# MP5496

2.8V to 5.5V, Power Management IC with Four 4.5A/2.5A/4A/2A Buck Converters, 5 LDOs, and Flexible System Settings via I<sup>2</sup>C and OTP

# DESCRIPTION

The MP5496 is a complete power management solution that integrates four high-efficiency, stepdown DC/DC converters, five low-dropout (LDO) regulators, and a flexible logic interface.

A DC/DC converter with constant-on-time (COT) control provides fast transient response. The default 1.5MHz fixed switching frequency during continuous conduction mode (CCM) greatly reduces the external inductor and capacitor values. Full protection features include under-voltage lockout (UVLO), over-current protection (OCP), and thermal shutdown.

The output voltage is adjustable through the I<sup>2</sup>C interface, or can be preset by the one-time programmable (OTP) function. The start-up and shutdown sequences can be configured via the OTP.

By using the I<sup>2</sup>C or OTP, the MP5496 can be utilized to configure the buck and LDO output voltages, mode, buck 1 and buck 3 current limits, and the enable function of all the buck converters and LDOs (ENBUCK/LDO).

When using only the I<sup>2</sup>C interface, the MP5496 allows users to configure the buck 2 and buck 4 current limits, slew rate (DVS slew rate), discharge (DISCHG), system enable (SYSEN), and software reset (SFRST). Status and ID2 registers can also be read via the I<sup>2</sup>C.

Additional features, such as AUTOON, frequency, power-on delay, RST delay, pushbutton time, LDORTC output voltage, OTP version, and the I<sup>2</sup>C slave address can only be configured via the OTP.

The MP5496 requires a minimal number of external components, and is available in a space-saving QFN-28 (4mmx4mm) package.

# FEATURES

## Four High-Efficiency Step-Down Converters

- Buck 1: 4.5A DC/DC Converter
- Buck 2: 2.5A DC/DC Converter
- Buck 3: 4A DC/DC Converter
- Buck 4: 2A DC/DC Converter
- 0.6V to 2.1875V/12.5mV Step VOUT Range
- 2.8V to 5.5V Operating Input Range
- Adjustable Switching Frequency
- Configurable Forced PWM, Auto-PFM/PWM Mode
- Hiccup Over-Current Protection (OCP)

#### **Five Low-Dropout Regulators**

- One RTC-Dedicated LDO
- Four Low-Noise LDOs
- Two Separate Input Power Supplies
- 100mV Dropout at 300mA Load

#### System

- I<sup>2</sup>C Bus and OTP
- Power-On/Off Button
- Power-On Reset Output
- Flexible Power-On/Off Sequencing via OTP
- Flexible DC/DC, LDO On/Off via OTP

## **APPLICATIONS**

- Cable Modems, Set-Top Boxes
- Televisions
- Wi-Fi Routers/Access Points
- PoS Machines
- SSDs
- IP Cameras

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# **TYPICAL APPLICATION**

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## OTP E-Fuse Selected Table by Default (MP5496GR-0001)

OTP Items	Buck 1	Buck 2	Buck 3	Buck 4	LDORTC	LDO2	LDO3	LDO4	LDO5
Output voltage	1.0V	0.9V	0.9V	1.8V	1.2V	3V	1.8V	1.2V	0.9V
Initial on/off	On	On	On	On	On	On	Off	On	On
Mode	FPWM	FPWM	FPWM	FPWM			N/A		
Start-up delay/time slot #	0ms/0	4ms/2	2ms/1	6ms/3	Always on	12ms/6	2ms/1	8ms/4	10ms/5
Automatic start-up					Yes				
Switching frequency					1.5MHz				
Push-button timer					2 seconds	S			
RSTO delay					100ms				
Buck 1 peak current limit					6.8A				
Buck 3 peak current limit		5.6A							
I <sup>2</sup> C slave address					0x69				
OTP version					0004				

#### Other Parameter Information for the MP5496GR-0001

	Buck 1	Buck 2	Buck 3	Buck 4	LDORTC	LDO2	LDO3	LDO4	LDO5
Minimum input voltage					3V				
Typical input voltage					3.3V				
Maximum input voltage					3.6V				



# **ORDERING INFORMATION**

Part Number*	Package	Top Marking	MSL Rating
MP5496GR-xxxx**	QFN-28 (4mmx4mm)	See Below	
MP5496GR-0001	QFN-28 (4mmx4mm)	See Below	1
EVKT-MP5496	Evaluation kit	N/A	

\* For Tape & Reel, add suffix –Z (e.g. MP5496GR-XXXX–Z).

\*\* "xxxx" is the configuration code identifier for the register setting stored in the OTP. The default number is "0001". Each "x" can be a hexadecimal value between 0 and F. Work with an MPS FAE to create this unique number, even if ordering the "0001" code. MP5496GR-0001 is the default version.

# **TOP MARKING**

MPSYWW MP5496 LLLLLL

MPS: MPS prefix Y: Year code WW: Week code MP5496: Part number LLLLLL: Lot number



# **EVKT-MP5496 EVALUATION KIT**

EVKT-MP5496 kit contents (items below can be ordered separately):

#	Part Number	Item	Quantity
1	EV5496-R-00B	MP5496GR-CCCC evaluation board	1
2	EVKT-USBI2C-02-BAG	Includes USB to I <sup>2</sup> C Communication Interface device, one USB cable, and one ribbon cable	1
3	MP5496GR-CCCC	IC with default configurations	2

#### Order directly from MonolithicPower.com or our distributors.



Figure 1: EVKT-MP5496 Evaluation Kit Set-Up

# PACKAGE REFERENCE





# **PIN FUNCTIONS**

Pin #	Name	Description
1	VIN1	<b>Buck 1 supply voltage input.</b> The MP5496 operates from a 2.8V to 5.5V input rail. A ceramic capacitor is required to decouple the input rail. Connect VIN1 using a wide PCB trace. VIN1, VIN2, VIN3, VIN4, and AVIN must be connected to the same bus voltage.
2	SW1	Buck 1 switch output. Connect SW1 using a wide PCB trace.
3	GND1	<b>Buck 1 power ground.</b> GND1 requires special consideration when designing the PCB layout. Connect GND1 to GND with copper traces and vias.
4	GND3	<b>Buck 3 power ground.</b> GND3 requires special consideration when designing the PCB layout. Connect GND3 to GND with copper traces and vias.
5	SW3	Buck 3 switch output. Connect SW3 using a wide PCB trace.
6	VIN3	<b>Buck 3 supply voltage input.</b> The MP5496 operates from a 2.8V to 5.5V input rail. A ceramic capacitor is required to decouple the input rail. Connect VIN3 using a wide PCB trace.
7	FB3	Buck 3 feedback. Connect the output of buck 3 directly to FB3.
8	RSTO	<b>PMIC to CPU output reset.</b> Once the buck 4 output is ready, the RSTO goes high after a delay. RSTO is an open-drain output that requires an external pull-up resistor.
9	SCL	I <sup>2</sup> C clock signal input. Use an external resistor to pull SCL up to AVIN if I <sup>2</sup> C functionality is not used.
10	SDA	I <sup>2</sup> C data. Use an external resistor to pull SDA up to AVIN if I <sup>2</sup> C functionality is not used.
11	OUT4	LDO4 output. LDO4 is powered by VIN5.
12	OUT5	LDO5 output. LDO5 is powered by VIN5.
13	VIN5	LDO4 and LDO5 power input.
14	FB4	Buck 4 feedback. Connect the output of buck 4 directly to FB4.
15	VIN4	<b>Buck 4 supply voltage input.</b> The MP5496 operates from a 2.8V to 5.5V input rail. A ceramic capacitor is required to decouple the input rail. Connect VIN4 using a wide PCB trace.
16	SW4	Buck 4 switch output. Connect SW4 using a wide PCB trace.
17	GND4	<b>Buck 4 power ground.</b> GND4 requires special consideration when designing the PCB layout. Connect GND4 to GND with copper traces and vias.
18	GND2	<b>Buck 2 power ground.</b> GND2 requires special consideration when designing the PCB layout. Connect GND2 to GND with copper traces and vias.
19	SW2	Buck 2 switch output. Connect SW2 using a wide PCB trace.
20	VIN2	<b>Supply voltage input for buck 2, LDORTC, LDO2, and LDO3.</b> The MP5496 operates from a 2.8V to 5.5V input rail. A ceramic capacitor is required to decouple the input rail. Connect VIN2 using a wide PCB trace.
21	FB2	Buck 2 feedback. Connect the output of buck 2 directly to FB2.
22	OUT3	LDO 3 output. LDO 3 is powered by VIN2.
23	OUT2	LDO 2 output. LDO 2 is powered by VIN2.
24	OUTRTC	<b>RTC LDO output.</b> This LDO is powered by VIN2. Set a sufficient output voltage to achieve a lower voltage gap between VIN2 and OUTRTC.
25	AGND	Analog ground. Connect AGND to power ground.
26	AVIN	<b>Power supply input for logic circuitry.</b> Bypass AVIN with a $0.1\mu$ F to $1\mu$ F ceramic capacitor to AGND. Connect AVIN to the system input.
27	FB1	Buck 1 feedback. Connect the output of buck 1 directly to FB1.
28	nPBIN	<b>Push-button input.</b> nPBIN is a logic input pin that can start up or shut down the device. A logic low over a preset deglitch time must be applied to nPBIN. nPBIN has a weak internal pull-up current.



