

## Double channel high side driver with analog current sense for automotive applications

### Features

Max supply voltage	$V_{CC}$	41 V
Operating voltage range	$V_{CC}$	4.5 to 36V
Max on-state resistance (per ch.)	$R_{ON}$	160 mΩ
Current limitation (typ)	$I_{LIMH}$	5 A
Off state supply current	$I_S$	2 μA <sup>(1)</sup>

1. Typical value with all loads connected.

#### ■ General features

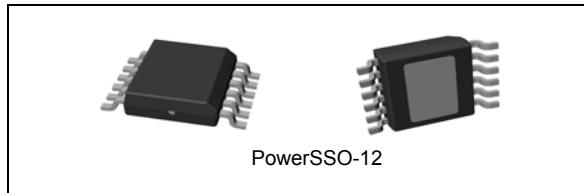
- Inrush current active management by power limitation
- Very low stand-by current
- 3.0V CMOS compatible input
- Optimized electromagnetic emission
- Very low electromagnetic susceptibility
- In compliance with the 2002/95/EC European directive

#### ■ Diagnostic functions

- Proportional load current sense
- High current sense precision for wide range currents
- Current sense disable
- Thermal shutdown indication
- Very low current sense leakage

#### ■ Protection

- Undervoltage shut-down
- Overvoltage clamp
- Load current limitation
- Self limiting of fast thermal transients
- Protection against loss of ground and loss of  $V_{CC}$
- Thermal shut down



- Reverse battery protection (see [Application schematic](#))
- Electrostatic discharge protection

### Application

- All types of resistive, inductive and capacitive loads
- Suitable as LED driver

### Description

The VND5160AJ-E is a monolithic device made using STMicroelectronics VIPower technology. It is intended for driving resistive or inductive loads with one side connected to ground. Active  $V_{CC}$  pin voltage clamp protects the device against low energy spikes (see ISO7637 transient compatibility table). This device integrates an analog current sense which delivers a current proportional to the load current (according to a known ratio) when CS\_DIS is driven low or left open. When CS\_DIS is driven high, the CURRENT SENSE pin is in a high impedance condition. Output current limitation protects the device in overload condition. In case of long overload duration, the device limits the dissipated power to safe level up to thermal shut-down intervention. Thermal shut-down with automatic restart allows the device to recover normal operation as soon as fault condition disappears.

**Table 1. Device summary**

Package	Order codes	
	Tube	Tape and Reel
PowerSSO-12	VND5160AJ-E	VND5160AJTR-E

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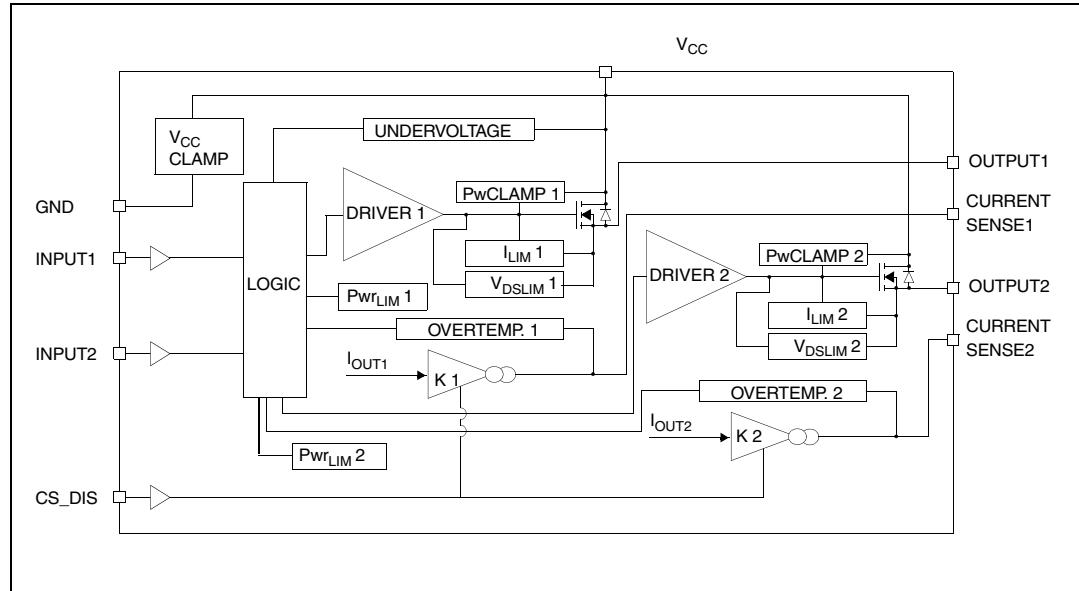
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# 1 Block diagram and pin description

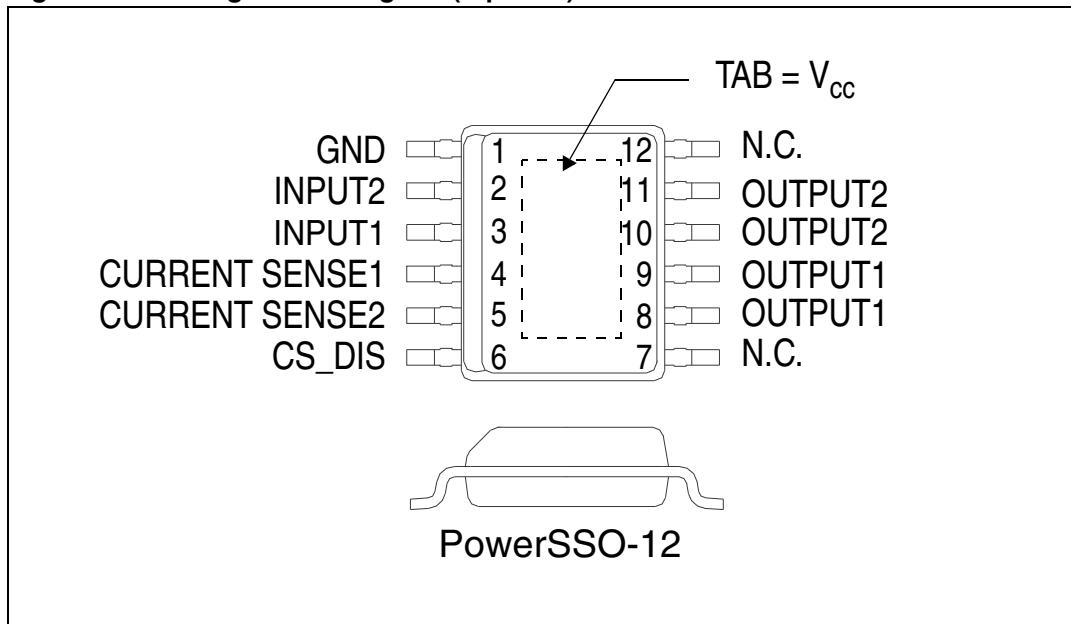
**Figure 1. Block diagram**



**Table 2. Pin function**

Name	Function
V <sub>CC</sub>	Battery connection.
OUTPUT <sub>n</sub>	Power output.
GND	Ground connection. Must be reverse battery protected by an external diode/resistor network.
INPUT <sub>n</sub>	Voltage controlled input pin with hysteresis, CMOS compatible. Controls output switch state.
CURRENT SENSE <sub>n</sub>	Analog current sense pin, delivers a current proportional to the load current.
CS_DIS	Active high CMOS compatible pin, to disable the current sense pin.

Figure 2. Configuration diagram (top view)

**Note:**

*The above pin configuration reflects the changes notified with PCN-APG-BOD/07/2886. The new pinout is backward compatible with existing PCB layouts where pins #7 and 12 are connected to Vcc. For new PCB designs, these pins should be left unconnected.*

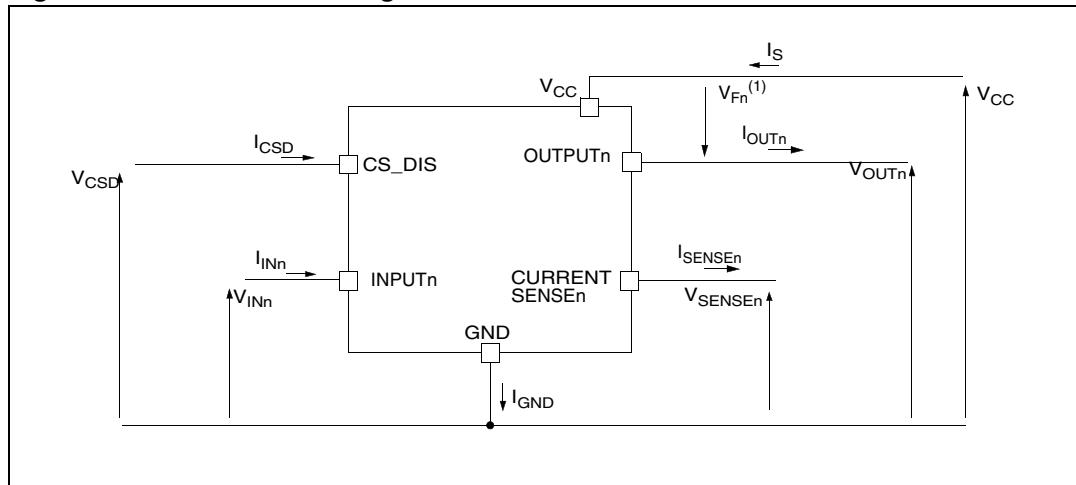
Table 3. Suggested connections for unused and N.C. pins

Connection / Pin	Current Sense	N.C.	Output	Input	CS_DIS
Floating	N.R. <sup>(1)</sup>	X	X	X	X
To ground	Through 1kΩ resistor	X	N.R. <sup>(1)</sup>	Through 10kΩ resistor	Through 10kΩ resistor

1. Not recommended.

## 2 Electrical specifications

**Figure 3. Current and voltage conventions**



Note:  $V_{Fn} = V_{OUTn} - V_{CC}$  during reverse battery condition.

### 2.1 Absolute maximum ratings

Stressing the device above the rating listed in the “Absolute maximum ratings” table may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other conditions above those indicated in the Operating sections of this specification is not implied. Exposure to the conditions in table below for extended periods may affect device reliability. Refer also to the STMicroelectronics SURE Program and other relevant quality document.

