General Description

The ICS844625I is a high frequency clock generator. The ICS844625I uses an external 25MHz crystal to synthesize 312.5MHz, 156.25MHz and 125MHz clocks. The ICS844625I has excellent cycle-to-cycle and RMS period jitter performance.

The ICS844625I operates at full 3.3V, 2.5V or mixed 3.3V, 2.5V supply modes and is available in a fully RoHS compliant 48-lead TQFP, E-Pad package.

Features

- **•** Ten selectable differential LVDS outputs
- **•** Output frequencies of 312.5MHz, 156.25MHz or 125MHz using a 25MHz crystal.
- **•** Crystal oscillator interface designed for 18pF, 25MHz parallel resonant crystal
- **•** Cycle-to-cycle jitter: 13ps (typical)
- **•** RMS phase jitter at 125MHz (1.875MHz 20MHz): 0.417ps (typical), $V_{DD} = 3.3V$
- **•** RMS phase jitter at 156.25MHz (1.875MHz 20MHz): 0.387ps (typical), $V_{DD} = 3.3V$
- **•** Output duty cycle: 50%, (typical)
- **•** Supply modes: $V_{DD} / V_{DDA} / V_{DDO}$ 3.3V / 3.3V / 3.3V 2.5V / 2.5V / 2.5V 3.3V / 3.3V / 2.5V
- **•** -40°C to 85°C ambient operating temperature
- **•** Available in lead-free (RoHS 6) package

Pin Assignment

Frequency Table for Bank A, B and C Outputs

Block Diagram

Table 1. Pin Descriptions

NOTE: Pullup and Pulldown refer to internal input resistors. See Table 2, Pin Characteristics, for typical values.

Table 2. Pin Characteristics

Function Tables

Table 3A. SELA Function Table

Table 3B. SELB Function Table

Table 3C. SELC Function Table

Absolute Maximum Ratings

NOTE: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the DC Characteristics or AC Characteristics is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

DC Electrical Characteristics

Table 4A. LVDS Power Supply DC Characteristics, $V_{DD} = V_{DDO} = 3.3V \pm 5\%$, $T_A = -40^{\circ}C$ to 85°C

Table 4B. LVDS Power Supply DC Characteristics, $V_{DD} = V_{DDO} = 2.5V \pm 5\%$, $T_A = -40^{\circ}C$ to 85°C

Table 4C. LVDS Power Supply DC Characteristics, $V_{DD} = 3.3V \pm 5\%$, $V_{DDO} = 2.5V \pm 5\%$, $T_A = -40^{\circ}$ C to 85°C

Table 4D. LVCMOS/LVTTL DC Characteristics, $T_A = -40^{\circ}C$ **to 85°C**

Table 4E. LVDS DC Characteristics, $V_{DD} = V_{DDO} = 3.3V \pm 5\%$, $T_A = -40^{\circ}C$ to 85°C

Table 4F. LVDS DC Characteristics, $V_{DD} = V_{DDO} = 2.5V \pm 5\%$, $T_A = -40^{\circ}C$ to 85°C

Table 4G. LVDS DC Characteristics, $V_{DD} = 3.3V \pm 5\%$, $V_{DDO} = 2.5V \pm 5\%$, $T_A = -40\degree$ C to 85 \degree C

Table 5. Crystal Characteristics

AC Electrical Characteristics

Table 6A. LVDS AC Characteristics, $V_{DD} = V_{DDO} = 3.3V \pm 5\%$, $T_A = -40^{\circ}C$ to 85°C

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE 1: This parameter is defined in accordance with JEDEC Standard 65.

Table 6B. LVDS AC Characteristics, $V_{DD} = V_{DDO} = 2.5V \pm 5\%$, $T_A = -40^{\circ}C$ to 85°C

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE 1: This parameter is defined in accordance with JEDEC Standard 65.

Table 6C. LVDS AC Characteristics, $V_{DD} = 3.3V \pm 5\%$, $V_{DDO} = 2.5V \pm 5\%$, $T_A = -40\degree$ C to 85 \degree C

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE 1: This parameter is defined in accordance with JEDEC Standard 65.

Typical Phase Noise at 125MHz, LVDS Output ($V_{DD} = 3.3V$, $V_{DDO} = 3.3V$)

Typical Phase Noise at 156.25MHz, LVDS Output (V_{DD} = 3.3V, V_{DDO} = 3.3V)

Offset Frequency (Hz)

Parameter Measurement Information

3.3V Core/ 3.3V LVDS Output Load AC Test Circuit

3.3V Core/ 2.5V LVDS Output Load AC Test Circuit

Output Duty Cycle/Pulse Width/Period

2.5V Core/ 2.5V LVDS Output Load AC Test Circuit

RMS Phase Jitter

Output Rise/Fall Time

Parameter Measurement Information, continued

Differential Output Voltage Setup Contract Contract Contract Property Offset Voltage Setup

Applications Information

Recommendations for Unused Input and Output Pins

Inputs:

REF_CLK Input

For applications not requiring the use of the reference clock, it can be left floating. Though not required, but for additional protection, a $1k\Omega$ resistor can be tied from the REF_CLK to ground.

