

S-19611A

Rev.1.2 00

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AUTOMOTIVE, 105°C OPERATION, LOW INPUT OFFSET VOLTAGE CMOS OPERATIONAL AMPLIFIER

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This IC incorporates a general purpose analog circuit in a small package. This is a zero-drift operational amplifier with Rail-to-Rail input and output, which uses auto-zeroing techniques to provide low input offset voltage. This IC is suitable for applications requiring less offset voltage. The S-19611AB is a dual operational amplifier (2 circuits).

Caution This product can be used in vehicle equipment and in-vehicle equipment. Before using the product for these purposes, it is imperative to contact our sales representatives.

Features

• Low input offset voltage:

• Operation power supply voltage range:

- Low current consumption (Per circuit):
- Internal phase compensation:
- Rail-to-Rail input and output
- Operation temperature range:
- Lead-free (Sn 100%), halogen-free
- AEC-Q100 qualified^{*1}

*1. Contact our sales representatives for details.

Applications

- High-accuracy current detection
- Various sensor interfaces
- Strain gauge amplifier

Package

• TMSOP-8

 $Ta = -40^{\circ}C \text{ to } +105^{\circ}C$

AUTOMOTIVE, 105°C OPERATION, LOW INPUT OFFSET VOLTAGE CMOS OPERATIONAL AMPLIFIER S-19611A Rev.1.2_00

Block Diagram

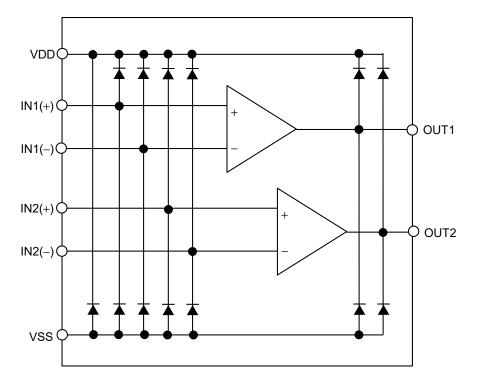


Figure 1

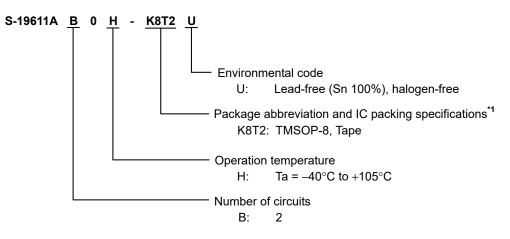
AEC-Q100 Qualified

This IC supports AEC-Q100 for the operation temperature grade 2. Contact our sales representatives for details of AEC-Q100 reliability specification.

Product Name Structure

Refer to "1. Product name" regarding the contents of product name, "2. Package" regarding the package drawings and "3. Product name list" regarding the product type.

1. Product name



*1. Refer to the tape drawing.

2. Package

Table 1 Package Drawing Codes

Package Name	Dimension	Таре	Reel
TMSOP-8	FM008-A-P-SD	FM008-A-C-SD	FM008-A-R-SD

3. Product name list

Table 2

Product Name	Package
S-19611AB0H-K8T2U	TMSOP-8

Pin Configuration

1. TMSOP-8

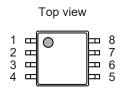


Figure 2

Pin No.	Symbol	Description
1	OUT1	Output pin 1
2	IN1(-)	Inverted input pin 1
3	IN1(+)	Non-inverted input pin 1
4	VSS	GND pin
5	IN2(+)	Non-inverted input pin 2
6	IN2(-)	Inverted input pin 2
7	OUT2	Output pin 2
8	VDD	Positive power supply pin

Table 3

Absolute Maximum Ratings

Table 4

		(Ta = -40°C to +105°C unless othe	rwise specified)
Item	Symbol	Absolute Maximum Rating	Unit
Power supply voltage	V _{DD}	V _{SS} – 0.3 to V _{SS} + 6.0	V
Input voltage	VIN(+), VIN(-)	$V_{SS} - 0.3$ to $V_{DD} + 0.3$	V
Output voltage	Vout	$V_{SS} - 0.3$ to $V_{DD} + 0.3$	V
Differential input voltage	VIND	±5.5	V
Output nin ourrent	ISOURCE	10.0	mA
Output pin current	Isink	10.0	mA
Operation ambient temperature	T _{opr}	-40 to +105	°C
Storage temperature	T _{stg}	–55 to +125	°C

Caution The absolute maximum ratings are rated values exceeding which the product could suffer physical damage. These values must therefore not be exceeded under any conditions.