**Table 4. Absolute maximum ratings**

Symbol	Parameter	Value	Unit
$V_{CC}$	DC supply voltage	41	V
$-V_{CC}$	Reverse DC supply voltage	0.3	V
$-I_{GND}$	DC reverse ground pin current	200	mA
$I_{OUT}$	DC output current	Internally limited	A
$-I_{OUT}$	Reverse DC output current	6	A
$I_{IN}$	DC input current	-1 to 10	mA
$I_{CSD}$	DC current sense disable input current	-1 to 10	mA
$-I_{CSENSE}$	DC reverse CS pin current	200	mA
$V_{CSENSE}$	Current sense maximum voltage	$V_{CC}-41$ $+V_{CC}$	V V
$E_{MAX}$	Maximum switching energy (single pulse) ( $L=12\text{mH}$ ; $R_L=0\Omega$ ; $V_{bat}=13.5\text{V}$ ; $T_{jstart}=150^\circ\text{C}$ ; $I_{OUT} = I_{limL}(Typ.)$ )	34	mJ

**Table 4. Absolute maximum ratings (continued)**

Symbol	Parameter	Value	Unit
$V_{ESD}$	Electrostatic discharge (Human Body Model: R=1.5KΩ; C=100pF)		
	- INPUT	4000	V
	- CURRENT SENSE	2000	V
	- CS_DIS	4000	V
	- OUTPUT	5000	V
	- $V_{CC}$	5000	V
$V_{ESD}$	Charge device model (CDM-AEC-Q100-011)	750	V
$T_j$	Junction operating temperature	-40 to 150	°C
$T_{stg}$	Storage temperature	-55 to 150	°C

## 2.2 Thermal data

**Table 5. Thermal data**

Symbol	Parameter	Max value	Unit
$R_{thj-case}$	Thermal resistance junction-case (MAX) (With one channel ON)	8	°C/W
$R_{thj-amb}$	Thermal resistance junction-ambient (MAX)	See <a href="#">Figure 29</a>	°C/W

## 2.3 Electrical characteristics

The values specified in this section are for  $8V < V_{CC} < 36V$ ;  $-40^{\circ}C < T_j < 150^{\circ}C$ , unless otherwise stated.

**Table 6. Power section**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$V_{CC}$	Operating supply voltage		4.5	13	36	V
$V_{USD}$	Undervoltage shutdown			3.5	4.5	V
$V_{USDhyst}$	Undervoltage shut-down hysteresis			0.5		V
$R_{ON}$	On state resistance <sup>(1)</sup>	$I_{OUT}= 0.5A; T_j= 25^{\circ}C$ $I_{OUT}= 0.5A; T_j= 150^{\circ}C$ $I_{OUT}= 0.5A; V_{CC}= 5V; T_j= 25^{\circ}C$			160 320 210	$m\Omega$ $m\Omega$ $m\Omega$
$V_{clamp}$	Clamp voltage	$I_S= 20 mA$	41	46	52	V
$I_S$	Supply current	Off State; $V_{CC}= 13V; T_j= 25^{\circ}C$ ; $V_{IN}=V_{OUT}=V_{SENSE}=V_{CSD}=0V$ On State; $V_{CC}=13V; V_{IN}=5V; I_{OUT}=0A$		2 <sup>(2)</sup> 3	5 <sup>(2)</sup> 6	$\mu A$ mA
$I_{L(off)}$	Off state output current <sup>(1)</sup>	$V_{IN}=V_{OUT}=0V; V_{CC}=13V; T_j=25^{\circ}C$ $V_{IN}=V_{OUT}=0V; V_{CC}=13V; T_j=125^{\circ}C$	0 0	0.01	3 5	$\mu A$
$V_F$	Output - $V_{CC}$ diode voltage <sup>(1)</sup>	$-I_{OUT}= 0.6A; T_j=150^{\circ}C$			0.7	V

1. For each channel.

2. PowerMOS leakage included.

**Table 7. Switching ( $V_{CC}=13V, T_j=25^{\circ}C$ )**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$t_{d(on)}$	Turn- On delay time	$R_L= 26\Omega$ (see <a href="#">Figure 8.</a> )		10		$\mu s$
$t_{d(off)}$	Turn- Off delay time	$R_L= 26\Omega$ (see <a href="#">Figure 8.</a> )		15		$\mu s$
$(dV_{OUT}/dt)_{on}$	Turn- On voltage slope	$R_L= 26\Omega$		See <a href="#">Figure 20.</a>		$V/\mu s$
$(dV_{OUT}/dt)_{off}$	Turn- Off voltage slope	$R_L= 26\Omega$		See <a href="#">Figure 22.</a>		$V/\mu s$
$W_{ON}$	Switching energy losses during $t_{won}$	$R_L= 26\Omega$ (see <a href="#">Figure 8.</a> )		0.03		mJ
$W_{OFF}$	Switching energy losses during $t_{woff}$	$R_L= 26\Omega$ (see <a href="#">Figure 8.</a> )		0.02		mJ

**Table 8. Logic input**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$V_{IL}$	Input low level voltage				0.9	V
$I_{IL}$	Low level input current	$V_{IN}= 0.9V$	1			$\mu A$
$V_{IH}$	Input high level voltage		2.1			V
$I_{IH}$	High level input current	$V_{IN}= 2.1V$			10	$\mu A$
$V_{I(hyst)}$	Input hysteresis voltage		0.25			V
$V_{ICL}$	Input clamp voltage	$I_{IN}= 1mA$ $I_{IN}= -1mA$	5.5	-0.7	7	V V
$V_{CSDL}$	CS_DIS low level voltage				0.9	V
$I_{CSDL}$	Low level CS_DIS current	$V_{CSD}= 0.9V$	1			$\mu A$
$V_{CSDH}$	CS_DIS high level voltage		2.1			V
$I_{CSDH}$	High level CS_DIS current	$V_{CSD}= 2.1V$			10	$\mu A$
$V_{CSD(hyst)}$	CS_DIS hysteresis voltage		0.25			V
$V_{CSCL}$	CS_DIS clamp voltage	$I_{CSD}= 1mA$ $I_{CSD}= -1mA$	5.5	-0.7	7	V V

**Table 9. Protection and diagnostics<sup>(1)</sup>**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$I_{limH}$	DC short circuit current	$V_{CC}= 13V$ $5V < V_{CC} < 36V$	3.8	5	7.5 7.5	A A
$I_{limL}$	Short circuit current during thermal cycling	$V_{CC}= 13V$ ; $T_R < T_j < T_{TSD}$		2		A
$T_{TSD}$	Shutdown temperature		150	175	200	°C
$T_R$	Reset temperature		$T_{RS} + 1$	$T_{RS} + 5$		°C
$T_{RS}$	Thermal reset of STATUS		135			°C
$T_{HYST}$	Thermal hysteresis ( $T_{TSD} - T_R$ )			7		°C
$V_{DEMAG}$	Turn-Off output voltage clamp	$I_{OUT}= 1A$ ; $V_{IN}= 0$ ; $L= 20mH$	$V_{CC}-41$	$V_{CC}-46$	$V_{CC}-52$	V
$V_{ON}$	Output voltage drop limitation	$I_{OUT}= 0.03A$ ; $T_j= -40^{\circ}C...150^{\circ}C$ (see <a href="#">Figure 9</a> .)		25		mV

1. To ensure long term reliability under heavy overload or short circuit conditions, protection and related diagnostic signals must be used together with a proper software strategy. If the device is subjected to abnormal conditions, this software must limit the duration and number of activation cycles.

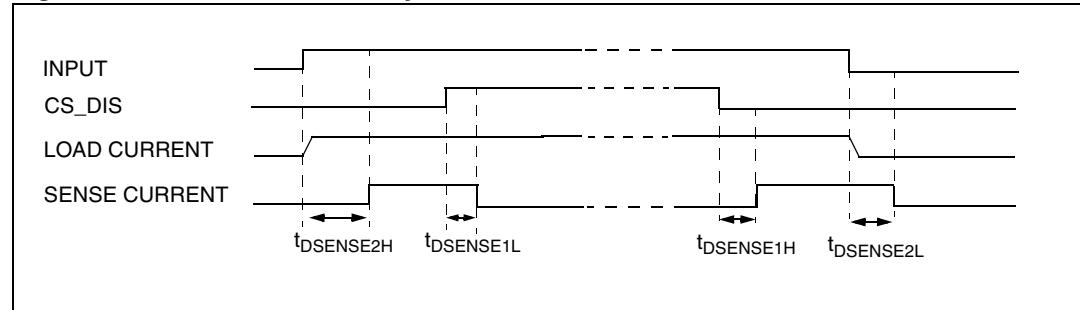
**Table 10. Current sense (8V<VCC<16V)**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$K_0$	$I_{OUT}/I_{SENSE}$	$I_{OUT}=0.025A; V_{SENSE}=0.5V; V_{CSD}=0V;$ $T_j= -40^{\circ}C...150^{\circ}C$	260	500	750	
$K_1$	$I_{OUT}/I_{SENSE}$	$I_{OUT}=0.35A; V_{SENSE}=0.5V; V_{CSD}=0V;$ $T_j= -40^{\circ}C...150^{\circ}C$	320	450	590	
		$I_{OUT}=0.35A; V_{SENSE}=0.5V; V_{CSD}=0V;$ $T_j= 25^{\circ}C...150^{\circ}C$	360	450	540	
$dK_1/K_1^{(1)}$	Current sense ratio drift	$I_{OUT}=0.35A; V_{SENSE}=0.5V;$ $V_{CSD}=0V;$ $T_j= -40^{\circ}C \text{ to } 150^{\circ}C$	-13		+13	%
$K_2$	$I_{OUT}/I_{SENSE}$	$I_{OUT}=0.5A; V_{SENSE}=4V; V_{CSD}=0V;$ $T_j= -40^{\circ}C...150^{\circ}C$	360	440	540	
		$I_{OUT}=0.5A; V_{SENSE}=4V; V_{CSD}=0V;$ $T_j= 25^{\circ}C...150^{\circ}C$	380	440	510	
$dK_2/K_2^{(1)}$	Current sense ratio drift	$I_{OUT}=0.5 A; V_{SENSE}=4 V;$ $V_{CSD}=0V;$ $T_j= -40^{\circ}C \text{ to } 150^{\circ}C$	-8		+8	%
$K_3$	$I_{OUT}/I_{SENSE}$	$I_{OUT}=1.5A; V_{SENSE}=4V; V_{CSD}=0V;$ $T_j= -40^{\circ}C...150^{\circ}C$	410	440	480	
		$I_{OUT}=1.5A; V_{SENSE}=4V; V_{CSD}=0V;$ $T_j= 25^{\circ}C...150^{\circ}C$	420	440	460	
$dK_3/K_3^{(1)}$	Current sense ratio drift	$I_{OUT}=1.5 A; V_{SENSE}=4 V;$ $V_{CSD}=0V;$ $T_j= -40^{\circ}C \text{ to } 150^{\circ}C$	-4		+4	%
$I_{SENSE0}$	Analog sense leakage current	$I_{OUT}=0A; V_{SENSE}=0V;$ $V_{CSD}=5V; V_{IN}=0V; T_j= -40^{\circ}C...150^{\circ}C$	0		1	$\mu A$
		$V_{CSD}=0V; V_{IN}=5V; T_j= -40^{\circ}C...150^{\circ}C$	0		2	$\mu A$
$I_{OL}$	Openload ON state current detection threshold	$V_{IN}=5V, I_{SENSE}=5 \mu A$	1		5	$mA$
$V_{SENSE}$	Max analog senseoutput voltage	$I_{OUT}=1.5A; V_{CSD}=0V;$	5			$V$
$V_{SENSEH}$	Analog sense output voltage in overtemperature condition	$V_{CC}=13V; R_{SENSE}=3.9K\Omega;$		9		$V$
$I_{SENSEH}$	Analog sense output current in overtemperature condition	$V_{CC}=13V; V_{SENSE}=5V;$		8		$mA$

