MPQ18024



100V, 4A, High-Frequency, Half-Bridge Gate Driver AEC-Q100 Qualified

DESCRIPTION

The MPQ18024 is a high-frequency, 100V, halfbridge, N-channel power MOSFET driver. Its low-side and high-side driver channels are controlled independently and matched with less than 5ns of time delay. Under-voltage lockout (UVLO) on the high-side and low-side supplies force their outputs low in the case of an insufficient supply. The integrated bootstrap diode reduces the external component count.

FEATURES

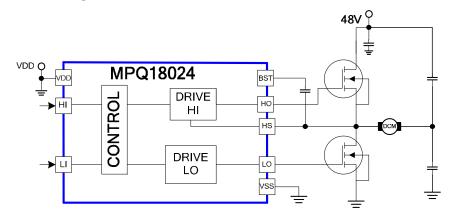
- Guaranteed Industrial / Automotive **Temperature Range Limits**
- Drives an N-Channel MOSFET Half-Bridge
- 100V V_{BST} Voltage Range
- On-Chip Bootstrap Diode
- Typical Propagation Delay of 20ns
- Gate Drive Matching of Less than 5ns
- Drives a 2.2nF Load with 15ns of Rise Time and 12ns of Fall Time at 12V VDD
- TTL-Compatible Input
- Quiescent Current of Less than 160µA
- UVLO for both High-Side and Low-Side
- Available in a SOIC-8E Package
- Available in AEC-Q100 Qualified Grade 1

APPLICATIONS

- Car DC/DC Power Systems
- Half-Bridge Motor Drivers

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TYPICAL APPLICATION





ORDERING INFORMATION

Part Number*	Package	Top Marking
MPQ18024HN-AEC1	SOIC-8 EP	See Below

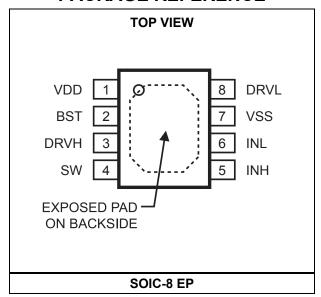
^{*} For Tape & Reel, add suffix -Z (e.g. MPQ18024HN-AEC1-Z).

TOP MARKING

MP18024 LLLLLLL MPSYWW

MP18024: Part number LLLLLLL: Lot number MPS: MPS prefix Y: Year code WW: Week code

PACKAGE REFERENCE





PIN FUNCTIONS

Pin#	Name	Description
1	VDD	Supply input. VDD supplies power to all of the internal circuitries. Place a decoupling capacitor to ground close to VDD to ensure a stable and clean supply.
2	BST	Bootstrap. BST is the positive power supply for the internal floating high-side MOSFET driver. Connect a bypass capacitor between BST and SW.
3	DRVH	Floating driver output.
4	SW	Switching node.
5	INH	Control signal input for the floating driver.
6	INL	Control signal input for the low side driver.
7	VSS	Chira arrayand. Connect the avanced and to VCC for avance the array on a setting
Exposed Pad		Chip ground. Connect the exposed pad to VSS for proper thermal operation.
8 DRVL		Low-side driver output.

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All other pins0.3V to $(V_{DD} + 0.3V)$
CDM rating (AEC-Q100-011C1) All pins
Continuous power dissipation $(T_A = 25^{\circ}C)^{(2)}$
Junction temperature 150°C Lead temperature 260°C Storage temperature -65°C to 150°C
Recommended Operating Conditions (3) Supply voltage (V _{DD})
(-10V / <100ns) to 100V - VDD SW slew rate<50V/ns

Operating junction temp. (T_J).....-40°C to 125°C

Thermal Resistance (4)	$oldsymbol{ heta}_{JA}$	$\boldsymbol{\theta}_{JC}$	
SOIC-8 EP	50	12	.°C/W

NOTES:

- 1) Exceeding these ratings may damage the device.
- 2) The maximum allowable power dissipation is a function of the maximum junction temperature T_J(MAX), the junction-to-ambient thermal resistance θ_{JA}, and the ambient temperature T_A. The maximum allowable continuous power dissipation at any ambient temperature is calculated by P_D(MAX)=(T_J(MAX)-T_A)/θ_{JA}. Exceeding the maximum allowable power dissipation produces an excessive die temperature, causing the regulator to go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
- circuitry protects the device from permanent damage.

 3) The device is not guaranteed to function outside of its operating conditions.
- 4) Measured on JESD51-7, 4-layer PCB.