## **ABSOLUTE MAXIMUM RATINGS** (1)

VIN1, VIN2, VIN3, VIN4, VIN5, AV	VIN
	0.3V to 6.25V
V <sub>SWx</sub> 0.6V (-5	V for <10ns) to
V <sub>IN</sub> + 0.3V	(7V for <10ns)
All other pins	0.3V to +6.25V
Continuous power dissipation (TA	= 25°C) <sup>(2)</sup>
	2.84W
Junction temperature	150°C
Lead temperature	260°C
Storage temperature6	5°C to +150°C

## ESD Ratings

Human body model (HBM)	±2000V
Charged device model (CDM)	±750V

#### **Recommended Operating Conditions** <sup>(3)</sup>

Step-down regulator (V <sub>IN</sub> )	2.8V to 5.5V
Step-down regulator (V <sub>OUT</sub> )	0.6V to 2.18V
LDO regulator (VOUTL)	0.8V to 3.9V
Operating junction temp (T <sub>J</sub> )	-40°C to +125°C

#### Thermal Resistance θ<sub>JA</sub> θ<sub>JC</sub>

### QFN-28 (4mmx4mm)

JESD51-7<sup>(4)</sup> ...... 44...... 9.... °C/W

#### Notes:

- 1) Exceeding these ratings may damage the device.
- 2) The maximum allowable power dissipation is a function of the maximum junction temperature, T<sub>J</sub> (MAX), the junction-toambient thermal resistance,  $\theta_{JA}$ , and the ambient temperature, T<sub>A</sub>. The maximum allowable continuous power dissipation at any ambient temperature is calculated by P<sub>D</sub> (MAX) = (T<sub>J</sub> (MAX) - T<sub>A</sub>) /  $\theta_{JA}$ . Exceeding the maximum allowable power dissipation may produce an excessive die temperature, which can cause the regulator to go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
- 3) The device is not guaranteed to function outside of its operating conditions.
- 4) Measured on JESD51-7, 4-layer PCB. The value of  $\theta_{JA}$  given in this table is only valid for comparison with other packages and cannot be used for design purposes. These values were calculated in accordance with JESD51-7, and simulated on a specified JEDEC board. They do not represent the performance obtained in an actual application.



# **ELECTRICAL CHARACTERISTICS**

#### VIN1 = VIN2 = VIN3 = VIN4 = VIN5 = AVIN = 5V, $T_J$ = -40°C to 125°C <sup>(5)</sup>, unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units
Supply current (no switching)	lin	No switching, feedback is high, $T_J = 25^{\circ}C$		420	500	μA
Default oscillation frequency	fsw		1.2	1.5	1.8	MHz
Thermal shutdown entry threshold <sup>(6)</sup>	Totp_r		145	153	162	°C
Thermal shutdown recovery threshold <sup>(6)</sup>	T <sub>HYS</sub>		121	130	139	°C
Step-Down Regulator						
AVIN UVLO rising	$V_{\text{AIN1}_{\text{R}}}$		2.4	2.55	2.7	V
AVIN UVLO hysteresis	VAIN1_HYS			300		mV
VIN1 UVLO rising	$V_{IN1_R}$			2.45		V
VIN1 UVLO hysteresis	VIN1_HYS			300		mV
VIN2 UVLO rising (7)	V <sub>IN2_R</sub>			2.45		V
VIN2 UVLO hysteresis (7)	VIN2_HYS			300		mV
VIN3 UVLO rising	VIN3_R			2.45		V
VIN3 UVLO hysteresis	VIN3_HYS			300		mV
VIN5 UVLO rising	V <sub>IN5_R</sub>			2.45		V
VIN5 UVLO hysteresis	V <sub>IN5_HYS</sub>			300		mV
	$V_{FB1}$	Default output of buck 1	0.985	1	1.015	V
Feedback voltage	$V_{FB2}$	Default output of buck 2	0.8865	0.9	0.9135	V
accuracy	$V_{FB3}$	Default output of buck 3	0.8865	0.9	0.9135	V
Feedback voltage accuracy	$V_{FB4}$	Default output of buck 4	1.7730	1.8	1.8270	V
Maximum duty cycle	D <sub>MAX</sub>	CH2 and CH4 only		100		%
Buck 1, Buck 3 (4.5A/4A	)	1	11		1	
	HS <sub>RDS-ON1</sub>	500 A T 0500	10	00	50	
	HS <sub>RDS-ON3</sub>	$500$ mA, $1_{\rm J} = 25^{\circ}$ C	10	30	50	mΩ
HS switch on resistance	HS <sub>RDS-ON1</sub>	500mA T 40%C to 105%C	10	20	5       1.8       N         3       162       0         0       139       0         0       139       0         0       139       0         0       139       0         0       139       0         0       139       0         0       139       0         0       0       0         0       0       0         0       0       0         1       0.9135       0         0       0.9135       0         0       0       0         0       50       0         0       50       0         0       50       0         1       1       1         8       8.3       0         6       7       0	
	HS <sub>RDS-ON3</sub>	$-500 \text{ mA}, 1_{\text{J}} = -40^{\circ} \text{C} \ 10 + 125^{\circ} \text{C}$	10	30	50	mt2
	LSRDS-ON1			10		
LC quitab an registeres	LS <sub>RDS-ON3</sub>	$500$ mA, $1_{\rm J} = 25^{\circ}$ C		12	20	mΩ
LS switch on resistance	LS <sub>RDS-ON1</sub>	500mA T 40%C to 105%C		10		
	LS <sub>RDS-ON3</sub>	$-500 \text{ mA}, 1 \text{ J} = -40^{\circ} \text{ C} 10 + 125^{\circ} \text{ C}$		12	20	mt2
Quitab lookaga 1	HSWILK1	EN = 0V, VIN = 5.5V, T <sub>J</sub> = 25°C,		0	-	
Switch leakage I	HSWILK3	SW = 0V or 5.5V		0	I	μΑ
Switch lookage 2	LSWILK1	EN = 0V, VIN = 5.5V, T <sub>J</sub> = 25°C,		0	4	
Switch leakage 2	<b>LSW</b> ILK3	SW = 0V or 5.5V		U	I	μΑ
High-side current limit		Under 20% duty cycle $T_{1} = 25^{\circ}$	5.5	6.8	8.3	Α
	I <sub>LIMIT3</sub>	$I_{\text{LIMIT3}}$ Under 20% duty cycle, 1J = 25°C		5.6	7	А



# ELECTRICAL CHARACTERISTICS (continued)

## VIN1 = VIN2 = VIN3 = VIN4 = VIN5 = AVIN = 5V, $T_J$ = -40°C to +125°C <sup>(5)</sup>, unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units
(0)	ton min1			40		ns
Minimum on time (°)	ton_мімз			33		ns
<b>NA</b> : 1 ( <b>0</b> )	toff_min1			120		ns
Minimum off time (°)	toff_min3			120		ns
Output discharge resistance	Ro_DIS1			7		Ω
	t <sub>SS B1</sub>	V <sub>OUT</sub> = 10% to 90%		450		μs
Soft-start time	tss вз	V <sub>OUT</sub> = 10% to 90%		450	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	μs
Buck 2, Buck 4 (2.5A/2A	.)					
	HSRDS-ON2	500 A T 0500				
	HSRDS-ON4	$-500$ mA, $I_{J} = 25$ °C	20	50	80	mΩ
HS switch on resistance	HS <sub>RDS-ON2</sub>	500mA T 4000 to 10500	00	50	Max Max Max Max 1 1 1 1 1 1 1 1 1 1 1 1 1	
	HSRDS-ON4	$-500$ mA, $1_{\rm J} = -40^{\circ}$ C to $125^{\circ}$ C	20	50		mΩ
	LS <sub>RDS-ON2</sub>	500mA T 0500		05	rp       Max $r         0$	
L C awitch an registeres	LS <sub>RDS-ON4</sub>	$-500$ mA, $T_{\rm J} = 25^{\circ}$ C		65		mΩ
LS switch on resistance	LSRDS-ON2	500mA T 40%C to 105%C		<u>CE</u>		
	LSRDS-ON4	$-500 \text{ mA}, 1 \text{ J} = -40^{\circ} \text{ C} \text{ to } 125^{\circ} \text{ C}$		65		mc2
Switch lookago 2	HSW <sub>ILK2</sub>	Shutdown, $V_{IN} = 5.5V$ ,		0	1	
Switch leakage S	HSWILK4	SW = 0V or 5.5V, $T_A = 25^{\circ}C$		0 1	μΛ	
Switch lookage 4	LSWILK2	Shutdown, $V_{IN} = 5.5V$ ,		0	5 105 1 1 2 5.4	
Switch leakage 4	LSWILK4	SW = 0V or 5.5V, $T_A = 25^{\circ}C$		0		μΑ
High-side current limit	I <sub>LIMIT2</sub>	$-$ Under 20% duty cycle $T_{1} = 25^{\circ}C$	3	12	Max         Imax         Imax <tr< td=""><td>Δ</td></tr<>	Δ
	ILIMIT4		5	7.2		~
Minimum on time <sup>(8)</sup>	ton_min2			32		ns
	ton_min4			32		ns
Minimum off time <sup>(8)</sup>	toff_min2			100		ns
	toff_min4			100		ns
Output discharge resistance	Ro_DIS2			7		Ω
	tss_b2	V <sub>OUT</sub> = 10 to 90%		450		μs
Soft-start time	t <sub>SS_B4</sub>	V <sub>OUT</sub> = 10 to 90%		450		μs
10mA RTC LDO		•				-
Default output voltage	$V_{\text{RTC}\_\text{LDO}}$	I <sub>OUT</sub> = 10mA, power-on state	1.164	1.2	1.236	V
Ground current	IQ_RTC	No load		6.5		μA
Dropout voltage (8)	V <sub>DROP1</sub>	$V_{OUT} = 3V$ , $I_{OUT} = 10mA$		100	1	mV
Current limit	ILIM_RTC	$\label{eq:VIN} \begin{array}{l} \text{VIN} = 3.3\text{V}, \ \text{V}_{\text{OUT}} \ \text{drops} \ 33\%, \\ \text{T}_{\text{J}} = 25^{\circ}\text{C} \end{array}$	25	55	85	mA
Soft-start slew rate	tss_rtc	Vout = 10% to 90%, Cout = 1µF		35		mV/µs



