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ON Semiconductor®

FDB3632-F085

N-Channel PowerTrench® MOSFET 100V, 80A, $9m\Omega$

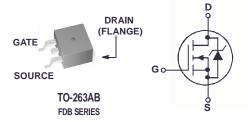
Features

- $r_{DS(ON)} = 7.5 m\Omega$ (Typ.), $V_{GS} = 10V$, $I_D = 80A$
- $Q_a(tot) = 84nC (Typ.), V_{GS} = 10V$
- Low Miller Charge
- Low Q_{RR} Body Diode
- UIS Capability (Single Pulse and Repetitive Pulse)
- Qualified to AEC Q101
- · RoHS Compliant



Applications

- DC/DC converters and Off-Line UPS
- Distributed Power Architectures and VRMs
- Primary Switch for 24V and 48V Systems
- High Voltage Synchronous Rectifier
- Direct Injection / Diesel Injection Systems
- 42V Automotive Load Control
- Electronic Valve Train Systems



MOSFET Maximum Ratings T_C = 25°C unless otherwise noted

Symbol	Parameter	Ratings	Units
V _{DSS}	Drain to Source Voltage	100	V
V_{GS}	Gate to Source Voltage	±20	V
	Drain Current		
I _D	Continuous ($T_C < 111^{\circ}C$, $V_{GS} = 10V$)	80	А
	Continuous ($T_{amb} = 25^{\circ}C$, $V_{GS} = 10V$, $R_{\theta JA} = 43^{\circ}C/W$)	12	А
	Pulsed	Figure 4	А
E _{AS}	Single Pulse Avalanche Energy (Note 1)	338	mJ
	Power dissipation	310	W
P_{D}	Derate above 25°C	2.07	W/°C
T _J , T _{STG}	Operating and Storage Temperature	-55 to +175	°C

Thermal Characteristics

$R_{\theta JC}$	Thermal Resistance Junction to Case TO-220, TO-263, TO-262	0.48	°C/W
	Thermal Resistance Junction to Ambient TO-220, TO-262 (Note 2)	62	°C/W
$R_{\theta JA}$	Thermal Resistance Junction to Ambient TO-263, 1in ² copper pad area	43	°C/W

Package Marking and Ordering Informa	tion
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Device Marking	Device	Package	Reel Size	Tape Width	Quantity
FDB3632	FDB3632-F085	TO-263AB	330mm	24mm	800 units

Electrical Characteristics $T_C = 25^{\circ}C$ unless otherwise noted

Symbol	Parameter	Test Conditions		n Typ	Max	Units
Off Chara	acteristics					
B _{VDSS}	Drain to Source Breakdown Voltage	$I_D = 250 \mu A, V_{GS} = 0 V$	100) -	-	V
I _{DSS}	Zana Cata Valtana Dusia Comunit	V _{DS} = 80V	-	-	1	
	Zero Gate Voltage Drain Current	$V_{GS} = 0V$ $T_{C} = 0$	150°C -	-	250	μΑ
I _{GSS}	Gate to Source Leakage Current	V _{GS} = ±20V	-	-	±100	nA

$V_{GS(TH)}$	Gate to Source Threshold Voltage	$V_{GS} = V_{DS}, I_{D} = 250 \mu A$	2	-	4	V
r _{DS(ON)}	Drain to Source On Resistance	I _D =80A, V _{GS} =10V	-	0.0075	0.009	
	Dialit to Source Off Resistance	I _D =80A, V _{GS} =10V, T _C =175°C	-	0.018	0.022	22

Dynamic Characteristics

C _{ISS}	Input Capacitance	V _{DS} = 25V, V _{GS} = 0V, f = 1MHz		-	6000	-	pF
C _{OSS}	Output Capacitance			-	820	-	pF
C _{RSS}	Reverse Transfer Capacitance			-	200	-	pF
$Q_{g(TOT)}$	Total Gate Charge at 10V	$V_{GS} = 0V \text{ to } 10V$		-	84	110	nC
$Q_{g(TH)}$	Threshold Gate Charge	$V_{GS} = 0V \text{ to } 2V$	$V_{DD} = 50V$	-	11	14	nC
Q _{gs}	Gate to Source Gate Charge		$I_{D} = 80A$	-	30	-	nC
Q _{gs} Q _{gs2}	Gate Charge Threshold to Plateau		$I_g = 1.0 \text{mA}$	-	20	-	nC
Q_{gd}	Gate to Drain "Miller" Charge			-	20	-	nC

