	gister			

AF[7:0]—Port Alternate Function enabled

- 0 = The port pin is in NORMAL mode and the DDx bit in the Port A-H Data Direction sub-register determines the direction of the pin.
- 1 = The alternate function is selected. Port pin operation is controlled by the alternate function.

# Port A-H Output Control Sub-Registers

The Port A–H Output Control sub-register (Table 18) is accessed through the Port A-H Control register by writing 03H to the Port A-H Address register. Setting the bits in the Port A–H Output Control sub-registers to 1 configures the specified port pins for open-drain operation. These sub-registers affect the pins directly and, as a result, alternate functions are also affected.

Table 18. Port A-H Output Control Sub-Registers

BITS	7	6	5	4	3	2	1	0		
FIELD	POC7	POC7         POC6         POC5         POC4         POC3         POC2         POC1         POC								
RESET	0									
R/W		R/W								
ADDR	If 03⊦	I in Port A-F	H Address R	egister, acce	essible throu	gh Port A-F	l Control Re	gister		

POC[7:0]—Port Output Control

These bits function independently of the alternate function bit and disables the drains if set to 1.

0 = The drains are enabled for any output mode.

1 = The drain of the associated pin is disabled (open-drain mode).

#### Port A-H High Drive Enable Sub-Registers

The Port A–H High Drive Enable sub-register (Table 19) is accessed through the Port A– H Control register by writing 04H to the Port A-H Address register. Setting the bits in the Port A–H High Drive Enable sub-registers to 1 configures the specified port pins for high current output drive operation. The Port A-H High Drive Enable sub-register affects the pins directly and, as a result, alternate functions are also affected.

Table 19. Port A-H High Drive Enable Sub-Registers

BITS	7	6	5	4	3	2	1	0		
FIELD	PHDE7	PHDE6	PHDE5	PHDE4	PHDE3	PHDE2	PHDE1	PHDE0		
RESET	0									
R/W		R/W								
ADDR	If 041	I in Port A-l	H Address R	egister, acce	essible throu	gh Port A-H	Control Reg	gister		

PHDE[7:0]—Port High Drive Enabled

0 = The Port pin is configured for standard output current drive.

1 = The Port pin is configured for high output current drive.

# Port A-H Stop Mode Recovery Source Enable Sub-Registers

The Port A–H Stop Mode Recovery Source Enable sub-register (Table 20) is accessed through the Port A–H Control register by writing 05H to the Port A–H Address register. Setting the bits in the Port A–H Stop Mode Recovery Source Enable sub-registers to 1 configures the specified Port pins as a Stop Mode Recovery source. During STOP Mode, any logic transition on a Port pin enabled as a Stop Mode Recovery source initiates Stop Mode Recovery.

Table 20. Port A–H Stop Mode Recovery Source Enable Sub-Registers

BITS	7	6	5	4	3	2	1	0		
FIELD	PSMRE7	PSMRE6	PSMRE5	PSMRE4	PSMRE3	PSMRE2	PSMRE1	PSMRE0		
RESET	0									
R/W		R/W								
ADDR	If 05H	I in Port A-l	H Address R	egister, acc	essible throu	igh Port A-F	l Control Re	gister		

PSMRE[7:0]—Port Stop Mode Recovery Source Enabled

- 0 = The Port pin is not configured as a Stop Mode Recovery source. Transitions on this pin during STOP mode do not initiate Stop Mode Recovery.
- 1 = The Port pin is configured as a Stop Mode Recovery source. Any logic transition on this pin during STOP mode initiates Stop Mode Recovery.

# Port A-H Input Data Registers

Reading from the Port A–H Input Data registers (Table 21) returns the sampled values from the corresponding port pins. The Port A–H Input Data registers are Read-only.

Table 21. Port A-H Input Data Registers (PxIN)

BITS	7	6	5	4	3	2	1	0			
FIELD	PIN7	N7 PIN6 PIN5 PIN4 PIN3 PIN2 PIN1 P									
RESET		X									
R/W		R									
ADDR		FD2	H, FD6H, FI	DAH, FDEH	, FE2H, FE6	H, FEAH, F	EEH				

PIN[7:0]—Port Input Data

Sampled data from the corresponding port pin input.

0 = Input data is logical 0 (Low).

1 = Input data is logical 1 (High).

# Port A-H Output Data Register

The Port A–H Output Data register (Table 22) writes output data to the pins.

Table 22. Port A-H Output Data Register (PxOUT)

BITS	7	6	5	4	3	2	1	0
FIELD	POUT7	POUT6	POUT5	POUT4	POUT3	POUT2	POUT1	POUT0
RESET	0							
R/W		R/W						
ADDR		FD3	H, FD7H, FI	DBH, FDFH,	, FE3H, FE7	H, FEBH, F	EFH	

POUT[7:0]—Port Output Data

These bits contain the data to be driven out from the port pins. The values are only driven if the corresponding pin is configured as an output and the pin is not configured for alternate function operation.

0 = Drive a logical 0 (Low).

1= Drive a logical 1 (High). High value is not driven if the drain has been disabled by setting the corresponding Port Output Control register bit to 1.

# **Interrupt Controller**

#### **Overview**

The interrupt controller on the 64K Series products prioritizes the interrupt requests from the on-chip peripherals and the GPIO port pins. The features of the interrupt controller include the following:

- 24 unique interrupt vectors:
  - 12 GPIO port pin interrupt sources
  - 12 on-chip peripheral interrupt sources
- Flexible GPIO interrupts
  - Eight selectable rising and falling edge GPIO interrupts
  - Four dual-edge interrupts
- Three levels of individually programmable interrupt priority
- Watchdog Timer can be configured to generate an interrupt

Interrupt requests (IRQs) allow peripheral devices to suspend CPU operation in an orderly manner and force the CPU to start an interrupt service routine (ISR). Usually this interrupt service routine is involved with the exchange of data, status information, or control information between the CPU and the interrupting peripheral. When the service routine is completed, the CPU returns to the operation from which it was interrupted.

The eZ8 CPU supports both vectored and polled interrupt handling. For polled interrupts, the interrupt control has no effect on operation. For more information on interrupt servicing by the eZ8 CPU, refer to eZ8 CPU Core User Manual (UM0128) available for download at www.zilog.com.

# **Interrupt Vector Listing**

Table 23 lists all of the interrupts available in order of priority. The interrupt vector is stored with the most-significant byte (MSB) at the even Program Memory address and the least-significant byte (LSB) at the following odd Program Memory address.

**Table 23. Interrupt Vectors in Order of Priority** 

	Program Memory	
Priority	Vector Address	Interrupt Source
Highest	0002H	Reset (not an interrupt)
	0004H	Watchdog Timer (see Watchdog Timer on page 97)
	0006H	Illegal Instruction Trap (not an interrupt)
	0008H	Timer 2
	000AH	Timer 1
	000CH	Timer 0
	000EH	UART 0 receiver
	0010H	UART 0 transmitter
	0012H	I <sup>2</sup> C
	0014H	SPI
	0016H	ADC
	0018H	Port A7 or Port D7, rising or falling input edge
	001AH	Port A6 or Port D6, rising or falling input edge
	001CH	Port A5 or Port D5, rising or falling input edge
	001EH	Port A4 or Port D4, rising or falling input edge
	0020H	Port A3 or Port D3, rising or falling input edge
	0022H	Port A2 or Port D2, rising or falling input edge
	0024H	Port A1 or Port D1, rising or falling input edge
	0026H	Port A0 or Port D0, rising or falling input edge
	0028H	Timer 3 (not available in 44-pin packages)
	002AH	UART 1 receiver
	002CH	UART 1 transmitter
	002EH	DMA
	0030H	Port C3, both input edges
	0032H	Port C2, both input edges
-	0034H	Port C1, both input edges
Lowest	0036H	Port C0, both input edges

#### **Architecture**

Figure 11 displays a block diagram of the interrupt controller.

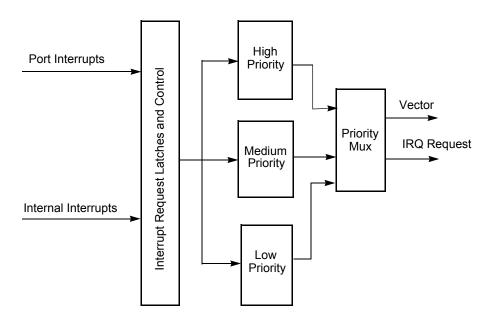


Figure 11. Interrupt Controller Block Diagram

# **Operation**

# Master Interrupt Enable

The master interrupt enable bit (IRQE) in the Interrupt Control register globally enables and disables interrupts.

Interrupts are globally enabled by any of the following actions:

- Executing an Enable Interrupt (EI) instruction.
- Executing an Return from Interrupt (IRET) instruction.
- Writing a 1 to the IRQE bit in the Interrupt Control register.

Interrupts are globally disabled by any of the following actions:

- Execution of a Disable Interrupt (DI) instruction.
- eZ8 CPU acknowledgement of an interrupt service request from the interrupt controller.
- Writing a 0 to the IRQE bit in the Interrupt Control register.
- Reset.

- Executing a Trap instruction.
- Illegal Instruction trap.

# **Interrupt Vectors and Priority**

The interrupt controller supports three levels of interrupt priority. Level 3 is the highest priority, Level 2 is the second highest priority, and Level 1 is the lowest priority. If all of the interrupts were enabled with identical interrupt priority (all as Level 2 interrupts, for example), then interrupt priority would be assigned from highest to lowest as specified in Table 23 on page 68. Level 3 interrupts always have higher priority than Level 2 interrupts which, in turn, always have higher priority than Level 1 interrupts. Within each interrupt priority level (Level 1, Level 2, or Level 3), priority is assigned as specified in Table 23 on page 68. Reset, Watchdog Timer interrupt (if enabled), and Illegal Instruction Trap always have highest priority.

# Interrupt Assertion

Interrupt sources assert their interrupt requests for only a single system clock period (single pulse). When the interrupt request is acknowledged by the eZ8 CPU, the corresponding bit in the Interrupt Request register is cleared until the next interrupt occurs. Writing a 0 to the corresponding bit in the Interrupt Request register likewise clears the interrupt request.



**Caution:** The following style of coding to clear bits in the Interrupt Request registers is NOT recommended. All incoming interrupts that are received between execution of the first LDX command and the last LDX command are lost.

#### Poor coding style that can result in lost interrupt requests:

```
LDX r0, IRQ0
AND r0, MASK
LDX IRQ0, r0
```

To avoid missing interrupts, the following style of coding to clear bits in the Interrupt Request 0 register is recommended:

#### Good coding style that avoids lost interrupt requests:

ANDX IRQ0, MASK

# **Software Interrupt Assertion**

Program code can generate interrupts directly. Writing a 1 to the desired bit in the Interrupt Request register triggers an interrupt (assuming that interrupt is enabled). When the interrupt request is acknowledged by the eZ8 CPU, the bit in the Interrupt Request register is automatically cleared to 0.



**Caution:** The following style of coding to generate software interrupts by setting bits in the Interrupt Request registers is NOT recommended. All incoming interrupts that are received between execution of the first LDX command and the last LDX command are lost.

# Poor coding style that can result in lost interrupt requests:

LDX r0, IRQ0 OR r0, MASK LDX IRQ0, r0

To avoid missing interrupts, the following style of coding to set bits in the *Interrupt Request registers is recommended:* 

#### Good coding style that avoids lost interrupt requests:

ORX IRQ0, MASK

# **Interrupt Control Register Definitions**

For all interrupts other than the Watchdog Timer interrupt, the interrupt control registers enable individual interrupts, set interrupt priorities, and indicate interrupt requests.

# Interrupt Request 0 Register

The Interrupt Request 0 (IRO0) register (Table 24) stores the interrupt requests for both vectored and polled interrupts. When a request is presented to the interrupt controller, the corresponding bit in the IRQ0 register becomes 1. If interrupts are globally enabled (vectored interrupts), the interrupt controller passes an interrupt request to the eZ8<sup>™</sup> CPU. If interrupts are globally disabled (polled interrupts), the eZ8 CPU can read the Interrupt Request 0 register to determine if any interrupt requests are pending

Table 24. Interrupt Request 0 Register (IRQ0)

BITS	7	6	5	4	3	2	1	0
FIELD	T2I	T1I	T0I	U0RXI	U0TXI	I2CI	SPII	ADCI
RESET	0							
R/W	R/W							
ADDR		FC0H						

T2I—Timer 2 Interrupt Request

0 = No interrupt request is pending for Timer 2.

1 = An interrupt request from Timer 2 is awaiting service.

- T1I—Timer 1 Interrupt Request
- 0 =No interrupt request is pending for Timer 1.
- 1 = An interrupt request from Timer 1 is awaiting service.

#### T0I—Timer 0 Interrupt Request

- 0 =No interrupt request is pending for Timer 0.
- 1 = An interrupt request from Timer 0 is awaiting service.

# U0RXI—UART 0 Receiver Interrupt Request

- 0 = No interrupt request is pending for the UART 0 receiver.
- 1 = An interrupt request from the UART 0 receiver is awaiting service.

#### U0TXI—UART 0 Transmitter Interrupt Request

- 0 = No interrupt request is pending for the UART 0 transmitter.
- 1 = An interrupt request from the UART 0 transmitter is awaiting service.

# I<sup>2</sup>CI—I<sup>2</sup>C Interrupt Request

- 0 = No interrupt request is pending for the  $I^2C$ .
- 1 = An interrupt request from the  $I^2C$  is awaiting service.

#### SPII—SPI Interrupt Request

- 0 =No interrupt request is pending for the SPI.
- 1 = An interrupt request from the SPI is awaiting service.

### ADCI—ADC Interrupt Request

- 0 = No interrupt request is pending for the Analog-to-Digital Converter.
- 1 = An interrupt request from the Analog-to-Digital Converter is awaiting service.

# Interrupt Request 1 Register

The Interrupt Request 1 (IRQ1) register (Table 25) stores interrupt requests for both vectored and polled interrupts. When a request is presented to the interrupt controller, the corresponding bit in the IRQ1 register becomes 1. If interrupts are globally enabled (vectored interrupts), the interrupt controller passes an interrupt request to the eZ8 CPU. If interrupts are globally disabled (polled interrupts), the eZ8 CPU can read the Interrupt Request 1 register to determine if any interrupt requests are pending.

Table 25. Interrupt Request 1 Register (IRQ1)

BITS	7	6	5	4	3	2	1	0
FIELD	PAD7I	PAD6I	PAD5I	PAD4I	PAD3I	PAD2I	PAD1I	PAD0I
RESET	0							
R/W	R/W							
ADDR				FC	3H			

PADxI—Port A or Port D Pin x Interrupt Request

0 = No interrupt request is pending for GPIO Port A or Port D pin x.

1 = An interrupt request from GPIO Port A or Port D pin x is awaiting service.

where x indicates the specific GPIO Port pin number (0 through 7). For each pin, only 1 of either Port A or Port D can be enabled for interrupts at any one time. Port selection (A or D) is determined by the values in the Interrupt Port Select Register.

# **Interrupt Request 2 Register**

The Interrupt Request 2 (IRQ2) register (Table 26) stores interrupt requests for both vectored and polled interrupts. When a request is presented to the interrupt controller, the corresponding bit in the IRQ2 register becomes 1. If interrupts are globally enabled (vectored interrupts), the interrupt controller passes an interrupt request to the eZ8 CPU. If interrupts are globally disabled (polled interrupts), the eZ8 CPU can read the Interrupt Request 1 register to determine if any interrupt requests are pending.

#### Table 26. Interrupt Request 2 Register (IRQ2)

BITS	7	6	5	4	3	2	1	0
FIELD	T3I	U1RXI	U1TXI	DMAI	PC3I	PC2I	PC1I	PC0I
RESET	0							
R/W	R/W							
ADDR		FC6H						

T3I—Timer 3 Interrupt Request

0 =No interrupt request is pending for Timer 3.

1 = An interrupt request from Timer 3 is awaiting service.

U1RXI—UART 1 Receive Interrupt Request

0 = No interrupt request is pending for the UART1 receiver.

1 = An interrupt request from UART1 receiver is awaiting service.

U1TXI—UART 1 Transmit Interrupt Request

0 = No interrupt request is pending for the UART 1 transmitter.

1 = An interrupt request from the UART 1 transmitter is awaiting service.

DMAI—DMA Interrupt Request

0 = No interrupt request is pending for the DMA.

1 = An interrupt request from the DMA is awaiting service.

PCxI—Port C Pin x Interrupt Request

0 = No interrupt request is pending for GPIO Port C pin x.

1 = An interrupt request from GPIO Port C pin x is awaiting service.

where *x* indicates the specific GPIO Port C pin number (0 through 3).

# IRQ0 Enable High and Low Bit Registers

The IRQ0 Enable High and Low Bit registers (see Table 28 and Table 29 on page 75) form a priority encoded enabling for interrupts in the Interrupt Request 0 register. Priority is generated by setting bits in each register. Table 27 describes the priority control for IRQ0.

Table 27. IRQ0 Enable and Priority Encoding

IRQ0ENH[x]	IRQ0ENL[x]	Priority	Description
0	0	Disabled	Disabled
0	1	Level 1	Low
1	0	Level 2	Nominal
1	1	Level 3	High

**Note:** where *x* indicates the register bits from 0 through 7.

Table 28. IRQ0 Enable High Bit Register (IRQ0ENH)

BITS	7	6	5	4	3	2	1	0
FIELD	T2ENH	T1ENH	T0ENH	U0RENH	U0TENH	I2CENH	SPIENH	ADCENH
RESET	0							
R/W		R/W						
ADDR				FC	1H			

T2ENH—Timer 2 Interrupt Request Enable High Bit

T1ENH—Timer 1 Interrupt Request Enable High Bit

T0ENH—Timer 0 Interrupt Request Enable High Bit

U0RENH—UART 0 Receive Interrupt Request Enable High Bit

U0TENH—UART 0 Transmit Interrupt Request Enable High Bit

I2CENH—I<sup>2</sup>C Interrupt Request Enable High Bit

SPIENH—SPI Interrupt Request Enable High Bit

ADCENH—ADC Interrupt Request Enable High Bit

Table 29. IRQ0 Enable Low Bit Register (IRQ0ENL)

BITS	7	6	5	4	3	2	1	0
FIELD	T2ENL	T1ENL	T0ENL	U0RENL	U0TENL	I2CENL	SPIENL	ADCENL
RESET	0							
R/W		R/W						
ADDR				FC	2H			

T2ENL—Timer 2 Interrupt Request Enable Low Bit

T1ENL—Timer 1 Interrupt Request Enable Low Bit

T0ENL—Timer 0 Interrupt Request Enable Low Bit

U0RENL—UART 0 Receive Interrupt Request Enable Low Bit

U0TENL—UART 0 Transmit Interrupt Request Enable Low Bit

I2CENL—I<sup>2</sup>C Interrupt Request Enable Low Bit

SPIENL—SPI Interrupt Request Enable Low Bit

ADCENL—ADC Interrupt Request Enable Low Bit

# IRQ1 Enable High and Low Bit Registers

The IRQ1 Enable High and Low Bit registers (see Table 31 and Table 32 on page 76) form a priority encoded enabling for interrupts in the Interrupt Request 1 register. Priority is generated by setting bits in each register. Table 30 describes the priority control for IRQ1.

Table 30. IRQ1 Enable and Priority Encoding

IRQ1ENH[x]	IRQ1ENL[x]	Priority	Description
0	0	Disabled	Disabled
0	1	Level 1	Low
1	0	Level 2	Nominal
1	1	Level 3	High

**Note:** where *x* indicates the register bits from 0 through 7.

Table 31. IRQ1 Enable High Bit Register (IRQ1ENH)

BITS	7	6	5	4	3	2	1	0
FIELD	PAD7ENH	PAD6ENH	PAD5ENH	PAD4ENH	PAD3ENH	PAD2ENH	PAD1ENH	PAD0ENH
RESET	0	0	0	0	0	0	0	0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
ADDR		FC4H						

PADxENH—Port A or Port D Bit[x] Interrupt Request Enable High Bit. For selection of either Port A or Port D as the interrupt source, see Interrupt Port Select Register on page 78.

Table 32. IRQ1 Enable Low Bit Register (IRQ1ENL)

BITS	7	6	5	4	3	2	1	0
FIELD	PAD7ENL	PAD6ENL	PAD5ENL	PAD4ENL	PAD3ENL	PAD2ENL	PAD1ENL	PAD0ENL
RESET	0	0 0 0 0 0 0 0						
R/W	R/W	R/W R/W R/W R/W R/W R/W R/W						
ADDR		FC5H						

PADxENL—Port A or Port D Bit[x] Interrupt Request Enable Low Bit For selection of either Port A or Port D as the interrupt source, see Interrupt Port Select Register on page 78.

# IRQ2 Enable High and Low Bit Registers

The IRQ2 Enable High and Low Bit registers (see Table 34 and Table 35 on page 77) form a priority encoded enabling for interrupts in the Interrupt Request 2 register. Priority is generated by setting bits in each register. Table 33 describes the priority control for IRQ2.

Table 33. IRQ2 Enable and Priority Encoding

IRQ2ENH[x]	IRQ2ENL[x]	Priority	Description
0	0	Disabled	Disabled
0	1	Level 1	Low
1	0	Level 2	Nominal

Table 33. IRQ2 Enable and Priority Encoding (Continued)

IRQ2ENH[x]	IRQ2ENL[x]	Priority	Description
1	1	Level 3	High

**Note:** where *x* indicates the register bits from 0 through 7.

# Table 34. IRQ2 Enable High Bit Register (IRQ2ENH)

BITS	7	6	5	4	3	2	1	0		
FIELD	T3ENH	T3ENH U1RENH U1TENH DMAENH C3ENH C2ENH C1ENH C0EN								
RESET		0								
R/W				R/	W					
ADDR				FC	7H					

T3ENH—Timer 3 Interrupt Request Enable High Bit

U1RENH—UART 1 Receive Interrupt Request Enable High Bit

U1TENH—UART 1 Transmit Interrupt Request Enable High Bit

DMAENH—DMA Interrupt Request Enable High Bit

C3ENH—Port C3 Interrupt Request Enable High Bit

C2ENH—Port C2 Interrupt Request Enable High Bit

C1ENH—Port C1 Interrupt Request Enable High Bit

C0ENH—Port C0 Interrupt Request Enable High Bit

#### Table 35. IRQ2 Enable Low Bit Register (IRQ2ENL)

BITS	7	6	5	4	3	2	1	0		
FIELD	T3ENL	T3ENL U1RENL U1TENL DMAENL C3ENL C2ENL C1ENL C0ENL								
RESET		0								
R/W				R/	W					
ADDR				FC	8H					

T3ENL—Timer 3 Interrupt Request Enable Low Bit

U1RENL—UART 1 Receive Interrupt Request Enable Low Bit

U1TENL—UART 1 Transmit Interrupt Request Enable Low Bit

DMAENL—DMA Interrupt Request Enable Low Bit

C3ENL—Port C3 Interrupt Request Enable Low Bit

C2ENL—Port C2 Interrupt Request Enable Low Bit



C1ENL—Port C1 Interrupt Request Enable Low Bit C0ENL—Port C0 Interrupt Request Enable Low Bit

# Interrupt Edge Select Register

The Interrupt Edge Select (IRQES) register (Table 36) determines whether an interrupt is generated for the rising edge or falling edge on the selected GPIO Port input pin. The Interrupt Port Select register selects between Port A and Port D for the individual interrupts.

Table 36. Interrupt Edge Select Register (IRQES)

BITS	7	6	5	4	3	2	1	0		
FIELD	IES7	IES7 IES6 IES5 IES4 IES3 IES2 IES1 IES0								
RESET		0								
R/W				R/	W					
ADDR				FC	DH					

IES*x*—Interrupt Edge Select *x* 

The minimum pulse width should be greater than 1 system clock to guarantee capture of the edge triggered interrupt. Shorter pulses may be captured but not guaranteed.

0 = An interrupt request is generated on the falling edge of the PAx/PDx input.

1 = An interrupt request is generated on the rising edge of the PAx/PDx input. where x indicates the specific GPIO Port pin number (0 through 7).

### **Interrupt Port Select Register**

The Port Select (IRQPS) register (Table 37) determines the port pin that generates the PAx/PDx interrupts. This register allows either Port A or Port D pins to be used as interrupts. The Interrupt Edge Select register controls the active interrupt edge.

Table 37. Interrupt Port Select Register (IRQPS)

BITS	7	6	5	4	3	2	1	0		
FIELD	PAD7S	PAD7S PAD6S PAD5S PAD4S PAD3S PAD2S PAD1S PAD0S								
RESET		0								
R/W				R/	W					
ADDR				FC	EH					

PADxS—PAx/PDx Selection

0 = PAx is used for the interrupt for PAx/PDx interrupt request.

1 = PDx is used for the interrupt for PAx/PDx interrupt request. where *x* indicates the specific GPIO Port pin number (0 through 7).

# **Interrupt Control Register**

The Interrupt Control (IRQCTL) register (Table 38) contains the master enable bit for all interrupts.

**Table 38. Interrupt Control Register (IRQCTL)** 

вітѕ	7	6	5	4	3	2	1	0
FIELD	IRQE		Reserved					
RESET		0						
R/W	R/W		R					
ADDR				FC	FH			

IRQE—Interrupt Request Enable

This bit is set to 1 by execution of an EI or IRET instruction, or by a direct register write of a 1 to this bit. It is reset to 0 by executing a DI instruction, eZ8 CPU acknowledgement of an interrupt request, or Reset.

0 = Interrupts are disabled

1 = Interrupts are enabled

Reserved—Must be 0.

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# **Timers**

#### Overview

The 64K Series products contain up to four 16-bit reloadable timers that can be used for timing, event counting, or generation of pulse width modulated signals. The timers' features include:

- 16-bit reload counter
- Programmable prescaler with prescale values from 1 to 128
- PWM output generation
- Capture and compare capability
- External input pin for timer input, clock gating, or capture signal. External input pin signal frequency is limited to a maximum of one-fourth the system clock frequency.
- Timer output pin
- Timer interrupt

In addition to the timers described in this chapter, the Baud Rate Generators for any unused UART, SPI, or I<sup>2</sup>C peripherals may also be used to provide basic timing functionality. For information on using the Baud Rate Generators as timers, see the respective serial communication peripheral. Timer 3 is unavailable in the 44-pin package devices.

#### **Architecture**

Figure 12 displays the architecture of the timers.

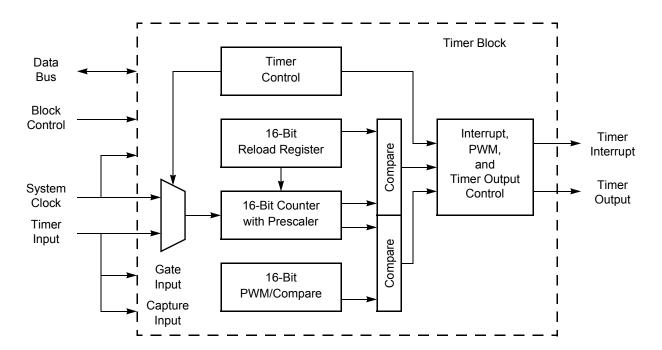


Figure 12. Timer Block Diagram

# Operation

The timers are 16-bit up-counters. Minimum time-out delay is set by loading the value 0001H into the Timer Reload High and Low Byte registers and setting the prescale value to 1. Maximum time-out delay is set by loading the value 0000H into the Timer Reload High and Low Byte registers and setting the prescale value to 128. If the Timer reaches FFFFH, the timer rolls over to 0000H and continues counting.

# **Timer Operating Modes**

The timers can be configured to operate in the following modes:

#### **ONE-SHOT Mode**

In ONE-SHOT mode, the timer counts up to the 16-bit Reload value stored in the Timer Reload High and Low Byte registers. The timer input is the system clock. Upon reaching the Reload value, the timer generates an interrupt and the count value in the Timer High and Low Byte registers is reset to 0001H. Then, the timer is automatically disabled and stops counting.

Also, if the Timer Output alternate function is enabled, the Timer Output pin changes state for one system clock cycle (from Low to High or from High to Low) upon timer Reload. If it is desired to have the Timer Output make a permanent state change upon

One-Shot time-out, first set the TPOL bit in the Timer Control 1 Register to the start value before beginning ONE-SHOT mode. Then, after starting the timer, set TPOL to the opposite bit value.

Follow the steps below for configuring a timer for ONE-SHOT mode and initiating the count:

- 1. Write to the Timer Control 1 register to:
  - Disable the timer
  - Configure the timer for ONE-SHOT mode
  - Set the prescale value
  - If using the Timer Output alternate function, set the initial output level (High or Low)
- 2. Write to the Timer High and Low Byte registers to set the starting count value
- 3. Write to the Timer Reload High and Low Byte registers to set the Reload value
- 4. If desired, enable the timer interrupt and set the timer interrupt priority by writing to the relevant interrupt registers
- 5. If using the Timer Output function, configure the associated GPIO port pin for the Timer Output alternate function
- 6. Write to the Timer Control 1 register to enable the timer and initiate counting

In ONE-SHOT mode, the system clock always provides the timer input. The timer period is given by the following equation:

ONE-SHOT Mode Time-Out Period (s) =  $\frac{(Reload\ Value - Start\ Value) \times Prescale}{System\ Clock\ Frequency\ (Hz)}$ 

#### **CONTINUOUS Mode**

In CONTINUOUS mode, the timer counts up to the 16-bit Reload value stored in the Timer Reload High and Low Byte registers. The timer input is the system clock. Upon reaching the Reload value, the timer generates an interrupt, the count value in the Timer High and Low Byte registers is reset to 0001H and counting resumes. Also, if the Timer Output alternate function is enabled, the Timer Output pin changes state (from Low to High or from High to Low) upon timer Reload.

Follow the steps below for configuring a timer for CONTINUOUS mode and initiating the count:

- 1. Write to the Timer Control 1 register to:
  - Disable the timer
  - Configure the timer for CONTINUOUS mode
  - Set the prescale value
  - If using the Timer Output alternate function, set the initial output level (High or Low)

- 2. Write to the Timer High and Low Byte registers to set the starting count value (usually 0001H), affecting only the first pass in CONTINUOUS mode. After the first timer Reload in CONTINUOUS mode, counting always begins at the reset value of 0001H.
- 3. Write to the Timer Reload High and Low Byte registers to set the Reload value.
- 4. If desired, enable the timer interrupt and set the timer interrupt priority by writing to the relevant interrupt registers.
- 5. If using the Timer Output function, configure the associated GPIO port pin for the Timer Output alternate function.
- 6. Write to the Timer Control 1 register to enable the timer and initiate counting.

In CONTINUOUS mode, the system clock always provides the timer input. The timer period is given by the following equation:

CONTINUOUS Mode Time-Out Period (s) = 
$$\frac{\text{Reload Value} \times \text{Prescale}}{\text{System Clock Frequency (Hz)}}$$

If an initial starting value other than 0001H is loaded into the Timer High and Low Byte registers, the ONE-SHOT mode equation must be used to determine the first time-out period.

#### **COUNTER Mode**

In COUNTER mode, the timer counts input transitions from a GPIO port pin. The timer input is taken from the GPIO Port pin Timer Input alternate function. The TPOL bit in the Timer Control 1 Register selects whether the count occurs on the rising edge or the falling edge of the Timer Input signal. In COUNTER mode, the prescaler is disabled.



**Caution:** The input frequency of the Timer Input signal must not exceed one-fourth the system clock frequency.

Upon reaching the Reload value stored in the Timer Reload High and Low Byte registers, the timer generates an interrupt, the count value in the Timer High and Low Byte registers is reset to 0001H and counting resumes. Also, if the Timer Output alternate function is enabled, the Timer Output pin changes state (from Low to High or from High to Low) at timer Reload.

Follow the steps below for configuring a timer for COUNTER mode and initiating the count:

- 1. Write to the Timer Control 1 register to:
  - Disable the timer
  - Configure the timer for COUNTER mode

- Q
- Select either the rising edge or falling edge of the Timer Input signal for the count.
   This also sets the initial logic level (High or Low) for the Timer Output alternate function. However, the Timer Output function does not have to be enabled
- 2. Write to the Timer High and Low Byte registers to set the starting count value. This only affects the first pass in COUNTER mode. After the first timer Reload in COUNTER mode, counting always begins at the reset value of 0001H. Generally, in COUNTER mode the Timer High and Low Byte registers must be written with the value 0001H.
- 3. Write to the Timer Reload High and Low Byte registers to set the Reload value.
- 4. If desired, enable the timer interrupt and set the timer interrupt priority by writing to the relevant interrupt registers.
- 5. Configure the associated GPIO port pin for the Timer Input alternate function.
- 6. If using the Timer Output function, configure the associated GPIO port pin for the Timer Output alternate function.
- 7. Write to the Timer Control 1 register to enable the timer.

In COUNTER mode, the number of Timer Input transitions since the timer start is given by the following equation:

COUNTER Mode Timer Input Transitions = Current Count Value – Start Value

#### **PWM Mode**

In PWM mode, the timer outputs a Pulse-Width Modulator (PWM) output signal through a GPIO Port pin. The timer input is the system clock. The timer first counts up to the 16-bit PWM match value stored in the Timer PWM High and Low Byte registers. When the timer count value matches the PWM value, the Timer Output toggles. The timer continues counting until it reaches the Reload value stored in the Timer Reload High and Low Byte registers. Upon reaching the Reload value, the timer generates an interrupt, the count value in the Timer High and Low Byte registers is reset to 0001H and counting resumes.

If the TPOL bit in the Timer Control 1 register is set to 1, the Timer Output signal begins as a High (1) and then transitions to a Low (0) when the timer value matches the PWM value. The Timer Output signal returns to a High (1) after the timer reaches the Reload value and is reset to 0001H.

If the TPOL bit in the Timer Control 1 register is set to 0, the Timer Output signal begins as a Low (0) and then transitions to a High (1) when the timer value matches the PWM value. The Timer Output signal returns to a Low (0) after the timer reaches the Reload value and is reset to 0001H.

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Follow the steps below for configuring a timer for PWM mode and initiating the PWM operation:

- 1. Write to the Timer Control 1 register to:
  - Disable the timer
  - Configure the timer for PWM mode
  - Set the prescale value
  - Set the initial logic level (High or Low) and PWM High/Low transition for the Timer Output alternate function
- 2. Write to the Timer High and Low Byte registers to set the starting count value (typically 0001H). This only affects the first pass in PWM mode. After the first timer reset in PWM mode, counting always begins at the reset value of 0001H.
- 3. Write to the PWM High and Low Byte registers to set the PWM value.
- 4. Write to the Timer Reload High and Low Byte registers to set the Reload value (PWM period). The Reload value must be greater than the PWM value.
- 5. If desired, enable the timer interrupt and set the timer interrupt priority by writing to the relevant interrupt registers.
- 6. Configure the associated GPIO port pin for the Timer Output alternate function.
- 7. Write to the Timer Control 1 register to enable the timer and initiate counting.

The PWM period is given by the following equation:

PWM Period (s) = 
$$\frac{\text{Reload Value} \times \text{Prescale}}{\text{System Clock Frequency (Hz)}}$$

If an initial starting value other than 0001H is loaded into the Timer High and Low Byte registers, the ONE-SHOT mode equation must be used to determine the first PWM timeout period.

If TPOL is set to 0, the ratio of the PWM output High time to the total period is given by:

PWM Output High Time Ratio (%) = 
$$\frac{\text{Reload Value} - \text{PWM Value}}{\text{Reload Value}} \times 100$$

If TPOL is set to 1, the ratio of the PWM output High time to the total period is given by:

PWM Output High Time Ratio (%) = 
$$\frac{\text{PWM Value}}{\text{Reload Value}} \times 100$$

#### **CAPTURE Mode**

In CAPTURE mode, the current timer count value is recorded when the desired external Timer Input transition occurs. The Capture count value is written to the Timer PWM High and Low Byte Registers. The timer input is the system clock. The TPOL bit in the Timer Control 1 register determines if the Capture occurs on a rising edge or a falling edge of the Timer Input signal. When the Capture event occurs, an interrupt is generated and the timer continues counting.

The timer continues counting up to the 16-bit Reload value stored in the Timer Reload High and Low Byte registers. Upon reaching the Reload value, the timer generates an interrupt and continues counting.

Follow the steps below for configuring a timer for CAPTURE mode and initiating the count:

- 1. Write to the Timer Control 1 register to:
  - Disable the timer
  - Configure the timer for CAPTURE mode.
  - Set the prescale value.
  - Set the Capture edge (rising or falling) for the Timer Input.
- 2. Write to the Timer High and Low Byte registers to set the starting count value (typically 0001H).
- 3. Write to the Timer Reload High and Low Byte registers to set the Reload value.
- 4. Clear the Timer PWM High and Low Byte registers to 0000H. This allows the software to determine if interrupts were generated by either a capture event or a reload. If the PWM High and Low Byte registers still contain 0000H after the interrupt, then the interrupt was generated by a Reload.
- 5. If desired, enable the timer interrupt and set the timer interrupt priority by writing to the relevant interrupt registers.
- 6. Configure the associated GPIO port pin for the Timer Input alternate function.
- 7. Write to the Timer Control 1 register to enable the timer and initiate counting.

In CAPTURE mode, the elapsed time from timer start to Capture event can be calculated using the following equation:

Capture Elapsed Time (s) = 
$$\frac{\text{(Capture Value - Start Value)} \times \text{Prescale}}{\text{System Clock Frequency (Hz)}}$$

#### **COMPARE Mode**

In COMPARE mode, the timer counts up to the 16-bit maximum Compare value stored in the Timer Reload High and Low Byte registers. The timer input is the system clock. Upon reaching the Compare value, the timer generates an interrupt and counting continues (the timer value is not reset to 0001H). Also, if the Timer Output alternate function is enabled, the Timer Output pin changes state (from Low to High or from High to Low) upon Compare.

If the Timer reaches FFFFH, the timer rolls over to 0000H and continue counting.

Follow the steps below for configuring a timer for COMPARE mode and initiating the count:

- 1. Write to the Timer Control 1 register to:
  - Disable the timer
  - Configure the timer for COMPARE mode
  - Set the prescale value
  - Set the initial logic level (High or Low) for the Timer Output alternate function, if desired
- 2. Write to the Timer High and Low Byte registers to set the starting count value.
- 3. Write to the Timer Reload High and Low Byte registers to set the Compare value.
- 4. If desired, enable the timer interrupt and set the timer interrupt priority by writing to the relevant interrupt registers.
- 5. If using the Timer Output function, configure the associated GPIO port pin for the Timer Output alternate function.
- 6. Write to the Timer Control 1 register to enable the timer and initiate counting.

In COMPARE mode, the system clock always provides the timer input. The Compare time is given by the following equation:

COMPARE Mode Time (s) = 
$$\frac{\text{(Compare Value - Start Value)} \times \text{Prescale}}{\text{System Clock Frequency (Hz)}}$$

#### **GATED Mode**

In GATED mode, the timer counts only when the Timer Input signal is in its active state (asserted), as determined by the TPOL bit in the Timer Control 1 register. When the Timer Input signal is asserted, counting begins. A timer interrupt is generated when the Timer Input signal is deasserted or a timer reload occurs. To determine if a Timer Input signal deassertion generated the interrupt, read the associated GPIO input value and compare to the value stored in the TPOL bit.

The timer counts up to the 16-bit Reload value stored in the Timer Reload High and Low Byte registers. The timer input is the system clock. When reaching the Reload value, the timer generates an interrupt, the count value in the Timer High and Low Byte registers is reset to 0001H and counting resumes (assuming the Timer Input signal is still asserted). Also, if the Timer Output alternate function is enabled, the Timer Output pin changes state (from Low to High or from High to Low) at timer reset.

Follow the steps below for configuring a timer for GATED mode and initiating the count:

- 1. Write to the Timer Control 1 register to:
  - Disable the timer
  - Configure the timer for GATED mode

- Set the prescale value
- 2. Write to the Timer High and Low Byte registers to set the starting count value. This only affects the first pass in GATED mode. After the first timer reset in GATED mode, counting always begins at the reset value of 0001H.
- 3. Write to the Timer Reload High and Low Byte registers to set the Reload value.
- 4. If desired, enable the timer interrupt and set the timer interrupt priority by writing to the relevant interrupt registers.
- 5. Configure the associated GPIO port pin for the Timer Input alternate function.
- 6. Write to the Timer Control 1 register to enable the timer.
- 7. Assert the Timer Input signal to initiate the counting.

#### **CAPTURE/COMPARE Mode**

In CAPTURE/COMPARE mode, the timer begins counting on the *first* external Timer Input transition. The desired transition (rising edge or falling edge) is set by the TPOL bit in the Timer Control 1 Register. The timer input is the system clock.

Every subsequent desired transition (after the first) of the Timer Input signal captures the current count value. The Capture value is written to the Timer PWM High and Low Byte Registers. When the Capture event occurs, an interrupt is generated, the count value in the Timer High and Low Byte registers is reset to 0001H, and counting resumes.

If no Capture event occurs, the timer counts up to the 16-bit Compare value stored in the Timer Reload High and Low Byte registers. Upon reaching the Compare value, the timer generates an interrupt, the count value in the Timer High and Low Byte registers is reset to 0001H and counting resumes.

Follow the steps below for configuring a timer for CAPTURE/COMPARE mode and initiating the count:

- 1. Write to the Timer Control 1 register to:
  - Disable the timer
  - Configure the timer for CAPTURE/COMPARE mode
  - Set the prescale value
  - Set the Capture edge (rising or falling) for the Timer Input
- 2. Write to the Timer High and Low Byte registers to set the starting count value (typically 0001H).
- 3. Write to the Timer Reload High and Low Byte registers to set the Compare value.
- 4. If desired, enable the timer interrupt and set the timer interrupt priority by writing to the relevant interrupt registers.
- 5. Configure the associated GPIO port pin for the Timer Input alternate function.

