# Features

- 80C52 Compatible
  - 8051 Instruction Compatible
  - Four 8-bit I/O Ports (44 Pins Version)
  - Three 16-bit Timer/Counters
  - 256 bytes Scratch Pad RAM
  - 11 Interrupt Sources With 4 Priority Levels
- ISP (In-System Programming) Using Standard V<sub>cc</sub> Power Supply
- Integrated Power Monitor (POR/PFD) to Supervise Internal Power Supply
- Boot ROM Contains Serial Loader for In-System Programming
- High-speed Architecture
  - In Standard Mode:

#### 40 MHz (Vcc 2.7V to 5.5V, Both Internal and External Code Execution) 60 MHz (Vcc 4.5V to 5.5V and Internal Code Execution Only)

- In X2 Mode (6 Clocks/Machine Cycle)
   20 MHz (Vcc 2.7V to 5.5V, Both Internal and External Code Execution)
   30 MHz (Vcc 4.5V to 5.5V and Internal Code Execution Only)
- 128K bytes On-chip Flash Program/Data Memory
  - 128 bytes Page Write with auto-erase
    - 100k Write Cycles
- On-chip 8192 bytes Expanded RAM (XRAM)
- Software Selectable Size (0, 256, 512, 768, 1024, 1792, 2048, 4096, 8192 bytes)
- Dual Data Pointer
- Extended stack pointer to 512 bytes
- Variable Length MOVX for Slow RAM/Peripherals
- Improved X2 Mode with Independant Selection for CPU and Each Peripheral
- Keyboard Interrupt Interface on Port 1
- SPI Interface (Master/Slave Mode)
- 8-bit Clock Prescaler
- Programmable Counter Array with:
  - High Speed Output
  - Compare/Capture
  - Pulse Width Modulator
  - Watchdog Timer Capabilities
- Asynchronous Port Reset
- Two Full Duplex Enhanced UART with Dedicated Internal Baud Rate Generator
- Low EMI (inhibit ALE)
- Hardware Watchdog Timer (One-time Enabled with Reset-Out), Power-Off Flag
- Power Control Modes: Idle Mode, Power-down Mode
- Power Supply: 2.7V to 5.5V
- Temperature Ranges: Industrial (-40 to +85°C)
- Packages: PLCC44, VQFP44



8-bit Flash Microcontroller

# AT89C51RE2





# Description

AT89C51RE2 is a high performance CMOS Flash version of the 80C51 CMOS single chip 8-bit microcontroller. It contains a 128 Kbytes Flash memory block for program.

The 128 Kbytes Flash memory can be programmed either in parallel mode or in serial mode with the ISP capability or with software. The programming voltage is internally generated from the standard  $V_{CC}$  pin.

The AT89C51RE2 retains all features of the Atmel 80C52 with 256 bytes of internal RAM, a 10-source 4-level interrupt controller and three timer/counters.

In addition, the AT89C51RE2 has a Programmable Counter Array, an XRAM of 8192 bytes, a Hardware Watchdog Timer, SPI and Keyboard, two serial channels that facilitates multiprocessor communication (EUART), a speed improvement mechanism (X2 mode) and an extended stack mode that allows the stack to be extended in the lower 256 bytes of XRAM.

The fully static design of the AT89C51RE2 allows to reduce system power consumption by bringing the clock frequency down to any value, even DC, without loss of data.

The AT89C51RE2 has 2 software-selectable modes of reduced activity and 8-bit clock prescaler for further reduction in power consumption. In the Idle mode the CPU is frozen while the peripherals and the interrupt system are still operating. In the power-down mode the RAM is saved and all other functions are inoperative.

The added features of the AT89C51RE2 make it more powerful for applications that need pulse width modulation, high speed I/O and counting capabilities such as alarms, motor control, corded phones, smart card readers.

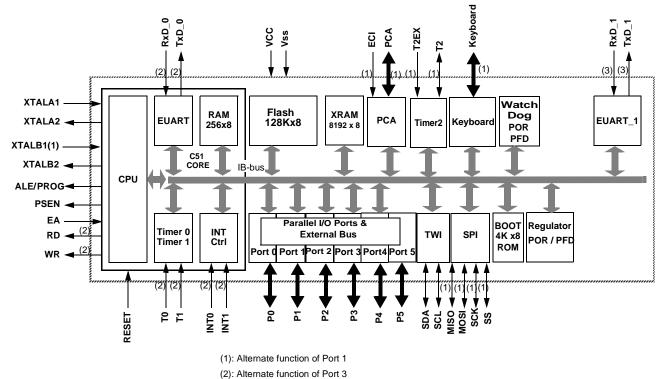
Table 1.	Memory Size and I/O pins	
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AT89C51RE2	Flash (bytes)	XRAM (bytes)	TOTAL RAM (bytes)	I/O
PLCC44 VQFP44	128K	8192	8192 + 256	34

# 2 AT89C51RE2

# **Block Diagram**

Figure 1. Block Diagram

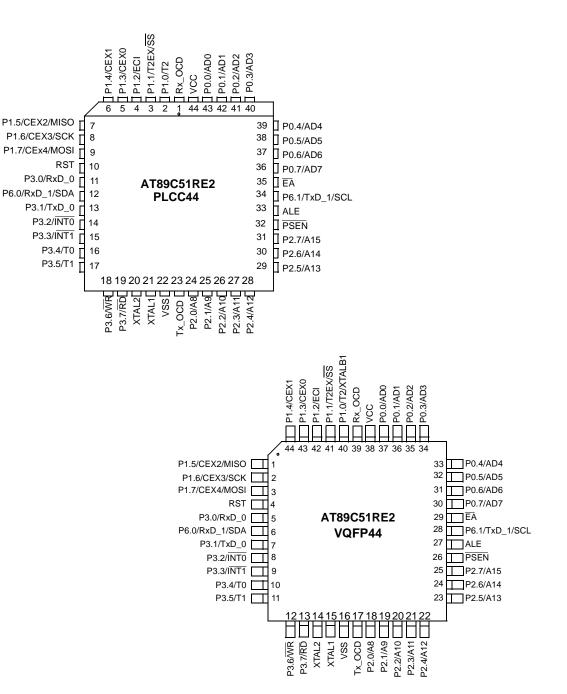


(3): Alternate function of Port 6





# **Pin Configurations**



	Pin	Number		
Mnemonic	LCC	VQFP 1.4	Туре	Name and Function
V <sub>SS</sub>	22	16	1	Ground: 0V reference
Vss1		39	I	Optional Ground: Contact the Sales Office for ground connection.
V <sub>cc</sub>	44	38	I	Power Supply: This is the power supply voltage for normal, idle and power-down operation
P0.0-P0.7	43-36	37-30	I/O	<b>Port 0</b> : Port 0 is an open-drain, bidirectional I/O port. Port 0 pins that have 1s written to them float and can be used as high impedance inputs. Port 0 must be polarized to $V_{CC}$ or $V_{SS}$ in order to prevent any parasitic current consumption. Port 0 is also the multiplexed low-order address and data bus during access to external program and data memory. In this application, it uses strong internal pull-up when emitting 1s. Port 0 also inputs the code bytes during EPROM programming. External pull-ups are required during program verification during which P0 outputs the code bytes.
P1.0-P1.7	2-9	40-44 1-3	I/O	Port 1: Port 1 is an 8-bit bidirectional I/O port with internal pull-ups. Port 1 pins that have 1s written to them are pulled high by the internal pull-ups and can be used as inputs. As inputs, Port 1 pins that are externally pulled low will source current because of the internal pull-ups. Port 1 also receives the low-order address byte during memory programming and verification. Alternate functions for TSC8x54/58 Port 1 include:
	2	40	I/O	T2 (P1.0): Timer/Counter 2 external count input/Clockout
	3	41	I	T2EX (P1.1): Timer/Counter 2 Reload/Capture/Direction Control
	4	42	I	ECI (P1.2): External Clock for the PCA
	5	43	I/O	CEX0 (P1.3): Capture/Compare External I/O for PCA module 0
	6	44	I/O	CEX1 (P1.4): Capture/Compare External I/O for PCA module 1
	7	1	I/O	CEX2 (P1.5): Capture/Compare External I/O for PCA module 2
	8	2	I/O	CEX3 (P1.6): Capture/Compare External I/O for PCA module 3
	9	3	I/O	CEX4 (P1.7): Capture/Compare External I/O for PCA module 4
P2.0-P2.7	24-31	18-25	I/O	Port 2: Port 2 is an 8-bit bidirectional I/O port with internal pull-ups. Port 2 pins that have 1s written to them are pulled high by the internal pull-ups and can be used as inputs. As inputs, Port 2 pins that are externally pulled low will source current because of the internal pull-ups. Port 2 emits the high-order address byte during fetches from external program memory and during accesses to external data memory that use 16-bit addresses (MOVX @DPTR).In this application, it uses strong internal pull-ups emitting 1s. During accesses to external data memory that use 8-bit addresses (MOVX @Ri), port 2 emits the contents of the P2 SFR. Some Port 2 pins receive the high order address bits during EPROM programming and verification: P2.0 to P2.5 for RB devices P2.0 to P2.7 for RD devices.
P3.0-P3.7	11, 13-19	5, 7-13	I/O	Port 3: Port 3 is an 8-bit bidirectional I/O port with internal pull-ups. Port 3 pins that have 1s written to them are pulled high by the internal pull-ups and can be used as inputs. As inputs, Port 3 pins that are externally pulled low will source current because of the internal pull-ups. Port 3 also serves the special features of the 80C51 family, as listed below.
	11	5	I	RXD_0 (P3.0): Serial input port
	13	7	0	TXD_0 (P3.1): Serial output port
	14	8	I	INT0 (P3.2): External interrupt 0





	Pir	n Number						
Mnemonic	LCC	VQFP 1.4	Туре	Name and Function				
	15	9	I	INT1 (P3.3): External interrupt 1				
	16	10	I	T0 (P3.4): Timer 0 external input				
	17	11	I	T1 (P3.5): Timer 1 external input				
	18	12	0	WR (P3.6): External data memory write strobe				
	19	13	0	RD (P3.7): External data memory read strobe				
P6.0-P6.1	12,34	6, 28		Port 6: Port 6 is an 2-bit bidirectional I/O port with internal pull-ups. Port 6 pins that have 1s written to them are pulled high by the internal pull-ups and can be used as inputs. As inputs, Port 6 pins that are externally pulled low will source current because of the internal pull-ups. Port 6 also serves some special features as listed below.				
	12	6	I	RXD_1 (P6.0): Serial input port				
	12	6	I/O	SDA (P6.0) : TWI Serial Data SDA is the bidirectional TWI data line.				
	34	28	0	TXD_1 (P6.1) : Serial output port				
	34	28	I/O	SCL (P6.1) : TWI Serial Clock SCL output the serial clock to slave peripherals. SCL input the serial clock from master.				
Reset	10	4	I/O	$\begin{array}{c} \textbf{Reset:} \ \text{A high on this pin for two machine cycles while the oscillator is running, resets} \\ \textbf{device. An internal diffused resistor to V_{SS} permits a power-on reset using only an exter capacitor to V_{CC}. This pin is an output when the hardware watchdog forces a system reset is the set of t$				
ALE/PROG	33	27	O (I)	Address Latch Enable/Program Pulse: Output pulse for latching the low byte of the address during an access to external memory. In normal operation, ALE is emitted at a constant rate of 1/6 (1/3 in X2 mode) the oscillator frequency, and can be used for external timing or clocking. Note that one ALE pulse is skipped during each access to external data memory. This pin is also the program pulse input (PROG) during Flash programming. ALE can be disabled by setting SFR's AUXR.0 bit. With this bit set, ALE will be inactive during internal fetches.				
PSEN	32	26	0	Program Store ENable: The read strobe to external program memory. When executing code from the external program memory, PSEN is activated twice each machine cycle, except that two PSEN activations are skipped during each access to external data memory. PSEN is not activated during fetches from internal program memory.				
EA	35	29	I	External Access Enable: EA must be externally held low to enable the device to fetch code				
				from external program memory locations 0000H to FFFFH (RD). If security level 1 is programmed, EA will be internally latched on Reset.				
XTAL1	21	15	I	<b>Crystal 1:</b> Input to the inverting oscillator amplifier and input to the internal clock generator circuits.				
XTAL2	20	14	0	Crystal 2: Output from the inverting oscillator amplifier				
Tx_OCD	23	17	0	Tx_OCD: On chip debug Serial output port				
Rx_OCD	1	39	I	Rx_OCD: On chip debug Serial input port				

# **SFR Mapping**

- The Special Function Registers (SFRs) of the AT89C51RE2 fall into the following categories:
  - C51 core registers: ACC, B, DPH, DPL, PSW, SP
  - I/O port registers: P0, P1, P2, P3, P4, P5, P6
  - Timer registers: T2CON, T2MOD, TCON, TH0, TH1, TH2, TMOD, TL0, TL1, TL2, RCAP2L, RCAP2H
  - Serial I/O port registers: SADDR\_0, SADEN\_0, SBUF\_0, SCON\_0, SADDR\_1, SADEN\_1, SBUF\_1, SCON\_1,
  - PCA (Programmable Counter Array) registers: CCON, CCAPMx, CL, CH, CCAPxH, CCAPxL (x: 0 to 4)
  - Power and clock control registers: PCON, CKAL, CKCON0\_1
  - Hardware Watchdog Timer registers: WDTRST, WDTPRG
  - Interrupt system registers: IE0, IPL0, IPH0, IE1, IPL1, IPH1
  - Keyboard Interface registers: KBE, KBF, KBLS
  - 2-wire Interface registers: SSCON, SSCS, SSDAT, SSADR
  - SPI registers: SPCON, SPSTR, SPDAT
  - BRG (Baud Rate Generator) registers: BRL\_0, BRL\_1, BDRCON\_0, BDRCON\_1
  - Memory register: FCON, FSTA
  - Clock Prescaler register: CKRL
  - Others: AUXR, AUXR1, CKCON0, CKCON1, BMSEL





### Table 3. C51 Core SFRs

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
ACC	E0h	Accumulator								
В	F0h	B Register								
PSW	D0h	Program Status Word	CY	AC	F0	RS1	RS0	OV	F1	Р
SP	81h	Stack Pointer								
DPL	82h	Data Pointer Low byte								
DPH	83h	Data Pointer High byte								

#### Table 4. System Management SFRs

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
PCON	87h	Power Control	SMOD1_0	SMOD0_0	-	POF	GF1	GF0	PD	IDL
AUXR	8Eh	Auxiliary Register 0	-	-	MO	XRS2	XRS1	XRS0	EXTRA M	AO
AUXR1	A2h	Auxiliary Register 1	EES	SP9	U2	-	GF2	0	-	DPS
CKRL	97h	Clock Reload Register	-	-	-	-	-	-	-	-
BMSEL	92h	Bank Memory Select	MBO2	MBO1	MBO0		-	FBS2	FBS1	FBS0
CKCON0	8Fh	Clock Control Register 0	TWIX2	WDX2	PCAX2	SIX2_0	T2X2	T1X2	T0X2	X2
CKCON1	AFh	Clock Control Register 1	-	-	-	-	-	-	SIX2_1	SPIX2

#### Table 5. Interrupt SFRs

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
IEN0	A8h	Interrupt Enable Control 0	EA	EC	ET2	ES	ET1	EX1	ET0	EX0
IEN1	B1h	Interrupt Enable Control 1	-	-	-	-	ES_1	ESPI	ETWI	EKBD
IPH0	B7h	Interrupt Priority Control High 0	-	PPCH	PT2H	PSH	PT1H	PX1H	PT0H	PX0H
IPL0	B8h	Interrupt Priority Control Low 0	-	PPCL	PT2L	PSL	PT1L	PX1L	PT0L	PX0L
IPH1	B3h	Interrupt Priority Control High 1	-	-	-	-	PSH_1	SPIH	IE2CH	KBDH
IPL1	B2h	Interrupt Priority Control Low 1	-	-	-	-	PSL_1	SPIL	IE2CL	KBDL

## Table 6. Port SFRs

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
P0	80h	8-bit Port 0								
P1	90h	8-bit Port 1								
P2	A0h	8-bit Port 2								
P3	B0h	8-bit Port 3								
P4	C0h	8-bit Port 4								

#### Table 6. Port SFRs

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
P5	E8h	8-bit Port 5								
P6	F8h	2-bit Port 5	-	-	-	-	-	-		

# Table 7. Flash and EEPROM Data Memory SFR

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
FCON	D1h	Flash Controller Control	FPL3	FPL2	FPL1	FPL0	FPS	FMOD2	FMOD1	FMOD0
FSTA	D3h	Flash Controller Status	FMR					FSE	FLOAD	FBUSY

## Table 8. Timer SFRs

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
TCON	88h	Timer/Counter 0 and 1 Control	TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0
TMOD	89h	Timer/Counter 0 and 1 Modes	GATE1	C/T1#	M11	M01	GATE0	C/T0#	M10	M00
TL0	8Ah	Timer/Counter 0 Low Byte								
TH0	8Ch	Timer/Counter 0 High Byte								
TL1	8Bh	Timer/Counter 1 Low Byte								
TH1	8Dh	Timer/Counter 1 High Byte								
WDTRST	A6h	WatchDog Timer Reset								
WDTPRG	A7h	WatchDog Timer Program	-	-	-	-	-	WTO2	WTO1	WTO0
T2CON	C8h	Timer/Counter 2 control	TF2	EXF2	RCLK	TCLK	EXEN2	TR2	C/T2#	CP/RL2#
T2MOD	C9h	Timer/Counter 2 Mode	-	-	-	-	-	-	T2OE	DCEN
RCAP2H	CBh	Timer/Counter 2 Reload/Capture High byte								
RCAP2L	CAh	Timer/Counter 2 Reload/Capture Low byte								
TH2	CDh	Timer/Counter 2 High Byte								
TL2	CCh	Timer/Counter 2 Low Byte								

## Table 9. PCA SFRs

Mnemo -nic	Add	Name	7	6	5	4	3	2	1	0
CCON	D8h	PCA Timer/Counter Control	CF	CR	-	CCF4	CCF3	CCF2	CCF1	CCF0
CMOD	D9h	PCA Timer/Counter Mode	CIDL	WDTE	-	-	-	CPS1	CPS0	ECF
CL	E9h	PCA Timer/Counter Low byte								





# Table 9. PCA SFRs (Continued)

Mnemo -nic	Add	Name	7	6	5	4	3	2	1	0
СН	F9h	PCA Timer/Counter High byte								
CCAPM0	DAh	PCA Timer/Counter Mode 0		ECOM0	CAPP0	CAPN0	MAT0	TOG0	PWM0	ECCF0
CCAPM1	DBh	PCA Timer/Counter Mode 1		ECOM1	CAPP1	CAPN1	MAT1	TOG1	PWM1	ECCF1
CCAPM2	DCh	PCA Timer/Counter Mode 2	-	ECOM2	CAPP2	CAPN2	MAT2	TOG2	PWM2	ECCF2
CCAPM3	DDh	PCA Timer/Counter Mode 3		ECOM3	CAPP3	CAPN3	MAT3	TOG3	PWM3	ECCF3
CCAPM4	DEh	PCA Timer/Counter Mode 4		ECOM4	CAPP4	CAPN4	MAT4	TOG4	PWM4	ECCF4
CCAP0H	FAh	PCA Compare Capture Module 0 H	CCAP0H7	CCAP0H6	CCAP0H5	CCAP0H4	CCAP0H3	CCAP0H2	CCAP0H1	CCAP0H0
CCAP1H	FBh	PCA Compare Capture Module 1 H	CCAP1H7	CCAP1H6	CCAP1H5	CCAP1H4	CCAP1H3	CCAP1H2	CCAP1H1	CCAP1H0
CCAP2H	FCh	PCA Compare Capture Module 2 H	CCAP2H7	CCAP2H6	CCAP2H5	CCAP2H4	CCAP2H3	CCAP2H2	CCAP2H1	CCAP2H0
ССАРЗН	FDh	PCA Compare Capture Module 3 H	CCAP3H7	CCAP3H6	CCAP3H5	CCAP3H4	CCAP3H3	CCAP3H2	CCAP3H1	CCAP3H0
CCAP4H	FEh	PCA Compare Capture Module 4 H	CCAP4H7	CCAP4H6	CCAP4H5	CCAP4H4	CCAP4H3	CCAP4H2	CCAP4H1	CCAP4H0
CCAP0L	EAh	PCA Compare Capture Module 0 L	CCAP0L7	CCAP0L6	CCAP0L5	CCAP0L4	CCAP0L3	CCAP0L2	CCAP0L1	CCAP0L0
CCAP1L	EBh	PCA Compare Capture Module 1 L	CCAP1L7	CCAP1L6	CCAP1L5	CCAP1L4	CCAP1L3	CCAP1L2	CCAP1L1	CCAP1L0
CCAP2L	ECh	PCA Compare Capture Module 2 L	CCAP2L7	CCAP2L6	CCAP2L5	CCAP2L4	CCAP2L3	CCAP2L2	CCAP2L1	CCAP2L0
CCAP3L	EDh	PCA Compare Capture Module 3 L	CCAP3L7	CCAP3L6	CCAP3L5	CCAP3L4	CCAP3L3	CCAP3L2	CCAP3L1	CCAP3L0
CCAP4L	EEh	PCA Compare Capture Module 4 L	CCAP4L7	CCAP4L6	CCAP4L5	CCAP4L4	CCAP4L3	CCAP4L2	CCAP4L1	CCAP4L0

