MAX21000 USER GUIDE

Revision 2.6, May 2015

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2 Introduction

MEMS sensors are revolutionizing the way people interact with everyday technology, making it easier and more user-friendly. Maxim can leverage its analog integration expertise to develop and manufacture new breakthrough MEMS sensors being smaller, lower power and more accurate than ever.

Owning the entire supply chain, Maxim brings its customers complete, reliable and cost-effective solutions, ensuring prompt time-to-volume and time-to-market to effectively address high-volume applications in consumer and industrial market segments.

Thanks to its leadership in analog integration and its manufacturing experience in MEMS, Maxim is capable of high-volume production to meet the market's demands. Maxim's manufacturing expertise and highest quality standards also guarantee high performance and product reliability.

Every MEMS sensor is tested and trimmed in factory so that for most consumer applications, no additional sensor calibrations are required. The end user can quickly verify the sensor's operation without physically tilting or rotating the sensor thanks to the built-in self-test feature, which allows accelerating the time-to-market for mass production.

This User Guide will provide a clear picture of the guidelines for its use in consumer applications and a comprehensive description of his unique features. The final section of this guide will present the structure of the register file, the purpose of each field or every register, including two examples about typical programming sequences.

3 Nomenclature

- ODR Output Data Rate
- BW Bandwidth
- FS Fulllscale
- UI User Interface
- OIS Optical Image Stabilization
- MSB Most Significant Bit or Byte
- LSB Least Significant Bit or Byte
- HPF Highpass Filter
- LPF Lowpass Filter
- dps Degrees per seconds
- RFU Reserved for future uses

4 MAX21000 Description

The MAX21000 is a low-power, low-noise, three-axis angular rate sensor able to offer unprecedented accuracy and sensitivity over temperature and time.

It is capable of working with a supply voltage as low as 1.71V for minimum power consumption. It includes a sensing element and an IC interface capable of providing the measured angular rate to the external world (I^2C/SPI) .

The MAX21000 has a configurable full scale of ±31.25/±62.5/±125/±250/±500/±1000/±2000dps and is capable of measuring rates with a finely tunable user-selectable bandwidth. The high output data rate (ODR) and the large bandwidth (BW), together with the low phase delay, make the MAX21000 suitable for both user interface (UI) and optical image stabilization (OIS) applications.

The MAX21000 is a highly integrated solution requiring only two external capacitors, available in a compact 3mm x 3mm x 0.9mm plastic land grid array (LGA) package and can operate within a temperature range of -40°C to +85°C.

Figure 1: Block Diagram

5 Pin Description

Table 1: Pin Description

6 I 2 C Interface

To connect a MAX21000 device to an I^2C master, the SDA_SDI_O pin of the MAX21000 device must be connected to the SDA pin of the I²C master and the SCL_CLK pin of the MAX21000 device must be connected to the SCL pin of the I^2C master. Both SDA and SCL lines must be connected to a pullup resistor. The SA0_SDO pin must be connected to VDD or GND to configure the MAX21000 I^2C slave address (see Table 3: I^2C Device Addresses).

Figure 2: I² C Interface Connection to an Application Processor

6.1 I 2 C Protocol

To start an I²C request, the master sends a START condition (S), followed by the MAX21000's I²C address. Then, the master sends the address of the register to be programmed. The master then terminates the communication by issuing a STOP condition (P) to relinquish the control of the bus, or a repeated START condition (Sr) to keep controlling it.

Figure 3: START (S), STOP (P), and Repeated START (Sr) Conditions

6.2 Slave Address

The slave address is used to identify the MAX21000 device in I^2C communications. The address is defined as the seven most significant bits followed by the read/write bit. Set the read/write bit to 1 to request a read operation, or 0 to request a write operation (see [Table 3\)](#page-12-3).

Table 3: I² C Device Addresses

6.3 Acknowledge

The acknowledge bit is sent after every byte. This bit allows the receiver to notify the transmitter that the byte has been received correctly and another byte may be sent. The master generates all clock pulses, including the acknowledge's ninth clock pulse (8 bits of data + ACK).

To allow the receiver to send the acknowledge, the transmitter releases the SDA line during the acknowledge pulse, so that the receiver can pull the SDA line LOW during the ninth clock pulse to signal an Acknowledge (ACK), or release the SDA line (HIGH) to signal a Not Acknowledge (NACK).

The NACK is sent if the device is busy or a system fault occurs. It is also used by the master to signal the end of the transfer during a read operation.

6.4 Register Address

The I²C register address for the MAX21000 is composed by 6 bits of address and 1 bit (if parity) whose meaning can be configured as:

Auto-increment: if 0, in case of burst operation the initial register address is auto-incremented after every data byte; if 1, the operation is executed always on the same register; **Even parity** this bit represents the even parity computed on the 6 bits of the register address; **Odd Parity** this bit represents the odd parity computed on the 6 bits of the register address;

6.5 I 2 C Operations

6.5.1 Write One Byte

To write one byte, the following steps must be executed:

- 1: The master sends a START condition.
- 2: The master sends the 7 bits slave ID plus a write bit (low)
- 3: The addressed slave asserts an ACK on the data line.
- 4: The master sends 8 bits of the Register Address.
- 5: The slave asserts an ACK on the data line *only if the address is valid (NACK if not).*
- 6: The master sends 8 bits of data.
- 7: The slave asserts an ACK on the data line.
- 8: The master generates a STOP condition.

Figure 4: I² C Write One Byte

6.5.2 Write a Burst of Data

To execute a write of a burst of data, the following steps must be executed:

- 1: The master sends a START condition.
- 2: The master sends the 7 bits slave ID plus a write bit (low).
- 3: The addressed slave asserts an ACK on the data line.
- 4: The master sends 8 bits of the Register Address.
5: The slave asserts an ACK on the data line **only if**
- 5: The slave asserts an ACK on the data line *only if the address is valid (NACK if not).*
- 6: The master sends 8 bits of data.
- 7: The slave asserts an ACK on the data line.
- 8: Repeat 6 and 7 as long as needed.