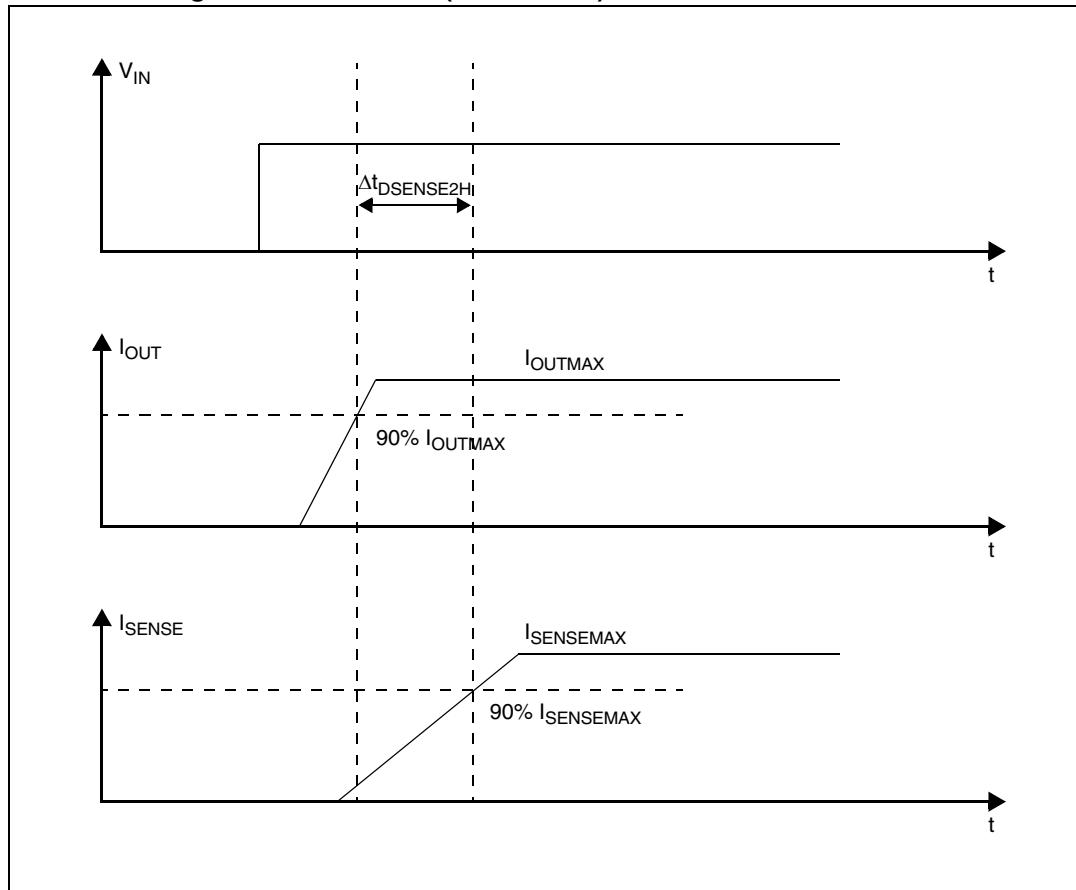
**Table 10. Current sense (8V<VCC<16V) (continued)**

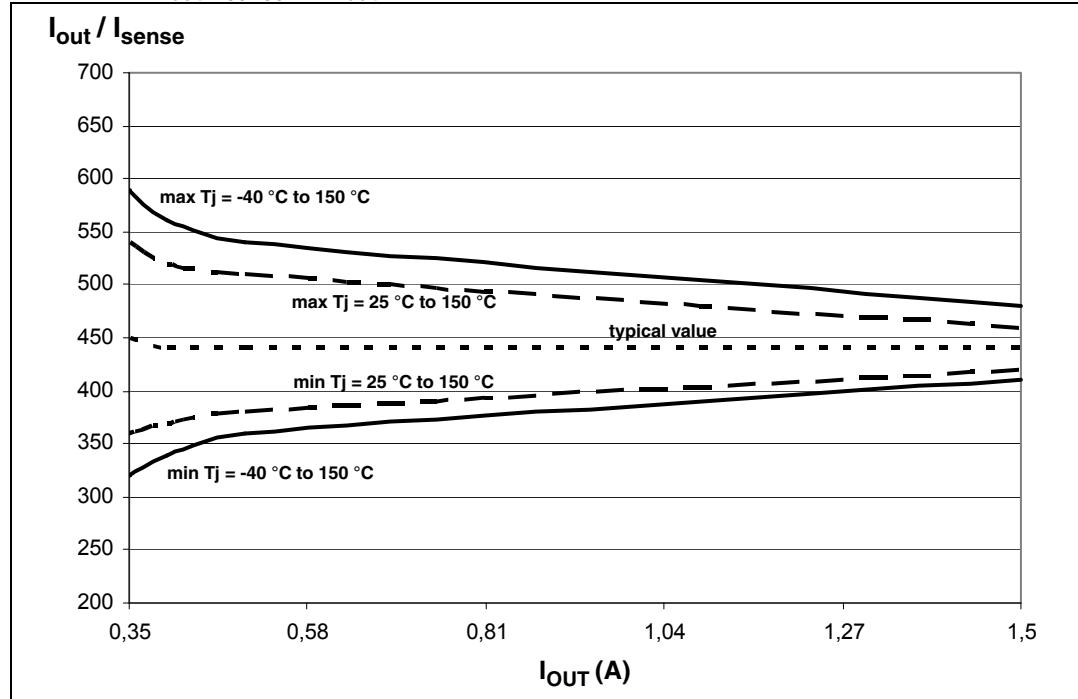
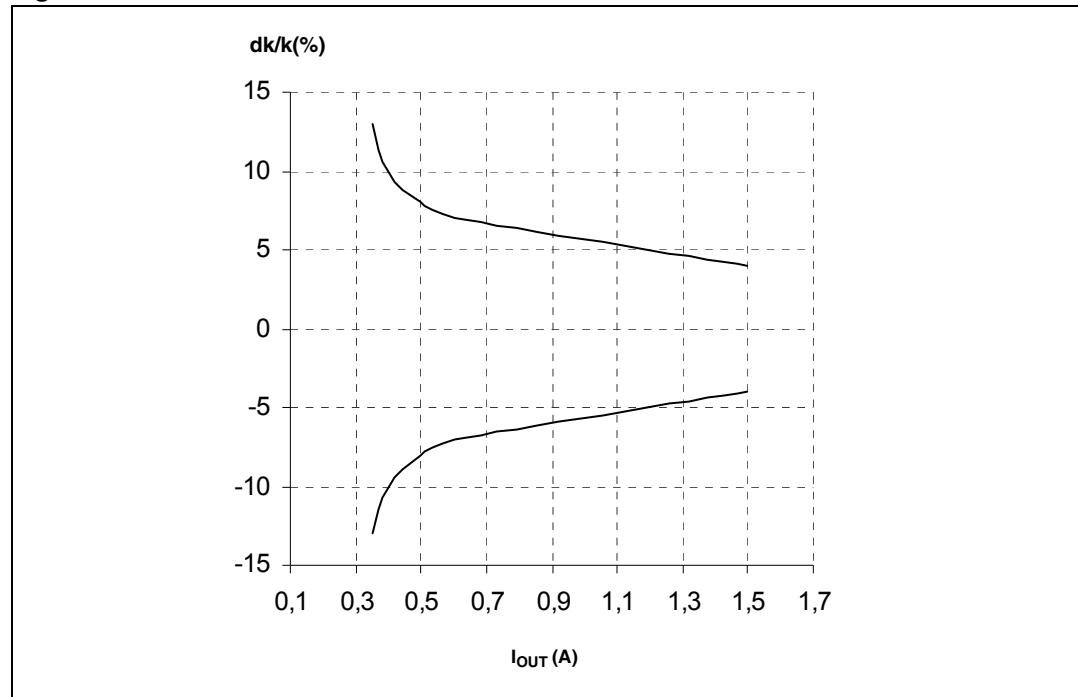
Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$t_{DSENSE1H}$	Delay response time from falling edge of CS_DIS pin	$V_{SENSE} < 4V, 0.08A < I_{out} < 1.5A$ $I_{SENSE} = 90\% \text{ of } I_{SENSE \text{ max}}$ (see <a href="#">Figure 4</a> .)		50	100	$\mu s$
$t_{DSENSE1L}$	Delay response time from rising edge of CS_DIS pin	$V_{SENSE} < 4V, 0.08A < I_{out} < 1.5A$ $I_{SENSE} = 10\% \text{ of } I_{SENSE \text{ max}}$ (see <a href="#">Figure 4</a> .)		5	20	$\mu s$
$t_{DSENSE2H}$	Delay response time from rising edge of INPUT pin	$V_{SENSE} < 4V, 0.08A < I_{out} < 1.5A$ $I_{SENSE} = 90\% \text{ of } I_{SENSE \text{ max}}$ (see <a href="#">Figure 4</a> .)		80	150	$\mu s$
$\Delta t_{DSENSE2H}$	Delay response time between rising edge of output current and rising edge of current sense	$V_{SENSE} < 4V,$ $I_{SENSE} = 90\% \text{ of } I_{SENSE MAX},$ $I_{OUT} = 90\% \text{ of } I_{OUT MAX}$ $I_{OUT MAX} = 2A$ (see <a href="#">Figure 5</a> )			20	$\mu s$
$t_{DSENSE2L}$	Delay response time from falling edge of INPUT pin	$V_{SENSE} < 4V, 0.08A < I_{out} < 1.5A$ $I_{SENSE} = 10\% \text{ of } I_{SENSE \text{ max}}$ (see <a href="#">Figure 4</a> .)		100	250	$\mu s$