Resistive Switching Characteristics $(V_{GS} = 10V)$

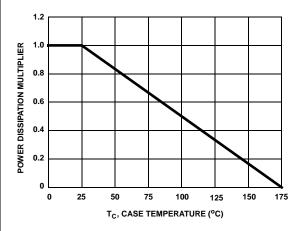
t _{ON}	Turn-On Time		-	-	102	ns
t _{d(ON)}	Turn-On Delay Time	$V_{DD} = 50V, I_{D} = 80A$ $V_{GS} = 10V, R_{GS} = 3.6\Omega$	-	30	-	ns
t _r	Rise Time		-	39	-	ns
t _{d(OFF)}	Turn-Off Delay Time		-	96	-	ns
t _f	Fall Time		-	46	-	ns
t _{OFF}	Turn-Off Time		-	-	213	ns

Drain-Source Diode Characteristics

V _{SD}	Source to Drain Diode Voltage	I _{SD} = 80A	-	-	1.25	V
	Source to Drain Diode Voltage	I _{SD} = 40A	-	-	1.0	V
t _{rr}	Reverse Recovery Time	$I_{SD} = 75A$, $dI_{SD}/dt = 100A/\mu s$	-	-	64	ns
Q_{RR}	Reverse Recovered Charge	$I_{SD} = 75A$, $dI_{SD}/dt = 100A/\mu s$	-	-	120	nC

Notes: 1: Starting $T_J = 25^{\circ}C$, L = 0.12mH, $I_{AS} = 75$ A. 2: Pulse Width = 100s

Typical Characteristics $T_A = 25$ °C unless otherwise noted



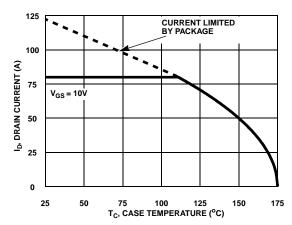


Figure 1. Normalized Power Dissipation vs **Ambient Temperature**

Figure 2. Maximum Continuous Drain Current vs **Case Temperature**

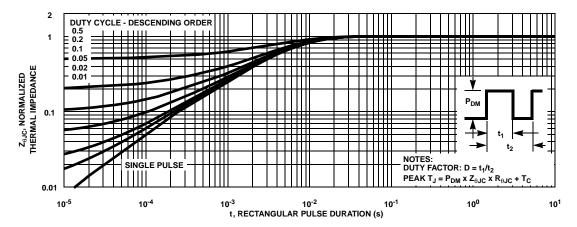


Figure 3. Normalized Maximum Transient Thermal Impedance

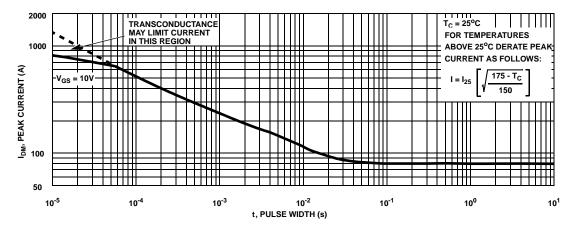
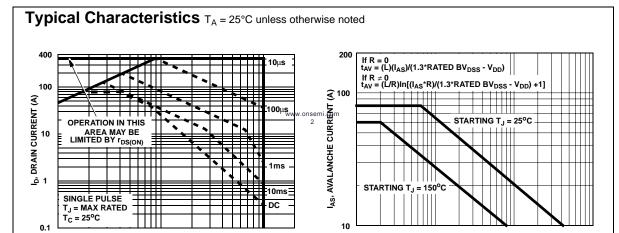


Figure 4. Peak Current Capability



0.01

100

Figure 5. Forward Bias Safe Operating Area

10 V_{DS}, DRAIN TO SOURCE VOLTAGE (V)

NOTE: Refer to ON Semiconductor Application Notes AN7514 and AN7515

Figure 6. Unclamped Inductive Switching Capability

0.1 1 t_{AV}, TIME IN AVALANCHE (ms)

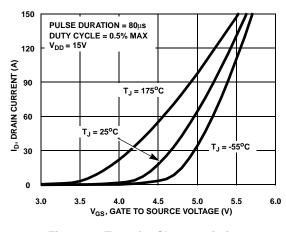


Figure 7. Transfer Characteristics

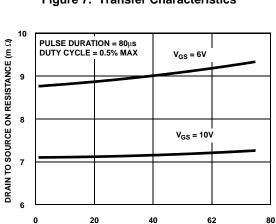


Figure 9. Drain to Source On Resistance vs Drain Current

ID, DRAIN CURRENT (A)

