

CS5361

114 dB, 192 kHz, Multi-Bit Audio A/D Converter

Features

- Advanced Multi-bit Delta-sigma Architecture
- 24-bit Conversion
- 114 dB Dynamic Range
- \bullet -105 dB THD+N
- System Sampling Rates up to 192 kHz
- 135 mW Power Consumption
- High-pass Filter and DC Offset Calibration
- Supports Logic Levels Between 5 and 2.5 V
- Differential Analog Architecture
- Overflow Detection
- Pin-compatible with the CS5381

General Description

The CS5361 is a complete analog-to-digital converter for digital audio systems. It performs sampling, analog-todigital conversion, and anti-alias filtering. The CS5361 generates 24-bit values for both left and right inputs in serial form at sample rates up to 192 kHz per channel.

The CS5361 uses a 5th-order, multi-bit, delta-sigma modulator followed by digital filtering and decimation. This removes the need for an external anti-alias filter. The ADC uses a differential architecture which provides excellent noise rejection.

The CS5361 is ideal for audio systems requiring wide dynamic range, negligible distortion, and low noise. These applications include A/V receivers, DVD-R, CD-R, digital mixing consoles, and effects processors.

ORDERING INFORMATION

CS5361K-KSZ -10° to 70°C24-pin SOIC Lead Free CS5361K-KSZR -10° to 70°C24-pin SOIC Lead Free CS5361K-KZZ -10° to 70°C24-pin TSSOP Lead Free CS5361K-KZZR -10° to 70°C24-pin TSSOP Lead Free CS5361K-DZZ -40° to 85°C24-pin TSSOP Lead Free CS5361K-DZZR -40° to 85°C24-pin TSSOP Lead Free CDB5361 Evaluation Board

CIRRUS LOGIC®

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1.0 CHARACTERISTICS AND SPECIFICATIONS

All Min/Max characteristics and specifications are guaranteed over the specified operating conditions. Typical performance characteristics and specifications are derived from measurements taken at typical supply voltages and $T_A = 25$ °C.

SPECIFIED OPERATING CONDITIONS

GND = 0 V, all voltages with respect to GND.

ABSOLUTE MAXIMUM RATINGS

GND = 0 V, All voltages with respect to GND. (Note 1)

Notes: 1. Operation beyond these limits may result in permanent damage to the device. Normal operation is not guaranteed at these extremes.

2. Any pin except supplies. Transient currents of up to ±100 mA on the analog input pins will not cause SCR latch-up.

3. The maximum over/under voltage is limited by the input current.

ANALOG CHARACTERISTICS (CS5361-KSZ/KZZ)

Test conditions (unless otherwise specified): Input test signal is a 1 kHz sine wave; measurement bandwidth is 10 Hz to 20 kHz.

Notes: 4. Referred to the typical full-scale input voltage.

5. Measured between AIN+ and AIN-

ANALOG CHARACTERISTICS (CS5361-DZZ)

Test conditions (unless otherwise specified): Input test signal is a 1 kHz sine wave; measurement bandwidth is 10 Hz to 20 kHz.

DIGITAL FILTER CHARACTERISTICS

Notes: 6. The filter frequency response scales precisely with Fs.

7. Response shown is for Fs equal to 48 kHz. Filter characteristics scale with Fs.

Figure 1. Single Speed Mode Stopband Rejection Figure 2. Single Speed Mode Transition Band

Figure 3. Single Speed Mode Transition Band (Detail) Figure 4. Single Speed Mode Passband Ripple

Figure 5. Double Speed Mode Stopband Rejection **Figure 6. Double Speed Mode Transition Band**

Figure 7. Double Speed Mode Transition Band (Detail) Figure 8. Double Speed Mode Passband Ripple

Figure 11. Quad Speed Mode Transition Band (Detail) Figure 12. Quad Speed Mode Passband Ripple

DC ELECTRICAL CHARACTERISTICS

GND = 0 V, all voltages with respect to ground. MCLK=12.288 MHz; Master Mode.

Notes: 8. Power Down Mode is defined as \overline{RST} = Low with all clocks and data lines held static.