■ Thermal Resistance Value

Table 5

ltem	Symbol	Condi	tion	Min.	Тур.	Max.	Unit
Junction-to-ambient thermal resistance*1	ALθ		Board A	_	160	-	°C/W
			Board B	_	133	-	°C/W
		TMSOP-8	Board C	_	-	-	°C/W
			Board D	_		-	°C/W
			Board E	_	_	_	°C/W

*1. Test environment: compliance with JEDEC STANDARD JESD51-2A

Remark Refer to "■ Power Dissipation" and "Test Board" for details.

Electrical Characteristics

1. Recommended operation condition

		Table 6					
		(Ta -	= -40°C to	+105°C ι	inless othei	wise sp	ecified)
Item	Symbol	Condition	Min.	Тур.	Max.	Unit	Test Circuit
Operation power supply voltage range	V _{DD}	_	2.65	5.00	5.50	V	_

2. $V_{DD} = 5.0 V$

DC Electrical Characteristics		Table 7 (Ta	= -40°C to	+105°C	unless othe	rwise s	pecified)
ltem	Symbol	Condition	Min.	Тур.	Max.	Unit	Test Circuit
Current consumption (2 circuits)	Idd	$V_{CMR} = V_{OUT} = \frac{V_{DD}}{2}$	-	400	600	μA	5
Input offset voltage	VIO	$V_{CMR} = \frac{V_{DD}}{2}$, Ta = +25°C	-17	±1	+17	μV	1
input onoot voltago	•10	$V_{CMR} = \frac{V_{DD}}{2}$	-100	±1	+100	μV	1
Input offset voltage drift	$\frac{\Delta V_{IO}}{\Delta Ta}$	$V_{CMR} = \frac{V_{DD}}{2}$	-	±0.1	_	μV/°C	1
Input bigg ourrent	1	Ta = +25°C	_	±70	_	pА	_
Input bias current	Ibias	_	-	±3000	-	pА	_
Input offset current	lio	Ta = +25°C	-	±140	-	рА	-
	10	_	-	±300	_	рА	_
Common-mode input voltage range	Vcmr	-	Vss - 0.1	-	$V_{\text{DD}} + 0.1$	V	2
Voltage gain (open loop)	Avol	$\begin{split} V_{SS} + 0.1 \ V \leq V_{OUT} \leq V_{DD} - 0.1 \ V, \\ V_{CMR} = \frac{V_{DD}}{2}, \ R_L = 10 \ k\Omega \end{split} \label{eq:VSS}$	106	130	_	dB	8
N	Vон	R _L = 10 kΩ	4.9	-	-	V	3
Maximum output swing voltage	Vol	R _L = 10 kΩ	-	-	0.1	V	4
Common-mode input signal rejection ratio	CMRR	$V_{SS} - 0.1 \ V \leq V_{CMR} \leq V_{DD} + 0.1 \ V$	100	130	_	dB	2
Power supply voltage rejection ratio	PSRR	$2.65~V \leq V_{DD} \leq 5.50~V$	95	120	_	dB	1
Source current	ISOURCE	$V_{OUT} = V_{DD} - 0.1 V$	0.8	2.5	-	mA	6
Sink current	Isink	V _{OUT} = 0.1 V	1.0	2.9	-	mA	7

Table 8

AC Electrical Characteristics		(Та	= -40°C to +	105°C unles	ss otherwise	specified)
Item	Symbol	Condition	Min.	Тур.	Max.	Unit
Slew rate	SR	R _L = 1.0 MΩ, C _L = 15 pF (Refer to Figure 11)	-	0.22	_	V/µs
Gain-bandwidth product	GBP	C _L = 0 pF	_	320	_	kHz

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Test Circuits (Per circuit)