ELECTRICAL CHARACTERISTICS

 V_{DD} = V_{BST} - V_{SW} = 12V, V_{SS} = V_{SW} = 0V, no load at DRVH and DRVL, T_J = -40°C to +125°C, typical values tested at T_J = +25°C, unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units
Supply Currents						
VDD quiescent current	I _{DDQ}	INL = INH = 0		120	160	μA
VDD operating current	I _{DDO}	fsw = 500kHz		9		mA
Floating driver quiescent current	I _{BSTQ}	INL = INH = 0		70	100	μA
Floating driver operating current	I _{BSTO}	fsw = 500kHz		8.5		mA
Leakage current	I _{LK}	BST = SW = 100V		0.05	2.5	μA
Inputs					I.	
INL/INH high				2.2	2.6	V
INL/INH low			1	1.5		V
INL/INH internal pull-down resistance	Rin			185		kΩ
Under-Voltage Protection (UV	P)				Į.	u.
VDD rising threshold	V _{DDR}		8.1	8.5	8.9	V
VDD hysteresis	V _{DDH}			0.5		V
(BST - SW) rising threshold	V _{BSTR}		6.8	7.4	8	V
(BST - SW) hysteresis	V _{BSTH}			0.55		V
Bootstrap Diode			•		•	
Bootstrap diode VF @ 100µA	V _{F1}			0.5		V
Bootstrap diode VF @ 100mA	V _{F2}			0.95		V
Bootstrap diode dynamic R	R_D	@ 100mA		2.3		Ω
Low-Side Gate Driver			•		•	
Low-level output voltage	Voll	I _O = 100mA		0.08		V
High-level output voltage to rail	Vohl	I _O = -100mA		0.23		V
Dook pull up ourront (5)		$V_{DRVL} = 0V$, $V_{DD} = 12V$		3		Α
Peak pull-up current (5)	IOHL	$V_{DRVL} = 0V$, $V_{DD} = 16V$		4.7		Α
Peak pull-down current (5)	I	$V_{DRVL} = V_{DD} = 12V$		4.5		Α
Peak pull-down current (5)	I _{OLL}	$V_{DRVL} = V_{DD} = 16V$		6		Α
Floating Gate Driver						
Low-level output voltage	V_{OLH}	I _O = 100mA		0.08		V
High-level output voltage to rail	Vонн	I _O = -100mA		0.23		V
Peak pull-up current (5)	la	$V_{DRVH} = 0V$, $V_{DD} = 12V$		2.6		Α
reak pull-up current (4)	Іонн	$V_{DRVH} = 0V$, $V_{DD} = 16V$		4		Α
Peak pull-down current (5)	Іогн	$V_{DRVH} = V_{DD} = 12V$		4.5		Α
reak pull-down current (4)		$V_{DRVH} = V_{DD} = 16V$		5.9		Α
Switching spec – low-side gate driver						
Turn-off propagation delay INL falling to DRVL falling	T_DLFF			20		ns
Turn-on propagation delay INL rising to DRVL rising	T_{DLRR}			20		
DRVL rise time		C _L = 2.2nF		15		ns
DRVL fall time		C _L = 2.2nF		9		ns



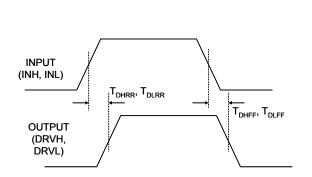
ELECTRICAL CHARACTERISTICS (continued)

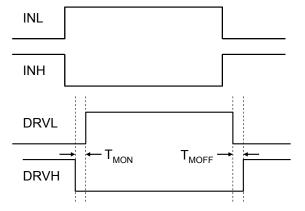
 $V_{DD} = V_{BST} - V_{SW} = 12V$, $V_{SS} = V_{SW} = 0V$, no load at DRVH and DRVL, $T_J = -40^{\circ}C$ to +125°C, typical values tested at $T_J = +25^{\circ}C$, unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units
Switching Spec – Floating Gate Driver						
Turn-off propagation delay INL falling to DRVH falling	T_{DHFF}			20		ns
Turn-on propagation delay INL rising to DRVH rising	T_{DHRR}			20		ns
DRVH rise time		C _L = 2.2nF		15		ns
DRVH fall time		C _L = 2.2nF		12		ns
Switching Spec – Matching						
Floating driver turn-off to low- side drive turn-on (5)	T _{MON}			1	5	ns
Low-side driver turn-off to floating driver turn-on (5)	T _{MOFF}			1	5	ns
Minimum input pulse width that changes the output	T_PW				50 ⁽⁵⁾	ns
Bootstrap diode turn-on or turn-off time	T _{BS}			10 ⁽⁵⁾		ns
Thermal shutdown (5)				170		°C
Thermal shutdown hysteresis				25		°C

NOTE:

TIMING DIAGRAM





⁵⁾ Guaranteed by design.



