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## **30V Li+ Linear Battery Charger** with GSM Test Mode in 3mm x 2mm TDFN

### **General Description**

The MAX8922L linear battery charger safely charges a single-cell lithium-ion (Li+) battery. Charging rate is optimized to accommodate the thermal characteristics of a given application. There is no need to reduce the maximum charge current at the worst-case charger power dissipation. Charging is optimized for a single Li+ cell using a control algorithm that includes low-battery precharging, voltage and current-limited fast charging, and top-off charging, while continuously monitoring for input overvoltage and device die-temperature conditions. The fast-charge current and top-off current thresholds are programmable by a simple single-pin serial interface. The charger status and valid input power are indicated by two open-drain outputs (CHG and POK).

The fast-charge current is defaulted to 400mA and programmable through the single-pin interface ( $\overline{\text{EN}}/\text{SET}$ ). The MAX8922L also can be programmable to GSM test mode through the single-pin interface.

The MAX8922L is available in a tiny (3mm x 2mm x 0.8mm) 10-pin TDFN package.

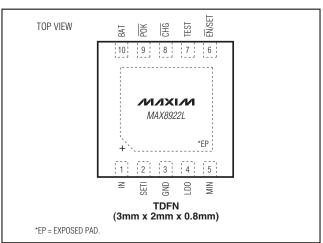
### **Applications**

GSM/EDGE/UMTS/CDMA Cell Phones

Digital Cameras

PDAs

Portable Media Players and MP3 Players Wireless Appliances



### M/XI/M

### **Pin Configuration**

### \_\_ Features

- Overvoltage-Protected 30V<sub>DC</sub> Rated Input (IN)
- Input Overvoltage-Protected Safe 4.94V LDO Output
- 2.3A GSM RF Test Mode
- No External FET, Blocking Diode, or Sense Resistor Required
- Single-Pin Easy Programmable Fast-Charge and GSM Test Mode (EN/SET)
- Resistor-Programmable Fast-Charge Current (SETI)
- Resistor-Programmable Top-Off Current Threshold (MIN)
- Prequalification Charge
- Power-OK Monitor Output (POK)
- Charging-Status Output (CHG)
- Die Temperature Regulation for Optimized Charge Rate
- Tiny (3mm x 2mm x 0.8mm) 10-Pin TDFN Package

### **\_Ordering Information**

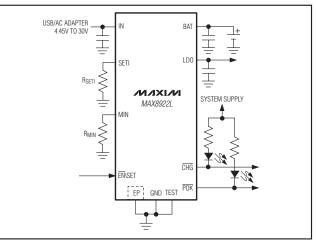
PART	PIN-PACKAGE	TOP MARK
MAX8922LETB+T	10 TDFN-EP*	AWN

+Denotes a lead(Pb)-free and RoHS-compliant package.

\*EP = Exposed pad.

**Note:** This device operates in the -40°C to +85°C extended operating temperature range.

### \_Typical Operating Circuit



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For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

### **ABSOLUTE MAXIMUM RATINGS**

IN to GND	0.3V to +30V
BAT, CHG, EN/SET, POK, SETI,	
MIN, LDO, TEST to GND	0.3V to +6V
IN to BAT Continuous Current	1A <sub>BMS</sub>
Continuous Power Dissipation ( $T_A = +70^{\circ}C$ )	
10-Pin (3mm x 2mm) TDFN	
(derate 14.9mW/°C above +70°C)	1188.7mW

Operating Temperature Range	40°C to +85°C
Junction Temperature Range	40°C to +150°C
Storage Temperature Range	
Lead Temperature (soldering, 10s)	+300°C
Soldering Temperature (reflow)	

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### **ELECTRICAL CHARACTERISTICS**

 $(V_{IN} = 5V, V_{BAT} = 4V, V_{\overline{EN}/SET} = 0V, T_A = -40^{\circ}C$  to  $+85^{\circ}C$ , typical values are at  $T_A = +25^{\circ}C$ , unless otherwise noted.) (Note 1)