1. Parameter guaranteed by design; it is not tested.

**Figure 4. Current sense delay characteristics**

**Figure 5. Delay response time between rising edge of output current and rising edge of current sense (CS enabled)**



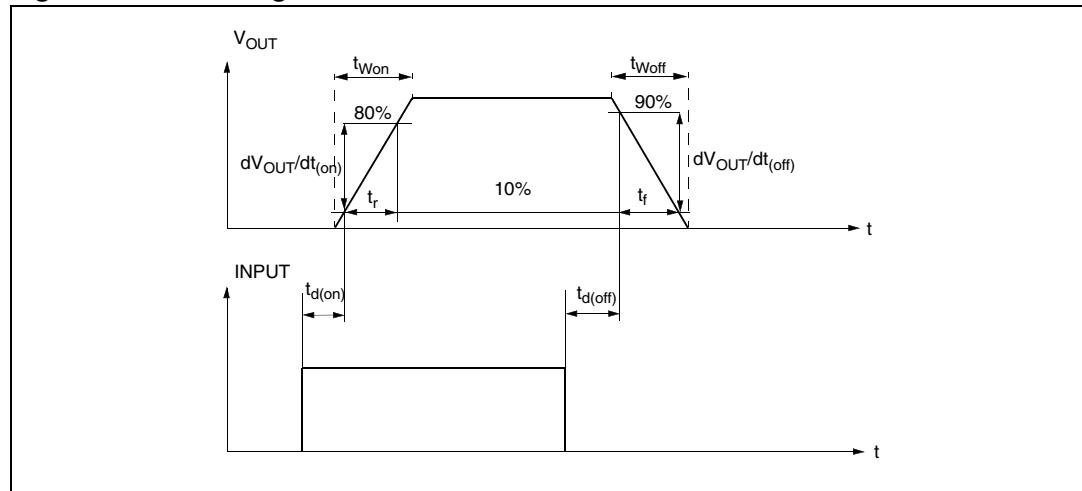
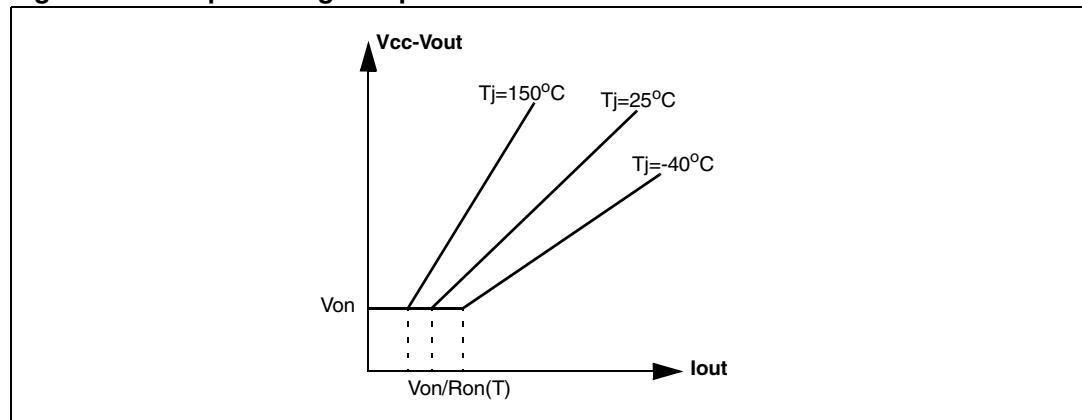
**Figure 6.**  $I_{out} / I_{sense}$  vs.  $I_{out}$  (see [Table 10](#) for details)**Figure 7.** Maximum current sense ratio drift vs load current

Note: Parameter guaranteed by design; it is not tested.

**Table 11. Truth table**

Conditions	INPUT	OUTPUT	SENSE ( $V_{CSD}=0V$ ) <sup>(1)</sup>
Normal operation	L	L	0
	H	H	Nominal
Overtemperature	L	L	0
	H	L	$V_{SENSEH}$
Undervoltage	L	L	0
	H	L	0
Short circuit to GND ( $R_{SC} \leq 10 \text{ m}\Omega$ )	L	L	0
	H	L	0 if $T_j < T_{TSD}$
	H	L	$V_{SENSEH}$ if $T_j > T_{TSD}$
Short circuit to $V_{CC}$	L	H	0
	H	H	< Nominal
Negative output voltage clamp	L	L	0

1. If the  $V_{CSD}$  is high, the SENSE output is at a high impedance, its potential depends on leakage currents and external circuit.

**Figure 8. Switching characteristics****Figure 9. Output voltage drop limitation**

**Table 12. Electrical transient requirements**

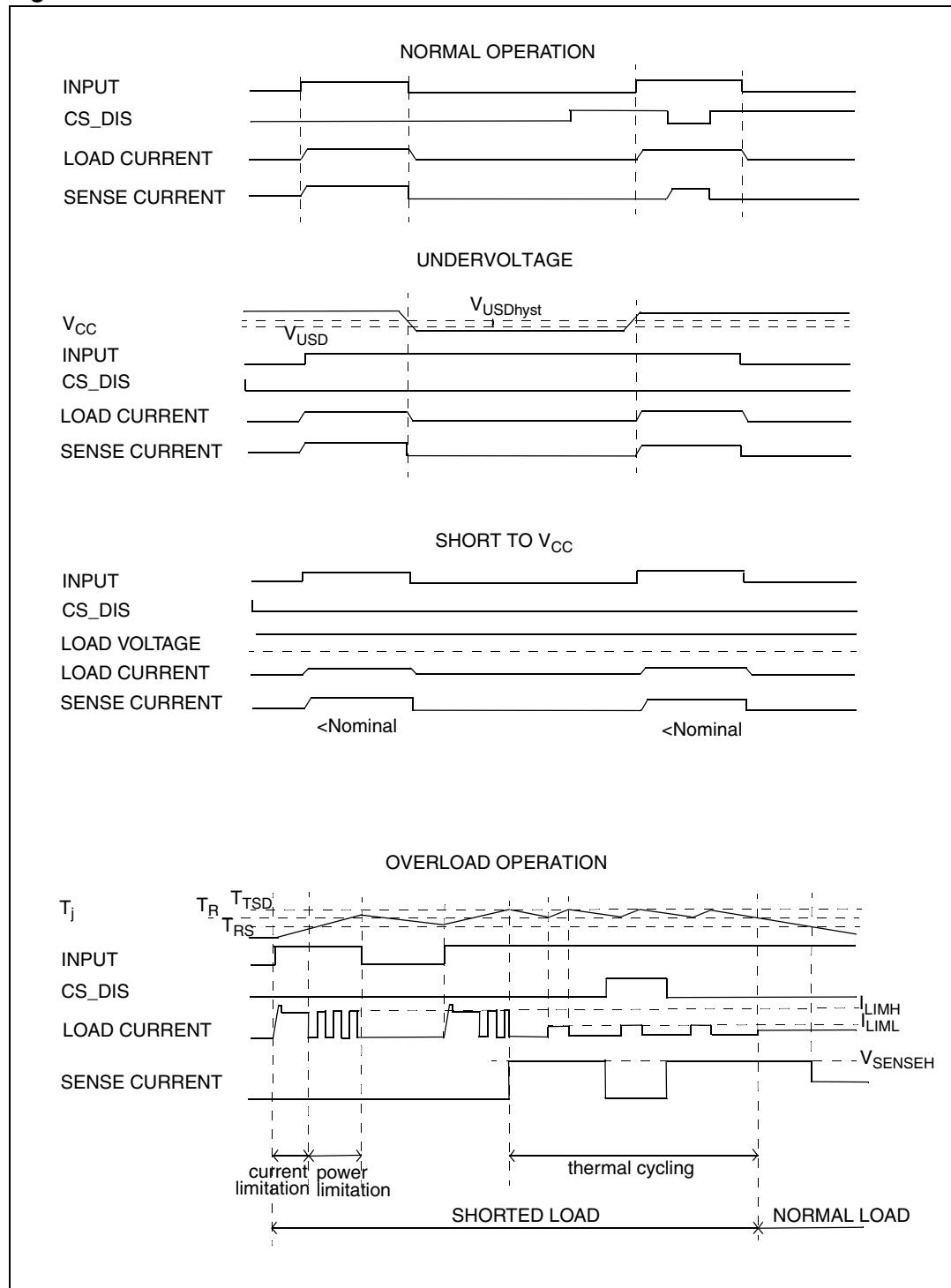
ISO 7637-2: 2004(E) Test pulse	Test levels <sup>(1)</sup>		Number of pulses or test times	Burst cycle/pulse repetition time		Delays and impedance
	III	IV				
1	-75V	-100V	5000 pulses	0.5 s	5 s	2 ms, 10 Ω
2a	+37V	+50V	5000 pulses	0.2 s	5 s	50 μs, 2 Ω
3a	-100V	-150V	1h	90 ms	100 ms	0.1 μs, 50 Ω
3b	+75V	+100V	1h	90 ms	100 ms	0.1 μs, 50 Ω
4	-6V	-7V	1 pulse			100 ms, 0.01 Ω
5b <sup>(2)</sup>	+65V	+87V	1 pulse			400 ms, 2 Ω

ISO 7637-2: 2004(E) Test pulse	Test level results <sup>(1)</sup>	
	III	IV
1	C	C
2a	C	C
3a	C	C
3b	C	C
4	C	C
5b <sup>(2)</sup>	C	C

1. The above test levels must be considered referred to Vcc = 13.5V except for pulse 5b.
2. Valid in case of external load dump clamp: 40V maximum referred to ground.

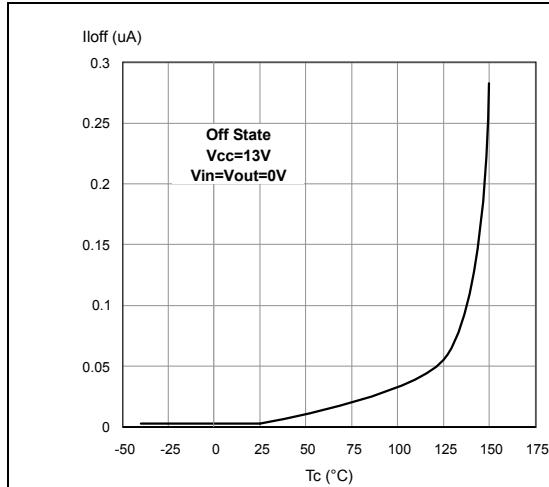
Class	Contents
C	All functions of the device are performed as designed after exposure to disturbance.
E	One or more functions of the device are not performed as designed after exposure to disturbance and cannot be returned to proper operation without replacing the device.