# ELECTRICAL CHARACTERISTICS (continued)

## VIN1 = VIN2 = VIN3 = VIN4 = VIN5 = AVIN = 5V, $T_J$ = -40°C to +125°C <sup>(5)</sup>, unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units
Low-Dropout (LDO) Regu	ulator: LDO2	to LDO5				
	VLDO2		2.940	3.00	3.060	V
Outrout wells as	VLDO3		1.764	1.80	1.836	V
Oulput voltage	V <sub>LDO4</sub>		1.176	1.20	1.224	V
PSBR <sup>(8)</sup>	VLDO5		0.882	0.90	0.918	V
	PSRR <sub>1k</sub>	$f_{sw} = 1 kHz$ , 100mA, $V_{OUT} = 1.8V$		47		dB
FORR (*)	PSRR <sub>10k</sub>	$f_{sw} = 10 kHz, 100 mA, V_{OUT} = 1.8 V$		51		dB
Dropout voltage	V <sub>DROP1</sub>	$V_{OUT} = 3V$ , $I_{OUT} = 300mA$		100		mV
Current limit	ILIMIT_LDO	VIN = 3.3V, VOUT drops 33%	320	430	640	mA
Output discharge resistance	Ro_DIS2			7		Ω
Soft-start time	t <sub>SS_B2</sub>	$V_{OUT} = 10\%$ to 90%, $C_{OUT} = 2.2\mu F$		70		μs
Line regulation		VIN2 = VIN5 = 2.8V to 5.5V		0.3		%/V
Load regulation		$VIN2 = VIN5 = 3.3V$ , $I_{OUT}$ from 10mA to 100mA		0.5		%
Logic Pins						
nPBIN pull-up current	PBIN	Internal pull-up to AVIN	5	9	13	μA
Push-button detection threshold	V <sub>PB</sub>		500	700	900	mV
Manual reset threshold	V <sub>MS</sub>	nPBIN pulls low, $T_J = 25^{\circ}C$			50	mV
RSTO rising threshold	V <sub>RSTO_R</sub>	Monitor buck 4's output		90%		V <sub>FB4</sub>
RSTO falling threshold	V <sub>RSTO_F</sub>			80%		$V_{FB4}$
RSTO rising delay	trsto	Adjustable through I <sup>2</sup> C/OTP	70	100	130	ms
I <sup>2</sup> C Interface Specificatio	ns <sup>(9)</sup>					
Input logic high	VIH		1.4			V
Input logic low	VIL				0.4	V
Output voltage logic low	Vout_l	RSTO pin sink 4mA			0.4	V
SCL clock frequency	fscl				3.4	MHz
SCL high time	tнigн		60			ns
SCL low time	t∟ow		160			ns
Data set-up time	<b>t</b> sudat		10			ns
Data hold time	<b>t</b> hddat			70		ns
Set-up time for repeated start condition	<b>t</b> susta		160			ns
Hold time for repeated start condition	<b>t</b> hdsta		160			ns
Bus free time between a start and stop condition	t <sub>BUF</sub>		160			ns
Set-up time for stop condition	tsusto		160			ns



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# ELECTRICAL CHARACTERISTICS (continued)

## VIN1 = VIN2 = VIN3 = VIN4 = VIN5 = AVIN = 5V, $T_J$ = -40°C to +125°C <sup>(5)</sup>, unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units
SCL and SDA rise time	tR		10		300	ns
SCL and SDA fall time	t⊧		10		300	ns
Pulse width of suppressed spike	tsp		0		50	ns
Capacitance bus for each bus line	Св				400	pF
SCL low time	t∟ow		160			ns

#### Notes:

Not tested in production. Guaranteed by over-temperature correlation. 5)

6) Guaranteed by design.

VIN2 and VIN4 share the same UVLO threshold. It is recommended to connect VIN2 and VIN4 together in applications. 7)

8) Guaranteed by engineering sample characterization.

See the I<sup>2</sup>C Timing Diagram below when reading the I<sup>2</sup>C interface specifications. It is recommended to begin operating the I<sup>2</sup>C after the 9) power-on sequence is finished, or RSTO switches high.



tunsta

# I<sup>2</sup>C TIMING DIAGRAM

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Figure 2: I<sup>2</sup>C Timing Diagram

tuici

tsuna

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# **TYPICAL CHARACTERISTICS**

Performance waveforms are tested on the evaluation board.  $V_{IN} = 3.3V$ ,  $T_A = 25^{\circ}C$ , tested using MP5496GR-0001 specified parts, unless otherwise noted.





**Quiescent Current vs. Input Voltage** 









# **TYPICAL PERFORMANCE CHARACTERISTICS**

Performance waveforms are tested on the evaluation board.  $V_{IN} = 3.3V$ ,  $T_A = 25^{\circ}C$ , tested using MP5496GR-0001 specified parts, unless otherwise noted.



Start-Up





**Start-Up** VIN = 3.3V, I<sub>BUCK1</sub> = 4.5A



### Shutdown









# TYPICAL PERFORMANCE CHARACTERISTICS (continued)

Performance waveforms are tested on the evaluation board.  $V_{IN} = 3.3V$ ,  $T_A = 25$ °C, tested using MP5496GR-0001 specified parts, unless otherwise noted.



**Load Transient** I<sub>OUT</sub> = 2A to 4.5A, 2.5A/µs



nPBIN Start-Up

Each channel with no load



Load Transient



nPBIN Start-Up Each channel with no load



### nPBIN Shutdown

Each channel with no load





# TYPICAL PERFORMANCE CHARACTERISTICS (continued)

Performance waveforms are tested on the evaluation board.  $V_{IN} = 3.3V$ ,  $T_A = 25$ °C, tested using MP5496GR-0001 specified parts, unless otherwise noted.



SCP Steady State SCP steady state, no load





#### SCP Recovery SCP recovery, no load



#### SCP Entry SCP entry, full load



SCP Steady State







# **TYPICAL PERFORMANCE CHARACTERISTICS** (continued)

Performance waveforms are tested on the evaluation board.  $V_{IN} = 3.3V$ ,  $T_A = 25^{\circ}C$ , tested using MP5496GR-0001 specified parts, unless otherwise noted.





# FUNCTIONAL BLOCK DIAGRAM



Figure 3: Functional Block Diagram



# OPERATION

The MP5496 provides a complete power management solution for many 5V systems, such as televisions, SSDs, and STBs. The MP5496 integrates four high-frequency, synchronous, rectified, step-down switch-mode converters and five low-dropout regulators. The MP5496 greatly reduces PCB space and the number of external components. The MP5496 can manage the power system for either a

#### **POWER CONTROL**

#### State Machine Diagram

single-cell Li-ion battery or a 5V regulated input voltage, allowing for greater flexibility of the system design.

The I<sup>2</sup>C interface and one-time programmable (OTP) memory interfaces provide an adjustable default output voltage and dynamic voltage scaling. The I<sup>2</sup>C also provides powerful logic functions. See the Register Map section starting on page 27 for more details.



Figure 4: Power Control State Machine Diagram

#### **State Machine Description**

Figure 4 shows the state machine, which has multiple features. These features are described below.

#### No Supply

The PMIC's input pin has an under-voltage lockout (UVLO) detection circuit. If the input voltage (AVIN) is below the UVLO rising threshold, the PMIC's functions are disabled.

#### Power Off

If AVIN exceeds its rising UVLO threshold and the AUTOON bit = 0, the PMIC first enters a power-off state. In this state, the PMIC always monitors the power-on factor. Once the poweron factor is detected, the device initiates the power-on sequence.



