the trip threshold voltage.



# **4-Pin Micropower Voltage Monitors**

#### **Features**

- ±1.25% Precision Voltage Threshold
  - ♦ SOT143 Package
  - Low Cost

TOP VIEW

GND

V<sub>CC</sub>

2

- ♦ < 5µA Typical Supply Current
- Open-Drain Output (MAX836) Push-Pull Output (MAX837)

### **Applications**

**General Description** 

The MAX836/MAX837 micropower voltage monitors

contain a 1.204V precision bandgap reference and a

comparator in a SOT143 package. The MAX836 has an open-drain, n-channel output driver, while the MAX837

has a push-pull output driver. Two external resistors set

Precision Battery Monitor Load Switching Battery-Powered Systems Threshold Detectors

	<u> </u>	 		
TEM	PR	PIN	I-	ТОР

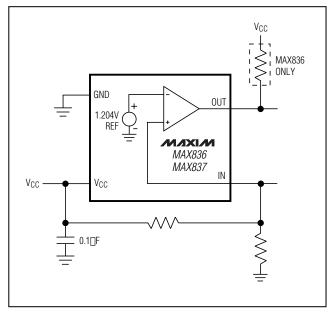
**Ordering Information** 

PART*	TEMP RANGE	PIN- PACKAGE	TOP MARK	
MAX836EUS-T	-40°C to +85°C	4 SOT143-4	EQAA	
MAX837EUS-T	-40°C to +85°C	4 SOT143-4	ERAA	
MAX837EUS-1	-40°C to +85°C	4 501 143-4	ERA	

\*All devices available in tape-and-reel only. Contact factory for availability.

Devices are available in both leaded and lead-free packaging. Specify lead-free by replacing "-T" with "+T" when ordering.

## **Typical Operating Circuit**



### M/IXI/M

\_ Maxim Integrated Products 1

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.



OUT

IN

4

3

MAXIM

MAX836

MAX837

**SOT143** 

### **ABSOLUTE MAXIMUM RATINGS**

V <sub>CC</sub> , OUT to GND (MAX836)0.3V to 12V IN, OUT to GND (MAX837)0.3V to (V <sub>CC</sub> + 0.3V)	Rate of Rise, V <sub>CC</sub> 100V/µs Continuous Power Dissipation
Input Current	4-Pin SOT143 (derate 4mW/°C above +70°C)320mW
	Operating Temperature Range40°C to +85°C
IN10mA	Storage Temperature Range65°C to +150°C
Output Current, OUT20mA	Lead Temperature (soldering, 10s)+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### **ELECTRICAL CHARACTERISTICS**

(V<sub>CC</sub> = +2.5V to +11.0V,  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at  $T_A = +25^{\circ}$ C.)

PARAMETER	SYMBOL	CONDITIONS			MIN	ТҮР	MAX	UNITS
Operating Voltage Range (Note 1)	Vcc				2.5		11.0	V
		$V_{IN} = 1.16V,$ OUT = low	V <sub>CC</sub> = 3.6V	$T_A = +25^{\circ}C$		3.5	6.5	
				$T_A = T_{MIN}$ to $T_{MAX}$			10	
Supply Current (Note 2)			$V_{CC} = full operat$	V <sub>CC</sub> = full operating range			15	
Supply Current (Note 2)	ICC	$V_{IN} = 1.25V,$ OUT = high	V <sub>CC</sub> = 3.6V	$T_A = +25^{\circ}C$		2.0	5.0	μΑ
			VCC = 3.6V	$T_A = T_{MIN}$ to $T_{MAX}$			8.0	
		oor - nign	$V_{CC} = full operat$	V <sub>CC</sub> = full operating range			13	
Trip Threshold Voltage	Vтн	VIN falling	$T_A = +25^{\circ}C$		1.185	1.204	1.215	V
The Theshold Voltage	VIH	VINTAILING	$T_{A} = -40^{\circ}C \text{ to } +85^{\circ}C$		1.169	1.204	1.231	v
Trip Threshold Voltage Hysteresis	VHYST	$V_{CC} = 5V$ , IN = low to high			6		mV	
IN Operating Voltage Range (Note 1)	VIN						V <sub>CC</sub> - 1	V
IN Leakage Current (Note 3)	I <sub>IN</sub>	$V_{IN} = V_{TH}$			±3	±12	nA	
Propagation Delay	t <sub>PL</sub>	V <sub>CC</sub> = 5.0V, 50mV overdrive			80		μs	
Glitch Immunity		V <sub>CC</sub> = 5.0V, 100mV overdrive			35		μs	
OUT Rise Time	t <sub>RT</sub>	V <sub>CC</sub> = 5.0V, no load (MAX837 only)			260		ns	
OUT Fall Time	tFT	$V_{CC} = 5.0V$ , no load (MAX836 pull-up = $10k\Omega$ )			680		ns	
Output Leakage Current (Note 4)	ILOUT	V <sub>IN</sub> > V <sub>THMAX</sub> (MAX836 only)				±1	μA	
Output-Voltage High	Voh	VIN > VTHMAX, ISOURCE = 500µA (MAX837 only)			Vcc - 0.5	5		V
Output-Voltage Low	Vol	VIN < VTHMIN, ISINK = 500µA					0.4	V