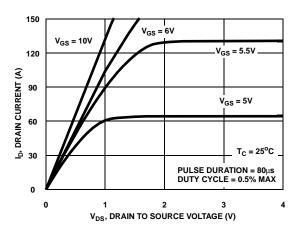


Figure 8. Saturation Characteristics

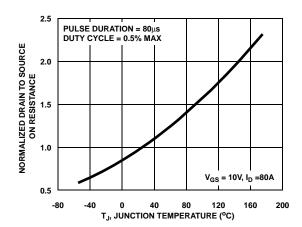


Figure 10. Normalized Drain to Source On Resistance vs Junction Temperature

Typical Characteristics $T_A = 25^{\circ}C$ unless otherwise noted

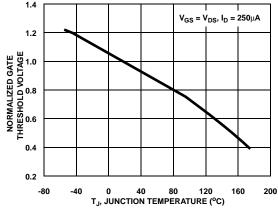
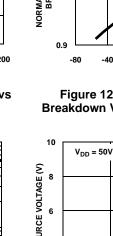


Figure 11. Normalized Gate Threshold Voltage vs **Junction Temperature**

10000

100

0.1



C, CAPACITANCE (pF) 1000

Figure 13. Capacitance vs Drain to Source Voltage

 $\begin{array}{cc} & 1 & 10 \\ V_{DS}, \, \text{DRAIN TO SOURCE VOLTAGE (V)} \end{array}$

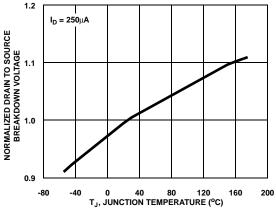


Figure 12. Normalized Drain to Source **Breakdown Voltage vs Junction Temperature**

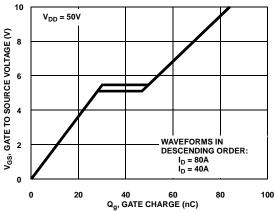
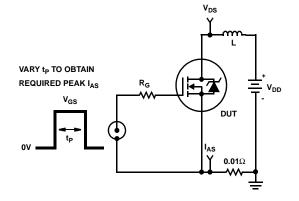


Figure 14. Gate Charge Waveforms for Constant **Gate Currents**

Test Circuits and Waveforms

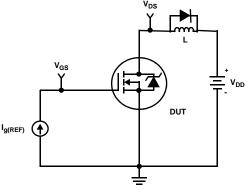


 BV_{DSS}

Figure 15. Unclamped Energy Test Circuit

Figure 16. Unclamped Energy Waveforms

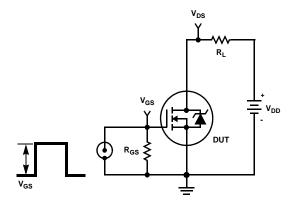
 $\mathbf{Q}_{\mathrm{g(TOT)}}$



I_{g(REF)}

Figure 17. Gate Charge Test Circuit

Figure 18. Gate Charge Waveforms



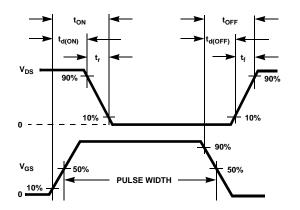


Figure 19. Switching Time Test Circuit

Figure 20. Switching Time Waveforms

Thermal Resistance vs. Mounting Pad Area

The maximum rated junction temperature, T_{JM} , and the thermal resistance of the heat dissipating path determines the maximum allowable device power dissipation, P_{DM} , in an application. Therefore the application's ambient temperature, T_A (°C), and thermal resistance $R_{\theta JA}$ (°C/W) must be reviewed to ensure that T_{JM} is never exceeded. Equation 1 mathematically represents the relationship and serves as the basis for establishing the rating of the part.

$$P_{DM} = \frac{(T_{JM} - T_A)}{R_{\theta JA}} \tag{EQ. 1}$$

In using surface mount devices such as the TO-263 package, the environment in which it is applied will have a significant influence on the part's current and maximum power dissipation ratings. Precise determination of P_{DM} is complex and influenced by many factors:

- Mounting pad area onto which the device is attached and whether there is copper on one side or both sides of the board.
- 2. The number of copper layers and the thickness of the board.
- 3. The use of external heat sinks.
- 4. The use of thermal vias.
- 5. Air flow and board orientation.
- 6. For non steady state applications, the pulse width, the duty cycle and the transient thermal response of the part, the board and the environment they are in.