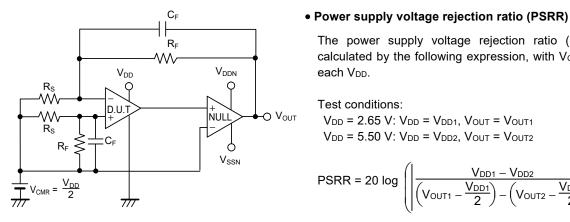


Figure 3 Test Circuit 1

1. Power supply voltage rejection ratio, input offset voltage

The power supply voltage rejection ratio (PSRR) can be

calculated by the following expression, with V_{OUT} measured at each VDD.

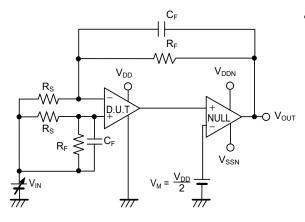
Test conditions: VDD = 2.65 V: VDD = VDD1, VOUT = VOUT1 V_{DD} = 5.50 V: V_{DD} = V_{DD2} , V_{OUT} = V_{OUT2}

$$\mathsf{PSRR} = 20 \log \left(\left| \frac{\mathsf{V}_{\mathsf{DD1}} - \mathsf{V}_{\mathsf{DD2}}}{\left(\mathsf{V}_{\mathsf{OUT1}} - \frac{\mathsf{V}_{\mathsf{DD1}}}{2} \right) - \left(\mathsf{V}_{\mathsf{OUT2}} - \frac{\mathsf{V}_{\mathsf{DD2}}}{2} \right)} \right| \times \frac{\mathsf{R}_{\mathsf{F}} + \mathsf{R}_{\mathsf{S}}}{\mathsf{R}_{\mathsf{S}}} \right)$$

• Input offset voltage (VIO)

$$V_{IO} = \left(V_{OUT} - \frac{V_{DD}}{2}\right) \times \frac{R_S}{R_F + R_S}$$

2. Common-mode input signal rejection ratio, common-mode input voltage range





• Common-mode input signal rejection ratio (CMRR)

The common-mode input signal rejection ratio (CMRR) can be calculated by the following expression, with VOUT measured at each VIN.

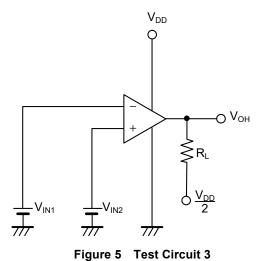
Test conditions: VIN = VCMR Max.: VIN = VIN1, VOUT = VOUT1 VIN = VCMR Min.: VIN = VIN2, VOUT = VOUT2

$$\mathsf{CMRR} = 20 \log \left(\left| \frac{\mathsf{V}_{\mathsf{IN1}} - \mathsf{V}_{\mathsf{IN2}}}{(\mathsf{V}_{\mathsf{OUT1}} - \mathsf{V}_{\mathsf{IN1}}) - (\mathsf{V}_{\mathsf{OUT2}} - \mathsf{V}_{\mathsf{IN2}})} \right| \times \frac{\mathsf{R}_{\mathsf{F}} + \mathsf{R}_{\mathsf{S}}}{\mathsf{R}_{\mathsf{S}}} \right)$$

• Common-mode input voltage range (V_{CMR})

The common-mode input voltage range is the range of VIN in which VOUT satisfies the common-mode input signal rejection ratio specifications when VIN is changed.

3. Maximum output swing voltage



• Maximum output swing voltage (VoH)

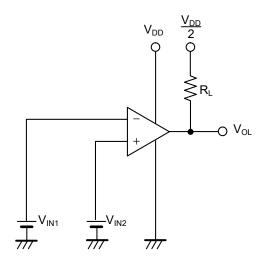
Test conditions:

$$V_{IN1} = \frac{V_{DD}}{2} - 0.1 V$$

$$V_{IN2} = \frac{V_{DD}}{2} + 0.1 V$$

$$R_L = 10 k\Omega$$

4. Maximum output swing voltage



• Maximum output swing voltage (VoL)