PARAMETER		CONDITIONS	MIN	ТҮР	MAX	UNITS		
IN								
Input Voltage Range			0		28	V		
Input Voltage Operating Range	(Note 2)		4.45		7	V		
Input Undervoltage Threshold (UVLO)	V <sub>IN</sub> rising, 500mV hy	<sub>N</sub> rising, 500mV hysteresis (typ)		3.90	4.00	V		
Input Overvoltage Threshold (OVP)	V <sub>IN</sub> rising, 200mV hysteresis (typ)		7.2	7.5	7.8	V		
	I <sub>BAT</sub> = 0mA, charge i	mode		700	1300			
Input Supply Current	$V_{\overline{EN}/SET} = 5V$ , standb	by mode		250	440	μΑ		
	$V_{IN} = V_{BAT}$ , shutdow	n mode		200				
IN-to-BAT On-Resistance	$V_{IN} = 4.15V, V_{BAT} = 4V$			0.35		Ω		
IN to DAT Comparator Thrashold	V <sub>IN</sub> rising		120	250	500	mV		
IN-to-BAT Comparator Threshold	V <sub>IN</sub> falling			100		IIIV		
BAT								
BAT Regulation Voltage	I <sub>BAT</sub> = 100mA	$T_A = +25^{\circ}C$	4.179	4.2	4.221	V		
	$T_A = -40^{\circ}C \text{ to } +85^{\circ}C$		4.158	4.2	4.242	v		
Battery Removal Detection	V <sub>BAT</sub> rising			4.67		V		
Threshold	Hysteresis		0.2		V			
			365	400	435	-		
			460	500	540			
Charging Current	$\overline{\text{EN}}/\text{SET}$ = two pulses with low > 4ms, V <sub>BAT</sub> = 3.5V		80	90	100	mA		
	$\overline{\text{EN}/\text{SET}}$ = three pulses with low > 4ms, V <sub>BAT</sub> = 3.5V (Note 3)			2350				
Soft-Start Time	Ramp time to fast-charge current			250		μs		
BAT Precharge Threshold	V <sub>BAT</sub> rising, 300mV h	iysteresis (typ)		2.5		V		
Precharge Current				80		mA		
BAT Leakage Current	$V_{IN} = 0V, V_{BAT} = 4.2$	V		1	5	μA		

### **ELECTRICAL CHARACTERISTICS (continued)**

 $(V_{IN} = 5V, V_{BAT} = 4V, V_{\overline{EN}/SET} = 0V, T_A = -40^{\circ}C$  to +85°C, typical values are at  $T_A = +25^{\circ}C$ , unless otherwise noted.) (Note 1)

PARAMETER		CONDITION	S	MIN	ТҮР	MAX	UNITS
LDO	1						1
Minimum LDO Bypass Capacitance					1		μF
LDO Regulated Output Voltage	$I_{LDO} = 10 \text{mA}, V_{IN}$	= 5V		4.8	4.94		V
LDO Output-Current Limit					100		mA
EN/SET	·						
	Rising					1.4	V
Logic Input Thresholds	Falling			0.4			V
Program Lock Time				4			ms
Shutdown Delay	$V_{IN} = 5V, \overline{EN}/SET$	$I_{\rm IN} = 5V, \overline{\rm EN}/\rm SET$ from low to high		4			ms
tLOW	(Note 4)	Note 4)		100		1400	
tнідн	(Note 4)	Note 4)		100		1400	μs
Pulldown Resistor					2		MΩ
POK, CHG							
Logic Output Voltage, Low	$I_{\overline{POK}} = I_{\overline{CHG}} = 5m$	۱A			0.05	0.2	V
Logio Output Current High		V <u>POK</u> = V <u>CHG</u> = 5.5V, V <sub>IN</sub> = 0V			0.001	1	
Logic Output Current, High	VPOK = VCHG = 5.	5V, VIN = 0V	$T_A = +85^{\circ}C$		0.01		μA
CHG							
		Default top-off (typ) = 80mA	threshold, hysteresis	60	80	100	
Top-Off Threshold	IBAT falling, battery is charged	$\overline{\text{EN}}/\text{SET} = \text{one}$ 1.875k $\Omega$ , hyste	pulse, R <sub>MIN</sub> = eresis (typ) = 130mA	60	80	100	mA
	charged	$\overline{EN}/SET = two pulses, hysteresis (typ) = 22mA$	50	60	70		
Detection Delay	IBAT falls below to	p-off threshold		2	4	6	ms
THERMAL LOOP							
Thermal-Limit Temperature	Junction temperat reduced, T <sub>J</sub> rising		arge current is		+105		°C
Thermal-Limit Gain	Reduction of IBAT	for increase of T,	j, default mode		-28		mA/°C

Note 1: Limits are 100% production tested at  $T_A = +25$  °C. Limits over the operating temperature range are guaranteed by design and characterization.

