

General Description

The MAX4200-MAX4205 are ultra-high-speed, openloop buffers featuring high slew rate, high output current, low noise, and excellent capacitive-load-driving capability. The MAX4200/MAX4201/MAX4202 are single buffers, while the MAX4203/MAX4204/MAX4205 are dual buffers. The MAX4201/MAX4204 have integrated 50Ω termination resistors, making them ideal for driving 50Ω transmission lines. The MAX4202/MAX4205 include 75Ω back-termination resistors for driving 75Ω transmission lines. The MAX4200/MAX4203 have no internal termination resistors.

The MAX4200-MAX4205 use a proprietary architecture to achieve up to 780MHz -3dB bandwidth, 280MHz 0.1dB gain flatness, 4200V/µs slew rate, and ±90mA output current drive capability. They operate from ±5V supplies and draw only 2.2mA of quiescent current. These features, along with low-noise performance, make these buffers suitable for driving high-speed analog-todigital converter (ADC) inputs or for data-communications applications.

Applications

High-Speed DAC Buffers

Wireless LANs

Digital-Transmission Line Drivers

High-Speed ADC Input Buffers

IF/Communications Systems

Selector Guide

PART	NO. OF BUFFERS	$\begin{array}{c} \text{INTERNAL} \\ \text{OUTPUT} \\ \text{TERMINATION} \\ (\Omega) \end{array}$	PIN-PACKAGE
MAX4200	1	_	8 SO, 5 SOT23
MAX4201	1	50	8 SO, 5 SOT23
MAX4202	1	75	8 SO, 5 SOT23
MAX4203	2	_	8 SO/µMAX
MAX4204	2	50	8 SO/µMAX
MAX4205	2	75	8 SO/µMAX

Pin Configurations appear at end of data sheet.

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Features

- ♦ 2.2mA Supply Current
- ♦ High Speed

780MHz -3dB Bandwidth (MAX4201/MAX4202) 280MHz 0.1dB Gain Flatness (MAX4201/MAX4202) 4200V/µs Slew Rate

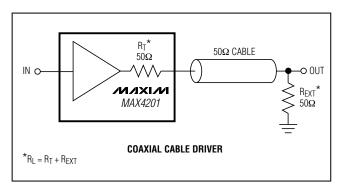
- ♦ Low 2.1nV/√Hz Voltage-Noise Density
- ♦ Low 0.8pA/√Hz Current-Noise Density
- ♦ High ±90mA Output Drive (MAX4200/MAX4203)
- **♦** Excellent Capacitive-Load-Driving Capability
- ♦ Available in Space-Saving SOT23 or µMAX® **Packages**

Ordering Information

PART	PIN-PACKAGE	TOP MARK	PKG CODE
MAX4200ESA	8 SO	_	S8-2
MAX4200EUK-T	5 SOT23-5	AABZ	U5-1
MAX4201ESA	8 SO	_	S8-2
MAX4201EUK-T	5 SOT23-5	ABAA	U5-1
MAX4202ESA	8 SO	_	S8-2
MAX4202EUK-T	5 SOT23-5	ABAB	U5-1
MAX4203ESA	8 SO	_	S8-2
MAX4203EUA-T	8 μMAX-8	_	U8-1
MAX4204ESA	8 SO	_	S8-2
MAX4204EUA-T	8 μMAX-8	_	U8-1
MAX4205ESA	8 SO	_	S8-2
MAX4205EUA-T	8 μMAX-8	_	U8-1

Note: All devices are specified over the -40°C to +85°C operating temperature range.

Typical Application Circuit



Maxim Integrated Products 1

ABSOLUTE MAXIMUM RATINGS

Supply Voltage (V _{CC} to V _{EE})+12V	Оре
Voltage on Any Pin to GND(VEE - 0.3V) to (VCC + 0.3V)	Stor
Output Short-Circuit Duration to GNDContinuous	Lea
Continuous Power Dissipation (T _A = +70°C)	
5-Pin SOT23 (derate 7.1mW/°C above +70°C)571mW	
8-Pin µMAX (derate 4.1mW/°C above +70°C)330mW	
8-Pin SO (derate 5.9mW/°C above +70°C)471mW	

Operating Temperature Range	40°C to +85°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

DC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = +5V, V_{EE} = -5V, R_L = \infty, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } T_A = +25^{\circ}C.)$