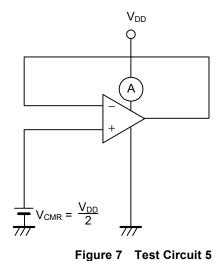
Test conditions:

$$V_{IN1} = \frac{V_{DD}}{2} + 0.1 V$$

 $V_{IN2} = \frac{V_{DD}}{2} - 0.1 V$

Figure 6 Test Circuit 4

5. Current consumption



• Current consumption (IDD)

• Source current (ISOURCE)

 $V_{\text{OUT}} = V_{\text{DD}} - 0.1 \text{ V}$ $V_{\text{IN1}} = \frac{V_{\text{DD}}}{2} - 0.1 \text{ V}$

 $V_{IN2} = \frac{V_{DD}}{2} + 0.1 \text{ V}$

Test conditions:

6. Source current

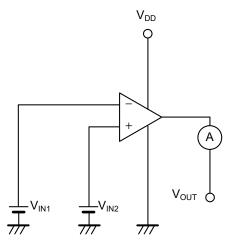
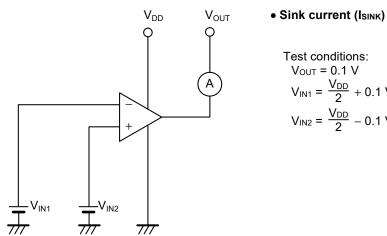


Figure 8 Test Circuit 6

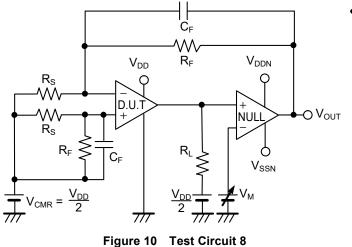
7. Sink current



- - Test conditions: V_{OUT} = 0.1 V $V_{IN1} = \frac{V_{DD}}{2} + 0.1 V$ $V_{IN2} = \frac{V_{DD}}{2} - 0.1 V$

Figure 9 Test Circuit 7

8. Voltage gain



• Voltage gain (open loop) (Avol)

The voltage gain (A_{VOL}) can be calculated by the following expression, with V_{OUT} measured at each $V_{M}.$

Test conditions:

$$A_{VOL} = 20 \log \left(\left| \frac{V_{M1} - V_{M2}}{V_{OUT1} - V_{OUT2}} \right| \times \frac{R_F + R_S}{R_S} \right)$$
$$R_L = 10 \text{ k}\Omega$$

9. Slew rate

Measured by the voltage follower circuit.

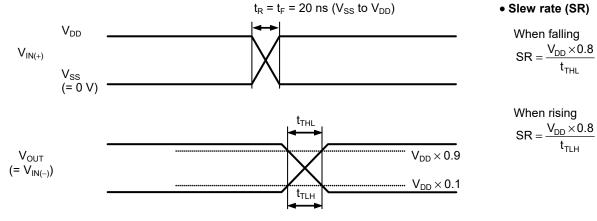
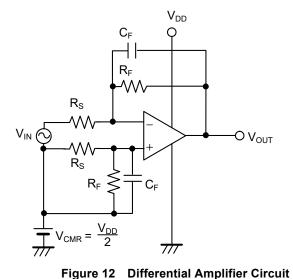


Figure 11

Usage Examples



 $\begin{array}{l} [\text{Example of Gain = 1000 times}] \\ R_{\text{S}} = 1 \ \text{k}\Omega \\ R_{\text{F}} = 1 \ \text{M}\Omega \\ C_{\text{F}} = 1000 \ \text{pF} \end{array}$

 $\begin{array}{l} \mbox{[Example of Gain = 100 times]} \\ R_{S} = 1 \ k\Omega \\ R_{F} = 100 \ k\Omega \\ C_{F} = 1000 \ pF \end{array}$

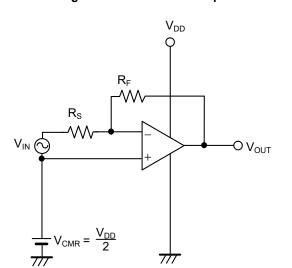
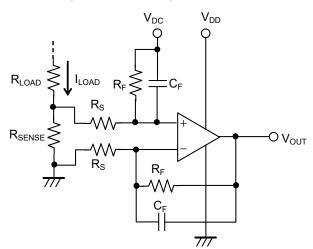
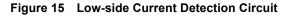


