

Features and Benefits

- 3.0 to 5.5 V logic supply range
- Schmitt trigger inputs for improved noise immunity
- Power-On Reset (POR)
- Up to 90 mA constant-current sinking outputs
- LED open circuit detection
- Low-power CMOS logic and latches
- High data input rate
- 20 ns typical staggering delay on the outputs
- Internal UVLO and thermal shutdown (TSD) circuitry

Packages:

28 pin QFN (suffix ET) 16 and 24 pin DIP (suffix A) 16 and 24 pin TSSOP (suffix LP) 16 and 24 pin SOIC (suffix LW)

Not to scale

Description

The A6278 and A6279 devices are specifically designed for LED display applications. Each of these BiCMOS devices includes a CMOS shift register, accompanying data latches, and NPN constant-current sink drivers. The A6278 contains 8 sink drivers, while there are 16 in the A6279.

The CMOS shift register and latches allow direct interfacing with microprocessor-based systems. With a 3.3 or 5 V logic supply, typical serial data-input rates can reach up to 25 MHz. The LED drive current is determined by the user's selection of a single resistor. A CMOS serial data output permits cascading between multiple devices in applications requiring additional drive lines. Open LED connections can be detected and signaled back to the host microprocessor through the SERIAL DATA OUT pin.

Four package styles are provided: a QFN surface mount, 0.90 mm overall height nominal (A6279 only); a DIP (type A) for through-hole applications; and for leaded surface-mount, an SOIC (type LW) and a TSSOP with exposed thermal pad (type LP). All package styles for the A6278 are electrically identical to each other, as are the A6279 package styles. All packages are lead (Pb) free, with 100% matte tin plated leadframes.

Functional Block Diagram

Selection Guide

1Variant is in production but has been determined to be LAST TIME BUY. This classification indicates that the variant is obsolete and notice has been given. Sale of the variant is currently restricted to existing customer applications. The variant should not be purchased for new design applications because of obsolescence in the near future. Samples are no longer available. Status date change November 1, 2010. Deadline for receipt of LAST TIME BUY orders is April 30, 2011. Recommended substitute: A6279.

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Absolute Maximum Ratings

Serial-Input, Constant-Current Latched LED Drivers with Open LED Detection

Terminal List Table

Serial-Input, Constant-Current Latched LED Drivers with Open LED Detection

OPERATING CHARACTERISTICS

Truth Table

L = Low logic (voltage) level

H = High logic (voltage) level

 $X = Don't care$

P = Present state

R = Previous state

n = 7 for the A6278, n = 15 for the A6279

Inputs and Outputs Equivalent Circuits

SERIAL DATA IN

Serial-Input, Constant-Current Latched LED Drivers with Open LED Detection

Functional Description

Normal Mode

Serial data present at the SERIAL DATA IN input is transferred to the shift register on the logic 0-to-logic 1 transition of the CLOCK input pulse. On succeeding CLOCK pulses, the register shifts data towards the SERIAL DATA OUT pin. The serial data must appear at the input prior to the rising edge of the CLOCK input waveform.

Data present in any register is transferred to the respective latch when the LATCH ENABLE input is high (serial-to-parallel conversion). The latches continue to accept new data as long as the LATCH ENABLE input is held high.

Applications where the latches are bypassed (LATCH ENABLE tied high) will require that the OUTPUT ENABLE input be high during serial data entry. When the OUTPUT ENABLE input is high, the output sink drivers are disabled (OFF).

The data stored in the latches is not affected by the OUTPUT ENABLE input. With the OUTPUT ENABLE input active (low), the outputs are controlled by the state of their respective latches.

LED Open Circuit Detection (Test) Mode

The LED Open Circuit Detection (OCD) mode, or *Test* mode, is entered by clocking in the LED OCD mode initialization sequence on the OUTPUT ENABLE (OE) and LATCH ENABLE (LE) pins. In Normal mode, the OE and LE pins do not change states while the CLOCK signal is cycling. The initialization sequence is shown in panel A of the LED OCD timing requirements diagram on page 7.

Note: Each step event during mode sequencing happens on the leading edge of the CLOCK signal. Five step events (CLOCK pulses) are required to enter OCD mode and five step events are required to return to Normal mode.