TYPICAL PERFORMANCE CHARACTERISTICS

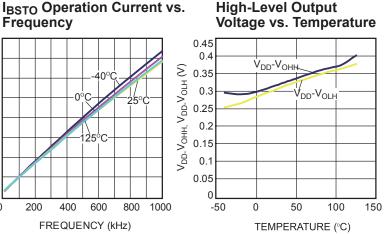
 V_{DD} = 12V, V_{SS} = V_{SW} = 0V, T_A = +25°C, unless otherwise noted.

Frequency 20 18 16 40°C 14 12 10 25°C 6 25°C 0 800 400 600 1000

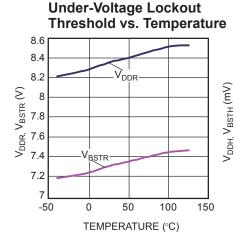
FREQUENCY (kHz)

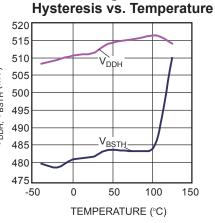
I_{DDO} Operation Current vs.

Frequency 14 12 I_{BSTO} (mA) 10 8 6 4 2 0 400 800 1000 0 200 600 FREQUENCY (kHz)

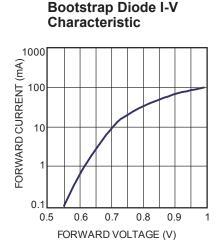


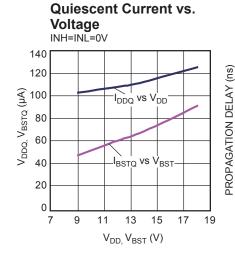
Low-Level Output Voltage vs. Temperature 0.18 0.16 0.14 0.12 0.10 0.08 0.06 0.06 VOLL 0.04 0.02 -50 50 100 150 TEMPERATURE (°C)

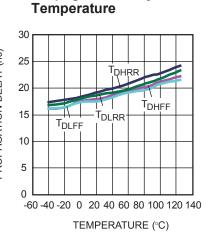




Under-Voltage Lockout







Propagation Delay vs.

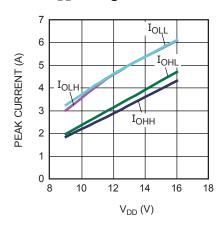
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TYPICAL PERFORMANCE CHARACTERISTICS (continued)

 V_{DD} = 12V, V_{SS} = V_{SW} = 0V, T_A = +25°C, unless otherwise noted.

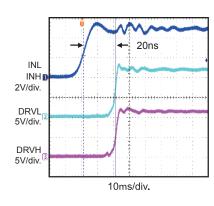
Peak Current vs. **V_{DD}** Voltage

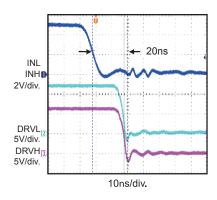


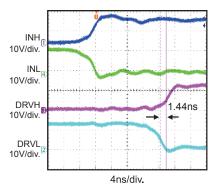
Turn-On Propagation Delay

Turn-Off Propagation Delay

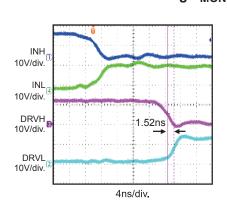
Gate Drive Matching Tmoff



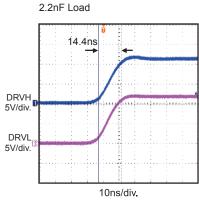




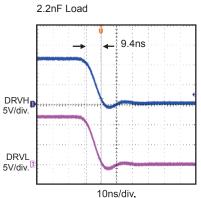
Gate Drive Matching T_{MON}



Drive Rise Time



Drive Fall Time



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BLOCK DIAGRAM

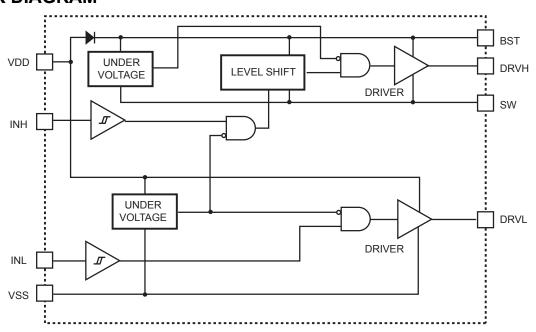


Figure 1: Functional Block Diagram



APPLICATION

The input signals of INH and INL can be controlled independently. If both INH and INL control the high-side MOSFET (HS-FET) and low-side MOSFET (LS-FET) of the same bridge, shoot through can be prevented by setting a sufficient dead time between INH and INL low, and vice versa (see Figure 2). Dead time is defined as the time interval between INH low and INL low.