Crystal Inputs

For applications not requiring the use of the crystal oscillator input, both XTAL_IN and XTAL_OUT can be left floating. Though not required, but for additional protection, a $1k\Omega$ resistor can be tied from XTAL_IN to ground.

LVCMOS Control Pins

All control pins have internal pullups or pulldowns; additional resistance is not required but can be added for additional protection. A 1k Ω resistor can be used.

Outputs:

LVDS Outputs

All unused LVDS output pairs can be either left floating or terminated with 100 Ω across. If they are left floating, there should be no trace attached.

Overdriving the XTAL Interface

The XTAL IN input can be overdriven by an LVCMOS driver or by one side of a differential driver through an AC coupling capacitor. The XTAL_OUT pin can be left floating. The amplitude of the input signal should be between 500mV and 1.8V and the slew rate should not be less than 0.2V/nS. For 3.3V LVCMOS inputs, the amplitude must be reduced from full swing to at least half the swing in order to prevent signal interference with the power rail and to reduce internal noise. Figure 1A shows an example of the interface diagram for a high speed 3.3V LVCMOS driver. This configuration requires that the sum of the output impedance of the driver (Ro) and the series resistance (Rs) equals the transmission line impedance. In addition, matched termination at the crystal input will attenuate the signal in half. This

can be done in one of two ways. First, R1 and R2 in parallel should equal the transmission line impedance. For most 50Ω applications, R1 and R2 can be 100 Ω . This can also be accomplished by removing R1 and changing R2 to 50 Ω . The values of the resistors can be increased to reduce the loading for a slower and weaker LVCMOS driver. Figure 1B shows an example of the interface diagram for an LVPECL driver. This is a standard LVPECL termination with one side of the driver feeding the XTAL_IN input. It is recommended that all components in the schematics be placed in the layout. Though some components might not be used, they can be utilized for debugging purposes. The datasheet specifications are characterized and guaranteed by using a quartz crystal as the input.

Figure 1A. General Diagram for LVCMOS Driver to XTAL Input Interface

Figure 1B. General Diagram for LVPECL Driver to XTAL Input Interface

LVDS Driver Termination

For a general LVDS interface, the recommended value for the termination impedance (Z_T) is between 90 Ω and 132 Ω . The actual value should be selected to match the differential impedance (Z_0) of your transmission line. A typical point-to-point LVDS design uses a 100 Ω parallel resistor at the receiver and a 100 Ω differential transmission-line environment. In order to avoid any transmission-line reflection issues, the components should be surface mounted and must be placed as close to the receiver as possible. IDT offers a full line of LVDS compliant devices with two types of output structures: current source and voltage source. The

standard termination schematic as shown in Figure 2A can be used with either type of output structure. Figure 2B, which can also be used with both output types, is an optional termination with center tap capacitance to help filter common mode noise. The capacitor value should be approximately 50pF. If using a non-standard termination, it is recommended to contact IDT and confirm if the output structure is current source or voltage source type. In addition, since these outputs are LVDS compatible, the input receiver's amplitude and common-mode input range should be verified for compatibility with the output.

LVDS Termination

EPAD Thermal Release Path

In order to maximize both the removal of heat from the package and the electrical performance, a land pattern must be incorporated on the Printed Circuit Board (PCB) within the footprint of the package corresponding to the exposed metal pad or exposed heat slug on the package, as shown in Figure 3. The solderable area on the PCB, as defined by the solder mask, should be at least the same size/shape as the exposed pad/slug area on the package to maximize the thermal/electrical performance. Sufficient clearance should be designed on the PCB between the outer edges of the land pattern and the inner edges of pad pattern for the leads to avoid any shorts.

While the land pattern on the PCB provides a means of heat transfer and electrical grounding from the package to the board through a solder joint, thermal vias are necessary to effectively conduct from the surface of the PCB to the ground plane(s). The land pattern must be connected to ground through these vias. The vias act as "heat pipes". The number of vias (i.e. "heat pipes") are application specific

and dependent upon the package power dissipation as well as electrical conductivity requirements. Thus, thermal and electrical analysis and/or testing are recommended to determine the minimum number needed. Maximum thermal and electrical performance is achieved when an array of vias is incorporated in the land pattern. It is recommended to use as many vias connected to ground as possible. It is also recommended that the via diameter should be 12 to 13mils (0.30 to 0.33mm) with 1oz copper via barrel plating. This is desirable to avoid any solder wicking inside the via during the soldering process which may result in voids in solder between the exposed pad/slug and the thermal land. Precautions should be taken to eliminate any solder voids between the exposed heat slug and the land pattern. Note: These recommendations are to be used as a guideline only. For further information, refer to the Application Note on the Surface Mount Assembly of Amkor's Thermally/Electrically Enhance Leadframe Base Package, Amkor Technology.