A pattern, such as all highs, should first be loaded into the registers and latched leaving LE low. The device is then sequenced into LED OCD mode. It should be noted that data is still being sent through the shift registers while entering the LED OCD mode. However, this data is not latched when the LE pin goes high and sees a CLOCK pulse during the initialization sequence. Open circuit detection does not take place until the sequence in Panel B on page 7 is performed. During this sequence, the OE pin must be held low for a minimum of 2 μs ($t_{W(OE1)}$) to ensure proper settling of the output currents and be given a minimum of three CLOCK pulses. During the period that the OE pin is low (active), OCD testing begins. The V_{CE} voltage on each of the output pins is compared to the Open LED Detection Theshold, $V_{CE(OCD)}$. If the V_{CE} of an enabled output is lower than $V_{CE(OCD)}$, an error bit value of 0 is set in the corresponding shift register. A value of 1 will be set if no error is detected. If a particular output is not enabled, a 0 will be set. The error codes are summarized in the following table:

Output State Test Condition Error Code Meaning

Output State	Test Condition	Error Code	Meaning
OFF	N/A		N/A
OΝ	V_{CE} < $V_{CE(OCD)}$	0	Open/TSD
	$V_{CE} \geq V_{CE(OCD)}$		Normal

After the testing process, setting the OE pin high causes the shift registers to latch the error code data where it can then be clocked out of the SERIAL DATA OUT pin. The OCD latching sequence (OE low, 3 CLOCK pulses, OE high as shown in panel B of the LED OCD timing diagram) can then be repeated if necessary to look for intermittent contact problems.

The state of the outputs can be programmed with new data at any time while in LED OCD mode (the same as in Normal mode). This allows specific patterns to be tested for open circuits. The pattern that is latched will then be tested during the OCD latching sequence and the resulting bit values can be clocked out of the SERIAL DATA OUT pin.

Note: LED Open Circuit Detection will not work properly if the current is being externally limited by resistors to within the set current limit for the device.

To return to Normal mode, perform the clocking sequence shown in panel C of the timing diagram on the OE and LE pins.

Constant Current (R_{EXT})

The A6278 and A6279 allow the user to set the magnitude of the constant current to the LEDs. Once set, the current remains constant regardless of the LED voltage variation, the supply voltage variation, or other circuit parameters that could otherwise affect LED current. The output current is determined by the value of an external current-control resistor (R_{EXT}) . The relationship of these parameters is shown in figure 1. Typical characteristics for output current and V_{CE} are shown in figure 2 for common values of R_{EXT} .

Figure 2. Output Current versus Device Voltage Drop $T_A = 25$ °C

Undervoltage Lockout

The A6278 and A6279 include an internal under-voltage lockout (UVLO) circuit that disables the outputs in the event that the logic supply voltage drops below a minimum acceptable level. This feature prevents the display of erroneous information, a necessary function for some critical applications.

Upon recovery of the logic supply voltage after a UVLO event, and on power-up, all internal shift registers and latches are set to 0. The A6278/A6279 is then in Normal mode.

Output Staggering Delay

The A6278/A6279 has a 20 ns delay between each output. The staggering of the outputs reduces the in-rush of currents onto the power and ground planes. This aids in power supply decoupling and EMI/EMC reduction.

The output staggering delay occurs under the following conditions:

- OUTPUT ENABLE is pulled low
- ï OUTPUT ENABLE is held low and LATCH ENABLE is pulled high
- ï OUTPUT ENABLE is held low, LATCH ENABLE is held high, and CLOCK is pulled high

The 20 ns delays are cumulative across all the outputs. Under any of the above conditions, the state of OUT0 gets set after a typical propagation delay, $t_{P(OF)}$. OUT1 will get set 20 ns after OUT0, and so forth. In the A6279, OUT15 will get set after 300 ns (15 \times 20 ns) plus $t_{P(OE)}$.

Note: The maximum CLOCK frequency is reduced in applications where both the OUTPUT ENABLE pin is held low and the LATCH ENABLE pin is held high continuously, and the outputs change state on the CLOCK edges. The staggering delay could cause spurious output responses at CLOCK speeds greater than 1 MHz.

Thermal Shutdown

When the junction temperature of the A6278/A6279 reaches the thermal shutdown temperature threshold, T_{ITSD} (165 \degree C typical), the outputs are shut off until the junction temperature cools down below the recovery threshold, $T_{JTSD} - T_{JTSDhys}$ (15°C typical). The shift register and output latches will remain active during a TSD event. Therefore, there is no need to reset the data in the output latches.