PARAMETER	SYMBOL		CON	DITIONS	MIN	TYP	MAX	UNITS	
Operating Supply Voltage	Vs	Guarantee	±4		±5.5	V			
Quiescent Supply Current	Is	Per buffer,	V _{IN} = 0V			2.2	4	mA	
Input Offset Voltage	Vos	$V_{IN} = 0V$				1	15	mV	
Input Offset Voltage Drift	TCVos	VIN = 0V				20		μV/°C	
Input Offset Voltage Matching		MAX4203/	MAX4204/MA	XX4205		0.4		mV	
Input Bias Current	IB					0.8	10	μΑ	
Input Resistance	RIN	(Note 1)				500		kΩ	
		-3.0V ≤	MAX4200/	MAX4203, $R_{EXT} = 150\Omega$	0.9	0.96	1.1		
Voltage Gain	Av	VOUT ≤ 3.0V	MAX4201,	MAX4204, $R_{EXT} = 50\Omega$	0.42	0.50	0.58	V/V	
			MAX4202/	MAX4205, $R_{EXT} = 75\Omega$	0.41	0.50	0.59		
Power-Supply Rejection	PSR	$V_S = \pm 4V \text{ to } \pm 5.5V$		55	72		dB		
		f = DC		MAX4200/MAX4203		8			
Output Resistance	Rout			MAX4201/MAX4204		50		Ω	
				MAX4202/MAX4205		75			
	lout	$R_L = 30\Omega$		MAX4200/MAX4203		±90			
Output Current				MAX4201/MAX4204		±52		mA	
				MAX4202/MAX4205		±44			
01 10: 10 1		Sinking or sourcing		MAX4200/MAX4203		150			
Short-Circuit Output Current	Isc			MAX4201/MAX4204		90		mA	
Current				MAX4202/MAX4205		75			
	Vout	MAX4200/MAX4203		$R_L = 150\Omega$	±3.3	±3.8		V	
				$R_L = 100\Omega$	±3.2	±3.7			
Output-Voltage Swing				$R_L = 37.5\Omega$		±3.3			
		MAX4201/MAX4204		$R_L = 50\Omega$	±1.9	±2.1			
		MAX4202/MAX4205		$R_L = 75\Omega$	±2.0	±2.3	·		

AC ELECTRICAL CHARACTERISTICS

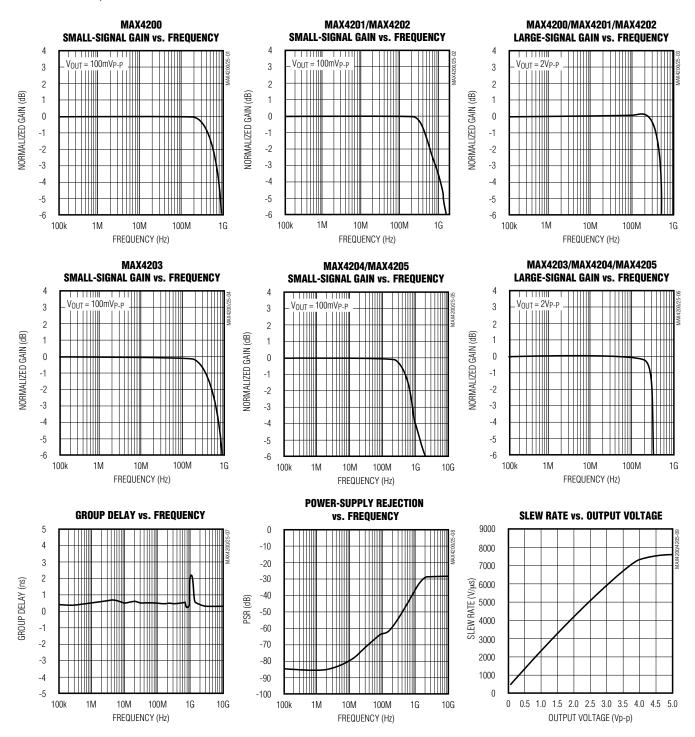
 $(V_{CC}=+5V,\ V_{EE}=-5V,\ R_L=100\Omega$ for MAX4200/MAX4201/MAX4203/MAX4204, $R_L=150\Omega$ for MAX4202/MAX4205, $T_A=T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A=+25^{\circ}C$.)