Figure 3. Assembly for Exposed Pad Thermal Release Path - Side View (drawing not to scale)

Schematic Layout

Figure 4 shows an example ICS844625I application schematic in which the device is operated at $V_{DD} = V_{DDA} = 3.3V.3V$ and $V_{DDO} =$ 2.5V. The schematic example focuses on functional connections and is intended as an example only and may not represent the exact user configuration. Refer to the pin description and functional tables in the datasheet to ensure the logic control inputs are properly set. For example MR, BYPASS and the output enables, OE[A:C] can be configured from an FPGA instead of pull up and pull down resistors as shown.

There are two LVDS termination options shown as examples of valid terminations.

- 1) The standard 100 Ω resistor termination of R2.
- 2) An AC termination, used when coupling the ICS844625I LVDS output stage to a different logic family receiver. Always check to make sure LVDS p-p swing is sufficient to drive the receiver to valid logic levels.
	- a) If the receiver is HCSL, then the R6-R7 voltage divider is set at the common mode center voltage of 0.35V.
	- b) If the receiver is 1.5V CML, then $R6 = 0\Omega$ and R7 and C12 are not populated. Typically in this case, the termination resistors, R3 and R4 are integrated into the receiver. In this case only the external coupling caps, C13 and C14 are necessary for the proper termination of the LVDS output.

This device package has an ePAD that is connected to ground internally. The ePAD is to be connected to GND through vias in order to improve heat dissipation.

As with any high speed analog circuitry, the power supply pins are vulnerable to random noise, so to achieve optimum jitter performance isolation of the V_{DD} pin from power supply is required. In order to achieve the best possible filtering, it is recommended that the placement of the filter components be on the device side of the PCB as close to the power pins as possible. If space is limited, all the 0.1uf capacitors associated with the V_{DD} , V_{DDA} and V_{DDO} pins as well as the 10 Ω resistor of the V_{DDA} filter must be placed on the device side with direct return to the ground plane though vias. The remaining filter components can be on the opposite side of the PCB.

Power supply filter component recommendations are a general guideline to be used for reducing external noise from coupling into the devices. The filter performance is designed for a wide range of noise frequencies. This low-pass filter starts to attenuate noise at approximately 10kHz. If a specific frequency noise component is known, such as switching power supplies frequencies, it is recommended that component values be adjusted and if required, additional filtering be added. Additionally, good general design practices for power plane voltage stability suggests adding bulk capacitance in the local area of all devices.

Figure 4. ICS844625I Application Schematic

Power Considerations

This section provides information on power dissipation and junction temperature for the ICS844625I. Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the ICS844625I is the sum of the core power plus the analog power plus the power dissipated in the load(s). The following is the power dissipation for $V_{DD} = 3.3V + 5\% = 3.465V$, which gives worst case results.

- $Power (core)_{MAX} = V_{DD MAX} * (I_{DD MAX} + I_{DDA MAX}) = 3.465V * (125mA + 31mA) = 540.54mW$
- Power (outputs) $_{MAX}$ = V_{DDO_MAX} $*$ I_{DDO_MAX} = 3.465V $*$ 140mA = 485.1mW

Total Power_MAX = 540.54mW + 485.1mW = **1025.64mW**

2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature, Tj, to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for Tj is as follows: Tj = θ_{JA} * Pd_total + T_A

 Tj = Junction Temperature

 θ_{JA} = Junction-to-Ambient Thermal Resistance

Pd_total = Total Device Power Dissipation (example calculation is in section 1 above)

 T_A = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance θ_{l} must be used. Assuming no air flow and a multi-layer board, the appropriate value is 29°C/W per Table 7 below.

Therefore, Tj for an ambient temperature of 85° C with all outputs switching is:

 85° C + 1.026W $*$ 29 $^{\circ}$ C/W = 114.7 $^{\circ}$ C. This is below the limit of 125 $^{\circ}$ C.

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

Table 7. Thermal Resistance θ_{JA} for 48 Lead TQFP, E-Pad Forced Convection

Reliability Information

Table 8. θ_{JA} vs. Air Flow Table for a 48 Lead TQFP, E-Pad

Transistor Count

The transistor count for ICS844625I is: 3,716

Package Outline and Package Dimensions

Package Outline - Y Suffix for 48 Lead TQFP, E-Pad

Table 9. Package Dimensions for 48 Lead TQFP, E-Pad

Reference Document: JEDEC Publication 95, MS-026

Ordering Information

Table 10. Ordering Information

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