Figure 13 Inverting Amplifier Circuit





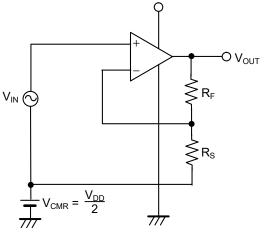


Figure 14 Non-inverting Amplifier Circuit

 V_{DD}

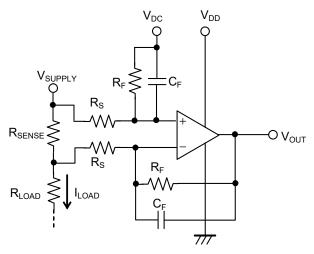
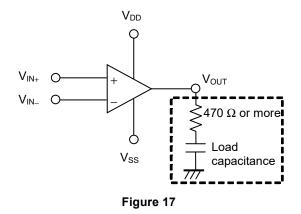


Figure 16 High-side Current Detection Circuit

Caution The above connection diagrams and constants will not guarantee successful operation. Perform thorough evaluation using the actual application to set the constants.

Precautions

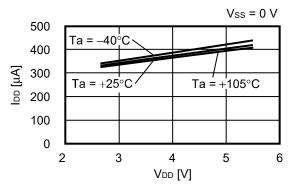
- During the operation of an operational amplifier circuit, when V_{OUT} ≤ V_{SS} + 100 mV or V_{OUT} ≥ V_{DD} 100 mV, the signal becomes difficult to be output, and the output voltage (V_{OUT}) may become V_{SS} or V_{DD}. If this happens, supply an appropriate input signal to the operational amplifier so that V_{OUT} is within the range of V_{SS} + 100 mV to V_{DD} 100 mV. Contact our sales representatives if you have any questions for use in the above operation conditions.
- Generally an operational amplifier may cause oscillation depending on the selection of external parts. Perform thorough evaluation using the actual application to set the constant.
- Do not apply an electrostatic discharge to this IC that exceeds performance ratings of the built-in electrostatic protection circuit.
- ABLIC Inc. claims no responsibility for any disputes arising out of or in connection with any infringement by products including this IC of patents owned by a third party.
- Use this IC with the output current of 10 mA or less.
- When using the voltage follower circuit (Gain = 1 time), connect a resistor of 470 Ω or more for the stable operation as shown in **Figure 17**. The operation may become unstable depending on the value of the load capacitance connected to the output pin, even when the voltage follower circuit is not used. Use the product under thorough evaluation.



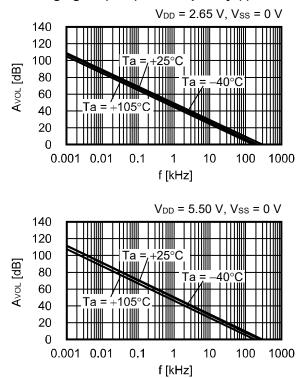
Caution The above connection diagram and constant will not guarantee successful operation. Perform thorough evaluation using the actual application to set the constant.

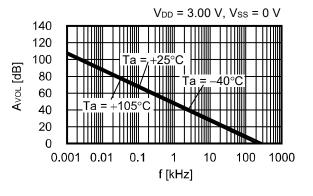
Characteristics (Typical Data)

1. Current consumption (IDD) (2 circuits) vs. Power supply voltage (VDD)

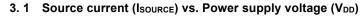


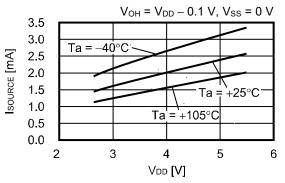
2. Voltage gain (Avol) vs. Frequency (f)