Figure 10. Waveforms

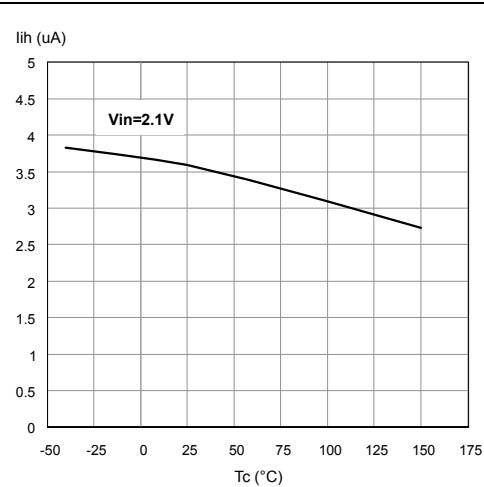


## 2.4 Electrical characteristics curves

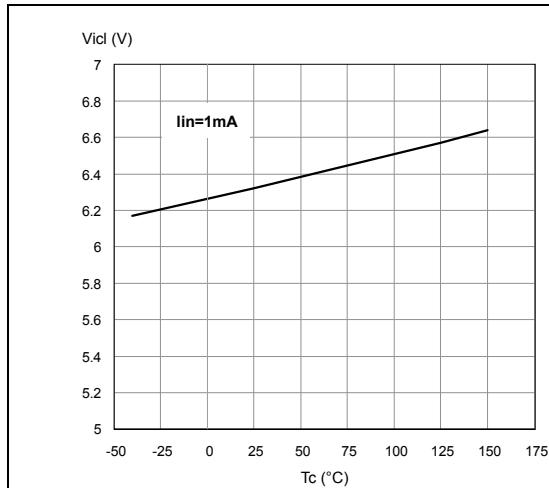
**Figure 11. Off state output current**



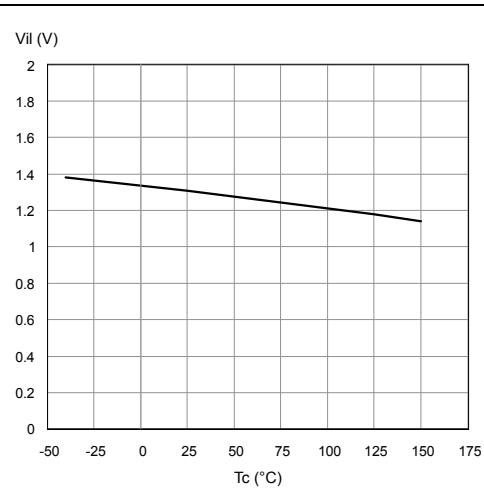
**Figure 12. High level input current**



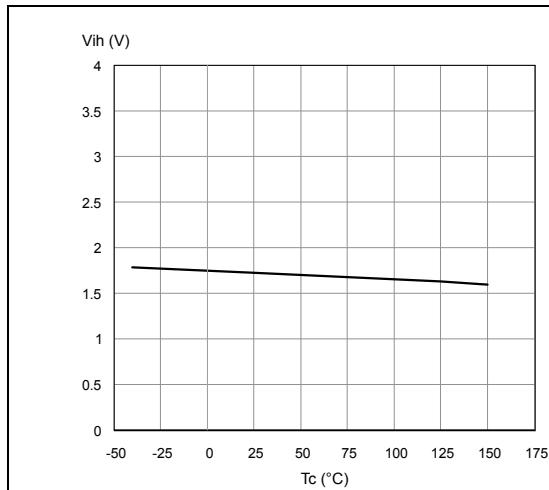
**Figure 13. Input clamp voltage**



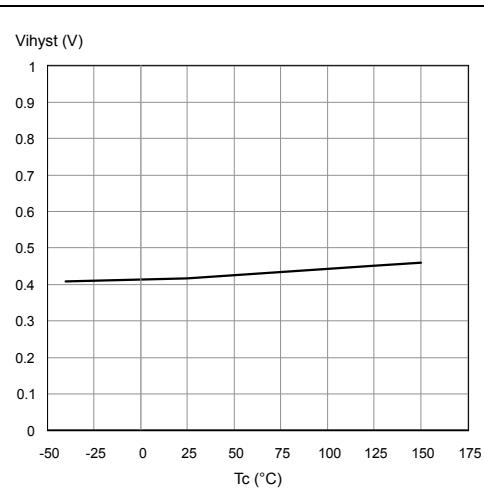
**Figure 14. Input low level**

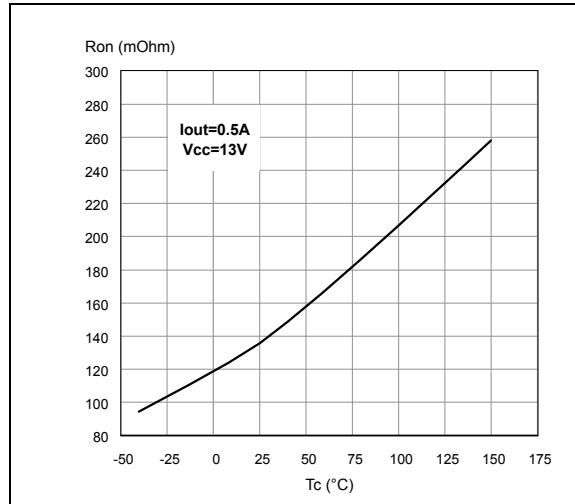
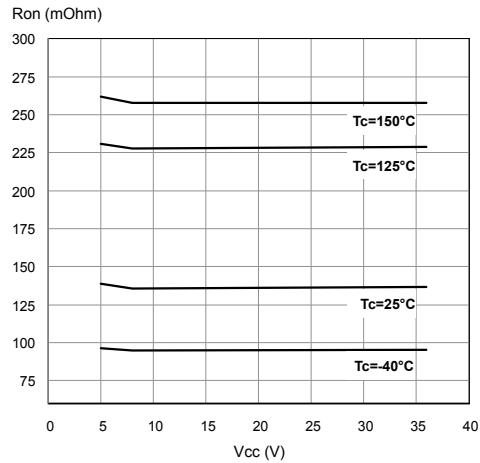
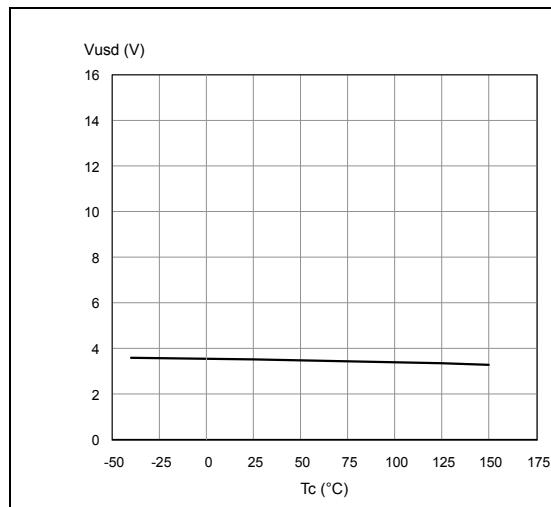
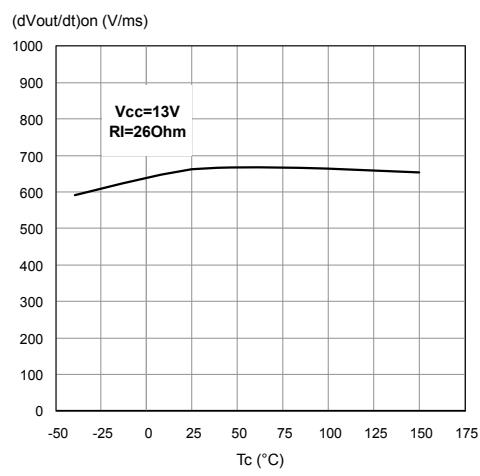
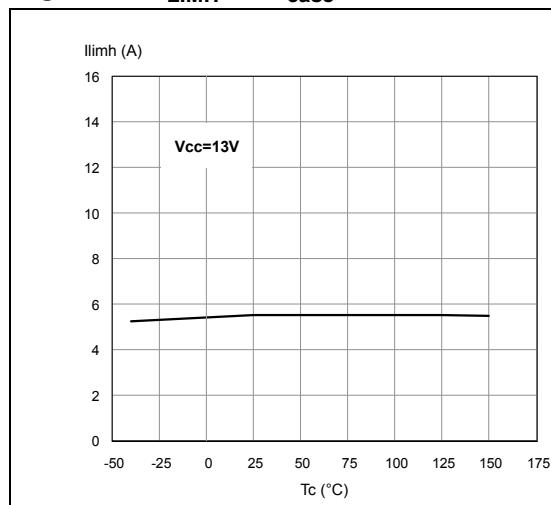
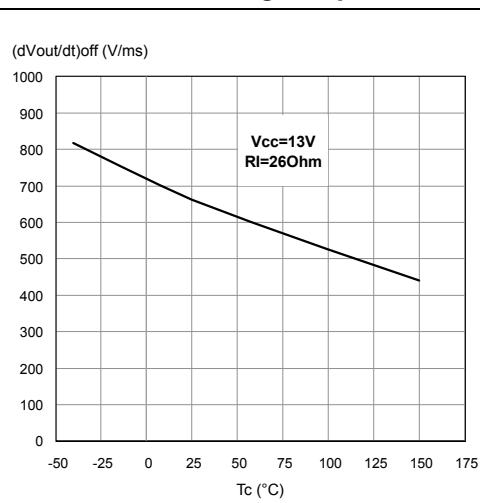


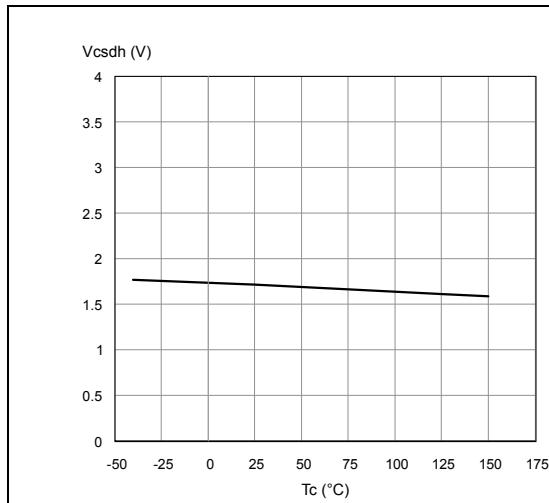
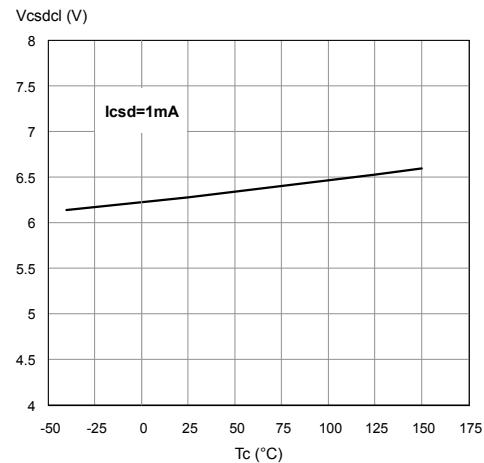
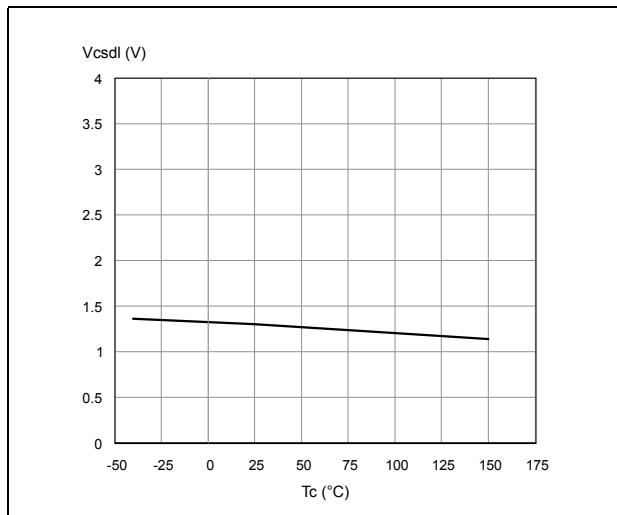
**Figure 15. Input high level**