**Note 2:** Guaranteed by undervoltage- and overvoltage-threshold testing. If  $V_{BAT} = 4.2V$ ,  $V_{IN}$  needs to be > 4.2V + 250mV (typ) to start normal operation. After the MAX8922L turns on, it can operate until  $V_{BAT}$  + 100mV (typ). For complete charging, the input voltage must be > 4.45V. See the *Input Sources* section.

Note 3: Used for factory GSM RF calibration. 217Hz, 12.5% current pulse,  $T_A = +25^{\circ}C$ . Not for continuous charge current.

**Note 4:** Not tested. Design guidance only.

**MAX8922L** 

**Typical Operating Characteristics** 

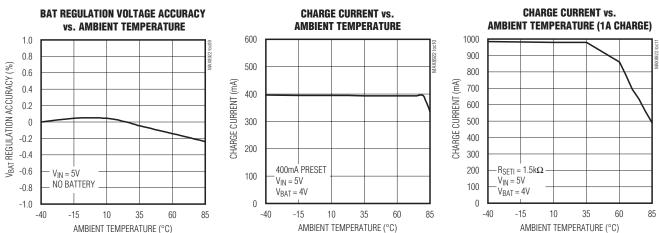
#### ( $V_{IN} = 5V$ , $V_{\overline{EN}/SET} = 0V$ . $V_{BAT} = 4V$ , MAX8922L Evaluation Kit. $T_A = +25^{\circ}C$ , unless otherwise noted.) **SUPPLY CURRENT** vs. **DISABLED SUPPLY CURRENT vs. CHARGE CURRENT vs. SUPPLY VOLTAGE BATTERY VOLTAGE** SUPPLY VOLTAGE 1.0 1.0 600 500mA SETI MODE 0.9 0.9 500 (mA) 400mA PRESET = 0.8 0.8 STANDBY SUPPLY CURRENT SUPPLY CURRENT (mA) CHARGE CURRENT (mA) 0.7 0.7 400 0.6 0.6 0.5 300 0.5 0.4 0.4 200 0.3 0.3 90mA PRESET 0.2 0.2 100 0.1 0.1 0 0 0 0 1 2 4 5 6 7 0 1 2 3 4 5 6 7 3 0 2 3 4 5 1 BATTERY VOLTAGE (V) SUPPLY VOLTAGE (V) SUPPLY VOLTAGE (V) **CHARGE CURRENT vs. CHARGE CURRENT WITH ONE CHARGE CURRENT WITH TWO** SUPPLY VOLTAGE **EN/SET PULSE EN/SET PULSES** 500 400mA DEFAULT 450 $V_{BAT} = 4V$ 5V/div 5V/div 400 0 VEN/SET 0 VFN/SFT CHARGE CURRENT (mA) 350 500mA 300 250 400mA 400mA 200 150 100 100mA/div 100mA/div 50 IBAT 0 IBAT 0 0 1ms/div 3 9 12 15 18 21 6 24 27 30 0 1ms/div SUPPLY VOLTAGE (V) **CHARGE CURRENT vs. GSM TRANSIENT RESPONSE INPUT VOLTAGE HEADROOM** 800 NO BATTERY C<sub>BAT</sub> = 68µF 700 200mv/div VBAT 600 CHARGE CURRENT (mA) AC-COUPLED 500 400 300 1A/div IBAT 200 100 $I_{FAST-CHARGE} = 677 mA$ 0A 0 1ms/div 0 50 100 150 200 300 250

V<sub>IN</sub> - V<sub>BAT</sub> (mV)

**MAX8922L** 

### **Typical Operating Characteristics (continued)**

( $V_{IN}$  = 5V,  $V_{\overline{EN}/SET}$  = 0V.  $V_{BAT}$  = 4V, MAX8922L Evaluation Kit.  $T_A$  = +25°C, unless otherwise noted.)