#### **Power-On Sequence**

The DC/DC converters and LDO regulators turn on sequentially according to the preconfigured order via the OTP e-fuse.

#### Power-On

The DC/DC converters and LDO regulators turn on. The RSTO output switches high. In this state, the PMIC always monitors the power-off or repower-on factors.

#### **Power-Off Sequence**

The PMIC enters the power-off sequence if it detects the power-off or repower-on factors while in the power-on state. First RSTO switches low, and then the DC/DC converters and LDO regulators turn off sequentially in the reverse order of the power-on sequence. After the power-off sequence is completed, the repower-on condition (software control) is monitored. Then the PMIC enters the power-on sequence automatically after a delay timer.

#### Shutdown Event

If the PMIC detects that the input voltage is below the UVLO falling threshold (the device enters a no supply state) or over-temperature protection is triggered (the device enters a power-off state), the PMIC switches to a no supply state or power-off state, regardless of the current state. <sup>(10)</sup>

#### Note:

10) If PMIC enters a power-off state due to over-temperature protection, LDORTC is off.

#### **Power-On Factors**

The PMIC has three power-on factors, described below.

### SYSEN

SYSEN is one data bit in the I<sup>2</sup>C register. If the SYSEN bit is set to 1, the system changes from a power-off state and initiates the power-on sequence. Two methods can set SYSEN from 0 to 1. The first method is by setting the AUTOON bit to 1 via the OTP. With this method, the system auto-loads the AUTOON bit into SYSEN when the input voltage exceeds the UVLO threshold. The second method is by using the push button to initiate a power-on sequence.

#### nPBIN On

nPBIN on includes two kinds of push-button events. If the nPBIN pin is pulled to logic low (but is not pulled to ground) and asserts low for longer than 2 seconds, the PMIC treats this as a power-on factor. If nPBIN is pulled to ground (below the manual reset threshold) in the poweroff state, and asserts for longer than the maximum 30ms debounce time, the PMIC also treats this as a power-on factor. The SYSEN bit is set high if any of the nPBIN-initiated power-on events described above are detected.

#### **Thermal Recovery**

If the MP5496 is in a power-off state because the die temperature has exceeded the thermal protection threshold, the PMIC enters a power-on sequence once the die's temperature decreases.

#### Power-On Sequence

There are eight slots for power-on sequence timing (see Figure 5). All the DC/DC converters and LDO regulators (except RTCLDO) can be configured to time slots 0 through 7 by the OTP e-fuse. The delay time between each time slot is related to the default switching frequency of the MP5496 (see Table 1).

#### Table 1: Slot Time Interval vs. Default Switching Frequency

Default Switching Frequency	Time Delay between Each Slot
1.0MHz	3ms
1.5MHz	2ms
2.0MHz	1.5ms
2.5MHz	1.2ms







Figure 5: Power-On Sequence (AUTOON Bit is Set to 1)

#### OUTRTC On

The OUTRTC LDO turns on if both VIN2 and AVIN exceed their respective UVLO rising thresholds, regardless of any other pin's status. OUTRTC turns off if either VIN2 or AVIN falls below its respective UVLO falling thresholds, or if thermal shutdown is triggered.

#### Other Buck Regulators and LDOs On

The MP5496 provides a configurable power-on sequence. If the power-on sequence is fixed, the power-off sequence is reversed. See the OTP Register Description section on page 27 to determine which bits set the time slot number for each channel.

## **Power-Off Factors**

#### nPBIN Long Press

If nPBIN is pulled to logic low (but not pulled to ground) and asserts longer than 8 seconds, the PMIC enters the power-off sequence.

#### nPBIN Short Press

If nPBIN is pulled to GND (below the manual reset threshold) and lasts longer than the maximum 30ms debounce time, the PMIC enters the power-off sequence after an 8ms delay time (see Figure 8).

#### SYSEN (Software-Initiated Shutdown)

The MP5496 supports a software-controlled shutdown through the I<sup>2</sup>C interface. SYSEN is one data bit in the I<sup>2</sup>C register. If the SYSEN bit is set to 0, the system enters a power-off sequence. To restart the PMIC, toggle the input power supply or use the nPBIN function (see the nPBIN Functions section on page 22 for more details).





Figure 6: Power-Off Sequence via nPBIN Pin Press

### **Power-Off Sequence**

RSTO is pulled low before the DC/DC converter or LDO regulator turns off (see Figure 6). The DC/DC converter and LDO regulator power-off sequence is in the reverse order of the power-on sequence.

### **Repower-On Factors**

#### Manual Reset

To manually reset the device, pull the nPBIN pin to ground (below the manual reset threshold) for longer than the maximum 30ms debounce time. It is released after a set time (see Figure 8).

#### SFRST

Software reset. If the SFRST bit is set to 1 through the I<sup>2</sup>C interface, the system detects this as a repower-on factor.

# Repower-On Sequence (Software-Initiated Power Cycle)

The MP5496 supports a software-controlled power reset through the I<sup>2</sup>C interface or a manual reset push button.

When using the software-controlled method, the SFRST bit is set high. The MP5496 waits for 8ms and powers off the system, then powers on all of the power rails after a 60ms delay. The SFRST bit is automatically reset to 0 by the RSTO rising edge. After the SFRST bit is reset to 0, the software can control power cycle again. Repower-on factor detection is blocked during the repower-on period (t1 to t2) (see Figure 7).





Figure 7: Repower-On Sequence (Software Control)

If the PMIC is working in a power-on state when using the manual reset control method, and the manual reset button is pressed down, the PMIC enters a power-off sequence. It enters this sequence after a maximum 30ms debounce time and 8ms of delay, then remains in the power-off state until the manual reset button is released. After a 30ms debounce time, the PMIC enters the power-on sequence again, and the manual reset function is also completed (see Figure 8).

### Shutdown Sequence

If the input voltage drops below the UVLO falling threshold, or the IC experiences an overtemperature condition, the PMIC immediately enters the shutdown sequence. All DC/DC converters and LDO regulators turn off at the same time (see Figure 9).



Figure 8: Repower-On Sequence (Manual Reset Control)





Figure 9: Shutdown Sequence

## HIGH-EFFICIENCY BUCK REGULATORS

Buck 1 to buck 4 are synchronous, step-down DC/DC converters with built-in under-voltage lockout (UVLO), soft start, compensation, and hiccup current limit protection. Fixed-frequency constant-on-time (COT) control provides fast transient response. The switching clock is phase-shifted from buck 1 to buck 4 during continuous conduction mode (CCM). Buck 2 and buck 4 support 100% duty cycle mode.

# Power Supply and Under-Voltage Lockout (UVLO)

VIN1 is the power supply for buck 1. VIN2 is the power supply for buck 2, LDORTC, LDO2, and LDO3. VIN3 is the power supply for buck 3. VIN4 is the supply for buck 4. VIN5 is the power supply for LDO4 and LDO5. AVIN is the power input for the biased internal logic blocks.

VIN1, VIN3, VIN5, and AVIN have their own under-voltage lockout (UVLO) thresholds with a proper hysteresis. VIN2 and VIN4 share the same UVLO threshold. If AVIN ramps up and exceeds the UVLO rising threshold, the nPBIN logic is enabled and ready to accept start-up and shutdown commands. LDORTC is active once VIN2 exceeds the rising threshold. Before the power key turns on, the input shutdown current is typically 15µA.

### Internal Soft Start (SS)

Soft start (SS) is implemented to prevent the PMIC output voltage from overshooting during start-up. When the PMIC starts up, the internal circuitry of each power rail generates a soft-start voltage that ramps up from 0V. The soft-start

period lasts until the voltage on the soft-start capacitor exceeds the reference voltage. At this point, the reference voltage takes over. For the four buck outputs, their soft-start times are fixed internally at 450µs. For the LDO2-5 outputs, their soft-start times are fixed internally at 70µs. For LDORTC, the soft-start slew rate is always 35mV/µs.

## **Output Discharge**

To discharge the energy of the output capacitor during a power-off sequence, there is an active discharge path from the DC/DC converters' and LDO regulators' output to ground. The discharge path turns on when the corresponding channel is disabled. The typical discharge resistance is  $7\Omega$ . The discharge function can be enabled or disabled through the I<sup>2</sup>C interface.

### SYSTEM CONTROL SIGNALS

### **nPBIN Functions**

nPBIN is a multi-function pin that supports pushbutton detection and manual reset functions. There is an internal pull-up current to pull up nPBIN's voltage to AVIN. The MP5496 distinguishes between the push-button and manual reset functions by the different pull-low resistances. Connect nPBIN to ground for a manual reset function. Connect nPBIN to ground through a  $34k\Omega$  resistor to generate a pushbutton signal (see Figure 10).

A push-button event and manual reset event can both generate an interrupt signal, and set the corresponding interrupt bit high. See the Status 2 register on page 34 for details.





#### Figure 10: nPBIN Functional Block Diagram

### **Push-Button Control**

Long Press 1/Start-Up: If AVIN exceeds the UVLO threshold and the push button asserts low for longer than 2 seconds when the PMIC is in the power-off state, the power-on sequence

begins. The power-on sequence must be completed before the backside CPU can take over control. The power-on sequence must be complete when the RSTO signal switches high (see Figure 11).