6.5.3 Read One Byte

To read one byte, the following steps must be executed:

- 1: The master sends a START condition.
- 2: The master sends the 7 bits slave ID plus a write bit (low).
- 3: The addressed slave asserts an ACK on the data line.
- 4: The master sends 8 data bits.
- 5: The active slave asserts an ACK on the data line *only if the address is valid (NACK if not).*
- 6: The master sends a restart condition.
- 7: The master sends the 7 bits slave ID plus a read bit (high).
- 8: The addressed slave asserts an ACK on the data line.
- 9: The slave sends 8 data bits.
- 10: The master asserts a NACK on the data line.
- 11: The master generates a STOP condition.

Figure 6: I² C Read One Byte

6.5.4 Read a Burst of Data

To execute a read of a burst of data, the following steps must be executed:

- 1: The master sends a START condition.
2: The master sends the 7 bits slave ID p
- 2: The master sends the 7 bits slave ID plus a write bit (low).
3: The addressed slave asserts an ACK on the data line.
- The addressed slave asserts an ACK on the data line.
- 4: The master sends 8 bits of the Register Address.
- 5: The slave asserts an ACK on the data line **only if the address is valid (NACK if not)**.
- 6: The master sends a repeated START condition.
7: The master sends the 7 bits slave ID plus a read
- 7: The master sends the $\overline{7}$ bits slave ID plus a read bit (high).
8: The slave asserts an ACK on the data line.
- 8: The slave asserts an ACK on the data line.
9: The slave sends 8 bits of data.
- The slave sends 8 bits of data.
- 10: The master asserts an ACK on the data line.
- 11: Repeat 9 and 10 as long as needed.
12: The master generates a STOP condit
- The master generates a STOP condition.

Figure 7: I² C Read a Burst of Data

7 SPI Interface

To connect a MAX21000 device to an SPI master, the CS pin of the MAX21000 device must be connected to the CSn pin of the SPI master, the SDA_SDI_O pin of the MAX21000 device must be connected to the MOSI pin of the SPI master, the SA0_SDO pin of the MAX21000 device must be connected to the MISO pin of the SPI master and the SCL CLK pin of the MAX21000 device must be connected to the SCLK pin of the SPI master. For external component parameters, refer to [Table 4: SPI External Component Properties.](#page-16-2)

Figure 8: SPI Interface Connection to an Application Processor

7.1 SPI Protocol

CSn is the serial port enable and is controlled by the SPI master. It goes low at the start of the transmission and returns to high at the end.

SCLK is the serial port clock and is controlled by the SPI master. It is kept high when CSn is high (no transmission). MISO and MOSI are, respectively, the serial port data input and output. These lines are driven at the falling edge of SCLK and are sampled at the rising edge of SCLK.

Both the read register and write register commands are completed in 16 clock pulses, or in multiples of 8 in case of multiple read/write bytes. Bit duration is the time between two falling edges of SCLK.

Figure 9: SPI Protocol

7.2 Register Address

The SPI register address for the MAX21000 is composed by 6 bits of address, 1 bit to select the direction of the operation (Read/Write) and 1 bit whose meaning can be configured as:

Auto-increment: if 0, in case of burst operation the initial register address is auto-incremented after every data byte; if 1, the operation is executed always on the same register; **Even parity** this bit represents the even parity computed on the 6 bits of the register address; **Odd Parity** this bit represents the odd parity computed on the 6 bits of the register address;

parity error and if parity are used to manage the parity bit during the SPI communication. The parity bit is an additional bit added to the end of a digital word and it indicates whether the number of bits in the word with the value one is even or odd. Parity bit is used to verify if there was a communication error. According to the datasheet, during a SPI communication, the first byte is the register address you want to read/write:

Here are shown some examples:

Bit 7 – indicated with x, can be 0 or 1, depending on if you want to write or read the correspondent register

For further explanation:

- to **write** register **0x00**, using **odd parity,** you have to send from your MCU the byte '**01000000**' (0x40);
- to **read** register **0x20**, using **even parity,** you have to send from your MCU the byte '**11100000**' (0xE0);
- to **read** register **0x23**, using **odd parity,** you have to send from your MCU the byte '**10100011**' (0xA3);
- to **read** register **0x3F**, using **even parity,** you have to send from your MCU the byte '**10111111**' (0xBF);

When the device receives the above bytes from the MCU, it will try to calculate the parity bit, in according with the rule set in the *if_parity* field. If the parity calculated by the device is the same of the parity bit, the communication was good and the content of *parity_error* bit will be 0. If instead the parity calculated by the device is different from the parity bit, the content of *parity error* bit is set to '1', reporting that there was an error in the serial communication.

8 Interrupts

The MAX21000 is equipped with an interrupt module to control a set of interrupt flags and two interrupt lines (INT1 and INT2).

This module allows to:

- 1: Configure the behavior of the interrupt lines (INT1 and INT2)
- 2: Map each interrupt flag to one of both the interrupt lines
- 3: Create a conditional interrupt (Rate interrupt) based on four different thresholds

8.1 Interrupt Flags

The interrupt module provides several interrupt flags in the [INT_STS_UL](#page-59-1) register. Each flag reports the occurrence of a significant event inside the device.

8.2 Interrupt Lines

An interrupt line is a dedicated pin where a notification to an external application processor can be provided. The MAX21000 is equipped with two interrupt lines that can be configured independently.

For each interrupt line, it is possible to (refer t[o INT_CFG2\)](#page-58-1):

- Enable/disable them
- Set the active level to low
- Set the output type to push-pull or open drain configuration

To map an interrupt flag to an interrupt line it is necessary to enable it through the [INT1_MSK](#page-61-2) for INT1 and [INT2_MSK](#page-61-3) for INT2: by setting its corresponding bit to 1 on the bit-mask, the interrupt flag is mapped to the related interrupt line. The output of the two INT1 and INT2 interrupt lines is then computed by applying the OR operator to all the enabled interrupt flags contained in the [INT1_STS](#page-60-2) and [INT2_STS](#page-60-3) registers, respectively. Please note that the enable is not applied to the **INT1_STS** and **INT2_STS** registers, so those registers contain also the value of the interrupt flags that are not enabled.