### **Pin Description**

PIN	NAME	FUNCTION
1	IN	DC Input Supply. Connect IN to $V_{IN} > 4V$ and $(V_{IN} - V_{BAT}) \ge 250 \text{mV}$ up to a 7V charging source. Bypass IN to GND with a 1µF or larger ceramic capacitor.
2 SETI		Charge-Current Program and Fast-Charge Current Monitor. Output current from SETI is 1000µA per ampere of battery-charging current. Set the charging current by connecting a resistor (R <sub>SETI</sub> in Figure 1) from SETI to GND. IFAST-CHARGE = 1500V/R <sub>SETI</sub> . <b>Connect to GND if pulse 1 mode (external SETI)</b> is not used.
3	GND	Ground
4	LDO	4.94V Regulated LDO Output with Input Overvoltage Protection. Bypass LDO to GND with a $1\mu$ F or larger ceramic capacitor. LDO can be used to supply low-voltage-rated USB systems.
5	MIN	Top-Off Current Threshold Programmable Input. $I_{MIN} = 150V/R_{MIN}$ . Connect to GND if pulse 1 mode (external SETI) is not used.
6	ĒN/SET	Active-Low Enable Input. EN/SET is used for programming fast-charge current and GSM test mode. For detailed descriptions, see the <i>Charger-Enable and Program Input</i> (EN/SET) section.
7	TEST	Factory Test Input. Connect to GND.
8	CHG	Charging-Status Output. CHG is internally pulled low when the charger is in prequalification or fast- charge mode. CHG is high impedance when the charger is in top-off or disabled.
9	POK	Input Power-OK Monitor. $\overline{\text{POK}}$ is an open-drain output that is internally pulled low when V <sub>IN</sub> is greater than V <sub>UVLO</sub> and lower than V <sub>OVP</sub> and V <sub>IN</sub> > V <sub>BAT</sub> + 250mV. $\overline{\text{POK}}$ is high impedance when V <sub>IN</sub> is less than V <sub>UVLO</sub> or greater than V <sub>OVP</sub> or V <sub>IN</sub> < V <sub>BAT</sub> + 100mV.
10	BAT	Battery Connection. The IC delivers charging current and monitors battery voltage using BAT. Bypass BAT to GND with a 2.2µF or larger ceramic capacitor. BAT is high impedance when the IC is disabled.
	EP	Exposed Pad. Connect to the GND plane for increased thermal dissipation.

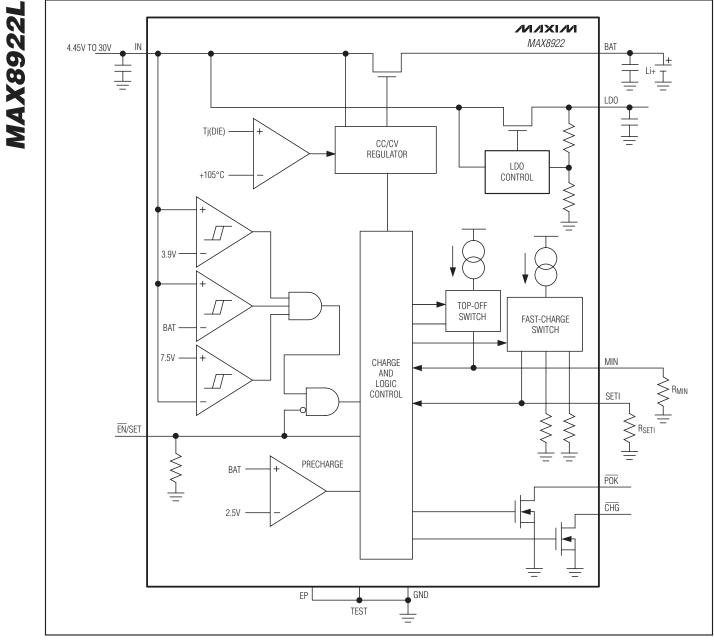


Figure 1. MAX8922L Functional Diagram

### **Detailed Description**

The MAX8922L is designed to charge a single-cell Li+ battery from a DC source voltage between 4.45V and 7V, while V<sub>IN</sub> can withstand up to 30V. The fast-charge current and top-off current thresholds are programmable with  $\overline{\text{EN}}$ /SET, SETI, and MIN.