- 6. Write to the Timer Control 1 register to enable the timer.
- 7. Counting begins on the first appropriate transition of the Timer Input signal. No interrupt is generated by this first edge.

In m/COMPARE mode, the elapsed time from timer start to Capture event can be calculated using the following equation:

Capture Elapsed Time (s) =  $\frac{\text{(Capture Value - Start Value)} \times \text{Prescale}}{\text{System Clock Frequency (Hz)}}$ 

# **Reading the Timer Count Values**

The current count value in the timers can be read while counting (enabled). This capability has no effect on timer operation. When the timer is enabled and the Timer High Byte register is read, the contents of the Timer Low Byte register are placed in a holding register. A subsequent read from the Timer Low Byte register returns the value in the holding register. This operation allows accurate reads of the full 16-bit timer count value while enabled. When the timers are not enabled, a read from the Timer Low Byte register returns the actual value in the counter.

# **Timer Output Signal Operation**

Timer Output is a GPIO Port pin alternate function. Generally, the Timer Output is toggled every time the counter is reloaded.

# **Timer Control Register Definitions**

Timers 0-2 are available in all packages. Timer 3 is only available in the 64-, 68-, and 80-pin packages.

# Timer 0-3 High and Low Byte Registers

The Timer 0-3 High and Low Byte (TxH and TxL) registers (see Table 39 and Table 40 on page 91) contain the current 16-bit timer count value. When the timer is enabled, a read from TxH causes the value in TxL to be stored in a temporary holding register. A read from TMRL always returns this temporary register when the timers are enabled. When the timer is disabled, reads from the TMRL reads the register directly.

Writing to the Timer High and Low Byte registers while the timer is enabled is not recommended. There are no temporary holding registers available for write operations, so simultaneous 16-bit writes are not possible. If either the Timer High or Low Byte registers are written during counting, the 8-bit written value is placed in the counter (High or Low Byte) at the next clock edge. The counter continues counting from the new value.

Timer 3 is unavailable in the 40- and 44-pin packages.

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Table 39. Timer 0-3 High Byte Register (TxH)

BITS	7	6	5	4	3	2	1	0		
FIELD		TH								
RESET		0								
R/W				R/	W					
ADDR			F	F00H, F08H,	F10H, F18H	1				

Table 40. Timer 0-3 Low Byte Register (TxL)

BITS	7	6	5	4	3	2	1	0		
FIELD		TL								
RESET	0									
R/W		R/W								
ADDR			F	F01H, F09H,	F11H, F19H	1				

TH and TL—Timer High and Low Bytes

These 2 bytes, {TMRH[7:0], TMRL[7:0]}, contain the current 16-bit timer count value.

# Timer Reload High and Low Byte Registers

The Timer 0-3 Reload High and Low Byte (TxRH and TxRL) registers (see Table 41 and Table 42 on page 92) store a 16-bit reload value, {TRH[7:0], TRL[7:0]}. Values written to the Timer Reload High Byte register are stored in a temporary holding register. When a write to the Timer Reload Low Byte register occurs, the temporary holding register value is written to the Timer High Byte register. This operation allows simultaneous updates of the 16-bit Timer Reload value.

In COMPARE mode, the Timer Reload High and Low Byte registers store the 16-bit Compare value.

Table 41. Timer 0-3 Reload High Byte Register (TxRH)

BITS	7	6	5	4	3	2	1	0		
FIELD		TRH								
RESET		1								
R/W				R/	W					
ADDR			F	02H, F0AH,	F12H, F1A	Н				

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Table 42. Timer 0-3 Reload Low Byte Register (TxRL)

BITS	7	6	5	4	3	2	1	0		
FIELD		TRL								
RESET		1								
R/W				R/	W					
ADDR			F	03H, F0BH,	F13H, F1BI	Н				

TRH and TRL—Timer Reload Register High and Low

These two bytes form the 16-bit Reload value, {TRH[7:0], TRL[7:0]}. This value sets the maximum count value which initiates a timer reload to 0001H. In COMPARE mode, these two byte form the 16-bit Compare value.

# Timer 0-3 PWM High and Low Byte Registers

The Timer 0-3 PWM High and Low Byte (TxPWMH and TxPWML) registers (see Table 43 and Table 44 on page 92) are used for Pulse-Width Modulator (PWM) operations. These registers also store the Capture values for the Capture and Capture/COMPARE modes.

Table 43. Timer 0-3 PWM High Byte Register (TxPWMH)

BITS	7	6	5	4	3	2	1	0	
FIELD		PWMH							
RESET		0							
R/W				R/	W				
ADDR			F	04H, F0CH,	F14H, F1C	Н			

# Table 44. Timer 0-3 PWM Low Byte Register (TxPWML)

BITS	7	6	5	4	3	2	1	0		
FIELD		PWML								
RESET		0								
R/W				R/	W					
ADDR			F	05H, F0DH,	F15H, F1D	Н				

PWMH and PWML—Pulse-Width Modulator High and Low Bytes These two bytes, {PWMH[7:0], PWML[7:0]}, form a 16-bit value that is compared to the current 16-bit timer count. When a match occurs, the PWM output changes state. The PWM output value is set by the TPOL bit in the Timer Control 1 Register (TxCTL1) register.

The TxPWMH and TxPWML registers also store the 16-bit captured timer value when operating in CAPTURE or CAPTURE/COMPARE modes.

# **Timer 0-3 Control 0 Registers**

The Timer 0-3 Control 0 (TxCTL0) registers (see Table 45 and Table 46) allow cascading of the Timers.

Table 45. Timer 0-3 Control 0 Register (TxCTL0)

BITS	7	6	5	4	3	2	1	0	
FIELD	Reserved			CSC	Reserved				
RESET	0								
R/W	R/W								
ADDR	F06H, F0EH, F16H, F1EH								

### CSC—Cascade Timers

0 = Timer Input signal comes from the pin.

1 = For Timer 0, Input signal is connected to Timer 3 output. For Timer 1, Input signal is connected to Timer 0 output. For Timer 2, Input signal is connected to Timer 1 output. For Timer 3, Input signal is connected to Timer 2 output.

# **Timer 0-3 Control 1 Registers**

The Timer 0-3 Control 1 (TxCTL1) registers enable/disable the timers, set the prescaler value, and determine the timer operating mode.

Table 46. Timer 0-3 Control 1 Register (TxCTL1)

BITS	7	6	5	4	3	2	1	0	
FIELD	TEN	TPOL	PRES			TMODE			
RESET	0								
R/W	R/W								
ADDR	F07H, F0FH, F17H, F1FH								

TEN—Timer Enable

0 = Timer is disabled.

1 = Timer enabled to count.

TPOL—Timer Input/Output Polarity

Operation of this bit is a function of the current operating mode of the timer.

#### **ONE-SHOT** mode

When the timer is disabled, the Timer Output signal is set to the value of this bit. When the timer is enabled, the Timer Output signal is complemented upon timer Reload.

#### **CONTINUOUS** mode

When the timer is disabled, the Timer Output signal is set to the value of this bit. When the timer is enabled, the Timer Output signal is complemented upon timer Reload.

## **COUNTER** mode

When the timer is disabled, the Timer Output signal is set to the value of this bit. When the timer is enabled, the Timer Output signal is complemented upon timer Reload.

0 = Count occurs on the rising edge of the Timer Input signal.

1 = Count occurs on the falling edge of the Timer Input signal.

#### **PWM** mode

0 = Timer Output is forced Low (0) when the timer is disabled. When enabled, the Timer Output is forced High (1) upon PWM count match and forced Low (0) upon Reload.

1 = Timer Output is forced High (1) when the timer is disabled. When enabled, the Timer Output is forced Low (0) upon PWM count match and forced High (1) upon Reload.

#### **CAPTURE** mode

- 0 = Count is captured on the rising edge of the Timer Input signal.
- 1 = Count is captured on the falling edge of the Timer Input signal.

#### **COMPARE** mode

When the timer is disabled, the Timer Output signal is set to the value of this bit. When the timer is enabled, the Timer Output signal is complemented upon timer Reload.

#### **GATED** mode

- 0 = Timer counts when the Timer Input signal is High (1) and interrupts are generated on the falling edge of the Timer Input.
- 1 = Timer counts when the Timer Input signal is Low (0) and interrupts are generated on the rising edge of the Timer Input.

#### **CAPTURE/COMPARE** mode

- 0 = Counting is started on the first rising edge of the Timer Input signal. The current count is captured on subsequent rising edges of the Timer Input signal.
- 1 = Counting is started on the first falling edge of the Timer Input signal. The current count is captured on subsequent falling edges of the Timer Input signal.



Caution: When the Timer Output alternate function TxOUT on a GPIO port pin is enabled, TxOUT will change to whatever state the TPOL bit is in. The timer does not need to be enabled for that to happen. Also, the Port data direction sub register is not needed to be set to output on TxOUT. Changing the TPOL bit with the timer enabled and running does not immediately change the TxOUT.

#### PRES—Prescale value.

The timer input clock is divided by 2<sup>PRES</sup>, where PRES can be set from 0 to 7. The prescaler is reset each time the Timer is disabled. This insures proper clock division each time the Timer is restarted.

000 = Divide by 1

001 = Divide by 2

010 = Divide by 4

011 = Divide by 8

100 = Divide by 16

101 = Divide by 32

PS019919-1207 **Timers**  110 = Divide by 64

111 = Divide by 128

TMODE—TIMER mode

000 = ONE-SHOT mode

001 = CONTINUOUS mode

010 = COUNTER mode

 $011 = PWM \mod e$ 

100 = CAPTURE mode

101 = COMPARE mode

 $110 = GATED \mod e$ 

111 = CAPTURE/COMPARE mode

PS019919-1207 Timers

# Watchdog Timer

#### Overview

The Watchdog Timer (WDT) helps protect against corrupt or unreliable software, power faults, and other system-level problems which may place the Z8 Encore! XP into unsuitable operating states. The features of Watchdog Timer include:

- On-chip RC oscillator.
- A selectable time-out response.
- WDT Time-out response: Reset or interrupt.
- 24-bit programmable time-out value.

# **Operation**

The Watchdog Timer (WDT) is a retriggerable one-shot timer that resets or interrupts the 64K Series devices when the WDT reaches its terminal count. The Watchdog Timer uses its own dedicated on-chip RC oscillator as its clock source. The Watchdog Timer has only two modes of operation—ON and OFF. Once enabled, it always counts and must be refreshed to prevent a time-out. An enable can be performed by executing the WDT instruction or by setting the WDT\_AO Option Bit. The WDT\_AO bit enables the Watchdog Timer to operate all the time, even if a WDT instruction has not been executed.

The Watchdog Timer is a 24-bit reloadable downcounter that uses three 8-bit registers in the eZ8<sup>TM</sup> CPU register space to set the reload value. The nominal WDT time-out period is given by the following equation:

WDT Time-out Period (ms) = 
$$\frac{\text{WDT Reload Value}}{10}$$

where the WDT reload value is the decimal value of the 24-bit value given by {WDTU[7:0], WDTH[7:0], WDTL[7:0]} and the typical Watchdog Timer RC oscillator frequency is 10 kHz. The Watchdog Timer cannot be refreshed once it reaches 000002H. The WDT Reload Value must not be set to values below 000004H. Table 47 provides information on approximate time-out delays for the minimum and maximum WDT reload values.

Table 47. Watchdog Timer Approximate Time-Out Delays

WDT Reload Value	WDT Reload Value		e Time-Out Delay WDT oscillator frequency)
(Hex)	(Decimal)	Typical	Description
000004	4	400 μs	Minimum time-out delay
FFFFF	16,777,215	1677.5 s	Maximum time-out delay

## Watchdog Timer Refresh

When first enabled, the Watchdog Timer is loaded with the value in the Watchdog Timer Reload registers. The Watchdog Timer then counts down to 000000H unless a WDT instruction is executed by the eZ8<sup>TM</sup> CPU. Execution of the WDT instruction causes the downcounter to be reloaded with the WDT Reload value stored in the Watchdog Timer Reload registers. Counting resumes following the reload operation.

When the 64K Series devices are operating in DEBUG Mode (through the On-Chip Debugger), the Watchdog Timer is continuously refreshed to prevent spurious Watchdog Timer time-outs.

## **Watchdog Timer Time-Out Response**

The Watchdog Timer times out when the counter reaches 000000H. A time-out of the Watchdog Timer generates either an interrupt or a Reset. The WDT\_RES Option Bit determines the time-out response of the Watchdog Timer. For information on programming of the WDT\_RES Option Bit, see Option Bits on page 195.

#### **WDT Interrupt in Normal Operation**

If configured to generate an interrupt when a time-out occurs, the Watchdog Timer issues an interrupt request to the interrupt controller and sets the WDT status bit in the Watchdog Timer Control register. If interrupts are enabled, the eZ8 CPU responds to the interrupt request by fetching the Watchdog Timer interrupt vector and executing code from the vector address. After time-out and interrupt generation, the Watchdog Timer counter rolls over to its maximum value of FFFFFH and continues counting. The Watchdog Timer counter is not automatically returned to its Reload Value.

#### **WDT Interrupt in STOP Mode**

If configured to generate an interrupt when a time-out occurs and the 64K Series devices are in STOP mode, the Watchdog Timer automatically initiates a Stop Mode Recovery and generates an interrupt request. Both the WDT status bit and the STOP bit in the Watchdog Timer Control register are set to 1 following WDT time-out in STOP mode. For more information on Stop Mode Recovery, see Reset and Stop Mode Recovery on page 47.

If interrupts are enabled, following completion of the Stop Mode Recovery the eZ8 CPU responds to the interrupt request by fetching the Watchdog Timer interrupt vector and executing code from the vector address.

#### **WDT Reset in Normal Operation**

If configured to generate a Reset when a time-out occurs, the Watchdog Timer forces the device into the Reset state. The WDT status bit in the Watchdog Timer Control register is set to 1. For more information on Reset, see Reset and Stop Mode Recovery on page 47.

#### WDT Reset in STOP Mode

If enabled in STOP mode and configured to generate a Reset when a time-out occurs and the device is in STOP mode, the Watchdog Timer initiates a Stop Mode Recovery. Both the WDT status bit and the STOP bit in the Watchdog Timer Control register are set to 1 following WDT time-out in STOP mode. Default operation is for the WDT and its RC oscillator to be enabled during STOP mode.

#### WDT RC Disable in STOP Mode

To minimize power consumption in STOP Mode, the WDT and its RC oscillator can be disabled in STOP mode. The following sequence configures the WDT to be disabled when the 64K Series devices enter STOP Mode following execution of a STOP instruction:

- 1. Write 55H to the Watchdog Timer Control register (WDTCTL).
- 2. Write AAH to the Watchdog Timer Control register (WDTCTL).
- 3. Write 81H to the Watchdog Timer Control register (WDTCTL) to configure the WDT and its oscillator to be disabled during STOP Mode. Alternatively, write 00H to the Watchdog Timer Control register (WDTCTL) as the third step in this sequence to reconfigure the WDT and its oscillator to be enabled during STOP mode.

This sequence only affects WDT operation in STOP mode.

## Watchdog Timer Reload Unlock Sequence

Writing the unlock sequence to the Watchdog Timer (WDTCTL) Control register address unlocks the three Watchdog Timer Reload Byte registers (WDTU, WDTH, and WDTL) to allow changes to the time-out period. These write operations to the WDTCTL register address produce no effect on the bits in the WDTCTL register. The locking mechanism prevents spurious writes to the Reload registers. Follow the steps below to unlock the Watchdog Timer Reload Byte registers (WDTU, WDTH, and WDTL) for write access.

- 1. Write 55H to the Watchdog Timer Control register (WDTCTL).
- 2. Write AAH to the Watchdog Timer Control register (WDTCTL).
- 3. Write the Watchdog Timer Reload Upper Byte register (WDTU).
- 4. Write the Watchdog Timer Reload High Byte register (WDTH).

### 5. Write the Watchdog Timer Reload Low Byte register (WDTL).

All steps of the Watchdog Timer Reload Unlock sequence must be written in the order just listed. There must be no other register writes between each of these operations. If a register write occurs, the lock state machine resets and no further writes can occur, unless the sequence is restarted. The value in the Watchdog Timer Reload registers is loaded into the counter when the Watchdog Timer is first enabled and every time a WDT instruction is executed.

## **Watchdog Timer Control Register Definitions**

## **Watchdog Timer Control Register**

The Watchdog Timer Control (WDTCTL) register (Table 48) is a Read-Only register that indicates the source of the most recent Reset event, indicates a Stop Mode Recovery event, and indicates a Watchdog Timer time-out. Reading this register resets the upper four bits to 0.

Writing the 55H, AAH unlock sequence to the Watchdog Timer Control (WDTCTL) register address unlocks the three Watchdog Timer Reload Byte registers (WDTU, WDTH, and WDTL) to allow changes to the time-out period. These write operations to the WDTCTL register address produce no effect on the bits in the WDTCTL register. The locking mechanism prevents spurious writes to the Reload registers.

Table 48. Watchdog Timer Control Register (WDTCTL)

BITS	7	6	5	4	3	3 2 1				
FIELD	POR	STOP	WDT	EXT		SM				
RESET	See d	lescriptions	below	0						
R/W				F	₹					
ADDR				FF	0H					

Reset or Stop Mode Recovery Event	POR	STOP	WDT	EXT
Power-On Reset	1	0	0	0
Reset using RESET pin assertion	0	0	0	1
Reset using Watchdog Timer time-out	0	0	1	0
Reset using the On-Chip Debugger (OCDCTL[1] set to 1)	1	0	0	0
Reset from STOP Mode using DBG Pin driven Low	1	0	0	0
Stop Mode Recovery using GPIO pin transition	0	1	0	0
Stop Mode Recovery using Watchdog Timer time-out	0	1	1	0

#### POR—Power-On Reset Indicator

If this bit is set to 1, a Power-On Reset event occurred. This bit is reset to 0 if a WDT timeout or Stop Mode Recovery occurs. This bit is also reset to 0 when the register is read.

#### STOP—Stop Mode Recovery Indicator

If this bit is set to 1, a Stop Mode Recovery occurred. If the STOP and WDT bits are both set to 1, the Stop Mode Recovery occurred due to a WDT time-out. If the STOP bit is 1 and the WDT bit is 0, the Stop Mode Recovery was not caused by a WDT time-out. This bit is reset by a Power-On Reset or a WDT time-out that occurred while not in STOP mode. Reading this register also resets this bit.

#### WDT—Watchdog Timer Time-Out Indicator

If this bit is set to 1, a WDT time-out occurred. A Power-On Reset resets this pin. A Stop Mode Recovery from a change in an input pin also resets this bit. Reading this register resets this bit.

#### EXT—External Reset Indicator

If this bit is set to 1, a Reset initiated by the external  $\overline{RESET}$  pin occurred. A Power-On Reset or a Stop Mode Recovery from a change in an input pin resets this bit. Reading this register resets this bit.

#### Reserved

These bits are reserved and must be 0.

#### SM—STOP Mode Configuration Indicator

- 0 = Watchdog Timer and its internal RC oscillator will continue to operate in STOP Mode.
- 1 = Watchdog Timer and its internal RC oscillator will be disabled in STOP Mode.

## Watchdog Timer Reload Upper, High and Low Byte Registers

The Watchdog Timer Reload Upper, High and Low Byte (WDTU, WDTH, WDTL) registers (see Table 49 on page 102 through Table 51 on page 102) form the 24-bit reload value that is loaded into the Watchdog Timer when a WDT instruction executes. The 24-bit reload value is {WDTU[7:0], WDTH[7:0], WDTL[7:0]}. Writing to these registers sets the desired Reload Value. Reading from these registers returns the current Watchdog Timer count value.



**Caution:** The 24-bit WDT Reload Value must not be set to a value less than 000004H.

## Table 49. Watchdog Timer Reload Upper Byte Register (WDTU)

BITS	7	6	5	4	3	2	1	0		
FIELD				WE	TU					
RESET				•	1					
R/W				R/	W*					
ADDR		FF1H								
Note: R/W	/* - Read returns the current WDT count value. Write sets the desired Reload Value.									

WDTU—WDT Reload Upper Byte

Most significant byte, Bits[23:16], of the 24-bit WDT reload value.

## Table 50. Watchdog Timer Reload High Byte Register (WDTH)

BITS	7	6	5	4	3	2	1	0	
FIELD				WE	TH				
RESET				•	I				
R/W				R/	W*				
ADDR		FF2H							
Note: R/W	/* - Read returns the current WDT count value. Write sets the desired Reload Value.								

WDTH—WDT Reload High Byte

Middle byte, Bits[15:8], of the 24-bit WDT reload value.

## Table 51. Watchdog Timer Reload Low Byte Register (WDTL)

BITS	7	6	5	4	3	2	1	0		
FIELD				WE	TL					
RESET				•	1					
R/W				R/	W*					
ADDR		FF3H								
Note: R/W	V* - Read returns the current WDT count value. Write sets the desired Reload Value.									

WDTL—WDT Reload Low

Least significant byte, Bits[7:0], of the 24-bit WDT reload value.

# **UART**

#### Overview

The Universal Asynchronous Receiver/Transmitter (UART) is a full-duplex communication channel capable of handling asynchronous data transfers. The UART uses a single 8-bit data mode with selectable parity. Features of the UART include:

- 8-bit asynchronous data transfer
- Selectable even- and odd-parity generation and checking
- Option of one or two Stop bits
- Separate transmit and receive interrupts
- Framing, parity, overrun and break detection
- Separate transmit and receive enables
- 16-bit Baud Rate Generator (BRG)
- Selectable MULTIPROCESSOR (9-bit) mode with three configurable interrupt schemes
- Baud Rate Generator timer mode
- Driver Enable output for external bus transceivers

#### **Architecture**

The UART consists of three primary functional blocks: Transmitter, Receiver, and Baud rate generator. The UART's transmitter and receiver function independently, but employ the same baud rate and data format. Figure 13 on page 104 displays the UART architecture.

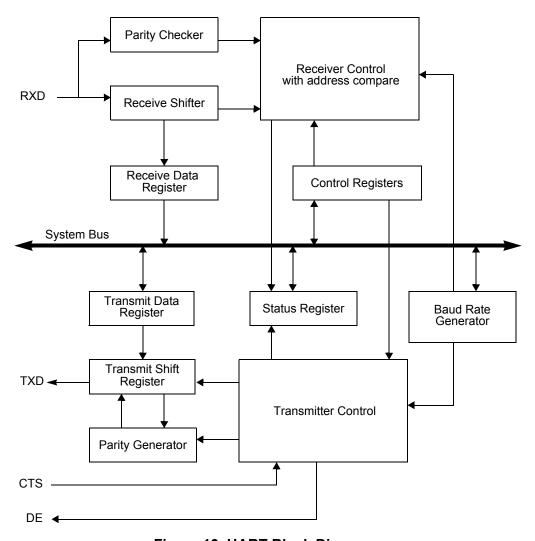


Figure 13. UART Block Diagram

# **Operation**

#### **Data Format**

The UART always transmits and receives data in an 8-bit data format, least-significant bit first. An even or odd parity bit can be optionally added to the data stream. Each character begins with an active Low Start bit and ends with either 1 or 2 active High Stop bits. Figure 14 and Figure 15 on page 105 displays the asynchronous data format employed by the UART without parity and with parity, respectively.

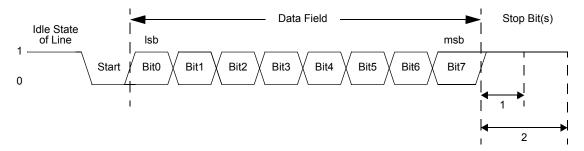


Figure 14. UART Asynchronous Data Format without Parity

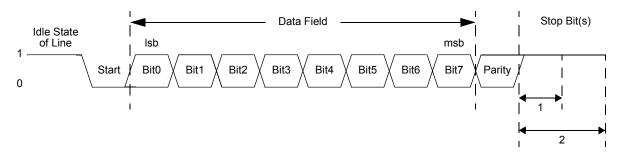


Figure 15. UART Asynchronous Data Format with Parity

## **Transmitting Data using the Polled Method**

Follow the steps below to transmit data using the polled method of operation:

- 1. Write to the UART Baud Rate High and Low Byte registers to set the desired baud rate
- 2. Enable the UART pin functions by configuring the associated GPIO Port pins for alternate function operation.
- 3. If MULTIPROCESSOR mode is desired, write to the UART Control 1 register to enable MULTIPROCESSOR (9-bit) mode functions.
  - Set the MULTIPROCESSOR Mode Select (MPEN) to Enable MULTIPROCESSOR mode.
- 4. Write to the UART Control 0 register to:
  - Set the transmit enable bit (TEN) to enable the UART for data transmission
  - If parity is desired and MULTIPROCESSOR mode is not enabled, set the parity enable bit (PEN) and select either Even or Odd parity (PSEL).

- Set or clear the CTSE bit to enable or disable control from the remote receiver using the CTS pin.
- 5. Check the TDRE bit in the UART Status 0 register to determine if the Transmit Data register is empty (indicated by a 1). If empty, continue to step 6. If the Transmit Data register is full (indicated by a 0), continue to monitor the TDRE bit until the Transmit Data register becomes available to receive new data.
- 6. Write the UART Control 1 register to select the outgoing address bit.
- 7. Set the MULTIPROCESSOR Bit Transmitter (MPBT) if sending an address byte, clear it if sending a data byte.
- 8. Write the data byte to the UART Transmit Data register. The transmitter automatically transfers the data to the Transmit Shift register and transmits the data.
- 9. If desired and MULTIPROCESSOR mode is enabled, make any changes to the MULTIPROCESSOR Bit Transmitter (MPBT) value.
- 10. To transmit additional bytes, return to step 5.

## Transmitting Data using the Interrupt-Driven Method

The UART transmitter interrupt indicates the availability of the Transmit Data register to accept new data for transmission. Follow the steps below to configure the UART for interrupt-driven data transmission:

- 1. Write to the UART Baud Rate High and Low Byte registers to set the desired baud rate.
- 2. Enable the UART pin functions by configuring the associated GPIO Port pins for alternate function operation.
- 3. Execute a DI instruction to disable interrupts.
- 4. Write to the Interrupt control registers to enable the UART Transmitter interrupt and set the desired priority.
- 5. If MULTIPROCESSOR mode is desired, write to the UART Control 1 register to enable MULTIPROCESSOR (9-bit) mode functions.
- 6. Set the MULTIPROCESSOR Mode Select (MPEN) to Enable MULTIPROCESSOR mode.
- 7. Write to the UART Control 0 register to:
  - Set the transmit enable bit (TEN) to enable the UART for data transmission.
  - Enable parity, if desired and if MULTIPROCESSOR mode is not enabled, and select either even or odd parity.
  - Set or clear the CTSE bit to enable or disable control from the remote receiver via the CTS pin.

8. Execute an EI instruction to enable interrupts.

The UART is now configured for interrupt-driven data transmission. Because the UART Transmit Data register is empty, an interrupt is generated immediately. When the UART Transmit interrupt is detected, the associated interrupt service routine performs the following:

- 1. Write the UART Control 1 register to select the outgoing address bit:
  - Set the MULTIPROCESSOR Bit Transmitter (MPBT) if sending an address byte, clear it if sending a data byte.
- 2. Write the data byte to the UART Transmit Data register. The transmitter automatically transfers the data to the Transmit Shift register and transmits the data.
- 3. Clear the UART Transmit interrupt bit in the applicable Interrupt Request register.
- 4. Execute the IRET instruction to return from the interrupt-service routine and wait for the Transmit Data register to again become empty.

## Receiving Data using the Polled Method

Follow the steps below to configure the UART for polled data reception:

- 1. Write to the UART Baud Rate High and Low Byte registers to set the desired baud rate.
- 2. Enable the UART pin functions by configuring the associated GPIO Port pins for alternate function operation.
- 3. Write to the UART Control 1 register to enable MULTIPROCESSOR mode functions, if desired.
- 4. Write to the UART Control 0 register to:
  - Set the receive enable bit (REN) to enable the UART for data reception.
  - Enable parity, if desired and if MULTIPROCESSOR mode is not enabled, and select either even or odd parity.
- 5. Check the RDA bit in the UART Status 0 register to determine if the Receive Data register contains a valid data byte (indicated by a 1). If RDA is set to 1 to indicate available data, continue to step 6. If the Receive Data register is empty (indicated by a 0), continue to monitor the RDA bit awaiting reception of the valid data.
- 6. Read data from the UART Receive Data register. If operating in MULTIPROCESSOR (9-bit) mode, further actions may be required depending on the MULTIPROCESSOR Mode bits MPMD[1:0].
- 7. Return to step 5 to receive additional data.

## Receiving Data using the Interrupt-Driven Method

The UART Receiver interrupt indicates the availability of new data (as well as error conditions). Follow the steps below to configure the UART receiver for interrupt-driven operation:

- 1. Write to the UART Baud Rate High and Low Byte registers to set the desired baud rate
- 2. Enable the UART pin functions by configuring the associated GPIO Port pins for alternate function operation.
- 3. Execute a DI instruction to disable interrupts.
- 4. Write to the Interrupt control registers to enable the UART Receiver interrupt and set the desired priority.
- 5. Clear the UART Receiver interrupt in the applicable Interrupt Request register.
- 6. Write to the UART Control 1 Register to enable MULTIPROCESSOR (9-bit) mode functions, if desired.
  - Set the MULTIPROCESSOR Mode Select (MPEN) to Enable MULTIPROCESSOR mode.
  - Set the MULTIPROCESSOR Mode Bits, MPMD[1:0], to select the desired address matching scheme.
  - Configure the UART to interrupt on received data and errors or errors only (interrupt on errors only is unlikely to be useful for Z8 Encore! devices without a DMA block).
- 7. Write the device address to the Address Compare Register (automatic multiprocessor modes only).
- 8. Write to the UART Control 0 register to:
  - Set the receive enable bit (REN) to enable the UART for data reception.
  - Enable parity, if desired and if MULTIPROCESSOR mode is not enabled, and select either even or odd parity.
- 9. Execute an EI instruction to enable interrupts.

The UART is now configured for interrupt-driven data reception. When the UART Receiver interrupt is detected, the associated interrupt service routine performs the following:

- 1. Check the UART Status 0 register to determine the source of the interrupt error, break, or received data.
- 2. If the interrupt was caused by data available, read the data from the UART Receive Data register. If operating in MULTIPROCESSOR (9-bit) mode, further actions may be required depending on the MULTIPROCESSOR Mode bits MPMD[1:0].

- 3. Clear the UART Receiver interrupt in the applicable Interrupt Request register.
- 4. Execute the IRET instruction to return from the interrupt-service routine and await more data.

# Clear To Send (CTS) Operation

The CTS pin, if enabled by the CTSE bit of the UART Control 0 register, performs flow control on the outgoing transmit datastream. The Clear To Send ( $\overline{\text{CTS}}$ ) input pin is sampled one system clock before beginning any new character transmission. To delay transmission of the next data character, an external receiver must deassert  $\overline{\text{CTS}}$  at least one system clock cycle before a new data transmission begins. For multiple character transmissions, this would typically be done during Stop Bit transmission. If  $\overline{\text{CTS}}$  deasserts in the middle of a character transmission, the current character is sent completely.

## **MULTIPROCESSOR (9-bit) Mode**

The UART has a MULTIPROCESSOR (9-bit) mode that uses an extra (9th) bit for selective communication when a number of processors share a common UART bus. In MULTIPROCESSOR mode (also referred to as 9-Bit mode), the multiprocessor bit (MP) is transmitted immediately following the 8-bits of data and immediately preceding the Stop bit(s) as displayed in Figure 16. The character format is:

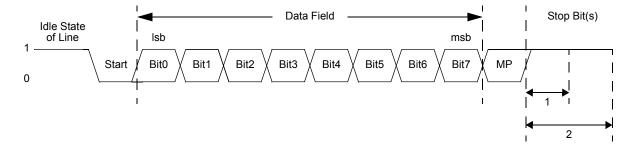


Figure 16. UART Asynchronous MULTIPROCESSOR Mode Data Format

In MULTIPROCESSOR (9-bit) mode, the Parity bit location (9th bit) becomes the MULTIPROCESSOR control bit. The UART Control 1 and Status 1 registers provide MULTIPROCESSOR (9-bit) mode control and status information. If an automatic address matching scheme is enabled, the UART Address Compare register holds the network address of the device.

## **MULTIPROCESSOR (9-bit) Mode Receive Interrupts**

When MULTIPROCESSOR mode is enabled, the UART only processes frames addressed to it. The determination of whether a frame of data is addressed to the UART can be made in hardware, software or some combination of the two, depending on the multiprocessor



configuration bits. In general, the address compare feature reduces the load on the CPU, since it does not need to access the UART when it receives data directed to other devices on the multi-node network. The following three MULTIPROCESSOR modes are available in hardware:

- Interrupt on all address bytes.
- Interrupt on matched address bytes and correctly framed data bytes.
- Interrupt only on correctly framed data bytes.

These modes are selected with MPMD[1:0] in the UART Control 1 Register. For all MULTIPROCESSOR modes, bit MPEN of the UART Control 1 Register must be set to 1.

The first scheme is enabled by writing 01b to MPMD[1:0]. In this mode, all incoming address bytes cause an interrupt, while data bytes never cause an interrupt. The interrupt service routine must manually check the address byte that caused triggered the interrupt. If it matches the UART address, the software clears MPMD[0]. At this point, each new incoming byte interrupts the CPU. The software is then responsible for determining the end of the frame. It checks for end-of-frame by reading the MPRX bit of the UART Status 1 Register for each incoming byte. If MPRX=1, a new frame has begun. If the address of this new frame is different from the UART's address, then set MPMD[0] to 1 causing the UART interrupts to go inactive until the next address byte. If the new frame's address matches the UART's, the data in the new frame is processed as well.

The second scheme is enabled by setting MPMD[1:0] to 10b and writing the UART's address into the UART Address Compare Register. This mode introduces more hardware control, interrupting only on frames that match the UART's address. When an incoming address byte does not match the UART's address, it is ignored. All successive data bytes in this frame are also ignored. When a matching address byte occurs, an interrupt is issued and further interrupts now occur on each successive data byte. The first data byte in the frame contains the NEWFRM=1 in the UART Status 1 Register. When the next address byte occurs, the hardware compares it to the UART's address. If there is a match, the interrupts continue sand the NEWFRM bit is set for the first byte of the new frame. If there is no match, then the UART ignores all incoming bytes until the next address match.

The third scheme is enabled by setting MPMD[1:0] to 11b and by writing the UART's address into the UART Address Compare Register. This mode is identical to the second scheme, except that there are no interrupts on address bytes. The first data byte of each frame is still accompanied by a NEWFRM assertion.

## **External Driver Enable**

The UART provides a Driver Enable (DE) signal for off-chip bus transceivers. This feature reduces the software overhead associated with using a GPIO pin to control the transceiver when communicating on a multi-transceiver bus, such as RS-485.

Driver Enable is an active High signal that envelopes the entire transmitted data frame including parity and Stop bits as displayed in Figure 17. The Driver Enable signal asserts

when a byte is written to the UART Transmit Data register. The Driver Enable signal asserts at least one UART bit period and no greater than two UART bit periods before the Start bit is transmitted. This timing allows a setup time to enable the transceiver. The Driver Enable signal deasserts one system clock period after the last Stop bit is transmitted. This one system clock delay allows both time for data to clear the transceiver before disabling it, as well as the ability to determine if another character follows the current character. In the event of back to back characters (new data must be written to the Transmit Data Register before the previous character is completely transmitted) the DE signal is not deasserted between characters. The DEPOL bit in the UART Control Register 1 sets the polarity of the Driver Enable signal.

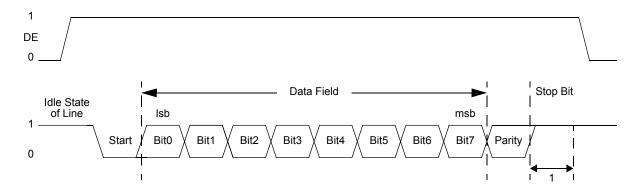


Figure 17. UART Driver Enable Signal Timing (shown with 1 Stop Bit and Parity)

The Driver Enable to Start bit setup time is calculated as follows:

$$\left(\frac{1}{\text{Baud Rate (Hz)}}\right) \le \text{DE to Start Bit Setup Time (s)} \le \left(\frac{2}{\text{Baud Rate (Hz)}}\right)$$

## **UART Interrupts**

The UART features separate interrupts for the transmitter and the receiver. In addition, when the UART primary functionality is disabled, the Baud Rate Generator can also function as a basic timer with interrupt capability.

## **Transmitter Interrupts**

The transmitter generates a single interrupt when the Transmit Data Register Empty bit (TDRE) is set to 1. This indicates that the transmitter is ready to accept new data for transmission. The TDRE interrupt occurs after the Transmit shift register has shifted the first bit of data out. At this point, the Transmit Data register may be written with the next character to send. This provides 7 bit periods of latency to load the Transmit Data register before the Transmit shift register completes shifting the current character. Writing to the UART Transmit Data register clears the TDRE bit to 0.



## **Receiver Interrupts**

The receiver generates an interrupt when any of the following occurs:

- A data byte has been received and is available in the UART Receive Data register.
   This interrupt can be disabled independent of the other receiver interrupt sources. The received data interrupt occurs once the receive character has been received and placed in the Receive Data register. Software must respond to this received data available condition before the next character is completely received to avoid an overrun error.
- **Note:** In MULTIPROCESSOR mode (MPEN = 1), the receive data interrupts are dependent on the multiprocessor configuration and the most recent address byte.
  - A break is received
  - An overrun is detected
  - A data framing error is detected

#### **UART Overrun Errors**

When an overrun error condition occurs the UART prevents overwriting of the valid data currently in the Receive Data register. The Break Detect and Overrun status bits are not displayed until after the valid data has been read.

After the valid data has been read, the UART Status 0 register is updated to indicate the overrun condition (and Break Detect, if applicable). The RDA bit is set to 1 to indicate that the Receive Data register contains a data byte. However, because the overrun error occurred, this byte may not contain valid data and should be ignored. The BRKD bit indicates if the overrun was caused by a break condition on the line. After reading the status byte indicating an overrun error, the Receive Data register must be read again to clear the error bits is the UART Status 0 register. Updates to the Receive Data register occur only when the next data word is received.

## **UART Data and Error Handling Procedure**

Figure 18 on page 113 displays the recommended procedure for use in UART receiver interrupt service routines.

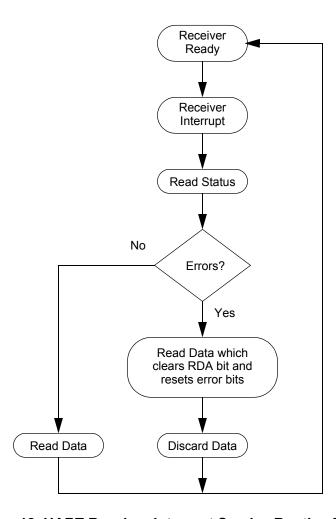


Figure 18. UART Receiver Interrupt Service Routine Flow

## **Baud Rate Generator Interrupts**

If the Baud Rate Generator interrupt enable is set, the UART Receiver interrupt asserts when the UART Baud Rate Generator reloads. This action allows the Baud Rate Generator to function as an additional counter if the UART functionality is not employed.

#### **UART Baud Rate Generator**

The UART Baud Rate Generator creates a lower frequency baud rate clock for data transmission. The input to the Baud Rate Generator is the system clock. The UART Baud Rate High and Low Byte registers combine to create a 16-bit baud rate divisor value

(BRG[15:0]) that sets the data transmission rate (baud rate) of the UART. The UART data rate is calculated using the following equation:

UART Data Rate (bits/s) = 
$$\frac{\text{System Clock Frequency (Hz)}}{16 \times \text{UART Baud Rate Divisor Value}}$$

When the UART is disabled, the Baud Rate Generator can function as a basic 16-bit timer with interrupt on time-out. To configure the Baud Rate Generator as a timer with interrupt on time-out, complete the following procedure:

- 1. Disable the UART by clearing the REN and TEN bits in the UART Control 0 register to 0.
- 2. Load the desired 16-bit count value into the UART Baud Rate High and Low Byte registers.
- 3. Enable the Baud Rate Generator timer function and associated interrupt by setting the BRGCTL bit in the UART Control 1 register to 1.

When configured as a general purpose timer, the interrupt interval is calculated using the following equation:

Interrupt Interval(s) = System Clock Period (s)  $\times$  BRG[15:0]

# **UART Control Register Definitions**

The UART control registers support the UART and the associated Infrared Encoder/ Decoders. For more information on the infrared operation, see Infrared Encoder/Decoder on page 125.

# **UART Transmit Data Register**

Data bytes written to the UART Transmit Data register (Table 52) are shifted out on the TXDx pin. The Write-only UART Transmit Data register shares a Register File address with the Read-only UART Receive Data register.

Table 52. UART Transmit Data Register (UxTXD)

BITS	7	6	5	4	3	2	1	0		
FIELD		TXD								
RESET		Х								
R/W				V	V					
ADDR				F40H ar	nd F48H					

TXD—Transmit Data

UART transmitter data byte to be shifted out through the TXDx pin.

# **UART Receive Data Register**

Data bytes received through the RXDx pin are stored in the UART Receive Data register (Table 53). The Read-only UART Receive Data register shares a Register File address with the Write-only UART Transmit Data register.