Note 1: The voltage-detector output remains in the direct state for V<sub>CC</sub> down to 1.2V when  $V_{IN} \le V_{CC}/2$ .

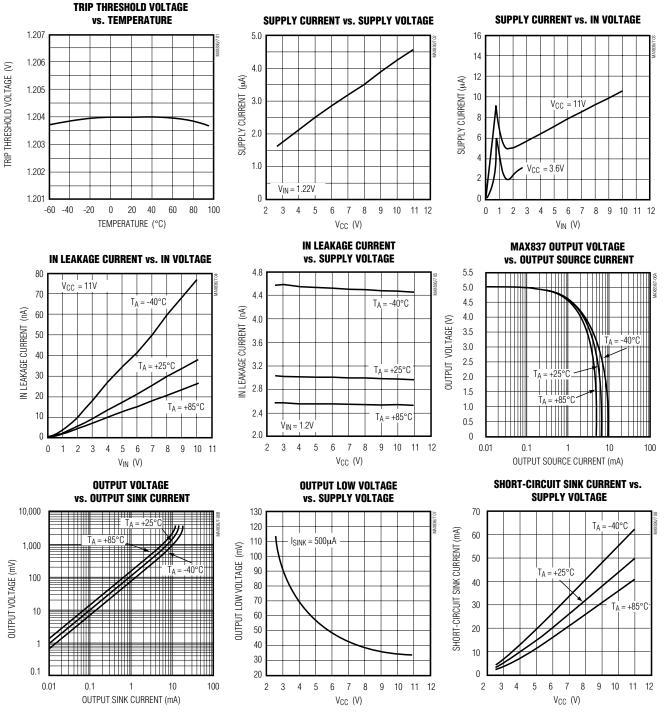
Note 2: Supply current has a monotonic dependence on V<sub>CC</sub> (see the *Typical Operating Characteristics*).

**Note 3:** IN leakage current has a monotonic dependence on V<sub>CC</sub> (see the *Typical Operating Characteristics*).

Note 4: The MAX836 open-drain output can be pulled up to a voltage greater than V<sub>CC</sub>, but may not exceed 11V.

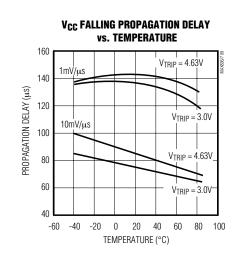
### **Typical Operating Characteristics**

 $(V_{CC} = +5V, R_{LOAD} = 1M\Omega, R_{PULLUP} = 10k\Omega$  (MAX836 only),  $T_A = +25^{\circ}C$ , unless otherwise noted.)



## **Typical Operating Characteristics (continued)**

 $(V_{CC} = +5V, R_{LOAD} = 1M\Omega, R_{PULLUP} = 10k\Omega$  (MAX836 only),  $T_A = +25^{\circ}C$ , unless otherwise noted.)



#### **Pin Description**

PIN	NAME	FUNCTION
1	GND	System Ground
2	Vcc	System Supply Input
3	IN	Noninverting Input to the Comparator. The inverting input connects to the internal 1.204V bandgap reference.
4	OUT	Open-Drain (MAX836) or Push-Pull (MAX837) Output

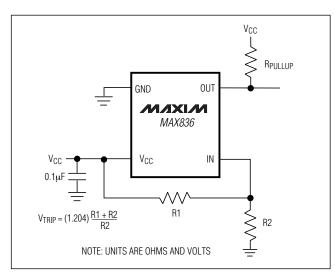
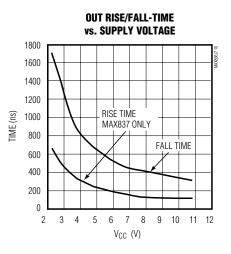


Figure 1. Programming the Trip Voltage, VTRIP



### **Detailed Description**

The MAX836/MAX837 micropower voltage monitors contain a 1.204V precision bandgap reference and a comparator (see the *Typical Operating Circuit*). The only difference between the two parts is the structure of the comparator output driver. The MAX836 has an open-drain n-channel output driver that can be pulled up to a voltage higher than V<sub>CC</sub>, but under 11V. The MAX837's output is push-pull, and can both source and sink current.