### Charger-Enable and Progra<u>m I</u>nput (EN/SET)

EN/SET is an active-low logic input that enables the charger. Drive EN/SET high longer than 4ms to disable the charger-control circuitry. If EN/SET is left unconnected, an internal  $2M\Omega$  pulldown resistor enables 400mA fast-charge current by default. The pulse programming scheme shown in Table 1 and Figure 3 is used to program the charge current and GSM test mode. There are four different fast-charge current states. Default fast-charge current state is 400mA mode. More than three pulses are interpreted to 90mA mode. After programming is locked, the MAX8922L ignores pulses until the IC is disabled/enabled or input power is cycled. Each fast-charge state is locked after a 4ms logic-low is asserted on EN/SET, followed by programming pulses. However, during default mode, if EN/SET does not receive any pulses, the charger stays in default mode unlocked indefinitely.

#### **Debounce Timer**

To prevent the MAX8922L from charging the battery momentarily upon IN power-up with EN/SET held low, a 2ms (typ) debounce timer delays the charging loop upon power-up. If EN/SET is logic-low or unconnected (pulled down by an internal pulldown resistor) during IN power-up, the charger starts charging the battery 2ms after VUVLO <  $V_{IN}$  <  $V_{OVP}$  and  $V_{BAT}$  + 250mV <  $V_{IN}$ . If EN/SET is logic-high during IN power-up, the charger does not charge the battery.

### Soft-Start

To prevent input transients, the rate of change of the charge current is limited when the charger is turned on or changes its current compliance. It takes approximately 250µs (typ) (tSOFTSTART) for the charger to go from 0mA to the maximum fast-charge current.

#### **Thermal-Limit Control**

The MAX8922L features a thermal limit that reduces the charge current when the die temperature exceeds  $+105^{\circ}$ C. As the temperature increases above  $+105^{\circ}$ C, the IC decreases the charge current by 28mA/°C.

### Charge-Indicator Output (CHG)

CHG is an open-drain output that indicates charger status. CHG goes low during charging in prequalification or fast-charge mode. The CHG internal open-drain MOSFET turns off when the charge current reaches the top-off threshold. The CHG status is latched after the top-off threshold is reached. The latch can be reset as follows:

- Disable and re-enable the MAX8922L.
- Input power is cycled.
- Battery-charge current increases greater than the top-off threshold + hysteresis.

When the MAX8922L is used in conjunction with a microprocessor, connect a pullup resistor between CHG and the logic I/O voltage to indicate charge status to the  $\mu$ P. Alternatively, CHG can sink 5mA or more for an LED charge indicator.

CHARGE CURRENT	DEFAULT	NUMBER OF PULSES + > 4ms LOGIG-LOW	FAST-CHARGE CURRENT SETTING
		One	SETI, resistor programmable
$l_{\text{D}} = (m \Lambda)$	100 1	Тwo	90mA
I <sub>BAT</sub> (mA)	400mA	Three	2.3A (GSM test)
		Four and more	90mA