#### Figure 11: nPBIN Push-Button Long Press 1 (Start-Up)

Long Press 2/Shutdown: If the push button asserts low for more than 8 seconds during the power-on state, the power-off sequence begins. The MP5496 turns off all regulators and LDOs (excluding OUTRTC). The power-off sequence turns off all components in the reverse order of start-up. If nPBIN is pulled low through a  $34k\Omega$  resistor at all times, the MP5496 remains in the power-off state (see Figure 12).







#### Manual Reset Control

Short Press and Release/Manual Reset: If the PMIC is in a power-off state, a short press on nPBIN pulls the nPBIN voltage below the manual reset threshold with a maximum 30ms debounce

time, and then the PMIC begins the power-on sequence. When the PMIC is powered on, a short press drops the nPBIN voltage below the manual reset threshold with a maximum 30ms debounce time, and triggers the manual reset function (see Figure 13).



Figure 13: nPBIN Push Button Short Press to Ground

If the manual reset function is triggered, the PMIC turns off and remains off until the manual reset button is released. The PMIC enters the power-on state again after a maximum 30ms debounce time.

#### Automatic Turn-On

If the AUTOON bit in the OTP configuration table is set high, the system changes the default value of SYSEN to 1. The PMIC enters the power-on sequence automatically once the input voltage (AVIN) exceeds its under-voltage lockout (UVLO) threshold. The system can start up automatically without the push button being pressed. After start-up, the push button can still support the manual power-on and power-off control. SYSEN can be read or written by the I<sup>2</sup>C.

#### **RSTO (Reset Output)**

When buck 4's output voltage is ready ( $V_{FB}$  exceeds 90% of  $V_{REF}$ ), RSTO outputs high to enable the processer after an RSTO delay time. RSTO is an open-drain structure with an external pull-up resistor. RSTO is pulled low when buck 4's output is below 80% of the nominal value, or when the system detects a power-off factor, shutdown factor, or repower-on factor.

#### **Thermal Warning and Shutdown**

Thermal warning and shutdown prevent the part from operating at exceedingly high temperatures. If the silicon die temperature exceeds 120°C, the MP5496 sets the OTWARNING bit to 1.

If the die temperature exceeds 153°C, the MP5496 sets the OTEMPP bit to 1. Meanwhile, the system enters the shutdown sequence. When the temperature recovers to 130°C, the regulator enters the power-on sequence again.

### I<sup>2</sup>C Timing Graph

The I<sup>2</sup>C interface of the PMIC is powered by an internal, fixed, 2V power supply. When VIN exceeds its UVLO threshold during VIN start-up, this 2V power supply is ready after a 0.5ms delay. After another 5ms of delay time, the I<sup>2</sup>C is available (see Figure 14).





#### I<sup>2</sup>C INTERFACE

#### I<sup>2</sup>C Serial Interface Description

The I<sup>2</sup>C is a two-wire, bidirectional, serial interface consisting of a data line (SDA) and a clock line (SCL). The lines are externally pulled to a bus voltage when they are idle. A master device connected to the line generates the SCL signal and device address, and arranges the communication sequence.

The MP5496 interface is an  $I^2C$  slave that supports both fast mode (400kHz) and highspeed mode (3.4MHz). The  $I^2C$  interface adds flexibility to the power supply solution. The output voltage, transition slew rate, and other parameters can be controlled by the  $I^2C$  interface instantaneously. When the master sends the address as an 8-bit value, the 7-bit address should be followed by a 0 or 1 to indicate a write or read operation, respectively.

#### Start and Stop Conditions

Start (S) and stop (P) conditions are signaled by the master device, which signifies the beginning and end of the I<sup>2</sup>C transfer. The start condition is defined as the SDA signal transitioning from high to low while the SCL is high. The stop condition is defined as the SDA signal transitioning from low to high while the SCL is high (see Figure 15).



Figure 15: Start and Stop Conditions

The master generates the SCL clocks, then transmits the device address and the read/write direction bit (R/W) on the SDA line.

#### **Transfer Data**

Data is transferred in 8-bit bytes by the SDA line. Each byte of data must be followed by an acknowledge (ACK) bit.

#### I<sup>2</sup>C Update Sequence

The MP5496 requires a start condition, valid I<sup>2</sup>C address, register address byte, and a data byte for a single data update. The MP5496 acknowledges each byte that has been received by pulling the SDA line low during the high period of a single clock pulse. A valid I<sup>2</sup>C address selects the MP5496. The MP5496 performs an update on the falling edge of the LSB byte. Figure 16, Figure 17, and Figure 18 show examples of I<sup>2</sup>C write and read sequences.







Figure 18: I<sup>2</sup>C Read Example (Read Single Register)



# **OTP REGISTER DESCRIPTION**

#### **OTP E-Fuse Configuration Table**

#	NAME	D7	D6	D5	D4	D3	D2	D1	D0
00	CTL1	AUTOON	FREQUE	INCY	PUS	HBUTTONTI	MER	RSTOR	DELAY
01	CTL2	ILIMBU	CK1	ILIME	SUCK3	N/A	PWR	ONDELAYBL	JCK1
02	CTL3	MODEBUCK1	N/A	PWF	RONDELAYE	BUCK2	PWR	ONDELAYBL	JCK3
03	CTL4	MODEBUCK2	N/A	PWF	RONDELAYE	BUCK4	PWF	RONDELAYLI	DO2
04	CTL5	MODEBUCK3	N/A	PW	RONDELAY	LDO3	PWF	RONDELAYLI	DO4
05	CTL6	MODEBUCK4	N/A	PW	RONDELAY	LDO5		N/A	
06	VSET1	ENBUCK1	Buck 1 V <sub>C</sub>	Buck 1 V <sub>OUT</sub> Set: 0.6V to 2.1V/100mV step I <sup>2</sup> C SL			I <sup>2</sup> C SLAVE	E ADDRESS /	A3, A2, A1
07	VSET2	ENBUCK2	Bu	ick 2 V <sub>OUT</sub> S	Set: 0.6V to 2	2.1/50mV step	)	N/.	A
08	VSET3	ENBUCK3	Buck 3 Vo	оот Set: 0.6	/ to 2.1V/100	)mV step		N/A	
09	VSET4	ENBUCK4	Buc	с <mark>k 4 V</mark> оит S	et: 0.6V to 2.	.1V/50mV ste	0	N/.	A
10	VSET5	RESERVED	LDO	RTC VOUT S	Set: 0.8V to 3	8.9V/100mV st	ep	N/.	A
11	VSET6	ENLDO2	LD	O2 Vout Se	t: 0.8V to 3.9	V/100mV ste	р	N/.	A
12	VSET7	ENLDO3	LDO3 VOUT Set: 0.8V to 3.9V/100mV step			N/.	A		
13	VSET8	ENLDO4	LDO4 VOUT Set: 0.8V to 3.9V/100mV step			OTP version	on D1, D0		
14	VSET9	ENLDO5	LDO	O5 VOUT Se	t: 0.8V to 3.9	V/100mV ste	p	OTP version	on D3, D2

#### OTP E-Fuse Selected Table by Default (MP5496GR-0001)

OTP Items	Buck 1	Buck 2	Buck 3	Buck 4	LDORTC	LDO2	LDO3	LDO4	LDO5
Output voltage	1.0V	0.9V	0.9V	1.8V	1.2V	3V	1.8V	1.2V	0.9V
Initial on/off	On	On	On	On	On	On	Off	On	On
Mode	FPWM	FPWM	FPWM	FPWM			N/A		
Power-on delay/time slot #	0ms/0	4ms/2	2ms/1	6ms/3	Always on	12ms/6	N/A	8ms/4	10ms/ 5
Automatic turn-on					Yes				
Switching frequency		1.5MHz							
Push-button timer				2	seconds				
RSTO delay					100ms				
Buck 1 peak current limit					6.8A				
Buck 3 peak current limit	5.6A								
I <sup>2</sup> C slave address	0x69								
OTP version					0004				

#### Other Parameter Information for the MP5496GR-0001

	Buck 1	Buck 2	Buck 3	Buck 4	LDORTC	LDO2	LDO3	LDO4	LDO5
Minimum input voltage					ЗV				
Typical input voltage	3.3V								
Maximum input voltage	3.6V								