9. Valid with the recommended capacitor values on FILT+ and VQ as shown in the Typical Connection Diagram.

DIGITAL CHARACTERISTICS

THERMAL CHARACTERISTICS

SWITCHING CHARACTERISTICS - SERIAL AUDIO PORT

Logic "0" = GND = 0 V; Logic "1" = VL, C^L = 20 pF

Figure 15. Master Mode, I2S SAI Figure 16. Slave Mode, I2S SAI

Figure 17. OVFL Output Timing

Figure 18. Left Justified Serial Audio Interface

Figure 19. I2S Serial Audio Interface

Figure 20. OVFL Output Timing, I2S Format

Figure 21. OVFL Output Timing, Left-Justified Format

2.0 PIN DESCRIPTIONS

3.0 TYPICAL CONNECTION DIAGRAM

Figure 22. Typical Connection Diagram

4.0 APPLICATIONS

4.1 Operational Mode/Sample Rate Range Select

The output sample rate, Fs, can be adjusted from 2 kHz to 204 kHz. The CS5361 must be set to the proper speed mode via the mode pins, M1 and M0. Refer to Table 1.

Table 1. CS5361 Mode Control

4.2 System Clocking

The device supports operation in either Master Mode, where the left/right and serial clocks are synchronously generated on-chip, or Slave Mode, which requires external generation of the left/right and serial clocks. The device also includes a master clock divider in Master Mode where the master clock will be internally divided prior to any other internal circuitry when MDIV is enabled, set to logic 1. In Slave Mode, the MDIV pin needs to be disabled, set to logic 0_l

4.2.1 Slave Mode

LRCK and SCLK operate as inputs in Slave mode. The left/right clock must be synchronously derived from the master clock and be equal to Fs. It is also recommended that the serial clock be synchronously derived from the master clock and be equal to 64x Fs to maximize system performance. Refer to Table 2 for required clock ratios.

4.2.2 Master Mode

In Master mode, LRCK and SCLK operate as outputs. The left/right and serial clocks are internally derived from the master clock with the left/right clock equal to Fs and the serial clock equal to 64x Fs, as shown in Figure 23. Refer to Table 3 for common master clock frequencies.

Figure 23. CS5361 Master Mode Clocking

Table 3. CS5361 Common Master Clock Frequencies

4.3 Power-up Sequence

Reliable power-up can be accomplished by keeping the device in reset until the power supplies, clocks and configuration pins are stable. It is also recommended that reset be enabled if the analog or digital supplies drop below the minimum specified operating voltages to prevent power glitch related issues.

The internal reference voltage must be stable for the device to produce valid data. Therefore, there is a delay between the release of reset and the generation of valid output, due to the finite output impedance of FILT+ and the presence of the external capacitance.

4.4 Analog Connections

The analog modulator samples the input at 6.144 MHz. The digital filter will reject signals within the stopband of the filter. However, there is no rejection for input signals which are ($n \times 6.144$ MHz) the digital passband frequency, where n=0,1,2,...Refer to Figure 24 which shows the suggested filter that will attenuate any noise energy at 6.144 MHz, in addition to providing the optimum source impedance for the modulators. The use of capacitors which have a large voltage coefficient (such as general purpose ceramics) must be avoided since these can degrade signal linearity.

Figure 24. CS5361 Recommended Analog Input Buffer

4.5 High-pass Filter and DC Offset Calibration

The operational amplifiers in the input circuitry driving the CS5361 may generate a small DC offset into the A/D converter. The CS5361 includes a high-pass filter after the decimator to remove any DC offset which could result in recording a DC level, possibly yielding "clicks" when switching between devices in a multichannel system.

The high-pass filter continuously subtracts a measure of the DC offset from the output of the decimation filter. If the HPF pin is taken high during normal operation, the current value of the DC offset register is frozen and this DC offset will continue to be subtracted from the conversion result. This feature makes it possible to perform a system DC offset calibration by:

- 1) Running the CS5361 with the high-pass filter enabled until the filter settles. See the Digital Filter Characteristics for filter settling time.
- 2) Disabling the high-pass filter and freezing the stored DC offset.

A system calibration performed in this way will eliminate offsets anywhere in the signal path between the calibration point and the CS5361.

4.6 Overflow Detection

The CS5361 includes overflow detection on both the left and right channels. This time multiplexed information is presented as open drain, active low on pin 15, OVFL. The OVFL_L and OVFL_R data will go to a logical low as soon as an overrange condition in either channel is detected. The data will remain low as specified in the Switching Characteristics - Serial Audio Port section. This ensures sufficient time to detect an overrange condition regardless of the speed mode. After the timeout, the OVFL_L and OVFL_R data will return to a logical high if there has not been any other overrange condition detected. Please note that an overrange condition on either channel will restart the timeout period for both channels.