MAX4203 530 MAX4204/MAX4205 720 MAX4204/MAX4205 720 MAX4204/MAX4205 220 MAX4201/MAX4202 280 MAX4203 130 MAX4203 130 MAX4203 130 MAX4203 MAX	PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS				
Second harmonic Distortion SFDR Harmonic Distortion HD Harmonic Distortion Differential Gain Error DG NTSC, RL = 150Ω Laptor (Range) Laptor (Ra						MAX4200			660			
MAX4203 S30 MAX4204 MAX4205 720 MAX4200 220 MAX4201 MAX4202 280 MAX4203	-3dB Bandwidth	BW _(-3dB)	V _{OUT} ≤ 100mV _{RMS}		MAX4201/MAX4202			780		NALI-		
MAX4200 MAX4201 MAX4202 MAX4201 MAX4202 MAX4203 MA					MAX4203			530		1 IVIHZ		
0.1dB Bandwidth BW(0.1dB) VOUT ≤ 100mVRMS MAX4201/MAX4202 280 MAX4203 130 MAX4203 MAX4204/MAX4205 230 MAX4204/MAX4205 230 MAX4204/MAX4205 230 MAX4204/MAX4205 310 MAX4203/MAX4204/MAX4205 310 MAX4203/MAX4204/MAX4205 310 MAX4203/MAX4204/MAX4205 310 MAX4203/MAX4204/MAX4205 310 MAX4203/MAX4204/MAX4205 MAX4203/MAX4204/MAX4205 Ps 12 ns MAX4203/MAX4204 f = 50MHz -45						MAX4204/MAX4205			720			
MAX4203 130 MHz			VOLIT ~ 100m\/pwo		MAX4200			220				
MAX4203 130 MAX4204/MAX4205 230 MAX4204/MAX4205 230 MAX4204/MAX4202 490 MAX4204/MAX4205 310 MAX4204/MAX4205 MAX4204/MAX4205 MAX4204/MAX4205 MAX4204/MAX4205 MAX4204/MAX4205 MAX4204/MAX4204 MAX4204/MAX4202 MAX4204/MAX4204 MAX4204/MAX4205 MAX42	0.1dB Bandwidth	DW(o 4 ID)			MAX4201/MAX4202			280		MHz		
Full-Power Bandwidth FPBW VOUT ≤ 2VP.P MAX4200/MAX4201 / MAX4202 490 MHz	o. rab banawam	DVV(0.10B)	V0013100	AOOL & LOOHINKWR		MAX420)3				1011 12	
FPBW VOUT ≤ 2VP-P MAX4203/MAX4204/MAX4205 310 MHz												
MAX4203/MAX4204 MAX4205 MAX4204 MAX4205 MAX4206 MAX42	Full-Power Bandwidth	FPRW	VOLIT < 2VE	MAX							MHz	
Settling Time to 0.1% ts Vout = 2V step 12 ns		11 BW	MAX		4203/MAX	(4204/MAX4205		310		IVIDZ		
Settling Time to 0.1% ts Vout = 2V step 12 ns		SR	V _{OUT} = 2V	step					4200		V/µs	
Spurious-Free Dynamic Range SFDR SFDR Spurious-Free Dynamic Range SFDR Spurious-Free Dynamic Range SFDR											ps	
Second harmonic Distortion SFDR SFDR SPDR	Settling Time to 0.1%	ts	V _{OUT} = 2V	step							ns	
		SFDR		MAX4	200/N	1AX4201/						
Range SFDR 2VP-P MAX4203/MAX4204/ f = 5MHz -47 dBc f = 20MHz f = 100MHz -44 f = 100MHz -32 max4200/Max4201/ Max4202, f = 500kHz, Vour = 2VP-P Total harmonic -72 max4203/Max4204/ Max4203/Max4201/ Total harmonic				MAX4202 MAX4203/MAX4		, 0 (120 17					dBc	
MAX4203/MAX4204 f = 20MHz												
$ \text{MAX4205} \qquad \frac{\text{f = 20MHz}}{\text{f = 100MHz}} \qquad \frac{-44}{\text{f = 100MHz}} $	Hange					1AX4204/						
Harmonic Distortion												
Harmonic Distortion $ \begin{array}{c ccccccccccccccccccccccccccccccccccc$												
Harmonic Distortion $ HD = \begin{array}{c ccccccccccccccccccccccccccccccccccc$		HD	MAX4202, f = 500kHz, VOUT = 2VP-P MAX4203/MAX4204/I MAX4205, f = 500kHz, VOUT = 2VP-P								dBc	