## Table 10. Serial I/O Port SFRs

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
SCON_0	98h	Serial Control 0	FE/SM0_0	SM1_0	SM2_0	REN_0	TB8_0	RB8_0	TI_0	RI_0
SBUF_0	99h	Serial Data Buffer 0								
SADEN_0	B9h	Slave Address Mask 0								
SADDR_0	A9h	Slave Address 0								
BDRCON_0	9Bh	Baud Rate Control 0				BRR_0	TBCK_0	RBCK_0	SPD_0	SRC_0
BRL_0	9Ah	Baud Rate Reload 0								
SCON_1	C0h	Serial Control 1	FE_1/SM0_1	SM1_1	SM2_1	REN_1	TB8_1	RB8_1	TI_1	RI_1
SBUF_1	C1h	Serial Data Buffer 1								
SADEN_1	BAh	Slave Address Mask 1								
SADDR_1	AAh	Slave Address 1								
BDRCON_1	BCh	Baud Rate Control 1	SMOD1_1	SMOD0_1		BRR_1	TBCK_1	RBCK_1	SPD_1	SRC_1
BRL_1	BBh	Baud Rate Reload 1								

### Table 11. SPI Controller SFRs

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
SPCON	C3h	SPI Control	SPR2	SPEN	SSDIS	MSTR	CPOL	CPHA	SPR1	SPR0
SPSCR	C4h	SPI Status	SPIF		OVR	MODF	SPTE	UARTM	SPTEIE	MODFIE
SPDAT	C5h	SPI Data	SPD7	SPD6	SPD5	SPD4	SPD3	SPD2	SPD1	SPD0

## Table 12. Two-Wire Interface Controller SFRs

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
SSCON	93h	Synchronous Serial control	SSCR2	SSPE	SSSTA	SSSTO	SSI	SSAA	SSCR1	SSCR0
SSCS	94h	Synchronous Serial Status	SSC4	SSC3	SSC2	SSC1	SSC0	0	0	0
SSDAT	95h	Synchronous Serial Data	SSD7	SSD6	SSD5	SSD4	SSD3	SSD2	SSD1	SSD0
SSADR	96h	Synchronous Serial Address	SSA7	SSA6	SSA5	SSA4	SSA3	SSA2	SSA1	SSGC

### Table 13. Keyboard Interface SFRs

Mnemonic	Add	Name	7	6	5	4	3	2	1	0
KBLS	9Ch	Keyboard Level Selector	KBLS7	KBLS6	KBLS5	KBLS4	KBLS3	KBLS2	KBLS1	KBLS0
KBE	9Dh	Keyboard Input Enable	KBE7	KBE6	KBE5	KBE4	KBE3	KBE2	KBE1	KBE0
KBF	9Eh	Keyboard Flag Register	KBF7	KBF6	KBF5	KBF4	KBF3	KBF2	KBF1	KBF0





#### Table below shows all SFRs with their address and their reset value.

# Table 14. SFR Mapping

		Bit addressable			No	n Bit addressa	ble			
		0/8	1/9	2/A	3/B	4/C	5/D	6/E	7/F	
	F8h	P6 XXXX XX11	CH 0000 0000	CCAP0H XXXX XXXX	CCAP1H XXXX XXXX	CCAP2H XXXX XXXX	CCAP3H XXXX XXXX	CCAP4H XXXX XXXX		FFh
	F0h	B 0000 0000								F7h
	E8h	P5 1111 1111	CL 0000 0000	CCAP0L XXXX XXXX	CCAP1L XXXX XXXX	CCAP2L XXXX XXXX	CCAP3L XXXX XXXX	CCAP4L XXXX XXXX		EFh
	E0h	ACC 0000 0000								E7h
	D8h	CCON 00X0 0000	CMOD 00XX X000	CCAPM0 X000 0000	CCAPM1 X000 0000	CCAPM2 X000 0000	CCAPM3 X000 0000	CCAPM4 X000 0000		DFh
	D0h	PSW 0000 0000	FCON 0000 0000		FSTA xxxx x000					D7h
	C8h	T2CON 0000 0000	T2MOD XXXX XX00	RCAP2L 0000 0000	RCAP2H 0000 0000	TL2 0000 0000	TH2 0000 0000			CFh
C0h	U2(AUXR1.5) =0	SCON_1 0000 0000	SBUF_1		SPCON	SPSCR	SPDAT			C7h
Con	U2(AUXR1.5) =1	P4 1111 1111	0000 0000		0001 0100	0000 0000	XXXX XXXX			0/11
	B8h	IPL0 X000 000	SADEN_0 0000 0000	SADEN1 0000 0000	BRL_1 0000 0000	BDRCON_1 XXX0 0000				BFh
	B0h	P3 1111 1111	IEN1 XXXX 0000	IPL1 XXXX 0000	IPH1 XXXX 0111				IPH0 X000 0000	B7h
	A8h	IEN0 0000 0000	SADDR_0 0000 0000	SADDR_1 0000 0000					CKCON1 XXXX XX00	AFh
	A0h	P2 1111 1111		AUXR1 000x 11x0				WDTRST XXXX XXXX	WDTPRG XXXX X000	A7h
	98h	SCON_0 0000 0000	SBUF_0 XXXX XXXX	BRL_0 0000 0000	BDRCON_0 XXX0 0000	KBLS 0000 0000	KBE 0000 0000	KBF 0000 0000		9Fh
	90h	P1 1111 1111		BMSEL 0000 0YYY	SSCON 0000 0000	SSCS 1111 1000	SSDAT 1111 1111	SSADR 1111 1110	CKRL 1111 1111	97h
	88h	TCON 0000 0000	TMOD 0000 0000	TL0 0000 0000	TL1 0000 0000	TH0 0000 0000	TH1 0000 0000	AUXR XX00 1000	CKCON0 0000 0000	8Fh
	80h	P0 1111 1111	SP 0000 0111	DPL 0000 0000	DPH 0000 0000				PCON 00X1 0000	87h
		0/8	1/9	2/A	3/B	4/C	5/D	6/E	7/F	

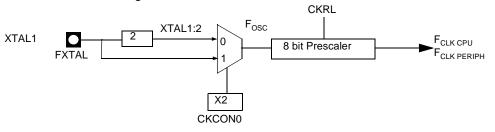
Reserved

12 AT89C51RE2

Enhanced Features	In comparison to the original 80C52, the AT89C51RE2 implements some new features, which are:
realures	X2 option
	Dual Data Pointer
	Extended RAM
	Extended stack
	Programmable Counter Array (PCA)
	Hardware Watchdog
	SPI interface
	<ul> <li>4-level interrupt priority system</li> </ul>
	<ul> <li>power-off flag</li> </ul>
	ONCE mode
	ALE disabling
	<ul> <li>Enhanced features on the UART and the timer 2</li> </ul>
X2 Feature	The AT89C51RE2 core needs only 6 clock periods per machine cycle. This feature called 'X2' provides the following advantages:
	<ul> <li>Divide frequency crystals by 2 (cheaper crystals) while keeping same CPU power.</li> </ul>
	• Save power consumption while keeping same CPU power (oscillator power saving).
	• Save power consumption by dividing dynamically the operating frequency by 2 in operating and idle modes.
	<ul> <li>Increase CPU power by 2 while keeping same crystal frequency.</li> </ul>
	In order to keep the original C51 compatibility, a divider by 2 is inserted between the XTAL1 sig- nal and the main clock input of the core (phase generator). This divider may be disabled by software.
Description	The clock for the whole circuit and peripherals is first divided by two before being used by the CPU core and the peripherals.
	This allows any cyclic ratio to be accepted on XTAL1 input. In X2 mode, as this divider is bypassed, the signals on XTAL1 must have a cyclic ratio between 40 to 60%.
	Figure 2 shows the clock generation block diagram. X2 bit is validated on the rising edge of the XTAL1 2 to avoid glitches when switching from X2 to STD mode. Figure 3 shows the switching

Figure 2. Clock Generation Diagram

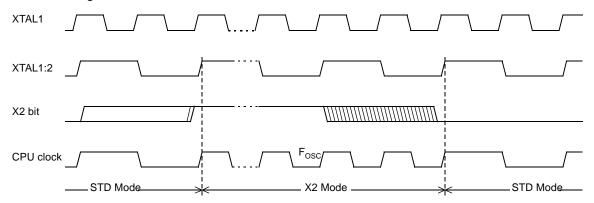
mode waveforms.







#### Figure 3. Mode Switching Waveforms



The X2 bit in the CKCON0 register (see Table 15) allows a switch from 12 clock periods per instruction to 6 clock periods and vice versa. At reset, the speed is set according to X2 bit of the Fuse Configuration Byte (FCB). By default, Standard mode is active. Setting the X2 bit activates the X2 feature (X2 mode).

The T0X2, T1X2, T2X2, UartX2, PcaX2, and WdX2 bits in the CKCON0 register (See Table 15.) and SPIX2 bit in the CKCON1 register (see Table 16) allows a switch from standard peripheral speed (12 clock periods per peripheral clock cycle) to fast peripheral speed (6 clock periods per peripheral clock cycle). These bits are active only in X2 mode.

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# Table 15. CKCON0 Register

CKCON0 - Clock Control Register (8Fh)

7	6	5	4	3	2	1	0				
TWIX2	WDX2	PCAX2	SIX2_0	T2X2	T1X2	T0X2	X2				
Bit Number	Bit Mnemonic	Description									
7	TWIX2	effect) Cleared to sel	This control bit is validated when the CPU clock X2 is set; when X2 is low, this bit has no								
6	WDX2	(This control b effect). Cleared to sel	<b>Vatchdog Clock</b> This control bit is validated when the CPU clock X2 is set; when X2 is low, this bit has no ffect). Eleared to select 6 clock periods per peripheral clock cycle. et to select 12 clock periods per peripheral clock cycle.								
5	PCAX2	(This control b effect). Cleared to sel	rogrammable Counter Array Clock This control bit is validated when the CPU clock X2 is set; when X2 is low, this bit has no ffect). leared to select 6 clock periods per peripheral clock cycle. Set to select 12 clock eriods per peripheral clock cycle.								
4	SIX2_0	(This control b effect). Cleared to sel	Enhanced UART0 Clock (Mode 0 and 2) (This control bit is validated when the CPU clock X2 is set; when X2 is low, this bit has no effect). Cleared to select 6 clock periods per peripheral clock cycle. Set to select 12 clock periods per peripheral clock cycle.								
3	T2X2	effect). Cleared to sel	it is validated v ect 6 clock per	vhen the CPU c iods per periph s per periphera	eral clock cycle		this bit has no				
2	T1X2	effect). Cleared to sel	it is validated v	vhen the CPU c iods per periph cycle.							
1	T0X2	effect). Cleared to sel	it is validated v	vhen the CPU c iods per periph cycle.							
0	X2	peripherals. S individual peri	<b>CPU Clock</b> Cleared to select 12 clock periods per machine cycle (STD mode) for CPU and all the peripherals. Set to select 6clock periods per machine cycle (X2 mode) and to enable the individual peripherals'X2' bits. Programmed by hardware after Power-up regarding Hardware Security Byte (HSB), Default setting, X2 is cleared.								

Reset Value = X000 000'HSB. X2'b (See "Fuse Configuration Byte: FCB") Not bit addressable





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# Table 16. CKCON1 Register

CKCON1 - Clock Control Register (AFh)

7	6	5	4	3	2	1	0				
-	-	-	-	-	-	SIX2_1	SPIX2				
Bit Number	Bit Mnemonic	Description									
7	-	Reserved									
6	-	Reserved									
5	-	Reserved									
4	-	Reserved	Reserved								
3	-	Reserved									
2	-	Reserved									
1	SIX2_1	(This control b effect). Cleared to sel		vhen the CPU c iods per periph							
0	SPIX2	has no effect). Clear to selec	t 6 clock period	ted when the C ds per periphera s per periphera	al clock cycle.	set; when X2 is	s low, this bit				

Reset Value = XXXX XX00b Not bit addressable



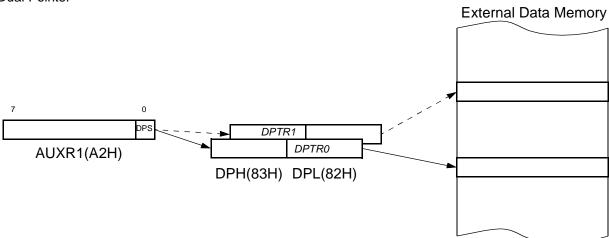


**Dual Data** The additional data pointer can be used to speed up code execution and reduce code size.

**Pointer Register** The dual DPTR structure is a way by which the chip will specify the address of an external data memory location. There are two 16-bit DPTR registers that address the external memory, and a single bit called DPS = AUXR1.0 (see Table 17) that allows the program code to switch between them (Refer to Figure 4).

Figure 4. Use of Dual Pointer

DPTR



#### Table 17. AUXR1 register

AUXR1- Auxiliary Register 1(0A2h)

7	6	5	4	3	2	1	0				
EES	SP9	U2	-	GF2	0	-	DPS				
Bit Number	Bit Mnemonic	Description	Description								
7	EES	This bit Set to enable	Enable Extended Stack         This bit allows the selection of the stack extended mode.         Set to enable the extended stack         Clear to disable the extended stack (default value)								
6	SP9	This bit has no Set when the	Stack Pointer 9th Bit This bit has no effect when the EES bit is cleared. Set when the stack pointer belongs to the XRAM memory space Cleared when the stack pointer belongs to the 256bytes of internal RAM.								
5	U2		SCON_1 regist	ter at C0h sfr ao at C0h address.							
4	-	<b>Reserved</b> The va	lue read from t	his bit is indete	rminate. Do not	set this bit.					
3	GF2	This b	it is a general	purpose user	flag. *						
2	0	Always	s cleared.								
1	-	<b>Reserved</b> The va	Reserved The value read from this bit is indeterminate. Do not set this bit.								
0	DPS	Data Pointer Selection Cleared to select DPTR0. Set to select DPTR1.									

Reset Value: XX0X XX0X0b

#### Not bit addressable

Note: \*Bit 2 stuck at 0; this allows to use INC AUXR1 to toggle DPS without changing GF3.

#### ASSEMBLY LANGUAGE

; Block move using dual data pointers ; Modifies DPTR0, DPTR1, A and PSW ; note: DPS exits opposite of entry state ; unless an extra INC AUXR1 is added ; 00A2 AUXR1 EQU 0A2H ; 0000 909000MOV DPTR,#SOURCE ; address of SOURCE

0000 909000MOV DPTR,#SOURCE ; address of SOURCE 0003 05A2 INC AUXR1 ; switch data pointers 0005 90A000 MOV DPTR,#DEST ; address of DEST 0008 LOOP: 0008 05A2 INC AUXR1 ; switch data pointers 000A E0 MOVX A,@DPTR ; get a byte from SOURCE 000B A3 INC DPTR ; increment SOURCE address 000C 05A2 INC AUXR1 ; switch data pointers 000E F0 MOVX @DPTR, 4 ; write the byte to DEST 000F A3 INC DPTR ; increment DEST address





0010 70F6JNZ LOOP ; check for 0 terminator 0012 05A2 INC AUXR1 ; (optional) restore DPS

INC is a short (2 bytes) and fast (12 clocks) way to manipulate the DPS bit in the AUXR1 SFR. However, note that the INC instruction does not directly force the DPS bit to a particular state, but simply toggles it. In simple routines, such as the block move example, only the fact that DPS is toggled in the proper sequence matters, not its actual value. In other words, the block move routine works the same whether DPS is '0' or '1' on entry. Observe that without the last instruction (INC AUXR1), the routine will exit with DPS in the opposite state.

# **Memory Architecture**

AT89C51RE2 features several on-chip memories:

- Flash memory: • containing 128 Kbytes of program memory (user space) organized into 128 bytes pages.
- Boot ROM: • 4K bytes for boot loader.
- 8K bytes internal XRAM •

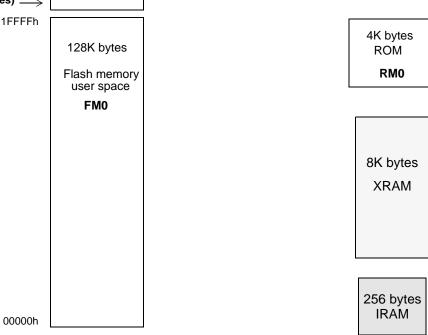
#### **Physical memory** organisation

Figure 5. Physical memory organisation

Fuse Configuration Byte(1 byte) FCB Hardware Security (1 byte)  $\longrightarrow$  HSB

Column Latches (128 bytes)  $\longrightarrow$ 

1FFFFh







# Expanded RAM<br/>(XRAM)The AT89C51RE2 provides additional Bytes of random access memory (RAM) space for<br/>increased data parameter handling and high level language usage.

AT89C51RE2 devices have expanded RAM in external data space configurable up to 8192bytes (see Table 18.).

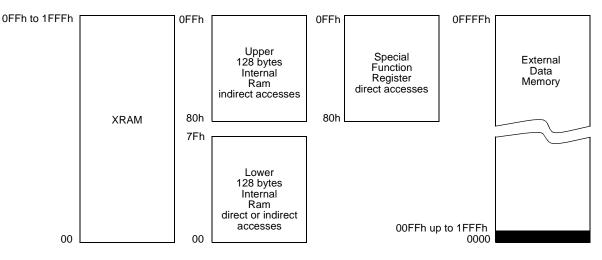
The AT89C51RE2 has internal data memory that is mapped into four separate segments.

The four segments are:

- 1. The Lower 128 bytes of RAM (addresses 00h to 7Fh) are directly and indirectly addressable.
- 2. The Upper 128 bytes of RAM (addresses 80h to FFh) are indirectly addressable only.
- The Special Function Registers, SFRs, (addresses 80h to FFh) are directly addressable only.
- 4. The expanded RAM bytes are indirectly accessed by MOVX instructions, and with the EXTRAM bit cleared in the AUXR register (see Table 18).

The lower 128 bytes can be accessed by either direct or indirect addressing. The Upper 128 bytes can be accessed by indirect addressing only. The Upper 128 bytes occupy the same address space as the SFR. That means they have the same address, but are physically separate from SFR space.





When an instruction accesses an internal location above address 7Fh, the CPU knows whether the access is to the upper 128 bytes of data RAM or to SFR space by the addressing mode used in the instruction.

- Instructions that use direct addressing access SFR space. For example: MOV 0A0H, # data, accesses the SFR at location 0A0h (which is P2).
- Instructions that use indirect addressing access the Upper 128 bytes of data RAM. For example: MOV @R0, # data where R0 contains 0A0h, accesses the data byte at address 0A0h, rather than P2 (whose address is 0A0h).
- The XRAM bytes can be accessed by indirect addressing, with EXTRAM bit cleared and MOVX instructions. This part of memory which is physically located on-chip, logically occupies the first bytes of external data memory. The bits XRS0 and XRS1 are used to hide a part of the available XRAM as explained in Table 18. This can be useful if external peripherals are mapped at addresses already used by the internal XRAM.