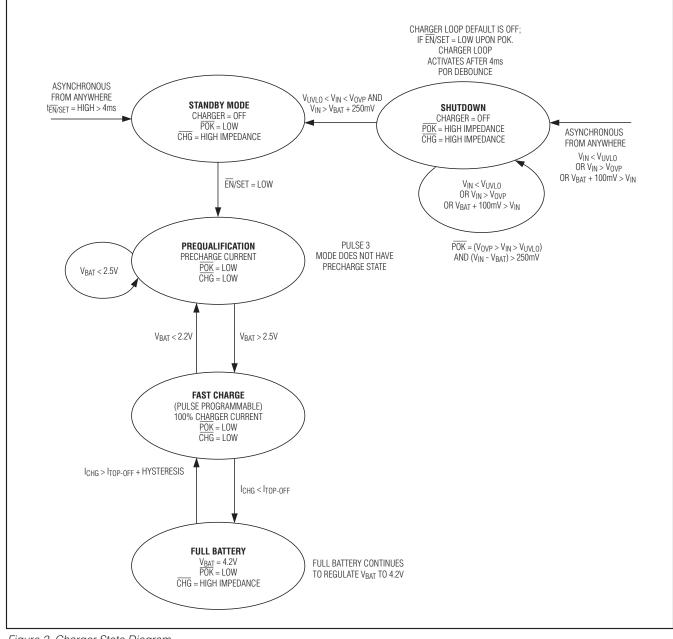


Figure 2. Charger State Diagram

### **Power-OK Indicator (POK)**

The MAX8922L contains an open-drain  $\overrightarrow{\text{POK}}$  output that goes low when V<sub>IN</sub> is greater than V<sub>UVLO</sub> and lower than V<sub>OVP</sub> and V<sub>IN</sub> exceeds the battery voltage by 250mV. Once charging has started, charging is sustained with

inputs as low as 3.5V, as long as the input voltage remains above the battery voltage by at least 100mV.  $\overrightarrow{POK}$  status should be maintained even though the charger is disabled by  $\overrightarrow{EN}/\overrightarrow{SET}$ . When  $V_{IN} > V_{OVP}$ ,  $\overrightarrow{POK}$  is high impedance.



**MAX8922L** 

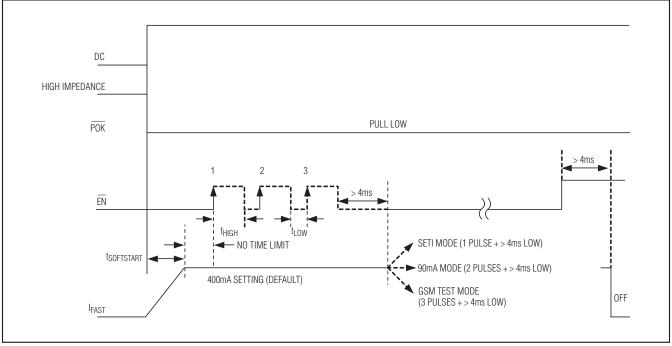


Figure 3. Charge-Current Programming

#### **LDO Output**

The LDO is preset to an output voltage of 4.94V and a 100mA current limit (typ). The LDO is powered from IN and has input overvoltage protection. The LDO is on if a valid input is present ( $V_{UVLO} < V_{IN} < V_{OVP}$ ).

Bypass LDO to GND with a  $1\mu F$  or larger ceramic capacitor. The LDO can be used to supply low-voltage-rated USB systems.

### **Applications Information**

#### Fast-Charge Current Settings

In pulse 1 mode, the maximum charging current is programmed by an external resistor connected from SETI to GND (RSETI). Calculate RSETI as follows:

R<sub>SETI</sub> = 1500V/I<sub>FAST-CHARGE</sub>

where IFAST-CHARGE is in amperes and RSETI is in ohms. SETI can be used to monitor the fast-charge current level in the one-pulse mode (R<sub>SETI</sub> mode). The output current from SETI is 1000 $\mu$ A per ampere of charging current.

The output voltage at SETI is proportional to the charging current (I<sub>CHARGE</sub>) when SETI mode is used for the fast-charge current:

VSETI = ICHARGE X RSETI/1000

The voltage at ISET is nominally 1.5V at the selected fast-charge current and decreases with charging current as the cell becomes fully charged or as the thermal-regulation circuitry activates.

#### Top-Off Current Settings

**MAX89221** 

The top-off charging current is programmed by an external resistor connected from MIN to GND ( $R_{MIN}$ ) in the one-pulse mode ( $R_{SETI}$  mode). Calculate  $R_{MIN}$  as follows:

#### RMIN = 150V/IMIN

where I<sub>MIN</sub> is in amperes and R<sub>MIN</sub> is in ohms.

#### **Capacitor Selection**

Connect a 2.2 $\mu$ F ceramic capacitor from BAT to GND for proper stability. Connect a 1 $\mu$ F ceramic capacitor from IN to GND. Use a larger input bypass capacitor for high charging currents to reduce supply noise. All capacitors should be X5R dielectric or better. Be aware that some capacitors have large-voltage coefficients, and should be avoided.