**Figure 16. Input hysteresis voltage**

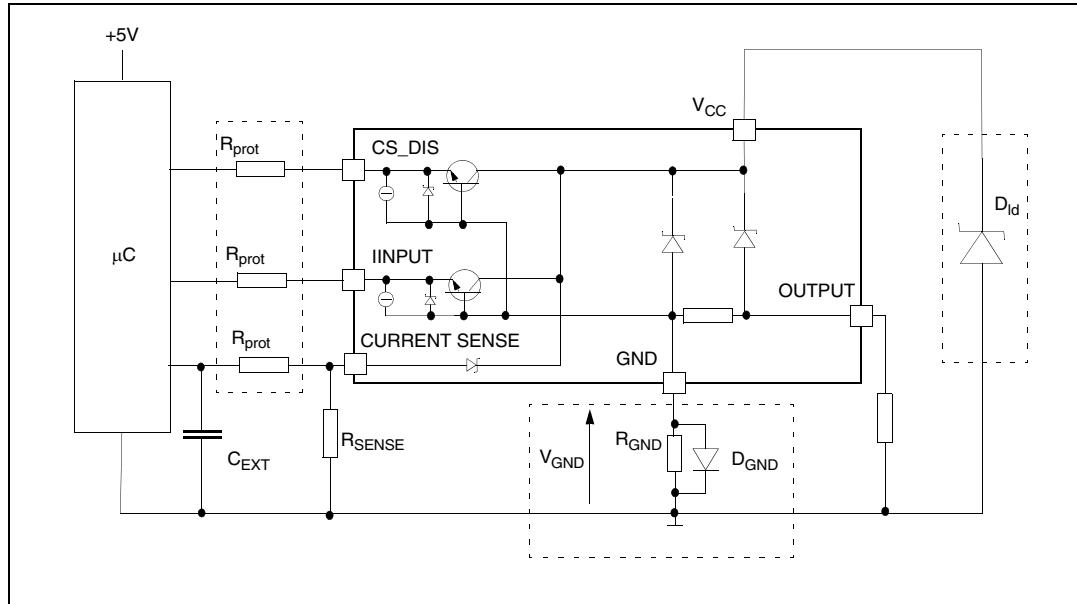


**Figure 17. On state resistance vs.  $T_{case}$** **Figure 18. On state resistance vs.  $V_{CC}$** **Figure 19. Undervoltage shutdown****Figure 20. Turn-On voltage slope****Figure 21.  $I_{LIMH}$  vs.  $T_{case}$** **Figure 22. Turn-Off voltage slope**

**Figure 23. CS\_DIS high level voltage****Figure 24. CS\_DIS clamp voltage****Figure 25. CS\_DIS low level voltage**

### 3 Application information

**Figure 26. Application schematic**



Note: Channel 2 has the same internal circuit as channel 1.

#### 3.1 GND protection network against reverse battery

##### 3.1.1 Solution 1 : resistor in the ground line ( $R_{GND}$ only)

This can be used with any type of load.

The following is an indication on how to dimension the  $R_{GND}$  resistor.

1.  $R_{GND} \leq 600\text{mV} / (I_{S(on)\max})$ .
2.  $R_{GND} \geq (-V_{CC}) / (-I_{GND})$

where  $-I_{GND}$  is the DC reverse ground pin current and can be found in the absolute maximum rating section of the device datasheet.

Power Dissipation in  $R_{GND}$  (when  $V_{CC}<0$ : during reverse battery situations) is:

$$P_D = (-V_{CC})^2 / R_{GND}$$

This resistor can be shared amongst several different HSDs. Please note that the value of this resistor should be calculated with formula (1) where  $I_{S(on)\max}$  becomes the sum of the maximum on-state currents of the different devices.

Please note that if the microprocessor ground is not shared by the device ground then the  $R_{GND}$  will produce a shift ( $I_{S(on)\max} * R_{GND}$ ) in the input thresholds and the status output values. This shift will vary depending on how many devices are ON in the case of several high side drivers sharing the same  $R_{GND}$ .

If the calculated power dissipation leads to a large resistor or several devices have to share the same resistor then ST suggests to utilize Solution 2 (see below).

### 3.1.2 Solution 2 : diode ( $D_{GND}$ ) in the ground line

A resistor ( $R_{GND}=1\text{k}\Omega$ ) should be inserted in parallel to  $D_{GND}$  if the device drives an inductive load.

This small signal diode can be safely shared amongst several different HSDs. Also in this case, the presence of the ground network will produce a shift ( $\approx 600\text{mV}$ ) in the input threshold and in the status output values if the microprocessor ground is not common to the device ground. This shift will not vary if more than one HSD shares the same diode/resistor network.

## 3.2 Load dump protection

$D_{ld}$  is necessary (Voltage Transient Suppressor) if the load dump peak voltage exceeds the  $V_{CC}$  max DC rating. The same applies if the device is subject to transients on the  $V_{CC}$  line that are greater than the ones shown in the ISO 7637-2: 2004(E) table.

## 3.3 MCU I/Os protection

If a ground protection network is used and negative transient are present on the  $V_{CC}$  line, the control pins will be pulled negative. ST suggests to insert a resistor ( $R_{prot}$ ) in line to prevent the MCU I/Os pins to latch-up.

The value of these resistors is a compromise between the leakage current of MCU and the current required by the HSD I/Os (Input levels compatibility) with the latch-up limit of MCU I/Os.

$$-V_{CCpeak}/I_{latchup} \leq R_{prot} \leq (V_{OH\mu C} - V_{IH} - V_{GND}) / I_{IHmax}$$

Calculation example:

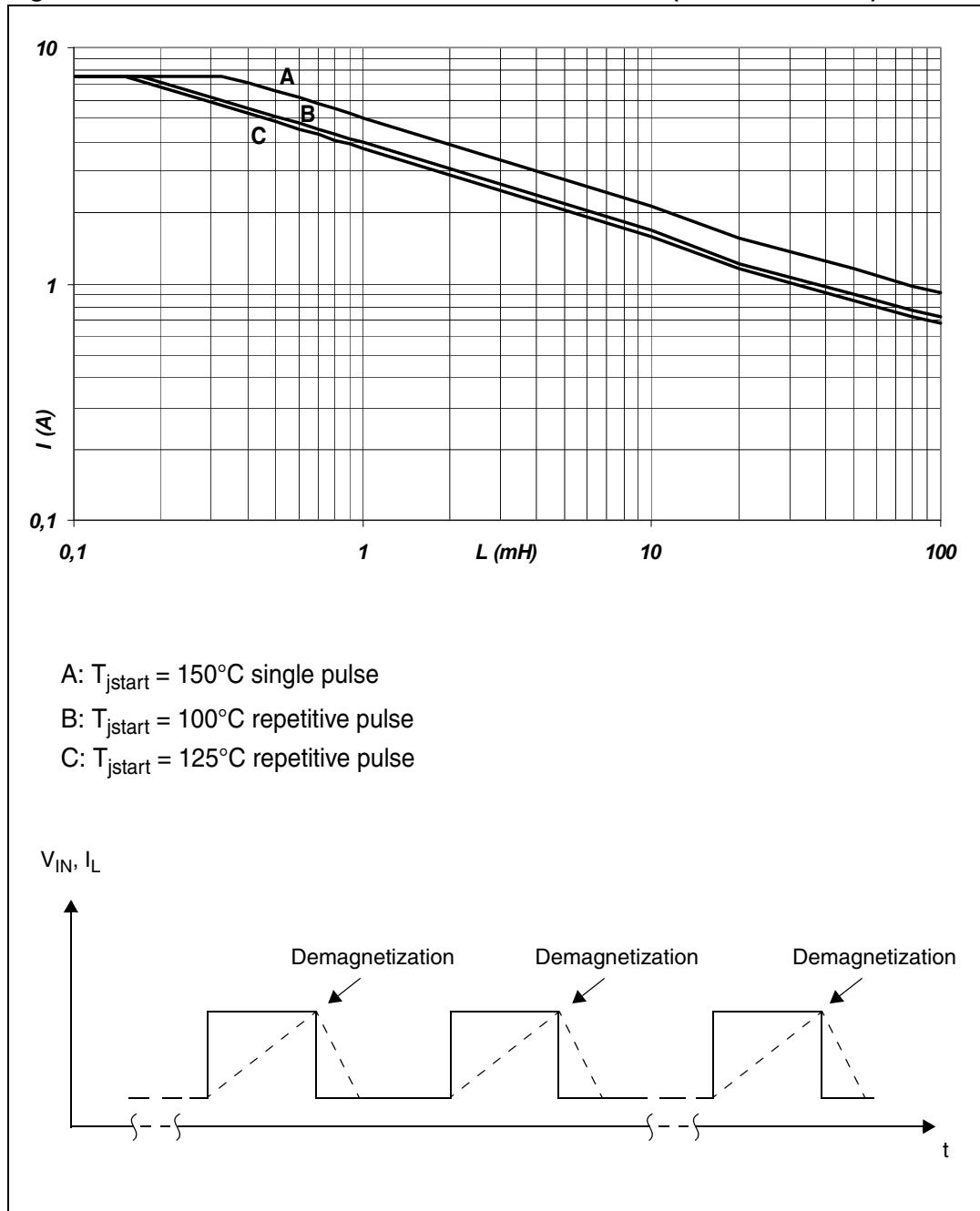
For  $V_{CCpeak} = -100\text{V}$  and  $I_{latchup} \geq 20\text{mA}$ ;  $V_{OH\mu C} \geq 4.5\text{V}$

$$5\text{k}\Omega \leq R_{prot} \leq 180\text{k}\Omega$$

Recommended values:  $R_{prot} = 10\text{k}\Omega$ ,  $C_{EXT} = 10\text{nF}$ .