- With <u>EXTRAM = 0</u>, the XRAM is indirectly addressed, using the MOVX instruction in combination with any of the registers R0, R1 of the selected bank or DPTR. An access to XRAM will not affect ports P0, P2, P3.6 (WR) and P3.7 (RD). For example, with EXTRAM = 0, MOVX @R0, # data where R0 contains 0A0H, accesses the XRAM at address 0A0H rather than external memory. An access to external data memory locations higher than the accessible size of the XRAM will be performed with the MOVX DPTR instructions in the same way as in the standard 80C51, with P0 and P2 as data/address busses, and P3.6 and P3.7 as write and read timing signals. Accesses to XRAM above 0FFH can only be done by the use of DPTR.
- With <u>EXTRAM = 1</u>, MOVX @Ri and MOVX @DPTR will be similar to the standard 80C51.MOVX @ Ri will provide an eight-bit address multiplexed with data on Port0 and any output port pins can be used to output higher order address bits. This is to provide the external paging capability. MOVX @DPTR will generate a sixteen-bit address. Port2 outputs the high-order eight address bits (the contents of DPH) while Port0 multiplexes the low-order eight address bits (DPL) with data. MOVX @ Ri and MOVX @DPTR will generate either read or write signals on P3.6 (WR) and P3.7 (RD).

The stack pointer (SP) may be located anywhere in the 256 bytes RAM (lower and upper RAM) internal data memory. The stack may be located in the 256 lower bytes of the XRAM by activating the extended stack mode (see EES bit in AUXR1).

The M0 bit allows to stretch the XRAM timings; if M0 is set, the read and write pulses are extended from 6 to 30 clock periods. This is useful to access external slow peripherals.





# Registers

# Table 18. AUXR Register

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AUXR - Auxiliary Register (8Eh)

7	6	5	4	3	2	1	0
-	-	MO	XRS2	XRS1	XRS0	EXTRAM	AO

Bit Number	Bit Mnemonic	Description
7	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
6	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
5	MO	Pulse length         Cleared to stretch MOVX control: the RD/ and the WR/ pulse length is 6 clock periods (default).         Set to stretch MOVX control: the RD/ and the WR/ pulse length is 30 clock periods.
4-2	XRS2:0	XRAM Size           XRS2         XRS1         XRS0         XRAM size           0         0         0         256 bytes           0         0         1         512 bytes           0         1         0         768 bytes           0         1         1024 bytes           1         0         1792 bytes           1         0         1         2048 bytes           1         1         0         4096 bytes           1         1         8192 bytes (default)
1	EXTRAM	EXTRAM bit Cleared to access internal XRAM using movx @ Ri/ @ DPTR. Set to access external memory. Programmed by hardware after Power-up regarding Hardware Security Byte (HSB), default setting, XRAM selected.
0	AO	ALE Output bit Cleared, ALE is emitted at a constant rate of 1/6 the oscillator frequency (or 1/3 if X2 mode is used). (default) Set, ALE is active only during a MOVX or MOVC instruction is used.

Reset Value = XX01 1100b Not bit addressable

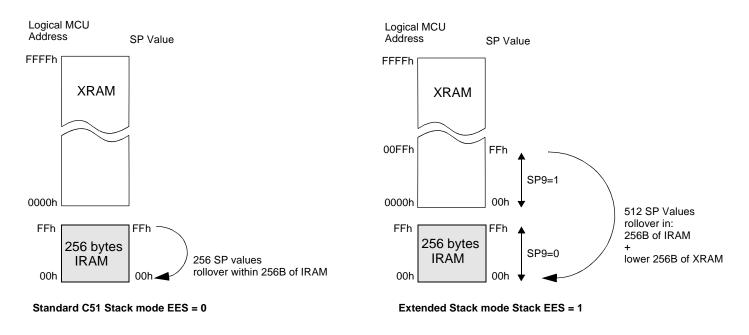
## **Extended Stack**

**ICk** The lowest bytes of the XRAM may be used to allow extension of the stack pointer.

The extended stack allows to extend the standard C51 stack over the 256 bytes of internal RAM. When the extended stack mode is activated (EES bit in AUXR1), the stack pointer (SP) can grow in the lower 256 bytes of the XRAM area.

The stack extension consists in a 9 bits stack pointer where the ninth bit is located in SP9 (bit 6 of AUXR1). The SP9 then indicates if the stack pointer belongs to the internal RAM (SP9 cleared) or to the XRAM memory (SP9 set).

To ensure backward compatibility with standard C51 architecture, the extended mode is disable at chip reset.



#### Figure 7. Stack modes

Figure 8. AUXR1 register

AUXR1- Auxiliary Register 1(0A2h)

7	6	5	4	3	2	1	0	
EES	SP9	U2	-	GF2	0	-	DPS	
Bit Number	Bit Mnemonic	Description						
7	EES	Set to enable	allows the selected s		ack extended m value)	iode.		





Bit Number	Bit Mnemonic	Description
6	SP9	Stack Pointer 9th Bit This bit has no effect when the EES bit is cleared. Set when the stack pointer belongs to the XRAM memory space Cleared when the stack pointer belongs to the 256bytes of internal RAM. Set and cleared by hardware. Can only be read.
5	U2	P4 bit addressable Clear to map SCON_1 register at C0h sfr address Set to map P4 port register at C0h address.
4	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
3	GF2	This bit is a general purpose user flag. *
2	0	Always cleared.
1	-	Reserved The value read from this bit is indeterminate. Do not set this bit.
0	DPS	Data Pointer Selection Cleared to select DPTR0. Set to select DPTR1.

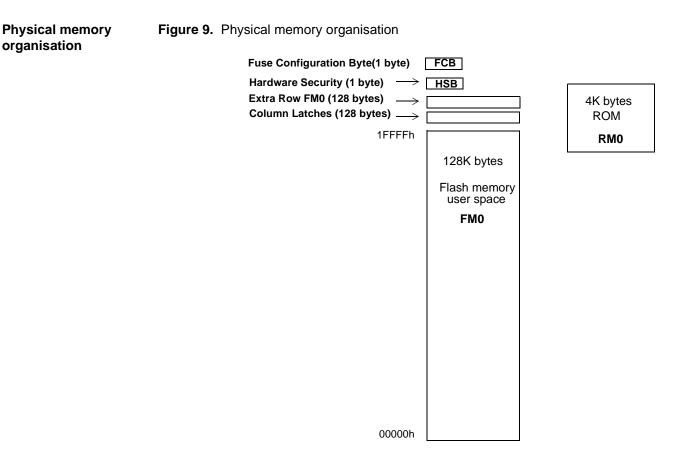
Reset Value = 00XX 00X0b Not bit addressable

# **Flash Memory**

General Description	The Flash memory increases EPROM and ROM functionality with in-circuit electrical erasure and programming. It contains 128K bytes of program memory organized in 1024 pages of 128 bytes. This memory is both parallel and serial In-System Programmable (ISP). ISP allows devices to alter their own program memory in the actual end product under software control. A default serial loader (bootloader) program allows ISP of the Flash.
	The programming does not require external high programming voltage. The necessary high pro- gramming voltage is generated on-chip using the standard V <sub>CC</sub> pins of the microcontroller.
Features	Flash internal program memory.
	<ul> <li>Boot vector allows user provided Flash loader code to reside anywhere in the Flash memory space. This configuration provides flexibility to the user.</li> </ul>
	<ul> <li>Default loader in Boot Flash allows programming via the serial port without the need of a user provided loader.</li> </ul>
	<ul> <li>Up to 64K byte external program memory if the internal program memory is disabled (EA = 0).</li> </ul>
	<ul> <li>Programming and erase voltage with standard 5V or 3V V<sub>CC</sub> supply.</li> </ul>
Flash memory	AT89C51RE2 features several on-chip memories:
organization	<ul> <li>Flash memory FM0: containing 128 Kbytes of program memory (user space) organized into 128 bytes pages.</li> <li>Boot ROM RM0: 4K bytes for boot loader.</li> <li>8K bytes internal XRAM</li> </ul>







# On-Chip Flash<br/>memoryThe AT89C51RE2 implements up to 128K bytes of on-chip program/code memory. Figure 9 and<br/>Figure 10 shows the partitioning of internal and external program/code memory spaces accord-<br/>ing to EA value.

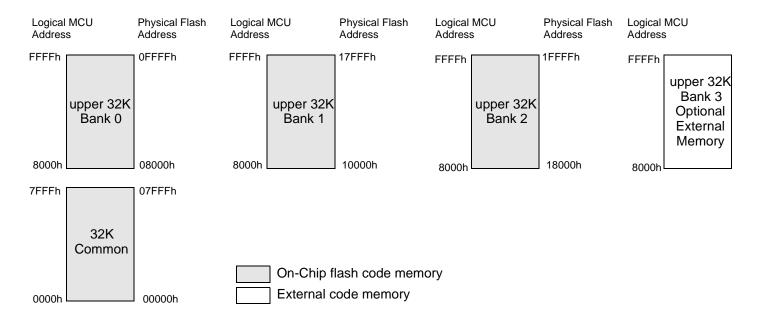
The memory partitioning of the 8051 core microcontroller is typical a Harvard architecture where program and data areas are held in separate memory areas. The program and data memory areas use the same physical address range from 0000H-FFFFH and a 8 bit instruction code/data format.

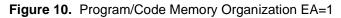
To access more than 64kBytes of code memory, without modifications of the MCU core, and development tools, the bank switching method is used.

The internal program memory is expanded to 128kByte in the 'Expanded Configuration', the data memory remains in the 'Normal Configuration'. The program memory is split into four 32 kByte banks (named Bank 0-2). The MCU core still addresses up to 64kBytes where the upper 32Kbytes can be selected between 3 32K bytes bank of on-chip flash memory. The lower 32K bank is used as common area for interrupt subroutines, bank switching and functions calls between banks.

The AT89C51RE2 also implements an extra upper 32K bank (Bank3) that allows external code execution.

# AT89C51RE2



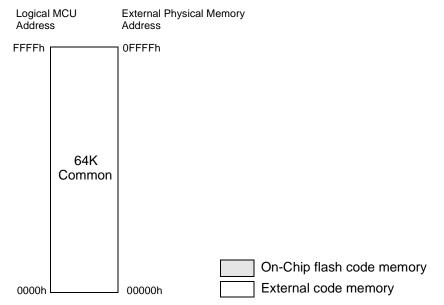






When EA=0, the on-chip flash memory is disabled and the MCU core can address only up to 64kByte of external memory (none of the on-chip flash memory FM0 banks or RM0 can be mapped and executed).

### Figure 11. Program/Code Memory Organization EA=0

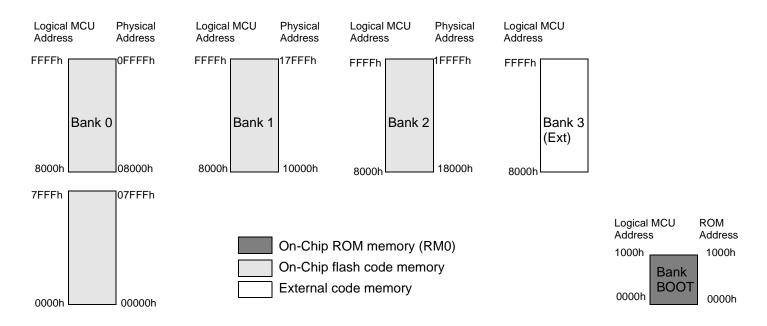


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# On-Chip ROM<br/>bootloaderThe On-chip ROM bootloader (RM0) is enable only for ISP operations after reset (bootloader<br/>execution). The RM0 memory area belongs to a logical addressable memory space called 'Bank<br/>Boot'.

**RM0 cannot be activated from the On-chip flash memory**. It means that it is not possible activate the Bank Boot area by software (it prevents any RM0 execution and flash corruption from the user application).

RM0 logical area consists in an independent code execution memory area of 4K bytes starting at logical 0x0000 address (it allows the use of the interrupts in the bootloader execution).







**Boot process** The BRV2-0 bits of the FCB (see Table 20 on page 34), the EA pin value upon reset and the presence of the external hardware conditions, allow to modify the default reset vector of the AT89C51RE2.

The Hardware conditions (EA = 1, PSEN = 0) during the Reset falling edge force the on-chip bootloader execution. This allows an application to be built that will normally execute the end user's code but can be manually forced into default ISP operation. The hardware conditions allows to force the enter in ISP mode whatever the configurations bits.

EA pin	Hardware conditions	BRV2-0	MCU reset vector		
0	Х	X External Code at address 0x0000			
	YES	Х	RM0 at address 0x0000 (ATMEL Bootloader)		
		111	FM0 at address 0x0000 with bank0 mapped		
		110	FM0 at address 0xFFFC in Bank 0		
		101	FM0 at address 0xFFFC in Bank 1		
1	NO	100	FM0 at address 0xFFFC in Bank 2		
	NO 0 1 1 RM0 at address 0x0000 (ATM		RM0 at address 0x0000 (ATMEL Bootloader)		
		010			
		0 0 1	Reserved (FM0 at address 0x0000 with bank 0 mapped)		
		000			

#### Figure 12. Boot Reset vector configuration

FM0 Memory Architecture	<ul> <li>The FM0 flash memory is made up of 5 blocks:</li> <li>1. The memory array (user space) 128K bytes</li> <li>2. The Extra Row also called FM0 XAF</li> <li>3. The Hardware security bits (HSB)</li> <li>4. The Fuse Configuration Byte (FCB)</li> <li>5. The column latch</li> </ul>
User Space	This space is composed of a 128K bytes Flash memory organized in 1024 pages of 128 bytes. It contains the user's application code. This block can be access in Read/write mode from FM0 and boot memory area. (When access in write mode from FM0, the CPU core enter pseudo idle mode).
Extra Row (XRow or XAF)	This row is a part of FM0 and has a size of 128 bytes. The extra row (XAF) may contain informa- tion for boot loader usage. This block can be access in Read/write mode from FM0 and boot memory area. (When access in write mode from FM0, the CPU core enter pseudo idle mode).
Hardware security Byte (HSB)	The Hardware security Byte is a part of FM0 and has a size of 1 byte. The 8 bits can be read/written by software (from FM0 or RM0) and written by hardware in paral- lel mode.
	The HSB bits can be written to '0' without any restriction (increase the security level of the chip), but can be written to '1' only when the corresponding memory area of the lock bits was full chip erased.

7	6	5	4	3	2	1	0		
-	-	-	-	-	FLB2	FLB1	FLB0		
Bit Number	Bit Mnemonic	Description	Description						
7	-	Unused	Unused						
6-4	-	Reserved	Reserved						
3	-	Unused							
2-0	FLB2-0	-	FM0 Memory Lock Bits See Table 32 on page 52						

 Table 19.
 Hardware Security Byte (HSB)





Fuse Configuration ByteThe Fuse configuration byte is a part of FM0.(FCB)The 8 bits read/written by software (from FM0

The 8 bits read/written by software (from FM0 or RM0) and written by hardware in parallel mode.

7	6	5	4	3	2	1	0		
X2	-	-	-	-	BRV2	BRV1	BRV0		
Bit Number	Bit Mnemonic	Description							
7	X2	X2 Mode Programmed ('0' value) to force X2 mode (6 clocks per instruction) after reset Unprogrammed ('1' value) to force X1 mode, Standard Mode, after reset (Default)							
6-3	-	Unused							
2-0	BRV2-0-	Boot Reset Vector These bits allow to configure the reset vector of the product according to the following values: 1 1 1: Reset at address 0x0000 of FM0 with Bank0 mapped 1 1 0: Reset at address 0xFFFC of Bank 0 1 0 1: Reset at address 0xFFFC of Bank 1 1 0 0: Reset at address 0xFFFC of Bank 2 0 1 1: Reset at address 0x0000 of RM0 (Internal ROM bootloader execution) 0 1 0: Reserved for further extension but same as 1 1 1 0 0 1: Reserved for further extension but same as 1 1 1 0 0 0: Reserved for further extension but same as 1 1 1							

# Table 20. Fuse Configuration Byte (FCB)

# AT89C51RE2

Column latches The column latches, also part of FM0, has a size of one page (128 bytes). The column latches are the entrance buffers of the three previous memory locations (user array, XROW, Hardware security byte and Fuse Configuration Byte).

This block is write only from FM0, RM0.

## **Cross Memory Access** The FM0 memory can be programmed from RM0 without entering idle mode.

Programming FM0 from FM0 makes the CPU core entering "pseudo idle" mode.

In the pseudo idle mode, the code execution is halted, the peripherals are still running (like standard idle mode) but all interrupt are delayed to the end of this mode. There are fours ways of exiting pseudo idle mode:

- At the end of the regular flash programming operation
- Reset the chip by external reset
- Reset the chip by hardware watchdog
- Reset the chip by PCA watchdog

Programming FM0 from external memory code (EA=0 or EA=1,with Bank3 active) is impossible.

If a reset occurs during flash programming the target page could be incompletely erased or programmed, but any other memory location (FM0, RAM, XRAM) remain unchanged.

The Table 21 shows all software flash access allowed.

 Table 21. Cross Memory Access

		Action	FM0 (user Flash)	RM0 (boot ROM)	
		Read	ok	Denied	
E	FM0 (user Flash)	Load column latch	ok	N.A.	
executing from	, , , , , , , , , , , , , , , , , , ,	Write	ok (pseudo idle mode)	N.A.	
ecutiı		Read	ok	ok	
	RM0 (boot ROM)	Load column latch	ok	N.A.	
Code	``````````````````````````````````````	Write	ok	N.A.	
	External memory	Read	(1)	Denied	
	EA = 0 or	Load column latch	Denied	N.A.	
	EA=1, Bank3	Write	Denied	N.A.	

1. Depends of general lock bits configuration

N.A. Not applicable



**Description overview** 



# Access and Operations Descriptions

#### **FM0 FLASH Registers**

The CPU interfaces to the flash memory through the FCON register, AUXR1 register and FSTA register.

These registers are used to map the columns latch, HSB, FCB and extra row in the working data or code space.

#### BMSEL Register

 Table 22.
 BMSEL Register

BMSEL Register (S:92h) Bank Memory Select

7	6	5	4	3	2	1	0	
MBO2	MBO1	MBO0			FBS2	FBS1	FBS0	
Bit Number	Bit Mnemonic	Description						
7-5	MBO2:0	<ul> <li>Memory Bank Operation</li> <li>These bits select the target memory bank for flash write or read operation. These bits allows to read or write the on-chip flash memory from one upper 32K bytes to another one.</li> <li>0 X X: The on-chip flash operation target banked is the same as FBS2:0</li> <li>1 0 0: The target memory bank is forced to Bank0</li> <li>1 0 1: The target memory bank is forced to Bank1</li> <li>1 1 0: The target memory bank is forced to Bank2</li> <li>1 1 1: The target memory bank is forced to Bank3 (optional External bank)</li> </ul>						
4-3		Reserved						
2-0	FBS2:0	FBS1:0 can be FBS2 is read-0 0 0 0 Bank0 0 0 1 Bank1 0 1 0 Bank2 0 1 1 Bank3 (c 1 X X Boot Ba	ect the upper 3 e read/write by only by softwar optionnal exter ank (Read onl	re (the Boot bar nal bank) <b>y)</b>	ution bank: nk can not be m o BRV2:0 config			

Reset Value= 0000 0YYYb (where YYY depends on BRV2:0 value in Fuse Configuration Byte)

#### FCON Register

Table 23. FCON Register

FCON Register (S:D1h) Flash Control Register

7	6	5	4	3	2	1	0	
FPL3	FPL2	FPL1	FPL0	FPS	FMOD2	FMOD1	FMOD0	
Bit Number	Bit Mnemonic	Description	Description					
7-4	FPL3:0		Programming Launch Command Bits Write 5Xh followed by AXh to launch the programming according to FMOD2:0. (see Table 26.)					
3	FPS	When this bit i The MOVX @ When this bit i	Flash Map Program Space When this bit is set: The MOVX @DPTR, A instruction writes in the columns latches space When this bit is cleared: The MOVX @DPTR, A instruction writes in the regular XDATA memory space					
2-0	FMOD2:0	Flash Mode These bits allo See Table 25.	Flash Mode These bits allow to select the target memory area and operation on FM0					

Reset Value= 0000 0000b





#### FSTA Register

Table 24. FSTA Register

FSTA Register (S:D3h) Flash Status Register

7	6	5	4	3	2	1	0	
FMR	-	-	-	-	FSE	FLOAD	FBUSY	
Bit Number	Bit Mnemonic	Description						
7	FMR	Flash Movc Redirection         When code is executed from RM0 (and only RM0), this bit allow the MOVC instruction to be redirected to FM0.         Clear this bit to allow MOVC instruction to read FM0         Set this bit to allow MOVC instruction to read RM0         This bit can be written only from RM0 (on-chip ROM bootloader execution).						
6-3	-	unused	unused					
2	FSE	Set by hardwa not correct (Se Clear by softw	<b>Flash sequence error</b> Set by hardware when the flash activation sequencers FCON 5X and MOV FCON AX) is not correct (See Error Report Section) Clear by software or clear by hardware if the last activation sequence was correct (previous error is canceled)					
1	FLOAD	Flash Columns latch loaded Set by hardware when the first data is loaded in the column latches. Clear by hardware when the activation sequence succeeds (flash write success, or reset column latch success)						
0	FBUSY	Flash Busy Set by hardware when programming is in progress. Clear by hardware when programming is done. Can not be changed by software.						