An interrupt line can be configured in order to keep the status until the master requests to clear it (latched) or after a timeout. Those modes can be selected through the *int1_latch_mode* and *int2_latch_mode* register fields [\(INT_TMO\)](#page-59-0). The only exception is the *gyro_dr* flag that is always unlatched regardless of the *int1_latch_mode* and *int2_latch_mode* register fields setting.

The lines can be configured also to auto-clear its status after a period of time where the duration to clear the interrupt is programmable (see Timed Mode at [INT_TMO\)](#page-59-0).

8.3 Rate Interrupts

The *rate interrupt* allows generating an interrupt event when the gyroscope data falls inside a range of values defined by a set of thresholds. By setting the absolute value of the threshold (|TH|) for each gyroscope axis, four ranges are defined:

Figure 10: Conditional Ranges

For each range, it is possible to select if it contributes to the generation of the rate interrupt for each axis. Then, the rate interrupts are computed as follow:

- int_and: Active if the defined conditions are satisfied for all the gyroscope axes at the same time;
- int_or: Active if the defined conditions are satisfied for at least one of the gyroscope axes.

The threshold absolute value can be set by writing the **INT_REF_X**, **INT_REF_Y** and **INT_REF_Z** registers. Those registers represent the most significant byte of the real threshold; so, to compute the actual absolute value of the threshold, the register value must be multiplied by 256. If the application requires a better resolution, it is possible to specify a 16-bit threshold by setting the int single ref register field; in this case, all the axes share the same threshold that is defined as the combination of [INT_REF_X](#page-53-3) and [INT_REF_Y](#page-53-1) where the first one is the most significant byte and the last one is the least significant byte of the threshold.

Once the thresholds are defined, the [INT_MSK_X,](#page-55-2) [INT_MSK_Y](#page-56-0) and [INT_MSK_Z](#page-56-1) permits to select which range contributes to the generation of the rate interrupts for each *axis* as follow:

- **int_***axis***_high_pos_en:** if enabled, the condition is satisfied if the data of the *axis* falls in the *high_pos* range;
- int *axis* high neg en: if enabled, the condition is satisfied if the data of the *axis* falls in the *high* neg range;
- int *axis* low pos en: if enabled, the condition is satisfied if the data of the *axis* falls in the *low* pos range;
- **int_***axis***_low_neg_en:** if enabled, the condition is satisfied if the data of the *axis* falls in the *low_neg* range.

Then, the desired axes must be enabled by setting to 1 the corresponding bit of the [int_mask_xyx_and](#page-57-1) for the **int_and** and [int_mask_xyz_or](#page-57-1) for the **int_or**.

Through th[e INT_MSK_X,](#page-55-2) [INT_MSK_Y](#page-56-0) and [INT_MSK_Z](#page-56-1) registers, it is also possible to read the current value of the conditions, regardless of the content of the [int_mask_xyz_or](#page-57-1) and [int_mask_xyz_and](#page-57-1) register fields. Each of these values can be configured as latched by setting the int freeze register field. For the rate interrupts, it is possible also to define a de-bounce value by defining the number of samples the axis data has to satisfy the condition before asserting the corresponding interrupt. Those values can be set in the INT DEB X, INT DEB Y and INT DEB Z registers. If the required value is greater than 255, it is possible to define a 16-bit debounce value by setting int single deb register field; in this case, all the conditions share the same de-bounce value that is defined as the combination of [INT_DEB_X](#page-54-2) and [INT_DEB_Y](#page-54-1) where the first one is the most significant byte and the last one is the least significant byte of the debounce value.

9 Reading Data from MAX21000 and FIFO Operation

MAX21000 sensor output data can be read by the processor in two different mechanisms: synchronous (DATA_READY interrupt) and asynchronous (polling) reading methods.

9.1 Synchronous Reading

In this mode, the processor reads the data set (e.g. 6 bytes for a 3 axes configuration) generated by the MAX21000 every time that gyro dr flag is set. The processor must read the output data only once the gyro dr flag is set in order to avoid data inconsistencies. Benefits of using this approach include the perfect reconstruction of the signal coming from the sensors and minimization of the data traffic at the interface.

9.2 Asynchronous Reading

In this mode, the processor reads the data generated by the MAX21000 regardless the status of the [gyro_dr](#page-37-0) flag. To minimize the error caused by different samples being read a different number of times, the access frequency to be used must be higher than the selected ODR.

9.3 FIFO Modes Description

The MAX21000 also embeds a 256-slot of a 16-bit data FIFO for each of the three output channel; X, Y and Z. This allows a consistent power saving for the system since the host processor does not need to continuously poll data from the sensor, but it can wake up only when needed and burst the data out from the FIFO. When configured in Snapshot mode, it offers the ideal mechanism to capture the data following a Rate Interrupt event. The data order in FIFO depends on the endian setting [\(endian\)](#page-49-0):

The FIFO buffer can work according to four main different modes (see FIFO CFG): off, normal, interrupt and snapshot. Both normal and Interrupt modes can be optionally configured to operate in overrun Mode, depending on whether, in case of buffer under-run, newer or older data are lost.

9.3.1 FIFO OFF Mode

In this mode, the FIFO is turned off; data are stored only in the data registers. No data are available from FIFO if read. When the FIFO is turned OFF, there are two options to use the device: synchronous/asynchronous reading.

9.3.2 Normal Mode

The behavior of the FIFO in Normal mode varies depending on the **Overrun** settings (*[fifo_overrun](#page-50-0)* register field). The following paragraphs show a description of the behavior for both settings of the overrun. For the next sections, the following descriptions are useful for the user's understanding of FIFO:

9.3.2.1 Stop on Full

- FIFO is turned on.
- FIFO is filled with the data at the selected Output Data Rate (ODR).
- When FIFO is full, an interrupt can be generated.
- When FIFO is full, all the new incoming data will be discharged. Reading only a subset of the data already stored into the FIFO keeps locked the possibility for new data to be written.
- Only if all the data are read the FIFO will restart saving data. If the communication speed is high, data loss can be prevented.
- In order to prevent a FIFO full condition, the required condition is to complete the reading of the data set before the next DATA_READY occurs.
- If this condition is not guaranteed, data can be lost.