In LED OCD mode, if the junction temperature reaches the Thermal Shut Down threshold, the outputs will turn off, as in Normal mode operation. However, all of the shift registers will be set with 0, the error bit value.

Application Information

Load Supply Voltage (V_{LED})

These devices are designed to operate with driver voltage drops (V_{CE}) of 0.7 to 3V, with an LED forward voltage, V_{F} , of 1.2 to 4.0 V. If higher voltages are dropped across the driver, package power dissipation will increase significantly. To minimize package power dissipation, it is recommended to use the lowest possible load supply voltage, V_{LED} , or to set any series voltage dropping, V_{DROP} , according to the following formula:

$$
V_{DROP} = V_{LED} - V_F - V_{CE},
$$

with $V_{DROP} = I_0 \times R_{DROP}$ for a single driver or for a Zener diode (V_Z) , or for a series string of diodes (approximately 0.7 V per diode) for a group of drivers (see figure 3). If the available voltage source, V_{LED} , will cause unacceptable power dissipation and series resistors or diodes are undesirable, a voltage regulator can be used to provide supply voltages.

For reference, typical LED forward voltages are:

Pattern Layout

This device has a common logic ground and power ground terminal, GND. For the LP package, the GND pin should be tied to the exposed metal pad, EP, allowing the ground plane copper to be used to dissipate heat. If the ground pattern layout contains large common mode resistance, and the voltage between the system ground and the LATCH ENABLE, OUTPUT ENABLE, or CLOCK terminals exceeds 2.5 V (because of switching noise), these devices may not work properly.

Package Power Dissipation (P_D)

The maximum allowable package power dissipation based on package type is determined by:

$$
P_{D(max)} = (150 - T_A) / R_{\theta JA},
$$

where R_{0JA} is the thermal resistance of the package, determined experimentally. Power dissipation levels based on the package are shown in the Package Thermal Characteristics section (see page 14).

The actual package power dissipation is determined by:

$$
P_{D(act)} = DC \times (V_{CE} \times I_0 \times 16) + (V_{DD} \times I_{DD}),
$$

where DC is the duty cycle. The value *16* represents the maximum number of available device outputs for the A6279, used for the worst-case scenario (displaying all 16 LEDs; this would be *8* for the A6278).

When the load suppy voltage, V_{LED} , is greater than 3 to 5 V, and $P_{D(act)} > P_{D(max)}$, an external voltage reducer (V_{DROP}) must be used (see figure 3).

Reducing the percent duty cycle, DC, will also reduce power dissipation. Typical results are shown on the following pages.

Figure 3. Typical appplications for voltage drops

Serial-Input, Constant-Current Latched LED Drivers with Open LED Detection

Allowable Output Current versus Duty Cycle, A6278

 $V_{DD} = 5 V$

Serial-Input, Constant-Current Latched LED Drivers with Open LED Detection

Allowable Output Current versus Duty Cycle, A6279

 $V_{DD} = 5 V$

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Package Thermal Characteristics

*Additional thermal information is available on the Allegro Web site.

 $0.25 \begin{array}{c} +0.10 \\ -0.05 \end{array}$

7.62

Package A, 16-pin DIP (A6278)

Package A, 24-pin DIP (A6279)

can improve thermal dissipation (reference EIA/JEDEC Standard JESD51-5)

Package LP, 16-pin TSSOP with Exposed Thermal Pad (A6278)

Package LP, 24-pin TSSOP with Exposed Thermal Pad (A6279)

PCB, thermal vias at the exposed thermal pad land can improve thermal dissipation (reference EIA/JEDEC Standard JESD51-5)

Package LW, 16-pin SOIC (A6278)

Package LW, 24-pin SOIC (A6279)

For Reference Only (Reference JEDEC MS-013 AD) Dimensions exclusive of mold flash, gate burrs, and dambar protrusions Exact case and lead configuration at supplier discretion within limits shown

 \underline{A} Terminal #1 mark area

B Reference pad layout (reference IPC SOIC127P1030X265-24M) All pads a minimum of 0.20 mm from all adjacent pads; adjust as necessary to meet application process requirements and PCB layout tolerances

Package ET, 28-pin MLPQ (A6279)

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