Reset Value= 'R'xxx x000b

Where 'R' depends on the reset conditions: If RM0 is executed after Reset R=1, if FM0 is executed after reset R=0

#### Mapping of the Memory Space

By default, the user space is accessed by MOVC A, @A+DPTR instruction for read only. Setting FPS bit in FCON register takes precedence on the EXTRAM bit in AUXR register.

The other memory spaces (user, extra row, hardware security) are made accessible in the code segment by programming bits FMOD2:0 in FCON register in accordance with Table 25. A MOVC instruction is then used for reading these spaces.

Thanks to the columns latches access, it is possible to write FM0 array, HSB and extra row blocks. The column latches space is made accessible by setting the FPS bit in FCON register. Writing is possible from 0000h to FFFFh, address bits 6 to 0 are used to select an address within a page while bits 14 to 7 are used to select the programming address of the page.

FMOD2	FMOD1	FMOD0	Adressable Space
0	0	0	FM0 array(0000h-FFFFh)
0	0	1	Extra Row(00h-80h)
0	1	0	Erase FM0
0	1	1	Column latches reset
1	0	0	HSB
1	0	1	FCB
1	1	0	Reserved
1	1	1	

Table 25.	.FM0 blo	cks select bits
Table 25.		CKS Select Dits





#### Launching flash commands (activation sequence)

FPL3:0 bits in FCON register are used to secure the launch of programming. A specific sequence must be written in these bits to unlock the write protection and to launch the operation. This sequence is 5xh followed by Axh. Table 26 summarizes the memory spaces to program according to FMOD2:0 bits.

P						
		١	Write to FCOI	N		
	FPL3:0	FPS	FMOD2	FMOD1	FMOD0	Operation
FM0	5	х	0	0	0	No action
FIVIO	А	х	0	0	0	Write the column latches in FM0
XAF	5	х	0	0	1	No action
FM0	А	х	0	0	1	Write the column latches in FM0 extra row space
Erase FM0	5	х	0	1	0	No action
ETASEFINIU	А	х	0	1	0	Full erase FM0 memory area
Reset	5	х	0	1	1	No action
FM0 Column Latches	А	х	0	1	1	Reset the FM0 column latches
	5	х	1	0	0	No action
HSB	А	х	1	0	0	Write the hardware Security byte (HSB) See <sup>(4)</sup>
	5	х	1	0	1	No action
FCB	А	х	1	0	1	Write the Fuse Configuration Byte (FCB)
Decensed	5	х	1	1	0	
Reserved	А	х	1	1	0	No action
Record	5	х	1	1	1	
Reserved	А	х	1	1	1	1

#### Table 26. FM0 Programming Sequences

Note: 1. The sequence 5xh and Axh must be executed without instructions between them otherwise the programming is not executed (see flash status register).

2. The sequence 5xh and Axh can be executed with the different FMOD0, FMOD1 values, the last FMOD1:0 value latches the destination target.

3. When the FMOD2 bit is set (corresponding to the serial number field code) no write operation can be performed.

4. Only the bits corresponding to the previously "full erase" memory space can be written to one.

## Loading the Column Latches

Any number of data from 0 byte to 128 bytes can be loaded in the column latches. The data written in the column latches can be written in a none consecutive order. The DPTR allows to select the address of the byte to load in the column latches.

The page address to be written (target page in FM0) is given by the last address loaded in the column latches and when this page belongs to the upper 32K bytes of the logical addressable MCU space, the target memory bank selection is performed upon the MBO2:0 value during the last address loaded.

When 0 byte is loaded in the column latches the activation sequence (5xh, Axh in FCON) does not launch any operations. The FSE bit in FSTA register is set.

When a current flash write operation is on-going (FBUSY is set), it is impossible to load the columns latches before the end of flash programming process (the write operation in the columns latches is not performed, and the previous columns latches content is not overwritten).

When programming is launched, an **automatic erase** of the entire memory page is first performed, then programming is effectively done. Thus no page or block erase is needed and only the loaded data are programmed in the corresponding page. The unloaded data of the target memory page are programmed at 0xFF value (automatic page erase value).

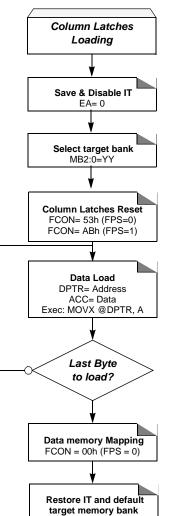
The following procedure is used to load the column latches and is summarized in Figure 13:

- Disable interrupt and map the column latch space by setting FPS bit.
- Select the target memory bank (for page address larger than 32K)
- Map the column latch
- Reset the column latch
- Load the DPTR with the address to write.
- Load Accumulator register with the data to write.
- Execute the MOVX @DPTR, A instruction, and only this one (no MOVX @Ri, A).
- If needed loop the last three instructions until the page is completely loaded.
- Unmap the column latch if needed (it can be left mapped) and Enable Interrupt





Figure 13. Column Latches Loading Procedure



Note: The last page address used when loading the column latch is the one used to select the page programming address.

Note: The value of MB02:0 during the last load gives the upper 32K bytes bank target selection.

Note: The execution of this sequence when BUSY flag is set leads to the no-execution of the write in the column latches (the previous loaded data remains unchanged).

# Writing the Flash Spaces

User

The following procedure is used to program the User space and is summarized in Figure 14:

- Load up to one page of data in the column latches from address 0000h to FFFFh (see Figure 13.).
- Disable the interrupts.

•

Launch the programming by writing the data sequence 50h followed by A0h in FCON register.

The end of the programming indicated by the FBUSY flag cleared.

• Enable the interrupts.

## 42 **AT89C51RE2**

#### Extra Row

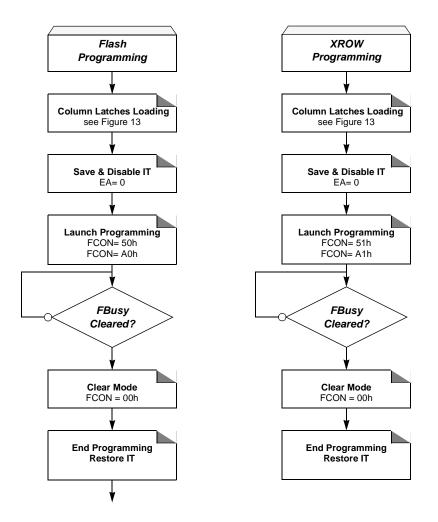
The following procedure is used to program the Extra Row space and is summarized in Figure 14:

- Load data in the column latches from address FF80h to FFFFh.
- Disable the interrupts.
- Launch the programming by writing the data sequence 51h followed by A1h in FCON register.

The end of the programming indicated by the FBUSY flag cleared.

Enable the interrupts.

Figure 14. Flash and Extra row Programming Procedure



Hardware Security ByteThe following procedure is used to program the Hardware Security Byte space and is<br/>summarized in Figure 15:

- Set FPS and map Hardware byte (FCON = 0x0C)
- Save and disable the interrupts.
- Load DPTR at address 0000h
- Load Accumulator register with the data to load.
- Execute the MOVX @DPTR, A instruction.



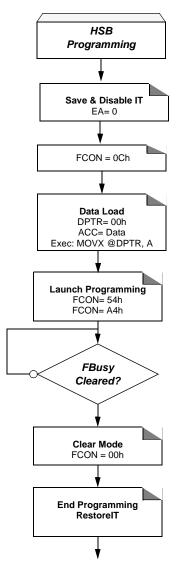


• Launch the programming by writing the data sequence 54h followed by A4h in FCON register.

The end of the programming indicated by the FBusy flag cleared.

Restore the interrupts

Figure 15. Hardware Security Byte Programming Procedure



*Fuse Configuration Byte* The following procedure is used to program the Fuse Configuration Byte space and is summarized in Figure 16:

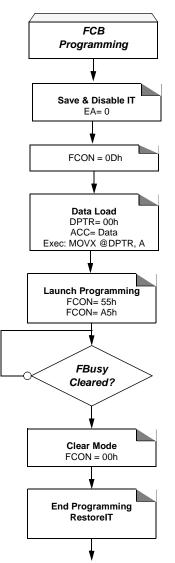
- Set FPS and map FCB (FCON = 0x0D)
- Save and disable the interrupts.
- Load DPTR at address 0000h
- Load Accumulator register with the data to load.
- Execute the MOVX @DPTR, A instruction.

• Launch the programming by writing the data sequence 55h followed by A5h in FCON register.

The end of the programming indicated by the FBusy flag cleared.

Restore the interrupts

Figure 16. Fuse Configuration Byte Programming Procedure



Reset of columns latches space

No automatic reset of the columns latches is performed after a successful flash write process. Resetting the columns latches during a flash write process is mandatory. User shall implement a reset of the column latch before each column latch load sequence.





In addition, the user application can reset the columns latches space manually. The following procedure is used to reset the columns latches space

Launch the programming by writing the data sequence 53h followed by A3h in FCON register (from FM0 and RM0).

#### Errors Report / Miscellaneous states

Flash Busy flag	The FBUSY flag indicates on-going flash write operation.
	The busy flag is set by hardware, the hardware clears this flag after the end of the programming operation.
Flash Programming	When a wrong sequence is detected the FSE in FSTA is set.
Sequence Error	The following events are considered as not correct activation sequence:
	- The two "MOV FCON,5x and MOV FCON, Ax" were not consecutive, or the second instruction differs from "MOV FCON Ax" (for example, an interrupt occurs during the sequence).
	- The sequence (write flash or reset column latches) occurred with no data loaded in the column latches
	The FSE bit can be cleared:
	- By software
	- By hardware when a correct programming sequence occurs.
	Note: When a good sequence occurs just after an incorrect sequence, the previous error is lost. The user software application should take care to check the FSE bit before initiating a new sequence.
Power Down Mode	
Request	In Power Down mode, the on-chip flash memory is deselected (to reduce power consumption), this leads to the lost of the columns latches content.
	In this case, if columns latches were previously loaded they are reset: FLOAD bit in FSTA regis- ter should be reset after power down mode.
	If a nower down mode is requested during flash programming (FBUSY-1), all power down

If a power down mode is requested during flash programming (FBUSY=1), all power down sequence instructions should be ignored until the end of flash process.





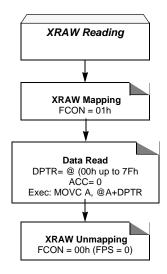
# Reading the Flash Spaces

User	<ul> <li>The following procedure is used to read the User space:</li> <li>Read one byte in Accumulator by executing MOVC A,@A+DPTR</li> <li>Note: FCON is supposed to be reset when not needed.</li> <li>Depending of the MBO2:0 bits, the MOVC A,@A+DPTR can address a specific upper 32K bytes bank. It allows to read the 32K bytes upper On-chip flash memory from one bank to another one.</li> </ul>
	When read from the bootloader area, the user memory shall be mapped before any read access by setting the FMR bit of the FSTA register.
	By default, when the bootloader is entered by hardware conditions, the ROM area is mapped for MOVC A,@A+DPTR operations. It is necessary to remap the user memory before each read access.
	The following presedure is used to read the Eutre Devrements and is supressized in Figure 17.
Extra Row (XAF)	<ul> <li>The following procedure is used to read the Extra Row space and is summarized in Figure 17:</li> <li>Map the Extra Row space by writing 01h in FCON register.</li> </ul>
	<ul> <li>Read one byte in Accumulator by executing MOVC A,@A+DPTR with A= 0 &amp; DPTR= 0000h</li> </ul>

• Clear FCON to unmap the Extra Row.

Figure 17. XAF Reading Procedure

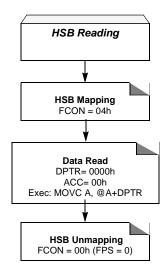
to 007Fh.



Hardware Security Byte The following procedure is used to read the Hardware Security space and is summarized in Figure 18:

- Map the Hardware Security space by writing 04h in FCON register.
- Read the byte in Accumulator by executing MOVC A, @A+DPTR with A= 0 & DPTR= 0000h.
- Clear FCON to unmap the Hardware Security Byte.

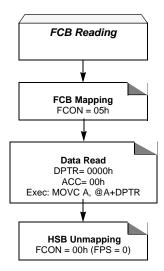
#### Figure 18. HSB Reading Procedure



*Fuse ConfigurationByte* The following procedure is used to read the Fuse Configuration byte and is summarized in Figure 18:

- Map the FCB by writing 05h in FCON register.
- Read the byte in Accumulator by executing MOVC A, @A+DPTR with A= 0 & DPTR= 0000h.
- Clear FCON to unmap the Hardware Security Byte.

**HSB** Reading Procedure







### **Operation Cross Memory Access**

Space addressable in read and write are:

- RAM
- ERAM (Expanded RAM access by movx)
- XRAM (eXternal RAM)
- FM0 (user flash)
- Hardware byte
- XROW FM0
- Boot RM0
- Flash Column latch

The table below provide the different kind of memory which can be accessed from different code location.

	Action	RAM	XRAM ERAM	boot RM0	FMO	HSB	FCB	XAF FM0
boot RM0	Read	ok	ok	ok	ok	ok	ok	ok
	Write	ok	ok	-	ok (RWW)	ok (RWW)	ok (RWW)	ok (RWW)
FMO	Read	ok	ok	-	ok	ok	ok	ok
FINU	Write	ok	ok	-	ok (idle)	ok	ok	ok
External	Read	ok	ok	-	-	-	-	-
EA = 0 or BANK3	Write	ok	ok	-	-	-	-	-

#### Table 27. Cross Memory Access

## **Sharing Instructions**

#### Table 28. Instructions shared

Action	RAM	XRAM	RM0	CL FM0	FM0	HSB	XAF FM0
Read	MOV	MOVX A,@DPTR	MOVC A, @A+DPTR	-	MOVC A, @A+DPTR	MOVC A, @A+DPTR	MOVC A, @A+DPTR
Write	MOV	MOVX @DPTR,A	-	MOVX @DPTR,A	by CL FM0	by CL FM0	by CL FM0

Note: by cl: using Column Latch

#### Table 29. Write MOVX @DPTR,A

FPS of FCCON	EA	XRAM ERAM	CL FM0
0	Х	winner	
1	1		winner
	0	winner	

Table 30. MOVC A, @A+DPTR executed from External code EA=0

FMOD2:0	FBS (Fetch)	MBO (Target)	MOVC A,@A+DPTR
Х	Х	Х	Read External Code

Table 31. MOVC A, @A+DPTR executed from External code EA=1, PC>=0x8000, FBS=Bank3

FMOD2:0	MBO (Target)	DPTR	MOVC A,@A+DPTR
x	х	< 0x8000	Depends on FLB2:0 Can Returns Random value, for secured part.
	Х	>= 0x8000	External code read





## Flash Protection from Parallel Programming

The three lock bits in Hardware Security Byte (see "In-System Programming" section) are programmed according to Table 32 provide different level of protection for the on-chip flash memory FM0.

They are set by default to level 4

Table 32. Program Lock Bit FLB2-0

Pro	Program Lock Bits			
Security level	FLB0	FLB1	FLB2	Protection Description
1	U	U	U	No program lock features enabled.
2	Ρ	U	U	MOVC instruction executed from external program memory are disabled from fetching code bytes from internal memory, EA is sampled and latched on reset, and further parallel programming of the Flash is disabled. ISP allows only flash verification (no write operations are allowed) but IAP from internal code still allowed.
3	U	Ρ	U	Same as 2, also verify through parallel programming interface is disabled and ISP read operation not allowed.
4	U	U	Р	Same as 3, also external execution is disabled (external bank not accessible)

Program Lock bits

U: unprogrammed

P: programmed

WARNING: Security level 2 and 3 should only be programmed after verification.

### **Bootloader Architecture**

Introduction

The bootloader manages a communication between a host platform running an ISP tool and a AT89C51RE2 target.

The bootloader implemented in AT89C51RE2 is designed to reside in the dedicated ROM bank. This memory area can only be executed (fetched) when the processor enters the boot process.

The implementation of the bootloader is based on standard set of libraries including INTEL hex based protocol, standard communication links and ATMEL ISP command set.

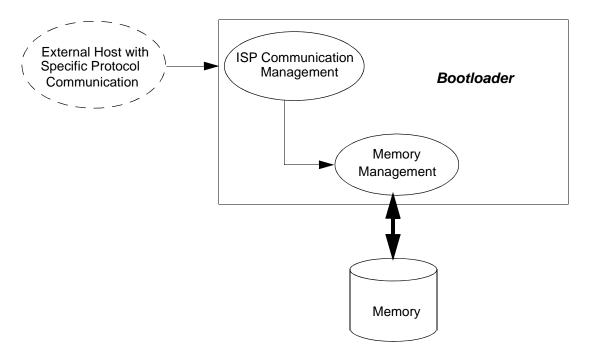


Figure 19. Bootloader Functional Description

On the above diagram, the on-chip bootloader processes are:

ISP Communication Management

The purpose of this process is to manage the communication and its protocol between the onchip bootloader and a external device. The on-chip ROM implement a serial protocol (see section Bootloader Protocol). This process translate serial communication frame (UART) into Flash memory access (read, write, erase...).

Memory Management

This process manages low level access to Flash memory (performs read and write access).





### Bootloader Description

Entry points After reset only one bootloader entry point is possible. This entry point stands at address 0x0000 of the boot ROM memory. This entry point executes the boot process of the bootloader.

The bootloader entry point can be selected through two processes:

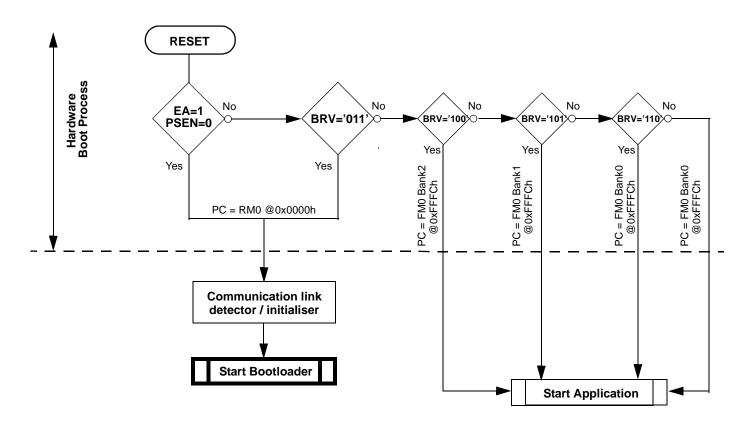
At reset, if the hardware conditions are applied, the bootloader entry point is accessed and executed.

At reset, if the hardware conditions are not set and the BRV2-0 is programmed '011', the bootloader entry point is accessed and the bootprocess is started.

#### Boot Process Description

The boot process consists in three main operations:

- The hardware boot process request detection
- The communication link detection (Uart or OCD)
- The start-up of the bootloader



Hardware boot process request detection The hardware boot process request is detected when the hardware conditions (under reset, EA=1 and PSEN=0) are received by the processor or when no hardware condition is applied and the BRV2:0 is configured '011'.

Communication link detection

- Two interfaces are available for ISP:
- UART0
- OCD UART

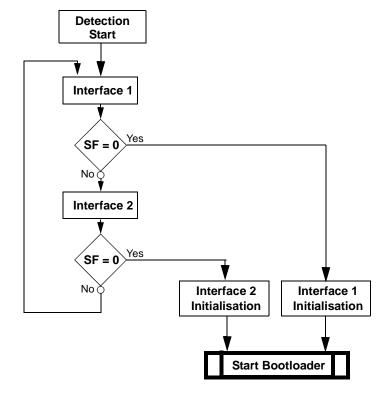
## 54 AT89C51RE2

The communication link detection is done by a circular polling on all the interfaces. On AT89C51RE2, the ISP interfaces are all based on simple UART mechanisms (Rx, Tx).