Figure 11: FIFO Normal Mode, Stop on Full

9.3.2.2 Overwrite

- FIFO is turned on.
- FIFO is filled with the data at the selected ODR.
- When FIFO is full, an interrupt can be generated.
- When FIFO is full, the oldest data will be overwritten with the new ones.
- If communication speed is high, data integrity can be preserved.
- In order to prevent a data lost condition, the requirement is to complete the reading of the data set before the next DATA_READY occurs.
- If this condition is not guaranteed, data can be overwritten.
- When an overrun condition occurs the Reading pointer is forced to Writing Pointer -1, to make sure only older data are discarded and newer data have a chance to be read.

Figure 12: FIFO Normal Mode, Overwrite

9.3.3 Interrupt Mode

The behavior of the FIFO in Interrupt mode varies depending on the **Overrun** settings (*[fifo_overrun](#page-50-0)* register field). The following paragraphs show a description of the behavior for both settings of the overrun.

9.3.3.1 Stop on Full

- FIFO is initially disabled. Data is stored only in the data registers.
- When a Rate Interrupt (either OR or AND) is generated, the FIFO is turned on automatically. It stores the data at the selected ODR. Rate interrupt are documented starting from [INT_REF_X.](#page-53-3)
- When FIFO is full, all the new incoming data will be discharged. Reading only a subset of the data already stored into the FIFO keeps locked the possibility for new data to be written.
- Only if all the data are read the FIFO will restart saving data.
- If communication speed is high, data loss can be prevented.
- In order to prevent a FIFO full condition, the required condition is to complete the reading of the data set before the next DATA_READY occurs.
- If this condition is not guaranteed, data can be lost.

Figure 13: FIFO Interrupt Mode, Stop on Full

9.3.3.2 Overwrite

- FIFO is initially disabled. Data is stored only in the data registers.
- When a Rate Interrupt (either OR or AND) is generated, the FIFO is turned on automatically. It stores the data at the selected ODR.
- When FIFO is full, an interrupt can be generated.
- When FIFO is full, the oldest data will be overwritten with the new ones.
- If communication speed is high, data integrity can be preserved.
- In order to prevent a data lost condition, the required condition is to complete the reading of the data set before the next DATA_READY occurs.
- If this condition is not guaranteed, data can be overwritten.
- When an overrun condition occurs the Reading pointer is forced to Writing Pointer -1, to make sure only older data are discarded and newer data have a chance to be read.

Figure 14: FIFO Interrupt Mode, Overwrite

Figure 15: FIFO Snapshot Mode

9.3.4 Snapshot Mode

- FIFO is initially in normal mode with overrun enabled.
- When a Rate Interrupt (either OR or AND) is generated, the FIFO switches automatically to not-overrun mode. It stores the data at the selected ODR until the FIFO becomes full.
- When FIFO is full, an interrupt can be generated.
- When FIFO is full, all the new incoming data will be discharged. Reading only a subset of the data already stored into the FIFO keeps locked the possibility for new data to be written.
- Only if all the data are read the FIFO will restart saving data.
- If communication speed is high, data loss can be prevented.
- In order to prevent a FIFO full condition, the required condition is to complete the reading of the data set before the next DATA_READY occurs.
- - If this condition is not guaranteed, data can be lost.

9.4 Example of FIFO Read/Write Pointers Evolution

The following drawing assumes:

- A reading frequency roughly twice the writing frequency (ODR);
- A FIFO threshold = 126
- FIFO is in normal mode

Figure 16: FIFO Read/Write Pointer Evolution

10 Programming Example

This chapter shows some examples on how to execute some operations on the device. Two functions are defined to abstract the SPI/I²C communication:

- *Write(<Register Address>, <Value>)*: Writes the *<Value>* to the register at the *<Register Address>;*
- *Read(<Register Address>)*: Returns the value of the register at the *<Register Address>*;
- *ReadBurstNoInc(<Register Address>, <count>)*: Returns an array of *count* elements with the result of a burst read with no auto-increment on *<Register Address>* register;

10.1 Simple Read-Out Sequence, No FIFO, No Interrupts

Figure 17: Simple Read-Out Sequence, No FIFO, No Interrupts

10.2 Simple Read-Out Sequence, FIFO Normal Mode, No Interrupts

Figure 18: Simple Read-Out Sequence, FIFO Normal, No Interrupts

10.3Simple Read-Out Sequence, Normal Mode, Data-Ready Interrupts

Figure 19: Simple Read-Out Sequence, Normal Mode, w/ Data Ready Interrupts and No FIFO

11 DSYNC

Data synchronization pin of MAX21000 can be used in various use cases such as controlling the power mode of MAX21000 through an external device, capturing the data in FIFO with an event triggered by camera or GPS, and lastly mapping the DSYNC level in LSB of the sensor output data in order to compare the data timing with respect to the DSYNC edge timing.

The DSYNC operation modes are described in three different sub-sections:

11.1 Power Mode Switching Using DSYNC

DSYNC pin can be configured to trigger the switching from a power mode to another. In order to achieve this, the following flow has to be executed:

In [Power mode table,](#page-44-0) the power mode transitions with respect to DSYNC active level have been described. In order to enable power mode transition with DSYNC level, [DSYNC_CFG](#page-51-0) register following bit have to be set properly:

DSW_EDG: 0: DSYNC is active on level, 1: DSYNC is active on edge *DSW_LOW:* When 1, DSYNC is an active low level control to wake up. When 0, DSYNC is an active high level to wake up. This bit affects both the edge and the level modes

11.1.1 Example Register Settings for Power Mode Switching

11.2Filling FIFO with DSYNC Edge

With this function turned on, DSYNC signal is used to capture data and fill the FIFO with a selected number of data. It is enabled with [DSYNC_CFG](#page-51-0) register and number of samples is configured using th[e DSYNC_CNT](#page-51-1) register.