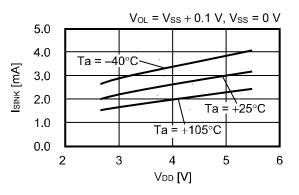


3. Output current

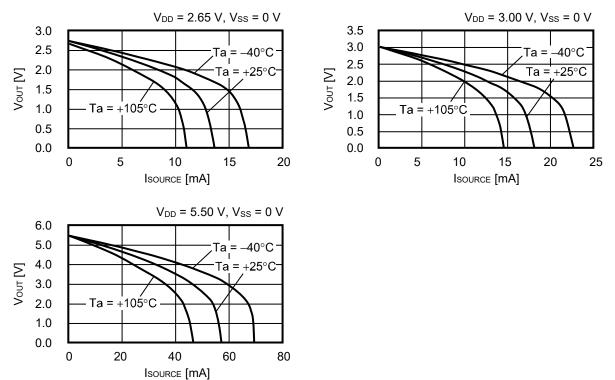


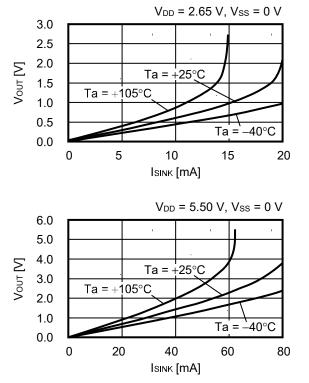


3. 2 Sink current (ISINK) vs. Power supply voltage (VDD)

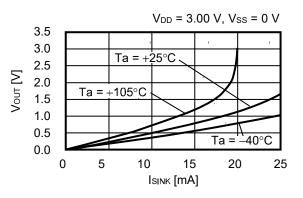


3. 3 Output voltage (Vout) vs. Source current (Isource)