The Rx line default state is '1' when no communication is in progress. A transition from '1' to '0' on the Rx line indicates a start of frame.

Once one of the interface detects a starts of frame ('0') on its Rx line, the interface is selected and configuration of the communication link starts.





Notes: 1. SF: Start of Frame ('0' = detected; '1' = not detected)

2. In AT89C51RE2 implementation, Interface 1 refers to UART0 and Interface 2 refers to the OCD UART interface.





### ISP Protocol Description

**Physical Layer** 

The UART used to transmit information has the following configuration:

- Character: 8-bit data
- Parity: none
- Stop: 1 bit
- Flow control: none
- Baud rate: autobaud is performed by the bootloader to compute the baud rate chosen by the host.

**Frame Description** The Serial Protocol is based on the Intel Extended Hex-type records.

Intel Hex records consist of ASCII characters used to represent hexadecimal values and are summarized below.

#### Table 33. Intel Hex Type Frame

Record Mark ':'	Record length	Load Offset	Record Type	Data or Info	Checksum
1 byte	1 byte	2 bytes	1 bytes	n byte	1 byte

- Record Mark:
  - Record Mark is the start of frame. This field must contain':'.
- Record length:
  - Record length specifies the number of Bytes of information or data which follows the Record Type field of the record.
- Load Offset:
  - Load Offset specifies the 16-bit starting load offset of the data Bytes, therefore this field is used only for
  - Data Program Record.
- Record Type:
  - Record Type specifies the command type. This field is used to interpret the remaining information within the frame.
- Data/Info:
  - Data/Info is a variable length field. It consists of zero or more Bytes encoded as pairs of hexadecimal digits. The meaning of data depends on the Record Type.
- Checksum:
  - Checksum is the two's complement of the 8-bit Bytes that result from converting each pair of ASCII hexadecimal digits to one Byte of binary, thus including all field from the Record Length field to the last Byte of the Data/Info field. Therefore, the sum of all the ASCII pairs in a record after converting to binary, including all field from the Record Length field to the Checksum field, is zero.

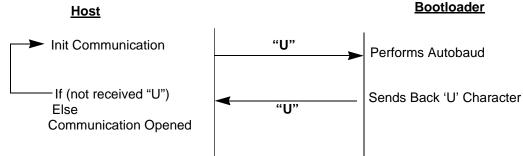
#### Protocol

### **Overview** An initialization step must be performed after each Reset. After microcontroller reset, the bootloader waits for an autobaud sequence (see Section "Autobaud Performances").

When the communication is initialized the protocol depends on the record type issued by the host.

**Communication** The host initiates the communication by sending a 'U' character to help the bootloader to compute the baudrate (autobaud).

Figure 21. Initialization

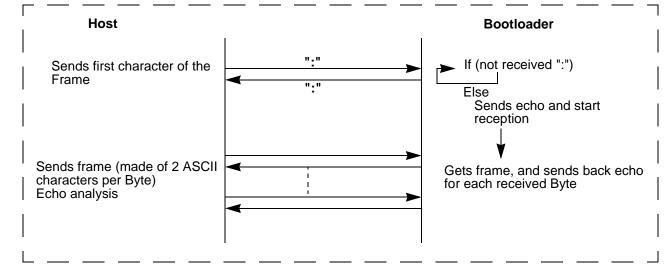


Autobaud The bootloader supports a wide range of baud rates. It is also adaptable to a wide range of oscillator frequencies. This is accomplished by measuring the bit-time of a single bit in a received character. This information is then used to program the baud rate in terms of timer counts based on the oscillator frequency.

#### Command Data Stream Protocol

All commands are sent using the same flow. To increase performance, the echo has been removed from the bootloader response.

#### Figure 22. Command Flow



Each command flow may end with:





- "X": If checksum error
- "L": If read security is set
- "P": If program security is set
- ".": If command ok
- byte + ".": read byte ok

#### Reading/Blank checking memory

To start the reading or blank checking operation,

Requests from Host

Command	Record Type	Record Length	Offset	Data[0]	Data[1]	Data[2]	Data[3]	Data[4]
Read selected memory	04h	05h	0000h	Start Address		End Address		00h
Blank Check selected memory	040	001	000011	Start P	1001035	End A	001035	01h

Answers from Bootloader The boot loader can answer to a read command with:

- 'Address = data '& 'CR' &'LF' the number of data by line depends of the bootloader.
- 'X' & 'CR' & 'LF' if the checksum is wrong
- 'L' & 'CR' & 'LF' if the Security is set

The bootloader answers to blank check command:

- '.' & 'CR' &'LF' when the blank check is ok
- 'First Address wrong' 'CR' & 'LF' when the blank check is fail
- 'X' & 'CR' & 'LF' if the checksum is wrong
- 'L' & 'CR' & 'LF' if the Security is set

Changing memory/page

To change the memory selected and/or the page, the Host can send two commands.

- Select New Page to keep the same memory.
- Select Memory to change the Memory and page

#### Requests from Host

Command	Record Type	Record Length	Offset	Data[0]	Data[1]
Select New Page	02h	02h	start address	Page (4 bits) + 0h	00h
Select Memory	04h	02h	0000h	Memory space	Page

Answers from Bootloader

- The boot loader can answer to a read command with:
- '. '& 'CR' &'LF' if the command is done
- 'X' & 'CR' & 'LF' if the checksum is wrong





# Programming/Erasing memory

Requests from Host

Command	Record Type	Record Length	Offset	Data[0]	Data[1]	Data[2]	Data[3]	Data[4]
Program selected memory	00h	nb of data	start address	x	x	x	x	x
Erase selected memory	04h	05h	0000h	00h	FFh	00h	00h	02h

Answers from Bootloader The boot loader answers with:

- '.' & 'CR' &'LF' when the data are programmed
- 'X' & 'CR' & 'LF' if the checksum is wrong
- 'P' & 'CR' & 'LF' if the Security is set

**Starting application** 

**on** The application can only be started by a Watchdog reset.

No answer is returned by the bootloader.

#### Requests from Host

Command	Record Type	Record Length	Offset
Start application with watchdog	01h	00h	0000h

#### **ISP** Commands description

Select Memory Space The 'Select Memory Space' command allows to route all read, write commands to a selected area. For each area (Family) a code is defined. This code corresponds to the memory area encoded value in the INTEL HEX frame.

The area supported and there coding are listed in the table below.

Table 34.	Memory	/ Families	&	coding
-----------	--------	------------	---	--------

Memory/Information Family	coding*	name
FLASH	0	MEM_FLASH
SECURITY	7	MEM_PROTECT
CONFIGURATION	8	MEM_CONF
BOOTLOADER	3	MEM_BOOT
SIGNATURE	6	MEM_SIGNATURE

The Bootloader information and the signature areas are read only. The value in the coding column is the value to report in the corresponding protocol field.

Note: \* the coding number doesn't include any information on the authorized address range of the family. A summary of these addresses is available in appendix (See "Address Mapping" on page 67.)





#### Select Page

The 'Select Page' command allows to define a page number in the selected area. A page is defined as a 64K linear memory space (According to the INTEL HEX format). It doesn't corresponds to a physical bank from the processor.

The following table summarizes the memory spaces for which the select page command can be applied.

Table 35.	Memory space &	& Select page
-----------	----------------	---------------

Memory/Information Family	Comments/Restriction		
FLASH	page 0 (0->64K) and 1(64k->128k) available		

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Write commands The following table summarizes the memory spaces for which the write command can be applied.

Table 36.	Memory	space &	Select page
-----------	--------	---------	-------------

Memory/Information Family	Comments/Restriction
FLASH	need security level check
SECURITY	only a higher level can be write
CONFIGURATION	

In case of write command to other area, nothing is done.

The bootloader returns a Write protection ('P') if the SECURITY do not allow any write operation from the bootloader.

The program/data Flash memory area can be programmed by the bootloader by data pages of up to 128bytes.

If the Flash memory security level is at least '2' (FLB2:0 = '110'), no write operation can be performed through the bootloader.

**Table 37.** Flash Write Authorization Summary

	Security level (HSB)				
		FLB2:0			
Command	111	110	101	011	
Write	Allowed	Forbidden	Forbidden	Forbidden	

CONFIGURATION The FCB configuration byte can always be written, whatever are the security levels.

SECURITY The Security byte can always be written with a value that enables a protection higher than the previous one.

If attempting to write a lower security, no action is performed and the bootloader returns a protection error code ('P')

#### Table 38. Security Write Authorization Summary

	Security level (HSB)					
write from	to FLB2:0					
FLB2:0	111	110	101	011		
111	Allowed	Allowed	Allowed	Allowed		
110	Forbidden	Allowed	Allowed	Allowed		
101	Forbidden	Forbidden	Allowed	Allowed		
011	Forbidden	Forbidden	Forbidden	Allowed		



FLASH



#### **Erasing commands** The erasing command is supported by the following areas:

#### Table 39. Memory space & Erase

Memory/Information Family	Comments/Restriction
FLASH	need security level check

Nothing is done on the other areas.

FLASH

The erasing command on the Flash memory:

- erases the four physical flash memory banks (from address 0000h to 1FFFFh).
- the HSB (Hardware Security Byte) is set at NO\_PROTECTION:

– FLB2.0 = '111'

# Blank Checking commands

FLASH

The blank checking command is supported by the following areas

#### Table 40. Memory space & Erase

Memory/Information Family	Comments/Restriction
FLASH	need security level check

Nothing is done on the other areas.

The first not erased address is returned if the blank check is failed.

The blank checking command on the Flash memory can be done from address 0000h to 1FFFFh.

The blank check operation is only possible if the HSB (Hardware Security Byte) has a security level lower than or equal to '2' (FLB2.0 = '110')

#### Table 41. Flash Blank check Authorization Summary

	Security level (HSB)				
		FLB2:0			
Command	111	110	101	011	
Blank Check	Allowed	Allowed	Forbidden	Forbidden	





#### **Reading commands** The reading command is supported by the following areas:

#### Table 42. Memory space & Select page

Memory/Information Family	Comments/Restriction
FLASH	need security level check
SECURITY	
CONFIGURATION	
BOOTLOADER	
SIGNATURE	

FLASH

The reading command on the Flash memory can be done from address 000h to 1FFFFh. The read operation is only possible if the HSB (Hardware Security Byte) has a security level lower than or equal to '2' (FLB2.0 = '110')

#### Table 43. Flash Read Authorization Summary

	Security level (HSB)			
	FLB2:0			
Command	111	110	101	011
Read	Allowed	Allowed	Forbidden	Forbidden

CONFIGURATION The CONFIGURATION family can always be read.

SECURITY The SECURITY family can always be read.

BOOTLOADER All the field from the BOOTLOARED family can be read from the bootloader. Each bootloader information shall be read unitary. Accesses must be done byte per byte according to the address definition

SIGNATURE All the field from the SIGNATURE family can be read from the bootloader. Each signature information shall be read unitary. Accesses must be done byte per byte according to the address definition

#### **Start Application**

The start application is performed by a watchdog reset.

The best way to start the application from a user defined entry point is to configure the FCB (Fuse Configuration Byte) before launching the watchdog. Then, depending on the configuration of the BRV2:0 field, the hardware boots from the selected memory area.

The start application command is used to quit the bootloader and start the application loaded.

## ISP Command summary

#### UART Protocol frames

#### Table 44. Summary of frames from Host

Command	Record Type	Record Length	Offset	Data[0]	Data[1]	Data[2]	Data[3]	Data[4]
Program selected memory	00h	nb of data	start address	х	х	х	х	х
Start application with watchdog	01h	00h	0000h	х	х	х	х	х
Select New Page	02h	02h	start address	Page (4 bits) + 0h	00h	х	х	х
Select Memory		02h	0000h	Memory space	Page	х	х	х
Read selected memory	04h 05			Start A	ddress	End	ddress	00h
Blank Check selected memory		05h	05h 0000h	0000h	Start P	luuress	Ella A	uuless
Erase Selected memory				00h	FFh	00h	00h	02h

#### Address Mapping

#### Table 45. Memory Families, Addresses & Coding

Memory/Parameter	coding	Address	Page number	Memory/Information Family
FLASH	0	0 up to 0x1FFFF	0 up to 1	FLASH
HSB	7	0	0	SECURITY
FCB	8	0	0	CONFIGURATION
Bootloader revision		00h		
Boot id1	3	01h	0	BOOTLOADER
Boot id2		02h		
Manuf. code		30h		
Family code		31h	0	
Product name	6	60h	0	SIGNATURE
Product rev		61h		

Attempting an access with any other 'coding', 'page number' or 'Address' results in no action and no answer from the bootloader.

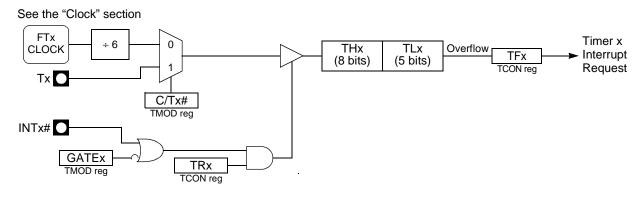




Timers/Counters	The AT89C51RE2 implements two general-purpose, 16-bit Timers/Counters. Such are identified as Timer 0 and Timer 1, and can be independently configured to operate in a variety of modes as a Timer or an event Counter. When operating as a Timer, the Timer/Counter runs for a pro- grammed length of time, then issues an interrupt request. When operating as a Counter, the Timer/Counter counts negative transitions on an external pin. After a preset number of counts, the Counter issues an interrupt request. The various operating modes of each Timer/Counter are described in the following sections.
Timer/Counter Operations	A basic operation is Timer registers THx and TLx ( $x = 0, 1$ ) connected in cascade to form a 16- bit Timer. Setting the run control bit (TRx) in TCON register (see Figure 46) turns the Timer on by allowing the selected input to increment TLx. When TLx overflows it increments THx; when THx overflows it sets the Timer overflow flag (TFx) in TCON register. Setting the TRx does not clear the THx and TLx Timer registers. Timer registers can be accessed to obtain the current count or to enter preset values. They can be read at any time but TRx bit must be cleared to pre- set their values, otherwise the behavior of the Timer/Counter is unpredictable.
	The C/Tx# control bit selects Timer operation or Counter operation by selecting the divided- down peripheral clock or external pin Tx as the source for the counted signal. TRx bit must be cleared when changing the mode of operation, otherwise the behavior of the Timer/Counter is unpredictable.
	For Timer operation (C/Tx# = 0), the Timer register counts the divided-down peripheral clock. The Timer register is incremented once every peripheral cycle (6 peripheral clock periods). The Timer clock rate is $F_{PER}/6$ , i.e. $F_{OSC}/12$ in standard mode or $F_{OSC}/6$ in X2 mode.
	For Counter operation (C/Tx# = 1), the Timer register counts the negative transitions on the Tx external input pin. The external input is sampled every peripheral cycles. When the sample is high in one cycle and low in the next one, the Counter is incremented. Since it takes 2 cycles (12 peripheral clock periods) to recognize a negative transition, the maximum count rate is $F_{PER}/12$ , i.e. $F_{OSC}/24$ in standard mode or $F_{OSC}/12$ in X2 mode. There are no restrictions on the duty cycle of the external input signal, but to ensure that a given level is sampled at least once before it changes, it should be held for at least one full peripheral cycle.
Timer 0	Timer 0 functions as either a Timer or event Counter in four modes of operation. Figure 23 to Figure 26 show the logical configuration of each mode.
	Timer 0 is controlled by the four lower bits of TMOD register (see Figure 47) and bits 0, 1, 4 and 5 of TCON register (see Figure 46). TMOD register selects the method of Timer gating (GATE0), Timer or Counter operation (T/C0#) and mode of operation (M10 and M00). TCON register provides Timer 0 control functions: overflow flag (TF0), run control bit (TR0), interrupt flag (IE0) and interrupt type control bit (IT0).
	For normal Timer operation (GATE0 = 0), setting TR0 allows TL0 to be incremented by the selected input. Setting GATE0 and TR0 allows external pin INT0# to control Timer operation.
	Timer 0 overflow (count rolls over from all 1s to all 0s) sets TF0 flag generating an interrupt request.
	It is important to stop Timer/Counter before changing mode.

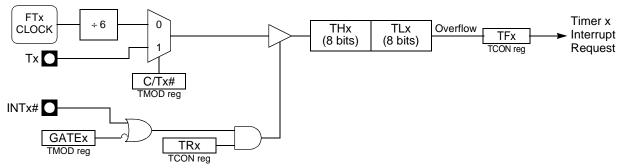
**Mode 0 (13-bit Timer)** Mode 0 configures Timer 0 as an 13-bit Timer which is set up as an 8-bit Timer (TH0 register) with a modulo 32 prescaler implemented with the lower five bits of TL0 register (see Figure 23). The upper three bits of TL0 register are indeterminate and should be ignored. Prescaler overflow increments TH0 register.

**Figure 23.** Timer/Counter x (x = 0 or 1) in Mode 0



**Mode 1 (16-bit Timer)** Mode 1 configures Timer 0 as a 16-bit Timer with TH0 and TL0 registers connected in cascade (see Figure 24). The selected input increments TL0 register.

Figure 24. Timer/Counter x (x = 0 or 1) in Mode 1 See the "Clock" section

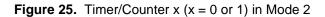


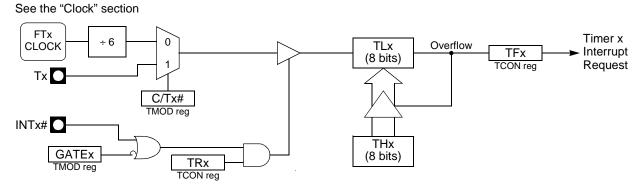




# Mode 2 (8-bit Timer with Auto-Reload)

Mode 2 configures Timer 0 as an 8-bit Timer (TL0 register) that automatically reloads from TH0 register (see Figure 25). TL0 overflow sets TF0 flag in TCON register and reloads TL0 with the contents of TH0, which is preset by software. When the interrupt request is serviced, hardware clears TF0. The reload leaves TH0 unchanged. The next reload value may be changed at any time by writing it to TH0 register.

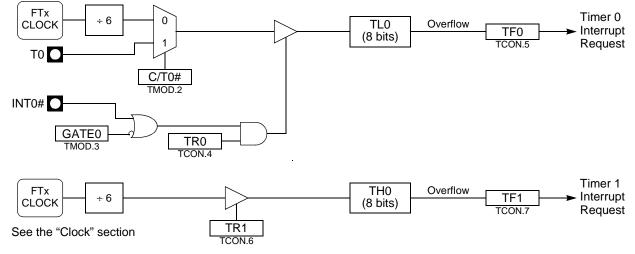




## Mode 3 (Two 8-bit Mode 3 c Timers) (see Figu

Mode 3 configures Timer 0 such that registers TL0 and TH0 operate as separate 8-bit Timers (see Figure 26). This mode is provided for applications requiring an additional 8-bit Timer or Counter. TL0 uses the Timer 0 control bits C/T0# and GATE0 in TMOD register, and TR0 and TF0 in TCON register in the normal manner. TH0 is locked into a Timer function (counting  $F_{PER}$  /6) and takes over use of the Timer 1 interrupt (TF1) and run control (TR1) bits. Thus, operation of Timer 1 is restricted when Timer 0 is in mode 3.