# **OTP REGISTERS**

Bits	Name	Default	Descri	ption							
D[7]	AUTOON	1	System sequen (nPBIN SYSEN	automatic turn-on ice when AVIN ex ) does not have to I register after AVII	bit. If AUTOON is xceeds the UVLO be pressed. The A N reaches the UVL	set high, the syste rising threshold. UTOON bit inform O rising threshold.	m enters the power-on Then the push button nation is loaded into the				
D[6:5]	FREQUENCY	01	Switch 00: fsw 01: fsw 10: fsw 11: fsw	ing frequency sele = 1MHz = 1.5MHz = 2MHz = 2.5MHz	ection bit.						
			This bi	t sets the push-bu	tton long press 1 p	oower-on deglitch	timer.				
				f <sub>sw</sub> = 1MHz	f <sub>sw</sub> = 1.5MHz	f <sub>sw</sub> = 2MHz	f <sub>sw</sub> = 2.5MHz				
			000	0.75s	0.5s	0.375s	0.3s				
			001	1.5s	1s	0.75s	0.6s				
			010	2.25s	1.5s	1.225s	0.9s				
D[4:2]		011	011	3s	2s	1.5s	1.2s				
			100	3.75s	2.5s	1.875s	1.5s				
			101	4.5s	3s	2.25s	1.8s				
			110	5.25s	3.5s	2.625s	2.1s				
			111	6s	4s	3s	2.4s				
			There i	There is no corresponding data in the I <sup>2</sup> C register table for these 3 bits.							
			This bi	t sets the reset ou	tput delay.						
				fsw = 1MHz	fsw = 1.5MHz	fsw = 2MHz	fsw = 2.5MHz				
			00	210ms	140ms	105ms	84ms				
D[1:0]	RSTODELAY	01	01	150ms	100ms	75ms	60ms				
			10	75ms	50ms	37.5ms	30ms				
			11	15ms	10ms	7.5ms	6ms				
			There is no corresponding data in the I <sup>2</sup> C register table for these 3 bits.								
	PWRONDELAY BUCK1										
	PWRONDELAY										
	PWRONDELAY		Delay t	ime from when S	YSEN goes high to	when the bucks/	LDOs begin to switch.				
			000: Or	ns (time slot 0)							
	BUCK4		001:2r	ns (time slot 1)							
	PWRONDELAY		010:41 011:6r	ns (time slot 2)							
3 bits	LDO2	N/A	100: 8r	ns (time slot 4)							
0 210	PWRONDELAY		101:10	Oms (time slot 5)							
			111:14	Ams (time slot 7)							
	LDO3		The de	lay time between	neighboring slots a	ire related to the P	MIC default switching				
	PWRONDELAY		freque	ncy. See the op	eration section o	n page 18 for	details. There is no				
	LDO4		corresp	oonding data in th	e l <sup>2</sup> C register table	for these 3 bits.					
	PWRONDELAY										
	LDO5										
			1								



Bits	Name	Default	Description
1 bit	MODE BUCK1 MODE BUCK2 MODE BUCK3 MODE BUCK4	N/A	Selects the mode (auto-PFM/PWM mode or forced PWM mode). 0: Auto-PFM/PWM mode 1: Forced PWM mode (default for buck 1, buck 2, buck 3, and buck 4) These bits are loaded into the I <sup>2</sup> C MODEBUCKx registers when VIN1 exceeds its UVLO threshold.
2 bits	ILIMBUCK1 ILIMBUCK3	11 (Buck 1) 10 (Buck 3)	Programs the current limit threshold of the buck regulator: 00: 3.8A typical high-side peak current limit 01: 4.6A typical high-side peak current limit 10: 5.6A typical high-side peak current limit 11: 6.8A typical high-side peak current limit
3 bits	I2C SLAVE ADDRESS A3, A2, A1	001	Sets the A1, A2, and A3 bits of the slave's $I^2C$ address. See the $I^2C$ Bus Slave Address section on page 30 for more details.



# I<sup>2</sup>C REGISTER MAP

Add (Hex)	Name	R/W	D7	D6	D5	D4	D3	D2	D1	D0
00	CTL0	R/W	SYSEN	SFRST	SFRST RESERVED					RESERVED
01	CTL1	R/W	RESERVED	MODE BUCK1	MODE BUCK2	MODE BUCK3	MODE BUCK4	DISCHG BUCK3	DISCHG BUCK2	DISCHG BUCK1
02	CTL2	R/W	DVS SLEV	V RATE	DISCHG BUCK4	DISCHG LDO2	DISCHG LDO3	DISCHG LDO4	DISCHG LDO5	RESERVED
03	ILIMIT	R/W	ILIMBU	CK1	ILIME	BUCK3	ILIMB	UCK2	ILIMB	UCK4
04	VSET1	R/W	ENBUCK1		В	uck 1 output vol	tage set: 0.6V to	2.1875V/12.5m	V step	
05	VSET2	R/W	ENBUCK2		В	uck 2 output vol	tage set: 0.6V to	2.1875V/12.5m	V step	
06	VSET3	R/W	ENBUCK3		В	uck 3 output vol	tage set: 0.6V to	2.1875V/12.5m	V step	
07	VSET4	R/W	ENBUCK4		В	uck 4 output vol	tage set: 0.6V to	2.1875V/12.5m	V step	
08	VSET5	R/W	ENLDO2			LDO2 output vo	oltage set: 0.8V t	o 3.975V/25mV :	step	
09	VSET6	R/W	ENLDO3			LDO3 output vo	oltage set: 0.8V t	o 3.975V/25mV :	step	
0A	VSET7	R/W	ENLDO4			LDO4 output vo	oltage set: 0.8V t	o 3.975V/25mV :	step	
0B	VSET8	R/W	ENLDO5			LDO5 output vo	oltage set: 0.8V t	o 3.975V/25mV :	step	
0C						RESERVED				
0D	Status1	R	PGLDO4	PGLDO3	PGLDO2	PGRTC	PG4	PG3	PG2	PG1
0E	Status2	R	KEYON	KEYOFF	MREST	SHORTKEY ON	SFRST_ON	RESERVED	RESERVED	PGLDO5
0F	Status3	R	OT WARNING	OTEMPP	RESERVE D	RESERVED	RESERVED	RESERVED	RESERVED	RESERVED
10						RESERVED				
11	ID2	R		VEND	OR ID			OTP VI	ERSION	

# **REGISTER DESCRIPTION**

#### I<sup>2</sup>C Bus Slave Address (11)

The slave address is 7 bits followed by an 8th data direction bit (read or write). The A3, A2, and A1 bits can be configured via the OTP e-fuse.

	A7	A6	A5	A4	A3	A2	A1
Setting Value	1	1	0	1	0 (12)	0 (12)	1 (12)

#### Notes:

11) By default, the slave address is 0x69, A[7:1] = 1101 001.

12) This bit is configurable via the OTP e-fuse.



# I<sup>2</sup>C REGISTERS

## Reg 00: CTL0

Bits	Name	Default	Description	Reset Condition
D[7]	SYSEN	BY OTP AUTO ON	System enable on/off bit. If the MP5496 detects a power-on event, this bit is set to 1. Then the power-on sequence starts. The DC/DC converters and LDO regulators turn on sequentially according to their enable bits (e.g. ENBUCK1 = 1) and power- on delay (POWERONDELAYBUCK1) settings. Set this bit from 1 to 0 to trigger a power-off sequence. Other I <sup>2</sup> C registers are not reset when the I <sup>2</sup> C sets SYSEN from 1 to 0. The MP5496 is enabled again until the push button is long- pressed or a manual reset is asserted for more than 30ms of the maximum debounce time.	AVIN < UVLO
D[6]	SFRST	0	Software reset. Once the SFRST bit is set high, the MP5496 waits for 8ms, then restarts all of the power rails. When the RSTO signal switches from low to high, the MP5496 resets SFRST and sets it to 0.	<ul> <li>AVIN &lt; UVLO</li> <li>RSTO from low to high</li> </ul>

## Reg 01: CTL1

Bits	Name	Default	Description	Reset Condition
D[6:3]	MODE BUCK1 MODE BUCK2 MODE BUCK3 MODE BUCK4	BY OTP	PFM/PWM mode or forced PWM mode. 0: Auto-PFM/PWM 1: Forced PWM mode	<ul> <li>AVIN &lt; UVLO</li> <li>SFRST</li> <li>Manual reset</li> <li>Long press 2</li> </ul>
D[2:0]	Regulator Discharge	1	Output discharge enable bit. The output discharge function is active during the power-off sequence.	