Table 53. UART Receive Data Register (UxRXD)

BITS	7	6	5	4	3	2	1	0		
FIELD		RXD								
RESET		X								
R/W				F	₹					
ADDR				F40H ar	nd F48H					

RXD—Receive Data UART receiver data byte from the RXDx pin

## **UART Status 0 Register**

The UART Status 0 and Status 1 registers (Table 54 and Table 55 on page 117) identify the current UART operating configuration and status.

Table 54. UART Status 0 Register (UxSTAT0)

BITS	7	6	5	4	3	2	1	0		
FIELD	RDA	PE	TDRE	TXE	CTS					
RESET		0								
R/W				F	₹					
ADDR				F41H ar	nd F49H					

RDA—Receive Data Available

This bit indicates that the UART Receive Data register has received data. Reading the UART Receive Data register clears this bit.

0 = The UART Receive Data register is empty.

1 = There is a byte in the UART Receive Data register.

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### PE—Parity Error

This bit indicates that a parity error has occurred. Reading the UART Receive Data register clears this bit.

0 = No parity error occurred.

1 = A parity error occurred.

#### OE—Overrun Error

This bit indicates that an overrun error has occurred. An overrun occurs when new data is received and the UART Receive Data register has not been read. If the RDA bit is reset to 0, then reading the UART Receive Data register clears this bit.

0 = No overrun error occurred.

1 = An overrun error occurred.

## FE—Framing Error

This bit indicates that a framing error (no Stop bit following data reception) was detected. Reading the UART Receive Data register clears this bit.

0 =No framing error occurred.

1 = A framing error occurred.

#### BRKD—Break Detect

This bit indicates that a break occurred. If the data bits, parity/multiprocessor bit, and Stop bit(s) are all zeros then this bit is set to 1. Reading the UART Receive Data register clears this bit.

0 = No break occurred.

1 = A break occurred.

#### TDRE—Transmitter Data Register Empty

This bit indicates that the UART Transmit Data register is empty and ready for additional data. Writing to the UART Transmit Data register resets this bit.

0 = Do not write to the UART Transmit Data register.

1 = The UART Transmit Data register is ready to receive an additional byte to be transmitted.

#### TXE—Transmitter Empty

This bit indicates that the transmit shift register is empty and character transmission is finished

0 = Data is currently transmitting.

1 = Transmission is complete.

#### CTS—CTS signal

When this bit is read it returns the level of the  $\overline{\text{CTS}}$  signal.

## **UART Status 1 Register**

This register contains multiprocessor control and status bits.

### Table 55. UART Status 1 Register (UxSTAT1)

BITS	7	6	5	4	3	2	1	0		
FIELD		Reserved NEWFRM MPRX								
RESET		0								
R/W		F	₹		F	R/W	F			
ADDR				F44H ar	nd F4CH					

Reserved—Must be 0.

NEWFRM—Status bit denoting the start of a new frame. Reading the UART Receive Data register resets this bit to 0.

0 = The current byte is not the first data byte of a new frame.

1 = The current byte is the first data byte of a new frame.

MPRX—Multiprocessor Receive

Returns the value of the last multiprocessor bit received. Reading from the UART Receive Data register resets this bit to 0.

## **UART Control 0 and Control 1 Registers**

The UART Control 0 and Control 1 registers (see Table 56 and Table 57 on page 118) configure the properties of the UART's transmit and receive operations. The UART Control registers must not been written while the UART is enabled.

Table 56. UART Control 0 Register (UxCTL0)

BITS	7	6	5	4	3	2	1	0			
FIELD	TEN	TEN REN CTSE PEN PSEL SBRK STOP LBEN									
RESET		0									
R/W				R/	W						
ADDR				F42H ar	nd F4AH						

TEN—Transmit Enable

This bit enables or disables the transmitter. The enable is also controlled by the  $\overline{CTS}$  signal and the CTSE bit. If the  $\overline{CTS}$  signal is low and the CTSE bit is 1, the transmitter is enabled.

0 = Transmitter disabled.

1 = Transmitter enabled.

REN—Receive Enable

This bit enables or disables the receiver.

0 =Receiver disabled.

1 = Receiver enabled.

#### CTSE—CTS Enable

 $0 = \text{The } \overline{\text{CTS}}$  signal has no effect on the transmitter.

 $1 = \text{The UART recognizes the } \overline{\text{CTS}}$  signal as an enable control from the transmitter.

#### PEN—Parity Enable

This bit enables or disables parity. Even or odd is determined by the PSEL bit. It is over-ridden by the MPEN bit.

0 = Parity is disabled.

1 = The transmitter sends data with an additional parity bit and the receiver receives an additional parity bit.

#### PSEL—Parity Select

0 = Even parity is transmitted and expected on all received data.

1 = Odd parity is transmitted and expected on all received data.

#### SBRK—Send Break

This bit pauses or breaks data transmission. Sending a break interrupts any transmission in progress, so ensure that the transmitter has finished sending data before setting this bit.

0 = No break is sent.

1 =The output of the transmitter is zero.

#### STOP—Stop Bit Select

0 =The transmitter sends one stop bit.

1 = The transmitter sends two stop bits.

#### LBEN—Loop Back Enable

0 = Normal operation.

1 = All transmitted data is looped back to the receiver.

## Table 57. UART Control 1 Register (UxCTL1)

BITS	7	6	5	4	3	2	1	0		
FIELD	MPMD[1]	MPMD[1] MPEN MPMD[0] MPBT DEPOL BRGCTL RDAIRQ IREN								
RESET		0								
R/W				R/	W					
ADDR				F43H ar	nd F4BH					

MPMD[1:0]—MULTIPROCESSOR Mode

If MULTIPROCESSOR (9-bit) mode is enabled,

00 = The UART generates an interrupt request on all received bytes (data and address).

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- 01 = The UART generates an interrupt request only on received address bytes.
- 10 = The UART generates an interrupt request when a received address byte matches the value stored in the Address Compare Register and on all successive data bytes until an address mismatch occurs.
- 11 = The UART generates an interrupt request on all received data bytes for which the most recent address byte matched the value in the Address Compare Register.

## MPEN—MULTIPROCESSOR (9-bit) Enable

This bit is used to enable MULTIPROCESSOR (9-bit) mode.

- 0 = Disable MULTIPROCESSOR (9-bit) mode.
- 1 = Enable MULTIPROCESSOR (9-bit) mode.

#### MPBT—MULTIPROCESSOR Bit Transmit

This bit is applicable only when MULTIPROCESSOR (9-bit) mode is enabled.

- 0 =Send a 0 in the multiprocessor bit location of the data stream ( $9^{th}$  bit).
- 1 =Send a 1 in the multiprocessor bit location of the data stream ( $9^{th}$  bit).

#### DEPOL—Driver Enable Polarity

0 = DE signal is Active High.

1 = DE signal is Active Low.

#### BRGCTL—Baud Rate Control

This bit causes different UART behavior depending on whether the UART receiver is enabled (REN = 1 in the UART Control 0 Register).

When the UART receiver is not enabled, this bit determines whether the Baud Rate Generator issues interrupts.

- 0 = Reads from the Baud Rate High and Low Byte registers return the BRG Reload Value
- 1 = The Baud Rate Generator generates a receive interrupt when it counts down to 0.

Reads from the Baud Rate High and Low Byte registers return the current BRG count value.

When the UART receiver is enabled, this bit allows reads from the Baud Rate Registers to return the BRG count value instead of the Reload Value.

- 0 = Reads from the Baud Rate High and Low Byte registers return the BRG Reload Value.
- 1 = Reads from the Baud Rate High and Low Byte registers return the current BRG count value. Unlike the Timers, there is no mechanism to latch the High Byte when the Low Byte is read.

## RDAIRQ—Receive Data Interrupt Enable

- 0 = Received data and receiver errors generates an interrupt request to the Interrupt Controller.
- 1 = Received data does not generate an interrupt request to the Interrupt Controller. Only receiver errors generate an interrupt request.

## IREN—Infrared Encoder/Decoder Enable

0 = Infrared Encoder/Decoder is disabled. UART operates normally operation.

1 = Infrared Encoder/Decoder is enabled. The UART transmits and receives data through the Infrared Encoder/Decoder.

# **UART Address Compare Register**

The UART Address Compare register (Table 58) stores the multi-node network address of the UART. When the MPMD[1] bit of UART Control Register 0 is set, all incoming address bytes are compared to the value stored in the Address Compare register. Receive interrupts and RDA assertions only occur in the event of a match.

Table 58. UART Address Compare Register (UxADDR)

BITS	7	6	5	4	3	2	1	0			
FIELD	COMP_ADDR										
RESET		0									
R/W		R/W									
ADDR				F45H ar	nd F4DH						

COMP\_ADDR—Compare Address

This 8-bit value is compared to the incoming address bytes.

# **UART Baud Rate High and Low Byte Registers**

The UART Baud Rate High and Low Byte registers (see Table 59 and Table 60 on page 121) combine to create a 16-bit baud rate divisor value (BRG[15:0]) that sets the data transmission rate (baud rate) of the UART. To configure the Baud Rate Generator as a timer with interrupt on time-out, complete the following procedure:

- 1. Disable the UART by clearing the REN and TEN bits in the UART Control 0 register to 0.
- 2. Load the desired 16-bit count value into the UART Baud Rate High and Low Byte registers.
- 3. Enable the Baud Rate Generator timer function and associated interrupt by setting the BRGCTL bit in the UART Control 1 register to 1.

When configured as a general purpose timer, the UART BRG interrupt interval is calculated using the following equation:

UART BRG Interrupt Interval(s) = System Clock Period (s) × BRG[15:0]

## Table 59. UART Baud Rate High Byte Register (UxBRH)

BITS	7	6	5	4	3	2	1	0			
FIELD	BRH										
RESET		1									
R/W		R/W									
ADDR				F46H ar	nd F4EH						

## Table 60. UART Baud Rate Low Byte Register (UxBRL)

BITS	7 6 5 4 3					2	1	0			
FIELD	BRL										
RESET		1									
R/W				R/	W						
ADDR				F47H ar	nd F4FH						

For a given UART data rate, the integer baud rate divisor value is calculated using the following equation:

UART Baud Rate Divisor Value (BRG) = 
$$Round$$
 (System Clock Frequency (Hz))  $\frac{16 \times UART Data Rate (bits/s)}{16 \times UART Data Rate (bits/s)}$ 

The baud rate error relative to the desired baud rate is calculated using the following equation:

UART Baud Rate Error (%) = 
$$100 \times \left(\frac{\text{Actual Data Rate} - \text{Desired Data Rate}}{\text{Desired Data Rate}}\right)$$

For reliable communication, the UART baud rate error must never exceed 5 percent. Table 61 provides information on data rate errors for popular baud rates and commonly used crystal oscillator frequencies.

**Table 61. UART Baud Rates** 

## 20.0 MHz System Clock

## 18.432 MHz System Clock

			_					
Desired Rate	BRG Divisor	Actual Rate	Error		Desired Rate	BRG Divisor	Actual Rate	Error
(kHz)	(Decimal)	(kHz)	(%)	_	(kHz)	(Decimal)	(kHz)	(%)
1250.0	1	1250.0	0.00		1250.0	1	1152.0	-7.84%
625.0	2	625.0	0.00	_	625.0	2	576.0	-7.84%
250.0	5	250.0	0.00	_	250.0	5	230.4	-7.84%
115.2	11	113.6	-1.36	_	115.2	10	115.2	0.00
57.6	22	56.8	-1.36	_	57.6	20	57.6	0.00
38.4	33	37.9	-1.36	_	38.4	30	38.4	0.00
19.2	65	19.2	0.16	_	19.2	60	19.2	0.00
9.60	130	9.62	0.16	_	9.60	120	9.60	0.00
4.80	260	4.81	0.16	=	4.80	240	4.80	0.00
2.40	521	2.40	-0.03	=	2.40	480	2.40	0.00
1.20	1042	1.20	-0.03	=	1.20	960	1.20	0.00
0.60	2083	0.60	0.02	=	0.60	1920	0.60	0.00
0.30	4167	0.30	-0.01	=	0.30	3840	0.30	0.00
-				_				

# 16.667 MHz System Clock

# 11.0592 MHz System Clock

Desired Rate	BRG Divisor	Actual Rate	Error	Desired Rate	BRG Divisor	Actual Rate	Error
(kHz)	(Decimal)	(kHz)	(%)	(kHz)	(Decimal)	(kHz)	(%)
1250.0	1	1041.69	-16.67	1250.0	N/A	N/A	N/A
625.0	2	520.8	-16.67	625.0	1	691.2	10.59
250.0	4	260.4	4.17	250.0	3	230.4	-7.84
115.2	9	115.7	0.47	115.2	6	115.2	0.00
57.6	18	57.87	0.47	57.6	12	57.6	0.00
38.4	27	38.6	0.47	38.4	18	38.4	0.00
19.2	54	19.3	0.47	19.2	36	19.2	0.00
9.60	109	9.56	-0.45	9.60	72	9.60	0.00
4.80	217	4.80	-0.83	4.80	144	4.80	0.00
2.40	434	2.40	0.01	2.40	288	2.40	0.00

Table 61. UART Baud Rates (Continued)

1.20	868	1.20	0.01		1.20	576	1.20	0.00
0.60	1736	0.60	0.01		0.60	1152	0.60	0.00
0.30	3472	0.30	0.01	-	0.30	2304	0.30	0.00

## 10.0 MHz System Clock

## 5.5296 MHz System Clock

Desired Rate	BRG Divisor	Actual Rate	Error	Desired Rate	BRG Divisor	Actual Rate	Error
(kHz)	(Decimal)	(kHz)	(%)	(kHz)	(Decimal)	(kHz)	(%)
1250.0	N/A	N/A	N/A	1250.0	N/A	N/A	N/A
625.0	1	625.0	0.00	625.0	N/A	N/A	N/A
250.0	3	208.33	-16.67	250.0	1	345.6	38.24
115.2	5	125.0	8.51	115.2	3	115.2	0.00
57.6	11	56.8	-1.36	57.6	6	57.6	0.00
38.4	16	39.1	1.73	38.4	9	38.4	0.00
19.2	33	18.9	0.16	19.2	18	19.2	0.00
9.60	65	9.62	0.16	9.60	36	9.60	0.00
4.80	130	4.81	0.16	4.80	72	4.80	0.00
2.40	260	2.40	-0.03	2.40	144	2.40	0.00
1.20	521	1.20	-0.03	1.20	288	1.20	0.00
0.60	1042	0.60	-0.03	0.60	576	0.60	0.00
0.30	2083	0.30	0.2	0.30	1152	0.30	0.00

## 3.579545 MHz System Clock

## 1.8432 MHz System Clock

Desired Rate	BRG Divisor	Actual Rate	Error	Desired Rate	BRG Divisor	Actual Rate	Error
(kHz)	(Decimal)	(kHz)	(%)	(kHz)	(Decimal)	(kHz)	(%)
1250.0	N/A	N/A	N/A	1250.0	N/A	N/A	N/A
625.0	N/A	N/A	N/A	625.0	N/A	N/A	N/A
250.0	1	223.72	-10.51	250.0	N/A	N/A	N/A
115.2	2	111.9	-2.90	115.2	1	115.2	0.00
57.6	4	55.9	-2.90	57.6	2	57.6	0.00
38.4	6	37.3	-2.90	38.4	3	38.4	0.00
19.2	12	18.6	-2.90	19.2	6	19.2	0.00



Table 61. UART Baud Rates (Continued)

9.60	23	9.73	1.32	9.60	12	9.60	0.00
4.80	47	4.76	-0.83	4.80	24	4.80	0.00
2.40	93	2.41	0.23	2.40	48	2.40	0.00
1.20	186	1.20	0.23	1.20	96	1.20	0.00
0.60	373	0.60	-0.04	0.60	192	0.60	0.00
0.30	746	0.30	-0.04	0.30	384	0.30	0.00

# Infrared Encoder/Decoder

## **Overview**

The 64K Series products contain two fully-functional, high-performance UART to Infrared Encoder/Decoders (Endecs). Each Infrared Endec is integrated with an on-chip UART to allow easy communication between the 64K Series and IrDA Physical Layer Specification, Version 1.3-compliant infrared transceivers. Infrared communication provides secure, reliable, low-cost, point-to-point communication between PCs, PDAs, cell phones, printers, and other infrared enabled devices.

## **Architecture**

Figure 19 displays the architecture of the Infrared Endec.

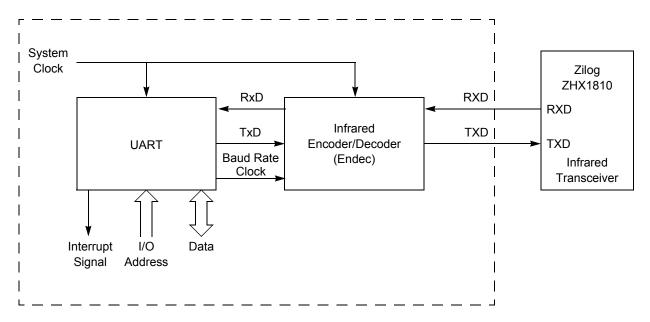


Figure 19. Infrared Data Communication System Block Diagram

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# Operation

When the Infrared Endec is enabled, the transmit data from the associated on-chip UART is encoded as digital signals in accordance with the IrDA standard and output to the infrared transceiver via the TXD pin. Likewise, data received from the infrared transceiver is passed to the Infrared Endec via the RXD pin, decoded by the Infrared Endec, and then passed to the UART. Communication is half-duplex, which means simultaneous data transmission and reception is not allowed.

The baud rate is set by the UART's Baud Rate Generator and supports IrDA standard baud rates from 9600 baud to 115.2 Kbaud. Higher baud rates are possible, but do not meet IrDA specifications. The UART must be enabled to use the Infrared Endec. The Infrared Endec data rate is calculated using the following equation:

Infrared Data Rate (bits/s) = 
$$\frac{\text{System Clock Frequency (Hz)}}{16 \times \text{UART Baud Rate Divisor Value}}$$

## **Transmitting IrDA Data**

The data to be transmitted using the infrared transceiver is first sent to the UART. The UART's transmit signal (TXD) and baud rate clock are used by the IrDA to generate the modulation signal (IR\_TXD) that drives the infrared transceiver. Each UART/Infrared data bit is 16-clock wide. If the data to be transmitted is 1, the IR\_TXD signal remains low for the full 16-clock period. If the data to be transmitted is 0, a 3-clock high pulse is output following a 7-clock low period. After the 3-clock high pulse, a 6-clock low pulse is output to complete the full 16-clock data period. Figure 20 displays IrDA data transmission. When the Infrared Endec is enabled, the UART's TXD signal is internal to the 64K Series products while the IR TXD signal is output through the TXD pin.

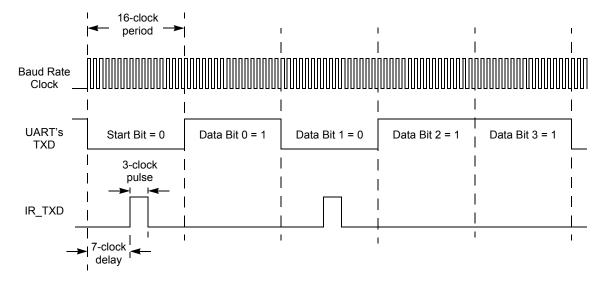


Figure 20. Infrared Data Transmission

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## Receiving IrDA Data

Data received from the infrared transceiver via the IR\_RXD signal through the RXD pin is decoded by the Infrared Endec and passed to the UART. The UART's baud rate clock is used by the Infrared Endec to generate the demodulated signal (RXD) that drives the UART. Each UART/Infrared data bit is 16-clocks wide. Figure 21 displays data reception. When the Infrared Endec is enabled, the UART's RXD signal is internal to the 64K Series products while the IR\_RXD signal is received through the RXD pin.

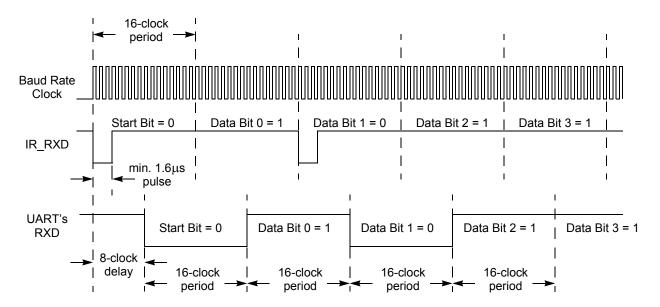


Figure 21. Infrared Data Reception

<u>^</u>

**Caution:** The system clock frequency must be at least 1.0 MHz to ensure proper reception of the 1.6 μs minimum width pulses allowed by the IrDA standard.

#### **Endec Receiver Synchronization**

The IrDA receiver uses a local baud rate clock counter (0 to 15 clock periods) to generate an input stream for the UART and to create a sampling window for detection of incoming pulses. The generated UART input (UART RXD) is delayed by 8 baud rate clock periods with respect to the incoming IrDA data stream. When a falling edge in the input data stream is detected, the Endec counter is reset. When the count reaches a value of 8, the UART RXD value is updated to reflect the value of the decoded data. When the count reaches 12 baud clock periods, the sampling window for the next incoming pulse opens. The window remains open until the count again reaches 8 (or in other words 24 baud clock periods since the previous pulse was detected). This gives the Endec a sampling window of minus four baudrate clocks to plus eight baudrate clocks around the expected time of an incoming pulse. If an incoming pulse is detected inside this window this process is

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repeated. If the incoming data is a logical 1 (no pulse), the Endec returns to the initial state and waits for the next falling edge. As each falling edge is detected, the Endec clock counter is reset, resynchronizing the Endec to the incoming signal. This action allows the Endec to tolerate jitter and baud rate errors in the incoming data stream. Resynchronizing the Endec does not alter the operation of the UART, which ultimately receives the data. The UART is only synchronized to the incoming data stream when a Start bit is received.

## Infrared Encoder/Decoder Control Register Definitions

All Infrared Endec configuration and status information is set by the UART control registers as defined in UART Control Register Definitions on page 114.



**Caution:** To prevent spurious signals during IrDA data transmission, set the IREN bit in the UARTx Control 1 register to 1 to enable the Infrared Encoder/Decoder before enabling the GPIO Port alternate function for the corresponding pin.

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# **Serial Peripheral Interface**

## **Overview**

The Serial Peripheral Interface is a synchronous interface allowing several SPI-type devices to be interconnected. SPI-compatible devices include EEPROMs, Analog-to-Digital Converters, and ISDN devices. Features of the SPI include:

- Full-duplex, synchronous, character-oriented communication
- Four-wire interface
- Data transfers rates up to a maximum of one-half the system clock frequency
- Error detection
- Dedicated Baud Rate Generator

## **Architecture**

The SPI may be configured as either a Master (in single or multi-master systems) or a Slave as displayed in Figure 22 through Figure 24.

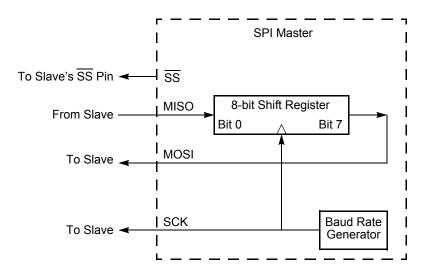


Figure 22. SPI Configured as a Master in a Single Master, Single Slave System

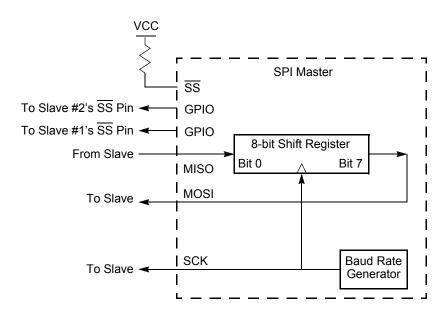


Figure 23. SPI Configured as a Master in a Single Master, Multiple Slave System

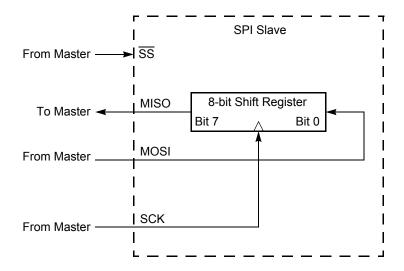


Figure 24. SPI Configured as a Slave

# **Operation**

The SPI is a full-duplex, synchronous, character-oriented channel that supports a four-wire interface (serial clock, transmit, receive and Slave select). The SPI block consists of a transmit/receive shift register, a Baud Rate (clock) Generator and a control unit.

During an SPI transfer, data is sent and received simultaneously by both the Master and the Slave SPI devices. Separate signals are required for data and the serial clock. When an SPI transfer occurs, a multi-bit (typically 8-bit) character is shifted out one data pin and an multi-bit character is simultaneously shifted in on a second data pin. An 8-bit shift register in the Master and another 8-bit shift register in the Slave are connected as a circular buffer. The SPI shift register is single-buffered in the transmit and receive directions. New data to be transmitted cannot be written into the shift register until the previous transmission is complete and receive data (if valid) has been read.

# SPI Signals

The four basic SPI signals are:

- Master-In/Slave-Out
- Master-Out/Slave-In
- Serial Clock
- Slave Select

Each signal is described in both Master and Slave modes.

#### Master-In/Slave-Out

The Master-In/Slave-Out (MISO) pin is configured as an input in a Master device and as an output in a Slave device. It is one of the two lines that transfer serial data, with the most significant bit sent first. The MISO pin of a Slave device is placed in a high-impedance state if the Slave is not selected. When the SPI is not enabled, this signal is in a high-impedance state.

# Master-Out/Slave-In

The Master-Out/Slave-In (MOSI) pin is configured as an output in a Master device and as an input in a Slave device. It is one of the two lines that transfer serial data, with the most significant bit sent first. When the SPI is not enabled, this signal is in a high-impedance state.

#### **Serial Clock**

The Serial Clock (SCK) synchronizes data movement both in and out of the device through its MOSI and MISO pins. In MASTER mode, the SPI's Baud Rate Generator creates the serial clock. The Master drives the serial clock out its own SCK pin to the Slave's SCK pin. When the SPI is configured as a Slave, the SCK pin is an input and the clock signal from the Master synchronizes the data transfer between the Master and Slave devices. Slave devices ignore the SCK signal, unless the  $\overline{SS}$  pin is asserted. When configured as a slave, the SPI block requires a minimum SCK period of greater than or equal to 8 times the system (XIN) clock period.

The Master and Slave are each capable of exchanging a character of data during a sequence of NUMBITS clock cycles (see NUMBITS field in the SPI Mode Register on page 140). In both Master and Slave SPI devices, data is shifted on one edge of the SCK and is sampled on the opposite edge where data is stable. Edge polarity is determined by the SPI phase and polarity control.

#### Slave Select

The active Low Slave Select  $(\overline{SS})$  input signal selects a Slave SPI device.  $\overline{SS}$  must be Low prior to all data communication to and from the Slave device.  $\overline{SS}$  must stay Low for the full duration of each character transferred. The  $\overline{SS}$  signal may stay Low during the transfer of multiple characters or may deassert between each character.

When the SPI is configured as the only Master in an SPI system, the  $\overline{SS}$  pin can be set as either an input or an output. For communication between the Z8F642x family Z8R642x family device's SPI Master and external Slave devices, the  $\overline{SS}$  signal, as an output, can assert the  $\overline{SS}$  input pin on one of the Slave devices. Other GPIO output pins can also be employed to select external SPI Slave devices.

When the SPI is configured as one Master in a multi-master SPI system, the  $\overline{SS}$  pin must be set as an input. The  $\overline{SS}$  input signal on the Master must be High. If the  $\overline{SS}$  signal goes Low (indicating another Master is driving the SPI bus), a Collision error Flag is set in the SPI Status register.

# SPI Clock Phase and Polarity Control

The SPI supports four combinations of serial clock phase and polarity using two bits in the SPI Control register. The clock polarity bit, CLKPOL, selects an active high or active Low clock and has no effect on the transfer format. Table 62 lists the SPI Clock Phase and Polarity Operation parameters. The clock phase bit, PHASE, selects one of two fundamentally different transfer formats. For proper data transmission, the clock phase and polarity must be identical for the SPI Master and the SPI Slave. The Master always places data on the MOSI line a half-cycle before the receive clock edge (SCK signal), in order for the Slave to latch the data.

Table 62. SPI Clock Phase (PHASE) and Clock Polarity (CLKPOL) Operation

PHASE	CLKPOL	SCK Transmit Edge	SCK Receive Edge	SCK Idle State
0	0	Falling	Rising	Low
0	1	Rising	Falling	High
1	0	Rising	Falling	Low
1	1	Falling	Rising	High

## **Transfer Format PHASE Equals Zero**

Figure 25 displays the timing diagram for an SPI transfer in which PHASE is cleared to 0. The two SCK waveforms show polarity with CLKPOL reset to 0 and with CLKPOL set to one. The diagram may be interpreted as either a Master or Slave timing diagram because the SCK Master-In/Slave-Out (MISO) and Master-Out/Slave-In (MOSI) pins are directly connected between the Master and the Slave.

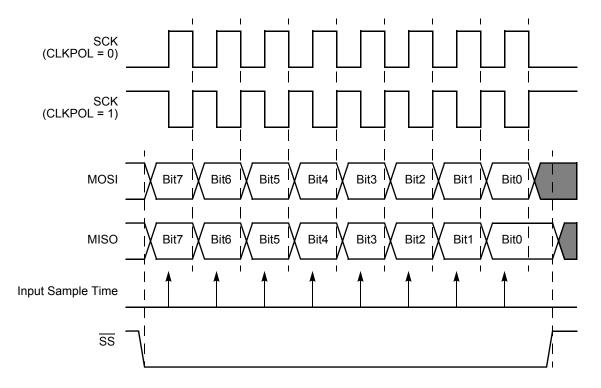


Figure 25. SPI Timing When PHASE is 0

# Transfer Format PHASE Equals One

Figure 26 on page 134 displays the timing diagram for an SPI transfer in which PHASE is one. Two waveforms are depicted for SCK, one for CLKPOL reset to 0 and another for CLKPOL set to 1.

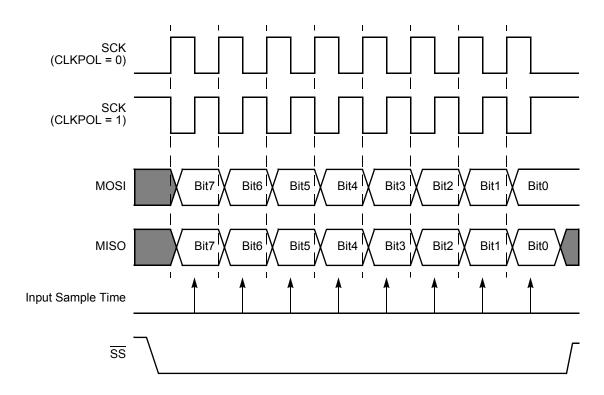


Figure 26. SPI Timing When PHASE is 1

# **Multi-Master Operation**

In a multi-master SPI system, all SCK pins are tied together, all MOSI pins are tied together and all MISO pins are tied together. All SPI pins must then be configured in OPEN-DRAIN mode to prevent bus contention. At any one time, only one SPI device is configured as the Master and all other SPI devices on the bus are configured as Slaves. The Master enables a single Slave by asserting the  $\overline{\rm SS}$  pin on that Slave only. Then, the single Master drives data out its SCK and MOSI pins to the SCK and MOSI pins on the Slaves (including those which are not enabled). The enabled Slave drives data out its MISO pin to the MISO Master pin.

For a Master device operating in a multi-master system, if the  $\overline{SS}$  pin is configured as an input and is driven Low by another Master, the COL bit is set to 1 in the SPI Status Register. The COL bit indicates the occurrence of a multi-master collision (mode fault error condition).

## **Slave Operation**

The SPI block is configured for SLAVE mode operation by setting the SPIEN bit to 1 and the MMEN bit to 0 in the SPICTL register and setting the SSIO bit to 0 in the SPIMODE

register. The IRQE, PHASE, CLKPOL, WOR bits in the SPICTL register and the NUM-BITS field in the SPIMODE register must be set to be consistent with the other SPI devices. The STR bit in the SPICTL register may be used if desired to force a "startup" interrupt. The BIRQ bit in the SPICTL register and the SSV bit in the SPIMODE register are not used in SLAVE mode. The SPI baud rate generator is not used in SLAVE mode so the SPIBRH and SPIBRL registers need not be initialized.

If the slave has data to send to the master, the data must be written to the SPIDAT register before the transaction starts (first edge of SCK when  $\overline{SS}$  is asserted). If the SPIDAT register is not written prior to the slave transaction, the MISO pin outputs whatever value is currently in the SPIDAT register.

Due to the delay resulting from synchronization of the SPI input signals to the internal system clock, the maximum SPICLK baud rate that can be supported in SLAVE mode is the system clock frequency (XIN) divided by 8. This rate is controlled by the SPI master.

#### **Error Detection**

The SPI contains error detection logic to support SPI communication protocols and recognize when communication errors have occurred. The SPI Status register indicates when a data transmission error has been detected.

## **Overrun (Write Collision)**

An overrun error (write collision) indicates a write to the SPI Data register was attempted while a data transfer is in progress (in either MASTER or SLAVE modes). An overrun sets the OVR bit in the SPI Status register to 1. Writing a 1 to OVR clears this error Flag. The data register is not altered when a write occurs while data transfer is in progress.

# Mode Fault (Multi-Master Collision)

A mode fault indicates when more than one Master is trying to communicate at the same time (a multi-master collision). The mode fault is detected when the enabled Master's  $\overline{SS}$  pin is asserted. A mode fault sets the COL bit in the SPI Status register to 1. Writing a 1 to COL clears this error Flag.

## **Slave Mode Abort**

In SLAVE mode of operation if the  $\overline{SS}$  pin deasserts before all bits in a character have been transferred, the transaction is aborted. When this condition occurs the ABT bit is set in the SPISTAT register as well as the IRQ bit (indicating the transaction is complete). The next time  $\overline{SS}$  asserts, the MISO pin outputs SPIDAT[7], regardless of where the previous transaction left off. Writing a 1 to ABT clears this error Flag.

#### SPI Interrupts

When SPI interrupts are enabled, the SPI generates an interrupt after character transmission/reception completes in both MASTER and SLAVE modes. A character can be

defined to be 1 through 8 bits by the NUMBITS field in the SPI Mode register. In slave mode it is not necessary for  $\overline{SS}$  to deassert between characters to generate the interrupt. The SPI in Slave mode can also generate an interrupt if the  $\overline{SS}$  signal deasserts prior to transfer of all the bits in a character (see description of slave abort error above). Writing a 1 to the IRQ bit in the SPI Status Register clears the pending SPI interrupt request. The IRQ bit must be cleared to 0 by the Interrupt Service Routine to generate future interrupts. To start the transfer process, an SPI interrupt may be forced by software writing a 1 to the STR bit in the SPICTL register.

If the SPI is disabled, an SPI interrupt can be generated by a Baud Rate Generator timeout. This timer function must be enabled by setting the BIRQ bit in the SPICTL register. This Baud Rate Generator time-out does not set the IRQ bit in the SPISTAT register, just the SPI interrupt bit in the interrupt controller.

#### **SPI Baud Rate Generator**

In SPI Master mode, the Baud Rate Generator creates a lower frequency serial clock (SCK) for data transmission synchronization between the Master and the external Slave. The input to the Baud Rate Generator is the system clock. The SPI Baud Rate High and Low Byte registers combine to form a 16-bit reload value, BRG[15:0], for the SPI Baud Rate Generator. The SPI baud rate is calculated using the following equation:

SPI Baud Rate (bits/s) = 
$$\frac{\text{System Clock Frequency (Hz)}}{2 \times \text{BRG}[15:0]}$$

Minimum baud rate is obtained by setting BRG[15:0] to 0000H for a clock divisor value of  $(2 \times 65536 = 131072)$ .

When the SPI is disabled, the Baud Rate Generator can function as a basic 16-bit timer with interrupt on time-out. Follow the steps below to configure the Baud Rate Generator as a timer with interrupt on time-out:

- 1. Disable the SPI by clearing the SPIEN bit in the SPI Control register to 0.
- 2. Load the desired 16-bit count value into the SPI Baud Rate High and Low Byte registers.
- 3. Enable the Baud Rate Generator timer function and associated interrupt by setting the BIRQ bit in the SPI Control register to 1.

When configured as a general purpose timer, the interrupt interval is calculated using the following equation:

Interrupt Interval (s) = System Clock Period (s)  $\times$  BRG[15:0]

# **SPI Control Register Definitions**

# SPI Data Register

The SPI Data register (Table 63) stores both the outgoing (transmit) data and the incoming (receive) data. Reads from the SPI Data register always return the current contents of the 8-bit shift register. Data is shifted out starting with bit 7. The last bit received resides in bit position 0.

With the SPI configured as a Master, writing a data byte to this register initiates the data transmission. With the SPI configured as a Slave, writing a data byte to this register loads the shift register in preparation for the next data transfer with the external Master. In either the Master or Slave modes, if a transmission is already in progress, writes to this register are ignored and the Overrun error Flag, OVR, is set in the SPI Status register.

When the character length is less than 8 bits (as set by the NUMBITS field in the SPI Mode register), the transmit character must be left justified in the SPI Data register. A received character of less than 8 bits is right justified (last bit received is in bit position 0). For example, if the SPI is configured for 4-bit characters, the transmit characters must be written to SPIDATA[7:4] and the received characters are read from SPIDATA[3:0].

Table 63. SPI Data Register (SPIDATA)

BITS	7	6	5	4	3	2	1	0			
FIELD				DA	λΤΑ						
RESET		X									
R/W				R/	W						
ADDR				F6	0H						

DATA—Data

Transmit and/or receive data.

# **SPI Control Register**

The SPI Control register (see Table 64 on page 138) configures the SPI for transmit and receive operations.

# Table 64. SPI Control Register (SPICTL)

вітѕ	7	6	5	4	3	2	1	0					
FIELD	IRQE	STR	BIRQ	PHASE	CLKPOL	WOR	MMEN	SPIEN					
RESET		0											
R/W				R/	W								
ADDR				F6	1H								

IRQE—Interrupt Request Enable

- 0 = SPI interrupts are disabled. No interrupt requests are sent to the Interrupt Controller.
- 1 = SPI interrupts are enabled. Interrupt requests are sent to the Interrupt Controller.

STR—Start an SPI Interrupt Request

- 0 = No effect.
- 1 = Setting this bit to 1 also sets the IRQ bit in the SPI Status register to 1. Setting this bit forces the SPI to send an interrupt request to the Interrupt Control. This bit can be used by software for a function similar to transmit buffer empty in a UART. Writing a 1 to the IRQ bit in the SPI Status register clears this bit to 0.

#### BIRQ—BRG Timer Interrupt Request

If the SPI is enabled, this bit has no effect. If the SPI is disabled:

- 0 = The Baud Rate Generator timer function is disabled.
- 1 = The Baud Rate Generator timer function and time-out interrupt are enabled.

#### PHASE—Phase Select

Sets the phase relationship of the data to the clock. For more information on operation of the PHASE bit, see SPI Clock Phase and Polarity Control on page 132.

CLKPOL—Clock Polarity

0 = SCK idles Low (0).

1 = SCK idle High (1).

WOR-Wire-OR (OPEN-DRAIN) Mode Enabled

0 = SPI signal pins not configured for open-drain.

 $1 = \text{All four SPI signal pins (SCK, } \overline{\text{SS}}, \text{MISO, MOSI)}$  configured for open-drain function. This setting is typically used for multi-master and/or multi-slave configurations.

MMEN—SPI Master Mode Enable

0 = SPI configured in Slave mode.

1 = SPI configured in Master mode.

SPIEN—SPI Enable

0 = SPI disabled.

1 = SPI enabled.

# **SPI Status Register**

The SPI Status register (Table 65) indicates the current state of the SPI. All bits revert to their reset state if the SPIEN bit in the SPICTL register = 0.

# Table 65. SPI Status Register (SPISTAT)

BITS	7	6	5	4	3	2	1	0				
FIELD	IRQ	OVR	COL	ABT	Reserved TXST			SLAS				
RESET				0		1						
R/W		R/\	W*				R					
ADDR		F62H										
Note: R/W	e: R/W* = Read access. Write a 1 to clear the bit to 0.											

IRQ—Interrupt Request

If SPIEN = 1, this bit is set if the STR bit in the SPICTL register is set, or upon completion of an SPI master or slave transaction. This bit does not set if SPIEN = 0 and the SPI Baud Rate Generator is used as a timer to generate the SPI interrupt.

0 = No SPI interrupt request pending.

1 = SPI interrupt request is pending.

#### OVR—Overrun

0 = An overrun error has not occurred.

1 = An overrun error has been detected.

#### COL—Collision

0 = A multi-master collision (mode fault) has not occurred.

1 = A multi-master collision (mode fault) has been detected.

#### ABT—Slave mode transaction abort

This bit is set if the SPI is configured in slave mode, a transaction is occurring and  $\overline{SS}$  deasserts before all bits of a character have been transferred as defined by the NUMBITS field of the SPIMODE register. The IRQ bit also sets, indicating the transaction has completed.

0 = A slave mode transaction abort has not occurred.

1 = A slave mode transaction abort has been detected.

Reserved—Must be 0.

#### TXST—Transmit Status

0 = No data transmission currently in progress.

1 = Data transmission currently in progress.

SLAS—Slave Select

If SPI enabled as a Slave,

- $0 = \overline{SS}$  input pin is asserted (Low).
- $1 = \overline{SS}$  input is not asserted (High).

If SPI enabled as a Master, this bit is not applicable.

# **SPI Mode Register**

The SPI Mode register (Table 66) configures the character bit width and the direction and value of the  $\overline{SS}$  pin.