Figure 26. Timer/Counter 0 in Mode 3: Two 8-bit Counters

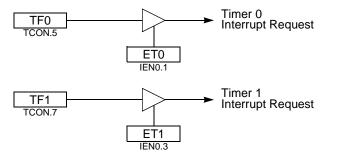


Timer 1	Timer 1 is identical to Timer 0 excepted for Mode 3 which is a hold-count mode. The following comments help to understand the differences:
	• Timer 1 functions as either a Timer or event Counter in three modes of operation. Figure 23 to Figure 25 show the logical configuration for modes 0, 1, and 2. Timer 1's mode 3 is a hold-count mode.
	• Timer 1 is controlled by the four high-order bits of TMOD register (see Figure 47) and bits 2, 3, 6 and 7 of TCON register (see Figure 46). TMOD register selects the method of Timer gating (GATE1), Timer or Counter operation (C/T1#) and mode of operation (M11 and M01). TCON register provides Timer 1 control functions: overflow flag (TF1), run control bit (TR1), interrupt flag (IE1) and interrupt type control bit (IT1).
	• Timer 1 can serve as the Baud Rate Generator for the Serial Port. Mode 2 is best suited for this purpose.
	• For normal Timer operation (GATE1 = 0), setting TR1 allows TL1 to be incremented by the selected input. Setting GATE1 and TR1 allows external pin INT1# to control Timer operation.
	<ul> <li>Timer 1 overflow (count rolls over from all 1s to all 0s) sets the TF1 flag generating an interrupt request.</li> </ul>
	• When Timer 0 is in mode 3, it uses Timer 1's overflow flag (TF1) and run control bit (TR1). For this situation, use Timer 1 only for applications that do not require an interrupt (such as a Baud Rate Generator for the Serial Port) and switch Timer 1 in and out of mode 3 to turn it off and on.
	It is important to stop Timer/Counter before changing mode.
Mode 0 (13-bit Timer)	Mode 0 configures Timer 1 as a 13-bit Timer, which is set up as an 8-bit Timer (TH1 register) with a modulo-32 prescaler implemented with the lower 5 bits of the TL1 register (see Figure 23). The upper 3 bits of TL1 register are ignored. Prescaler overflow increments TH1 register.
Mode 1 (16-bit Timer)	Mode 1 configures Timer 1 as a 16-bit Timer with TH1 and TL1 registers connected in cascade (see Figure 24). The selected input increments TL1 register.
Mode 2 (8-bit Timer with Auto-Reload)	Mode 2 configures Timer 1 as an 8-bit Timer (TL1 register) with automatic reload from TH1 reg- ister on overflow (see Figure 25). TL1 overflow sets TF1 flag in TCON register and reloads TL1 with the contents of TH1, which is preset by software. The reload leaves TH1 unchanged.
Mode 3 (Halt)	Placing Timer 1 in mode 3 causes it to halt and hold its count. This can be used to halt Timer 1 when TR1 run control bit is not available i.e. when Timer 0 is in mode 3.
Interrupt	Each Timer handles one interrupt source that is the timer overflow flag TF0 or TF1. This flag is set every time an overflow occurs. Flags are cleared when vectoring to the Timer interrupt routine. Interrupts are enabled by setting $ETx$ bit in IEN0 register. This assumes interrupts are globally enabled by setting EA bit in IEN0 register.





### Figure 27. Timer Interrupt System



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## Registers

Table 46. TCON Register

TCON (S:88h) Timer/Counter Control Register

7	6	5	4	3	2	1	0				
TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0				
Bit Number	Bit Mnemonic	Description	escription								
7	TF1	Cleared by ha	Timer 1 Overflow Flag Cleared by hardware when processor vectors to interrupt routine. Set by hardware on Timer/Counter overflow, when Timer 1 register overflows.								
6	TR1	Clear to turn o	Timer 1 Run Control Bit Clear to turn off Timer/Counter 1. Set to turn on Timer/Counter 1.								
5	TF0	Cleared by ha	Timer 0 Overflow Flag Cleared by hardware when processor vectors to interrupt routine. Set by hardware on Timer/Counter overflow, when Timer 0 register overflows.								
4	TR0		Control Bit ff Timer/Count Timer/Counter								
3	IE1	-	rdware when i	nterrupt is proce	0	<b>UU</b>	1).				
2	IT1		t low level activ	<b>t</b> ve (level triggere ive (edge trigge			ī1#).				
1	IE0	Cleared by ha	Interrupt 0 Edge Flag Cleared by hardware when interrupt is processed if edge-triggered (see IT0). Set by hardware when external interrupt is detected on INT0# pin.								
0	ITO	Clear to select		<b>t</b> ve (level triggere ive (edge trigge	,	• •	<sup></sup> 0#).				





## Table 47. TMOD Register

TMOD (S:89h) Timer/Counter Mode Control Register

7	6	5	4	3	2	1	0					
GATE1	C/T1#	M11	M01	GATE0	С/Т0#	M10	M00					
Bit Number	Bit Mnemonic	Description	escription									
7	GATE1		e Timer 1 whe	never TR1 bit is hile INT1# pin is		bit is set.						
6	C/T1#	Clear for Time	imer 1 Counter/Timer Select Bit lear for Timer operation: Timer 1 counts the divided-down system clock. et for Counter operation: Timer 1 counts negative transitions on external pin T1.									
5	M11		imer 1 Mode Select Bits									
4	M01	0 0 1 1	01Mode 1: 16-bit Timer/Counter.10Mode 2: 8-bit auto-reload Timer/Counter (TL1) <sup>(1)</sup>									
3	GATE0	Clear to enabl		never TR0 bit is 0 only while IN		and TR0 bit is s	set.					
2	C/T0#	Clear for Time		ect Bit mer 0 counts the mer 0 counts ne			pin T0.					
1	M10	<b>Timer 0 Mode</b> <u>M10</u> 0	0 0 Mode 0: 8-bit Timer/Counter (TH0) with 5-bit prescaler (TL0).									
0	M00	01Mode 1: 16-bit Timer/Counter.10Mode 2: 8-bit auto-reload Timer/Counter (TL0) <sup>(2)</sup> 11Mode 3: TL0 is an 8-bit Timer/CounterTH0 is an 8-bit Timer using Timer 1's TR0 and TF0 bits.										
		om TH1 at ov	verflow.									

2. Reloaded from TH0 at overflow.

 Table 48.
 TH0 Register

TH0 (S:8Ch) Timer 0 High Byte Register

7	6	5	4	3	2	1	0			
-	-	-	-	-	-	-	-			
Bit Number	Bit Mnemonic	Description	Description							
7:0		High Byte of	High Byte of Timer 0.							

Reset Value = 0000 0000b

#### Table 49. TL0 Register

TL0 (S:8Ah) Timer 0 Low Byte Register

7	6	5	4	3	2	1	0			
_	-									
Bit Number	Bit Mnemonic	Description	Description							
7:0		Low Byte of Timer 0.								

Reset Value = 0000 0000b

#### Table 50. TH1 Register

TH1 (S:8Dh) Timer 1 High Byte Register

7	6	5	4	3	2	1	0			
-	-									
Bit Number	Bit Mnemonic	Description	Description							
7:0		High Byte of	Timer 1.							





## Table 51. TL1 Register

TL1 (S:8Bh) Timer 1 Low Byte Register

7	6	5	4	3	2	1	0		
-	-	-	-	-	-	-	-		
Bit Number	Bit Mnemonic	Description	Description						
7:0		Low Byte of	Timer 1.						

Reset Value = 0000 0000b

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## Timer 2

The Timer 2 in the AT89C51RE2 is the standard C52 Timer 2.

It is a 16-bit timer/counter: the count is maintained by two eight-bit timer registers, TH2 and TL2 are cascaded. It is controlled by T2CON (Table 52) and T2MOD (Table 53) registers. Timer 2 operation is similar to Timer 0 and Timer 1.C/T2 selects  $F_{OSC}/12$  (timer operation) or external pin T2 (counter operation) as the timer clock input. Setting TR2 allows TL2 to increment by the selected input.

Timer 2 has 3 operating modes: capture, autoreload and Baud Rate Generator. These modes are selected by the combination of RCLK, TCLK and CP/RL2 (T2CON).

Refer to the Atmel 8-bit Microcontroller Hardware description for the description of Capture and Baud Rate Generator Modes.

Timer 2 includes the following enhancements:

- Auto-reload mode with up or down counter
- Programmable clock-output

Auto-Reload Mode The auto-reload mode configures Timer 2 as a 16-bit timer or event counter with automatic reload. If DCEN bit in T2MOD is cleared, Timer 2 behaves as in 80C52 (refer to the Atmel C51 Microcontroller Hardware description). If DCEN bit is set, Timer 2 acts as an Up/down timer/counter as shown in Figure 28. In this mode the T2EX pin controls the direction of count.

When T2EX is high, Timer 2 counts up. Timer overflow occurs at FFFFh which sets the TF2 flag and generates an interrupt request. The overflow also causes the 16-bit value in RCAP2H and RCAP2L registers to be loaded into the timer registers TH2 and TL2.

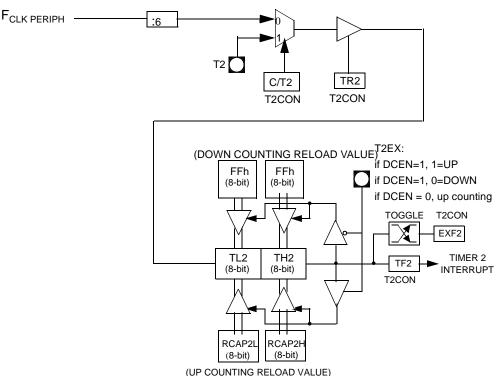
When T2EX is low, Timer 2 counts down. Timer underflow occurs when the count in the timer registers TH2 and TL2 equals the value stored in RCAP2H and RCAP2L registers. The underflow sets TF2 flag and reloads FFFFh into the timer registers.

The EXF2 bit toggles when Timer 2 overflows or underflows according to the direction of the count. EXF2 does not generate any interrupt. This bit can be used to provide 17-bit resolution.





Figure 28. Auto-Reload Mode Up/Down Counter (DCEN = 1)



### Programmable Clock-Output

In the clock-out mode, Timer 2 operates as a 50%-duty-cycle, programmable clock generator (See Figure 29). The input clock increments TL2 at frequency F<sub>CLK PERIPH</sub>/2.The timer repeatedly counts to overflow from a loaded value. At overflow, the contents of RCAP2H and RCAP2L registers are loaded into TH2 and TL2.In this mode, Timer 2 overflows do not generate interrupts. The formula gives the clock-out frequency as a function of the system oscillator frequency and the value in the RCAP2H and RCAP2L registers:

$$Clock-OutFrequency = \frac{F_{CLKPERIPH}}{4 \times (65536 - RCAP2H/RCAP2L)}$$

For a 16 MHz system clock, Timer 2 has a programmable frequency range of 61 Hz  $(F_{CLK PERIPH}/2^{16})$  to 4 MHz  $(F_{CLK PERIPH}/4)$ . The generated clock signal is brought out to T2 pin (P1.0).

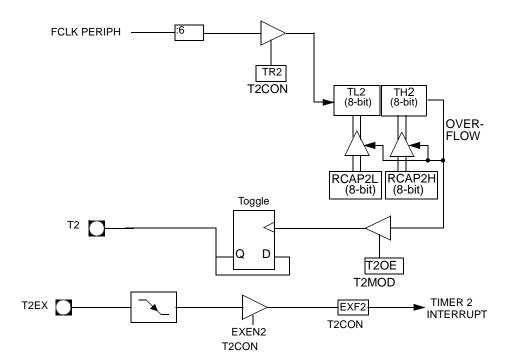
Timer 2 is programmed for the clock-out mode as follows:

- Set T2OE bit in T2MOD register.
- Clear C/T2 bit in T2CON register.
- Determine the 16-bit reload value from the formula and enter it in RCAP2H/RCAP2L registers.
- Enter a 16-bit initial value in timer registers TH2/TL2.It can be the same as the reload value or a different one depending on the application.
- To start the timer, set TR2 run control bit in T2CON register.

It is possible to use Timer 2 as a baud rate generator and a clock generator simultaneously. For this configuration, the baud rates and clock frequencies are not independent since both functions use the values in the RCAP2H and RCAP2L registers.

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**Figure 29.** Clock-Out Mode  $C/\overline{T2} = 0$ 







## Registers

## Table 52. T2CON Register

T2CON - Timer 2 Control Register (C8h)

7	6	5	4	3	2	1	0				
TF2	EXF2	RCLK	TCLK	EXEN2	TR2	C/T2#	CP/RL2#				
Bit Number	Bit Mnemonic	Description	escription								
7	TF2	Must b	Fimer 2 overflow Flag Must be cleared by software. Set by hardware on Timer 2 overflow, if RCLK = 0 and TCLK = 0.								
6	EXF2	Set when a ca EXEN2=1. When set, cau is enabled.	When set, causes the CPU to vector to Timer 2 interrupt routine when Timer 2 interrupt s enabled. Aust be cleared by software. EXF2 doesn't cause an interrupt in Up/down counter mode								
5	RCLK	Cleared to use	Receive Clock bit Cleared to use timer 1 overflow as receive clock for serial port in mode 1 or 3. Set to use Timer 2 overflow as receive clock for serial port in mode 1 or 3.								
4	TCLK		e timer 1 overfl	ow as transmit as transmit cloc		•					
3	EXEN2	Cleared to ign Set to cause a		T2EX pin for Tir oad when a neg	•		s detected, if				
2	TR2	Timer 2 Run Cleared to tur Set to turn on	n off Timer 2.								
1	C/T2#	Cleared for tin Set for counte	r operation (inp	input from interr out from T2 inpu							
0	CP/RL2#	If RCLK=1 or overflow. Cleared to aut EXEN2=1.	Cleared to auto-reload on Timer 2 overflows or negative transitions on T2EX pin if								

Reset Value = 0000 0000b Bit addressable

## Table 53. T2MOD Register

T2MOD - Timer 2 Mode Control Register (C9h)

7	6	5	4	3	2	1	0					
-	-	-	-	-	-	T2OE	DCEN					
Bit Number	Bit Mnemonic	Description	Description									
7	-	<b>Reserved</b> The value rea	Reserved The value read from this bit is indeterminate. Do not set this bit.									
6	-	<b>Reserved</b> The value rea	Reserved The value read from this bit is indeterminate. Do not set this bit.									
5	-	Reserved The value rea	Reserved The value read from this bit is indeterminate. Do not set this bit.									
4	-	<b>Reserved</b> The value rea	d from this bit i	s indeterminate	. Do not set this	s bit.						
3	-	Reserved The value rea	d from this bit i	s indeterminate	. Do not set this	s bit.						
2	-	Reserved The value rea	d from this bit i	s indeterminate	. Do not set this	s bit.						
1	T2OE	Cleared to pro	<b>Timer 2 Output Enable bit</b> Cleared to program P1.0/T2 as clock input or I/O port. Set to program P1.0/T2 as clock output.									
0	DCEN	Cleared to dis	<b>Down Counter Enable bit</b> Cleared to disable Timer 2 as up/down counter. Set to enable Timer 2 as up/down counter.									

Reset Value = XXXX XX00b Not bit addressable





# Programmable Counter Array PCA

The PCA provides more timing capabilities with less CPU intervention than the standard timer/counters. Its advantages include reduced software overhead and improved accuracy. The PCA consists of a dedicated timer/counter which serves as the time base for an array of five compare/capture modules. Its clock input can be programmed to count any one of the following signals:

- Peripheral clock frequency (F<sub>CLK PERIPH</sub>) ÷ 6
- Peripheral clock frequency  $(F_{CLK PERIPH}) \div 2$
- Timer 0 overflow
- External input on ECI (P1.2)

Each compare/capture modules can be programmed in any one of the following modes:

- Rising and/or falling edge capture
- Software timer
- High-speed output
- Pulse width modulator

Module 4 can also be programmed as a watchdog timer (See Section "PCA Watchdog Timer", page 93).

When the compare/capture modules are programmed in the capture mode, software timer, or high speed output mode, an interrupt can be generated when the module executes its function. All five modules plus the PCA timer overflow share one interrupt vector.

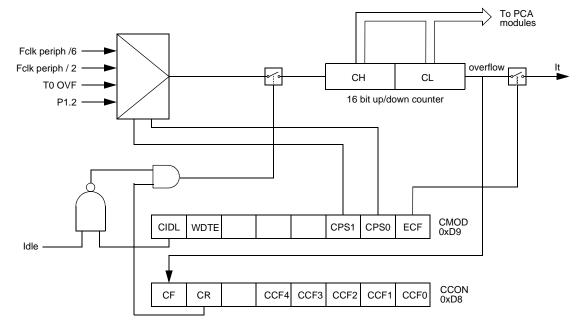
The PCA timer/counter and compare/capture modules share Port 1 for external I/O. These pins are listed below. If the port is not used for the PCA, it can still be used for standard I/O.

PCA component	External I/O Pin				
16-bit Counter	P1.2 / ECI				
16-bit Module 0	P1.3 / CEX0				
16-bit Module 1	P1.4 / CEX1				
16-bit Module 2	P1.5 / CEX2				
16-bit Module 3	P1.6 / CEX3				

The PCA timer is a common time base for all five modules (See Figure 30). The timer count source is determined from the CPS1 and CPS0 bits in the CMOD register (Table 54) and can be programmed to run at:

- 1/6 the peripheral clock frequency (F<sub>CLK PERIPH</sub>)
- 1/2 the peripheral clock frequency (F<sub>CLK PERIPH</sub>)
- The Timer 0 overflow
- The input on the ECI pin (P1.2)

Figure 30. PCA Timer/Counter







#### Table 54. CMOD Register

CMOD - PCA Counter Mode Register (D9h)

7	6	5	4	3	2	1	0					
CIDL	WDTE	-	-	-	CPS1	CPS0	ECF					
Bit Number	Bit Mnemonic	Description	Description									
7	CIDL	Cleared to pro	Counter Idle Control Cleared to program the PCA Counter to continue functioning during idle Mode. Set to program PCA to be gated off during idle.									
6	WDTE	Cleared to dis	<b>Vatchdog Timer Enable</b> Cleared to disable Watchdog Timer function on PCA Module 4. Set to enable Watchdog Timer function on PCA Module 4.									
5	-	<b>Reserved</b> The value rea	Reserved The value read from this bit is indeterminate. Do not set this bit.									
4	-	<b>Reserved</b> The value rea	d from this bit i	s indeterminate	. Do not set this	s bit.						
3	-	<b>Reserved</b> The value rea	d from this bit i	s indeterminate	. Do not set this	s bit.						
2	CPS1	PCA Count P	ulse Select									
1	CPS0	CPS1         CPS0           0         0           1         0           1         1	0     1     Internal clock fCLK PERIPH/2       1     0     Timer 0 Overflow									
0	ECF	PCA Enable Counter Overflow Interrupt Cleared to disable CF bit in CCON to inhibit an interrupt. Set to enable CF bit in CCON to generate an interrupt.										

Reset Value = 00XX X000b Not bit addressable

The CMOD register includes three additional bits associated with the PCA (See Figure 30 and Table 54).

- The CIDL bit which allows the PCA to stop during idle mode.
- The WDTE bit which enables or disables the watchdog function on module 4.
- The ECF bit which when set causes an interrupt and the PCA overflow flag CF (in the CCON SFR) to be set when the PCA timer overflows.

The CCON register contains the run control bit for the PCA and the flags for the PCA timer (CF) and each module (Refer to Table 55).

- Bit CR (CCON.6) must be set by software to run the PCA. The PCA is shut off by clearing this bit.
- Bit CF: The CF bit (CCON.7) is set when the PCA counter overflows and an interrupt will be generated if the ECF bit in the CMOD register is set. The CF bit can only be cleared by software.
- Bits 0 through 4 are the flags for the modules (bit 0 for module 0, bit 1 for module 1, etc.) and are set by hardware when either a match or a capture occurs. These flags also can only be cleared by software.

## Table 55. CCON Register

CCON - PCA Counter Control Register (D8h)

7	6	5	4	3	2	1	0					
CF	CR	-	CCF4	CCF3	CCF2	CCF1	CCF0					
Bit Number	Bit Mnemonic	Description	Description									
7	CF	Set by hardwa set. CF	<b>PCA Counter Overflow flag</b> Set by hardware when the counter rolls over. CF flags an interrupt if bit ECF in CMOD is set. CF may be set by either hardware or software but can only be cleared by software.									
6	CR	Must be clear	PCA Counter Run control bit Must be cleared by software to turn the PCA counter off. Set by software to turn the PCA counter on.									
5	-	Reserved The value rea	Reserved The value read from this bit is indeterminate. Do not set this bit.									
4	CCF4	Must be clear	<b>4 interrupt fla</b> ed by software are when a mat	•	ccurs.							
3	CCF3	Must be clear	<b>3 interrupt fla</b> ed by software are when a ma	•	ccurs.							
2	CCF2	Must be clear	<b>2 interrupt fla</b> ed by software are when a ma	•	ccurs.							
1	CCF1	Must be clear	PCA Module 1 interrupt flag Must be cleared by software. Set by hardware when a match or capture occurs.									
0	CCF0	Must be clear	<b>0 interrupt fla</b> ed by software are when a mat	-	ccurs.							