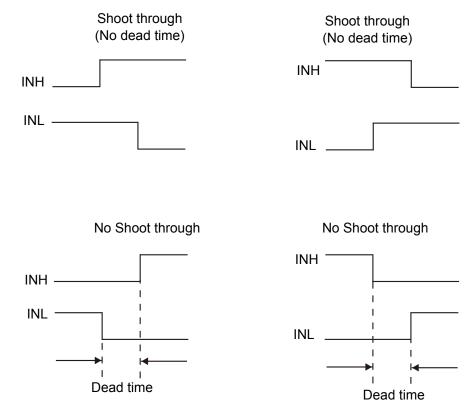


Figure 2: Shoot-Through Timing Diagram



REFERENCE DESIGN CIRCUITS

Half-Bridge Converter

The MPQ18024 drives the MOSFETs with alternating signals (with dead time) in a halfbridge converter topology. Because the pulsewidth modulation (PWM) controller drives INH and INL with alternating signals, the input voltage can rise as high as 100V (see Figure 3 through Figure 5).

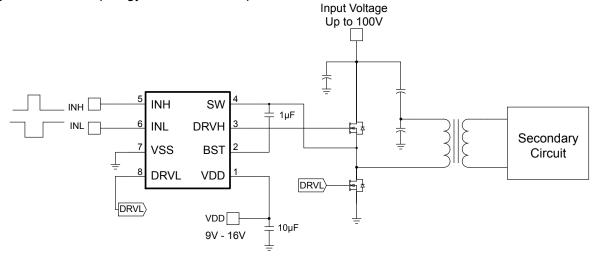


Figure 3: Half-Bridge Converter

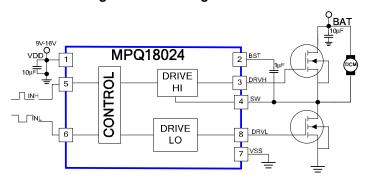


Figure 4: Half-Bridge for Unidirectional Motor

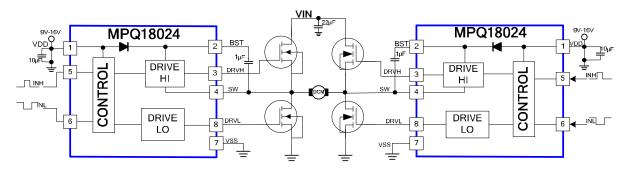
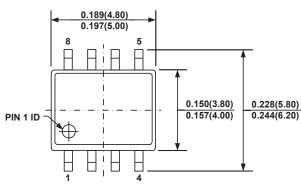


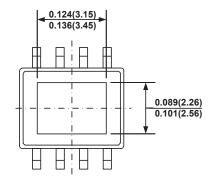
Figure 5: 2x MPQ18024 for One Bidirectional DC Motor



PACKAGE INFORMATION

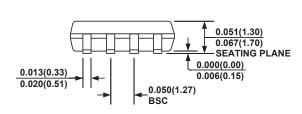
SOIC-8E

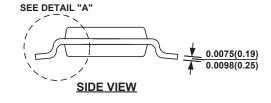




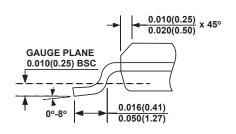
TOP VIEW

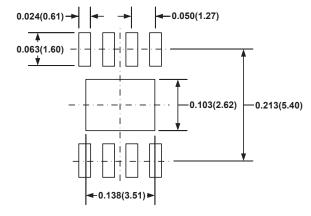
BOTTOM VIEW





FRONT VIEW





DETAIL "A"

NOTE:

- 1) CONTROL DIMENSION IS IN INCHES. DIMENSION IN BRACKET IS IN MILLIMETERS.
- 2) PACKAGE LENGTH DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS.
- 3) PACKAGE WIDTH DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS.
- 4) LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.004" INCHES MAX.
- 5) DRAWING CONFORMS TO JEDEC MS-012, VARIATION BA.
- 6) DRAWING CONFORMS TO SELE.

RECOMMENDED LAND PATTERN

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