4.6.1 OVFL Output Timing

In left-justified format, the $\overline{\text{OVEL}}$ pin is updated one SCLK period after an LRCK transition. In l^2 S format, the $\overline{\text{OVEL}}$ pin is updated two SCLK periods after an LRCK transition. Refer to Figures 23 and 24. In both cases the OVFL data can be easily demultiplexed by using the LRCK to latch the data. In left-justified format, the rising edge of LRCK would latch the right channel overflow status, and the falling edge of LRCK would latch the left channel overflow status. In I²S format, the falling edge of LRCK would latch the right channel overflow status and the rising edge of LRCK would latch the left channel overflow status.

4.7 Grounding and Power Supply Decoupling

As with any high resolution converter, the CS5361 requires careful attention to power supply and grounding arrangements if its potential performance is to be realized. Figure 22 shows the recommended power arrangements, with VA and VL connected to clean supplies. VD, which powers the digital filter, may be run from the system logic supply or may be powered from the analog supply via a resistor. In this case, no additional devices should be powered from VD. Decoupling capacitors should be as near to the ADC as possible, with the low value ceramic capacitor being the nearest. All signals, especially clocks, should be kept away from the FILT+ and VQ pins in order to avoid unwanted coupling into the modulators. The FILT+ and VQ decoupling capacitors, particularly the 0.01 µF, must be positioned to minimize the electrical path from FILT+ and REFGND. The CDB5361 evaluation board demonstrates the optimum layout and power supply arrangements. To minimize digital noise, connect the ADC digital outputs only to CMOS inputs.

4.8 Synchronization of Multiple Devices

In systems where multiple ADCs are required, care must be taken to achieve simultaneous sampling. To ensure synchronous sampling, the MCLK and LRCK must be the same for all of the CS5361's in the system. If only one master clock source is needed, one solution is to place one CS5361 in Master mode, and slave all of the other CS5361's to the one master. If multiple master clock sources are needed, a possible solution would be to supply all clocks from the same external source and time the CS5361 reset with the inactive edge of MCLK. This will ensure that all converters begin sampling on the same clock edge.

5.0 PARAMETER DEFINITIONS

Dynamic Range

The ratio of the rms value of the signal to the rms sum of all other spectral components over the specified bandwidth. Dynamic Range is a signal-to-noise ratio measurement over the specified bandwidth made with a -60 dBFS signal. 60 dB is added to resulting measurement to refer the measurement to full-scale. This technique ensures that the distortion components are below the noise level and do not affect the measurement. This measurement technique has been accepted by the Audio Engineering Society, AES17-1991, and the Electronic Industries Association of Japan, EIAJ CP-307. Expressed in decibels.

Total Harmonic Distortion + Noise

The ratio of the rms value of the signal to the rms sum of all other spectral components over the specified bandwidth (typically 10 Hz to 20 kHz), including distortion components. Expressed in decibels. Measured at -1 and -20 dBFS as suggested in AES17-1991 Annex A.

Frequency Response

A measure of the amplitude response variation from 10 Hz to 20 kHz relative to the amplitude response at 1 kHz. Units in decibels.

Interchannel Isolation

A measure of crosstalk between the left and right channels. Measured for each channel at the converter's output with no signal to the input under test and a full-scale signal applied to the other channel. Units in decibels.

Interchannel Gain Mismatch

The gain difference between left and right channels. Units in decibels.

Gain Error

The deviation from the nominal full-scale analog output for a full-scale digital input.

Gain Drift

The change in gain value with temperature. Units in ppm/°C.

Offset Error

The deviation of the mid-scale transition (111...111 to 000...000) from the ideal. Units in mV.

6.0 PACKAGE DIMENSIONS

24L SOIC (300 MIL BODY) PACKAGE DRAWING

24L TSSOP (4.4 mm BODY) PACKAGE DRAWING

JEDEC #: MO-153

Controlling Dimension is Millimeters.

- Notes: 1."D" and "E1" are reference datums and do not included mold flash or protrusions, but do include mold mismatch and are measured at the parting line, mold flash or protrusions shall not exceed 0.20 mm per side.
	- 2.Dimension "b" does not include dambar protrusion/intrusion. Allowable dambar protrusion shall be 0.13 mm total in excess of "b" dimension at maximum material condition. Dambar intrusion shall not reduce dimension "b" by more than 0.07 mm at least material condition.
	- 3.These dimensions apply to the flat section of the lead between 0.10 and 0.25 mm from lead tips.

7.0 REVISION HISTORY

Contacting Cirrus Logic Support

For all product questions and inquiries, contact a Cirrus Logic Sales Representative. To find the one nearest you, go to [www.cirrus.com.](http://www.cirrus.com)

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