In order to configure data filling with FIFO, the following procedure has to be performed:

The number of dataset stored in the FIFO is given by: *N_data_set_stored_in_FIFO = DSYNC_CNT + 1*

A dataset is composed by the numbers of the axes enabled for each sensor (3 for Gyro). The DSYNC frequency must be: ODR / (DSYNC CNT + 1) if one edge only is selected. Minimum DSYNC pulse duration is 500ns.

11.2.1 Example Register Settings for FIFO Filling with DSYNC

Table 6: Register Settings for FIFO Filling with DSYNC

11.3 Map DSYNC on LSb

When this function is turned on, the DSYNC signal is mapped into the LSb of all the sensors that are enabled for this functionality. It is configured through DSYNC_CFG register, and the following have to be set accordingly:

```
DSM_ENB: When 1, the DSYNC signal is mapped onto the Gyro LSB
DS TEMP When 1, the DSYNC signal is mapped onto the temperature LSB
```
In this configuration, the DSYNC is active only on the level. To map the DSYNC signal in the LSb correctly, it is suggested to use an ODR that is bigger than the DSYNC frequency.

11.3.1 Example Register Settings for Mapping DSYNC level on LSB

Table 7: Register Settings for Mapping LSB

11.4 Example of DSYNC Application: Generation of Non-Standard ODR

One of the possible uses of the DSYNC functionality is to generate the output data at a particular rate that it is not defined in the list of the possible ODRs. Using the "data queuing in FIFO with DSYNC" function and the FIFO over-threshold interrupt, it is possible to generate a data-ready signal at a specific data rate. In order to minimize the ODR jitter, the max ODR of the device should be selected, so that the max jitter is equal to $\pm 1/(2^*ODR_{max})$.

For configuring this functionality the following procedure is suggested:

Based on this procedure, the following operation described with a timing diagram can be obtained:

Figure 20: Timing Diagram to Generate a Nonstandard ODR

12 Register Map

12.1COMMON BANK

Table 8: Common Bank

12.1.1 WHO_AM_I

Description

WHO_AM_I identifies the MAX21000 with the value of 0xB1.

12.1.2 BANK_SELECT

Description

There are total 48 registers in the User Space of MAX21000: Bank 0, Bank 1 and Common Bank. *bank_sel* is used to access the registers in two different banks : Bank 0, Bank 1, whereas Common Bank registers can be accessed independent of the *bank_sel* value. Bank 0 and Bank 1 have total 32 registers (address range from 0x00 to 0x1F) and Common Bank has the register address range from 0x20 to 0x3F.

Fields

bank_sel: Selects the total number of 16 banks each of which has total 32 bytes (registers). For the user, the valid values of BANK_SEL are:

12.1.3 SYSTEM_STATUS

Description

System Status register reports two fundamental flags necessary to properly manage the communication with the MAX21000. Ideally, every new data-reading operation from the MAX21000 should only take place when at least a new DATA_READY (*gyro_dr*) event occurs. Failure to read data at every DATA_READY event may result in either reading the same data more than once or missing at least one ouput data. That is particularly true when the FIFO is disabled. The way the **gyro_dr** flag is reset can be configured using register [DR_CFG \(0x13\).](#page-47-0)

The *gyro_err* flag indicates the occurence of either one of the events described above. If the FIFO is used, multiple data can be read safely. In order to set up FIFO and burst read the data loaded on the FIFO (refer t[o FIFO section\)](#page-21-0).

Fields

12.1.4 GYRO_X_H

Description

GYRO X H stores the MSB (bit[15:8]) of the most recent X-axis gyroscope output whe[n endian](#page-49-0) bit is set to 0, or the LSB (bit[7:0]) whe[n endian](#page-49-0) bit is set to 1.

12.1.5 GYRO_X_L

Description

GYRO_X_L stores the LSB (bit[7:0]) of the most recent X-axis gyroscope output when [endian](#page-49-0) bit is set to 0, or the MSB (bit[15:8]) when [endian](#page-49-0) bit is set to 1.

12.1.6 GYRO_Y_H

Description

GYRO_Y_H stores the MSB (bit[15:8]) of the most recent Y-axis gyroscope output whe[n endian](#page-49-0) bit is set to 0, or the LSB (bit[7:0]) whe[n endian](#page-49-0) bit is set to 1.

12.1.7 GYRO_Y_L

Description

GYRO_Y_L stores the LSB (bit[7:0]) of the last Y-axis gyroscope output whe[n endian](#page-49-0) bit is set to 0, or the MSB (bit[15:8]) when $endian$ bit is set to 1.

12.1.8 GYRO_Z_H

Description

GYRO Z H stores the MSB (bit[15:8]) of the most recent Z-axis gyroscope output when [endian](#page-49-0) bit is set to 0, or the LSB (bit[7:0]) whe[n endian](#page-49-0) bit is set to 1.

12.1.9 GYRO_Z_L

Description

GYRO_Z_L stores the LSB (bit[7:0]) of the most recent Z-axis gyroscope output when [endian](#page-49-0) bit is set to 0, or the MSB (bit[15:8]) when [endian](#page-49-0) bit is set to 1.

12.1.10 TEMP_H

Description

TEMP_H stores the MSB (bit[15:8]) of the most recent temperature sensor output when [endian](#page-49-0) bit is set to 0, or the LSB (bit[7:0]) when [endian](#page-49-0) bit is set to 1.

12.1.11 TEMP_L

Description

TEMP_L stores the LSB (bit[7:0]) of the most recent temperature sensor output whe[n endian](#page-49-0) bit is set to 0, or the MSB (bit[15:8]) when [endian](#page-49-0) bit is set to 1.

12.1.12 HP_RST

Description

Reading HP_RST register resets the Highpass filter output.

12.1.13 FIFO_COUNT

Description

FIFO_COUNT provides the total number of FIFO 16-bits words available in FIFO.

12.1.14 FIFO_STATUS

Description

FIFO_STATUS register provides the status of all the potential FIFO events.