# Table 66. SPI Mode Register (SPIMODE)

BITS	7	6	5	4	3	2	1	0			
FIELD	Rese	erved	DIAG	NUMBITS[2:0]			NUMBITS[2:0] SSIO				SSV
RESET		0									
R/W	F	₹			F	R/W					
ADDR		F63H									

Reserved—Must be 0.

DIAG—Diagnostic Mode Control bit

This bit is for SPI diagnostics. Setting this bit allows the Baud Rate Generator value to be read using the SPIBRH and SPIBRL register locations.

- 0 = Reading SPIBRH, SPIBRL returns the value in the SPIBRH and SPIBRL registers
- 1 = Reading SPIBRH returns bits [15:8] of the SPI Baud Rate Generator; and reading SPIBRL returns bits [7:0] of the SPI Baud Rate Counter. The Baud Rate Counter High and Low byte values are not buffered.



**Caution:** *Exercise caution if reading the values while the BRG is counting.* 

NUMBITS[2:0]—Number of Data Bits Per Character to Transfer

This field contains the number of bits to shift for each character transfer. For information on valid bit positions when the character length is less than 8-bits, see SPI Data Register description.

000 = 8 bits

001 = 1 bit

010 = 2 bits

011 = 3 bits

100 = 4 bits

101 = 5 bits

110 = 6 bits

111 = 7 bits

SSIO—Slave Select I/O

 $0 = \overline{SS}$  pin configured as an input.

 $1 = \overline{SS}$  pin configured as an output (Master mode only).

SSV—Slave Select Value

If SSIO = 1 and SPI configured as a Master:

 $0 = \overline{SS}$  pin driven Low (0).

 $1 = \overline{SS}$  pin driven High (1).

This bit has no effect if SSIO = 0 or SPI configured as a Slave.

# **SPI Diagnostic State Register**

The SPI Diagnostic State register (Table 67) provides observability of internal state. This is a read only register used for SPI diagnostics.

Table 67. SPI Diagnostic State Register (SPIDST)

BITS	7	6	5	5 4 3 2 1 0								
FIELD	SCKEN	TCKEN		SPISTATE								
RESET				0								
R/W				F	₹							
ADDR				F6	4H							

#### SCKEN—Shift Clock Enable

- 0 = The internal Shift Clock Enable signal is deasserted
- 1 = The internal Shift Clock Enable signal is asserted (shift register is updates on next system clock)

#### TCKEN—Transmit Clock Enable

- 0 = The internal Transmit Clock Enable signal is deasserted.
- 1 = The internal Transmit Clock Enable signal is asserted. When this is asserted the serial data out is updated on the next system clock (MOSI or MISO).

# SPISTATE—SPI State Machine

Defines the current state of the internal SPI State Machine.

# **SPI Baud Rate High and Low Byte Registers**

The SPI Baud Rate High and Low Byte registers (Table 68 and Table 69) combine to form a 16-bit reload value, BRG[15:0], for the SPI Baud Rate Generator.

When configured as a general purpose timer, the SPI BRG interrupt interval is calculated using the following equation:

SPI BRG Interrupt Interval (s) = System Clock Period (s)  $\times$  BRG[15:0]

# Table 68. SPI Baud Rate High Byte Register (SPIBRH)

BITS	7	6	5	4	3	2	1	0			
FIELD				BF	RH						
RESET		1									
R/W				R/	W						
ADDR				F6	6H						

BRH = SPI Baud Rate High Byte

Most significant byte, BRG[15:8], of the SPI Baud Rate Generator's reload value.

# Table 69. SPI Baud Rate Low Byte Register (SPIBRL)

BITS	7	6	5	4	3	2	1	0
FIELD				ВГ	RL			
RESET				•	I			
R/W				R/	W			
ADDR				F6	7H			

BRL = SPI Baud Rate Low Byte

Least significant byte, BRG[7:0], of the SPI Baud Rate Generator's reload value.

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# I<sup>2</sup>C Controller

## **Overview**

The  $I^2C$  Controller makes the 64K Series products bus-compatible with the  $I^2C$  protocol. The  $I^2C$  Controller consists of two bidirectional bus lines—a serial data signal (SDA) and a serial clock signal (SCL). Features of the  $I^2C$  Controller include:

- Transmit and Receive Operation in MASTER mode
- Maximum data rate of 400 kbit/sec
- 7- and 10-bit addressing modes for Slaves
- Unrestricted number of data bytes transmitted per transfer

The I<sup>2</sup>C Controller in the 64K Series products does not operate in SLAVE mode.

## **Architecture**

Figure 27 displays the architecture of the I<sup>2</sup>C Controller.

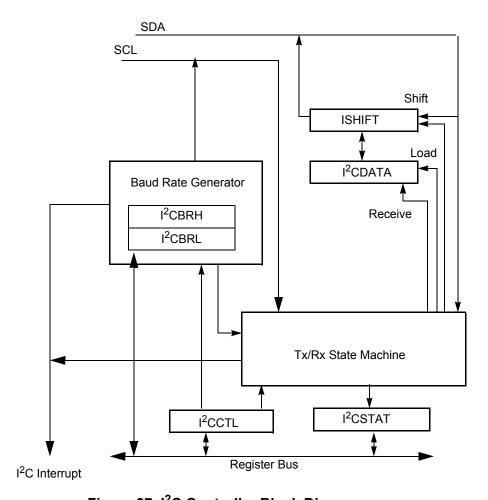


Figure 27. I<sup>2</sup>C Controller Block Diagram

# **Operation**

The I<sup>2</sup>C Controller operates in MASTER mode to transmit and receive data. Only a single master is supported. Arbitration between two masters must be accomplished in software. I<sup>2</sup>C supports the following operations:

- Master transmits to a 7-bit slave
- Master transmits to a 10-bit slave

- Master receives from a 7-bit slave
- Master receives from a 10-bit slave

# SDA and SCL Signals

I<sup>2</sup>C sends all addresses, data and acknowledge signals over the SDA line, most-significant bit first. SCL is the common clock for the I<sup>2</sup>C Controller. When the SDA and SCL pin alternate functions are selected for their respective GPIO ports, the pins are automatically configured for open-drain operation.

The master (I<sup>2</sup>C) is responsible for driving the SCL clock signal, although the clock signal can become skewed by a slow slave device. During the low period of the clock, the slave pulls the SCL signal Low to suspend the transaction. The master releases the clock at the end of the low period and notices that the clock remains low instead of returning to a high level. When the slave releases the clock, the I<sup>2</sup>C Controller continues the transaction. All data is transferred in bytes and there is no limit to the amount of data transferred in one operation. When transmitting data or acknowledging read data from the slave, the SDA signal changes in the middle of the low period of SCL and is sampled in the middle of the high period of SCL.

# I<sup>2</sup>C Interrupts

The I<sup>2</sup>C Controller contains four sources of interrupts—Transmit, Receive, Not Acknowledge and baud rate generator. These four interrupt sources are combined into a single interrupt request signal to the Interrupt Controller. The Transmit interrupt is enabled by the IEN and TXI bits of the Control register. The Receive and Not Acknowledge interrupts are enabled by the IEN bit of the Control register. The baud rate generator interrupt is enabled by the BIRQ and IEN bits of the Control register.

Not Acknowledge interrupts occur when a Not Acknowledge condition is received from the slave or sent by the  $I^2C$  Controller and neither the START or STOP bit is set. The Not Acknowledge event sets the NCKI bit of the  $I^2C$  Status register and can only be cleared by setting the START or STOP bit in the  $I^2C$  Control register. When this interrupt occurs, the  $I^2C$  Controller waits until either the STOP or START bit is set before performing any action. In an interrupt service routine, the NCKI bit should always be checked prior to servicing transmit or receive interrupt conditions because it indicates the transaction is being terminated.

Receive interrupts occur when a byte of data has been received by the I<sup>2</sup>C Controller (master reading data from slave). This procedure sets the RDRF bit of the I<sup>2</sup>C Status register. The RDRF bit is cleared by reading the I<sup>2</sup>C Data register. The RDRF bit is set during the acknowledge phase. The I<sup>2</sup>C Controller pauses after the acknowledge phase until the receive interrupt is cleared before performing any other action.



Transmit interrupts occur when the TDRE bit of the I<sup>2</sup>C Status register sets and the TXI bit in the I<sup>2</sup>C Control register is set. Transmit interrupts occur under the following conditions when the transmit data register is empty:

- The I<sup>2</sup>C Controller is enabled.
- The first bit of the byte of an address is shifting out and the RD bit of the I<sup>2</sup>C Status register is deasserted.
- The first bit of a 10-bit address shifts out.
- The first bit of write data shifts out.

**Note:** Writing to the  $I^2C$  Data register always clears the TRDE bit to 0. When TDRE is asserted, the  $I^2C$  Controller pauses at the beginning of the Acknowledge cycle of the byte currently shifting out until the Data register is written with the next value to send or the STOP or START bits are set indicating the current byte is the last one to send.

> The fourth interrupt source is the baud rate generator. If the I<sup>2</sup>C Controller is disabled (IEN bit in the I2CCTL register = 0) and the BIRQ bit in the I2CCTL register = 1, an interrupt is generated when the baud rate generator counts down to 1. This allows the I<sup>2</sup>C baud rate generator to be used by software as a general purpose timer when IEN = 0.

# Software Control of I<sup>2</sup>C Transactions

Software can control I<sup>2</sup>C transactions by using the I<sup>2</sup>C Controller interrupt, by polling the I<sup>2</sup>C Status register or by DMA. Note that not all products include a DMA Controller.

To use interrupts, the I<sup>2</sup>C interrupt must be enabled in the Interrupt Controller. The TXI bit in the I<sup>2</sup>C Control register must be set to enable transmit interrupts.

To control transactions by polling, the interrupt bits (TDRE, RDRF and NCKI) in the I<sup>2</sup>C Status register should be polled. The TDRE bit asserts regardless of the state of the TXI bit.

Either or both transmit and receive data movement can be controlled by the DMA Controller. The DMA Controller channel(s) must be initialized to select the I<sup>2</sup>C transmit and receive requests. Transmit DMA requests require that the TXI bit in the I<sup>2</sup>C Control register be set.



**Caution:** A transmit (write) DMA operation hangs if the slave responds with a Not Acknowledge before the last byte has been sent. After receiving the Not Acknowledge, the I<sup>2</sup>C Controller sets the NCKI bit in the Status register and pauses until either the STOP or START bits in the Control register are set.



In order for a receive (read) DMA transaction to send a Not Acknowledge on the last byte, the receive DMA must be set up to receive n-1 bytes, then software must set the NAK bit and receive the last (nth) byte directly.

# **Start and Stop Conditions**

The master (I<sup>2</sup>C) drives all Start and Stop signals and initiates all transactions. To start a transaction, the I<sup>2</sup>C Controller generates a START condition by pulling the SDA signal Low while SCL is High. To complete a transaction, the I<sup>2</sup>C Controller generates a Stop condition by creating a low-to-high transition of the SDA signal while the SCL signal is high. The START and STOP bits in the I<sup>2</sup>C Control register control the sending of the Start and Stop conditions. A master is also allowed to end one transaction and begin a new one by issuing a Restart. This is accomplished by setting the START bit at the end of a transaction, rather than the STOP bit. Note that the Start condition not sent until the START bit is set and data has been written to the I<sup>2</sup>C Data register.

#### **Master Write and Read Transactions**

The following sections provide a recommended procedure for performing I<sup>2</sup>C write and read transactions from the I<sup>2</sup>C Controller (master) to slave I<sup>2</sup>C devices. In general software should rely on the TDRE, RDRF and NCKI bits of the status register (these bits generate interrupts) to initiate software actions. When using interrupts or DMA, the TXI bit is set to start each transaction and cleared at the end of each transaction to eliminate a 'trailing' Transmit interrupt.

Caution should be used in using the ACK status bit within a transaction because it is difficult for software to tell when it is updated by hardware.

When writing data to a slave, the  $I^2C$  pauses at the beginning of the Acknowledge cycle if the data register has not been written with the next value to be sent (TDRE bit in the  $I^2C$  Status register = 1). In this scenario where software is not keeping up with the  $I^2C$  bus (TDRE asserted longer than one byte time), the Acknowledge clock cycle for byte n is delayed until the Data register is written with byte n + 1, and appears to be grouped with the data clock cycles for byte n+1. If either the START or STOP bit is set, the  $I^2C$  does not pause prior to the Acknowledge cycle because no additional data is sent.

When a Not Acknowledge condition is received during a write (either during the address or data phases), the  $I^2C$  Controller generates the Not Acknowledge interrupt (NCKI = 1) and pause until either the STOP or START bit is set. Unless the Not Acknowledge was received on the last byte, the Data register will already have been written with the next address or data byte to send. In this case the FLUSH bit of the Control register should be set at the same time the STOP or START bit is set to remove the stale transmit data and enable subsequent Transmit interrupts.

When reading data from the slave, the I<sup>2</sup>C pauses after the data Acknowledge cycle until the receive interrupt is serviced and the RDRF bit of the status register is cleared by

reading the I<sup>2</sup>C Data register. Once the I<sup>2</sup>C data register has been read, the I<sup>2</sup>C reads the next data byte.

# Address Only Transaction with a 7-bit Address

In the situation where software determines if a slave with a 7-bit address is responding without sending or receiving data, a transaction can be done which only consists of an address phase. Figure 28 displays this 'address only' transaction to determine if a slave with a 7-bit address will acknowledge. As an example, this transaction can be used after a 'write' has been done to a EEPROM to determine when the EEPROM completes its internal write operation and is once again responding to I<sup>2</sup>C transactions. If the slave does not Acknowledge, the transaction can be repeated until the slave does Acknowledge.



Figure 28. 7-Bit Address Only Transaction Format

Follow the steps below for an address only transaction to a 7-bit addressed slave:

- 1. Software asserts the IEN bit in the I<sup>2</sup>C Control register.
- 2. Software asserts the TXI bit of the  $I^2C$  Control register to enable Transmit interrupts.
- 3. The  $I^2C$  interrupt asserts, because the  $I^2C$  Data register is empty (TDRE = 1)
- 4. Software responds to the TDRE bit by writing a 7-bit slave address plus write bit (=0) to the I<sup>2</sup>C Data register. As an alternative this could be a read operation instead of a write operation.
- 5. Software sets the START and STOP bits of the I<sup>2</sup>C Control register and clears the TXI bit.
- 6. The I<sup>2</sup>C Controller sends the START condition to the I<sup>2</sup>C slave.
- 7. The I<sup>2</sup>C Controller loads the I<sup>2</sup>C Shift register with the contents of the I<sup>2</sup>C Data register.
- 8. Software polls the STOP bit of the I<sup>2</sup>C Control register. Hardware deasserts the STOP bit when the address only transaction is completed.
- 9. Software checks the ACK bit of the I<sup>2</sup>C Status register. If the slave acknowledged, the ACK bit is = 1. If the slave does not acknowledge, the ACK bit is = 0. The NCKI interrupt does not occur in the not acknowledge case because the STOP bit was set.

#### Write Transaction with a 7-Bit Address

Figure 29 displays the data transfer format for a 7-bit addressed slave. Shaded regions indicate data transferred from the I<sup>2</sup>C Controller to slaves and unshaded regions indicate data transferred from the slaves to the I<sup>2</sup>C Controller.

S	Slave Address	W = 0	Α	Data	A	Data	Α	Data	A/A	P/S

Figure 29. 7-Bit Addressed Slave Data Transfer Format

Follow the steps below for a transmit operation to a 7-bit addressed slave:

- 1. Software asserts the IEN bit in the I<sup>2</sup>C Control register.
- 2. Software asserts the TXI bit of the I<sup>2</sup>C Control register to enable Transmit interrupts.
- 3. The I<sup>2</sup>C interrupt asserts, because the I<sup>2</sup>C Data register is empty
- 4. Software responds to the TDRE bit by writing a 7-bit slave address plus write bit (=0) to the  $I^2C$  Data register.
- 5. Software asserts the START bit of the I<sup>2</sup>C Control register.
- 6. The I<sup>2</sup>C Controller sends the START condition to the I<sup>2</sup>C slave.
- 7. The I<sup>2</sup>C Controller loads the I<sup>2</sup>C Shift register with the contents of the I<sup>2</sup>C Data register.
- 8. After one bit of address has been shifted out by the SDA signal, the Transmit interrupt is asserted (TDRE = 1).
- 9. Software responds by writing the transmit data into the I<sup>2</sup>C Data register.
- 10. The I<sup>2</sup>C Controller shifts the rest of the address and write bit out by the SDA signal.
- 11. If the I<sup>2</sup>C slave sends an acknowledge (by pulling the SDA signal low) during the next high period of SCL the I<sup>2</sup>C Controller sets the ACK bit in the I<sup>2</sup>C Status register. Continue with step 12.

If the slave does not acknowledge, the Not Acknowledge interrupt occurs (NCKI bit is set in the Status register, ACK bit is cleared). Software responds to the Not Acknowledge interrupt by setting the STOP and FLUSH bits and clearing the TXI bit. The I<sup>2</sup>C Controller sends the STOP condition on the bus and clears the STOP and NCKI bits. The transaction is complete (ignore the following steps).

12. The I<sup>2</sup>C Controller loads the contents of the I<sup>2</sup>C Shift register with the contents of the I<sup>2</sup>C Data register.

- 13. The I<sup>2</sup>C Controller shifts the data out of using the SDA signal. After the first bit is sent, the Transmit interrupt is asserted.
- 14. If more bytes remain to be sent, return to step 9.
- 15. Software responds by setting the STOP bit of the I<sup>2</sup>C Control register (or START bit to initiate a new transaction). In the STOP case, software clears the TXI bit of the I<sup>2</sup>C Control register at the same time.
- 16. The I<sup>2</sup>C Controller completes transmission of the data on the SDA signal.
- 17. The slave may either Acknowledge or Not Acknowledge the last byte. Because either the STOP or START bit is already set, the NCKI interrupt does not occur.
- 18. The I<sup>2</sup>C Controller sends the STOP (or RESTART) condition to the I<sup>2</sup>C bus. The STOP or START bit is cleared.

# Address Only Transaction with a 10-bit Address

In the situation where software wants to determine if a slave with a 10-bit address is responding without sending or receiving data, a transaction can be done which only consists of an address phase. Figure 30 displays this 'address only' transaction to determine if a slave with 10-bit address will acknowledge. As an example, this transaction can be used after a 'write' has been done to a EEPROM to determine when the EEPROM completes its internal write operation and is once again responding to I<sup>2</sup>C transactions. If the slave does not Acknowledge the transaction can be repeated until the slave is able to Acknowledge.

s	Slave Address 1st 7 bits	W = 0	A/A	Slave Address 2nd Byte	A/A	Р
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Figure 30. 10-Bit Address Only Transaction Format

Follow the steps below for an address only transaction to a 10-bit addressed slave:

- 1. Software asserts the IEN bit in the I<sup>2</sup>C Control register.
- 2. Software asserts the TXI bit of the I<sup>2</sup>C Control register to enable Transmit interrupts.
- 3. The  $I^2C$  interrupt asserts, because the  $I^2C$  Data register is empty (TDRE = 1)
- 4. Software responds to the TDRE interrupt by writing the first slave address byte. The least-significant bit must be 0 for the write operation.
- 5. Software asserts the START bit of the I<sup>2</sup>C Control register.
- 6. The I<sup>2</sup>C Controller sends the START condition to the I<sup>2</sup>C slave.

- 7. The I<sup>2</sup>C Controller loads the I<sup>2</sup>C Shift register with the contents of the I<sup>2</sup>C Data register.
- 8. After one bit of address is shifted out by the SDA signal, the Transmit interrupt is asserted
- 9. Software responds by writing the second byte of address into the contents of the I<sup>2</sup>C Data register.
- 10. The I<sup>2</sup>C Controller shifts the rest of the first byte of address and write bit out the SDA signal.
- 11. If the I<sup>2</sup>C slave sends an acknowledge by pulling the SDA signal low during the next high period of SCL the I<sup>2</sup>C Controller sets the ACK bit in the I<sup>2</sup>C Status register. Continue with step 12.

If the slave does not acknowledge the first address byte, the I<sup>2</sup>C Controller sets the NCKI bit and clears the ACK bit in the I<sup>2</sup>C Status register. Software responds to the Not Acknowledge interrupt by setting the STOP and FLUSH bits and clearing the TXI bit. The I<sup>2</sup>C Controller sends the STOP condition on the bus and clears the STOP and NCKI bits. The transaction is complete (ignore following steps).

- 12. The I<sup>2</sup>C Controller loads the I<sup>2</sup>C Shift register with the contents of the I<sup>2</sup>C Data register (2nd byte of address).
- 13. The I<sup>2</sup>C Controller shifts the second address byte out the SDA signal. After the first bit has been sent, the Transmit interrupt is asserted.
- 14. Software responds by setting the STOP bit in the I<sup>2</sup>C Control register. The TXI bit can be cleared at the same time.
- 15. Software polls the STOP bit of the I<sup>2</sup>C Control register. Hardware deasserts the STOP bit when the transaction is completed (STOP condition has been sent).
- 16. Software checks the ACK bit of the  $I^2C$  Status register. If the slave acknowledged, the ACK bit is = 1. If the slave does not acknowledge, the ACK bit is = 0. The NCKI interrupt do not occur because the STOP bit was set.

# Write Transaction with a 10-Bit Address

Figure 31 displays the data transfer format for a 10-bit addressed slave. Shaded regions indicate data transferred from the I<sup>2</sup>C Controller to slaves and unshaded regions indicate data transferred from the slaves to the I<sup>2</sup>C Controller.

s	Slave Address 1st 7 bits	W = 0	Α	Slave Address 2nd Byte	Α	Data	Α	Data	A/A	P/S	
---	-----------------------------	-------	---	---------------------------	---	------	---	------	-----	-----	--

Figure 31. 10-Bit Addressed Slave Data Transfer Format

The first seven bits transmitted in the first byte are 11110xx. The two bits xx are the two most-significant bits of the 10-bit address. The lowest bit of the first byte transferred is the read/write control bit (=0). The transmit operation is carried out in the same manner as 7-bit addressing.

Follow the steps below for a transmit operation on a 10-bit addressed slave:

- 1. Software asserts the IEN bit in the I<sup>2</sup>C Control register.
- 2. Software asserts the TXI bit of the I<sup>2</sup>C Control register to enable Transmit interrupts.
- 3. The I<sup>2</sup>C interrupt asserts because the I<sup>2</sup>C Data register is empty.
- 4. Software responds to the TDRE interrupt by writing the first slave address byte to the I<sup>2</sup>C Data register. The least-significant bit must be 0 for the write operation.
- 5. Software asserts the START bit of the I<sup>2</sup>C Control register.
- 6. The I<sup>2</sup>C Controller sends the START condition to the I<sup>2</sup>C slave.
- 7. The I<sup>2</sup>C Controller loads the I<sup>2</sup>C Shift register with the contents of the I<sup>2</sup>C Data register.
- 8. After one bit of address is shifted out by the SDA signal, the Transmit interrupt is asserted.
- 9. Software responds by writing the second byte of address into the contents of the I<sup>2</sup>C Data register.
- 10. The I<sup>2</sup>C Controller shifts the rest of the first byte of address and write bit out the SDA signal.
- 11. If the I<sup>2</sup>C slave acknowledges the first address byte by pulling the SDA signal low during the next high period of SCL, the I<sup>2</sup>C Controller sets the ACK bit in the I<sup>2</sup>C Status register. Continue with step 12.

If the slave does not acknowledge the first address byte, the I<sup>2</sup>C Controller sets the NCKI bit and clears the ACK bit in the I<sup>2</sup>C Status register. Software responds to the Not Acknowledge interrupt by setting the STOP and FLUSH bits and clearing the TXI bit. The I<sup>2</sup>C Controller sends the STOP condition on the bus and clears the STOP and NCKI bits. The transaction is complete (ignore the following steps).

- 12. The I<sup>2</sup>C Controller loads the I<sup>2</sup>C Shift register with the contents of the I<sup>2</sup>C Data register.
- 13. The I<sup>2</sup>C Controller shifts the second address byte out the SDA signal. After the first bit has been sent, the Transmit interrupt is asserted.
- 14. Software responds by writing a data byte to the I<sup>2</sup>C Data register.
- 15. The I<sup>2</sup>C Controller completes shifting the contents of the shift register on the SDA signal.

16. If the I<sup>2</sup>C slave sends an acknowledge by pulling the SDA signal low during the next high period of SCL, the I<sup>2</sup>C Controller sets the ACK bit in the I<sup>2</sup>C Status register. Continue with step 17.

If the slave does not acknowledge the second address byte or one of the data bytes, the  $I^2C$  Controller sets the NCKI bit and clears the ACK bit in the  $I^2C$  Status register. Software responds to the Not Acknowledge interrupt by setting the STOP and FLUSH bits and clearing the TXI bit. The  $I^2C$  Controller sends the STOP condition on the bus and clears the STOP and NCKI bits. The transaction is complete (ignore the following steps).

- 17. The I<sup>2</sup>C Controller shifts the data out by the SDA signal. After the first bit is sent, the Transmit interrupt is asserted.
- 18. If more bytes remain to be sent, return to step 14.
- 19. If the last byte is currently being sent, software sets the STOP bit of the I<sup>2</sup>C Control register (or START bit to initiate a new transaction). In the STOP case, software also clears the TXI bit of the I<sup>2</sup>C Control register at the same time.
- 20. The I<sup>2</sup>C Controller completes transmission of the last data byte on the SDA signal.
- 21. The slave may either Acknowledge or Not Acknowledge the last byte. Because either the STOP or START bit is already set, the NCKI interrupt does not occur.
- 22. The I<sup>2</sup>C Controller sends the STOP (or RESTART) condition to the I<sup>2</sup>C bus and clears the STOP (or START) bit.

#### Read Transaction with a 7-Bit Address

Figure 32 displays the data transfer format for a read operation to a 7-bit addressed slave. The shaded regions indicate data transferred from the  $I^2C$  Controller to slaves and unshaded regions indicate data transferred from the slaves to the  $I^2C$  Controller.

S	Slave Address	R = 1	Α	Data	Α	Data	Ā	P/S

Figure 32. Receive Data Transfer Format for a 7-Bit Addressed Slave

Follow the steps below for a read operation to a 7-bit addressed slave:

- 1. Software writes the I<sup>2</sup>C Data register with a 7-bit slave address plus the read bit (=1).
- 2. Software asserts the START bit of the I<sup>2</sup>C Control register.
- 3. If this is a single byte transfer, Software asserts the NAK bit of the I<sup>2</sup>C Control register so that after the first byte of data has been read by the I<sup>2</sup>C Controller, a Not Acknowledge is sent to the I<sup>2</sup>C slave.

- 4. The I<sup>2</sup>C Controller sends the START condition.
- 5. The I<sup>2</sup>C Controller shifts the address and read bit out the SDA signal.
- 6. If the I<sup>2</sup>C slave acknowledges the address by pulling the SDA signal Low during the next high period of SCL, the I<sup>2</sup>C Controller sets the ACK bit in the I<sup>2</sup>C Status register. Continue with step 7.

If the slave does not acknowledge, the Not Acknowledge interrupt occurs (NCKI bit is set in the Status register, ACK bit is cleared). Software responds to the Not Acknowledge interrupt by setting the STOP bit and clearing the TXI bit. The I<sup>2</sup>C Controller sends the STOP condition on the bus and clears the STOP and NCKI bits. The transaction is complete (ignore the following steps).

- 7. The  $I^2C$  Controller shifts in the byte of data from the  $I^2C$  slave on the SDA signal. The I<sup>2</sup>C Controller sends a Not Acknowledge to the I<sup>2</sup>C slave if the NAK bit is set (last byte), else it sends an Acknowledge.
- 8. The I<sup>2</sup>C Controller asserts the Receive interrupt (RDRF bit set in the Status register).
- 9. Software responds by reading the I<sup>2</sup>C Data register which clears the RDRF bit. If there is only one more byte to receive, set the NAK bit of the I<sup>2</sup>C Control register.
- 10. If there are more bytes to transfer, return to step 7.
- 11. After the last byte is shifted in, a Not Acknowledge interrupt is generated by the I<sup>2</sup>C Controller.
- 12. Software responds by setting the STOP bit of the I<sup>2</sup>C Control register.
- 13. A STOP condition is sent to the I<sup>2</sup>C slave, the STOP and NCKI bits are cleared.

#### Read Transaction with a 10-Bit Address

Figure 33 displays the read transaction format for a 10-bit addressed slave. The shaded regions indicate data transferred from the I<sup>2</sup>C Controller to slaves and unshaded regions indicate data transferred from the slaves to the I<sup>2</sup>C Controller

s	Slave Address 1st 7 bits	W=0	Α	Slave Address 2nd Byte	A	s	Slave Address 1st 7 bits	R=1	Α	Data	A	Data	Ā	Р	
---	-----------------------------	-----	---	---------------------------	---	---	-----------------------------	-----	---	------	---	------	---	---	--

Figure 33. Receive Data Format for a 10-Bit Addressed Slave

The first seven bits transmitted in the first byte are 11110xx. The two bits xx are the two most-significant bits of the 10-bit address. The lowest bit of the first byte transferred is the write control bit.

Follow the steps below for the data transfer for a read operation to a 10-bit addressed slave:

- 1. Software writes 11110B followed by the two address bits and a 0 (write) to the I<sup>2</sup>C Data register.
- 2. Software asserts the START and TXI bits of the I<sup>2</sup>C Control register.
- 3. The I<sup>2</sup>C Controller sends the Start condition.
- 4. The I<sup>2</sup>C Controller loads the I<sup>2</sup>C Shift register with the contents of the I<sup>2</sup>C Data register.
- 5. After the first bit has been shifted out, a Transmit interrupt is asserted.
- 6. Software responds by writing the lower eight bits of address to the I<sup>2</sup>C Data register.
- 7. The  $I^2C$  Controller completes shifting of the two address bits and a 0 (write).
- 8. If the I<sup>2</sup>C slave acknowledges the first address byte by pulling the SDA signal low during the next high period of SCL, the I<sup>2</sup>C Controller sets the ACK bit in the I<sup>2</sup>C Status register. Continue with step 9.

If the slave does not acknowledge the first address byte, the I<sup>2</sup>C Controller sets the NCKI bit and clears the ACK bit in the I<sup>2</sup>C Status register. Software responds to the Not Acknowledge interrupt by setting the STOP and FLUSH bits and clearing the TXI bit. The I<sup>2</sup>C Controller sends the STOP condition on the bus and clears the STOP and NCKI bits. The transaction is complete (ignore following steps).

- 9. The I<sup>2</sup>C Controller loads the I<sup>2</sup>C Shift register with the contents of the I<sup>2</sup>C Data register (second address byte).
- 10. The I<sup>2</sup>C Controller shifts out the second address byte. After the first bit is shifted, the I<sup>2</sup>C Controller generates a Transmit interrupt.
- 11. Software responds by setting the START bit of the I<sup>2</sup>C Control register to generate a repeated START and by clearing the TXI bit.
- 12. Software responds by writing 11110B followed by the 2-bit slave address and a 1 (read) to the I<sup>2</sup>C Data register.
- 13. If only one byte is to be read, software sets the NAK bit of the I<sup>2</sup>C Control register.
- 14. After the I<sup>2</sup>C Controller shifts out the 2nd address byte, the I<sup>2</sup>C slave sends an acknowledge by pulling the SDA signal low during the next high period of SCL, the I<sup>2</sup>C Controller sets the ACK bit in the I<sup>2</sup>C Status register. Continue with step 15.

If the slave does not acknowledge the second address byte, the  $I^2C$  Controller sets the NCKI bit and clears the ACK bit in the  $I^2C$  Status register. Software responds to the Not Acknowledge interrupt by setting the STOP and FLUSH bits and clearing the TXI bit. The  $I^2C$  Controller sends the STOP condition on the bus and clears the STOP and NCKI bits. The transaction is complete (ignore the following steps).

- 15. The I<sup>2</sup>C Controller sends the repeated START condition.
- 16. The I<sup>2</sup>C Controller loads the I<sup>2</sup>C Shift register with the contents of the I<sup>2</sup>C Data register (third address transfer).
- 17. The I<sup>2</sup>C Controller sends 11110B followed by the two most significant bits of the slave read address and a 1 (read).
- 18. The I<sup>2</sup>C slave sends an acknowledge by pulling the SDA signal Low during the next high period of SCL

If the slave were to Not Acknowledge at this point (this should not happen because the slave did acknowledge the first two address bytes), software would respond by setting the STOP and FLUSH bits and clearing the TXI bit. The I<sup>2</sup>C Controller sends the STOP condition on the bus and clears the STOP and NCKI bits. The transaction is complete (ignore the following steps).

- 19. The I<sup>2</sup>C Controller shifts in a byte of data from the I<sup>2</sup>C slave on the SDA signal. The I<sup>2</sup>C Controller sends a Not Acknowledge to the I<sup>2</sup>C slave if the NAK bit is set (last byte), else it sends an Acknowledge.
- 20. The I<sup>2</sup>C Controller asserts the Receive interrupt (RDRF bit set in the Status register).
- 21. Software responds by reading the I<sup>2</sup>C Data register which clears the RDRF bit. If there is only one more byte to receive, set the NAK bit of the I<sup>2</sup>C Control register.
- 22. If there are one or more bytes to transfer, return to step 19.
- 23. After the last byte is shifted in, a Not Acknowledge interrupt is generated by the I<sup>2</sup>C Controller.
- 24. Software responds by setting the STOP bit of the I<sup>2</sup>C Control register.
- 25. A STOP condition is sent to the I<sup>2</sup>C slave and the STOP and NCKI bits are cleared.

# I<sup>2</sup>C Control Register Definitions

# I<sup>2</sup>C Data Register

The I<sup>2</sup>C Data register (see Table 70 on page 157) holds the data that is to be loaded into the I<sup>2</sup>C Shift register during a write to a slave. This register also holds data that is loaded from the I<sup>2</sup>C Shift register during a read from a slave. The I<sup>2</sup>C Shift Register is not accessible in the Register File address space, but is used only to buffer incoming and outgoing data.

Table 70. I<sup>2</sup>C Data Register (I2CDATA)

BITS	7	6	5	4	3	2	1	0			
FIELD	DATA										
RESET	0										
R/W	R/W										
ADDR				F5	0H						

# I<sup>2</sup>C Status Register

The Read-only I<sup>2</sup>C Status register (Table 71) indicates the status of the I<sup>2</sup>C Controller.

Table 71. I<sup>2</sup>C Status Register (I2CSTAT)

BITS	7	6	5	4	3	2	1	0			
FIELD	TDRE	RDRF	RDRF ACK 10B RD TAS DSS NCKI								
RESET	1		0								
R/W		R									
ADDR				F5	1H						

## TDRE—Transmit Data Register Empty

When the  $I^2C$  Controller is enabled, this bit is 1 when the  $I^2C$  Data register is empty. When this bit is set, an interrupt is generated if the TXI bit is set, except when the  $I^2C$  Controller is shifting in data during the reception of a byte or when shifting an address and the RD bit is set. This bit is cleared by writing to the I2CDATA register.

#### RDRF—Receive Data Register Full

This bit is set = 1 when the  $I^2C$  Controller is enabled and the  $I^2C$  Controller has received a byte of data. When asserted, this bit causes the  $I^2C$  Controller to generate an interrupt. This bit is cleared by reading the  $I^2C$  Data register (unless the read is performed using execution of the On-Chip Debugger's Read Register command).

#### ACK—Acknowledge

This bit indicates the status of the Acknowledge for the last byte transmitted or received. When set, this bit indicates that an Acknowledge occurred for the last byte transmitted or received. This bit is cleared when IEN = 0 or when a Not Acknowledge occurred for the last byte transmitted or received. It is not reset at the beginning of each transaction and is not reset when this register is read.



#### Caution:

Software must be cautious in making decisions based on this bit within a transaction because software cannot tell when the bit is updated by hardware. In the case of write transactions, the  $l^2C$  pauses at the beginning of the Acknowledge cycle if the next transmit data or address byte has not been written (TDRE = 1) and STOP and START = 0. In this case the ACK bit is not updated until the transmit interrupt is serviced and the Acknowledge cycle for the previous byte completes. For examples of how the ACK bit can be used, see Address Only Transaction with a 7-bit Address on page 148 and Address Only Transaction with a 10-bit Address on page 150.

#### 10B—10-Bit Address

This bit indicates whether a 10- or 7-bit address is being transmitted. After the START bit is set, if the five most-significant bits of the address are 11110B, this bit is set. When set, it is reset once the first byte of the address has been sent.

#### RD—Read

This bit indicates the direction of transfer of the data. It is active high during a read. The status of this bit is determined by the least-significant bit of the I<sup>2</sup>C Shift register after the START bit is set.

#### TAS—Transmit Address State

This bit is active high while the address is being shifted out of the I<sup>2</sup>C Shift register.

#### DSS—Data Shift State

This bit is active high while data is being shifted to or from the I<sup>2</sup>C Shift register.

#### NCKI—NACK Interrupt

This bit is set high when a Not Acknowledge condition is received or sent and neither the START nor the STOP bit is active. When set, this bit generates an interrupt that can only be cleared by setting the START or STOP bit, allowing you to specify whether to perform a STOP or a repeated START.

# I<sup>2</sup>C Control Register

The I<sup>2</sup>C Control register (Table 72) enables the I<sup>2</sup>C operation.

# Table 72. I<sup>2</sup>C Control Register (I2CCTL)

BITS	7	6	5	4	3	2	1	0		
FIELD	IEN	IEN START STOP BIRQ TXI NAK FLUSH FILTEN								
RESET	0									
R/W	R/W	R/W R/W1 R/W1 R/W R/W1 W1 R/W								
ADDR		F52H								

# IEN—I<sup>2</sup>C Enable

 $1 = \text{The } I^2C$  transmitter and receiver are enabled.

 $0 = \text{The I}^2\text{C}$  transmitter and receiver are disabled.

#### START—Send Start Condition

This bit sends the Start condition. Once asserted, it is cleared by the I<sup>2</sup>C Controller after it sends the START condition or if the IEN bit is deasserted. If this bit is 1, it cannot be cleared to 0 by writing to the register. After this bit is set, the Start condition is sent if there is data in the I<sup>2</sup>C Data or I<sup>2</sup>C Shift register. If there is no data in one of these registers, the I<sup>2</sup>C Controller waits until the Data register is written. If this bit is set while the I<sup>2</sup>C Controller is shifting out data, it generates a START condition after the byte shifts and the acknowledge phase completes. If the STOP bit is also set, it also waits until the STOP condition is sent before the sending the START condition.

#### STOP—Send Stop Condition

This bit causes the I<sup>2</sup>C Controller to issue a Stop condition after the byte in the I<sup>2</sup>C Shift register has completed transmission or after a byte has been received in a receive operation. Once set, this bit is reset by the I<sup>2</sup>C Controller after a Stop condition has been sent or by deasserting the IEN bit. If this bit is 1, it cannot be cleared to 0 by writing to the register.

#### BIRQ—Baud Rate Generator Interrupt Request

This bit allows the I<sup>2</sup>C Controller to be used as an additional timer when the I<sup>2</sup>C Controller is disabled. This bit is ignored when the I<sup>2</sup>C Controller is enabled.

1 = An interrupt occurs every time the baud rate generator counts down to one.

0 = No band rate generator interrupt occurs.

#### TXI—Enable TDRE interrupts

This bit enables the transmit interrupt when the  $I^2C$  Data register is empty (TDRE = 1).

- 1 = Transmit interrupt (and DMA transmit request) is enabled.
- 0 = Transmit interrupt (and DMA transmit request) is disabled.

#### NAK—Send NAK

This bit sends a Not Acknowledge condition after the next byte of data has been read from the I<sup>2</sup>C slave. Once asserted, it is deasserted after a Not Acknowledge is sent or the IEN bit is deasserted. If this bit is 1, it cannot be cleared to 0 by writing to the register.

#### FLUSH—Flush Data

Setting this bit to 1 clears the I<sup>2</sup>C Data register and sets the TDRE bit to 1. This bit allows flushing of the I<sup>2</sup>C Data register when a Not Acknowledge interrupt is received after the data has been sent to the I<sup>2</sup>C Data register. Reading this bit always returns 0.

# FILTEN—I<sup>2</sup>C Signal Filter Enable

This bit enables low-pass digital filters on the SDA and SCL input signals. These filters reject any input pulse with periods less than a full system clock cycle. The filters introduce a 3-system clock cycle latency on the inputs.

- 1 = low-pass filters are enabled.
- 0 = low-pass filters are disabled.

# I<sup>2</sup>C Baud Rate High and Low Byte Registers

The I<sup>2</sup>C Baud Rate High and Low Byte registers (Tables 73 and 73) combine to form a 16-bit reload value, BRG[15:0], for the I<sup>2</sup>C Baud Rate Generator.

When the I<sup>2</sup>C is disabled, the Baud Rate Generator can function as a basic 16-bit timer with interrupt on time-out. To configure the Baud Rate Generator as a timer with interrupt on time-out, complete the following procedure:

- 1. Disable the  $I^2C$  by clearing the IEN bit in the  $I^2C$  Control register to 0.
- 2. Load the desired 16-bit count value into the I<sup>2</sup>C Baud Rate High and Low Byte registers.
- 3. Enable the Baud Rate Generator timer function and associated interrupt by setting the BIRQ bit in the  $I^2C$  Control register to 1.