Harmonic Distortion HD Total harmonic 4-8 dBc MAX4203/MAX4204/I MAX4205, f = 500kHz, Vout = 2VP-P Second harmonic -83 Third harmonic -47 Total harmonic -47					KHz,							
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	Harmonic Distortion											
VOUT = 2VP-P												
Differential Gain Error DG NTSC, $R_L = 150\Omega$ 1.3 % Differential Phase Error DP NTSC, $R_L = 150\Omega$ 0.15 degrees Input Voltage-Noise Density e_n $f = 1MHz$ 2.1 nV/\sqrt{Hz} Input Current-Noise Density i_n $f = 1MHz$ 0.8 pA/\sqrt{Hz} Input Capacitance C_{IN} 2 pF Output Impedance Z_{OUT} $f = 10MHz$ 6 Ω Amplifier Crosstalk Z_{TALK} $V_{OUT} = 2V_{PR}$								-47		_		
Differential Phase Error DP NTSC, $R_L = 150\Omega$ 0.15 degrees Input Voltage-Noise Density en $f = 1MHz$ 2.1 nV/\sqrt{Hz} Input Current-Noise Density in $f = 1MHz$ 0.8 pA/\sqrt{Hz} Input Capacitance C_{IN} 2 pF Output Impedance Z_{OUT} $f = 10MHz$ 6 Ω Amplifier Crosstalk X_{TALK} $V_{OUT} = 2V_{PR}$ $f = 10MHz$ -87 dB					Total harmonic							
Input Voltage-Noise Density $e_{\rm n}$ $f=1 \text{MHz}$ 2.1 nV/MHz Input Current-Noise Density $e_{\rm n}$ $f=1 \text{MHz}$ 0.8 pA/MHz Input Capacitance $e_{\rm n}$		DG								%		
Input Current-Noise Density in $f = 1MHz$ 0.8 pA/VHz Input Capacitance C_{IN} 2 pF Output Impedance Z_{OUT} $f = 10MHz$ 6 Ω Amplifier Crosstalk Z_{OUT} $Q_{OUT} = 2V_{PP}$ $Q_{OUT} = 2V_{PP}$ $Q_{OUT} = 2V_{PP}$		DP	NTSC, $R_L = 150\Omega$					0.15		degrees		
Input Capacitance C_{IN} 2 pF Output Impedance Z_{OUT} $f = 10MHz$ 6 Ω Amplifier Crosstalk Z_{OUT} $Q_{OUT} = 2V_{P}$ Q_{OUT} $Q_{OUT} = 2V_{P}$ Q_{OUT} $Q_{OUT} = 2V_{P}$ Q_{OUT}	Input Voltage-Noise Density	en	f = 1MHz				2.1		nV/√Hz			
Output Impedance Z_{OUT} $f = 10MHz$ 6 Ω Amplifier Crosstalk Z_{OUT} $Q_{OUT} = 2V_{PP}$ Q_{OUT	Input Current-Noise Density	in	f = 1MHz			0.8		pA/√Hz				
Amplifier Crosstalk XTALK VOLIT = 2VP P	Input Capacitance	CIN					2		pF			
Amplifier Crosstalk XTALK VOLIT = 2VPP HONOR Amplifier Crosstalk XTALK VOLIT = 2VPP HONOR Amplifier Crosstalk XTALK VOLIT = 2VPP HONOR Amplifier Crosstalk XTALK VOLIT = 2VPP HONOR Amplifier Crosstalk XTALK VOLIT = 2VPP HONOR Amplifier Crosstalk XTALK VOLIT = 2VPP HONOR Amplifier Crosstalk XTALK VOLIT = 2VPP HONOR Amplifier Crosstalk XTALK VOLIT = 2VPP HONOR Amplifier Crosstalk XTALK VOLIT = 2VPP HONOR Amplifier Crosstalk XTALK VOLIT = 2VPP HONOR Amplifier Crosstalk XTALK VOLIT = 2VPP Amplifier Crosstalk XTALK XTALK VOLIT = 2VPP HONOR Amplifier Crosstalk XTALK	Output Impedance	Zout	f = 10MHz						6		Ω	
f = 100MHz	Amplifier Crosstalk	Хтлік	V _{OUT} = 2V _{P-P}		f = 10MHz			-87		dB		
	7 mpmor orosotant	MIALIN				f = 100MHz			-65			