Fields

Figure 21: FIFO Flags

(F) FIFO is empty and all the available new data have been read

12.1.15 FIFO_DATA

Description

FIFO_DATA register is used to read and write data on the FIFO buffer. The contents of the sensor data registers are written into the FIFO buffer when their corresponding FIFO enable flags are set to 1 through the [FIFO_CFG](#page-50-0) register. When the FIFO buffer has overflowed, the oldest data will be lost and new data will be written to the FIFO if the override bit is set in th[e FIFO_CFG](#page-50-0) register, otherwise the new data is discarded. If the FIFO buffer has overflowed, the status bit *fifo_data_lost* (bit 4 o[f FIFO_STATUS](#page-41-0) register) is automatically set to 1.

If the FIFO buffer is empty, reading this register will return the last byte that was previously read from the FIFO until new data is available. The user should check [FIFO_COUNT](#page-40-2) to ensure that the FIFO buffer is not read when empty.

12.1.16 PAR_RST

Description

Reading PAR_RST register clears the *parity_error* flag [\(ITF_OTP\)](#page-49-0).

12.2USER BANK #0 (*bank_sel* **= 0000)**

Table 9: User Bank 0

12.2.1 POWER_CFG

Description

Gyroscope Full scale, power mode and axis configuration register.

Fields

Table 11: Power Mode Configuration

12.2.2 SENSE_CFG1

Description

Lowpass filter, OIS and Self Test configuration register. When a self-test mode is selected, an offset is generated on the digital output whose amount depends on the selected full-scale. The output of this parameter is affected by a strong spread, in the order of +/- 50%. This test allows detecting both electrical and mechanical issues.

Fields

sns_ois For Optical Image Stabilization (OIS) applications the main requirement is the resolution. So, when this mode is enabled (*sns_ois* = 1), the available full-scale settings are: 250 dps, 125 dps, 62.5 dps and 31.25 dps (see POWER CFG register)

Table 13: Bandwidth Configuration

12.2.3 SENSE_CFG2

Description

Output Data Rate configuration register selects the preferred Output Data Rate (ODR) according to the description below.

In ECO Mode: SENSE CFG2[1:0] = $b00 \rightarrow ODR = 200Hz$ $SENSE_CFG2[1:0] = b01 \rightarrow ODR = 100Hz$ SENSE_CFG2[1:0] = $b10 \rightarrow$ ODR = 50Hz SENSE_CFG2[1:0] = $b11 \rightarrow ODR = 25Hz$

12.2.4 SENSE_CFG3

Description

Highpass filter configuration register comprises 3 fields. The least significant 4 bits can be used to select the cut-off frequency of the high-pass filter. Bit[5:4] is used to bypass the lowpass filter and highpass filter, respectively.

Table 15: HPF Cut-Off Frequencies

12.2.5 DR_CFG

Description

Data Ready reset and temperature sensor settings..

12.2.6 IO_CFG

Description

I/O Configuration Register controls the pullup and pulldown resistors of the pins DSYNC, SCL, SDA, INT1, and INT2.

Fields

12.2.7 I2C_CFG

Description

I2C_CFG sets for I^2C bus speed and output drive strength.

12.2.8 ITF_OTP

Description

Interface and OTP configuration register.

Fields

12.2.9 FIFO_TH

Description

FIFO Threshold configuration register. When the number of 16-bits samples stored in FIFO is above the threshold, the *fifo_ovthold* interrupt is generated. In this case, a sample is the entire set of axes; for example, if the threshold is set to 5 and all the axes are stored in FIFO, the interrupt is generated when the FIFO contains 5 sets of {X, Y, Z} gyroscope data. This value must be different from 0.

12.2.10 FIFO_CFG

Description

FIFO configuration register determines which sensor measurements are loaded into the FIFO buffer and selects the desired FIFO behavior.

Data stored inside the sensor data registers will be loaded into the FIFO buffer if a sensor's respective *fifo_store_n* bit is set to 1 in this register. The behavior of FIFO writes, when the FIFO buffer is full, can be configured with the *fifo_mode* bit. In order to read the data in the FIFO buffer, the *fifo_mode* must be set to a value diffferent than 00.

The sensors are written into the FIFO at the Output Data Rate defined in **SENSE_CFG2** register.

12.2.11 DSYNC_CFG

Description

DSYNC configuration register has to be used to configure the way the MAX21000 manages events occurring on the DSYNC pin. Multiple different actions can be taken simultaneously, like changing the power mode, mapping the DSYNC pin value onto the gyroscope LSB data and concurrently triggering the capture of new data.

When the DSYNC pin is configured as active on edge and a dynamic power mode is configured, only the active edge determines the transition. The opposite transition must be done within SW or by reversing the active edge.

Fields

12.2.12 DSYNC_CNT

Description

This register sets the number of words to be filled in FIFO after a DSYNC pin event occurs. DSYNC counter configuration can be used to track the evolution of the rate signal from the gyroscope immediately after an external event captured on the DSYNC pin. The number of dataset stored in the FIFO is given by the following formula:

N_data_set_stored_in_FIFO = DSYNC_CNT + 1

12.3 USER BANK #1 (*bank_sel* **= 0001)**

Table 16: User Bank 1

12.3.1 INT_REF_X

Description

Most significant byte of the reference for X-axis. The actual reference is then computed as **INT_REF_X** * 256.

When *int_single_ref* is set, the reference will be {INT_REF_X, INT_REF_Y}, where INT_REF_X is the most significant byte and INT_REF_Y is the least significant byte.

12.3.2 INT_REF_Y

Description

Most significant byte of the reference for Y-axis. The actual reference is then computed as INT_REF_Y * 256.

When *int_single_ref* is set, the reference will be {INT_REF_X, INT_REF_Y}, where INT_REF_X is the most significant byte and INT_REF_Y is the least significant byte.

12.3.3 INT_REF_Z

Description

Most significant byte of the reference for Y-axis. The actual reference is then computed as INT_REF_Z * 256.