ON Semiconductor provides thermal information to assist the designer's preliminary application evaluation. Figure 21

defines the $R_{\theta JA}$ for the device as a function of the top copper (component side) area. This is for a horizontally positioned FR-4 board with 1oz copper after 1000 seconds of steady state power with no air flow. This graph provides the necessary information for calculation of the steady state junction temperature or power dissipation. Pulse applications can be evaluated using the ON Semiconductor device Spice thermal model or manually utilizing the normalized maximum transient thermal impedance curve.

Thermal resistances corresponding to other copper areas can be obtained from Figure 21 or by calculation using Equation 2 or 3. Equation 2 is used for copper area defined in inches square and equation 3 is for area in centimeters square. The area, in square inches or square centimeters is the top copper area including the gate and source pads.

$$R_{\theta JA} = 26.51 + \frac{19.84}{(0.262 + Area)}$$
 (EQ. 2)

Area in Inches Squared

$$R_{\Theta JA} = 26.51 + \frac{128}{(1.69 + Area)}$$
 (EQ. 3)

Area in Centimeters Squared

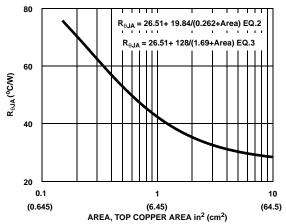
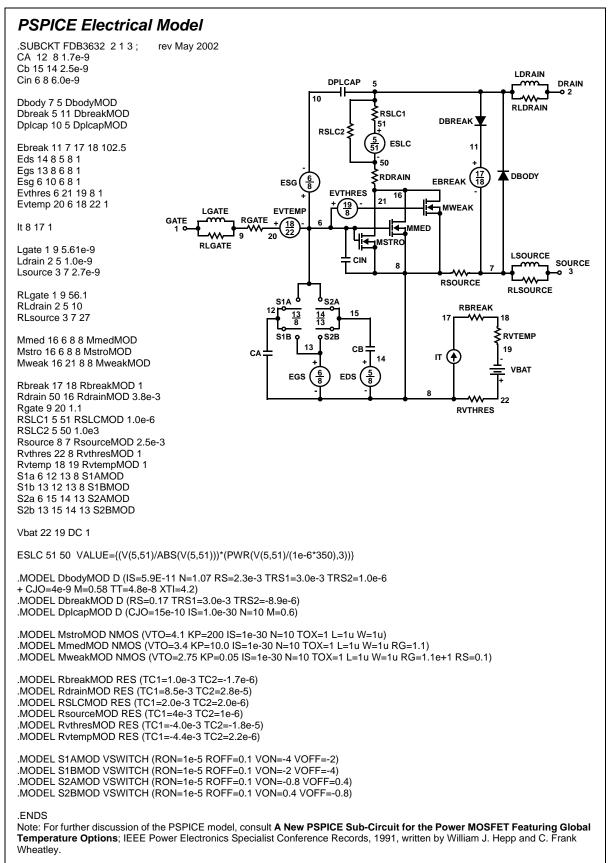
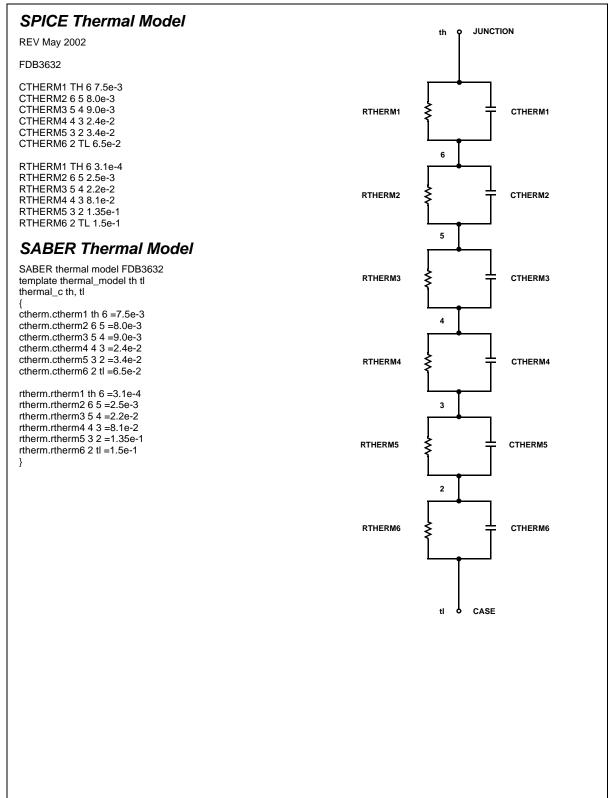