#### **Programming the Trip Voltage**

Two external resistors set the trip voltage, VTRIP (Figure 1). VTRIP is the point at which the applied voltage (typically VCC) toggles OUT. The MAX836/MAX837's high input impedance allows large-value resistors without compromising trip-voltage accuracy. To minimize current consumption, select a value for R2 between 500k $\Omega$  and 1M $\Omega$ , then calculate R1 as follows:

$$R1 = R2\left(\frac{V_{TRIP}}{V_{TH}} - 1\right)$$

where  $V_{TRIP}$  = desired trip voltage (in volts),  $V_{TH}$  = threshold trip voltage (1.204V).

### **Applications Information**

#### **Adding Hysteresis**

Hysteresis adds noise immunity to the MAX836/MAX837 and prevents repeated triggering when  $V_{\rm IN}$  is near the threshold trip voltage. Figure 2 shows how to add hysteresis to the comparator. The technique is similar for



both parts. For the MAX836, select the ratio of resistors R1 and R2 so that IN sees 1.204V when the monitor voltage falls to or rises above the desired trip point (VTRIP). R3 adds hysteresis and is typically an order of magnitude larger than R1 or R2. The current through R1 and R2 should be at least 500nA to ensure that the 12nA maximum input current does not shift the trip point significantly. Capacitor C1 adds additional noise rejection.

#### Monitoring Voltages Other than VCC

The MAX836/MAX837 can monitor voltages other than V<sub>CC</sub> (Figure 3). Calculate V<sub>TRIP</sub> as shown in the *Programming the Trip Voltage* section. The monitored voltage (V<sub>MON</sub>) is independent of V<sub>CC</sub>. V<sub>IN</sub> must be 1V less than V<sub>CC</sub>.

#### **Heater Temperature Control**

Figure 4 shows a basic heater temperature-control circuit. Upon power-up, OUT is high and the n-channel MOSFET turns on. Current flows through the heating element (R4), warming the surrounding area. R2 is a negative-temperature-coefficient thermistor and as temperature increases, its resistance decreases. As the thermistor heats up and its resistance decreases, the MAX837's voltage at IN decreases until it reaches the 1.204V threshold voltage. At this point, OUT goes low, turning off the heating element. The thermistor cools and the voltage at IN rises until it overcomes the MAX837's hysteresis (6mV). OUT returns high when this point is reached, turning on the heating element again. This cycle repeats as long as power is applied.

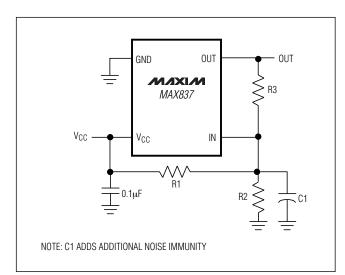


Figure 2. Adding Hysteresis to the Comparator

Determine the thermistor's resistance (R2) at the desired temperature. Then, using R2's resistance and half the resistance of R3, calculate R1's value with the following formula:

R1 = (R2 + R3) 
$$\left(\frac{V_{CC}}{1.204} - 1\right)$$

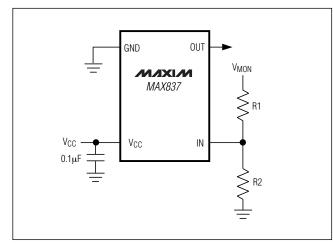


Figure 3. Monitoring Voltages Other than V<sub>CC</sub>

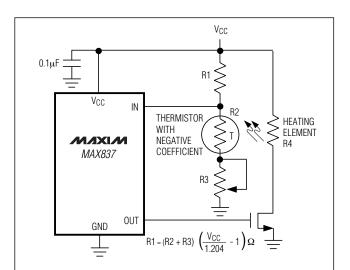


Figure 4. Heater Temperature Control

### Chip Information

#### Package Information

TRANSISTOR COUNT: 54

For the latest package outline information and land patterns, go to **www.maxim-ic.com/packages**.

PACKAGE TYPE	PACKAGE CODE	DOCUMENT NO.
4 SOT143	U4-1	<u>21-0052</u>

## **Revision History**

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	9/96	Initial release	—
1	3/04	Updated top mark information in the Ordering Information.	1
2	12/05	Added lead-free notation.	1
3	5/08	Updated top mark information in the Ordering Information.	1

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

Maxim Integrated Products, 120 San Gabriel Drive, Sunnyvale, CA 94086 408-737-7600 \_\_\_\_

© 2008 Maxim Integrated Products

is a registered trademark of Maxim Integrated Products, Inc.

\_\_\_\_7