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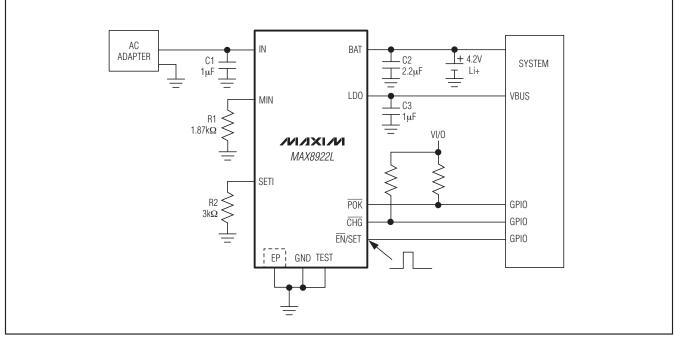


Figure 4. AC Adapter Charger Application

#### **Thermal Considerations**

The MAX8922L is in a thermally enhanced TDFN package with an exposed pad. Connect the exposed pad of the package to a large copper ground plane to provide a thermal contact between the device and the circuit board. The exposed pad transfers heat away from the device, allowing the IC to charge the battery with maximum current, while minimizing the increase in die temperature.

#### **Input Sources**

The MAX8922L operates from well-regulated DC sources. The charger input voltage range is 4.45V to 7V. The device survives input voltages up to 30V without damage to the IC. If the input voltage is greater than 7.5V (typ), the IC stops charging. An appropriate power supply must provide at least 4.2V plus the voltage drop across the internal-pass transistor when sourcing the desired peak charging current.

#### VIN(MIN) > 4.2V + IFAST-CHARGE(MAX) X RON

where R<sub>ON</sub> is the input-to-BAT resistance. Failure to meet this requirement results in an incomplete charge or increased charge time.

#### **Recommended PCB Layout and Routing**

Place all bypass capacitors for IN and BAT as close as possible to the IC. Connect the battery to BAT as close as possible to the IC to provide accurate battery voltage sensing. Provide a large copper ground plane to allow the exposed pad to sink heat away from the device. Make all high-current traces short and wide to minimize voltage drops. A sample layout is available in the MAX8922L Evaluation Kit to speed designs.

### **Typical Application Circuits**

#### AC Adapter Charge

Figure 4 shows the MAX8922L as a Li+ battery charger with an AC adapter. The MAX8922L detects the presence of an input supply resulting in POK pulled low. Once POK is pulled low, the MAX8922L begins charging the battery when EN/SET is low or unconnected. The system can program the charge current by EN/SET pulses. By monitoring CHG, the system can detect the top-off threshold and terminate the charge through EN/SET. The MAX8922L also provides an overvoltage-protected 4.94V LDO output to a low-voltage-rated USB system input.

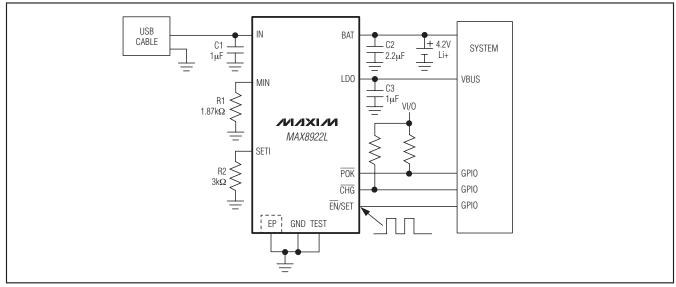


Figure 5. USB-Powered Li+ Battery-Charger Application

#### USB Charge

The universal serial bus (USB) provides a high-speed serial communications port as well as power for the remote device. The MAX8922L can be configured to charge a single Li+ battery at the highest current possible from the host port. Figure 5 shows the MAX8922L as a USB battery charger. The microprocessor enumerates the host to determine its current capability. The system can program the charge current to 90mA, ISETI, or 400mA by EN/SET pulses if the host port is capable. The MAX8922L also provides an overvoltage-protected 4.94V LDO output to a low-voltage-rated USB system input.

#### GSM Test Mode

Figure 6 shows the MAX8922L in a GSM test mode. By sending three pulses to EN/SET, the MAX8922L goes into GSM test mode. GSM PA can pull up to 2.3A for 576µs once every 217Hz from the MAX8922L's output. The configuration in Figure 6 is used for system development, testing, and calibrations in the production or design stage.