Note 1: Tested with no load; increasing load will decrease input impedance.

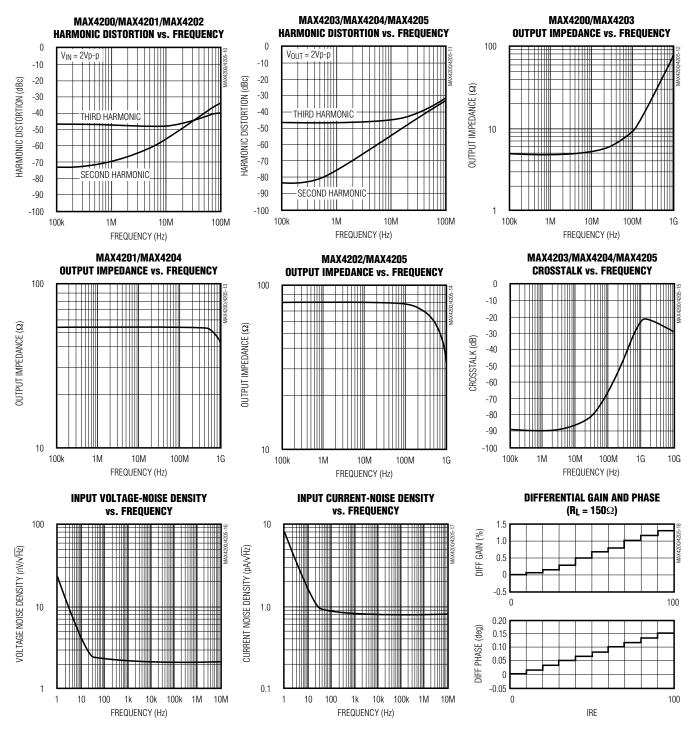
Typical Operating Characteristics

 $(V_{CC} = +5V, V_{EE} = -5V, R_L = 100\Omega$ for MAX4200/MAX4201/MAX4203/MAX4204, $R_L = 150\Omega$ for MAX4202/MAX4205, unless otherwise noted.)



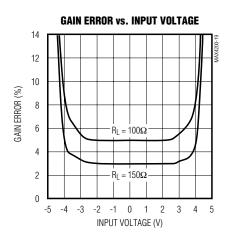
Typical Operating Characteristics (continued)

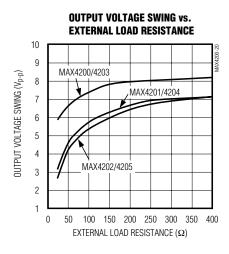
 $(V_{CC}=+5V,\,V_{EE}=-5V,\,R_L=100\Omega$ for MAX4200/MAX4201/MAX4203/MAX4204, $R_L=150\Omega$ for MAX4202/MAX4205, unless otherwise noted.)

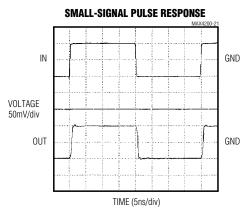


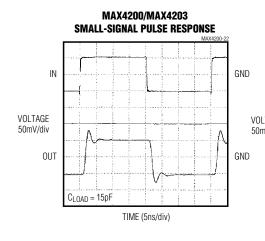
Typical Operating Characteristics (continued)

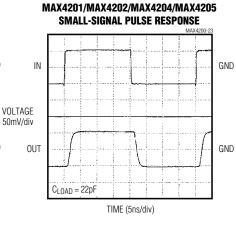
 $(V_{CC}=+5V,\,V_{EE}=-5V,\,R_L=100\Omega$ for MAX4200/MAX4201/MAX4203/MAX4204, $R_L=150\Omega$ for MAX4202/MAX4205, unless otherwise noted.)

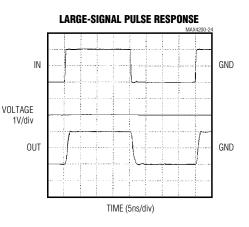






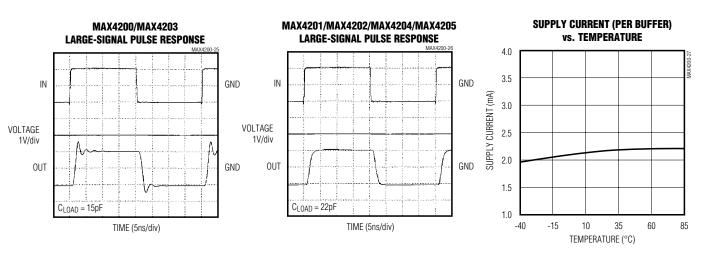


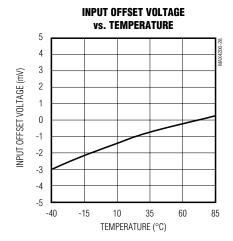


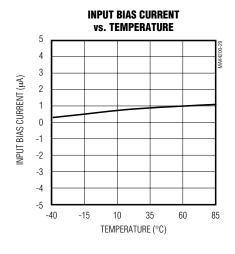


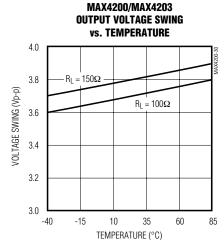
Typical Operating Characteristics (continued)

 $(V_{CC} = +5V, V_{EE} = -5V, R_L = 100\Omega$ for MAX4200/MAX4201/MAX4203/MAX4204, $R_L = 150\Omega$ for MAX4202/MAX4205, unless otherwise noted.)