Figure 21. Thermal Resistance vs Mounting
Pad Area



SABER Electrical Model REV May 2002 template FDB3632 n2,n1,n3 electrical n2,n1,n3 var i iscl dp..model dbodymod = (isl=5.9e-11,nl=1.07,rs=2.3e-3,trs1=3.0e-3,trs2=1.0e-6,cjo=4e-9,m=0.58,tt=4.8e-8,xti=4.2) dp..model dbreakmod = (rs=0.17,trs1=3.0e-3,trs2=-8.9e-6) dp..model dplcapmod = (cjo=15e-10,isl=10.0e-30,nl=10,m=0.6) m..model mstrongmod = (type=_n,vto=4.1,kp=200,is=1e-30, tox=1) m..model mmedmod = $(type=_n, vto=3.4, kp=10.0, is=1e-30, tox=1)$ m..model mweakmod = $(type=_n, vto=2.75, kp=0.05, is=1e-30, tox=1, rs=0.1)$ sw_vcsp..model s1amod = (ron=1e-5,roff=0.1,von=-4,voff=-2) I DRAIN sw_vcsp..model s1bmod = (ron=1e-5,roff=0.1,von=-2,voff=-4) **DPLCAP** DRAIN sw_vcsp..model s2amod = (ron=1e-5,roff=0.1,von=-0.8,voff=0.4) 10 sw_vcsp..model s2bmod = (ron=1e-5,roff=0.1,von=0.4,voff=-0.8) RLDRAIN c.ca n12 n8 = 1.7e-9₹RSLC1 c.cb n15 n14 = 2.5e-9 RSLC2 € c.cin n6 n8 = 6.0e-9ISCL dp.dbody n7 n5 = model=dbodymod DBREAK dp.dbreak n5 n11 = model=dbreakmod RDRAIN <u>6</u> 8 dp.dplcap n10 n5 = model=dplcapmod FSG **DBODY EVTHRES** spe.ebreak n11 n7 n17 n18 = 102.5 19 8 **MWEAK LGATE EVTEMP** spe.eds n14 n8 n5 n8 = 1 **RGATE** 18 22 EBREAK ■MMED spe.egs n13 n8 n6 n8 = 1 MSTRO spe.esg n6 n10 n6 n8 = 1 RLGATE spe.evthres n6 n21 n19 n8 = 1 LSOURCE CIN spe.evtemp n20 n6 n18 n22 = 1 SOURCE RSOURCE i.it n8 n17 = 1RLSOURCE I.lgate n1 n9 = 5.61e-9RBREAK 15 I.ldrain n2 n5 = 1.0e-9I.Isource n3 n7 = 2.7e-9**₹**RVTEMP S2B 19 res.rlgate n1 n9 = 56.1 IT (♠ res.rldrain n2 n5 = 10 VBAT res.rlsource n3 n7 = 27 EGS **EDS** m.mmed n16 n6 n8 n8 = model=mmedmod, I=1u, w=1um.mstrong n16 n6 n8 n8 = model=mstrongmod, l=1u, w=1u **RVTHRES** m.mweak n16 n21 n8 n8 = model=mweakmod, l=1u, w=1u res.rbreak n17 n18 = 1, tc1=1.0e-3,tc2=-1.7e-6 res.rdrain n50 n16 = 3.8e-3, tc1=8.5e-3,tc2=2.8e-5 res.rgate n9 n20 = 1.1 res.rslc1 n5 n51 = 1.0e-6, tc1=2.0e-3,tc2=2.0e-6 res.rslc2 n5 n50 = 1.0e3res.rsource n8 n7 = 2.5e-3, tc1=4e-3,tc2=1e-6 res.rvthres n22 n8 = 1, tc1=-4.0e-3,tc2=-1.8e-5 res.rvtemp n18 n19 = 1, tc1=-4.4e-3,tc2=2.2e-6 sw vcsp.s1a n6 n12 n13 n8 = model=s1amod sw_vcsp.s1b n13 n12 n13 n8 = model=s1bmod sw_vcsp.s2a n6 n15 n14 n13 = model=s2amod sw_vcsp.s2b n13 n15 n14 n13 = model=s2bmod v.vbat n22 n19 = dc=1 equations { i (n51->n50) +=iscl iscl: v(n51,n50) = ((v(n5,n51)/(1e-9+abs(v(n5,n51))))*((abs(v(n5,n51)*1e6/350))**3))



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