When configured as a general purpose timer, the interrupt interval is calculated using the following equation:

Interrupt Interval (s) = System Clock Period (s)  $\times$  BRG[15:0]

Table 73. I<sup>2</sup>C Baud Rate High Byte Register (I2CBRH)

BITS	7	6	5	4	3	2	1	0			
FIELD	BRH										
RESET	FFH										
R/W		R/W									
ADDR				F5	3H						

BRH =  $I^2C$  Baud Rate High Byte Most significant byte, BRG[15:8], of the  $I^2C$  Baud Rate Generator's reload value.

**Note:** If the DIAG bit in the  $l^2C$  Diagnostic Control Register is set to 1, a read of the  $l^2C$  BRH register returns the current value of the  $l^2C$  Baud Rate Counter[15:8].

Table 74. I<sup>2</sup>C Baud Rate Low Byte Register (I2CBRL)

BITS	7	6	5	4	3	2	1	0		
FIELD	BRL									
RESET	FFH									
R/W	R/W									
ADDR				F5	4H					

BRL =  $I^2C$  Baud Rate Low Byte Least significant byte, BRG[7:0], of the  $I^2C$  Baud Rate Generator's reload value.

**Note:** If the DIAG bit in the I<sup>2</sup>C Diagnostic Control Register is set to 1, a read of the I2CBRL register returns the current value of the I<sup>2</sup>C Baud Rate Counter[7:0].

# I<sup>2</sup>C Diagnostic State Register

The  $I^2C$  Diagnostic State register (Table 75) provides observability of internal state. This is a read only register used for  $I^2C$  diagnostics and manufacturing test.

Table 75. I<sup>2</sup>C Diagnostic State Register (I2CDST)

BITS	7	6	5	4 3 2 1 0					
FIELD	SCLIN	SDAIN	STPCNT	TXRXSTATE					
RESET	>	(	0						
R/W	R								
ADDR	F55H								

SCLIN—Value of Serial Clock input signal

SDAIN—Value of the Serial Data input signal

STPCNT—Value of the internal Stop Count control signal

TXRXSTATE—Value of the internal I<sup>2</sup>C state machine

TXRXSTATE	State Description
0_0000	Idle State
0_0001	START State
0_0010	Send/Receive data bit 7
0_0011	Send/Receive data bit 6
0_0100	Send/Receive data bit 5
0_0101	Send/Receive data bit 4
0_0110	Send/Receive data bit 3
0_0111	Send/Receive data bit 2
0_1000	Send/Receive data bit 1
0_1001	Send/Receive data bit 0
0_1010	Data Acknowledge State
0_1011	Second half of data Acknowledge State used only for not acknowledge
0_1100	First part of STOP state
0_1101	Second part of STOP state
0_1110	10-bit addressing: Acknowledge State for 2nd address byte 7-bit addressing: Address Acknowledge State
0_1111	10-bit address: Bit 0 (Least significant bit) of 2nd address byte 7-bit address: Bit 0 (Least significant bit) (R/W) of address byte
1_0000	10-bit addressing: Bit 7 (Most significant bit) of 1st address byte
1_0001	10-bit addressing: Bit 6 of 1st address byte
1_0010	10-bit addressing: Bit 5 of 1st address byte
1_0011	10-bit addressing: Bit 4 of 1st address byte
1_0100	10-bit addressing: Bit 3 of 1st address byte
1_0101	10-bit addressing: Bit 2 of 1st address byte
1_0110	10-bit addressing: Bit 1 of 1st address byte
1_0111	10-bit addressing: Bit 0 (R/W) of 1st address byte
1_1000	10-bit addressing: Acknowledge state for 1st address byte
1_1001	10-bit addressing: Bit 7 of 2nd address byte 7-bit addressing: Bit 7 of address byte
1_1010	10-bit addressing: Bit 6 of 2nd address byte 7-bit addressing: Bit 6 of address byte
1_1011	10-bit addressing: Bit 5 of 2nd address byte 7-bit addressing: Bit 5 of address byte
1_1100	10-bit addressing: Bit 4 of 2nd address byte 7-bit addressing: Bit 4 of address byte

TXRXSTATE	State Description
1_1101	10-bit addressing: Bit 3 of 2nd address byte 7-bit addressing: Bit 3 of address byte
1_1110	10-bit addressing: Bit 2 of 2nd address byte 7-bit addressing: Bit 2 of address byte
1_1111	10-bit addressing: Bit 1 of 2nd address byte 7-bit addressing: Bit 1 of address byte

# I<sup>2</sup>C Diagnostic Control Register

The  $I^2C$  Diagnostic register (Table 76) provides control over diagnostic modes. This register is a read/write register used for  $I^2C$  diagnostics.

Table 76. I<sup>2</sup>C Diagnostic Control Register (I2CDIAG)

BITS	7	6	5	4	3	2	1	0		
FIELD	Reserved									
RESET	0									
R/W	R R/W									
ADDR				F5	6H					

DIAG = Diagnostic Control Bit - Selects read back value of the Baud Rate Reload registers

- 0 = NORMAL mode. Reading the Baud Rate High and Low Byte registers returns the baud rate reload value.
- 1 = DIAGNOSTIC mode. Reading the Baud Rate High and Low Byte registers returns the baud rate counter value.

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# **Direct Memory Access Controller**

### **Overview**

The 64K Series Direct Memory Access (DMA) Controller provides three independent Direct Memory Access channels. Two of the channels (DMA0 and DMA1) transfer data between the on-chip peripherals and the Register File. The third channel (DMA\_ADC) controls the ADC operation and transfers SINGLE-SHOT mode ADC output data to the Register File.

# Operation

# **DMA0** and **DMA1** Operation

DMA0 and DMA1, referred to collectively as DMAx, transfer data either from the on-chip peripheral control registers to the Register File, or from the Register File to the on-chip peripheral control registers. The sequence of operations in a DMAx data transfer is:

- 1. DMAx trigger source requests a DMA data transfer.
- 2. DMAx requests control of the system bus (address and data) from the eZ8 CPU.
- 3. After the eZ8 CPU acknowledges the bus request, DMAx transfers either a single byte or a two-byte word (depending upon configuration) and then returns system bus control back to the eZ8 CPU.
- 4. If Current Address equals End Address:
  - DMAx reloads the original Start Address
  - If configured to generate an interrupt, DMAx sends an interrupt request to the Interrupt Controller
  - If configured for single-pass operation, DMAx resets the DEN bit in the DMAx
     Control register to 0 and the DMA is disabled.

If Current Address does not equal End Address, the Current Address increments by 1 (single-byte transfer) or 2 (two-byte word transfer).

# **Configuring DMA0 and DMA1 for Data Transfer**

Follow the steps below to configure and enable DMA0 or DMA1:

- 1. Write to the DMAx I/O Address register to set the Register File address identifying the on-chip peripheral control register. The upper nibble of the 12-bit address for on-chip peripheral control registers is always FH. The full address is {FH, DMAx IO[7:0]}.
- 2. Determine the 12-bit Start and End Register File addresses. The 12-bit Start Address is given by {DMAx\_H[3:0], DMA\_START[7:0]}. The 12-bit End Address is given by {DMAx\_H[7:4], DMA\_END[7:0]}.
- 3. Write the Start and End Register File address high nibbles to the DMA*x* End/Start Address High Nibble register.
- 4. Write the lower byte of the Start Address to the DMAx Start/Current Address register.
- 5. Write the lower byte of the End Address to the DMAx End Address register.
- 6. Write to the DMAx Control register to complete the following:
  - Select loop or single-pass mode operation
  - Select the data transfer direction (either from the Register File RAM to the onchip peripheral control register; or from the on-chip peripheral control register to the Register File RAM)
  - Enable the DMAx interrupt request, if desired
  - Select Word or Byte mode
  - Select the DMAx request trigger
  - Enable the DMAx channel

# **DMA ADC Operation**

DMA\_ADC transfers data from the ADC to the Register File. The sequence of operations in a DMA\_ADC data transfer is:

- 1. ADC completes conversion on the current ADC input channel and signals the DMA controller that two-bytes of ADC data are ready for transfer.
- 2. DMA\_ADC requests control of the system bus (address and data) from the eZ8 CPU.
- 3. After the eZ8 CPU acknowledges the bus request, DMA\_ADC transfers the two-byte ADC output value to the Register File and then returns system bus control back to the eZ8 CPU.
- 4. If the current ADC Analog Input is the highest numbered input to be converted:
  - DMA\_ADC resets the ADC Analog Input number to 0 and initiates data conversion on ADC Analog Input 0.
  - If configured to generate an interrupt, DMA\_ADC sends an interrupt request to the Interrupt Controller

If the current ADC Analog Input is not the highest numbered input to be converted, DMA ADC initiates data conversion in the next higher numbered ADC Analog Input.

# Configuring DMA ADC for Data Transfer

Follow the steps below to configure and enable DMA ADC:

- 1. Write the DMA ADC Address register with the 7 most-significant bits of the Register File address for data transfers.
- 2. Write to the DMA ADC Control register to complete the following:
  - Enable the DMA ADC interrupt request, if desired
  - Select the number of ADC Analog Inputs to convert
  - Enable the DMA ADC channel



**Caution:** When using the DMA ADC to perform conversions on multiple ADC inputs, the Analog-to-Digital Converter must be configured for SINGLE-SHOT mode. If the ADC IN field in the DMA ADC Control Register is greater than 000b, the ADC must be in SINGLE-SHOT mode.

> CONTINUOUS mode operation of the ADC can only be used in conjunction with DMA ADC if the ADC IN field in the DMA ADC Control Register is reset to 000b to enable conversion on ADC Analog Input 0 only.

# **DMA Control Register Definitions**

#### DMAx Control Register

The DMAx Control register (see Table 77 on page 167) enables and selects the mode of operation for DMAx.

Table 77. DMAx Control Register (DMAxCTL)

BITS	7	6	5	4	3	2	1	0		
FIELD	DEN	DLE	DDIR	IRQEN	WSEL	RSS				
RESET		0								
R/W				R/	W					
ADDR				FB0H,	FB8H					

DEN—DMA*x* Enable

0 = DMAx is disabled and data transfer requests are disregarded.

1 = DMAx is enabled and initiates a data transfer upon receipt of a request from the trigger source.

#### DLE—DMA*x* Loop Enable

- 0 = DMAx reloads the original Start Address and is then disabled after the End Address data is transferred.
- 1 = DMAx, after the End Address data is transferred, reloads the original Start Address and continues operating.

#### DDIR—DMAx Data Transfer Direction

- $0 = \text{Register File} \rightarrow \text{on-chip peripheral control register}$ .
- $1 = \text{on-chip peripheral control register} \rightarrow \text{Register File.}$

#### IRQEN—DMAx Interrupt Enable

- 0 = DMAx does not generate any interrupts.
- 1 = DMAx generates an interrupt when the End Address data is transferred.

#### WSEL—Word Select

- 0 = DMAx transfers a single byte per request.
- 1 = DMAx transfers a two-byte word per request. The address for the on-chip peripheral control register must be an even address.

#### RSS—Request Trigger Source Select

The Request Trigger Source Select field determines the peripheral that can initiate a DMA transfer. The corresponding interrupts do not need to be enabled within the Interrupt Controller to initiate a DMA transfer. However, if the Request Trigger Source can enable or disable the interrupt request sent to the Interrupt Controller, the interrupt request must be enabled within the Request Trigger Source block.

- 000 = Timer 0.
- 001 = Timer 1.
- 010 = Timer 2.
- 011 = Timer 3.
- 100 = DMA0 Control register: UARTO Received Data register contains valid data.

  DMA1 Control register: UARTO Transmit Data register empty.
- 101 = DMA0 Control register: UART1 Received Data register contains valid data. DMA1 Control register: UART1 Transmit Data register empty.
- 110 = DMA0 Control register: I<sup>2</sup>C Receiver Interrupt. DMA1 Control register: I<sup>2</sup>C Transmitter Interrupt register empty.
- 111 = Reserved.

# DMAx I/O Address Register

The DMAx I/O Address register (Table 78) contains the low byte of the on-chip peripheral address for data transfer. The full 12-bit Register File address is given by {FH,

DMA*x*\_IO[7:0]}. When the DMA is configured for two-byte word transfers, the DMA*x* I/O Address register must contain an even numbered address.

Table 78. DMAx I/O Address Register (DMAxIO)

BITS	7	6	5	4	3	2	1	0		
FIELD		DMA_IO								
RESET		X								
R/W				R/	W					
ADDR				FB1H,	FB9H					

DMA\_IO—DMA on-chip peripheral control register address

This byte sets the low byte of the on-chip peripheral control register address on Register File Page FH (addresses FOOH to FFFH).

# DMAx Address High Nibble Register

The DMAx Address High register (Table 79) specifies the upper four bits of address for the Start/Current and End Addresses of DMAx.

Table 79. DMAx Address High Nibble Register (DMAxH)

BITS	7	6	5	4	3	2	1	0		
FIELD		DMA_END_H DMA_START_H								
RESET		X								
R/W				R/	W					
ADDR				FB2H,	FBAH					

DMA END H—DMAx End Address High Nibble

These bits, used with the DMAx End Address Low register, form a 12-bit End Address. The full 12-bit address is given by {DMA\_END\_H[3:0], DMA\_END[7:0]}.

DMA\_START\_H—DMAx Start/Current Address High Nibble These bits, used with the DMAx Start/Current Address Low register, form a 12-bit Start/Current Address. The full 12-bit address is given by {DMA\_START\_H[3:0], DMA\_START[7:0]}.

# DMAx Start/Current Address Low Byte Register

The DMAx Start/Current Address Low register, in conjunction with the DMAx Address High Nibble register, forms a 12-bit Start/Current Address. Writes to this register set the Start Address for DMA operations. Each time the DMA completes a data transfer, the 12-bit Start/Current Address increments by either 1 (single-byte transfer) or 2 (two-byte word transfer). Reads from this register return the low byte of the Current Address to be used for the next DMA data transfer.

Table 80. DMAx Start/Current Address Low Byte Register (DMAxSTART)

BITS	7	6	5	4	3	2	1	0		
FIELD		DMA_START								
RESET		X								
R/W				R/	W					
ADDR				FB3H,	FBBH					

DMA START—DMAx Start/Current Address Low

These bits, with the four lower bits of the DMAx\_H register, form the 12-bit Start/Current address. The full 12-bit address is given by {DMA START H[3:0], DMA START[7:0]}.

# DMAx End Address Low Byte Register

The DMAx End Address Low Byte register (Table 80), in conjunction with the DMAx\_H register (Table 81), forms a 12-bit End Address.

Table 81. DMAx End Address Low Byte Register (DMAxEND)

BITS	7	6	5	4	3	2	1	0		
FIELD		DMA_END								
RESET		Х								
R/W				R/	W					
ADDR				FB4H,	FBCH					

DMA END—DMAx End Address Low

These bits, with the four upper bits of the DMAx\_H register, form a 12-bit address. This address is the ending location of the DMAx transfer. The full 12-bit address is given by {DMA\_END\_H[3:0], DMA\_END[7:0]}.

# **DMA\_ADC Address Register**

The DMA\_ADC Address register (Table 83) points to a block of the Register File to store ADC conversion values as displayed in Table 82. This register contains the seven most-significant bits of the 12-bit Register File addresses. The five least-significant bits are calculated from the ADC Analog Input number (5-bit base address is equal to twice the ADC Analog Input number). The 10-bit ADC conversion data is stored as two bytes with the most significant byte of the ADC data stored at the even numbered Register File address.

Table 82 provides an example of the Register File addresses if the DMA\_ADC Address register contains the value 72H.

Table 82. DMA\_ADC Register File Address Example

ADC Analog Input	Register File Address (Hex) <sup>1</sup>
0	720H-721H
1	722H-723H
2	724H-725H
3	726H-727H
4	728H-729H
5	72AH-72BH
6	72CH-72DH
7	72EH-72FH
8	730H-731H
9	732H-733H
10	734H-735H
11	736H-737H

<sup>&</sup>lt;sup>1</sup>DMAA\_ADDR set to 72H.

Table 83. DMA\_ADC Address Register (DMAA\_ADDR)

BITS	7	6	5	4	3	2	1	0			
FIELD		DMAA_ADDR F									
RESET		×									
R/W				R/	W						
ADDR				FB	DH						

#### DMAA ADDR—DMA ADC Address

These bits specify the seven most-significant bits of the 12-bit Register File addresses used for storing the ADC output data. The ADC Analog Input Number defines the five least-significant bits of the Register File address. Full 12-bit address is {DMAA\_ADDR[7:1], 4-bit ADC Analog Input Number, 0}.

#### Reserved

This bit is reserved and must be 0.

# DMA\_ADC Control Register

The DMA\_ADC Control register (Table 84 on page 172) enables and sets options (DMA enable and interrupt enable) for ADC operation.

Table 84. DMA\_ADC Control Register (DMAACTL)

BITS	7	6	5	4	3 2 1					
FIELD	DAEN	IRQEN	Reserved ADC_IN							
RESET		0								
R/W		R/W								
ADDR				FB	EH					

## DAEN—DMA ADC Enable

0 = DMA ADC is disabled and the ADC Analog Input Number (ADC IN) is reset to 0.

1 = DMA ADC is enabled.

#### IRQEN—Interrupt Enable

0 = DMA ADC does not generate any interrupts.

1 = DMA\_ADC generates an interrupt after transferring data from the last ADC Analog Input specified by the ADC IN field.

#### Reserved

These bits are reserved and must be 0.

## ADC IN—ADC Analog Input Number

These bits set the number of ADC Analog Inputs to be used in the continuous update (data conversion followed by DMA data transfer). The conversion always begins with ADC Analog Input 0 and then progresses sequentially through the other selected ADC Analog Inputs.

0000 = ADC Analog Input 0 updated.

0001 = ADC Analog Inputs 0-1 updated.

0010 = ADC Analog Inputs 0-2 updated.

0011 = ADC Analog Inputs 0-3 updated.

0100 = ADC Analog Inputs 0-4 updated.

0101 = ADC Analog Inputs 0-5 updated. 0110 = ADC Analog Inputs 0-6 updated. 0111 = ADC Analog Inputs 0-7 updated. 1000 = ADC Analog Inputs 0-8 updated. 1001 = ADC Analog Inputs 0-9 updated. 1010 = ADC Analog Inputs 0-10 updated. 1011 = ADC Analog Inputs 0-11 updated. 1100-1111 = Reserved.

# **DMA Status Register**

The DMA Status register (Table 85 on page 173) indicates the DMA channel that generated the interrupt and the ADC Analog Input that is currently undergoing conversion. Reads from this register reset the Interrupt Request Indicator bits (IRQA, IRQ1, and IRQ0) to 0. Therefore, software interrupt service routines that read this register must process all three interrupt sources from the DMA.

Table 85. DMA\_ADC Status Register (DMAA\_STAT)

BITS	7	6	5	4	3	2	1	0	
FIELD		CADO	C[3:0]		Reserved	IRQA	IRQ1	IRQ0	
RESET		0							
R/W		R							
ADDR				FB	FH				

CADC[3:0]—Current ADC Analog Input

This field identifies the Analog Input that the ADC is currently converting.

#### Reserved

This bit is reserved and must be 0.

IRQA—DMA ADC Interrupt Request Indicator

This bit is automatically reset to 0 each time a read from this register occurs.

0 = DMA ADC is not the source of the interrupt from the DMA Controller.

1 = DMA\_ADC completed transfer of data from the last ADC Analog Input and generated an interrupt.

IRQ1—DMA1 Interrupt Request Indicator

This bit is automatically reset to 0 each time a read from this register occurs.

0 = DMA1 is not the source of the interrupt from the DMA Controller.

1 = DMA1 completed transfer of data to/from the End Address and generated an interrupt.

#### IRQ0—DMA0 Interrupt Request Indicator

This bit is automatically reset to 0 each time a read from this register occurs.

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0 = DMA0 is not the source of the interrupt from the DMA Controller.

1 = DMA0 completed transfer of data to/from the End Address and generated an interrupt.

# **Analog-to-Digital Converter**

#### **Overview**

The Analog-to-Digital Converter (ADC) converts an analog input signal to a 10-bit binary number. The features of the sigma-delta ADC include:

- 12 analog input sources are multiplexed with general-purpose I/O ports
- Interrupt upon conversion complete
- Internal voltage reference generator
- Direct Memory Access (DMA) controller can automatically initiate data conversion and transfer of the data from 1 to 12 of the analog inputs

## **Architecture**

Figure 34 displays the three major functional blocks (converter, analog multiplexer, and voltage reference generator) of the ADC. The ADC converts an analog input signal to its digital representation. The 12-input analog multiplexer selects one of the 12 analog input sources. The ADC requires an input reference voltage for the conversion. The voltage reference for the conversion may be input through the external VREF pin or generated internally by the voltage reference generator.

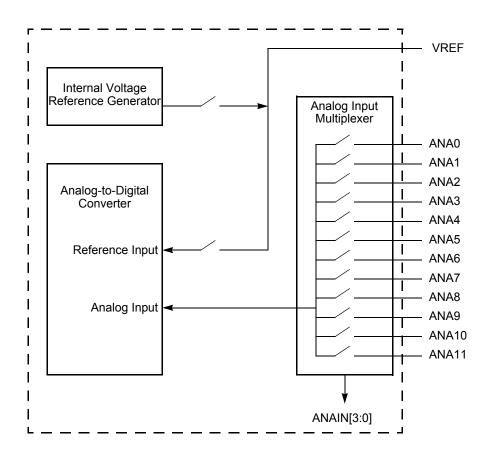


Figure 34. Analog-to-Digital Converter Block Diagram

The sigma-delta ADC architecture provides alias and image attenuation below the amplitude resolution of the ADC in the frequency range of DC to one-half the ADC clock rate (one-fourth the system clock rate). The ADC provides alias free conversion for frequencies up to one-half the ADC clock rate. Thus the sigma-delta ADC exhibits high noise immunity making it ideal for embedded applications. In addition, monotonicity (no missing codes) is guaranteed by design.

# **Operation**

#### **Automatic Power-Down**

If the ADC is idle (no conversions in progress) for 160 consecutive system clock cycles, portions of the ADC are automatically powered-down. From this power-down state, the ADC requires 40 system clock cycles to power-up. The ADC powers up when a conversion is requested using the ADC Control register.

# Single-Shot Conversion

When configured for single-shot conversion, the ADC performs a single analog-to-digital conversion on the selected analog input channel. After completion of the conversion, the ADC shuts down. Follow the steps below for setting up the ADC and initiating a singleshot conversion:

- 1. Enable the desired analog inputs by configuring the general-purpose I/O pins for alternate function. This configuration disables the digital input and output drivers.
- 2. Write to the ADC Control register to configure the ADC and begin the conversion. The bit fields in the ADC Control register can be written simultaneously:
  - Write to the ANAIN[3:0] field to select one of the 12 analog input sources.
  - Clear CONT to 0 to select a single-shot conversion.
  - Write to the VREF bit to enable or disable the internal voltage reference generator.
  - Set CEN to 1 to start the conversion.
- 3. CEN remains 1 while the conversion is in progress. A single-shot conversion requires 5129 system clock cycles to complete. If a single-shot conversion is requested from an ADC powered-down state, the ADC uses 40 additional clock cycles to power-up before beginning the 5129 cycle conversion.
- 4. When the conversion is complete, the ADC control logic performs the following operations:
  - 10-bit data result written to {ADCD H[7:0], ADCD L[7:6]}.
  - CEN resets to 0 to indicate the conversion is complete.
  - An interrupt request is sent to the Interrupt Controller.
- 5. If the ADC remains idle for 160 consecutive system clock cycles, it is automatically powered-down.

## **Continuous Conversion**

When configured for continuous conversion, the ADC continuously performs an analogto-digital conversion on the selected analog input. Each new data value over-writes the previous value stored in the ADC Data registers. An interrupt is generated after each conversion.



**Caution:** *In CONTINUOUS mode, you must be aware that ADC updates are limited by* the input signal bandwidth of the ADC and the latency of the ADC and its digital filter. Step changes at the input are not seen at the next output from the ADC. The response of the ADC (in all modes) is limited by the input signal bandwidth and the latency.

Follow the steps below for setting up the ADC and initiating continuous conversion:

- 1. Enable the desired analog input by configuring the general-purpose I/O pins for alternate function. This disables the digital input and output driver.
- 2. Write to the ADC Control register to configure the ADC for continuous conversion. The bit fields in the ADC Control register may be written simultaneously:
  - Write to the ANAIN[3:0] field to select one of the 12 analog input sources.
  - Set CONT to 1 to select continuous conversion.
  - Write to the VREF bit to enable or disable the internal voltage reference generator.
  - Set CEN to 1 to start the conversions.
- 3. When the first conversion in continuous operation is complete (after 5129 system clock cycles, plus the 40 cycles for power-up, if necessary), the ADC control logic performs the following operations:
  - CEN resets to 0 to indicate the first conversion is complete. CEN remains 0 for all subsequent conversions in continuous operation.
  - An interrupt request is sent to the Interrupt Controller to indicate the conversion is complete.
- 4. Thereafter, the ADC writes a new 10-bit data result to {ADCD\_H[7:0], ADCD\_L[7:6]} every 256 system clock cycles. An interrupt request is sent to the Interrupt Controller when each conversion is complete.
- 5. To disable continuous conversion, clear the CONT bit in the ADC Control register to 0.

#### DMA Control of the ADC

The Direct Memory Access (DMA) Controller can control operation of the ADC including analog input selection and conversion enable. For more information on the DMA and configuring for ADC operations, see Direct Memory Access Controller on page 165.

# **ADC Control Register Definitions**

# **ADC Control Register**

The ADC Control register selects the analog input channel and initiates the analog-to-digital conversion.

#### Table 86. ADC Control Register (ADCCTL)

BITS	7	6	5	4	3 2 1 0				
FIELD	CEN	Reserved	VREF	CONT	ANAIN[3:0]				
RESET	(	)	1		0				
R/W		R/W							
ADDR				F7	0H				

#### CEN—Conversion Enable

0 = Conversion is complete. Writing a 0 produces no effect. The ADC automatically clears this bit to 0 when a conversion has been completed.

1 = Begin conversion. Writing a 1 to this bit starts a conversion. If a conversion is already in progress, the conversion restarts. This bit remains 1 until the conversion is complete.

Reserved—Must be 0.

#### VREF

0 = Internal voltage reference generator enabled. The VREF pin should be left unconnected (or capacitively coupled to analog ground) if the internal voltage reference is selected as the ADC reference voltage.

1 = Internal voltage reference generator disabled. An external voltage reference must be provided through the VREF pin.

#### CONT

0 = Single-shot conversion. ADC data is output once at completion of the 5129 system clock cycles.

1 = Continuous conversion. ADC data updated every 256 system clock cycles.

#### ANAIN—Analog Input Select

These bits select the analog input for conversion. Not all Port pins in this list are available in all packages for the Z8F642x family Z8R642x family of products. For information on the Port pins available with each package style, see Signal and Pin Descriptions on page 7. Do not enable unavailable analog inputs.

0000 = ANA0

0001 = ANA1

0010 = ANA2

0011 = ANA3

0100 = ANA4 0101 = ANA5 0110 = ANA6 0111 = ANA7 1000 = ANA8 1001 = ANA9 1010 = ANA10 1011 = ANA11 11XX = Reserved.

# **ADC Data High Byte Register**

The ADC Data High Byte register (Table 87) contains the upper eight bits of the 10-bit ADC output. During a single-shot conversion, this value is invalid. Access to the ADC Data High Byte register is read-only. The full 10-bit ADC result is given by {ADCD\_H[7:0], ADCD\_L[7:6]}. Reading the ADC Data High Byte register latches data in the ADC Low Bits register.

Table 87. ADC Data High Byte Register (ADCD\_H)

BITS	7	6	5	4	3	2	1	0		
FIELD		ADCD_H								
RESET		X								
R/W				F	₹					
ADDR				F7	2H					

ADCD H—ADC Data High Byte

This byte contains the upper eight bits of the 10-bit ADC output. These bits are not valid during a single-shot conversion. During a continuous conversion, the last conversion output is held in this register. These bits are undefined after a Reset.

# ADC Data Low Bits Register

The ADC Data Low Bits register (Table 88) contains the lower two bits of the conversion value. The data in the ADC Data Low Bits register is latched each time the ADC Data High Byte register is read. Reading this register always returns the lower two bits of the conversion last read into the ADC High Byte register. Access to the ADC Data Low Bits register is read-only. The full 10-bit ADC result is given by {ADCD\_H[7:0], ADCD\_L[7:6]}.

# Table 88. ADC Data Low Bits Register (ADCD\_L)

BITS	7	6	5 4 3 2 1 0							
FIELD	ADC	ADCD_L Reserved								
RESET		Х								
R/W				F	₹					
ADDR				F7	3H					

ADCD\_L—ADC Data Low Bits

These are the least significant two bits of the 10-bit ADC output. These bits are undefined after a Reset.

#### Reserved

These bits are reserved and are always undefined.

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# Flash Memory

#### **Overview**

The products in the Z8 Encore! XP 64K Series Flash Microcontrollers feature up to 64 KB (65,536 bytes) of non-volatile Flash memory with read/write/erase capability. The Flash memory can be programmed and erased in-circuit by either user code or through the On-Chip Debugger.

The Flash memory array is arranged in 512-byte per page. The 512-byte page is the minimum Flash block size that can be erased. The Flash memory is also divided into 8 sectors which can be protected from programming and erase operations on a per sector basis.

Table 89 describes the Flash memory configuration for each device in the 64K Series. Table 90 on page 184 lists the sector address ranges. Figure 35 on page 184 displays the Flash memory arrangement.

**Table 89. Flash Memory Configurations** 

Part Number	Flash Size	Number of Pages	Flash Memory Addresses	Sector Size	Number of Sectors	Pages per Sector
Z8F162x	16K (16,384)	32	0000H - 3FFFH	2K (2048)	8	4
Z8F242x	24K (24,576)	48	0000H - 5FFFH	4K (4096)	6	8
Z8F322x	32K (32,768)	64	0000H - 7FFFH	4K (4096)	8	8
Z8F482x	48K (49,152)	96	0000H - BFFFH	8K (8192)	6	16
Z8F642x	64K (65,536)	128	0000H - FFFFH	8K (8192)	8	16

**Table 90. Flash Memory Sector Addresses** 

	Flash Sector Address Ranges								
Sector Number	Z8F162x	Z8F242x	Z8F322x	Z8F482x	Z8F642x				
0	0000H-07FFH	0000H-0FFFH	0000H-0FFFH	0000H-1FFFH	0000H-1FFFH				
1	0800H-0FFFH	1000H-1FFFH	1000H-1FFFH	2000H-3FFFH	2000H-3FFFH				
2	1000H-17FFH	2000H-2FFFH	2000H-2FFFH	4000H-5FFFH	4000H-5FFFH				
3	1800H-1FFFH	3000H-3FFFH	3000H-3FFFH	6000H-7FFFH	6000H-7FFFH				
4	2000H-27FFH	4000H-4FFFH	4000H-4FFFH	8000H-9FFFH	8000H-9FFFH				
5	2800H-2FFFH	5000H-5FFFH	5000H-5FFFH	A000H-BFFFH	A000H-BFFFH				
6	3000H-37FFH	N/A	6000H-6FFFH	N/A	C000H-DFFFH				
7	3800H-3FFFH	N/A	7000H-7FFFH	N/A	E000H-FFFFH				

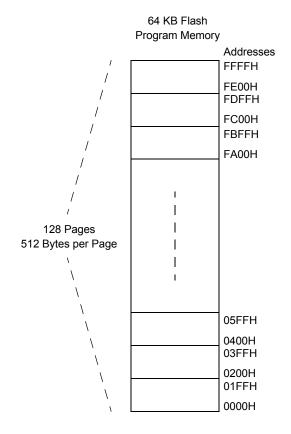


Figure 35. Flash Memory Arrangement

#### **Information Area**

Table 91 describes the 64K Series Information Area. This 512-byte Information Area is accessed by setting bit 7 of the Page Select Register to 1. When access is enabled, the Information Area is mapped into Flash Memory and overlays the 512 bytes at addresses FE00H to FFFFH. When the Information Area access is enabled, LDC instructions return data from the Information Area. CPU instruction fetches always comes from Flash Memory regardless of the Information Area access bit. Access to the Information Area is readonly.

Table 91. Z8 Encore! XP 64K Series Flash Microcontrollers Information Area Map

Flash Memory Address (Hex)	Function
FE00H-FE3FH	Reserved
FE40H-FE53H	Part Number 20-character ASCII alphanumeric code Left justified and filled with zeros
FE54H-FFFFH	Reserved

# **Operation**

The Flash Controller provides the proper signals and timing for Byte Programming, Page Erase, and Mass Erase of the Flash memory. The Flash Controller contains a protection mechanism, via the Flash Control register (FCTL), to prevent accidental programming or erasure. The following subsections provide details on the various operations (Lock, Unlock, Sector Protect, Byte Programming, Page Erase, and Mass Erase).

# **Timing Using the Flash Frequency Registers**

Before performing a program or erase operation on the Flash memory, you must first configure the Flash Frequency High and Low Byte registers. The Flash Frequency registers allow programming and erasure of the Flash with system clock frequencies ranging from 20 kHz through 20 MHz (the valid range is limited to the device operating frequencies).

The Flash Frequency High and Low Byte registers combine to form a 16-bit value, FFREQ, to control timing for Flash program and erase operations. The 16-bit Flash Frequency value must contain the system clock frequency in kHz. This value is calculated using the following equation:.

$$FFREQ[15:0] = \frac{System Clock Frequency (Hz)}{1000}$$



**Caution:** Flash programming and erasure are not supported for system clock frequencies below 20 kHz, above 20 MHz, or outside of the device operating frequency range. The Flash Frequency High and Low Byte registers must be loaded with the correct value to insure proper Flash programming and erase operations.

#### Flash Read Protection

The user code contained within the Flash memory can be protected from external access. Programming the Flash Read Protect Option Bit prevents reading of user code by the On-Chip Debugger or by using the Flash Controller Bypass mode. For more information, see Option Bits on page 195 and On-Chip Debugger on page 199.

#### Flash Write/Erase Protection

The 64K Series provides several levels of protection against accidental program and erasure of the Flash memory contents. This protection is provided by the Flash Controller unlock mechanism, the Flash Sector Protect register, and the Flash Write Protect option bit.

#### Flash Controller Unlock Mechanism

At Reset, the Flash Controller locks to prevent accidental program or erasure of the Flash memory. To program or erase the Flash memory, the Flash controller must be unlocked. After unlocking the Flash Controller, the Flash can be programmed or erased. Any value written by user code to the Flash Control register or Page Select Register out of sequence will lock the Flash Controller.

Follow the steps below to unlock the Flash Controller from user code:

- 1. Write 00H to the Flash Control register to reset the Flash Controller.
- 2. Write the page to be programmed or erased to the Page Select register.



- 3. Write the first unlock command 73H to the Flash Control register.
- 4. Write the second unlock command 8CH to the Flash Control register.
- 5. Re-write the page written in step 2 to the Page Select register.

#### **Flash Sector Protection**

The Flash Sector Protect register can be configured to prevent sectors from being programmed or erased. Once a sector is protected, it cannot be unprotected by user code. The Flash Sector Protect register is cleared after reset and any previously written protection values is lost. User code must write this register in their initialization routine if they want to enable sector protection.

The Flash Sector Protect register shares its Register File address with the Page Select register. The Flash Sector Protect register is accessed by writing the Flash Control register with 5EH. Once the Flash Sector Protect register is selected, it can be accessed at the Page Select Register address. When user code writes the Flash Sector Protect register, bits can only be set to 1. Thus, sectors can be protected, but not unprotected, via register write operations. Writing a value other than 5EH to the Flash Control register de-selects the Flash Sector Protect register and re-enables access to the Page Select register.

Follow the steps below to setup the Flash Sector Protect register from user code:

- 1. Write 00H to the Flash Control register to reset the Flash Controller.
- 2. Write 5EH to the Flash Control register to select the Flash Sector Protect register.
- 3. Read and/or write the Flash Sector Protect register which is now at Register File address FF9H.
- 4. Write 00H to the Flash Control register to return the Flash Controller to its reset state.

#### **Flash Write Protection Option Bit**

The Flash Write Protect option bit can be enabled to block all program and erase operations from user code. For more information, see Option Bits on page 195.

#### Byte Programming

When the Flash Controller is unlocked, writes to Flash Memory from user code will program a byte into the Flash if the address is located in the unlocked page. An erased Flash byte contains all ones (FFH). The programming operation can only be used to change bits from one to zero. To change a Flash bit (or multiple bits) from zero to one requires a Page Erase or Mass Erase operation.

Byte Programming can be accomplished using the eZ8 CPU's LDC or LDCI instructions. For a description of the LDC and LDCI instructions, refer to  $eZ8^{TM}$  CPU Core User Manual (UM0128).



While the Flash Controller programs the Flash memory, the eZ8 CPU idles but the system clock and on-chip peripherals continue to operate. Interrupts that occur when a Programming operation is in progress are serviced once the Programming operation is complete. To exit Programming mode and lock the Flash Controller, write 00H to the Flash Control register.

User code cannot program Flash Memory on a page that lies in a protected sector. When user code writes memory locations, only addresses located in the unlocked page are programmed. Memory writes outside of the unlocked page are ignored.



**Caution:** Each memory location must not be programmed more than twice before an erase occurs.

Follow the steps below to program the Flash from user code:

- 1. Write 00H to the Flash Control register to reset the Flash Controller.
- 2. Write the page of memory to be programmed to the Page Select register.
- 3. Write the first unlock command 73H to the Flash Control register.
- 4. Write the second unlock command 8CH to the Flash Control register.
- 5. Re-write the page written in step 2 to the Page Select register.
- 6. Write Flash Memory using LDC or LDCI instructions to program the Flash.
- 7. Repeat step 6 to program additional memory locations on the same page.
- 8. Write 00H to the Flash Control register to lock the Flash Controller.

#### Page Erase

The Flash memory can be erased one page (512 bytes) at a time. Page Erasing the Flash memory sets all bytes in that page to the value FFH. The Page Select register identifies the page to be erased. While the Flash Controller executes the Page Erase operation, the eZ8 CPU idles but the system clock and on-chip peripherals continue to operate. The eZ8 CPU resumes operation after the Page Erase operation completes. Interrupts that occur when the Page Erase operation is in progress are serviced once the Page Erase operation is complete. When the Page Erase operation is complete, the Flash Controller returns to its locked state. Only pages located in unprotected sectors can be erased.

Follow the steps below to perform a Page Erase operation:

- 1. Write 00H to the Flash Control register to reset the Flash Controller.
- 2. Write the page to be erased to the Page Select register.
- 3. Write the first unlock command 73H to the Flash Control register.
- 4. Write the second unlock command 8CH to the Flash Control register.

- 5. Re-write the page written in step 2 to the Page Select register.
- 6. Write the Page Erase command 95H to the Flash Control register.

#### **Mass Erase**

The Flash memory cannot be Mass Erased by user code.

# Flash Controller Bypass

The Flash Controller can be bypassed and the control signals for the Flash memory brought out to the GPIO pins. Bypassing the Flash Controller allows faster Programming algorithms by controlling the Flash programming signals directly.

Flash Controller Bypass is recommended for gang programming applications and large volume customers who do not require in-circuit programming of the Flash memory.

For more information on bypassing the Flash Controller, refer to *Third-Party Flash Programming Support for Z8 Encore!* available for download at <a href="www.zilog.com">www.zilog.com</a>.

# Flash Controller Behavior in Debug Mode

The following changes in behavior of the Flash Controller occur when the Flash Controller is accessed using the On-Chip Debugger:

- The Flash Write Protect option bit is ignored.
- The Flash Sector Protect register is ignored for programming and erase operations.
- Programming operations are not limited to the page selected in the Page Select register.
- Bits in the Flash Sector Protect register can be written to one or zero.
- The second write of the Page Select register to unlock the Flash Controller is not necessary.
- The Page Select register can be written when the Flash Controller is unlocked.
- The Mass Erase command is enabled through the Flash Control register.



**Caution:** For security reasons, Flash controller allows only a single page to be opened for write/erase. When writing multiple Flash pages, the Flash controller must go through the unlock sequence again to select another page.

# Flash Control Register Definitions

## Flash Control Register

The Flash Control register (Table 92) unlocks the Flash Controller for programming and erase operations, or to select the Flash Sector Protect register.

The Write-only Flash Control Register shares its Register File address with the Read-only Flash Status Register.

**Table 92. Flash Control Register (FCTL)** 

BITS	7	6	5	4	3	2	1	0			
FIELD		FCMD									
RESET		0									
R/W		W									
ADDR				FF	8H						

FCMD—Flash Command

73H = First unlock command.

8CH = Second unlock command.

95H = Page erase command.

63H = Mass erase command

5EH = Flash Sector Protect register select.

# Flash Status Register

The Flash Status register (Table 93) indicates the current state of the Flash Controller. This register can be read at any time. The Read-only Flash Status Register shares its Register File address with the Write-only Flash Control Register.

Table 93. Flash Status Register (FSTAT)

BITS	7	6	5	4	3	2	1	0	
FIELD	Rese	erved	FSTAT						
RESET		0							
R/W		R							
ADDR				FF	8H				

<sup>\*</sup> All other commands, or any command out of sequence, lock the Flash Controller.

#### Reserved

These bits are reserved and must be 0.

FSTAT—Flash Controller Status

00 0000 = Flash Controller locked

00 0001 = First unlock command received

00 0010 = Second unlock command received

00 0011 = Flash Controller unlocked

00 0100 = Flash Sector Protect register selected

00 1xxx = Program operation in progress

 $01 \ 0xxx = Page erase operation in progress$ 

10 0xxx = Mass erase operation in progress

# **Page Select Register**

The Page Select (FPS) register (Table 94) selects one of the 128 available Flash memory pages to be erased or programmed. Each Flash Page contains 512 bytes of Flash memory. During a Page Erase operation, all Flash memory locations with the 7 most significant bits of the address given by the PAGE field are erased to FFH.