## Reg 02: CTL2

Bits	Name	Default	Description	Reset Condition
D[7:6]	DVS SLEW RATE	01	Voltage scaling slew rate for the buck 1 to buck 4 converters. 00: 32mV/µs 01: 16mV/µs 10: 8mV/µs 11: 4mV/µs	<ul> <li>AVIN &lt; UVLO</li> <li>SFRST</li> <li>Manual reset</li> </ul>
D[5:0]	Regulator Discharge	1	Output discharge enable bit. The output discharge function is active during the power-off sequence and remains active after shutdown.	Long press 2



#### Reg 03: ILIMIT

Bits	Name	Default	Description	Reset Condition
D[7:6] D[5:4]	ILIMBUCK1 ILIMBUCK3	ВҮ ОТР	Sets the current-limit threshold of the buck regulator. 00: 3.8A typical high-side peak current limit 01: 4.6A typical high-side peak current limit 10: 5.6A typical high-side peak current limit 11: 6.8A typical high-side peak current limit	<ul> <li>AVIN &lt; UVLO</li> <li>SFRST</li> <li>Manual reset</li> <li>Long press 2</li> </ul>
D[3:2] D[1:0]	ILIMBUCK2 ILIMBUCK4	10	Sets the current-limit threshold of the buck regulator. 00: 2.2A typical high-side peak current limit 01: 3.2A typical high-side peak current limit 10: 4.2A typical high-side peak current limit 11: 5.2A typical high-side peak current limit	

#### Reg 04 to Reg 0B: VOUTSET and EN

Bits	Name	Default	Description	Reset Condition
D[7]	ENBUCK1 ENBUCK2 ENBUCK3 ENBUCK4 ENLDO2 ENLDO3 ENLDO4 ENLDO5	BY OTP	Enable bit for buck regulators and LDOs. The default value is 1, but the default SYSEN is set to 0. Regulators are enabled when both ENx = 1 and SYSEN = 1.	<ul> <li>AVIN &lt; UVLO</li> <li>SFRST</li> <li>Manual react</li> </ul>
D[6:0]	BUCK1 VOUTSET BUCK2 VOUTSET BUCK3 VOUTSET BUCK4 VOUTSET	BY OTP	Sets the output voltage from 0.6V to 2.1875V with 12.5mV steps (see Table 2).	Long press 2

#### Table 2: Output Voltage Chart for Buck 1, Buck 2, Buck 3, and Buck 4

D[6:0]	V <sub>OUT</sub> (V)	D[6:0]	V <sub>OUT</sub> (V)	D[6:0]	V <sub>оит</sub> (V)	D[6:0]	V <sub>OUT</sub> (V)
0000000	0.6000	0100000	1.0000	1000000	1.4000	1100000	1.8000
0000001	0.6125	0100001	1.0125	1000001	1.4125	1100001	1.8125
0000010	0.6250	0100010	1.0250	1000010	1.4250	1100010	1.8250
0000011	0.6375	0100011	1.0375	1000011	1.4375	1100011	1.8375
0000100	0.6500	0100100	1.0500	1000100	1.4500	1100100	1.8500
0000101	0.6625	0100101	1.0625	1000101	1.4625	1100101	1.8625
0000110	0.6750	0100110	1.0750	1000110	1.4750	1100110	1.8750
0000111	0.6875	0100111	1.0875	1000111	1.4875	1100111	1.8875
0001000	0.7000	0101000	1.1000	1001000	1.5000	1101000	1.9000
0001001	0.7125	0101001	1.1125	1001001	1.5125	1101001	1.9125
0001010	0.7250	0101010	1.1250	1001010	1.5250	1101010	1.9250
0001011	0.7375	0101011	1.1375	1001011	1.5375	1101011	1.9375
0001100	0.7500	0101100	1.1500	1001100	1.5500	1101100	1.9500
0001101	0.7625	0101101	1.1625	1001101	1.5625	1101101	1.9625
0001110	0.7750	0101110	1.1750	1001110	1.5750	1101110	1.9750
0001111	0.7875	0101111	1.1875	1001111	1.5875	1101111	1.9875
0010000	0.8000	0110000	1.2000	1010000	1.6000	1110000	2.0000
0010001	0.8125	0110001	1.2125	1010001	1.6125	1110001	2.0125
0010010	0.8250	0110010	1.2250	1010010	1.6250	1110010	2.0250
0010011	0.8375	0110011	1.2375	1010011	1.6375	1110011	2.0375



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0010100	0.8500	0110100	1.2500	1010100	1.6500	1110100	2.0500
0010101	0.8625	0110101	1.2625	1010101	1.6625	1110101	2.0625
0010110	0.8750	0110110	1.2750	1010110	1.6750	1110110	2.0750
0010111	0.8875	0110111	1.2875	1010111	1.6875	1110111	2.0875
0011000	0.9000	0111000	1.3000	1011000	1.7000	1111000	2.1000
0011001	0.9125	0111001	1.3125	1011001	1.7125	1111001	2.1125
0011010	0.9250	0111010	1.3250	1011010	1.7250	1111010	2.1250

#### Table 2: Output Voltage Chart for Buck 1, Buck 2, Buck 3, and Buck 4 (continued)

D[6:0]	V <sub>оυт</sub> (V)	D[6:0]	V <sub>оит</sub> (V)	D[6:0]	V <sub>оит</sub> (V)	D[6:0]	V <sub>OUT</sub> (V)
0011011	0.9375	0111011	1.3375	1011011	1.7375	1111011	2.1375
0011100	0.9500	0111100	1.3500	1011100	1.7500	1111100	2.1500
0011101	0.9625	0111101	1.3625	1011101	1.7625	1111101	2.1625
0011110	0.9750	0111110	1.3750	1011110	1.7750	1111110	2.1750
0011111	0.9875	0111111	1.3875	1011111	1.7875	1111111	2.1875

## LDO Regulator Output Voltage Set

Bits	Name	Default	Description	Reset Condition
D[6:0]	LDOVOUT SET	BY OTP	Sets the output voltage from 0.8V to 3.975V with 25mV steps (see Table 3).	<ul> <li>AVIN &lt; UVLO</li> <li>SFRST</li> <li>Manual reset</li> <li>Long press 2</li> </ul>

#### Table 3: Output Voltage Chart of all LDOs

D[6:0]	V <sub>OUT</sub> (V)						
0000000	0.800	0100000	1.600	1000000	2.400	1100000	3.200
0000001	0.825	0100001	1.625	1000001	2.425	1100001	3.225
0000010	0.850	0100010	1.650	1000010	2.450	1100010	3.250
0000011	0.875	0100011	1.675	1000011	2.475	1100011	3.275
0000100	0.900	0100100	1.700	1000100	2.500	1100100	3.300
0000101	0.925	0100101	1.725	1000101	2.525	1100101	3.325
0000110	0.950	0100110	1.750	1000110	2.550	1100110	3.350
0000111	0.975	0100111	1.775	1000111	2.575	1100111	3.375
0001000	1.000	0101000	1.800	1001000	2.600	1101000	3.400
0001001	1.025	0101001	1.825	1001001	2.625	1101001	3.425
0001010	1.050	0101010	1.850	1001010	2.650	1101010	3.450
0001011	1.075	0101011	1.875	1001011	2.675	1101011	3.475
0001100	1.100	0101100	1.900	1001100	2.700	1101100	3.500
0001101	1.125	0101101	1.925	1001101	2.725	1101101	3.525
0001110	1.150	0101110	1.950	1001110	2.750	1101110	3.550
0001111	1.175	0101111	1.975	1001111	2.775	1101111	3.575
0010000	1.200	0110000	2.000	1010000	2.800	1110000	3.600
0010001	1.225	0110001	2.025	1010001	2.825	1110001	3.625
0010010	1.250	0110010	2.050	1010010	2.850	1110010	3.650
0010011	1.275	0110011	2.075	1010011	2.875	1110011	3.675
0010100	1.300	0110100	2.100	1010100	2.900	1110100	3.700
0010101	1.325	0110101	2.125	1010101	2.925	1110101	3.725
0010110	1.350	0110110	2.150	1010110	2.950	1110110	3.750
0010111	1.375	0110111	2.175	1010111	2.975	1110111	3.775
0011000	1.400	0111000	2.200	1011000	3.000	1111000	3.800
0011001	1.425	0111001	2.225	1011001	3.025	1111001	3.825
0011010	1.450	0111010	2.250	1011010	3.050	1111010	3.850
0011011	1.475	0111011	2.275	1011011	3.075	1111011	3.875
0011100	1.500	0111100	2.300	1011100	3.100	1111100	3.900
0011101	1.525	0111101	2.325	1011101	3.125	1111101	3.925
0011110	1.550	0111110	2.350	1011110	3.150	1111110	3.950
0011111	1.575	0111111	2.375	1011111	3.175	1111111	3.975



## Reg 0D: Status 1

Status registers are non-latch. This register automatically updates according to its real-time status.

Bits	Name	Description	Reset Condition
D[7:0]	PG1 PG2 PG3 PG4 PGLDO2 PGLDO3 PGLDO4 PGLDOBTC	Power good indicator for the bucks and LDOs. $PG = 1$ when the output voltage exceeds 90% of the reference voltage. $PG = 0$ when the output voltage is below 80% of the reference voltage. During the I <sup>2</sup> C-controlled dynamic voltage scaling, the PG deglitch timer blanks the possible PG glitch.	<ul> <li>AVIN &lt; UVLO</li> <li>SFRST</li> <li>Manual reset</li> <li>Long press 2</li> </ul>

#### Reg 0E: Status 2

Bits	Name	Description		Reset Condition	
D[7]	KEYON	A push-button power-on event (long press 1) is detected.			
D[6]	KEYOFF	A push-button power-off event (long press 2) is detected.	These bits are		
D[5]	MREST	A manual reset event is detected.			
D[4]	SHORTKEYON	If the MP5496 is in a power-off state, nPBIN is pulled to ground and lasts longer than the maximum 30ms debounce time. Then the SHORTKEYON bit is set high.	AVIN < UVLO		
D[3]	SFRST_ON	A software reset event is detected. When the SFRST bit is set high, the SFRST_ON bit goes high, and the high state is latched.			
D[0]	PGLDO5	Power good indicator for LDO5. Non-latch bits.	<ul> <li>AVIN &lt; UVLO</li> <li>SFRST</li> <li>Manual reset</li> <li>Long press 2</li> </ul>		

## Reg 0F: Status 3

Bits	Name	Description	Reset Condition	
D[7]	OTWARNING	Die temperature early warning bit. If the bit is high, the die temperature has exceeded 120°C.	This bit is latched once it is	
D[6]	OTEMPP	Over-temperature indication. If the bit is high, the IC is in thermal shutdown.	triggered. It is cleared by a read action to the status register.	AVIN < UVLO

#### Reg 11: ID2

Bits	Name	Description
D[7:4]	Vendor ID	1000.
D[3:0]	OTP version	0100.