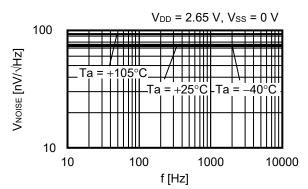


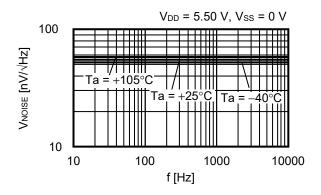


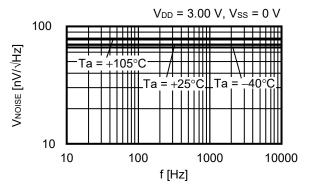




4. Input noise voltage density (V_{NOISE}) vs. Frequency (f)

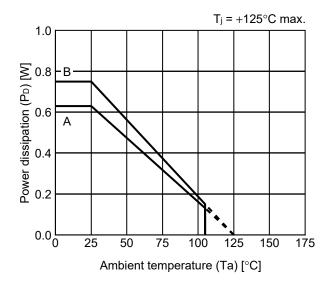






Power Dissipation

TMSOP-8

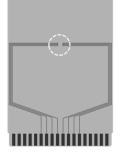


Board	Power Dissipation (P _D)
А	0.63 W
В	0.75 W
С	_
D	_
E	_

TMSOP-8 Test Board

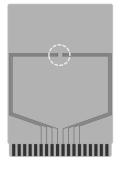
(1) Board A

🔘 IC Mount Area



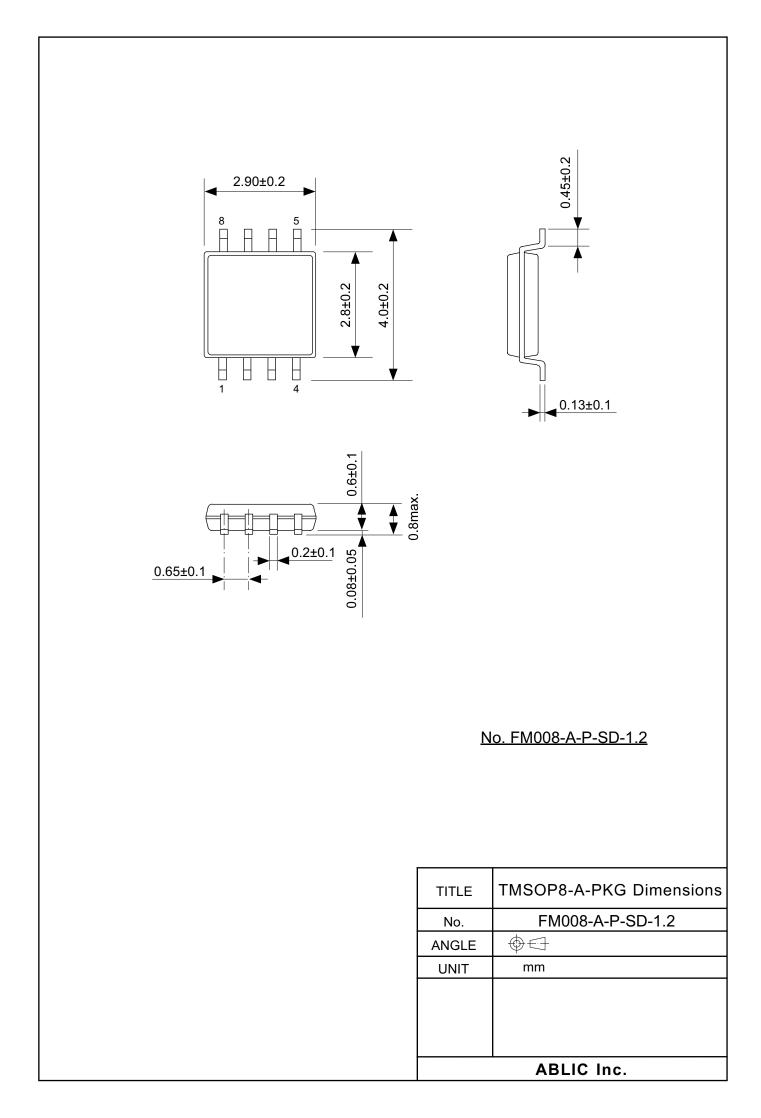
Item		Specification		
Size [mm]		114.3 x 76.2 x t1.6		
Material		FR-4		
Number of copper foil layer		2		
	1	Land pattern and wiring for testing: t0.070		
Coppor foil lovor [mm]	2	-		
Copper foil layer [mm]	3	-		
	4	74.2 x 74.2 x t0.070		
Thermal via		-		