### 3.4 Maximum demagnetization energy ( $V_{CC} = 13.5V$ )

Figure 27. Maximum turn-Off current versus inductance (for each channel)



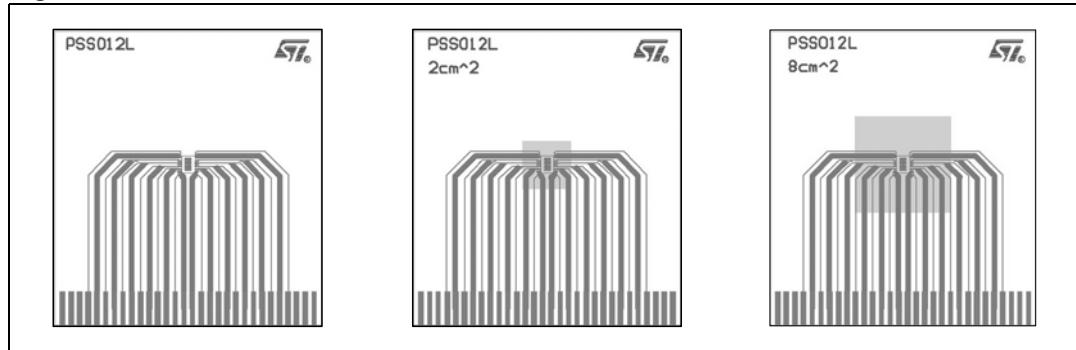
Note: Values are generated with  $R_L = 0 \Omega$ .

In case of repetitive pulses,  $T_{jstart}$  (at beginning of each demagnetization) of every pulse must not exceed the temperature specified above for curves A and B.

## 4 Package and PC board thermal data

### 4.1 PowerSSO-12™ thermal data

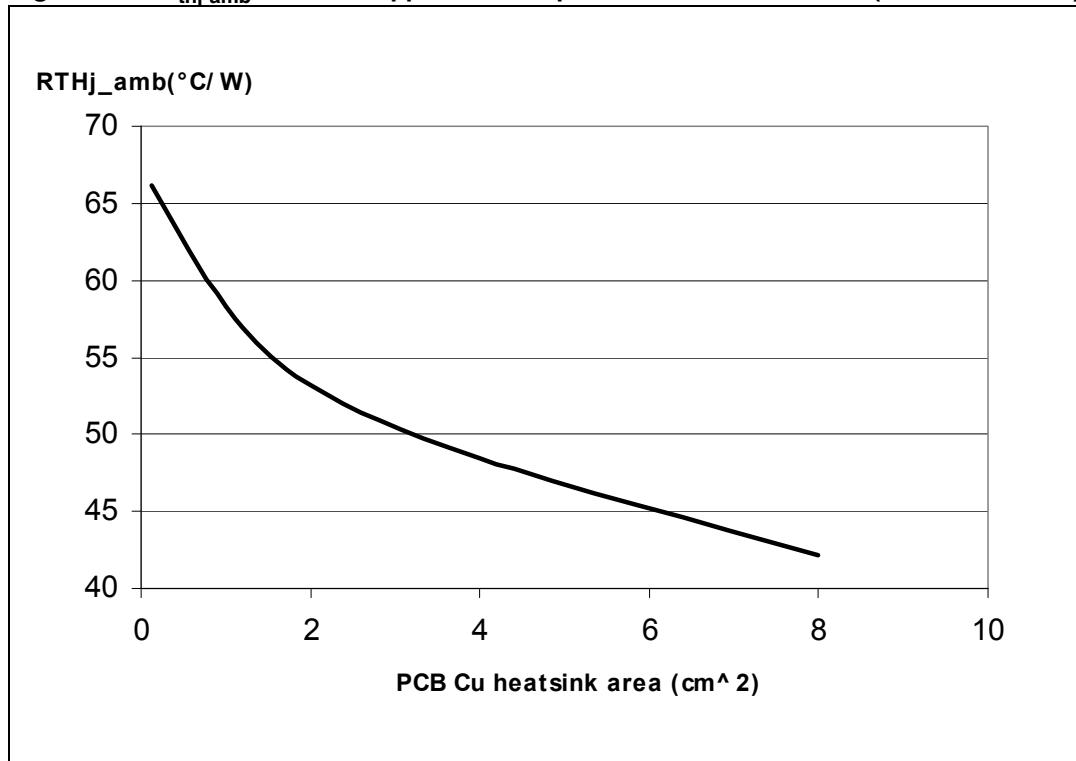
Figure 28. PowerSSO-12™ PC board



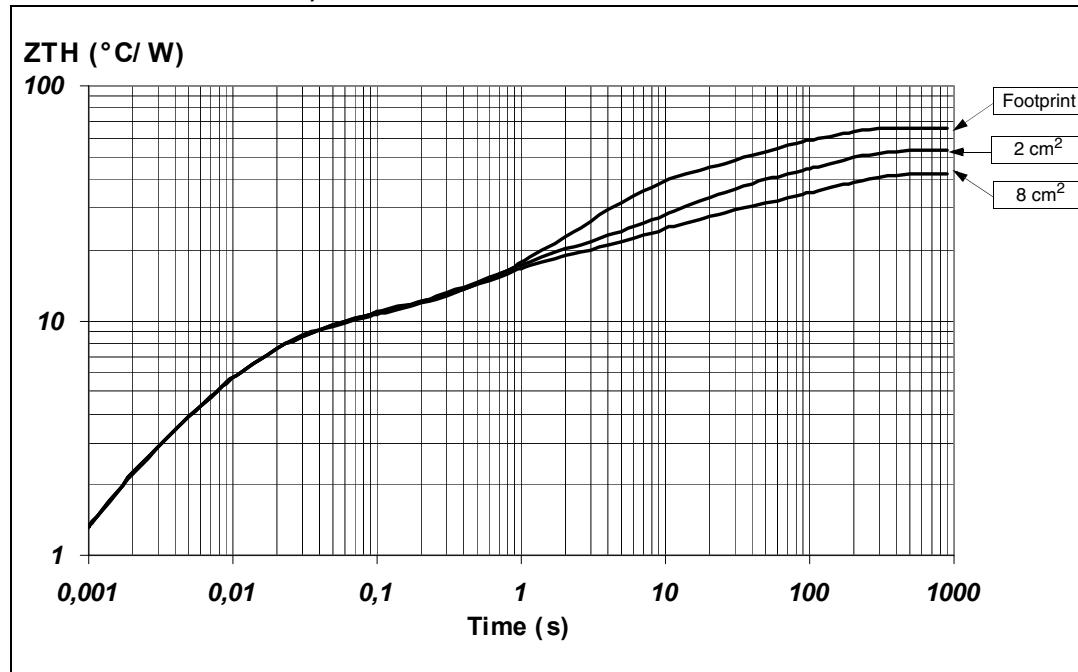
Note:

Layout condition of  $R_{th}$  and  $Z_{th}$  measurements (PCB: Double layer, Thermal Vias, FR4 area= 77mm x 86mm, PCB thickness=1.6mm, Cu thickness=70 $\mu$ m (front and back side), Copper areas: from minimum pad lay-out to 8cm $^2$ ).

Figure 29.  $R_{thj\_amb}$  vs. PCB copper area in open box free air condition (one channel ON)



**Figure 30. PowerSSO-12™ thermal impedance junction ambient single pulse (one channel ON)**

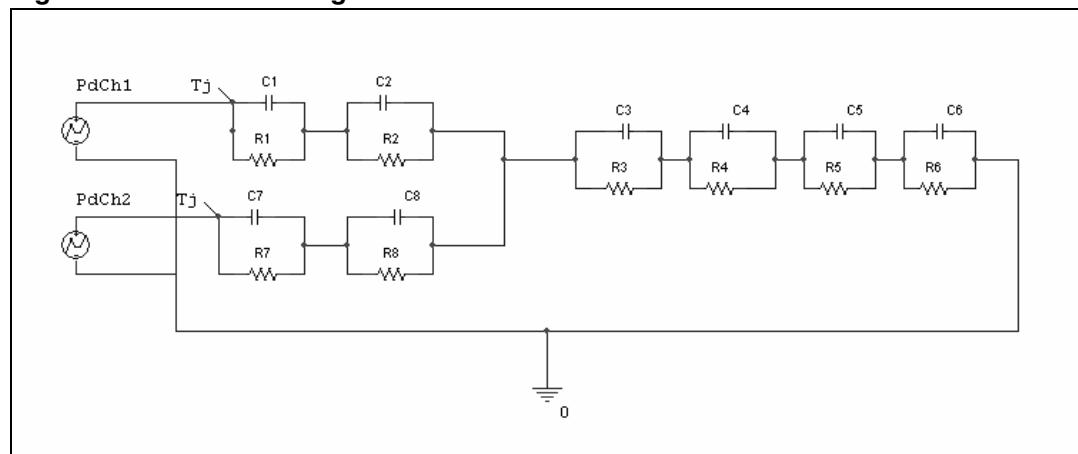


**Equation 1: pulse calculation formula**

$$Z_{TH\delta} = R_{TH} \cdot \delta + Z_{THtp}(1 - \delta)$$

where  $\delta = t_p/T$

**Figure 31. Thermal fitting model of a double channel HSD in PowerSSO-12™ (a)**



- a. The fitting model is a simplified thermal tool and is valid for transient evolutions where the embedded protections (power limitation or thermal cycling during thermal shutdown) are not triggered.

**Table 13. Thermal parameters**

Area/island (cm <sup>2</sup> )	Footprint	2	8
R1= R7 (°C/W)	1.2		
R2= R8 (°C/W)	6		
R3 (°C/W)	3		
R4 (°C/W)	8	8	7
R5 (°C/W)	22	15	10
R6 (°C/W)	26	20	15
C1= C7 (W.s/°C)	0.0008		
C2= C8 (W.s/°C)	0.0016		
C3 (W.s/°C)	0.0166		
C4 (W.s/°C)	0.2	0.1	0.1
C5 (W.s/°C)	0.27	0.8	1
C6 (W.s/°C)	3	6	9