Pin Description

	PIN					
MAX4200/MAX4201/MAX4202		MAX4203 MAX4204	NAME	FUNCTION		
SOT23-5	so	MAX4205				
		SO/μMAX				
1	1, 2, 5, 8	_	N.C.	No Connection. Not Internally Connected		
3	3	_	IN	Buffer Input		
_	_	1	IN1	Buffer 1 Input		
_	_	2	OUT1	Buffer 1 Output		
2	4	_	VEE	Negative Power Supply		
_	_	3	V _{EE1}	Negative Power Supply for Buffer 1		
_	_	4	V _{EE2}	Negative Power Supply for Buffer 2		
_	_	5	IN2	Buffer 2 Input		
_	_	6	OUT2	Buffer 2 Output		
5	6	_	OUT	Buffer Output		
4	7	_	Vcc	Positive Power Supply		
_	_	7	VCC2	Positive Power Supply for Buffer 2		
_	_	8	V _{CC1}	Positive Power Supply for Buffer 1		

Detailed Description

The MAX4200–MAX4205 wide-band, open-loop buffers feature high slew rates, high output current, low 2.1nV $\sqrt{\text{Hz}}$ voltage-noise density, and excellent capacitive-load-driving capability. The MAX4200/MAX4203 are single/dual buffers with up to 660MHz bandwidth, 230MHz 0.1dB gain flatness, and a 4200V/µs slew rate. The MAX4201/MAX4204 single/dual buffers with integrated 50 Ω output termination resistors, up to 780MHz bandwidth, 280MHz gain flatness, and a 4200V/µs slew rate, are ideally suited for driving high-speed signals over 50 Ω cables. The MAX4202/MAX4205 provide bandwidths up to 720MHz, 230MHz gain flatness, 4200V/µs slew rate, and integrated 75 Ω output termination resistors for driving 75 Ω cables.

With an open-loop gain that is slightly less than +1V/V, these devices do not have to be compensated with the internal dominant pole (and its associated phase shift) that is present in voltage-feedback devices. This feature allows the MAX4200–MAX4205 to achieve a nearly constant group delay time of 405ps over their full frequency range, making them well suited for a variety of RF and IF signal-processing applications.

These buffers operate with ±5V supplies and consume only 2.2mA of quiescent supply current per buffer while providing up to ±90mA of output current drive capability.

Applications Information

Power Supplies

The MAX4200–MAX4205 operate with dual supplies from ±4V to ±5.5V. Both VCC and VEE should be bypassed to the ground plane with a 0.1µF capacitor located as close to the device pin as possible.

Layout Techniques

Maxim recommends using microstrip and stripline techniques to obtain full bandwidth. To ensure that the PC board does not degrade the amplifier's performance, design it for a frequency greater than 6GHz. Pay careful attention to inputs and outputs to avoid large parasitic capacitance. Whether or not you use a constant-impedance board, observe the following guidelines when designing the board:

- Do not use wire-wrap boards, because they are too inductive.
- Do not use IC sockets, because they increase parasitic capacitance and inductance.

- Use surface-mount instead of through-hole components for better high-frequency performance.
- Use a PC board with at least two layers; it should be as free from voids as possible.
- Keep signal lines as short and as straight as possible. Do not make 90° turns; round all corners.

Input Impedance

The MAX4200–MAX4205 input impedance looks like a 500k Ω resistor in parallel with a 2pF capacitor. Since these devices operate without negative feedback, there is no loop gain to transform the input impedance upward, as in closed-loop buffers. As a consequence, the input impedance is directly related to the output impedance. If the output load impedance decreases, the input impedance also decreases. Inductive input sources (such as an unterminated cable) may react with the input capacitance and produce some peaking in the buffer's frequency response. This effect can usually be minimized by using a properly terminated transmission line at the buffer input, as shown in Figure 1.

Output Current and Gain Sensitivity

The absence of negative feedback means that open-loop buffers have no loop gain to reduce their effective output impedance. As a result, open-loop devices usually suffer from decreasing gain as the output current is decreased. The MAX4200–MAX4205 include local feedback around the buffer's class-AB output stage to ensure low output impedance and reduce gain sensitivity to load variations. This feedback also produces demand-driven current bias to the output transistors for ±90mA (MAX4200/MAX4203) drive capability that is relatively independent of the output voltage (see *Typical Operating Characteristics*).