## **APPLICATION INFORMATION**

#### Selecting the Inductor

For most applications, use a  $0.47\mu$ H to  $2.2\mu$ H inductor with a DC current rating at least 25% greater than the maximum load current. For the highest efficiency, use an inductor with a DC resistance below  $15m\Omega$ . For most designs, the inductance value can be calculated with Equation (1):

$$L_{1} = \frac{V_{\text{OUT}} \times (V_{\text{IN}} - V_{\text{OUT}})}{V_{\text{IN}} \times \Delta I_{\text{L}} \times f_{\text{OSC}}}$$
(1)

Where  $\Delta I_{L}$  is the inductor ripple current.

Choose the inductor ripple current to be approximately 30% of the maximum load current. The maximum inductor peak current can be estimated with Equation (2):

$$I_{L(MAX)} = I_{LOAD} + \frac{\Delta I_{L}}{2}$$
(2)

Use a larger inductor (<100mA) to improve efficiency under light-load conditions.

# Selecting the Step-Down Converter Input Capacitor

The step-down converter has discontinuous input current, and requires a capacitor to supply AC current to the converter while maintaining the DC input voltage. Use-low ESR capacitors for the best performance. Ceramic capacitors with X5R or X7R dielectrics are recommended because of their low ESR and small temperature coefficients. For most applications, use a 22µF capacitor.

Since C1 absorbs the input switching current, it requires an adequate ripple current rating. The RMS current in the input capacitor can be estimated with Equation (3):

$$I_{C1} = I_{LOAD} \times \sqrt{\frac{V_{OUT}}{V_{IN}}} \times \left(1 - \frac{V_{OUT}}{V_{IN}}\right)$$
(3)

The worst-case condition occurs at  $V_{IN} = 2V_{OUT}$ , calculated with Equation (4):

$$I_{C1} = \frac{I_{LOAD}}{2}$$
(4)

For simplification, choose an input capacitor with an RMS current rating greater than half of the maximum load current.

The input capacitor can be electrolytic, tantalum, or ceramic. When using electrolytic or tantalum capacitors, add a small, high-quality, ceramic capacitor (e.g.  $0.1\mu$ F) placed as close to the IC as possible. When using ceramic capacitors, ensure that they have enough capacitance to provide sufficient charge to prevent excessive voltage ripple at the input. The input voltage ripple caused by the capacitance can be estimated with Equation (5):

$$\Delta V_{\rm IN} = \frac{I_{\rm LOAD}}{f_{\rm SW} \times C1} \times \frac{V_{\rm OUT}}{V_{\rm IN}} \times \left(1 - \frac{V_{\rm OUT}}{V_{\rm IN}}\right)$$
(5)

# Selecting the Step-Down Converter Output Capacitor

The output capacitor for the step-down regulator maintains the DC output voltage. Use ceramic, tantalum, or low-ESR electrolytic capacitors. For the best results, use low-ESR capacitors to keep the output voltage ripple low. The output voltage ripple can be estimated with Equation (6):

$$\Delta V_{\text{OUT}} = \frac{V_{\text{OUT}}}{f_{\text{SW}} \times L_{1}} \times \left(1 - \frac{V_{\text{OUT}}}{V_{\text{IN}}}\right) \times \left(R_{\text{ESR}} + \frac{1}{8 \times f_{\text{SW}} \times C2}\right)$$
(6)

Where  $L_1$  is the inductor value, and  $R_{\text{ESR}}$  is the equivalent series resistance (ESR) value of the output capacitor.

For ceramic capacitors, the capacitance dominates the impedance at the switching frequency and causes the majority of the output voltage ripple. For simplification, the output voltage ripple can be estimated with Equation (7):

$$\Delta V_{\text{OUT}} = \frac{V_{\text{OUT}}}{8 \times f_{\text{SW}}^2 \times L_1 \times C2} \times \left(1 - \frac{V_{\text{OUT}}}{V_{\text{IN}}}\right)$$
(7)

For tantalum or electrolytic capacitors, the ESR dominates the impedance at the switching frequency. For simplification, the output ripple can be calculated with Equation (8):

$$\Delta V_{\text{OUT}} = \frac{V_{\text{OUT}}}{f_{\text{SW}} \times L_{1}} \times \left(1 - \frac{V_{\text{OUT}}}{V_{\text{IN}}}\right) \times R_{\text{ESR}}$$
(8)

The characteristics of the output capacitor also affect the stability of the regulation.

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# Recommended External Components for DC/DC and LDO Converters

Table 4 lists recommended external components for the DC/DC converters and LDO regulators.

Component	Value	Description
VIN1 C <sub>IN</sub>	22µF	0805 size/10V ceramic capacitor
VIN2 CIN	22µF	0805 size/10V ceramic capacitor
VIN3 CIN	22µF	0805 size/10V ceramic capacitor
VIN4 C <sub>IN</sub>	22µF	0805 size/10V ceramic capacitor
VIN5 CIN	10µF	0805 size/10V ceramic capacitor
	1µF	0603 size/10V ceramic capacitor
Buck 1 COUT	22µF x 2	0805 size/10V ceramic capacitor
Buck 1 inductance	1µH	Isat > current limit
Buck 2 Cout	22µF	0805 size/10V ceramic capacitor
Buck 2 inductance	1µH	I <sub>SAT</sub> > current limit
Buck 3 Cout	22µF x 2	0805 size/10V ceramic capacitor
Buck 3 inductance	1μH	Isat > current limit
Buck 4 Cout	22µF	0805 size/10V ceramic capacitor
Buck 4 inductance	1.5µH	Isat > current limit
RTCLDO COUT	1µF	0603 size/6.3V ceramic capacitor
LDO2 COUT	2.2µF	0603 size/6.3V ceramic capacitor
LDO3 Cout	2.2µF	0603 size/6.3V ceramic capacitor
LDO4 Cout	2.2µF	0603 size/6.3V ceramic capacitor
LDO5 COUT	2.2µF	0603 size/6.3V ceramic capacitor
RSTO pull-up resistor	100kΩ	0603 or 0402 size film resistor
nPBIN pull-low resistor	34kΩ	Push-button function, 0603 or 0402 size film resistor
AVIN series resistor to VIN1	10Ω	0603 or 0402 size film resistor

#### **Table 4: Recommended External Components**



#### PCB Layout Guidelines (13)

Efficient PCB layout is critical for stable operation. A 4-layer layout is recommended for the best performance. For the best results, refer to Figure 14 and follow the guidelines below:

- 1. Connect the input ground to GND using the shortest and widest trace possible.
- 2. Connect the input capacitor to VIN using the shortest and widest trace possible.
- 3. Ensure that FB1, FB2, FB3, and FB4 are Kelvin-connected to the buck output capacitors. Do not directly connect FB to the inductor's output node.
- 4. Route SW away from sensitive analog areas, such as FB1 to FB4.

#### Note:

13) The recommended layout is based on Figure 20 on page 38.



Figure 19: Recommended Layout (14)

#### Note:

14) It is recommended to separate Buck 1/3 and Buck 2/4's PGND on the top layer.



# **TYPICAL APPLICATION CIRCUIT**



Figure 20: Typical Application Circuit (15)

#### Note:

15) The minimum input voltage of VIN5 is equal to the maximum nominal output voltage of LDO4 and LDO5.



# **PACKAGE INFORMATION**

#### **QFN-28 (4mmx4mm)**



TOP VIEW



#### **BOTTOM VIEW**







#### **RECOMMENDED LAND PATTERN**

#### NOTE:

 LAND PATTERNS OF PINS 1, 6, 15, AND 20 HAVE THE SAME LENGTH AND WIDTH.
 LAND PATTERNS OF PINS 7, 14, 21, AND 28 HAVE THE SAME LENGTH AND WIDTH.
 ALL DIMENSIONS ARE IN MILLIMETERS.
 LEAD COPLANARITY SHALL BE 0.10 MILLIMETERS MAX.
 DRAWING CONFORMS TO JEDEC MO-220.
 DRAWING IS NOT TO SCALE.



# **CARRIER INFORMATION**



Part Number	Package Description	Quantity/ Reel	Quantity/ Tray	Quantity/ Tube	Reel Diameter	Carrier Tape Width	Carrier Tape Pitch
MP5496GR- 0001–Z	QFN-28 (4mmx4mm)	5000	N/A	N/A	13in	12mm	8mm



# **Revision History**

Revision #	Revision Date	Description	Pages Updated
1.0	8/25/2020	Initial Release	-

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