## 5 Package and packing information

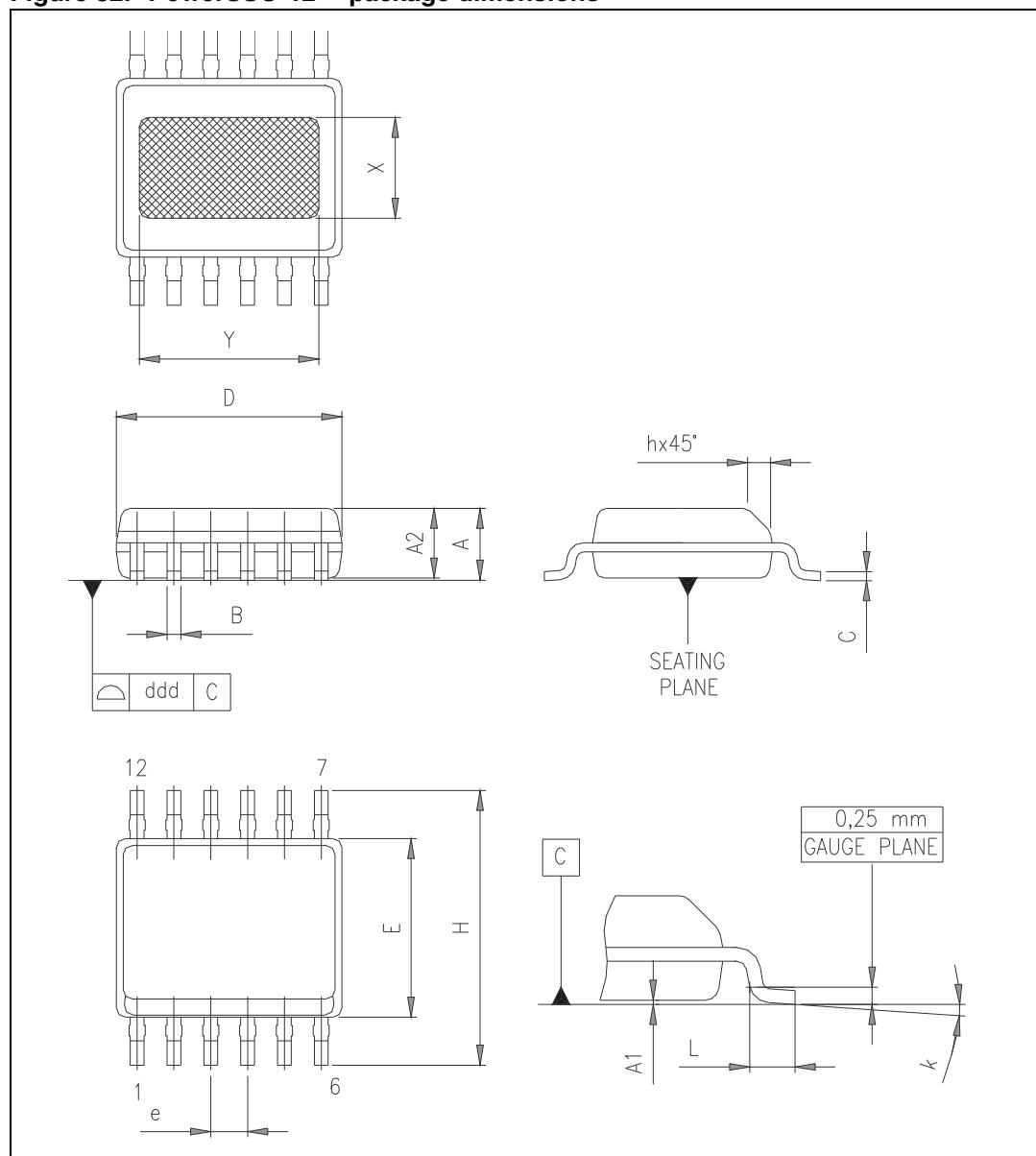
### 5.1 ECOPACK® packages

In order to meet environmental requirements, ST offers these devices in ECOPACK® packages. These packages have a Lead-free second-level interconnect. The category of Second-Level Interconnect is marked on the package and on the inner box label, in compliance with JEDEC Standard JESD97.

The maximum ratings related to soldering conditions are also marked on the inner box label. ECOPACK is an ST trademark. ECOPACK specifications are available at: [www.st.com](http://www.st.com).

### 5.2 Package mechanical data

**Figure 32. PowerSSO-12™ package dimensions**

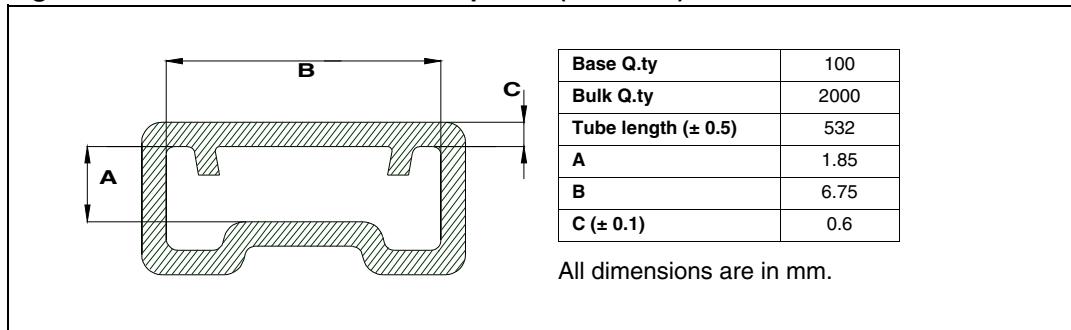


**Table 14. PowerSSO-12™ mechanical data**

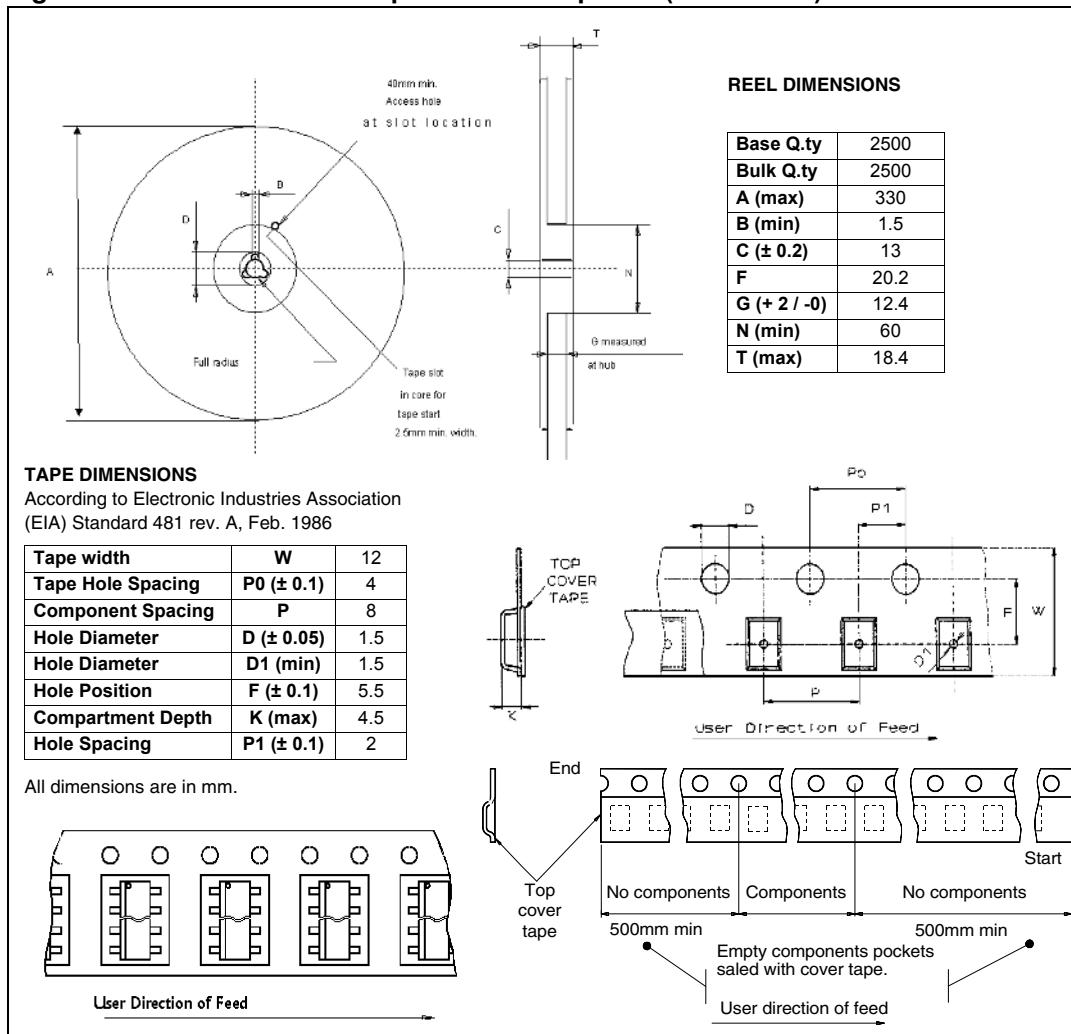
Symbol	Millimeters		
	Min.	Typ.	Max.
A	1.250		1.620
A1	0.000		0.100
A2	1.100		1.650
B	0.230		0.410
C	0.190		0.250
D	4.800		5.000
E	3.800		4.000
e		0.800	
H	5.800		6.200
h	0.250		0.500
L	0.400		1.270
k	0°		8°
X	2.200		2.800
Y	2.900		3.500
ddd			0.100

## 5.3 Packing information

**Figure 33. PowerSSO-12™ tube shipment (no suffix)**



**Figure 34. PowerSSO-12™ tape and reel shipment (suffix “TR”)**



## 6 Revision history

**Table 15. Document revision history**

Date	Revision	Changes
13-Sep-2004	1	Initial release.
10-Apr-2006	2	Layout changed. Major update to <a href="#">Section 2: Electrical specifications</a> .
01-Mar-2007	3	Reformatted. Contents, List of tables and List of figures added. Added <a href="#">Section 3.4: Maximum demagnetization energy (VCC = 13.5V)</a> . ECOPACK® package information added.
10-Dec-2007	4	Document reformatted and restructured. Updated <a href="#">Figure 2: Configuration diagram (top view)</a> : pins 7-12 left unconnected (N.C) and added note. <a href="#">Table 4: Absolute maximum ratings</a> : corrected $E_{MAX}$ value from 14 to 34 mJ. Added <a href="#">Figure 5: Delay response time between rising edge of output current and rising edge of current sense (CS enabled)</a> . Updated <a href="#">Figure 6: Iout/ Isense vs. Iout</a> (see Table 10 for details). Added <a href="#">Figure 7: Maximum current sense ratio drift vs load current</a> . Updated <a href="#">Table 10: Current sense (8V&lt;VCC&lt;16V)</a> : – changed $t_{DSENSE2H}$ max value from 300 $\mu$ s to 150 $\mu$ s. – added dk1/k1, dk2/k2, dk3/k3, $\Delta t_{DSENSE2H}$ , $I_{OL}$ parameters. <a href="#">Table 12: Electrical transient requirements</a> : updated test level values III and IV for test pulse 5b and notes. Updated <a href="#">Section 4.1: PowerSSO-12™ thermal data</a> : – changed <a href="#">Figure 29: Rthj-amb vs. PCB copper area in open box free air condition (one channel ON)</a> . – changed <a href="#">Figure 30: PowerSSO-12™ thermal impedance junction ambient single pulse (one channel ON)</a> . – <a href="#">Figure 31: Thermal fitting model of a double channel HSD in PowerSSO-12™</a> : added note. – updated <a href="#">Table 13: Thermal parameters</a> : $R_3$ value changed from 7 to 3 °C/W. $R_4$ values changed from 10 /10 /9 to 8 /8 /7 °C/W. $C_3$ value changed from 0.05 to 0.0166 W.s/°C.
12-Feb-2008	5	Corrected typing error in <a href="#">Table 10: Current sense (8V&lt;VCC&lt;16V)</a> : changed $I_{OL}$ test condition from $V_{IN} = 0V$ to $V_{IN} = 5V$ .
24-Sep-2013	6	Updated disclaimer.

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