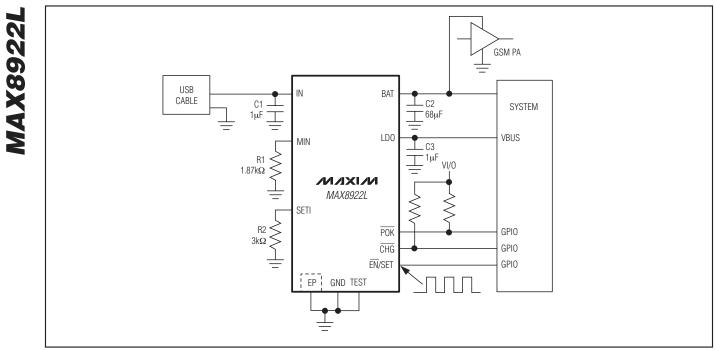


Figure 6. GSM Test Mode

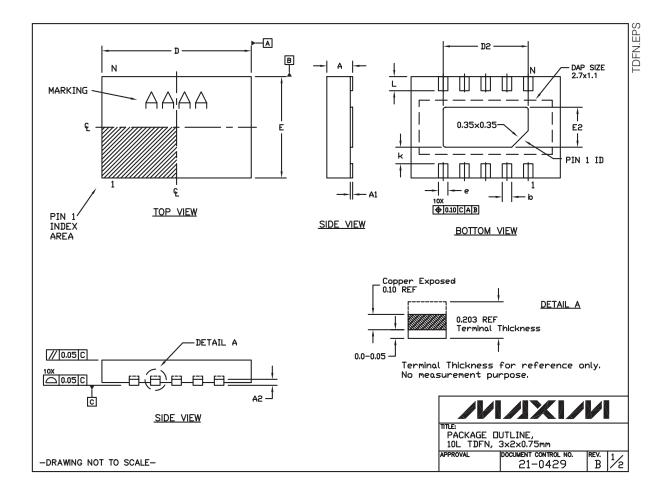
**Chip Information** 

PROCESS: BICMOS

### **Package Information**

For the latest package outline information and land patterns, go to <u>www.maxim-ic.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	DOCUMENT NO.	
10 TDFN-EP	T1032N+1	<u>21-0429</u>	



### **Package Information (continued)**

For the latest package outline information and land patterns, go to <u>www.maxim-ic.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

NOTES:

- 1. N IS THE TOTAL NUMBER OF LEADS.
- 2. ND AND NE REFER TO THE NUMBER OF TERMINALS OF EACH D AND E SIDE RESPECTIVELY.
- 3. REFER TO JEDEC MO-229(WECD-2) WITH VARIATION OF D2, E2 & L.
- 4. WARPAGE SHALL NOT EXCEED 0.10 mm.
- 5. COPLANARITY APPLIES TO THE EXPOSED PAD AS WELL AS THE TERMINALS. COPLANARITY SHALL NOT EXCEED 0.08 mm.
- 6. PKG. LENGTH/PKG. WIDTH ARE CONSIDERED AS SPECIAL CHARACTERISTIC.
- 7. NUMBER OF LEADS SHOWN ARE FOR REFERENCE ONLY.
- 8. MARKING SHOWN IS FOR PACKAGE ORIENTATION REFERENCE ONLY.
- 9. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND PbFREE (+) PKG. CODES.

COMMON DIMENSIONS				
SYMBOL	MIN.	NDM.	MAX.	
Α	0.70	0.75	0.80	
A1	0.0		0.05	
A2	0	10 REF	<del>.</del> .	
D	2.90	3.00	3.10	
E	1.90	2.00	2.10	
D2	1.90	2.00	2.10	
E2	0.80	0.90	1.00	
Ν		10		
b	0	.25±0.0	5	
e	0.50 REF			
L	0.20	0.30	0.40	
PKG. CODE: T1032N-1				

		-
	ππ <u>ε</u> PACKAGE DUTLINE, 10L TDFN, 3×2×0.75mm	•
ING NOT TO SCALE-	APPROVAL DOCUMENT CONTROL NO. REV. 2 21-0429 B 22	

### **Revision History**

1	REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
	0	9/09	Initial release	—
	1	4/10	Replaced 1-Wire references with single-pin, updated <i>Absolute Maximum Ratings</i> section, and added soldering temperature	1, 2

**MAX8922L** 

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