Output Capacitive Loading and Stability

The MAX4200–MAX4205 provide maximum AC performance with no load capacitance. This is the case when the load is a properly terminated transmission line. However, these devices are designed to drive any load capacitance without oscillating, but with reduced AC performance.

Since the MAX4200–MAX4205 operate in an open-loop configuration, there is no negative feedback to be transformed into positive feedback through phase shift introduced by a capacitive load. Therefore, these devices will not oscillate with capacitive loading, unlike similar buffers operating in a closed-loop configuration. However, a capacitive load reacting with the buffer's output impedance can still affect circuit performance. A capacitive load will form a lowpass filter with the buffer's output resistance, thereby limiting system

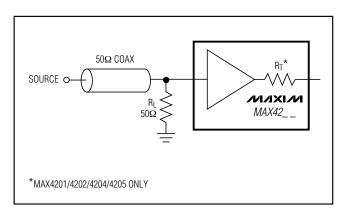


Figure 1. Using a Properly Terminated Input Source

bandwidth. With higher capacitive loads, bandwidth is dominated by the RC network formed by RT and C_L ; the bandwidth of the buffer itself is much higher. Also note that the isolation resistor forms a divider that decreases the voltage delivered to the load.

Another concern when driving capacitive loads results from the amplifier's output impedance, which looks inductive at high frequency. This inductance forms an L-C resonant circuit with the capacitive load and causes peaking in the buffer's frequency response.

Figure 2 shows the frequency response of the MAX4200/MAX4203 under different capacitive loads. To settle out some of the peaking, the output requires an isolation resistor like the one shown in Figure 3. Figure 4 is a plot of the MAX4200/MAX4203 frequency response with capacitive loading and a 10Ω isolation resistor. In many applications, the output termination resistors included in the MAX4201/MAX4202/ MAX4204/MAX4205 will serve this purpose, reducing component count and board space. Figure 5 shows the MAX4201/MAX4202/ MAX4204/MAX4205 frequency response with capacitive loads of 47pF, 68pF, and 120pF.

Coaxial Cable Drivers

Coaxial cable and other transmission lines are easily driven when properly terminated at both ends with their characteristic impedance. Driving back-terminated transmission lines essentially eliminates the line's capacitance. The MAX4201/MAX4204, with their integrated 50Ω output termination resistors, are ideal for driving 50Ω cables. The MAX4202/MAX4205 include integrated 75Ω termination resistors for driving 75Ω cables. Note that the output termination resistor forms a voltage divider with the load resistance, thereby decreasing the amplitude of the signal at the receiving end of the cable by one half (see the *Typical Application Circuit*).

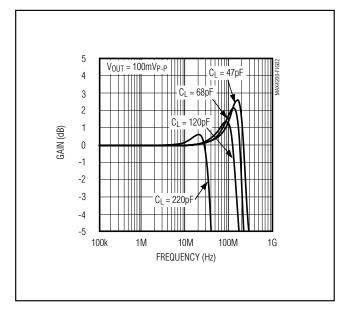


Figure 2. MAX4200/MAX4203 Small-Signal Gain vs. Frequency with Load Capacitance and No Isolation Resistor

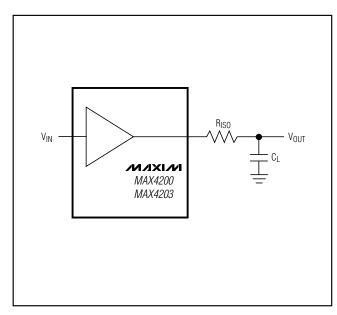


Figure 3. Driving a Capacitive Load Through an Isolation Resistor

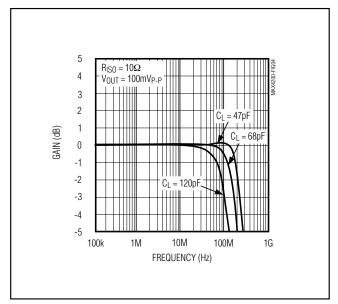


Figure 4. MAX4200/MAX4203 Small-Signal Gain vs. Frequency with Load Capacitance and 10Ω Isolation Resistor