When *int_single_ref* is set, the reference will be {INT_REF_X, INT_REF_Y}, where INT_REF_X is the most significant byte and INT_REF_Y is the least significant byte.

12.3.4 INT_DEB_X

Description

Rate interrupt duration reference for X-axis. This register determines how long (measured in number of samples) the selected AND/OR rate interrupt configuration has to stay asserted before the corresponding interrupt source bit is set and the interrupt on either INT1 or INT2 is eventually generated. When the selected AND/OR rate interrupt configuration is deasserted (goes to 0) the corresponding interrupt source bit is deasserted immediately, without any delay.

The duration in seconds can be computed as Number of samples * ODR.

12.3.5 INT_DEB_Y

Description

Rate interrupt duration reference for Y-axis. This register determines how long (measured in number of samples) the selected AND/OR rate interrupt configuration has to stay asserted before the corresponding interrupt source bit is set and the interrupt on either INT1 or INT2 is eventually generated. When the selected AND/OR rate interrupt configuration is deasserted (goes to 0) the corresponding interrupt source bit is deasserted immediately, without any delay.

The duration in seconds can be computed as Number of samples * ODR.

12.3.6 INT_DEB_Z

Description

Rate interrupt duration reference for Z-axis. This register determines how long (measured in number of samples) the selected AND/OR rate interrupt configuration has to stay asserted before the corresponding interrupt source bit is set and the interrupt on either INT1 or INT2 is eventually generated. When the selected AND/OR rate interrupt configuration is deasserted (goes to 0) the corresponding interrupt source bit is deasserted immediately, without any delay.

The duration in seconds can be computed as Number of samples * ODR.

12.3.7 INT_MSK_X

Description

Rate Interrupt, X-axis configuration register comprises 2 fields. The 4 LSBs are one-hot encoded bits which indicate whether the rate is positive or negative and whether it is above or below the rate threshold. The most significant 4 bits enable the interrupt bits when the corresponding condition is met.

Fields

int_x_high_pos_en: Enable the int_x_high_pos interrupt generation for threshold event detection on X-axis. *int_x_low_pos_en:* Enable the int_x_low_pos interrupt generation for threshold event detection on X-axis. *int_x_high_neg_en:* Enable the int_x_high_neg interrupt generation for threshold event detection on X-axis. *int_x_low_neg_en:* Enable the int_x_low_neg interrupt generation for threshold event detection on X-axis. x_high_pos: Signal is positive, higher than threshold (see [Interrupt Zones\)](#page-19-3) *x_low_pos:* Signal is positive, lower than threshold *x_high_neg:* Signal is negative, higher than threshold *x_low_neg:* Signal is negative, lower than threshold

12.3.8 INT_MSK_Y

Description

Rate Interrupt, Y-axis configuration register comprises 2 fields. The 4 LSBs are one-hot encoded bits which indicate whether the rate is positive or negative and whether it is above or below the rate threshold. The most significant 4 bits enable the interrupt bits when the corresponding condition is met.

Fields

int_y_high_pos_en: Enable the int_y_high_pos interrupt generation for threshold event detection on Y-axis. *int_y_low_pos_en:* Enable the int_y_low_pos interrupt generation for threshold event detection on Y-axis. *int_y_high_neg_en:* Enable the int_y_high_neg interrupt generation for threshold event detection on Y-axis. *int* y low neg en: Enable the int y low neg interrupt generation for threshold event detection on Y-axis. y_high_pos: Signal is positive, higher than threshold (see [Interrupt Zones\)](#page-19-3) *y_low_pos:* Signal is positive, lower than threshold *y_high_neg:* Signal is negative, higher than threshold y_low_neg: Signal is negative, lower than threshold

12.3.9 INT_MSK_Z

Description

Rate Interrupt, Z-axis configuration register comprises 2 fields. The 4 LSBs are one-hot encoded bits which indicate whether the rate is positive or negative and whether it is above or below the rate threshold. The most significant 4 bits enable the interrupt bits when the corresponding condition is met.

Fields

12.3.10 INT_MSK_AO

Description

Interrupt AND/OR masks register.

Fields

int_freeze: Set the interrupt on threshold as latched. When enabled, all the rate interrupt flags are latched. When triggered, the interrupt is latched until the INT_MSK_{X,Y,Z} register is read. *int_mask_xyz_and:* Each bit activates an axis. The active axes are ANDed together to generate the AND interrupt. *int_mask_xyz_or:* Each bit activates an axis. The active axes are ORed together to generate the OR interrupt.

12.3.11 INT_CFG1

Description

Interrupt 1 configuration register. In this register, the rate interrupt functionality configuration such as FS and data rate for all axes can be separately set from the output data signal path.

Fields

- b10: data at 10 kHz without HPF;
- b11: data at 10 kHz with HPF

12.3.12 INT_CFG2

Description

INT2_CFG register controls both INT1/2 enable bits, push-pull/open-drain configuration and interrupt active levels. When the interrupts are disabled, they will remain inactive regardless of the setting at [INT1_MSK](#page-61-1) an[d INT2_MSK.](#page-62-0)

12.3.13 INT_TMO

Description

Interrupt Timeout and Interrupt Mode configuration register allows to configure the interrupt lines to operate as either un-latched, latched or timed. As un-latched, they can be further configured in such a way that interrupt sources [\(INT1_STS](#page-60-1) and [INT2_STS\)](#page-61-0) can be cleared when they are read or cleared when they are written with a logic 1. Clearing an interrupt source by writing a logic 1 allows clearing single bits rather than the entire register.

12.3.14 INT_STS_UL

Description

INT_STS_UL bits are the un-latched version of the interrupt status registers; as these signals are shared by the INT1 generator and the INT2 generator, there is a unique register shared. This register is the actual source for the interrupt lines when the interrupts are configured as un-latched. When the interrupt lines are configured as latched, be it both or just one of them, these bits can be used to keep monitoring the status of an interrupt source after the event.

Fields

12.3.15 INT1_STS

Description

INT1_STS bits are the latched interrupt sources; when the latched mode is used, they can be cleared either by reading INT1_STS register, or writing these bits as 1. Bits can be cleared at the same time by forming the proper mask. When INT1 is configured as unlatched or timed, these registers are set as 0.