The Page Select register shares its Register File address with the Flash Sector Protect Register. The Page Select register cannot be accessed when the Flash Sector Protect register is enabled.

#### Table 94. Page Select Register (FPS)

BITS	7	6	5	4	3	2	1	0		
FIELD	INFO_EN PAGE									
RESET	0									
R/W	R/W									
ADDR				FF	9H					

INFO EN—Information Area Enable

0 = Information Area is not selected.

1 = Information Area is selected. The Information area is mapped into the Flash Memory address space at addresses FE00H through FFFFH.

PAGE—Page Select

This 7-bit field selects the Flash memory page for Programming and Page Erase operations. Flash Memory Address[15:9] = PAGE[6:0].

# Flash Sector Protect Register

The Flash Sector Protect register (Table 95) protects Flash memory sectors from being programmed or erased from user code. The Flash Sector Protect register shares its Register File address with the Page Select register. The Flash Sector protect register can be accessed only after writing the Flash Control register with 5EH.

User code can only write bits in this register to 1 (bits cannot be cleared to 0 by user code).

#### Table 95. Flash Sector Protect Register (FPROT)

BITS	7	6	5	4	3	2	1	0			
FIELD	SECT7	SECT6	SECT5	SECT4	SECT3	SECT2	SECT1	SECT0			
RESET		0									
R/W				R/\	W1						
ADDR	FF9H										
Note: R/W	1 = Register i	s accessible f	or Read oper	ations. Regist	ter can be wri	tten to 1 only	(via user code	e).			

SECT*n*—Sector Protect

- 0 = Sector n can be programmed or erased from user code.
- 1 = Sector n is protected and cannot be programmed or erased from user code.

# Flash Frequency High and Low Byte Registers

The Flash Frequency High and Low Byte registers (Table 96 and Table 97) combine to form a 16-bit value, FFREQ, to control timing for Flash program and erase operations. The 16-bit Flash Frequency registers must be written with the system clock frequency in kHz for Program and Erase operations. Calculate the Flash Frequency value using the following equation:

$$FFREQ[15:0] = \{FFREQH[7:0], FFREQL[7:0]\} = \frac{System\ Clock\ Frequency}{1000}$$

 $\triangle$ 

Caution: Flash programming and erasure is not supported for system clock frequencies below 20 kHz, above 20 MHz, or outside of the valid operating frequency range for the device. The Flash Frequency High and Low Byte registers must be loaded with the correct value to insure proper program and erase times.

<sup>\*</sup> User code can only write bits from 0 to 1.

# Table 96. Flash Frequency High Byte Register (FFREQH)

BITS	7	6	5	4	3	2	1	0			
FIELD		FFREQH									
RESET		0									
R/W		R/W									
ADDR				FF.	AH						

# Table 97. Flash Frequency Low Byte Register (FFREQL)

BITS	7	6	5	4	3	2	1	0			
FIELD		FFREQL									
RESET		0									
R/W		R/W									
ADDR				FF	ВН						

FFREQH and FFREQL—Flash Frequency High and Low Bytes These 2 bytes, {FFREQH[7:0], FFREQL[7:0]}, contain the 16-bit Flash Frequency value.

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# **Option Bits**

#### Overview

Option Bits allow user configuration of certain aspects of the 64K Series operation. The feature configuration data is stored in the Flash Memory and read during Reset. The features available for control via the Option Bits are:

- Watchdog Timer time-out response selection—interrupt or Reset.
- Watchdog Timer enabled at Reset.
- The ability to prevent unwanted read access to user code in Flash Memory.
- The ability to prevent accidental programming and erasure of the user code in Flash Memory.
- Voltage Brownout configuration-always enabled or disabled during STOP mode to reduce STOP mode power consumption.
- Oscillator mode selection-for high, medium, and low power crystal oscillators, or external RC oscillator.

# Operation

# **Option Bit Configuration By Reset**

Each time the Option Bits are programmed or erased, the device must be Reset for the change to take place. During any reset operation (System Reset, Reset, or Stop Mode Recovery), the Option Bits are automatically read from the Flash Memory and written to Option Configuration registers. The Option Configuration registers control operation of the devices within the 64K Series. Option Bit control is established before the device exits Reset and the eZ8 CPU begins code execution. The Option Configuration registers are not part of the Register File and are not accessible for read or write access.

# **Option Bit Address Space**

The first two bytes of Flash Memory at addresses 0000H (see Table 98 on page 196) and 0001H (see Table 99 on page 197) are reserved for the user Option Bits. The byte at Flash Memory address 0000H configures user options. The byte at Flash Memory address 0001H is reserved for future use and must remain unprogrammed.

# Flash Memory Address 0000H

Table 98. Flash Option Bits At Flash Memory Address 0000H

BITS	7	6	5	4	3	2	1	0			
FIELD	WDT_RE S	WDT_AO	OSC_S	SEL[1:0]	VBO_AO	RP	Reserved	FWP			
RESET		U									
R/W				R	/W						
ADDR		Program Memory 0000H									
Note: U = I	Unchanged by	y Reset. R/W	= Read/Write	<del>)</del> .							

# WDT RES-Watchdog Timer Reset

- 0 = Watchdog Timer time-out generates an interrupt request. Interrupts must be globally enabled for the eZ8 CPU to acknowledge the interrupt request.
- 1 = Watchdog Timer time-out causes a Short Reset. This setting is the default for unprogrammed (erased) Flash.

#### WDT AO—Watchdog Timer Always On

- 0 = Watchdog Timer is automatically enabled upon application of system power. Watchdog Timer can not be disabled except during STOP Mode (if configured to power down during STOP Mode).
- 1 = Watchdog Timer is enabled upon execution of the WDT instruction. Once enabled, the Watchdog Timer can only be disabled by a Reset or Stop Mode Recovery. This setting is the default for unprogrammed (erased) Flash.

#### OSC SEL[1:0]—Oscillator Mode Selection

- 00 = On-chip oscillator configured for use with external RC networks (<4 MHz).
- 01 = Minimum power for use with very low frequency crystals (32 kHz to 1.0 MHz).
- 10 = Medium power for use with medium frequency crystals or ceramic resonators (0.5 MHz to 10.0 MHz).
- 11 = Maximum power for use with high frequency crystals (8.0 MHz to 20.0 MHz). This setting is the default for unprogrammed (erased) Flash.

# VBO\_AO—Voltage Brownout Protection Always On

- 0 = Voltage Brownout Protection is disabled in STOP mode to reduce total power consumption.
- 1 = Voltage Brownout Protection is always enabled including during STOP mode. This setting is the default for unprogrammed (erased) Flash.

#### RP—Read Protect

0 = User program code is inaccessible. Limited control features are available through

the On-Chip Debugger.

1 = User program code is accessible. All On-Chip Debugger commands are enabled. This setting is the default for unprogrammed (erased) Flash.

#### Reserved

These Option Bits are reserved for future use and must always be 1. This setting is the default for unprogrammed (erased) Flash.

FWP—Flash Write Protect (Flash version only)

FWP	Description
0	Programming, Page Erase, and Mass Erase through User Code is disabled. Mass Erase is available through the On-Chip Debugger.
1	Programming, and Page Erase are enabled for all of Flash Program Memory.

# Flash Memory Address 0001H

Table 99. Options Bits at Flash Memory Address 0001H

вітѕ	7	6	5	4	3	2	1	0			
FIELD	Reserved										
RESET		U									
R/W				R/	W						
ADDR	Program Memory 0001H										
Note: U = l	Unchanged by	y Reset. R = I	Read-Only. R	/W = Read/W	rite.						

Reserved

These Option Bits are reserved for future use and must always be 1. This setting is the default for unprogrammed (erased) Flash.

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# **On-Chip Debugger**

#### **Overview**

The 64K Series products contain an integrated On-Chip Debugger (OCD) that provides advanced debugging features including:

- Reading and writing of the Register File
- Reading and writing of Program and Data Memory
- Setting of Breakpoints
- Execution of eZ8 CPU instructions

#### **Architecture**

The On-Chip Debugger consists of four primary functional blocks: transmitter, receiver, auto-baud generator, and debug controller. Figure 36 displays the architecture of the On-Chip Debugger.

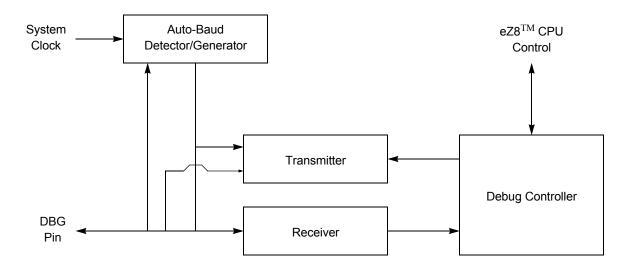


Figure 36. On-Chip Debugger Block Diagram

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# **Operation**

#### OCD Interface

The On-Chip Debugger uses the DBG pin for communication with an external host. This one-pin interface is a bi-directional open-drain interface that transmits and receives data. Data transmission is half-duplex, in that transmit and receive cannot occur simultaneously. The serial data on the DBG pin is sent using the standard asynchronous data format defined in RS-232. This pin can interface the 64K Series products to the serial port of a host PC using minimal external hardware. Two different methods for connecting the DBG pin to an RS-232 interface are depicted in Figure 37 and Figure 38 on page 201.



**Caution:** For operation of the On-Chip Debugger, all power pins ( $V_{DD}$  and  $AV_{DD}$ ) must be supplied with power, and **all** ground pins ( $V_{SS}$  and  $AV_{SS}$ ) must be properly grounded.

> The DBG pin is open-drain and must always be connected to  $V_{DD}$  through an external pull-up resistor to ensure proper operation.

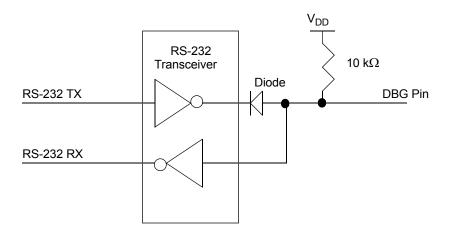


Figure 37. Interfacing the On-Chip Debugger's DBG Pin with an RS-232 Interface (1)

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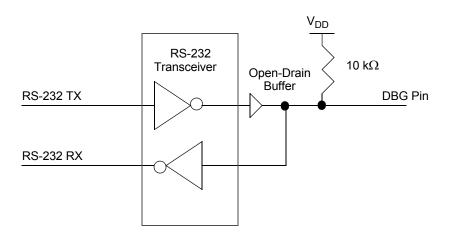


Figure 38. Interfacing the On-Chip Debugger's DBG Pin with an RS-232 Interface (2)

#### **DEBUG Mode**

The operating characteristics of the 64K Series devices in DEBUG mode are:

- The eZ8 CPU fetch unit stops, idling the eZ8 CPU, unless directed by the OCD to execute specific instructions.
- The system clock operates unless in STOP mode.
- All enabled on-chip peripherals operate unless in STOP mode.
- Automatically exits HALT mode.
- Constantly refreshes the Watchdog Timer, if enabled.

#### **Entering DEBUG Mode**

The device enters DEBUG mode following any of the following operations:

- Writing the DBGMODE bit in the OCD Control Register to 1 using the OCD interface.
- eZ8 CPU execution of a BRK (Breakpoint) instruction (when enabled).
- If the DBG pin is Low when the device exits Reset, the On-Chip Debugger automatically puts the device into DEBUG mode.

## **Exiting DEBUG Mode**

The device exits DEBUG mode following any of the following operations:

- Clearing the DBGMODE bit in the OCD Control Register to 0.
- Power-On Reset
- Voltage Brownout reset

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- Asserting the  $\overline{RESET}$  pin Low to initiate a Reset.
- Driving the DBG pin Low while the device is in STOP mode initiates a system reset.

#### **OCD Data Format**

The OCD interface uses the asynchronous data format defined for RS-232. Each character is transmitted as 1 Start bit, 8 data bits (least-significant bit first), and 1 Stop bit (see Figure 39).

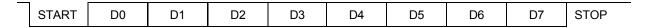


Figure 39. OCD Data Format

#### OCD Auto-Baud Detector/Generator

To run over a range of baud rates (bits per second) with various system clock frequencies, the On-Chip Debugger has an Auto-Baud Detector/Generator. After a reset, the OCD is idle until it receives data. The OCD requires that the first character sent from the host is the character 80H. The character 80H has eight continuous bits Low (one Start bit plus 7 data bits). The Auto-Baud Detector measures this period and sets the OCD Baud Rate Generator accordingly.

The Auto-Baud Detector/Generator is clocked by the system clock. The minimum baud rate is the system clock frequency divided by 512. For optimal operation, the maximum recommended baud rate is the system clock frequency divided by 8. The theoretical maximum baud rate is the system clock frequency divided by 4. This theoretical maximum is possible for low noise designs with clean signals. Table 100 lists minimum and recommended maximum baud rates for sample crystal frequencies.

Table 100. OCD Baud-Rate Limits

System Clock Frequency (MHz)	Recommended Maximum Baud Rate (kbits/s)	Minimum Baud Rate (kbits/s)
20.0	2500	39.1
1.0	125.0	1.96
0.032768 (32 kHz)	4.096	0.064

If the OCD receives a Serial Break (nine or more continuous bits Low) the Auto-Baud Detector/Generator resets. The Auto-Baud Detector/Generator can then be reconfigured by sending 80H.

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### **OCD Serial Errors**

The On-Chip Debugger can detect any of the following error conditions on the DBG pin:

- Serial Break (a minimum of nine continuous bits Low).
- Framing Error (received Stop bit is Low).
- Transmit Collision (OCD and host simultaneous transmission detected by the OCD).

When the OCD detects one of these errors, it aborts any command currently in progress, transmits a Serial Break 4096 system clock cycles long back to the host, and resets the Auto-Baud Detector/Generator. A Framing Error or Transmit Collision may be caused by the host sending a Serial Break to the OCD. Because of the open-drain nature of the interface, returning a Serial Break break back to the host only extends the length of the Serial Break if the host releases the Serial Break early.

The host transmits a Serial Break on the DBG pin when first connecting to the 64K Series devices or when recovering from an error. A Serial Break from the host resets the Auto-Baud Generator/Detector but does not reset the OCD Control register. A Serial Break leaves the device in DEBUG mode if that is the current mode. The OCD is held in Reset until the end of the Serial Break when the DBG pin returns High. Because of the opendrain nature of the DBG pin, the host can send a Serial Break to the OCD even if the OCD is transmitting a character.

### **Breakpoints**

Execution Breakpoints are generated using the BRK instruction (opcode 00H). When the eZ8 CPU decodes a BRK instruction, it signals the On-Chip Debugger. If Breakpoints are enabled, the OCD idles the eZ8 CPU and enters DEBUG mode. If Breakpoints are not enabled, the OCD ignores the BRK signal and the BRK instruction operates as an NOP.

If breakpoints are enabled, the OCD can be configured to automatically enter DEBUG mode, or to loop on the break instruction. If the OCD is configured to loop on the BRK instruction, then the CPU is still enabled to service DMA and interrupt requests.

The loop on BRK instruction can be used to service interrupts in the background. For interrupts to be serviced in the background, there cannot be any breakpoints in the interrupt service routine. Otherwise, the CPU stops on the breakpoint in the interrupt routine. For interrupts to be serviced in the background, interrupts must also be enabled. Debugging software should not automatically enable interrupts when using this feature, since interrupts are typically disabled during critical sections of code where interrupts should not occur (such as adjusting the stack pointer or modifying shared data).

Software can poll the IDLE bit of the OCDSTAT register to determine if the OCD is looping on a BRK instruction. When software wants to stop the CPU on the BRK instruction it is looping on, software should not set the DBGMODE bit of the OCDCTL register. The CPU may have vectored to and be in the middle of an interrupt service routine when this bit gets set. Instead, software must clear the BRKLP bit. This action allows the CPU to

finish the interrupt service routine it may be in and return the BRK instruction. When the CPU returns to the BRK instruction it was previously looping on, it automatically sets the DBGMODE bit and enter DEBUG mode.

Software detects that the majority of the OCD commands are still disabled when the eZ8<sup>TM</sup> CPU is looping on a BRK instruction. The eZ8 CPU must be stopped and the part must be in DEBUG mode before these commands can be issued.

### **Breakpoints in Flash Memory**

The BRK instruction is opcode 00H, which corresponds to the fully programmed state of a byte in Flash memory. To implement a Breakpoint, write 00H to the desired address, overwriting the current instruction. To remove a Breakpoint, the corresponding page of Flash memory must be erased and reprogrammed with the original data.

### **On-Chip Debugger Commands**

The host communicates to the On-Chip Debugger by sending OCD commands using the DBG interface. During normal operation, only a subset of the OCD commands are available. In DEBUG mode, all OCD commands become available unless the user code and control registers are protected by programming the Read Protect Option Bit (RP). The Read Protect Option Bit prevents the code in memory from being read out of the 64K Series products. When this option is enabled, several of the OCD commands are disabled. Table 101 contains a summary of the On-Chip Debugger commands. Each OCD command is described in detail in the bulleted list following Table 101.

Table 101 indicates those commands that operate when the device is not in DEBUG mode (normal operation) and those commands that are disabled by programming the Read Protect Option Bit.

Table 101. On-Chip Debugger Commands

Debug Command	Command Byte	Enabled when NOT in DEBUG mode?	Disabled by Read Protect Option Bit
Read OCD Revision	00H	Yes	-
Read OCD Status Register	02H	Yes	-
Read Runtime Counter	03H	-	-
Write OCD Control Register	04H	Yes	Cannot clear DBGMODE bit
Read OCD Control Register	05H	Yes	-

**Table 101. On-Chip Debugger Commands (Continued)** 

Debug Command	Command Byte	Enabled when NOT in DEBUG mode?	Disabled by Read Protect Option Bit
Write Program Counter	06H	-	Disabled
Read Program Counter	07H	-	Disabled
Write Register	08H	-	Only writes of the Flash Memory Control registers are allowed. Additionally, only the Mass Erase command is allowed to be written to the Flash Control register.
Read Register	09H	-	Disabled
Write Program Memory	0AH	-	Disabled
Read Program Memory	0BH	-	Disabled
Write Data Memory	0CH	-	Disabled
Read Data Memory	0DH	-	Disabled
Read Program Memory CRC	0EH	-	-
Reserved	0FH	-	-
Step Instruction	10H	-	Disabled
Stuff Instruction	11H	-	Disabled
Execute Instruction	12H	-	Disabled
Reserved	13H - FFH	-	-

In the following list of OCD Commands, data and commands sent from the host to the On-Chip Debugger are identified by 'DBG ← Command/Data'. Data sent from the On-Chip Debugger back to the host is identified by 'DBG → Data'

• **Read OCD Revision (00H)**—The Read OCD Revision command determines the version of the On-Chip Debugger. If OCD commands are added, removed, or changed, this revision number changes.

```
DBG \leftarrow 00H
DBG \rightarrow OCDREV[15:8] (Major revision number)
DBG \rightarrow OCDREV[7:0] (Minor revision number)
```

• **Read OCD Status Register (02H)**—The Read OCD Status Register command reads the OCDSTAT register.

```
DBG ← 02H
DBG → OCDSTAT[7:0]
```

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• Write OCD Control Register (04H)—The Write OCD Control Register command writes the data that follows to the OCDCTL register. When the Read Protect Option Bit is enabled, the DBGMODE bit (OCDCTL[7]) can only be set to 1, it cannot be cleared to 0 and the only method of putting the device back into normal operating mode is to reset the device.

```
DBG ← 04H
DBG ← OCDCTL[7:0]
```

• **Read OCD Control Register (05H)**—The Read OCD Control Register command reads the value of the OCDCTL register.

```
DBG \leftarrow 05H
DBG \rightarrow OCDCTL[7:0]
```

• Write Program Counter (06H)—The Write Program Counter command writes the data that follows to the eZ8 CPU's Program Counter (PC). If the device is not in DEBUG mode or if the Read Protect Option Bit is enabled, the Program Counter (PC) values are discarded.

```
DBG ← 06H
DBG ← ProgramCounter[15:8]
DBG ← ProgramCounter[7:0]
```

• **Read Program Counter (07H)**—The Read Program Counter command reads the value in the eZ8 CPU's Program Counter (PC). If the device is not in DEBUG mode or if the Read Protect Option Bit is enabled, this command returns FFFFH.

```
DBG ← 07H

DBG → ProgramCounter[15:8]

DBG → ProgramCounter[7:0]
```

Write Register (08H)—The Write Register command writes data to the Register File.
Data can be written 1-256 bytes at a time (256 bytes can be written by setting size to zero). If the device is not in DEBUG mode, the address and data values are discarded. If the Read Protect Option Bit is enabled, then only writes to the Flash Control Registers are allowed and all other register write data values are discarded.

```
DBG ← 08H

DBG ← {4'h0,Register Address[11:8]}

DBG ← Register Address[7:0]

DBG ← Size[7:0]

DBG ← 1-256 data bytes
```

• Read Register (09H)—The Read Register command reads data from the Register File. Data can be read 1-256 bytes at a time (256 bytes can be read by setting size to zero). If the device is not in DEBUG mode or if the Read Protect Option Bit is enabled, this command returns FFH for all the data values.

```
DBG ← 09H

DBG ← {4'h0,Register Address[11:8]

DBG ← Register Address[7:0]
```

```
DBG \leftarrow Size[7:0]
DBG \rightarrow 1-256 data bytes
```

• Write Program Memory (0AH)—The Write Program Memory command writes data to Program Memory. This command is equivalent to the LDC and LDCI instructions. Data can be written 1-65536 bytes at a time (65536 bytes can be written by setting size to zero). The on-chip Flash Controller must be written to and unlocked for the programming operation to occur. If the Flash Controller is not unlocked, the data is discarded. If the device is not in DEBUG mode or if the Read Protect Option Bit is enabled, the data is discarded.

```
DBG ← 0AH

DBG ← Program Memory Address[15:8]

DBG ← Program Memory Address[7:0]

DBG ← Size[15:8]

DBG ← Size[7:0]

DBG ← 1-65536 data bytes
```

• **Read Program Memory (0BH)**—The Read Program Memory command reads data from Program Memory. This command is equivalent to the LDC and LDCI instructions. Data can be read 1-65536 bytes at a time (65536 bytes can be read by setting size to zero). If the device is not in DEBUG mode or if the Read Protect Option Bit is enabled, this command returns FFH for the data.

```
DBG ← 0BH

DBG ← Program Memory Address[15:8]

DBG ← Program Memory Address[7:0]

DBG ← Size[15:8]

DBG ← Size[7:0]

DBG → 1-65536 data bytes
```

 Write Data Memory (0CH)—The Write Data Memory command writes data to Data Memory. This command is equivalent to the LDE and LDEI instructions. Data can be written 1-65536 bytes at a time (65536 bytes can be written by setting size to zero). If the device is not in DEBUG mode or if the Read Protect Option Bit is enabled, the data is discarded.

```
DBG ← 0CH

DBG ← Data Memory Address[15:8]

DBG ← Data Memory Address[7:0]

DBG ← Size[15:8]

DBG ← Size[7:0]

DBG ← 1-65536 data bytes
```

• **Read Data Memory (0DH)**—The Read Data Memory command reads from Data Memory. This command is equivalent to the LDE and LDEI instructions. Data can be read 1-65536 bytes at a time (65536 bytes can be read by setting size to zero). If the device is not in DEBUG mode, this command returns FFH for the data.

```
\begin{array}{l} \mathtt{DBG} \leftarrow \mathtt{ODH} \\ \mathtt{DBG} \leftarrow \mathtt{Data} \ \mathtt{Memory} \ \mathtt{Address}[15 \mathpunct{:} 8] \end{array}
```

```
DBG \leftarrow Data Memory Address[7:0]

DBG \leftarrow Size[15:8]

DBG \leftarrow Size[7:0]

DBG \rightarrow 1-65536 data bytes
```

• Read Program Memory CRC (0EH)—The Read Program Memory CRC command computes and returns the CRC (cyclic redundancy check) of Program Memory using the 16-bit CRC-CCITT polynomial. If the device is not in DEBUG mode, this command returns FFFFH for the CRC value. Unlike most other OCD Read commands, there is a delay from issuing of the command until the OCD returns the data. The OCD reads the Program Memory, calculates the CRC value, and returns the result. The delay is a function of the Program Memory size and is approximately equal to the system clock period multiplied by the number of bytes in the Program Memory.

```
DBG \leftarrow 0EH
DBG \rightarrow CRC[15:8]
DBG \rightarrow CRC[7:0]
```

• Step Instruction (10H)—The Step Instruction command steps one assembly instruction at the current Program Counter (PC) location. If the device is not in DEBUG mode or the Read Protect Option Bit is enabled, the OCD ignores this command.

```
DBG ← 10H
```

• Stuff Instruction (11H)—The Stuff Instruction command steps one assembly instruction and allows specification of the first byte of the instruction. The remaining 0-4 bytes of the instruction are read from Program Memory. This command is useful for stepping over instructions where the first byte of the instruction has been overwritten by a Breakpoint. If the device is not in DEBUG mode or the Read Protect Option Bit is enabled, the OCD ignores this command.

```
DBG ← 11H
DBG ← opcode[7:0]
```

• Execute Instruction (12H)—The Execute Instruction command allows sending an entire instruction to be executed to the eZ8 CPU. This command can also step over Breakpoints. The number of bytes to send for the instruction depends on the opcode. If the device is not in DEBUG mode or the Read Protect Option Bit is enabled, the OCD ignores this command

```
DBG \leftarrow 12H
DBG \leftarrow 1-5 byte opcode
```

# **On-Chip Debugger Control Register Definitions**

### **OCD Control Register**

The OCD Control register (Table 102) controls the state of the On-Chip Debugger. This register enters or exits DEBUG mode and enables the BRK instruction. It can also reset the Z8F642x family, Z8R642x family device.

A 'reset and stop' function can be achieved by writing 81H to this register. A 'reset and go' function can be achieved by writing 41H to this register. If the device is in DEBUG mode, a 'run' function can be implemented by writing 40H to this register.

### Table 102. OCD Control Register (OCDCTL)

BITS	7	6	5	4	3	2	1	0
FIELD	DBGMODE	BRKEN	DBGACK	BRKLOOP			Reserved	RST
RESET				0				
R/W		R/W			R			R/W

#### DBGMODE—DEBUG Mode

Setting this bit to 1 causes the device to enter DEBUG mode. When in DEBUG mode, the eZ8 CPU stops fetching new instructions. Clearing this bit causes the eZ8 CPU to start running again. This bit is automatically set when a BRK instruction is decoded and Breakpoints are enabled. If the Read Protect Option Bit is enabled, this bit can only be cleared by resetting the device, it cannot be written to 0.

0 = The 64K Series device is operating in NORMAL mode.

1 = The 64K Series device is in DEBUG mode.

#### BRKEN—Breakpoint Enable

This bit controls the behavior of the BRK instruction (opcode 00H). By default, Breakpoints are disabled and the BRK instruction behaves like a NOP. If this bit is set to 1 and a BRK instruction is decoded, the OCD takes action dependent upon the BRKLOOP bit.

0 = BRK instruction is disabled.

1 = BRK instruction is enabled.

#### DBGACK—Debug Acknowledge

This bit enables the debug acknowledge feature. If this bit is set to 1, then the OCD sends an Debug Acknowledge character (FFH) to the host when a Breakpoint occurs.

0 = Debug Acknowledge is disabled.

1 = Debug Acknowledge is enabled.

## BRKLOOP—Breakpoint Loop

This bit determines what action the OCD takes when a BRK instruction is decoded if breakpoints are enabled (BRKEN is 1). If this bit is 0, then the DBGMODE bit is automatically set to 1 and the OCD entered DEBUG mode. If BRKLOOP is set to 1, then the

eZ8 CPU loops on the BRK instruction.

0 = BRK instruction sets DBGMODE to 1.

1 = eZ8 CPU loops on BRK instruction.

#### Reserved

These bits are reserved and must be 0.

#### RST—Reset

Setting this bit to 1 resets the 64K Series devices. The devices go through a normal Power-On Reset sequence with the exception that the On-Chip Debugger is not reset. This bit is automatically cleared to 0 when the reset finishes.

0 = No effect

1 = Reset the 64K Series device

### **OCD Status Register**

The OCD Status register (Table 103) reports status information about the current state of the debugger and the system.

### Table 103. OCD Status Register (OCDSTAT)

BITS	7	6	5	4	3	2	1	0
FIELD	IDLE	HALT	RPEN			Reserved		
RESET				(	)			
R/W				F	२			

IDLE—CPU idling

This bit is set if the part is in DEBUG mode (DBGMODE is 1), or if a BRK instruction occurred since the last time OCDCTL was written. This can be used to determine if the CPU is running or if it is idling.

0 = The eZ8 CPU is running.

1 = The eZ8 CPU is either stopped or looping on a BRK instruction.

#### HALT—HALT Mode

0 =The device is not in HALT mode.

1 = The device is in HALT mode.

#### RPEN—Read Protect Option Bit Enabled

0 = The Read Protect Option Bit is disabled (1).

1 = The Read Protect Option Bit is enabled (0), disabling many OCD commands.

#### Reserved

These bits are always 0.

# **On-Chip Oscillator**

### Overview

The products in the 64K Series feature an on-chip oscillator for use with external crystals with frequencies from 32 kHz to 20 MHz. In addition, the oscillator can support external RC networks with oscillation frequencies up to 4 MHz or ceramic resonators with oscillation frequencies up to 20 MHz. This oscillator generates the primary system clock for the internal eZ8  $^{\text{TM}}$  CPU and the majority of the on-chip peripherals. Alternatively, the  $X_{\text{IN}}$  input pin can also accept a CMOS-level clock input signal (32 kHz–20 MHz). If an external clock generator is used, the  $X_{\text{OUT}}$  pin must be left unconnected.

When configured for use with crystal oscillators or external clock drivers, the frequency of the signal on the  $X_{\rm IN}$  input pin determines the frequency of the system clock (that is, no internal clock divider). In RC operation, the system clock is driven by a clock divider (divide by 2) to ensure 50% duty cycle.

# **Operating Modes**

The 64K Series products support four different oscillator modes:

- On-chip oscillator configured for use with external RC networks (<4 MHz).
- Minimum power for use with very low frequency crystals (32 kHz to 1.0 MHz).
- Medium power for use with medium frequency crystals or ceramic resonators (0.5 MHz to 10.0 MHz).
- Maximum power for use with high frequency crystals or ceramic resonators (8.0 MHz to 20.0 MHz).

The oscillator mode is selected through user-programmable Option Bits. For more information, see Option Bits on page 195.

# **Crystal Oscillator Operation**

Figure 40 on page 212 displays a recommended configuration for connection with an external fundamental-mode, parallel-resonant crystal operating at 20 MHz. Recommended 20 MHz crystal specifications are provided in Table 104 on page 212. Resistor R1 is optional and limits total power dissipation by the crystal. The printed circuit board layout

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must add no more than 4 pF of stray capacitance to either the  $X_{IN}$  or  $X_{OUT}$  pins. If oscillation does not occur, reduce the values of capacitors C1 and C2 to decrease loading.

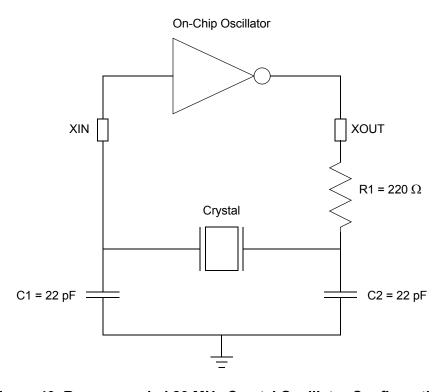


Figure 40. Recommended 20 MHz Crystal Oscillator Configuration

Table 104. Recommended Crystal Oscillator Specifications (20 MHz Operation)

Parameter	Value	Units	Comments
Frequency	20	MHz	
Resonance	Parallel		
Mode	Fundamental		
Series Resistance (R <sub>S</sub> )	25	Ω	Maximum
Load Capacitance (C <sub>L</sub> )	20	pF	Maximum
Shunt Capacitance (C <sub>0</sub> )	7	pF	Maximum
Drive Level	1	mW	Maximum

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# Oscillator Operation with an External RC Network

The External RC oscillator mode is applicable to timing insensitive applications. Figure 41 displays a recommended configuration for connection with an external resistor-capacitor (RC) network.

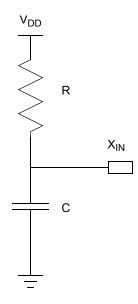


Figure 41. Connecting the On-Chip Oscillator to an External RC Network

An external resistance value of 45 k $\Omega$  is recommended for oscillator operation with an external RC network. The minimum resistance value to ensure operation is 40 k $\Omega$ . The typical oscillator frequency can be estimated from the values of the resistor (R in k $\Omega$ ) and capacitor (C in pF) elements using the following equation:

Oscillator Frequency (kHz) = 
$$\frac{1 \times 10^{6}}{(0.4 \times R \times C) + (4 \times C)}$$

Figure 42 displays the typical (3.3 V and 25 °C) oscillator frequency as a function of the capacitor (C in pF) employed in the RC network assuming a 45 k $\Omega$  external resistor. For very small values of C, the parasitic capacitance of the oscillator XIN pin and the printed circuit board should be included in the estimation of the oscillator frequency.

It is possible to operate the RC oscillator using only the parasitic capacitance of the package and printed circuit board. To minimize sensitivity to external parasitics, external capacitance values in excess of 20 pF are recommended.

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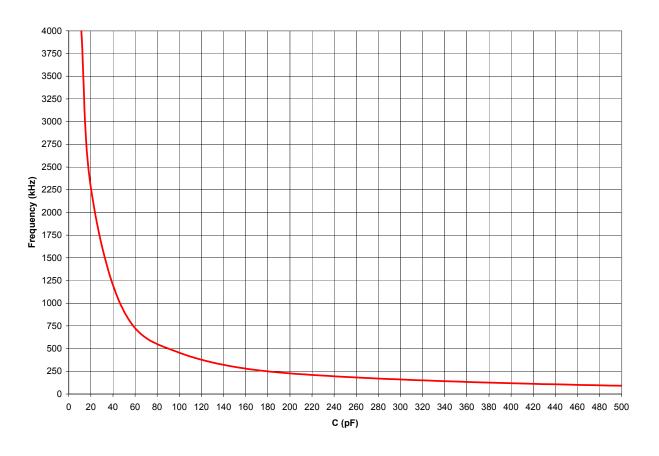


Figure 42. Typical RC Oscillator Frequency as a Function of the External Capacitance with a 45 k $\Omega$  Resistor



**Caution:** When using the external RC oscillator mode, the oscillator may stop oscillating if the power supply drops below 2.7 V, but before the power supply drops to the voltage brown-out threshold. The oscillator will resume oscillation as soon as the supply voltage exceeds 2.7 V.

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# **Electrical Characteristics**

# **Absolute Maximum Ratings**

Stresses greater than those listed in Table 105 may cause permanent damage to the device. These ratings are stress ratings only. Operation of the device at any condition outside those indicated in the operational sections of these specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability. For improved reliability, unused inputs must be tied to one of the supply voltages ( $V_{DD}$  or  $V_{SS}$ ).

**Table 105. Absolute Maximum Ratings** 

Parameter	Minimum	Maximum	Units	Notes
Ambient temperature under bias	-40	+125	С	
Storage temperature	-65	+150	С	
Voltage on any pin with respect to V <sub>SS</sub>	-0.3	+5.5	V	1
Voltage on V <sub>DD</sub> pin with respect to V <sub>SS</sub>	-0.3	+3.6	V	
Maximum current on input and/or inactive output pin	-5	+5	μΑ	
Maximum output current from active output pin	-25	+25	mA	
80-Pin QFP Maximum Ratings at –40 °C to 70 °C				
Total power dissipation		550	mW	
Maximum current into V <sub>DD</sub> or out of V <sub>SS</sub>		150	mA	
80-Pin QFP Maximum Ratings at 70 °C to 125 °C				
Total power dissipation		200	mW	
Maximum current into V <sub>DD</sub> or out of V <sub>SS</sub>		56	mA	
68-Pin PLCC Maximum Ratings at –40 °C to 70 °C				
Total power dissipation		1000	mW	
Maximum current into V <sub>DD</sub> or out of V <sub>SS</sub>		275	mA	
68-Pin PLCC Maximum Ratings at 70 °C to 125 °C				
Total power dissipation		500	mW	

**Table 105. Absolute Maximum Ratings (Continued)** 

Parameter	Minimum	Maximum	Units	Notes
Maximum current into V <sub>DD</sub> or out of V <sub>SS</sub>		140	mA	
64-Pin LQFP Maximum Ratings at –40 °C to 70 °C				
Total power dissipation		1000	mW	
Maximum current into V <sub>DD</sub> or out of V <sub>SS</sub>		275	mA	
64-Pin LQFP Maximum Ratings at 70 °C to 125 °C				
Total power dissipation		540	mW	
Maximum current into V <sub>DD</sub> or out of V <sub>SS</sub>		150	mA	
44-Pin PLCC Maximum Ratings at –40 °C to 70 °C				
Total power dissipation		750	mW	
Maximum current into V <sub>DD</sub> or out of V <sub>SS</sub>		200	mA	
44-Pin PLCC Maximum Ratings at 70 °C to 125 °C				
Total power dissipation		295	mW	
Maximum current into V <sub>DD</sub> or out of V <sub>SS</sub>		83	mA	
44-Pin LQFP Maximum Ratings at –40 °C to 70 °C				
Total power dissipation		750	mW	
Maximum current into V <sub>DD</sub> or out of V <sub>SS</sub>		200	mA	
44-Pin LQFP Maximum Ratings at 70 °C to 125 °C				
Total power dissipation		360	mW	
Maximum current into V <sub>DD</sub> or out of V <sub>SS</sub>		100	mA	
40-Pin PDIP Maximum Ratings at -40 °C to 70 °C				
Total power dissipation		1000	mW	
Maximum current into V <sub>DD</sub> or out of V <sub>SS</sub>		275	mA	
40-Pin PDIP Maximum Ratings at 70 °C to 125 °C				
Total power dissipation		540	mW	
Maximum current into V <sub>DD</sub> or out of V <sub>SS</sub>		150	mA	

**Note:** This voltage applies to all pins except the following: VDD, AVDD, pins supporting analog input (Ports B and H), RESET, and where noted otherwise.

# **DC Characteristics**

Table 106 lists the DC characteristics of the 64K Series products. All voltages are referenced to  $V_{SS}$ , the primary system ground.

### **Table 106. DC Characteristics**

		T <sub>A</sub> = -40 °	C to 125 °	С		
Symbol	Parameter	Minimum	Typical	Maximum	Units	Conditions
$V_{DD}$	Supply Voltage	3.0	_	3.6	V	
V <sub>IL1</sub>	Low Level Input Voltage	-0.3	_	0.3*V <sub>DD</sub>	V	For all input pins except RESET, DBG, XIN
$V_{IL2}$	Low Level Input Voltage	-0.3	_	0.2*V <sub>DD</sub>	V	For RESET, DBG, and XIN.
V <sub>IH1</sub>	High Level Input Voltage	0.7*V <sub>DD</sub>	_	5.5	V	Port A, C, D, E, F, and G pins.
V <sub>IH2</sub>	High Level Input Voltage	0.7*V <sub>DD</sub>	_	V <sub>DD</sub> +0.3	V	Port B and H pins.
V <sub>IH3</sub>	High Level Input Voltage	0.8*V <sub>DD</sub>	_	V <sub>DD</sub> +0.3	V	RESET, DBG, and XIN pins
V <sub>OL1</sub>	Low Level Output Voltage Standard Drive	_	_	0.4	V	I <sub>OL</sub> = 2 mA; VDD = 3.0 V High Output Drive disabled.
V <sub>OH1</sub>	High Level Output Voltage Standard Drive	2.4	_	_	V	I <sub>OH</sub> = -2 mA; VDD = 3.0 V High Output Drive disabled.
V <sub>OL2</sub>	Low Level Output Voltage High Drive	_	_	0.6	V	$I_{OL}$ = 20 mA; VDD = 3.3 V High Output Drive enabled $T_A$ = -40 °C to +70 °C
V <sub>OH2</sub>	High Level Output Voltage High Drive	2.4	_	-	V	$I_{OH}$ = -20 mA; VDD = 3.3 V High Output Drive enabled; $T_A$ = -40 °C to +70 °C
V <sub>OL3</sub>	Low Level Output Voltage High Drive	_	_	0.6	V	$I_{OL}$ = 15 mA; VDD = 3.3 V High Output Drive enabled; $T_A$ = +70 °C to +105 °C
V <sub>OH3</sub>	High Level Output Voltage High Drive	2.4	_	-	V	$I_{OH}$ = 15 mA; VDD = 3.3 V High Output Drive enabled; $T_A$ = +70 °C to +105 °C
$V_{RAM}$	RAM Data Retention	0.7	_	_	V	
I <sub>IL</sub>	Input Leakage Current	-5	_	+5	μΑ	V <sub>DD</sub> = 3.6 V; V <sub>IN</sub> = VDD or VSS <sup>1</sup>
I <sub>TL</sub>	Tri-State Leakage Current	-5	_	+5	μА	V <sub>DD</sub> = 3.6 V



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# **Table 106. DC Characteristics (Continued)**

		T <sub>A</sub> = -40 °	C to 125 °	С		
Symbol	Parameter	Minimum	Typical	Maximum	Units	Conditions
C <sub>PAD</sub>	GPIO Port Pad Capacitance	_	8.0 <sup>2</sup>	_	pF	
C <sub>XIN</sub>	XIN Pad Capacitance	_	8.0 <sup>2</sup>	_	pF	
C <sub>XOUT</sub>	XOUT Pad Capacitance	_	9.5 <sup>2</sup>	_	pF	
I <sub>PU</sub>	Weak Pull-up Current	30	100	350	mA	V <sub>DD</sub> = 3.0 - 3.6 V
I <sub>DDA</sub>	Active Mode Supply Current (See Figure 43 on	_	11	16 12	mA	V <sub>DD</sub> = 3.6 V, Fsysclk = 20 MHz V <sub>DD</sub> = 3.3 V
	page 220 and Figure 44 on page 221) GPIO pins configured as outputs	_	9	11 9	mA	$V_{DD}$ = 3.6 V, Fsysclk = 10 MHz $V_{DD}$ = 3.3 V
I <sub>DDH</sub>	HALT Mode Supply Current (See Figure 45 on		4	7 5	mA	V <sub>DD</sub> = 3.6 V, Fsysclk = 20 MHz V <sub>DD</sub> = 3.3 V
	page 222 and Figure 46 on page 223) GPIO pins configured as outputs	_	3	5 4	mA	V <sub>DD</sub> = 3.6 V, Fsysclk = 10 MHz V <sub>DD</sub> = 3.3 V

# **Table 106. DC Characteristics (Continued)**

		T <sub>A</sub> = -40 °	C to 125 °	С		
Symbol	Parameter	Minimum	Typical	Maximum	Units	Conditions
I <sub>DDS</sub>	Stop Mode Supply Current	_	520	700	μА	V <sub>DD</sub> = 3.6 V, VBO and WDT Enabled
	(See Figure 47 and			650		V <sub>DD</sub> = 3.3 V
	Figure 48) GPIO pins configured as outputs	_	10	25	μА	$V_{DD}$ = 3.6 V, $T_A$ = 0 to 70 °C VBO
						Disabled
				20		WDT
						Enabled V <sub>DD</sub> = 3.3 V
		_		80	μΑ	V <sub>DD</sub> = 3.6 V, T <sub>A</sub> = -40 to +105 °C
						VBO
				70		Disabled
						WDT
						Enabled V <sub>DD</sub> = 3.3 V
		_		250	μА	$V_{DD}$ = 3.6 V, $T_{A}$ = -40 to +125 °C
						VBO
				150		Disabled
						WDT
						Enabled V <sub>DD</sub> = 3.3 V

<sup>&</sup>lt;sup>1</sup>This condition excludes all pins that have on-chip pull-ups, when driven Low.