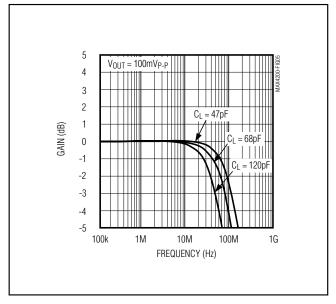
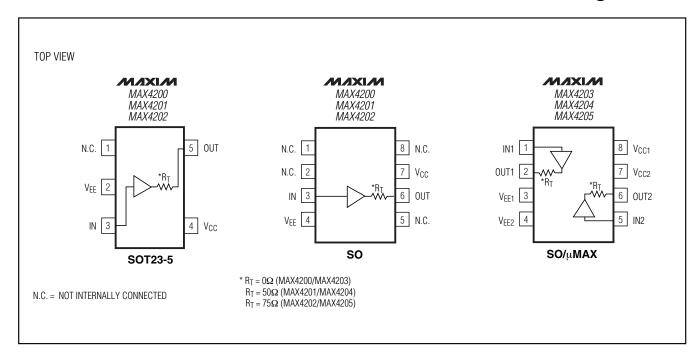


Figure 5. MAX4201/MAX4202/MAX4204/MAX4205 Small-Signal Gain vs. Frequency with Capacitive Load and No External Isolation Resistor

Pin Configurations



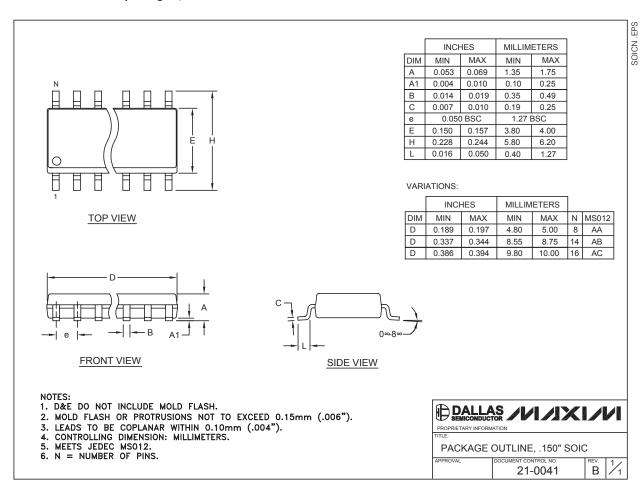
Chip Information

TRANSISTOR COUNTS:

MAX4200/MAX4201/MAX4202: 33 MAX4203/MAX4204/MAX4205: 67 SUBSTRATE CONNECTED TO VEE

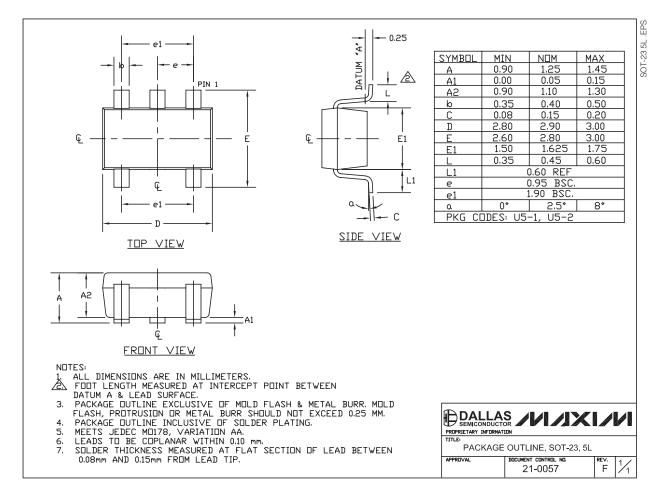
Package Information

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to www.maxim-ic.com/packages.)



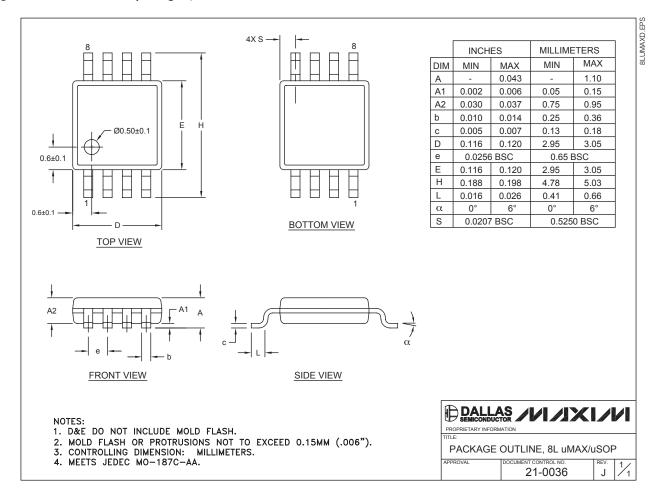
Package Information (continued)

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to www.maxim-ic.com/packages.)



Package Information (continued)

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to www.maxim-ic.com/packages.)



_Revision History

Pages changed at Rev 3: 1-5, 8, 10-14

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