Reset Value = 00X0 0000b Not bit addressable

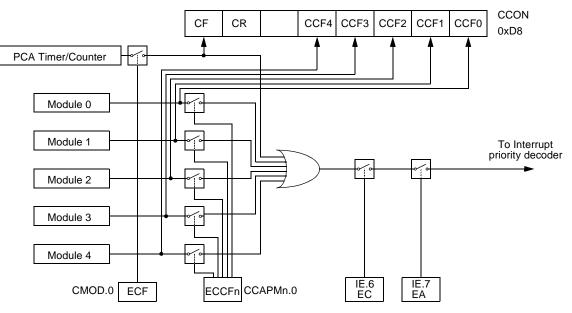
The watchdog timer function is implemented in module 4 (See Figure 33).

The PCA interrupt system is shown in Figure 31.





#### Figure 31. PCA Interrupt System



**PCA Modules:** each one of the five compare/capture modules has six possible functions. It can perform:

- 16-bit Capture, positive-edge triggered
- 16-bit Capture, negative-edge triggered
- 16-bit Capture, both positive and negative-edge triggered
- 16-bit Software Timer
- 16-bit High Speed Output
- 8-bit Pulse Width Modulator

In addition, module 4 can be used as a Watchdog Timer.

Each module in the PCA has a special function register associated with it. These registers are: CCAPM0 for module 0, CCAPM1 for module 1, etc. (See Table 56). The registers contain the bits that control the mode that each module will operate in.

- The ECCF bit (CCAPMn.0 where n=0, 1, 2, 3, or 4 depending on the module) enables the CCF flag in the CCON SFR to generate an interrupt when a match or compare occurs in the associated module.
- PWM (CCAPMn.1) enables the pulse width modulation mode.
- The TOG bit (CCAPMn.2) when set causes the CEX output associated with the module to toggle when there is a match between the PCA counter and the module's capture/compare register.
- The match bit MAT (CCAPMn.3) when set will cause the CCFn bit in the CCON register to be set when there is a match between the PCA counter and the module's capture/compare register.
- The next two bits CAPN (CCAPMn.4) and CAPP (CCAPMn.5) determine the edge that a
  capture input will be active on. The CAPN bit enables the negative edge, and the CAPP bit
  enables the positive edge. If both bits are set both edges will be enabled and a capture will
  occur for either transition.
- The last bit in the register ECOM (CCAPMn.6) when set enables the comparator function.

Table 56 shows the CCAPMn settings for the various PCA functions.

**Table 56.** CCAPMn Registers (n = 0-4)

CCAPM0 - PCA Module 0 Compare/Capture Control Register (0DAh)

CCAPM1 - PCA Module 1 Compare/Capture Control Register (0DBh)

CCAPM2 - PCA Module 2 Compare/Capture Control Register (0DCh)

CCAPM3 - PCA Module 3 Compare/Capture Control Register (0DDh)

CCAPM4 - PCA Module 4 Compare/Capture Control Register (0DEh)

7	6	5	4	3	2	1	0					
-	ECOMn	CAPPn	CAPNn	MATn	TOGn	PWMn	ECCFn					
Bit Number	Bit Mnemonic	Description	escription									
7	-	<b>Reserved</b> The value rea	d from this bit i	is indeterminate	. Do not set thi	s bit.						
6	ECOMn	Cleared to dis	able Comparator eared to disable the comparator function. t to enable the comparator function.									
5	CAPPn	Cleared to dis	apture Positive eared to disable positive edge capture. et to enable positive edge capture.									
4	CAPNn	Cleared to dis	Capture Negative Cleared to disable negative edge capture. Set to enable negative edge capture.									
3	MATn	register cause	es the	the PCA counte		lule's compare/	capture					
2	TOGn	Toggle When TOGn = register cause CEXn pin to to	es the	the PCA count	er with this mod	dule's compare,	/capture					
1	PWMn	Cleared to dis	Pulse Width Modulation Mode Cleared to disable the CEXn pin to be used as a pulse width modulated output. Set to enable the CEXn pin to be used as a pulse width modulated output.									
0	CCF0	interrupt.	able compare/	capture flag CC ure flag CCFn in		0 0						

Reset Value = X000 0000b Not bit addressable





ECOMn	CAPPn	CAPNn	MATn	TOGn	PWMm	ECCFn	Module Function
0	0	0	0	0	0	0	No Operation
х	1	0	0	0	0	х	16-bit capture by a positive-edge trigger on CEXn
х	0	1	0	0	0	х	16-bit capture by a negative trigger on CEXn
х	1	1	0	0	0	Х	16-bit capture by a transition on CEXn
1	0	0	1	0	0	х	16-bit Software Timer / Compare mode.
1	0	0	1	1	0	X 16-bit High Speed Output	
1	0	0	0	0	1	0 8-bit PWM	
1	0	0	1	Х	0	Х	Watchdog Timer (module 4 only)

Table 57. PC/	A Module Modes	s (CCAPMn Register	s)
---------------	----------------	--------------------	----

There are two additional registers associated with each of the PCA modules. They are CCAPnH and CCAPnL and these are the registers that store the 16-bit count when a capture occurs or a compare should occur. When a module is used in the PWM mode these registers are used to control the duty cycle of the output (See Table 58 & Table 59).

Table 58. CCAPnH Registers (n = 0-4)

CCAP0H - PCA Module 0 Compare/Capture Control Register High (0FAh)

CCAP1H - PCA Module 1 Compare/Capture Control Register High (0FBh)

CCAP2H - PCA Module 2 Compare/Capture Control Register High (0FCh)

CCAP3H - PCA Module 3 Compare/Capture Control Register High (0FDh)

CCAP4H - PCA Module 4 Compare/Capture Control Register High (0FEh)

7	6	5	4	3	2	1	0		
-	-	-	-	-	-	-	-		
Bit Number	Bit Mnemonic	Description							
7-0	-	PCA Module n Compare/Capture Control CCAPnH Value							

Reset Value = 0000 0000b Not bit addressable

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 Table 59.
 CCAPnL Registers (n = 0-4)

CCAP0L - PCA Module 0 Compare/Capture Control Register Low (0EAh)

CCAP1L - PCA Module 1 Compare/Capture Control Register Low (0EBh)

CCAP2L - PCA Module 2 Compare/Capture Control Register Low (0ECh)

CCAP3L - PCA Module 3 Compare/Capture Control Register Low (0EDh)

CCAP4L - PCA Module 4 Compare/Capture Control Register Low (0EEh)

7	6	5	4	3	2	1	0	
-	-	-	-	-	-	-	-	
Bit Number	Bit Mnemonic	Description						
7-0	-	PCA Module n Compare/Capture Control CCAPnL Value						

Reset Value = 0000 0000b Not bit addressable

#### Table 60. CH Register

CH - PCA Counter Register High (0F9h)

7	6	5	4	3	2	1	0
-	-	-	-	-	-	-	-
Bit Number	Bit Mnemonic	Description					
7-0	-	PCA counter CH Val					

Reset Value = 0000 0000b Not bit addressable

#### Table 61. CL Register

CL - PCA Counter Register Low (0E9h)

7	6	5	4	3	2	1	0	
-	-	-	-	-	-	-	-	
Bit Number	Bit Mnemonic	Description						
7-0	-	PCA Counter CL Valu						

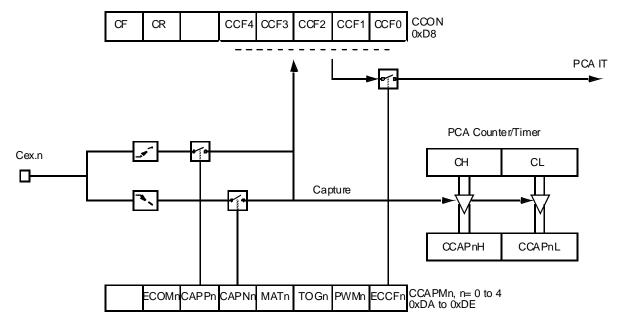
Reset Value = 0000 0000b Not bit addressable





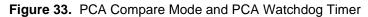
**PCA Capture Mode** To use one of the PCA modules in the capture mode either one or both of the CCAPM bits CAPN and CAPP for that module must be set. The external CEX input for the module (on port 1) is sampled for a transition. When a valid transition occurs the PCA hardware loads the value of the PCA counter registers (CH and CL) into the module's capture registers (CCAPnL and CCAPnH). If the CCFn bit for the module in the CCON SFR and the ECCFn bit in the CCAPMn SFR are set then an interrupt will be generated (Refer to Figure 32).

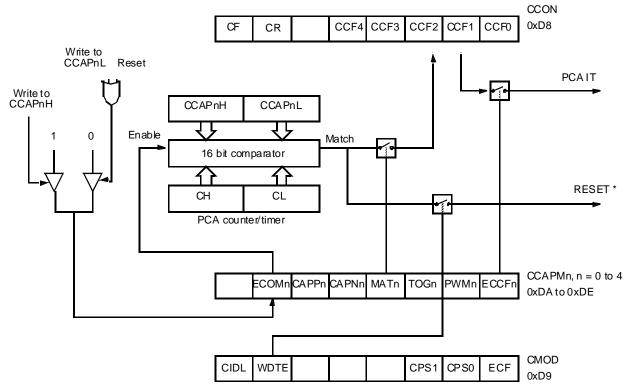
#### Figure 32. PCA Capture Mode



16-bit Software Timer/ Compare Mode

The PCA modules can be used as software timers by setting both the ECOM and MAT bits in the modules CCAPMn register. The PCA timer will be compared to the module's capture registers and when a match occurs an interrupt will occur if the CCFn (CCON SFR) and the ECCFn (CCAPMn SFR) bits for the module are both set (See Figure 33).





Before enabling ECOM bit, CCAPnL and CCAPnH should be set with a non zero value, otherwise an unwanted match could happen. Writing to CCAPnH will set the ECOM bit.

Once ECOM set, writing CCAPnL will clear ECOM so that an unwanted match doesn't occur while modifying the compare value. Writing to CCAPnH will set ECOM. For this reason, user software should write CCAPnL first, and then CCAPnH. Of course, the ECOM bit can still be controlled by accessing to CCAPMn register.

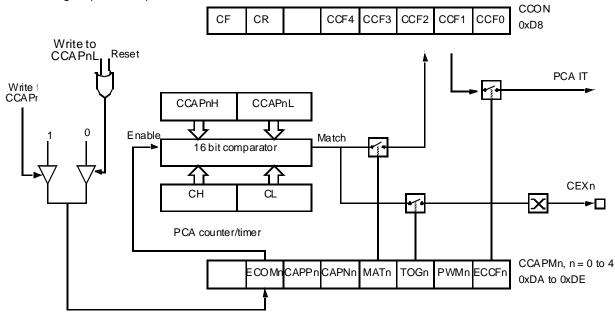
**High Speed Output** Mode In this mode the CEX output (on port 1) associated with the PCA module will toggle each time a match occurs between the PCA counter and the module's capture registers. To activate this mode the TOG, MAT, and ECOM bits in the module's CCAPMn SFR must be set (See Figure 34).

A prior write must be done to CCAPnL and CCAPnH before writing the ECOMn bit.





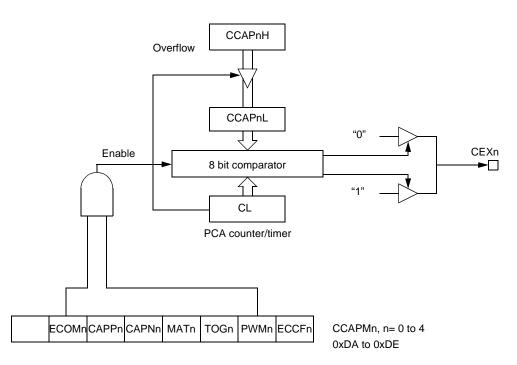
Figure 34. PCA High Speed Output Mode



Before enabling ECOM bit, CCAPnL and CCAPnH should be set with a non zero value, otherwise an unwanted match could happen.

Once ECOM set, writing CCAPnL will clear ECOM so that an unwanted match doesn't occur while modifying the compare value. Writing to CCAPnH will set ECOM. For this reason, user software should write CCAPnL first, and then CCAPnH. Of course, the ECOM bit can still be controlled by accessing to CCAPMn register.

**Pulse Width Modulator Mode** All of the PCA modules can be used as PWM outputs. Figure 35 shows the PWM function. The frequency of the output depends on the source for the PCA timer. All of the modules will have the same frequency of output because they all share the PCA timer. The duty cycle of each module is independently variable using the module's capture register CCAPLn. When the value of the PCA CL SFR is less than the value in the module's CCAPLn SFR the output will be low, when it is equal to or greater than the output will be high. When CL overflows from FF to 00, CCAPLn is reloaded with the value in CCAPHn. This allows updating the PWM without glitches. The PWM and ECOM bits in the module's CCAPMn register must be set to enable the PWM mode. Figure 35. PCA PWM Mode



## PCA Watchdog Timer

An on-board watchdog timer is available with the PCA to improve the reliability of the system without increasing chip count. Watchdog timers are useful for systems that are susceptible to noise, power glitches, or electrostatic discharge. Module 4 is the only PCA module that can be programmed as a watchdog. However, this module can still be used for other modes if the watchdog is not needed. Figure 33 shows a diagram of how the watchdog works. The user preloads a 16-bit value in the compare registers. Just like the other compare modes, this 16-bit value is compared to the PCA timer value. If a match is allowed to occur, an internal reset will be generated. This will not cause the RST pin to be driven high.

In order to hold off the reset, the user has three options:

- 1. periodically change the compare value so it will never match the PCA timer,
- 2. periodically change the PCA timer value so it will never match the compare values, or

3. disable the watchdog by clearing the WDTE bit before a match occurs and then re-enable it.

The first two options are more reliable because the watchdog timer is never disabled as in option #3. If the program counter ever goes astray, a match will eventually occur and cause an internal reset. The second option is also not recommended if other PCA modules are being used. Remember, the PCA timer is the time base for all modules; changing the time base for other modules would not be a good idea. Thus, in most applications the first solution is the best option.

This watchdog timer won't generate a reset out on the reset pin.





## Serial I/O Port

The serial I/O ports in the AT89C51RE2 are compatible with the serial I/O port in the 80C52. They provide both synchronous and asynchronous communication modes. They operates as a Universal Asynchronous Receiver and Transmitter (UART) in three full-duplex modes (Modes 1, 2 and 3). Asynchronous transmission and reception can occur simultaneously and at different baud rates

Both serial I/O port include the following enhancements:

- Framing error detection
- Automatic address recognition

As these improvements apply to both UART, most of the time in the following lines, there won't be any reference to UART\_0 or UART\_1, but only to UART, generally speaking.

Framing ErrorFraming bit error detection is provided for the three asynchronous modes (modes 1, 2 and 3). To<br/>enable the framing bit error detection feature, set SMOD0 bit in PCON register (See Figure 36)<br/>for UART 0 or set SMOD0\_1 in BDRCON\_1 register for UART 1 (See Figure 37).

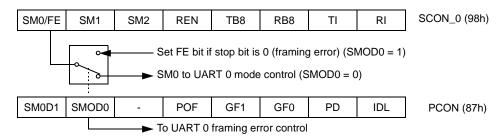
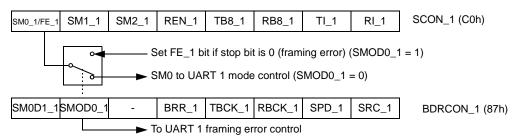


Figure 36. UART 0 Framing Error Block Diagram

#### Figure 37. UART 1 Framing Error Block Diagram



When this feature is enabled, the receiver checks each incoming data frame for a valid stop bit. An invalid stop bit may result from noise on the serial lines or from simultaneous transmission by two CPUs. If a valid stop bit is not found, the Framing Error bit (FE) in SCON register (See Table 68.) bit is set.

Software may examine FE bit after each reception to check for data errors. Once set, only software or a reset can clear FE bit. Subsequently received frames with valid stop bits cannot clear FE bit. When FE feature is enabled, RI rises on stop bit instead of the last data bit (See Figure 38 and Figure 39).



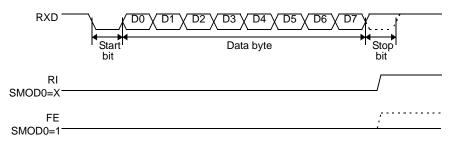
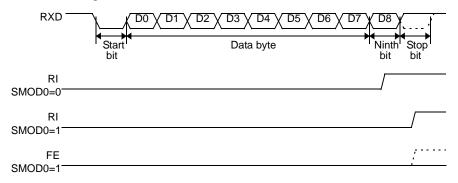


Figure 39. UART Timings in Modes 2 and 3



Automatic Address Recognition	<ul> <li>The automatic address recognition feature is enabled when the multiprocessor communication feature is enabled (SM2 bit in SCON register is set).</li> <li>Implemented in hardware, automatic address recognition enhances the multiprocessor communication feature by allowing the serial port to examine the address of each incoming command frame. Only when the serial port recognizes its own address, the receiver sets RI bit in SCON register to generate an interrupt. This ensures that the CPU is not interrupted by command frames addressed to other devices.</li> <li>If desired, the user may enable the automatic address recognition feature in mode 1.In this configuration, the stop bit takes the place of the ninth data bit. Bit RI is set only when the received command frame address matches the device's address and is terminated by a valid stop bit. To support automatic address recognition, a device is identified by a given address and a broadcast address.</li> </ul>
	Note: The multiprocessor communication and automatic address recognition features cannot be enabled in mode 0 (i. e. setting SM2 bit in SCON register in mode 0 has no effect).
Given Address	Each device has an individual address that is specified in SADDR register; the SADEN register is a mask byte that contains don't-care bits (defined by zeros) to form the device's given address. The don't-care bits provide the flexibility to address one or more slaves at a time. The following example illustrates how a given address is formed. To address a device by its individual address, the SADEN mask byte must be 1111 1111b. For example: SADDR0101 0110b SADEN1111 1100b Given0101 01XXb
	The following is an example of how to use given addresses to address different slaves:





Slave A:SADDR1111 0001b SADEN1111 1010b Given1111 0X0Xb

Slave B:SADDR1111 0011b SADEN1111 1001b Given1111 0XX1b

Slave C:SADDR1111 0010b <u>SADEN1111 1101b</u> Given1111 00X1b

The SADEN byte is selected so that each slave may be addressed separately.

For slave A, bit 0 (the LSB) is a don't-care bit; for slaves B and C, bit 0 is a 1.To communicate with slave A only, the master must send an address where bit 0 is clear (e.g., 1111 0000b). For slave A, bit 1 is a 1; for slaves B and C, bit 1 is a don't care bit. To communicate with slaves B and C, but not slave A, the master must send an address with bits 0 and 1 both set (e.g. 1111 0011b).

To communicate with slaves A, B and C, the master must send an address with bit 0 set, bit 1 clear, and bit 2 clear (e.g. 1111 0001b).

**Broadcast Address** A broadcast address is formed from the logical OR of the SADDR and SADEN registers with zeros defined as don't-care bits, e.g.:

SADDR0101 0110b SADEN1111 1100b Broadcast =SADDR OR SADEN1111 111Xb

The use of don't-care bits provides flexibility in defining the broadcast address, however in most applications, a broadcast address is FFh. The following is an example of using broadcast addresses:

Slave A:SADDR1111 0001b SADEN1111 1010b Broadcast1111 1X11b,

Slave B:SADDR1111 0011b SADEN1111 1001b Broadcast1111 1X11B,

Slave C:SADDR=1111 0011b <u>SADEN1111 1101b</u> Broadcast1111 1111b

For slaves A and B, bit 2 is a don't care bit; for slave C, bit 2 is set. To communicate with all of the slaves, the master must send an address FFh. To communicate with slaves A and B, but not slave C, the master can send and address FBh.

**Reset Addresses** On reset, the SADDR and SADEN registers are initialized to 00h, i. e. the given and broadcast addresses are XXXX XXXb (all don't-care bits). This ensures that the serial port will reply to any address, and so, that it is backwards compatible with the 80C51 microcontrollers that do not support automatic address recognition.

## Registers

#### Table 62. SADEN\_0 Register

SADEN - Slave Address Mask Register UART 0(B9h)

7	6	5	4	3	2	1	0

Reset Value = 0000 0000b Not bit addressable

#### Table 63. SADDR\_0 Register

#### SADDR - Slave Address Register UART 0(A9h)

7	6	5	4	3	2	1	0

Reset Value = 0000 0000b Not bit addressable

#### Table 64. SADEN\_1 Register

SADEN\_1 - Slave Address Mask Register UART 1(BAh)

7	6	5	4	3	2	1	0

Reset Value = 0000 0000b Not bit addressable

#### Table 65. SADDR\_1 Register

SADDR\_1 - Slave Address Register UART 1(AAh)

7	6	5	4	3	2	1	0

Reset Value = 0000 0000b Not bit addressable

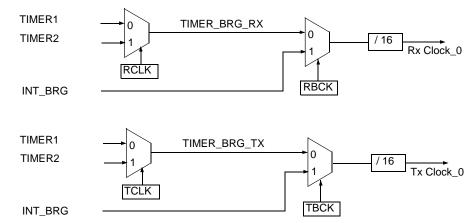




## Baud Rate Selection for UART 0 for Mode 1 and 3

The Baud Rate Generator for transmit and receive clocks can be selected separately via the T2CON and BDRCON\_0 registers.