<sup>&</sup>lt;sup>2</sup>These values are provided for design guidance only and are not tested in production.

Figure 43 displays the typical active mode current consumption while operating at 25 °C versus the system clock frequency. All GPIO pins are configured as outputs and driven High.

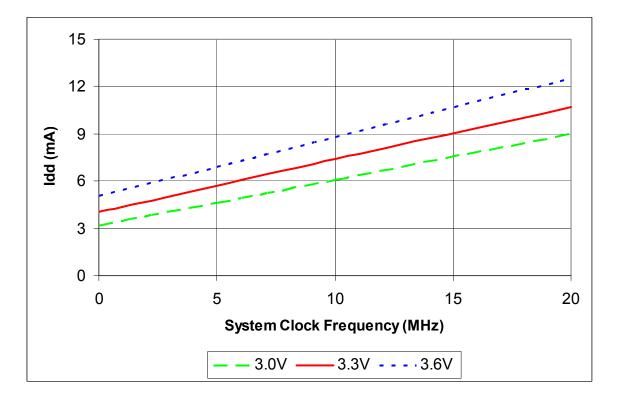


Figure 43. Typical Active Mode Idd Versus System Clock Frequency

Figure 44 displays the maximum active mode current consumption across the full operating temperature range of the device and versus the system clock frequency. All GPIO pins are configured as outputs and driven High.

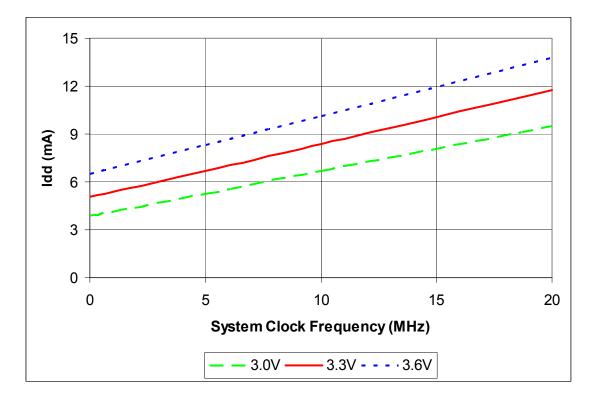


Figure 44. Maximum Active Mode Idd Versus System Clock Frequency

Figure 45 displays the typical current consumption in HALT mode while operating at 25 °C versus the system clock frequency. All GPIO pins are configured as outputs and driven High.

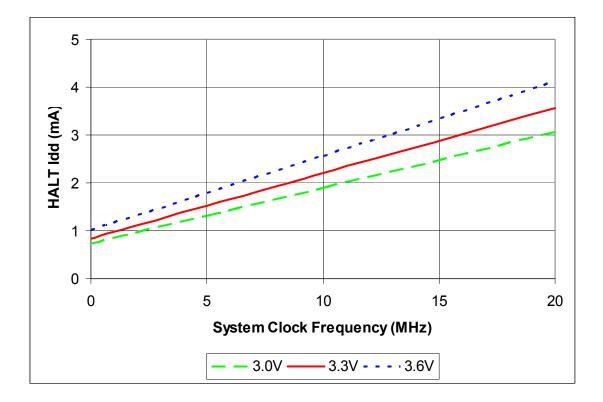


Figure 45. Typical HALT Mode Idd Versus System Clock Frequency

Figure 45 displays the maximum HALT mode current consumption across the full operating temperature range of the device and versus the system clock frequency. All GPIO pins are configured as outputs and driven High.

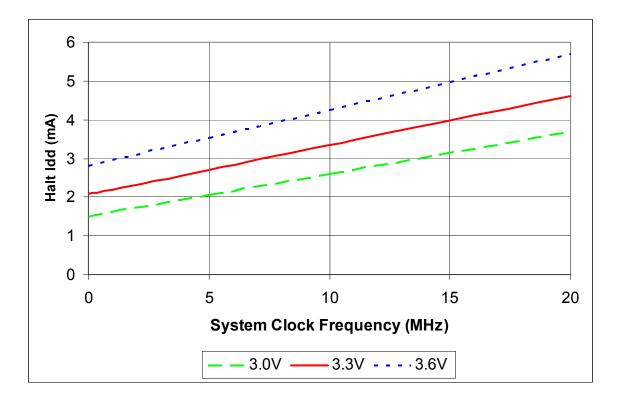


Figure 46. Maximum HALT Mode Icc Versus System Clock Frequency

Figure 47 displays the maximum current consumption in STOP mode with the VBO and Watchdog Timer enabled versus the power supply voltage. All GPIO pins are configured as outputs and driven High.

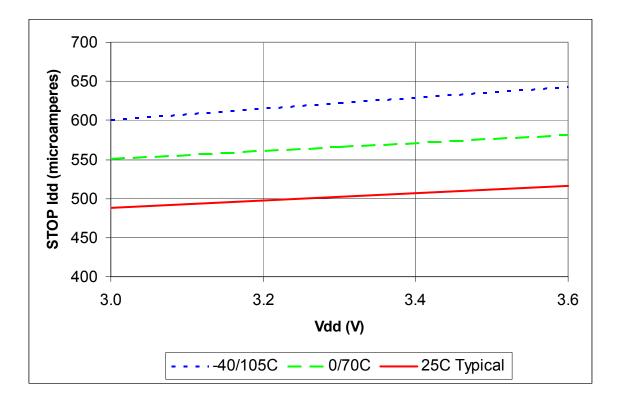


Figure 47. Maximum STOP Mode Idd with VBO enabled versus Power Supply Voltage

Figure 48 displays the maximum current consumption in STOP mode with the VBO disabled and Watchdog Timer enabled versus the power supply voltage. All GPIO pins are configured as outputs and driven High. Disabling the Watchdog Timer and its internal RC oscillator in STOP mode will provide some additional reduction in STOP mode current consumption. This small current reduction would be indistinguishable on the scale of Figure 48.

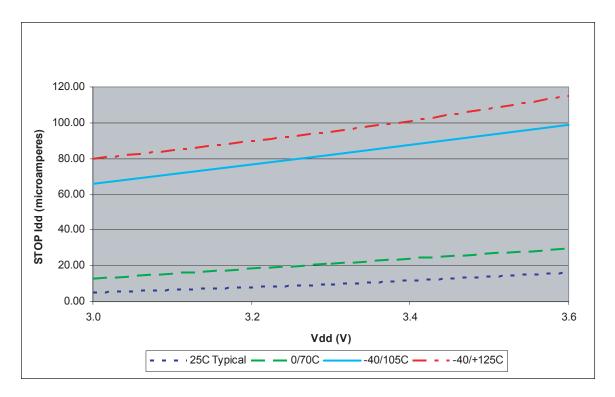


Figure 48. Maximum STOP Mode Idd with VBO Disabled versus Power Supply Voltage

# On-Chip Peripheral AC and DC Electrical Characteristics

Table 107. Power-On Reset and Voltage Brownout Electrical Characteristics and Timing

		T <sub>A</sub> = -40 °	C to 125 °	С		
Symbol	Parameter	Minimum	Typical <sup>1</sup>	Maximum	Units	Conditions
V <sub>POR</sub>	Power-On Reset Voltage Threshold	2.40	2.70	2.90	V	$V_{DD} = V_{POR}$
$V_{VBO}$	Voltage Brownout Reset Voltage Threshold	2.30	2.60	2.85	V	$V_{DD} = V_{VBO}$
	V <sub>POR</sub> to V <sub>VBO</sub> hysteresis	50	100	_	mV	
	Starting V <sub>DD</sub> voltage to ensure valid Power-On Reset.	_	V <sub>SS</sub>	_	V	
T <sub>ANA</sub>	Power-On Reset Analog Delay	_	50	_	μS	V <sub>DD</sub> > V <sub>POR</sub> ; T <sub>POR</sub> Digital Reset delay follows T <sub>ANA</sub>
T <sub>POR</sub>	Power-On Reset Digital Delay	_	6.6	_	ms	66 WDT Oscillator cycles (10 kHz) + 16 System Clock cycles (20 MHz)
T <sub>VBO</sub>	Voltage Brownout Pulse Rejection Period	_	10	_	μS	V <sub>DD</sub> < V <sub>VBO</sub> to generate a Reset.
T <sub>RAMP</sub>	Time for VDD to transition from V <sub>SS</sub> to V <sub>POR</sub> to ensure valid Reset	0.10	_	100	ms	

<sup>&</sup>lt;sup>1</sup>Data in the typical column is from characterization at 3.3 V and 0 °C. These values are provided for design guidance only and are not tested in production.

Table 108. External RC Oscillator Electrical Characteristics and Timing

		T <sub>A</sub> = -40 °C to 125 °C				
Symbol	Parameter	Minimum	Typical <sup>1</sup>	Maximum	Units	Conditions
V <sub>DD</sub>	Operating Voltage Range	2.70 <sup>1</sup>	_	-	V	
R <sub>EXT</sub>	External Resistance from XIN to VDD	40	45	200	kΩ	$V_{DD} = V_{VBO}$
C <sub>EXT</sub>	External Capacitance from XIN to VSS	0	20	1000	pF	
Fosc	External RC Oscillation Frequency	-	_	4	MHz	

 $<sup>^{\</sup>mathrm{l}}$  When using the external RC oscillator mode, the oscillator may stop oscillating if the power supply drops below 2.7 V, but before the power supply drops to the voltage brown-out threshold. The oscillator will resume oscillation as soon as the supply voltage exceeds 2.7 V.

Table 109. Reset and Stop Mode Recovery Pin Timing

		T <sub>A</sub> = -40 °C to 125 °C				
Symbol	Parameter	Minimum	Typical	Maximum	Units	Conditions
T <sub>RESET</sub>	RESET pin assertion to initiate a system reset.	4	_	_	T <sub>CLK</sub>	Not in STOP Mode. T <sub>CLK</sub> = System Clock period.
T <sub>SMR</sub>	Stop Mode Recovery pin Pulse Rejection Period	10	20	40	ns	RESET, DBG, and GPIO pins configured as SMR sources.

Table 110 list the Flash Memory electrical characteristics and timing.

**Table 110. Flash Memory Electrical Characteristics and Timing** 

	V <sub>DD</sub> = 3.0- T <sub>A</sub> = -40 °		С		
Parameter	Minimum Typical Maximu		Maximum	Units	Notes
Flash Byte Read Time	50	_	_	ns	
Flash Byte Program Time	20	_	40	μs	
Flash Page Erase Time	10	_	_	ms	
Flash Mass Erase Time	200	_	_	ms	
Writes to Single Address Before Next Erase	_	_	2		
Flash Row Program Time	-	_	8	ms	Cumulative program time for single row cannot exceed limit before next erase. This parameter is only an issue when bypassing the Flash Controller.
Data Retention	100	_	_	years	25 °C
Endurance, –40 °C to 105 °C	10,000	_	_	cycles	Program/erase cycles
Endurance, 106 °C to 125 °C	1,000	-	_	cycles	Program/erase cycles

Table 111 lists the Watchdog Timer electrical characteristics and timing.

**Table 111. Watchdog Timer Electrical Characteristics and Timing** 

		V <sub>DD</sub> = 3.0- T <sub>A</sub> = -40 °	V <sub>DD</sub> = 3.0–3.6 V T <sub>A</sub> = -40 °C to 125 °C			
Symbol	Parameter	Minimum	Typical	Maximum	Units	Conditions
F <sub>WDT</sub>	WDT Oscillator Frequency	5	10	20	kHz	
I <sub>WDT</sub>	WDT Oscillator Current including internal RC oscillator	-	< 1	5	μА	

Table 112 provides electrical characteristics and timing information for the Analog-to-Digital Converter. Figure 49 displays the input frequency response of the ADC.

Table 112. Analog-to-Digital Converter Electrical Characteristics and Timing

		V <sub>Γ</sub> T <sub>A</sub> = -	<sub>DD</sub> = 3.0–3 –40 °C to	.6 V 125 °C			
Symbol	Parameter	Minimum	Typical	Maximum	Units	Conditions	
	Resolution	10	_	_	bits	External V <sub>REF</sub> = 3.0 V;	
	Differential Nonlinearity (DNL)	-1.0		+1.0	Isb	Guaranteed by design	
	Integral Nonlinearity (INL)	-3.0	<u>+</u> 1.0	3.0	lsb	External V <sub>REF</sub> = 3.0 V	
	DC Offset Error	-35	_	25	mV		
	DC Offset Error	-50	-	25	mV	44-pin LQFP, 44-pin PLCC, and 68-pin PLCC packages.	
V <sub>REF</sub>	Internal Reference Voltage	1.9	2.0	2.4	V	V <sub>DD</sub> = 3.0 - 3.6 V T <sub>A</sub> = -40 °C to 105 °C	
VC <sub>REF</sub>	Voltage Coefficient of Internal Reference Voltage	_	78	_	mV/V	V <sub>REF</sub> variation as a function of AVDD.	
TC <sub>REF</sub>	Temperature Coefficient of Internal Reference Voltage	_	1	_	mV/°C		
	Single-Shot Conversion Period	_	5129	_	cycles	System clock cycles	
	Continuous Conversion Period	_	256	_	cycles	System clock cycles	
R <sub>S</sub>	Analog Source Impedance	_	_	150	Ω	Recommended	
Zin	Input Impedance		150		kΩ		
V <sub>REF</sub>	External Reference Voltage			AVDD	V	AVDD <= VDD. When using an external reference voltage, decoupling capacitance should be placed from VREF to AVSS.	
I <sub>REF</sub>	Current draw into VREF pin when driving with external source.		25.0	40.0	μА		

### ADC Magnitude Transfer Function (Linear Scale)

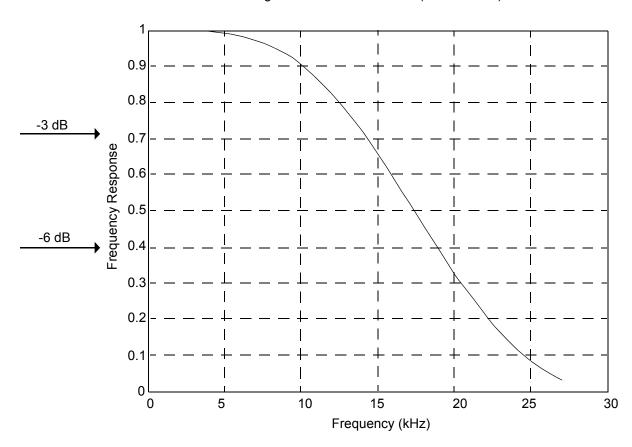


Figure 49. Analog-to-Digital Converter Frequency Response

# **AC Characteristics**

The section provides information on the AC characteristics and timing. All AC timing information assumes a standard load of 50 pF on all outputs. Table 113 lists the 64K Series AC characteristics and timing.

### **Table 113. AC Characteristics**

		V <sub>DD</sub> = 3.0–3.6V T <sub>A</sub> = -40 °C to 125 °C							
Symbol	Parameter	Minimum	Maximum	Units	Conditions				
F <sub>sysclk</sub>	System Clock Frequency	_	20.0	MHz	Read-only from Flash memory.				
		0.032768	20.0	MHz	Program or erasure of the Flash memory.				
F <sub>XTAL</sub>	Crystal Oscillator Frequency	0.032768	20.0	MHz	System clock frequencies below the crystal oscillator minimum require an external clock driver.				
T <sub>XIN</sub>	Crystal Oscillator Clock Period	50	_	ns	T <sub>CLK</sub> = 1/F <sub>sysclk</sub>				
T <sub>XINH</sub>	System Clock High Time	20		ns					
T <sub>XINL</sub>	System Clock Low Time	20		ns					
T <sub>XINR</sub>	System Clock Rise Time	_	3	ns	T <sub>CLK</sub> = 50 ns. Slower rise times can be tolerated with longer clock periods.				
T <sub>XINF</sub>	System Clock Fall Time	_	3	ns	T <sub>CLK</sub> = 50 ns. Slower fall times can be tolerated with longer clock periods.				

## General-Purpose I/O Port Input Data Sample Timing

Figure 50 displays timing of the GPIO Port input sampling. Table 114 lists the GPIO port input timing.

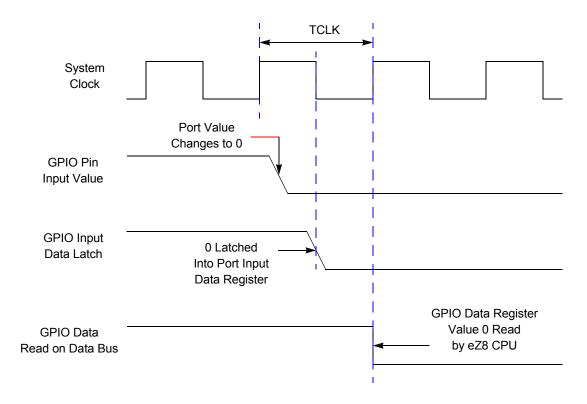


Figure 50. Port Input Sample Timing

**Table 114. GPIO Port Input Timing** 

		Dela	y (ns)
Parameter	Abbreviation	Min	Max
T <sub>S_PORT</sub>	Port Input Transition to XIN Fall Setup Time (Not pictured)	5	-
T <sub>H_PORT</sub>	XIN Fall to Port Input Transition Hold Time (Not pictured)	6	_
T <sub>SMR</sub>	GPIO Port Pin Pulse Width to Insure Stop Mode Recovery (for GPIO Port Pins enabled as SMR sources)	1 μs	

# **General-Purpose I/O Port Output Timing**

Figure 51 and Table 115 provide timing information for GPIO Port pins.

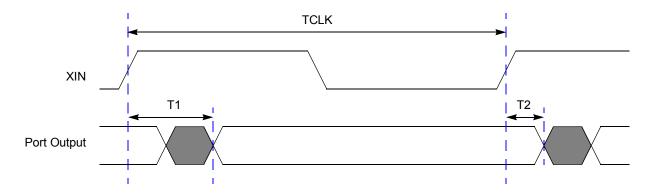


Figure 51. GPIO Port Output Timing

**Table 115. GPIO Port Output Timing** 

		Dela	y (ns)
Parameter	Abbreviation	Minimum	Maximum
GPIO Port p	oins		
T <sub>1</sub>	XIN Rise to Port Output Valid Delay	_	20
T <sub>2</sub>	XIN Rise to Port Output Hold Time	2	_
-			

# **On-Chip Debugger Timing**

Figure 52 and Table 116 provide timing information for the DBG pin. The DBG pin timing specifications assume a 4 µs maximum rise and fall time.

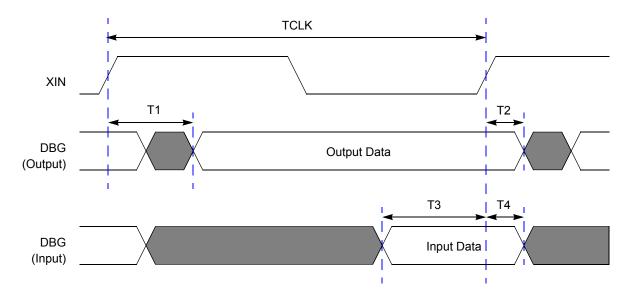


Figure 52. On-Chip Debugger Timing

Table 116. On-Chip Debugger Timing

		Delay (ns)		
Parameter	Abbreviation	Minimum	Maximum	
DBG				
T <sub>1</sub>	XIN Rise to DBG Valid Delay	_	30	
T <sub>2</sub>	XIN Rise to DBG Output Hold Time	2	_	
T <sub>3</sub>	DBG to XIN Rise Input Setup Time	10	_	
T <sub>4</sub>	DBG to XIN Rise Input Hold Time	5	_	
	DBG frequency		System Clock/4	

## **SPI Master Mode Timing**

Figure 53 and Table 117 provide timing information for SPI Master mode pins. Timing is shown with SCK rising edge used to source MOSI output data, SCK falling edge used to sample MISO input data. Timing on the SS output pin(s) is controlled by software.

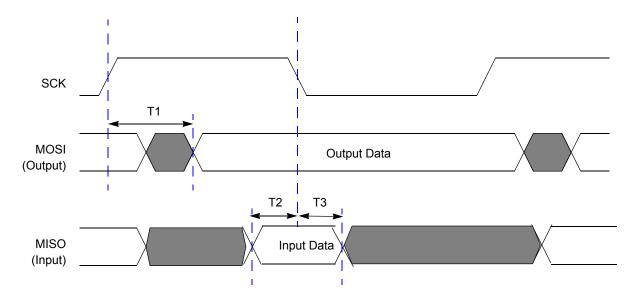


Figure 53. SPI Master Mode Timing

**Table 117. SPI Master Mode Timing** 

	Delay (ns)		
Abbreviation	Min	Max	
SCK Rise to MOSI output Valid Delay	-5	+5	
MISO input to SCK (receive edge) Setup Time	20		
MISO input to SCK (receive edge) Hold Time	0		
	SCK Rise to MOSI output Valid Delay MISO input to SCK (receive edge) Setup Time	Abbreviation Min  SCK Rise to MOSI output Valid Delay -5  MISO input to SCK (receive edge) Setup Time 20	

# **SPI Slave Mode Timing**

Figure 54 and Table 118 provide timing information for the SPI slave mode pins. Timing is shown with SCK rising edge used to source MISO output data, SCK falling edge used to sample MOSI input data.

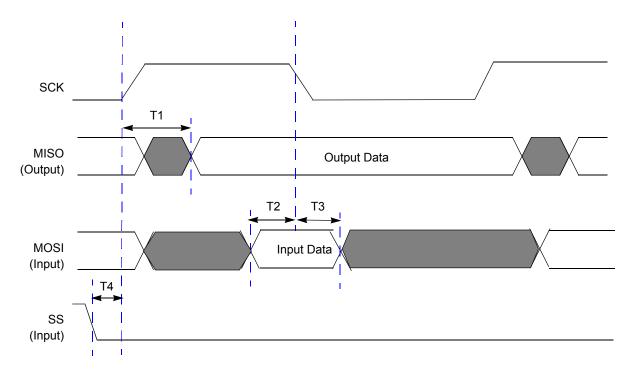


Figure 54. SPI Slave Mode Timing

**Table 118. SPI Slave Mode Timing** 

		Delay (ns)		
Parameter	Abbreviation	Minimum	Maximum	
SPI Slave				
T <sub>1</sub>	SCK (transmit edge) to MISO output Valid Delay	2 * Xin period	3 * Xin period + 20 nsec	
T <sub>2</sub>	MOSI input to SCK (receive edge) Setup Time	0		
T <sub>3</sub>	MOSI input to SCK (receive edge) Hold Time	3 * Xin period		
T <sub>4</sub>	SS input assertion to SCK setup	1 * Xin period		

# I<sup>2</sup>C Timing

Figure 55 and Table 119 provide timing information for I<sup>2</sup>C pins.

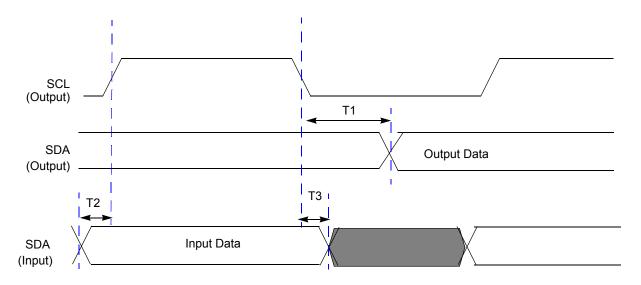


Figure 55. I<sup>2</sup>C Timing

Table 119. I<sup>2</sup>C Timing

		Delay (ns)		
Parameter	Abbreviation	Minimum Maximum		
I <sup>2</sup> C				
T <sub>1</sub>	SCL Fall to SDA output delay	SCL period/4		
T <sub>2</sub>	SDA Input to SCL rising edge Setup Time	0		
T <sub>3</sub>	SDA Input to SCL falling edge Hold Time	0		

# **UART Timing**

Figure 56 and Table 120 provide timing information for UART pins for the case where the Clear To Send input pin (CTS) is used for flow control. In this example, it is assumed that the Driver Enable polarity has been configured to be Active Low and is represented here by DE. The CTS to DE assertion delay (T1) assumes the UART Transmit Data register has been loaded with data prior to CTS assertion.

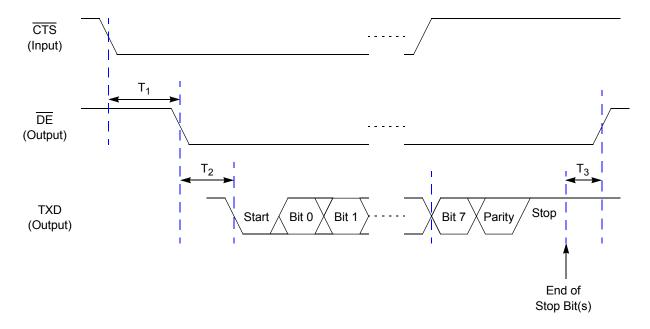


Figure 56. UART Timing with CTS

Table 120. UART Timing with CTS

		Dela	y (ns)
Paramet	er Abbreviation	Minimum	Maximum
T <sub>1</sub>	CTS Fall to DE Assertion Delay	2 * XIN period	2 * XIN period + 1 Bit period
T <sub>2</sub>	DE Assertion to TXD Falling Edge (Start) Delay	1 Bit period	1 Bit period + 1 * XIN period
T <sub>3</sub>	End of Stop Bit(s) to DE Deassertion Delay	1 * XIN period	2 * XIN period

Figure 57 and Table 121 provide timing information for UART pins for the case where the Clear To Send input signal ( $\overline{CTS}$ ) is not used for flow control. In this example, it is assumed that the Driver Enable polarity has been configured to be Active Low and is represented here by  $\overline{DE}$ .  $\overline{DE}$  asserts after the UART Transmit Data Register has been written.  $\overline{DE}$  remains asserted for multiple characters as long as the Transmit Data register is written with the next character before the current character has completed.

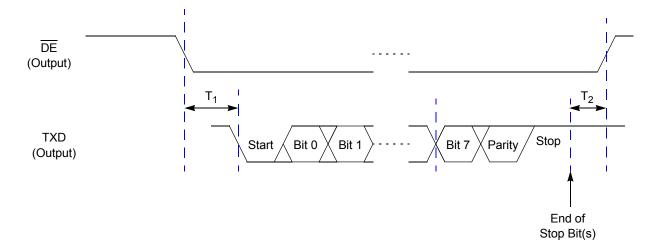


Figure 57. UART Timing without CTS

Table 121. UART Timing without CTS

		Dela	y (ns)
Parameter	Abbreviation	Minimum	Maximum
T <sub>1</sub>	DE Assertion to TXD Falling Edge (Start) Delay	1 Bit period	1 Bit period + 1 * XIN period
T <sub>2</sub>	End of Stop Bit(s) to $\overline{\text{DE}}$ Deassertion Delay	1 * XIN period	2 * XIN period

PS019919-1207 Electrical Characteristics

### Z8 Encore! XP<sup>®</sup> 64K Series Flash Microcontrollers **Product Specification**

zilog 240

PS019919-1207 **Electrical Characteristics** 

## eZ8™ CPU Instruction Set

#### **Assembly Language Programming Introduction**

The eZ8 CPU assembly language provides a means for writing an application program without having to be concerned with actual memory addresses or machine instruction formats. A program written in assembly language is called a source program. Assembly language allows the use of symbolic addresses to identify memory locations. It also allows mnemonic codes (opcodes and operands) to represent the instructions themselves. The opcodes identify the instruction while the operands represent memory locations, registers, or immediate data values.

Each assembly language program consists of a series of symbolic commands called statements. Each statement can contain labels, operations, operands and comments.

Labels can be assigned to a particular instruction step in a source program. The label identifies that step in the program as an entry point for use by other instructions.

The assembly language also includes assembler directives that supplement the machine instruction. The assembler directives, or pseudo-ops, are not translated into a machine instruction. Rather, the pseudo-ops are interpreted as directives that control or assist the assembly process.

The source program is processed (assembled) by the assembler to obtain a machine language program called the object code. The object code is executed by the eZ8 CPU. An example segment of an assembly language program is detailed in the following example.

#### **Assembly Language Source Program Example**

JP START	; Everything after the semicolon is a comment.
START:	; A label called "START". The first instruction (JP START) in this ; example causes program execution to jump to the point within the ; program where the START label occurs.
LD R4, R7	; A Load (LD) instruction with two operands. The first operand, ; Working Register R4, is the destination. The second operand, ; Working Register R7, is the source. The contents of R7 is ; written into R4.
LD 234H, #%01	; Another Load (LD) instruction with two operands. ; The first operand, Extended Mode Register Address 234H, ; identifies the destination. The second operand, Immediate Data

; value 01H, is the source. The value 01H is written into the

; Register at address 234H.

#### **Assembly Language Syntax**

For proper instruction execution, eZ8 CPU assembly language syntax requires that the operands be written as 'destination, source'. After assembly, the object code usually has the operands in the order 'source, destination', but ordering is opcode-dependent. The following instruction examples illustrate the format of some basic assembly instructions and the resulting object code produced by the assembler. This binary format must be followed if you prefer manual program coding or intend to implement your own assembler.

**Example 1**: If the contents of Registers 43H and 08H are added and the result is stored in 43H, the assembly syntax and resulting object code is:

#### **Assembly Language Syntax Example 1**

Assembly Language Code	ADD	43H,	H80	(ADD dst, src)
Object Code	04	80	43	(OPC src, dst)

**Example 2**: In general, when an instruction format requires an 8-bit register address, that address can specify any register location in the range 0-255 or, using Escaped Mode Addressing, a Working Register R0 - R15. If the contents of Register 43H and Working Register R8 are added and the result is stored in 43H, the assembly syntax and resulting object code is:

#### **Assembly Language Syntax Example 2**

Assembly Language Code	ADD	43H,	R8	(ADD dst, src)
Object Code	04	E8	43	(OPC src, dst)

Refer to the device-specific Product Specification to determine the exact register file range available. The register file size varies, depending on the device type.

#### **eZ8 CPU Instruction Notation**

In the eZ8 CPU Instruction Summary and Description sections, the operands, condition codes, status Flags, and address modes are represented by a notational shorthand that is described in Table 122.

**Table 122. Notational Shorthand** 

Notation	Description	Operand	Range
b	Bit	b	b represents a value from 0 to 7 (000B to 111B).
СС	Condition Code	_	Refer to Condition Codes overview in the eZ8 CPU User Manual.
DA	Direct Address	Addrs	Addrs. represents a number in the range of 0000H to FFFFH
ER	Extended Addressing Register	Reg	Reg. represents a number in the range of 000H to FFFH
IM	Immediate Data	#Data	Data is a number between 00H to FFH
Ir	Indirect Working Register	@Rn	n = 0 –15
IR	Indirect Register	@Reg	Reg. represents a number in the range of 00H to FFH
Irr	Indirect Working Register Pair	@RRp	p = 0, 2, 4, 6, 8, 10, 12, or 14
IRR	Indirect Register Pair	@Reg	Reg. represents an even number in the range 00H to FEH
р	Polarity	р	Polarity is a single bit binary value of either 0B or 1B.
r	Working Register	Rn	n = 0 – 15
R	Register	Reg	Reg. represents a number in the range of 00H to FFH
RA	Relative Address	Х	X represents an index in the range of +127 to -128 which is an offset relative to the address of the next instruction
rr	Working Register Pair	RRp	p = 0, 2, 4, 6, 8, 10, 12, or 14
RR	Register Pair	Reg	Reg. represents an even number in the range of 00H to FEH
Vector	Vector Address	Vector	Vector represents a number in the range of 00H to FFH
X	Indexed	#Index	The register or register pair to be indexed is offset by the signed Index value (#Index) in a +127 to -128 range.

Table 123 contains additional symbols that are used throughout the Instruction Summary and Instruction Set Description sections.

**Table 123. Additional Symbols** 

Symbol	Definition
dst	Destination Operand
src	Source Operand
@	Indirect Address Prefix
SP	Stack Pointer
PC	Program Counter
FLAGS	Flags Register
RP	Register Pointer
#	Immediate Operand Prefix
В	Binary Number Suffix
%	Hexadecimal Number Prefix
Н	Hexadecimal Number Suffix

Assignment of a value is indicated by an arrow. For example,

$$dst \leftarrow dst + src$$

indicates the source data is added to the destination data and the result is stored in the destination location.

#### **Condition Codes**

The C, Z, S and V Flags control the operation of the conditional jump (JP cc and JR cc) instructions. Sixteen frequently useful functions of the Flag settings are encoded in a 4-bit field called the condition code (cc), which forms Bits 7:4 of the conditional jump instructions. The condition codes are summarized in Table 124. Some binary condition codes can be created using more than one assembly code mnemonic. The result of the Flag test operation decides if the conditional jump is executed.

**Table 124. Condition Codes** 

Binary	Hex	Assembly Mnemonic	Definition	Flag Test Operation
0000	0	F	Always False	_
0001	1	LT	Less Than	(S XOR V) = 1
0010	2	LE	Less Than or Equal	(Z OR (S XOR V)) = 1

**Table 124. Condition Codes (Continued)** 

Binary	Hex	Assembly Mnemonic	Definition	Flag Test Operation
0011	3	ULE	Unsigned Less Than or Equal	(C OR Z) = 1
0100	4	OV	Overflow	V = 1
0101	5	MI	Minus	S = 1
0110	6	Z	Zero	Z = 1
0110	6	EQ	Equal	Z = 1
0111	7	С	Carry	C = 1
0111	7	ULT	Unsigned Less Than	C = 1
1000	8	T (or blank)	Always True	-
1001	9	GE	Greater Than or Equal	(S XOR V) = 0
1010	Α	GT	Greater Than	(Z OR (S XOR V)) = 0
1011	В	UGT	Unsigned Greater Than	(C = 0  AND  Z = 0) = 1
1100	С	NOV	No Overflow	V = 0
1101	D	PL	Plus	S = 0
1110	Е	NZ	Non-Zero	Z = 0
1110	Е	NE	Not Equal	Z = 0
1111	F	NC	No Carry	C = 0
1111	F	UGE	Unsigned Greater Than or Equal	C = 0
		· · · · · · · · · · · · · · · · · · ·	·	· · · · · · · · · · · · · · · · · · ·

### **eZ8 CPU Instruction Classes**

eZ8 CPU instructions can be divided functionally into the following groups:

- Arithmetic
- Bit Manipulation
- Block Transfer
- CPU Control
- Load
- Logical
- Program Control
- Rotate and Shift

Table 125 through Table 132 contain the instructions belonging to each group and the number of operands required for each instruction. Some instructions appear in more than one table as these instruction can be considered as a subset of more than one category. Within these tables, the source operand is identified as 'src', the destination operand is 'dst' and a condition code is 'cc'.

**Table 125. Arithmetic Instructions** 

Mnemonic	Operands	Instruction
ADC	dst, src	Add with Carry
ADCX	dst, src	Add with Carry using Extended Addressing
ADD	dst, src	Add
ADDX	dst, src	Add using Extended Addressing
СР	dst, src	Compare
CPC	dst, src	Compare with Carry
CPCX	dst, src	Compare with Carry using Extended Addressing
СРХ	dst, src	Compare using Extended Addressing
DA	dst	Decimal Adjust
DEC	dst	Decrement
DECW	dst	Decrement Word
INC	dst	Increment
INCW	dst	Increment Word
MULT	dst	Multiply
SBC	dst, src	Subtract with Carry
SBCX	dst, src	Subtract with Carry using Extended Addressing
SUB	dst, src	Subtract
SUBX	dst, src	Subtract using Extended Addressing

**Table 126. Bit Manipulation Instructions** 

Mnemonic	Operands	Instruction
BCLR	bit, dst	Bit Clear
BIT	p, bit, dst	Bit Set or Clear
BSET	bit, dst	Bit Set

**Table 126. Bit Manipulation Instructions (Continued)** 

Mnemonic	Operands	Instruction
BSWAP	dst	Bit Swap
CCF	_	Complement Carry Flag
RCF	_	Reset Carry Flag
SCF	<del></del>	Set Carry Flag
TCM	dst, src	Test Complement Under Mask
TCMX	dst, src	Test Complement Under Mask using Extended Addressing
TM	dst, src	Test Under Mask
TMX	dst, src	Test Under Mask using Extended Addressing

**Table 127. Block Transfer Instructions** 

Mnemonic	Operands	Instruction
LDCI	dst, src	Load Constant to/from Program Memory and Auto- Increment Addresses
LDEI	dst, src	Load External Data to/from Data Memory and Auto- Increment Addresses

**Table 128. CPU Control Instructions** 

Mnemonic	Operands	Instruction
ATM	<del>_</del>	Atomic Execution
CCF	<del></del>	Complement Carry Flag
DI	<del></del>	Disable Interrupts
El	<del></del>	Enable Interrupts
HALT	_	HALT Mode
NOP	_	No Operation
RCF	_	Reset Carry Flag
SCF	_	Set Carry Flag
SRP	src	Set Register Pointer

**Table 128. CPU Control Instructions** 

Mnemonic	Operands	Instruction
STOP	_	STOP Mode
WDT	_	Watchdog Timer Refresh

**Table 129. Load Instructions** 

Mnemonic	Operands	Instruction
CLR	dst	Clear
LD	dst, src	Load
LDC	dst, src	Load Constant to/from Program Memory
LDCI	dst, src	Load Constant to/from Program Memory and Auto-Increment Addresses
LDE	dst, src	Load External Data to/from Data Memory
LDEI	dst, src	Load External Data to/from Data Memory and Auto-Increment Addresses
LDWX	dst, src	Load Word using Extended Addressing
LDX	dst, src	Load using Extended Addressing
LEA	dst, X(src)	Load Effective Address
POP	dst	Рор
POPX	dst	Pop using Extended Addressing
PUSH	src	Push
PUSHX	src	Push using Extended Addressing

**Table 130. Logical Instructions** 

Mnemonic	Operands	Instruction
AND	dst, src	Logical AND
ANDX	dst, src	Logical AND using Extended Addressing
COM	dst	Complement
OR	dst, src	Logical OR
ORX	dst, src	Logical OR using Extended Addressing

**Table 130. Logical Instructions (Continued)** 

Mnemonic	Operands	Instruction
XOR	dst, src	Logical Exclusive OR
XORX	dst, src	Logical Exclusive OR using Extended Addressing

**Table 131. Program Control Instructions** 

Mnemonic	Operands	Instruction
BRK	_	On-Chip Debugger Break
BTJ	p, bit, src, DA	Bit Test and Jump
BTJNZ	bit, src, DA	Bit Test and Jump if Non-Zero
BTJZ	bit, src, DA	Bit Test and Jump if Zero
CALL	dst	Call Procedure
DJNZ	dst, src, RA	Decrement and Jump Non-Zero
IRET	_	Interrupt Return
JP	dst	Jump
JP cc	dst	Jump Conditional
JR	DA	Jump Relative
JR cc	DA	Jump Relative Conditional
RET	_	Return
TRAP	vector	Software Trap

**Table 132. Rotate and Shift Instructions** 

Mnemonic	Operands	Instruction
BSWAP	dst	Bit Swap
RL	dst	Rotate Left
RLC	dst	Rotate Left through Carry
RR	dst	Rotate Right
RRC	dst	Rotate Right through Carry
SRA	dst	Shift Right Arithmetic



Table 132. Rotate and Shift Instructions (Continued)

Mnemonic	Operands	Instruction
SRL	dst	Shift Right Logical
SWAP	dst	Swap Nibbles

### **eZ8 CPU Instruction Summary**

Table 133 summarizes the eZ8 CPU instructions. The table identifies the addressing modes employed by the instruction, the effect upon the Flags register, the number of CPU clock cycles required for the instruction fetch, and the number of CPU clock cycles required for the instruction execution.