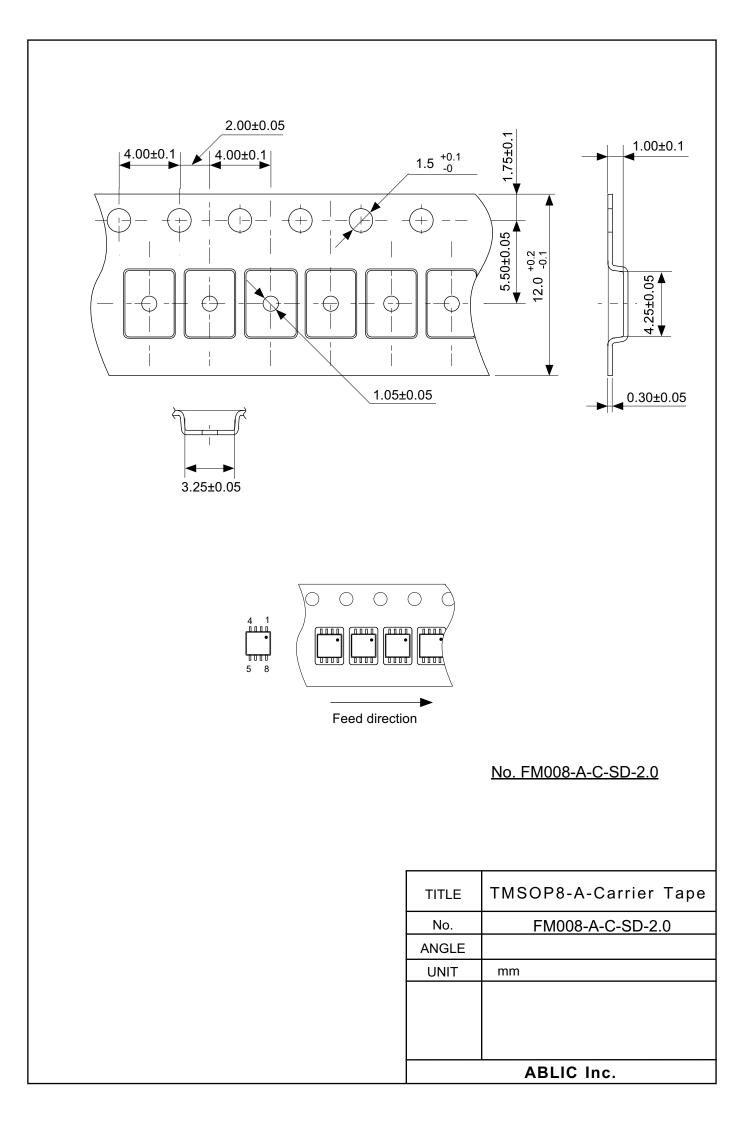
(2) Board B

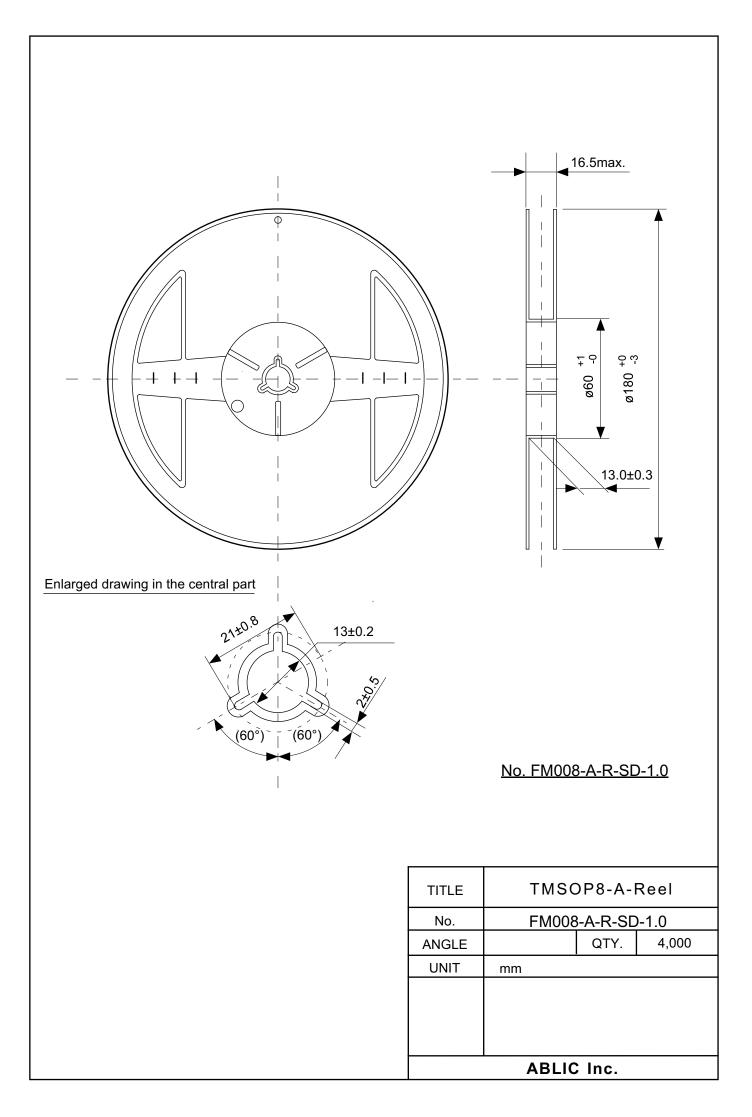


Item		Specification
Size [mm]		114.3 x 76.2 x t1.6
Material		FR-4
Number of copper foil layer		4
	1	Land pattern and wiring for testing: t0.070
Copper foil layer [mm]	2	74.2 x 74.2 x t0.035
	3	74.2 x 74.2 x t0.035
	4	74.2 x 74.2 x t0.070
Thermal via		-

No. TMSOP8-A-Board-SD-1.0







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The entire system in which the products are used must be sufficiently evaluated and judged whether the products are allowed to apply for the system on customer's own responsibility.

- 10. The products are not designed to be radiation-proof. The necessary radiation measures should be taken in the product design by the customer depending on the intended use.
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2.4-2019.07