Fields

sts_i1_data_ready: DATA_READY status *sts_i1_int_and:* Rate interrupt AND status **sts_i1_fifo_empty:** FIFO_EMPTY status **and interval interval of the interrupt OR status compared interval inter** *sts_i1_fifo_ths:* FIFO_THRESHOLD status *sts_i1_data_sync:* DSYNC status

sts_i1_fifo_overrun: FIFO_OVERRUN status *sts_i1_otp_downloading:* OTP_DOWNLOADING status

12.3.16 INT2_STS

Description

INT2_STS bits are the latched interrupt sources; when the latched mode is used, they can be cleared either by reading INT2_STS register, or writing these bits as 1. Bits can be cleared at the same time by forming the proper mask. When INT2 is configured as unlatched or timed, these registers are set as 0.

Fields

12.3.17 INT1_MSK

Description

Interrupt 1 generation, mask register is used to enable selected interrupt sources in the [INT1_STS](#page-60-1) register to activate the INT1 interrupt line. Interrupt sources are masked (prevented from generating an interrupt) as long as the corresponding bit in the mask register is 0. Valid configurations are with None, One, Multiple or Every bit set to 1; having multiple possible interrupt sources will require the Interrupt Service routine to identify the correct one(s).

Fields

msk_i1_data_ready: DATA_READY status **msk_i1_int_and:** Rate interrupt AND status msk _i1_fifo_empty: FIFO_EMPTY status **msk _i1_int_or: Rate interrupt OR status msk** *msk _i1_fifo_ths:* FIFO_THRESHOLD status *msk _i1_data_sync:* DSYNC status

msk _i1_fifo_overrun: FIFO_OVERRUN status *msk_i1_otp_downloading:* OTP_DOWNLOADING status

12.3.18 INT2_MSK

Description

Interrupt 2 generation, mask register is used to enable selected interrupt sources in the [INT2_STS](#page-61-0) register to activate the INT2 interrupt line. Interrupt sources are masked (prevented from generating an interrupt) as long as the corresponding bit in the mask register is 0. Valid configurations are with None, One, Multiple or Every bit set to 1; having multiple possible interrupt sources will require the Interrupt Service routine to identify the correct one(s).

Fields

12.3.19 OTP_STATUS

Description

OTP status register provides the OTP download status.

- b01: OTP download completed successfully, 1 bit corrected
- b10: OTP download completed successfully, after some attempts
- b11: OTP download completed with errors

12.3.20 SILICON_REV_OTP

Description

SILICON_REV_OTP identifies the MAX21000 silicon revision.

12.3.21 SERIAL_[0:5]

Description

SERIAL_0, SERIAL_1, SERIAL_2, SERIAL_3, SERIAL_4 and SERIAL_5 are 6 registers used to assign a unique identifier to every single MAX21000 sample to enable a complete tracability in terms of LOTs, Assembly history and Test equipment.

13 Definitions

Power supply [V]: This parameter defines the operating DC power supply voltage range of the MEMS gyroscope. Although it is always a good practice to keep V_{DD} clean with minimum ripple, unlike most of the competitors, who require an ultra-low noise, low-dropout regulator to power the MEMS gyroscope, the MAX21000 can not only operate at 1.71V but that supply can also be provided by a switching regular, to minimize the system power consumption.

Power supply current [mA]: This parameter defines the typical current consumption when the MEMS gyroscope is operating in normal mode.

Power supply current in Standby mode [mA]: This parameter defines the current consumption when the MEMS gyroscope is in Standby mode. To reduce power consumption and have a faster turn-on time, in Standby mode only an appropriate subset of the sensor is turned off.

Power supply current in Low Power Mode [mA]: This parameter defines the current consumption when the MEMS gyroscope is in a special mode named Low Power Mode. Whilst in Low Power Mode, the MAX21000 reduces significantly the power consumption, at the price of a slightly higher RND.

Power supply current in stand-by mode [µA]: This parameter defines the current consumption when the MEMS gyroscope is powered down. In this mode, both the mechanical sensing structure and reading chain are turned off. Users can configure the control register through the I²C/SPI interface for this mode. Full access to the control registers through the I^2 C/SPI interface is guaranteed also in power-down mode.

Full-scale range [dps]: This parameter defines the measurement range of the gyroscope in degrees per second (dps). When the applied angular velocity is beyond the full-scale range, the gyroscope output signal will be saturated.

Zero-rate level [LSBs]: This parameter defines the zero rate level when there is no angular velocity applied to the gyroscope.

Sensitivity [mdps/LSB]: Sensitivity (mdps/LSB) is the relationship between 1 LSB and milli-dps. It can be used to convert a digital gyroscope's measurement in LSBs to angular velocity.

Sensitivity change vs. Temperature [%/°C]: This parameter defines the sensitivity change in percentage (%) over the operating temperature range specified in the datasheet.

Zero-rate level change vs. Temperature [dps/°C]: This parameter defines the zero rate level change in degree per second (dps/°C) over the operating temperature range.

Non-linearity [% FS]: This parameter defines the maximum error between the gyroscope's outputs and the bestfit straight line in percentage with respect to the full-scale (FS) range.

System bandwidth [Hz]: This parameter defines the frequency of the angular velocity signal from DC to the builtin bandwidth (BW) that the gyroscopes can measure. A dedicated register can be modified to adjust the gyroscope's bandwidth.

Rate noise density [dps/√Hz]: This parameter defines the standard resolution that users can get from the gyroscopes outputs together with the BW parameter.

Self-test [dps]: This feature can be used to verify if the gyroscope is working properly or in order to not physically rotate the gyroscope after it is assembled on a PCB. When the self-test is enabled, an internal electrostatic force is generated to move the masses to simulate the Coriolis Effect. If the gyroscope's outputs are within the specified self-test values in the data sheet, then the gyroscope is working properly. Therefore, the self-test feature is an important consideration in a user's end-product mass production line.