Table 133. eZ8 CPU Instruction Summary

Assembly		Address Mode		- Opcode(s)			Fla	ags	_ Fetch	Instr.		
Mnemonic	Symbolic Operation	dst	src	(Hex)	С	Z	s	٧	D	Н		
ADC dst, src	dst ← dst + src + C	r	r	12	*	*	*	*	0	*	2	3
	-	r	lr	13							2	4
	-	R	R	14							3	3
	-	R	IR	15							3	4
	- -	R	IM	16							3	3
	- -	IR	IM	17							3	4
ADCX dst, sro	c dst ← dst + src + C	ER	ER	18	*	*	*	*	0	*	4	3
	-	ER	IM	19							4	3
ADD dst, src	dst ← dst + src	r	r	02	*	*	*	*	0	*	2	3
	-	r	lr	03							2	4
	-	R	R	04							3	3
	-	R	IR	05							3	4
	-	R	IM	06							3	3
	-	IR	IM	07							3	4
ADDX dst, sro	c dst ← dst + src	ER	ER	08	*	*	*	*	0	*	4	3
	·	ER	IM	09							4	3

Table 133. eZ8 CPU Instruction Summary (Continued)

Assembly		Add Mo	ress de	- Opcode(s)			Fla	ıgs	- Fetch	Instr.		
Mnemonic	Symbolic Operation	dst	src	(Hex)	С	Z	s	٧	D	Н	Cycles	
AND dst, src	dst ← dst AND src	r	r	52	-	*	*	0	-	-	2	3
		r	Ir	53							2	4
	-	R	R	54							3	3
	-	R	IR	55							3	4
	-	R	IM	56							3	3
	-	IR	IM	57							3	4
ANDX dst, src	$dst \leftarrow dst \; AND \; src$	ER	ER	58	-	*	*	0	-	-	4	3
	-	ER	IM	59							4	3
ATM	Block all interrupt and DMA requests during execution of the next 3 instructions			2F	-	-	-	-	-	-	1	2
BCLR bit, dst	dst[bit] ← 0	r		E2	-	*	*	0	-	-	2	2
BIT p, bit, dst	dst[bit] ← p	r		E2	-	*	*	0	-	-	2	2
BRK	Debugger Break			00	-	-	-	-	-	-	1	1
BSET bit, dst	dst[bit] ← 1	r		E2	-	*	*	0	-	-	2	2
BSWAP dst	dst[7:0] ← dst[0:7]	R		D5	Χ	*	*	0	-	-	2	2
BTJ p, bit, src,			r	F6	-	-	-	-	-	-	3	3
dst	$PC \leftarrow PC + X$		Ir	F7							3	4
BTJNZ bit,	if src[bit] = 1		r	F6	-	-	-	-	-	-	3	3
src, dst	$PC \leftarrow PC + X$		Ir	F7							3	4
BTJZ bit, src,	if src[bit] = 0		r	F6	-	-	-	-	-	-	3	3
dst	$PC \leftarrow PC + X$		Ir	F7							3	4
CALL dst	SP ← SP -2	IRR		D4	-	-	-	-	-	-	2	6
	@SP ← PC PC ← dst	DA		D6							3	3
CCF	C ← ~C			EF	*	-	-	-	-	-	1	2
CLR dst	dst ← 00H	R		В0	-	-	-	-	-	-	2	2
	-	IR		B1							2	3



Table 133. eZ8 CPU Instruction Summary (Continued)

Assembly		Add Mo	ress de	- Opcode(s)			Fla	ags			- Fetch	Instr.
Mnemonic	Symbolic Operation	dst	src		С	Z	S	٧	D	Н		Cycles
COM dst	dst ← ~dst	R		60	-	*	*	0	-	-	2	2
	·	IR		61	•						2	3
CP dst, src	dst - src	r	r	A2	*	*	*	*	-	-	2	3
	·	r	lr	A3	•						2	4
	•	R	R	A4	•						3	3
	·	R	IR	A5	•						3	4
	·	R	IM	A6	•						3	3
	·	IR	IM	A7	•						3	4
CPC dst, src	dst - src - C	r	r	1F A2	*	*	*	*	-	-	3	3
	·	r	lr	1F A3	•						3	4
	·	R	R	1F A4	•						4	3
	·	R	IR	1F A5	•						4	4
	·	R	IM	1F A6	•						4	3
	·	IR	IM	1F A7	•						4	4
CPCX dst, src	dst - src - C	ER	ER	1F A8	*	*	*	*	-	-	5	3
	·	ER	IM	1F A9	•						5	3
CPX dst, src	dst - src	ER	ER	A8	*	*	*	*	-	-	4	3
	·	ER	IM	A9	•						4	3
DA dst	dst ← DA(dst)	R		40	*	*	*	Χ	-	-	2	2
	·	IR		41	•						2	3
DEC dst	dst ← dst - 1	R		30	-	*	*	*	-	-	2	2
	·	IR		31	•						2	3
DECW dst	dst ← dst - 1	RR		80	-	*	*	*	-	-	2	5
	·	IRR		81	•						2	6
DI	$IRQCTL[7] \leftarrow 0$			8F	-	-	-	-	-	-	1	2
DJNZ dst, RA	$\begin{aligned} & \text{dst} \leftarrow \text{dst} - 1 \\ & \text{if dst} \neq 0 \\ & \text{PC} \leftarrow \text{PC} + X \end{aligned}$	r		0A-FA	-	-	-	-	-	-	2	3

Table 133. eZ8 CPU Instruction Summary (Continued)

Accombly			ress ode	Opcode(s)			Fla	ags			- Fetch	Instr.
Assembly Mnemonic	Symbolic Operation	dst	src	(Hex)		Z	s	٧	D	Н		Cycles
El	IRQCTL[7] ← 1			9F	-	-	-	-	-	-	1	2
HALT	HALT Mode			7F	-	-	-	-	-	-	1	2
INC dst	dst ← dst + 1	R		20	-	*	*	*	-	-	2	2
	•	IR		21	•						2	3
	•	r		0E-FE	•						1	2
INCW dst	dst ← dst + 1	RR		A0	-	*	*	*	-	-	2	5
	•	IRR		A1	•						2	6
IRET	FLAGS $\leftarrow$ @SP SP $\leftarrow$ SP + 1 PC $\leftarrow$ @SP SP $\leftarrow$ SP + 2 IRQCTL[7] $\leftarrow$ 1			BF	*	*	*	*	*	*	1	5
JP dst	PC ← dst	DA		8D	-	-	-	-	-	-	3	2
	•	IRR		C4	•						2	3
JP cc, dst	if cc is true PC ← dst	DA		0D-FD	-	-	-	-	-	-	3	2
JR dst	$PC \leftarrow PC + X$	DA		8B	-	-	-	-	-	-	2	2
JR cc, dst	if cc is true PC ← PC + X	DA		0B-FB	-	-	-	-	-	-	2	2
LD dst, rc	dst ← src	r	IM	0C-FC	-	-	-	-	-	-	2	2
		r	X(r)	C7							3	3
		X(r)	r	D7							3	4
		r	lr	E3							2	3
		R	R	E4							3	2
		R	IR	E5							3	4
	_	R	IM	E6	_						3	2
	•	IR	IM	E7	_						3	3
	•	Ir	r	F3	_						2	3
		IR	R	F5							3	3

Table 133. eZ8 CPU Instruction Summary (Continued)

Assembly			ress ode	- Opcode(s)			Fla	ags			_ Fetch	Instr.
Mnemonic	Symbolic Operation	dst	src	(Hex)		Z	S	٧	D	Н	Cycles	
LDC dst, src	dst ← src	r	Irr	C2	-	-	-	-	-	-	2	5
	•	lr	Irr	C5							2	9
		Irr	r	D2							2	5
LDCI dst, src		lr	Irr	C3	-	-	-	-	-	-	2	9
	$r \leftarrow r + 1$ $rr \leftarrow rr + 1$	Irr	lr	D3							2	9
LDE dst, src	dst ← src	r	Irr	82	-	-	-	-	-	-	2	5
	•	Irr	r	92							2	5
LDEI dst, src	dst ← src	lr	Irr	83	-	-	-	-	-	-	2	9
	$r \leftarrow r + 1$ $rr \leftarrow rr + 1$	Irr	lr	93							2	9
LDWX dst, src	c dst ← src	ER	ER	1F E8	-	-	-	-	-	-	5	4
LDX dst, src	dst ← src	r	ER	84	-	-	-	-	-	-	3	2
		lr	ER	85							3	3
		R	IRR	86							3	4
		IR	IRR	87							3	5
	•	r	X(rr)	88							3	4
	•	X(rr)	r	89							3	4
	•	ER	r	94							3	2
	•	ER	lr	95							3	3
	•	IRR	R	96							3	4
	•	IRR	IR	97							3	5
	•	ER	ER	E8							4	2
	•	ER	IM	E9	•						4	2
LEA dst,	$dst \leftarrow src + X$	r	X(r)	98	-	-	-	-	-	-	3	3
X(src)	•	rr	X(rr)	99							3	5
MULT dst	dst[15:0] ← dst[15:8] * dst[7:0]	RR		F4	-	-	-	-	-	-	2	8
NOP	No operation			0F	-	-	-	-	-	-	1	2



Table 133. eZ8 CPU Instruction Summary (Continued)

Assembly			ress ode	Opcode(s)			Fla	ıgs	_ Fetch	Instr.		
Mnemonic	Symbolic Operation	dst	src	(Hex)		Z	s	٧	D	Н	Cycles	
OR dst, src	dst ← dst OR src	r	r	42	-	*	*	0	-	-	2	3
		r	lr	43							2	4
	<del>-</del>	R	R	44							3	3
		R	IR	45							3	4
		R	IM	46							3	3
	<del>-</del>	IR	IM	47							3	4
ORX dst, src	dst ← dst OR src	ER	ER	48	-	*	*	0	-	-	4	3
	<del>-</del>	ER	IM	49							4	3
POP dst	dst ← @SP	R		50	-	-	-	-	-	-	2	2
	SP ← SP + 1	IR		51							2	3
POPX dst	dst ← @SP SP ← SP + 1	ER		D8	-	-	-	-	-	-	3	2
PUSH src	SP ← SP – 1	R		70	-	-	-	-	-	-	2	2
	@SP ← src	IR		71							2	3
	<del>-</del>	IM		1F 70							3	2
PUSHX src	SP ← SP − 1 @SP ← src	ER		C8	-	-	-	-	-	-	3	2
RCF	C ← 0			CF	0	-	-	-	-	-	1	2
RET	PC ← @SP SP ← SP + 2			AF	-	-	-	-	-	-	1	4
RL dst		R		90	*	*	*	*	-	-	2	2
	D7 D6 D5 D4 D3 D2 D1 D0 dst	IR		91							2	3
RLC dst		R		10	*	*	*	*	-	-	2	2
	—_C	IR		11							2	3
RR dst		R		E0	*	*	*	*	-	-	2	2
	D7 D6 D5 D4 D3 D2 D1 D0 C	IR		E1							2	3

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Table 133. eZ8 CPU Instruction Summary (Continued)

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Assembly			ress ode	- Opcode(s)			Fla	ıgs			- Fetch	Instr.
Mnemonic	Symbolic Operation	dst	src	(Hex)	С	Z	S	٧	D	Н	Cycles	
RRC dst		R		C0	*	*	*	*	-	-	2	2
	D7 D6 D5 D4 D3 D2 D1 D0 → C dst	IR		C1	I						2	3
SBC dst, src	$dst \leftarrow dst - src - C$	r	r	32	*	*	*	*	1	*	2	3
	_	r	lr	33							2	4
	<del>-</del>	R	R	34							3	3
	<del>-</del>	R	IR	35							3	4
	<del>-</del>	R	IM	36							3	3
	<del>-</del>	IR	IM	37							3	4
SBCX dst, src	dst ← dst – src - C	ER	ER	38	*	*	*	*	1	*	4	3
	_	ER	IM	39							4	3
SCF	C ← 1			DF	1	-	-	-	-	-	1	2
SRA dst		R		D0	*	*	*	0	-	-	2	2
	D7 D6 D5 D4 D3 D2 D1 D0  → C dst	IR		D1							2	3
SRL dst	) → D7 D6 D5 D4 D3 D2 D1 D0 → C	R		1F C0	*	*	0	*	-	-	3	2
	dst	IR		1F C1							3	3
SRP src	RP ← src		IM	01	-	-	-	-	-	-	2	2
STOP	STOP Mode			6F	-	-	-	-	-	-	1	2
SUB dst, src	dst ← dst – src	r	r	22	*	*	*	*	1	*	2	3
	_	r	lr	23							2	4
	_	R	R	24							3	3
	_	R	IR	25							3	4
	<del>-</del>	R	IM	26							3	3
	<del>-</del>	IR	IM	27							3	4
SUBX dst, src	dst ← dst – src	ER	ER	28	*	*	*	*	1	*	4	3
	<del>-</del>	ER	IM	29							4	3

eZ8<sup>™</sup> CPU Instruction Set

Table 133. eZ8 CPU Instruction Summary (Continued)

Assembly			dress ode	- Opcode(s)			Fla	ıgs			_ Fetch	Instr.
Mnemonic	Symbolic Operation	dst	src	(Hex)	С	Z	s	٧	D	Н		
SWAP dst	$dst[7:4] \leftrightarrow dst[3:0]$	R		F0	Χ	*	*	Χ	-	-	2	2
	-	IR		F1	•						2	3
TCM dst, src	(NOT dst) AND src	r	r	62	-	*	*	0	-	-	2	3
	-	r	lr	63	•						2	4
	-	R	R	64	•						3	3
	-	R	IR	65	•						3	4
	-	R	IM	66	•						3	3
	-	IR	IM	67	•						3	4
TCMX dst, src	(NOT dst) AND src	ER	ER	68	-	*	*	0	-	-	4	3
	-	ER	IM	69	•						4	3
TM dst, src	dst AND src	r	r	72	-	*	*	0	-	-	2	3
	-	r	lr	73	•						2	4
	-	R	R	74	•						3	3
	-	R	IR	75	•						3	4
	-	R	IM	76	•						3	3
	-	IR	IM	77	•						3	4
TMX dst, src	dst AND src	ER	ER	78	-	*	*	0	-	-	4	3
	-	ER	IM	79	•						4	3
TRAP Vector	$SP \leftarrow SP - 2$ $@SP \leftarrow PC$ $SP \leftarrow SP - 1$ $@SP \leftarrow FLAGS$ $PC \leftarrow @Vector$		Vector	F2	-	-	-	-	-	-	2	6
WDT				5F	-	-	-	-	-	-	1	2



Table 133. eZ8 CPU Instruction Summary (Continued)

Assembly			ress ode	- Opcode(s)			Fla	ags		- Fetch	Instr.	
Mnemonic	Symbolic Operation	dst	src	(Hex)	С	Z	S	٧	D	Н	Cycles	
XOR dst, src	dst ← dst XOR src	r	r	B2	-	*	*	0	-	-	2	3
	-	r	lr	В3	•						2	4
	-	R	R	B4	•						3	3
	-	R	IR	B5	•						3	4
	-	R	IM	B6	•						3	3
	-	IR	IM	В7	•						3	4
XORX dst, srd	c dst ← dst XOR src	ER	ER	B8	-	*	*	0	-	-	4	3
	-	ER	IM	В9							4	3
Flags Notation:	* = Value is a function of - = Unaffected X = Undefined	the resu	ılt of the	operation.		-	Res Set	et to to 1	0			

#### Flags Register

The Flags Register contains the status information regarding the most recent arithmetic, logical, bit manipulation or rotate and shift operation. The Flags Register contains six bits of status information that are set or cleared by CPU operations. Four of the bits (C, V, Z and S) can be tested for use with conditional jump instructions. Two Flags (H and D) cannot be tested and are used for Binary-Coded Decimal (BCD) arithmetic.

The two remaining bits, User Flags (F1 and F2), are available as general-purpose status bits. User Flags are unaffected by arithmetic operations and must be set or cleared by instructions. The User Flags cannot be used with conditional Jumps. They are undefined at initial power-up and are unaffected by Reset. Figure 58 displays the Flags and their bit positions in the Flags Register.

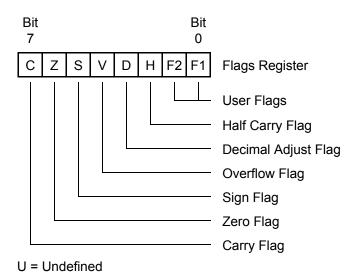


Figure 58. Flags Register

Interrupts, the Software Trap (TRAP) instruction, and Illegal Instruction Traps all write the value of the Flags Register to the stack. Executing an Interrupt Return (IRET) instruction restores the value saved on the stack into the Flags Register.

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## **Opcode Maps**

A description of the opcode map data and the abbreviations are provided in Figure 59 and Table 134 on page 262. Figure 60 on page 263 and Figure 61 on page 264 provide information on each of the eZ8<sup>TM</sup> CPU instructions.

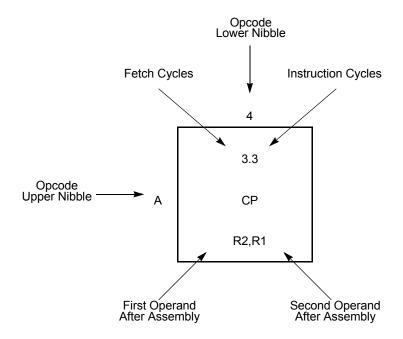


Figure 59. Opcode Map Cell Description



**Table 134. Opcode Map Abbreviations** 

Abbreviation	Description	Abbreviation	Description
b	Bit position	IRR	Indirect Register Pair
СС	Condition code	р	Polarity (0 or 1)
X	8-bit signed index or displacement	r	4-bit Working Register
DA	Destination address	R	8-bit register
ER	Extended Addressing register	r1, R1, Ir1, Irr1, IR1, rr1, RR1, IRR1, ER1	Destination address
IM	Immediate data value	r2, R2, Ir2, Irr2, IR2, rr2, RR2, IRR2, ER2	Source address
Ir	Indirect Working Register	RA	Relative
IR	Indirect register	rr	Working Register Pair
Irr	Indirect Working Register Pair	RR	Register Pair

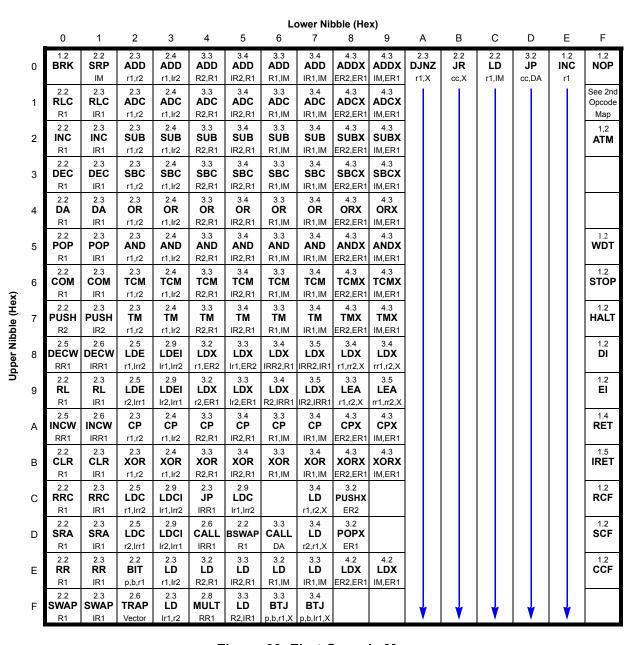


Figure 60. First Opcode Map

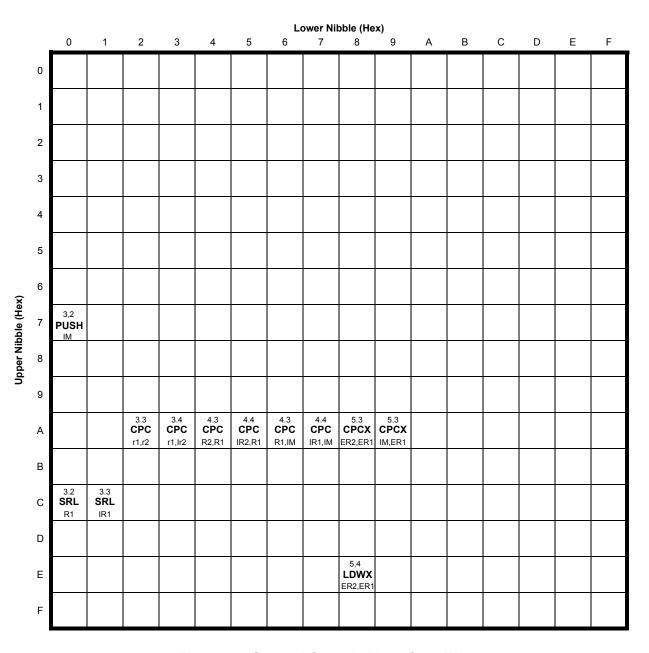


Figure 61. Second Opcode Map after 1FH

## **Packaging**

Figure 62 displays the 40-pin Plastic Dual-inline Package (PDIP) available for the Z8X1601, Z8X2401, Z8X3201, Z8X4801, and Z8X6401 devices.

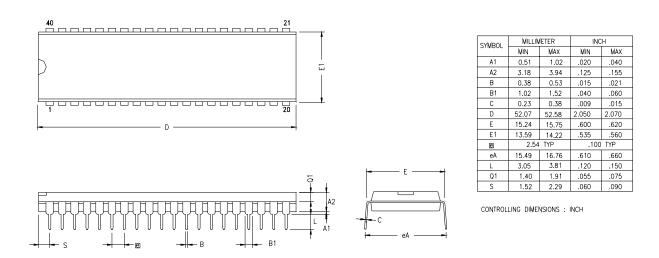


Figure 62. 40-Lead Plastic Dual-Inline Package (PDIP)

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Figure 63 displays the 44-pin Low Profile Quad Flat Package (LQFP) available for the Z8X1621, Z8X2421, Z8X3221, Z8X4821, and Z8X6421 devices.

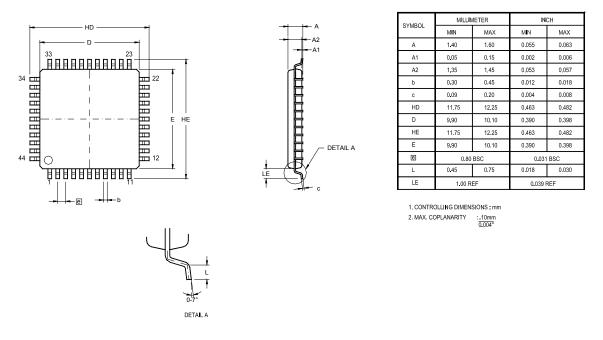
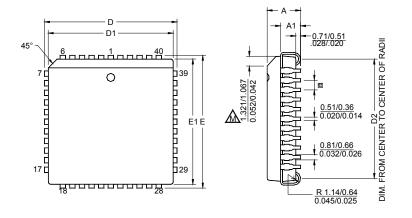


Figure 63. 44-Lead Low-Profile Quad Flat Package (LQFP)

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Figure 64 displays the 44-pin Plastic Lead Chip Carrier (PLCC) package available for the Z8X1621, Z8X2421, Z8X3221, Z8X4821, and Z8X6421 devices.



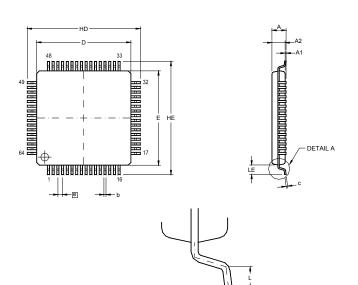
SYMBOL	MILLIN	METER	IN	CH			
OTWIDOL	MIN	MAX	MIN	MAX			
Α	4.27	4.57	0.168	0.180			
A1	2.41	2.92	0.095	0.115			
D/E	17.40	17.65	0.685	0.695			
D1/E1	16.51	16.66	0.650	0.656			
D2	15.24	16.00	0.600	0.630			
8	1.27	BSC	0.050 BSC				

NOTES:

1. CONTROLLING DIMENSION: INCH 2. LEADS ARE COPLANAR WITHIN 0.004". 3. DIMENSION: MM INCH

Figure 64. 44-Lead Plastic Lead Chip Carrier Package (PLCC)

Figure 64 displays the 64-pin Low-Profile Quad Flat Package (LQFP) available for the Z8X1622, Z8X2422, Z8X3222, Z8X4822, and Z8X6422 devices.



DETAIL A

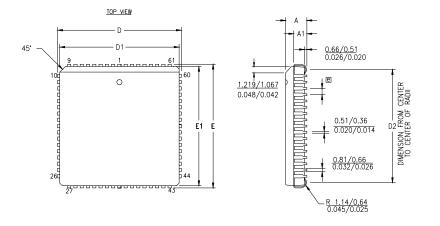
SYMBOL	MILLIN	IETER .	INC	CH		
OTHIDOL	MIN	MAX	MIN	MAX		
Α	1.40	1.60	0.055	0.063		
A1	0.05	0.15	0.002	0.006		
A2	1.35	1.45	0.053	0.057		
b	0.17	0.27	0.007	0.011		
c	0.09	0.20	0.004	0.008		
HD	11.75	12.25	0.463	0.482		
D	9.90	10.10	0.390	0.398		
HE	11.75	12.25	0.463	0.482		
Е	9.90	10.10	0.390	0.398		
е	0.50 B	3C	0.0197	BSC		
L	0.45	0.75	0.018	0.030		
LE	1.00 RI	EF	0.039 REF			

2. MAX. COPLANARITY : 0.10mm 0.004"

Figure 65. 64-Lead Low-Profile Quad Flat Package (LQFP)

PS019919-1207 Packaging

Figure 66 displays the 68-pin Plastic Lead Chip Carrier (PLCC) package available for the Z8X1622, Z8X2422, Z8X3222, Z8X4822, and Z8X6422 devices.



SYMBOL	MILLIM	(ETER	INC	Н
STWIDOL	MIN	MAX	MIN	MAX
A	4.32	4.57	.170	.180
A1	2.43	2.92	.095	.115
D/E	25.02	25.40	.985	1.000
D1/E1	24.13	24.33	.950	.958
D2	22.86	23.62	.900	.930
е	1.27	BSC	.050	BSC

- NOTE:
  1. CONTROLLING DIMENSIONS : INCH.
  2. LEADS ARE COPLANAR WITHIN 0.004 IN. RANGE.
  3. DIMENSION : MM INCH.

Figure 66. 68-Lead Plastic Lead Chip Carrier Package (PLCC)

Packaging PS019919-1207

Figure 67 displays the 80-pin Quad Flat Package (QFP) available for the Z8X4823 and Z8X6423 devices.

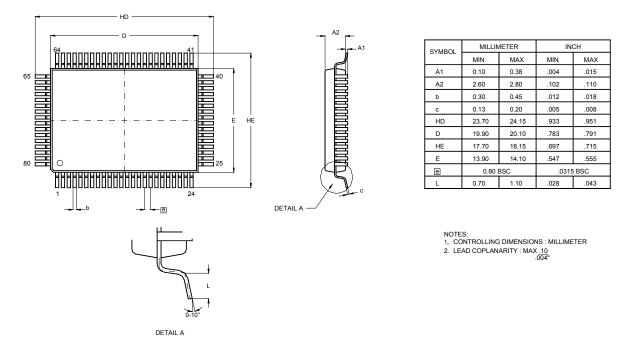


Figure 67. 80-Lead Quad-Flat Package (QFP)

PS019919-1207 Packaging

# **Ordering Information**

Part Number	Flash	RAM	I/O Lines	Interrupts	16-Bit Timers w/PWM	10-Bit A/D Channels	l²c	SPI	UARTS with IrDA	Description
Z8F642x with 64 KB Flash			-to-D	igita	I Co	nver	ter			
Standard Temperature: 0 °C										
Z8F6421PM020SC	64 KB	4 KB	29	23	3	8	1	1	2	PDIP 40-pin package
Z8F6421AN020SC	64 KB	4 KB	31	23	3	8	1	1	2	LQFP 44-pin package
Z8F6421VN020SC	64 KB	4 KB	31	23	3	8	1	1	2	PLCC 44-pin package
Z8F6422AR020SC	64 KB	4 KB	46	24	4	12	1	1	2	LQFP 64-pin package
Z8F6422VS020SC	64 KB	4 KB	46	24	4	12	1	1	2	PLCC 68-pin package
Z8F6423FT020SC	64 KB	4 KB	60	24	4	12	1	1	2	QFP 80-pin package
Extended Temperature: -4	0 °C to +	105 °C								
Z8F6421PM020EC	64 KB	4 KB	29	23	3	8	1	1	2	PDIP 40-pin package
Z8F6421AN020EC	64 KB	4 KB	31	23	3	8	1	1	2	LQFP 44-pin package
Z8F6421VN020EC	64 KB	4 KB	31	23	3	8	1	1	2	PLCC 44-pin package
Z8F6422AR020EC	64 KB	4 KB	46	24	4	12	1	1	2	LQFP 64-pin package
Z8F6422VS020EC	64 KB	4 KB	46	24	4	12	1	1	2	PLCC 68-pin package
Z8F6423FT020EC	64 KB	4 KB	60	24	4	12	1	1	2	QFP 80-pin package
Automotive/Industrial Temp	erature:	–40 °C t	to +1	25 °C	)					
Z8F6421PM020AC	64 KB	4 KB	29	23	3	8	1	1	2	PDIP 40-pin package
Z8F6421AN020AC	64 KB	4 KB	31	23	3	8	1	1	2	LQFP 44-pin package
Z8F6421VN020AC	64 KB	4 KB	31	23	3	8	1	1	2	PLCC 44-pin package
Z8F6422AR020AC	64 KB	4 KB	46	24	4	12	1	1	2	LQFP 64-pin package
Z8F6422VS020AC	64 KB	4 KB	46	24	4	12	1	1	2	PLCC 68-pin package
Z8F6423FT020AC	64 KB	4 KB	60	24	4	12	1	1	2	QFP 80-pin package

Part Number	_		Lines	nterrupts	6-Bit Timers w/PWM	0-Bit A/D Channels			UARTs with IrDA	Description
Part	Flash	RAM	1/0 L	Interi	16-Bi	10-Bi	1 <sub>2</sub> C	SPI	UAR	Desc
Z8F482x with 48 KB Flash	n, 10-Bit	Analog	-to-D	igita	I Co	nver	ter	"		
Standard Temperature: 0 °C	C to 70 °C	)								
Z8F4821PM020SC	48 KB	4 KB	29	23	3	8	1	1	2	PDIP 40-pin package
Z8F4821AN020SC	48 KB	4 KB	31	23	3	8	1	1	2	LQFP 44-pin package
Z8F4821VN020SC	48 KB	4 KB	31	23	3	8	1	1	2	PLCC 44-pin package
Z8F4822AR020SC	48 KB	4 KB	46	24	4	12	1	1	2	LQFP 64-pin package
Z8F4822VS020SC	48 KB	4 KB	46	24	4	12	1	1	2	PLCC 68-pin package
Z8F4823FT020SC	48 KB	4 KB	60	24	4	12	1	1	2	QFP 80-pin package
Extended Temperature: -40	0 °C to +	105 °C								
Z8F4821PM020EC	48 KB	4 KB	29	23	3	8	1	1	2	PDIP 40-pin package
Z8F4821AN020EC	48 KB	4 KB	31	23	3	8	1	1	2	LQFP 44-pin package
Z8F4821VN020EC	48 KB	4 KB	31	23	3	8	1	1	2	PLCC 44-pin package
Z8F4822AR020EC	48 KB	4 KB	46	24	4	12	1	1	2	LQFP 64-pin package
Z8F4822VS020EC	48 KB	4 KB	46	24	4	12	1	1	2	PLCC 68-pin package
Z8F4823FT020EC	48 KB	4 KB	60	24	4	12	1	1	2	QFP 80-pin package
Automotive/Industrial Temp	erature: -	–40 °C t	:o +1	25 °C	)					
Z8F4821PM020AC	48 KB	4 KB	29	23	3	8	1	1	2	PDIP 40-pin package
Z8F4821AN020AC	48 KB	4 KB	31	23	3	8	1	1	2	LQFP 44-pin package
Z8F4821VN020AC	48 KB	4 KB	31	23	3	8	1	1	2	PLCC 44-pin package
Z8F4822AR020AC	48 KB	4 KB	46	24	4	12	1	1	2	LQFP 64-pin package
Z8F4822VS020AC	48 KB	4 KB	46	24	4	12	1	1	2	PLCC 68-pin package
Z8F4823FT020AC	48 KB	4 KB	60	24	4	12	1	1	2	QFP 80-pin package

Part Number	Flash	RAM	I/O Lines	Interrupts	16-Bit Timers w/PWM	10-Bit A/D Channels	l²c	SPI	UARTs with IrDA	Description
Z8F322x with 32 KB Flas	-		-to-D	igita	I Co	nver	ter			
Standard Temperature: 0 °	C to 70 °0	<b>C</b>								
Z8F3221PM020SC	32 KB	2 KB	29	23	3	8	1	1	2	PDIP 40-pin package
Z8F3221AN020SC	32 KB	2 KB	31	23	3	8	1	1	2	LQFP 44-pin package
Z8F3221VN020SC	32 KB	2 KB	31	23	3	8	1	1	2	PLCC 44-pin package
Z8F3222AR020SC	32 KB	2 KB	46	24	4	12	1	1	2	LQFP 64-pin package
Z8F3222VS020SC	32 KB	2 KB	46	24	4	12	1	1	2	PLCC 68-pin package
Extended Temperature: -4	0 °C to 10	05 °C								
Z8F3221PM020EC	32 KB	2 KB	29	23	3	8	1	1	2	PDIP 40-pin package
Z8F3221AN020EC	32 KB	2 KB	31	23	3	8	1	1	2	LQFP 44-pin package
Z8F3221VN020EC	32 KB	2 KB	31	23	3	8	1	1	2	PLCC 44-pin package
Z8F3222AR020EC	32 KB	2 KB	46	24	4	12	1	1	2	LQFP 64-pin package
Z8F3222VS020EC	32 KB	2 KB	46	24	4	12	1	1	2	PLCC 68-pin package
Automotive/Industrial Temp	perature:	–40 °C 1	to 12	5°C						
Z8F3221PM020AC	32 KB	2 KB	29	23	3	8	1	1	2	PDIP 40-pin package
Z8F3221AN020AC	32 KB	2 KB	31	23	3	8	1	1	2	LQFP 44-pin package
Z8F3221VN020AC	32 KB	2 KB	31	23	3	8	1	1	2	PLCC 44-pin package
Z8F3222AR020AC	32 KB	2 KB	46	24	4	12	1	1	2	LQFP 64-pin package
Z8F3222VS020AC	32 KB	2 KB	46	24	4	12	1	1	2	PLCC 68-pin package

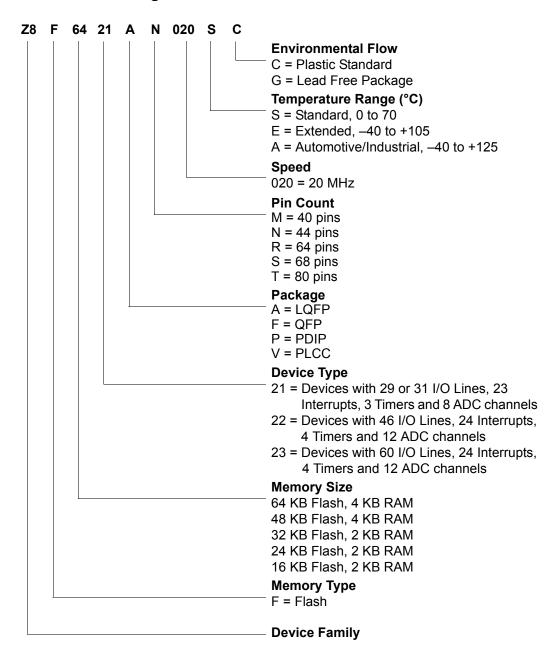
Part Number	ų,	5	Lines	nterrupts	16-Bit Timers w/PWM	0-Bit A/D Channels			JARTs with IrDA	Description
_	Flash	RAM	9		•		l <sub>2</sub> C	SPI	PA	Des
Z8F242x with 24 KB Flash, 10-Bit Analog-to-Digital Converter										
Standard Temperature: 0 °	C to 70 °C	2								
Z8F2421PM020SC	24 KB	2 KB	29	23	3	8	1	1	2	PDIP 40-pin package
Z8F2421AN020SC	24 KB	2 KB	31	23	3	8	1	1	2	LQFP 44-pin package
Z8F2421VN020SC	24 KB	2 KB	31	23	3	8	1	1	2	PLCC 44-pin package
Z8F2422AR020SC	24 KB	2 KB	46	24	4	12	1	1	2	LQFP 64-pin package
Z8F2422VS020SC	24 KB	2 KB	46	24	4	12	1	1	2	PLCC 68-pin package
Extended Temperature: –40 °C to 105 °C										
Z8F2421PM020EC	24 KB	2 KB	29	23	3	8	1	1	2	PDIP 40-pin package
Z8F2421AN020EC	24 KB	2 KB	31	23	3	8	1	1	2	LQFP 44-pin package
Z8F2421VN020EC	24 KB	2 KB	31	23	3	8	1	1	2	PLCC 44-pin package
Z8F2422AR020EC	24 KB	2 KB	46	24	4	12	1	1	2	LQFP 64-pin package
Z8F2422VS020EC	24 KB	2 KB	46	24	4	12	1	1	2	PLCC 68-pin package
Automotive/Industrial Temperature: –40 °C to 125 °C										
Z8F2421PM020AC	24 KB	2 KB	29	23	3	8	1	1	2	PDIP 40-pin package
Z8F2421AN020AC	24 KB	2 KB	31	23	3	8	1	1	2	LQFP 44-pin package
Z8F2421VN020AC	24 KB	2 KB	31	23	3	8	1	1	2	PLCC 44-pin package
Z8F2422AR020AC	24 KB	2 KB	46	24	4	12	1	1	2	LQFP 64-pin package
Z8F2422VS020AC	24 KB	2 KB	46	24	4	12	1	1	2	PLCC 68-pin package

Ī	1	1	1				1	1		1
Part Number	Flash	RAM	I/O Lines	Interrupts	16-Bit Timers w/PWM	10-Bit A/D Channels	1 <sup>2</sup> C	SPI	UARTs with IrDA	Description
Z8F162x with 16 KB Flash, 10-Bit Analog-to-Digital Converter										
Standard Temperature: 0 °C to 70 °C										
Z8F1621PM020SC	16 KB	2 KB	29	23	3	8	1	1	2	PDIP 40-pin package
Z8F1621AN020SC	16 KB	2 KB	31	23	3	8	1	1	2	LQFP 44-pin package
Z8F1621VN020SC	16 KB	2 KB	31	23	3	8	1	1	2	PLCC 44-pin package
Z8F1622AR020SC	16 KB	2 KB	46	24	4	12	1	1	2	LQFP 64-pin package
Z8F1622VS020SC	16 KB	2 KB	46	24	4	12	1	1	2	PLCC 68-pin package
Extended Temperature: –40 °C to +105 °C										
Z8F1621PM020EC	16 KB	2 KB	29	23	3	8	1	1	2	PDIP 40-pin package
Z8F1621AN020EC	16 KB	2 KB	31	23	3	8	1	1	2	LQFP 44-pin package
Z8F1621VN020EC	16 KB	2 KB	31	23	3	8	1	1	2	PLCC 44-pin package
Z8F1622AR020EC	16 KB	2 KB	46	24	4	12	1	1	2	LQFP 64-pin package
Z8F1622VS020EC	16 KB	2 KB	46	24	4	12	1	1	2	PLCC 68-pin package
Automotive/Industrial Temp	erature:	–40 °C 1	to +1	25 °C	)					
Z8F1621PM020AC	16 KB	2 KB	29	23	3	8	1	1	2	PDIP 40-pin package
Z8F1621AN020AC	16 KB	2 KB	31	23	3	8	1	1	2	LQFP 44-pin package
Z8F1621VN020AC	16 KB	2 KB	31	23	3	8	1	1	2	PLCC 44-pin package
Z8F1622AR020AC	16 KB	2 KB	46	24	4	12	1	1	2	LQFP 64-pin package
Z8F1622VS020AC	16 KB	2 KB	46	24	4	12	1	1	2	PLCC 68-pin package
Z8F64200100KITG										Development Kit
ZUSBSC00100ZACG										USB Smart Cable Accessory Kit
ZUSBOPTSC01ZACG										Opto-Isolated USB Smart Cable Accessory Kit
Note: Replace C with G for le	ad-free pa	ckaging.								

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For technical and customer support, hardware and software development tools, refer to the Zilog<sup>®</sup> website at <a href="www.zilog.com">www.zilog.com</a>. The latest released version of ZDS can be downloaded from this website.

#### **Part Number Suffix Designations**



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**Example**: Part number Z8F6421AN020SC is an 8-bit microcontroller product in an LQFP package, using 44 pins, operating with a maximum 20 MHz external clock frequency over a 0 °C to +70 °C temperature range and built using the Plastic-Standard environmental flow.

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