Figure 40. Baud Rate Selection for UART 0



#### Table 66. Baud Rate Selection Table UART 0

TCLK (T2CON)	RCLK (T2CON)	TBCK (BDRCON)	RBCK (BDRCON)	Clock Source UART Tx	Clock Source UART Rx
0	0	0	0	Timer 1	Timer 1
1	0	0	0	Timer 2	Timer 1
0	1	0	0	Timer 1	Timer 2
1	1	0	0	Timer 2	Timer 2
х	0	1	0	INT_BRG	Timer 1
х	1	1	0	INT_BRG	Timer 2
0	Х	0	1	Timer 1	INT_BRG
1	Х	0	1	Timer 2	INT_BRG
Х	Х	1	1	INT_BRG	INT_BRG

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## Baud Rate Selection for UART 1 for Mode 1 and 3

The Baud Rate Generator for transmit and receive clocks can be selected separately via the T2CON and BDRCON\_1 registers.

Figure 41. Baud Rate Selection for UART 1

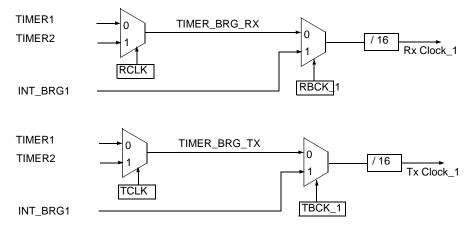


Table 67. Baud Rate Selection Table UART 1

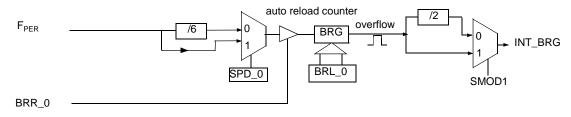
TCLK (T2CON)	RCLK (T2CON)	TBCK_1 (BDRCON_1)	RBCK_1 (BDRCON_1)	Clock Source UART Tx_1	Clock Source UART Rx_1
0	0	0	0	Timer 1	Timer 1
1	0	0	0	Timer 2	Timer 1
0	1	0	0	Timer 1	Timer 2
1	1	0	0	Timer 2	Timer 2
Х	0	1	0	INT_BRG_1	Timer 1
Х	1	1	0	INT_BRG_1	Timer 2
0	Х	0	1	Timer 1	INT_BRG_1
1	Х	0	1	Timer 2	INT_BRG_1
Х	Х	1	1	INT_BRG_1	INT_BRG_1



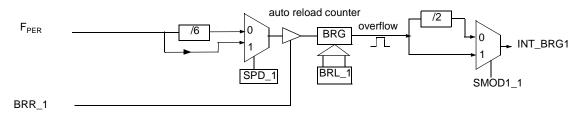


#### Internal Baud Rate Generator (BRG) The AT89C51RE2 implements two internal baudrate generators. Each one is dedicated to the corresponding UART. The configuration and operating mode for both BRG are similar. When an internal Baud Rate Generator is used, the Baud Rates are determined by the BRG overflow depending on the BRL (BRL or BRL\_1 registers) reload value, the value of SPD (or SPD\_1) bit (Speed Mode) in BDRCON (BDRCON\_1) register and the value of the SMOD1 bit in PCON register.

#### Figure 42. Internal Baud Rate generator 0



#### Figure 43. Internal Baud Rate generator 1



• The baud rate for UART is token by formula:

$$Baud_Rate = \frac{2^{SMOD1} \cdot F_{PER}}{6^{(1-SPD)} \cdot 32 \cdot (256 - BRL)}$$

$$BRL = 256 - \frac{2^{OMOS} \cdot F_{PER}}{6^{(1-SPD)} \cdot 32 \cdot Baud Rate}$$

## Table 68. SCON\_0 register

SCON\_0 - Serial Control Register for UART 0(98h)

7	6	ť	5	4	3	2	1	0		
FE/SM0_0	SM1_0	SM	12_0	REN_0	TB8_0	RB8_0	TI_0	RI_0		
Bit Number	Bit Mnemo	nic	Descri	ption						
7	FE_0	0	Clear to Set by	Framing Error bit (SMOD0=1) Clear to reset the error state, not cleared by a valid stop bit. Set by hardware when an invalid stop bit is detected. SMOD0 must be set to enable access to the FE bit.						
	SM0_	_0	Refer t		<b>0</b> erial port mode s eared to enable		SM0_0 bit.			
6	SM1_	_0	<u>SM0</u> 0 1	port Mode bit           SM1         Mode           0         0           1         1           0         2           1         3	1 <u>Description</u> Shift Register 8-bit UART 9-bit UART 9-bit UART	<u>Baud Rate</u> F <sub>CPU PERIPH</sub> /6 Variable F <sub>CPU PERIPH</sub> /3 Variable	2 or /16			
5	SM2_	_0	Serial port Mode 2 bit / Multiprocessor Communication Enable bit Clear to disable multiprocessor communication feature. Set to enable multiprocessor communication feature in mode 2 and 3, and eventually mode 1.This bit should be cleared in mode 0.							
4	REN_	_0	Clear t	tion Enable bi o disable serial enable serial re	reception.					
3	TB8_	_0	Clear t	o transmit a log	th bit to transm gic 0 in the 9th b : 1 in the 9th bit.	oit.	nd 3			
2	RB8_	_0	Receiver Bit 8 / Ninth bit received in modes 2 and 3 Cleared by hardware if 9th bit received is a logic 0. Set by hardware if 9th bit received is a logic 1. In mode 1, if SM2_0 = 0, RB8 is the received stop bit. In mode 0 RB8 is not used.							
1	TI_C	)	Transmit Interrupt flag Clear to acknowledge interrupt. Set by hardware at the end of the 8th bit time in mode 0 or at the beginning of the stop bit in the other modes.							
0	RI_(	)	Clear to Set by	ve Interrupt fla o acknowledge hardware at th 39. in the othe	interrupt. e end of the 8th	bit time in moo	de 0, see Figure	e 38. and		

Reset Value = 0000 0000b Bit addressable





## Table 69. SCON\_1 Register

SCON\_1 - Serial Control Register for UART 1(C0h)

7	6	5	4	3	2	1	0
FE/SM0_1	SM1_1	SM2_1	REN_1	TB8_1	RB8_1	TI_1	RI_1
Bit Number	Bit Mnemo	nic Des	cription				
7	FE_	1 Clea Set	Framing Error bit (SMOD0=1) Clear to reset the error state, not cleared by a valid stop bit. Set by hardware when an invalid stop bit is detected. SMOD0_1 must be set to enable access to the FE_1 bit.				
	SM0_	_1 Refe	Serial port Mode bit 0 Refer to SM1_1 for serial port mode selection. SMOD0_1 must be cleared to enable access to the SM0_1 bit.				
6	SM1_	<u>SM(</u>	al port Mode bit <u>SM1</u> <u>Mode</u> 0 0 1 1 0 2 1 3	1 <u>Description</u> Shift Register 8-bit UART 9-bit UART 9-bit UART	<u>Baud Rate</u> F <sub>CPU PERIPH</sub> /6 Variable F <sub>CPU PERIPH</sub> /3 Variable	2 or /16	
5	SM2_	_1 Clea	Serial port Mode 2 bit / Multiprocessor Communication Enable bit Clear to disable multiprocessor communication feature. Set to enable multiprocessor communication feature in mode 2 and 3, and eventually mode 1.This bit should be cleared in mode 0.				
4	REN_	_1 Clea	eption Enable bi Ir to disable seria Io enable serial re	l reception.			
3	TB8_	1 Clea	smitter Bit 8 / Nir r to transmit a log o transmit a logic	gic 0 in the 9th b	oit.	nd 3	
2	RB8_	1 Clea Set	Receiver Bit 8 / Ninth bit received in modes 2 and 3 Cleared by hardware if 9th bit received is a logic 0. Set by hardware if 9th bit received is a logic 1. In mode 1, if SM2_1 = 0, RB8 is the received stop bit. In mode 0 RB8 is not used.				
1	TI_1	Clea Set	Transmit Interrupt flag Clear to acknowledge interrupt. Set by hardware at the end of the 8th bit time in mode 0 or at the beginning of the stop bit in the other modes.				
0	RI_1	Clea Set	eive Interrupt fla r to acknowledge by hardware at th re 39. in the othe	e interrupt. e end of the 8th	bit time in moo	de 0, see Figur	e 38. and

Reset Value = 0000 0000b Bit addressable

Baud Rates	F <sub>osc</sub> = 16	. 384 MHz	F <sub>osc</sub> = 24MHz			
	BRL	BRL Error (%)		Error (%)		
115200	247	1.23	243	0.16		
57600	238	1.23	230	0.16		
38400	229	1.23	217	0.16		
28800	220	1.23	204	0.16		
19200	203	0.63	178	0.16		
9600	149	0.31	100	0.16		
4800	43	1.23	-	-		

Table 70. Example of Computed Value When X2=1, SMOD1=1, SPD=
--

Table 71.	Example of (	Computed	Value When	X2=0, S	MOD1=0, SPD=0
-----------	--------------	----------	------------	---------	---------------

Baud Rates	F <sub>OSC</sub> = 16	. 384 MHz	F <sub>osc</sub> = 24MHz		
	BRL	Error (%)	BRL	Error (%)	
4800	247	1.23	243	0.16	
2400	238	1.23	230	0.16	
1200	220	1.23	202	3.55	
600	185	0.16	152	0.16	

The baud rate generator can be used for mode 1 or 3 (refer to Figure 40.), but also for mode 0 for UART, thanks to the bit SRC located in BDRCON register (Table 78.)

## UART Registers Table 72. SBUF\_0 register

SBUF\_0 - Serial Buffer Register for UART 0(99h)

7	6	5	4	3	2	1	0

Reset Value = XXXX XXXXb

#### Table 73. BRL\_0 register

BRL\_0 - Baud Rate Reload Register for the internal baud rate generator 0 (9Ah)

7	6	5	4	3	2	1	0





## Table 74. SBUF\_1 Register

SBUF - Serial Buffer Register for UART 1(C1h)

7	6	5	4	3	2	1	0

Reset Value = XXXX XXXXb

#### Table 75. BRL\_1 Register

BRL - Baud Rate Reload Register for the internal baud rate generator 1 (BBh)

7	6	5	4	3	2	1	0

## Table 76. T2CON Register

T2CON - Timer 2 Control Register (C8h)

7	6	5	4	3	2	1	0		
TF2	EXF2	RCLK	RCLK TCLK EXEN2 TR2 C/T2# CP/RI						
Bit Number	Bit Mnemonic		Description						
7	TF2	Must be clea	<b>mer 2 overflow Flag</b> ust be cleared by software. et by hardware on timer 2 overflow, if RCLK = 0 and TCLK = 0.						
6	EXF2	Set when a c EXEN2=1. When set, ca enabled. Must be clear	ner 2 External Flag t when a capture or a reload is caused by a negative transition on T2EX pin if EN2=1. hen set, causes the CPU to vector to timer 2 interrupt routine when timer 2 interrupt is						
5	RCLK	Cleared to us	Receive Clock bit for UART Cleared to use timer 1 overflow as receive clock for serial port in mode 1 or 3. Set to use timer 2 overflow as receive clock for serial port in mode 1 or 3.						
4	TCLK	Cleared to us	Transmit Clock bit for UART Cleared to use timer 1 overflow as transmit clock for serial port in mode 1 or 3. Set to use timer 2 overflow as transmit clock for serial port in mode 1 or 3.						
3	EXEN2	Cleared to ig Set to cause	a capture or re	<b>it</b> 12EX pin for ti load when a ne the serial port.	•		is detected, if		
2	TR2	Timer 2 Run Cleared to tu Set to turn or	rn off timer 2.						
1	C/T2#	Cleared for ti Set for count	<b>Timer/Counter 2 select bit</b> Cleared for timer operation (input from internal clock system: F <sub>CLK PERIPH</sub> ). Set for counter operation (input from T2 input pin, falling edge trigger). Must be 0 for clock out mode.						
0	CP/RL2#	If RCLK=1 or overflow. Cleared to au EXEN2=1.	ito-reload on ti	it RL2# is ignored mer 2 overflows transitions on T2	s or negative tra	ansitions on T2			

Reset Value = 0000 0000b Bit addressable





#### Table 77. PCON Register

PCON - Power Control Register (87h)

7	6	5	4	3	2	1	0	
SMOD1_0	SMOD0_0	-	POF	GF1	GF0	PD	IDL	
Bit Number	Bit Mnemonic	Description						
7	SMOD1_0		Serial port Mode bit 1 for UART Set to select double baud rate in mode 1, 2 or 3.					
6	SMOD0_0	Cleared to se	Serial port Mode bit 0 for UART Cleared to select SM0 bit in SCON register. Set to select FE bit in SCON register.					
5	-	Reserved The value rea	Reserved The value read from this bit is indeterminate. Do not set this bit.					
4	POF	Cleared to re	Power-Off Flag Cleared to recognize next reset type. Set by hardware when VCC rises from 0 to its nominal voltage. Can also be set by software.					
3	GF1	Cleared by u	General purpose Flag Cleared by user for general purpose usage. Set by user for general purpose usage.					
2	GF0	General purpose Flag Cleared by user for general purpose usage. Set by user for general purpose usage.						
1	PD	Cleared by h	Power-Down mode bit Cleared by hardware when reset occurs. Set to enter power-down mode.					
0	IDL	Idle mode bit Cleared by hardware when interrupt or reset occurs. Set to enter idle mode.						

Reset Value = 00X1 0000b Not bit addressable

Power-off flag reset value will be 1 only after a power on (cold reset). A warm reset doesn't affect the value of this bit.

## Table 78. BDRCON\_0 Register

BDRCON\_0 - Baud Rate Control Register (9Bh)

7	6	5	4	3	2	1	0	
-	-	-	BRR_0	TBCK_0	RBCK_0	SPD_0	SRC_0	
Bit Number	Bit Mnemonic	Description						
7	-	<b>Reserved</b> The value read from this bit is indeterminate. Do not set this bit						
6	-	Reserved The value read from this bit is indeterminate. Do not set this bit						
5	-	<b>Reserved</b> The value read from this bit is indeterminate. Do not set this bit.						
4	BRR_0	Baud Rate Run Control bit Cleared to stop the internal Baud Rate Generator. Set to start the internal Baud Rate Generator.						
3	TBCK_0	Transmission Baud rate Generator Selection bit for UART Cleared to select Timer 1 or Timer 2 for the Baud Rate Generator. Set to select internal Baud Rate Generator.						
2	RBCK_0	Reception Baud Rate Generator Selection bit for UART Cleared to select Timer 1 or Timer 2 for the Baud Rate Generator. Set to select internal Baud Rate Generator.						
1	SPD_0	Baud Rate Speed Control bit for UART Cleared to select the SLOW Baud Rate Generator. Set to select the FAST Baud Rate Generator.						
0	SRC_0	Baud Rate Source select bit in Mode 0 for UART Cleared to select $F_{OSC}/12$ as the Baud Rate Generator ( $F_{CLK PERIPH}/6$ in X2 mode). Set to select the internal Baud Rate Generator for UARTs in mode 0.						

Reset Value = XXX0 0000b Not bit addressable





## Table 79. BDRCON\_1 Register

BDRCON - Baud Rate Control Register (BCh)

7	6	5	4	3	2	1	0	
SMOD1_1	SMOD0_1	-	BRR_1	TBCK_1	RBCK_1	SPD_1	SRC_1	
Bit Number	Bit Mnemonic	Description						
7	SMOD1_1	Serial port Mode bit 1 for UART 1 Set to select double baud rate in mode 1, 2 or 3.						
6	SMOD0_1	Serial port Mode bit 0 for UART 1 Cleared to select SM0 bit in SCON register. Set to select FE bit in SCON register.						
5	-	Reserved The value read from this bit is indeterminate. Do not set this bit.						
4	BRR_1	Cleared to st	Baud Rate Run Control bit Cleared to stop the internal Baud Rate Generator. Set to start the internal Baud Rate Generator.					
3	TBCK_1	Transmission Baud rate Generator Selection bit for UART 1 Cleared to select Timer 1 or Timer 2 for the Baud Rate Generator. Set to select internal Baud Rate Generator.						
2	RBCK_1	Reception Baud Rate Generator Selection bit for UART 1 Cleared to select Timer 1 or Timer 2 for the Baud Rate Generator. Set to select internal Baud Rate Generator.						
1	SPD_1	Baud Rate Speed Control bit for UART 1 Cleared to select the SLOW Baud Rate Generator. Set to select the FAST Baud Rate Generator.						
0	SRC_1	Baud Rate Source select bit in Mode 0 for UART 1 Cleared to select $F_{OSC}/12$ as the Baud Rate Generator ( $F_{CLK PERIPH}/6$ in X2 mode). Set to select the internal Baud Rate Generator for UARTs in mode 0.						

Reset Value = 0000 0000b Not bit addressable

#### Interrupt System

The AT89C51RE2 has a total of 10 interrupt vectors: two external interrupts (INT0 and INT1), three timer interrupts (timers 0, 1 and 2), two serial ports interrupts, SPI interrupt, Keyboard interrupt and the PCA global interrupt. These interrupts are shown in Figure 44.

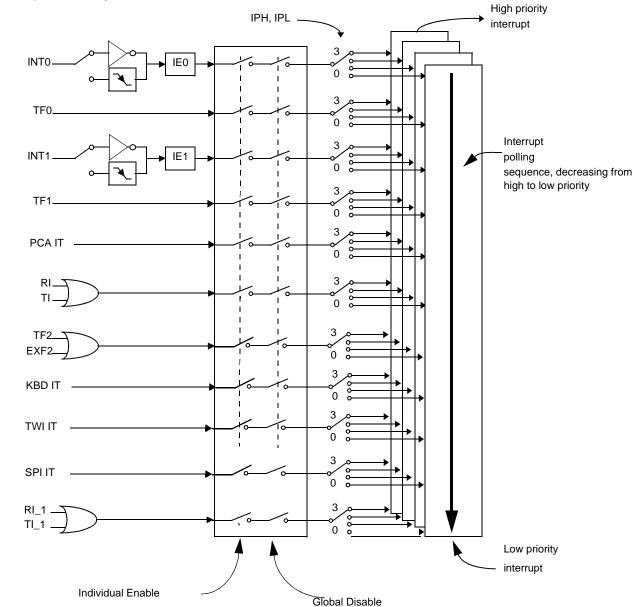


Figure 44. Interrupt Control System

Each of the interrupt sources can be individually enabled or disabled by setting or clearing a bit in the Interrupt Enable register (Table 84 and Table 82). This register also contains a global disable bit, which must be cleared to disable all interrupts at once.

Each interrupt source can also be individually programmed to one out of four priority levels by setting or clearing a bit in the Interrupt Priority register (Table 85) and in the Interrupt Priority High register (Table 83 and Table 84) shows the bit values and priority levels associated with each combination.





#### Registers

Table 80. Priority Level Bit Values

iph. x	ipl. x	interrupt level priority
0	0	0 (lowest)
0	1	1
1	0	2
1	1	3 (highest)

A low-priority interrupt can be interrupted by a high priority interrupt, but not by another low-priority interrupt. A high-priority interrupt can't be interrupted by any other interrupt source.

If two interrupt requests of different priority levels are received simultaneously, the request of higher priority level is serviced. If interrupt requests of the same priority level are received simultaneously, an internal polling sequence determines which request is serviced. Thus within each priority level there is a second priority structure determined by the polling sequence.

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#### Table 81. IEN0 Register

IEN0 - Interrupt Enable Register (A8h)

7	6	5	4	3	2	1	0	
EA	EC	ET2	ES	ET1	EX1	ET0	EX0	
Bit Number	Bit Mnemonic	Description						
7	EA	Enable All in Cleared to dis Set to enable	able all interru	pts.				
6	EC	PCA interrup Cleared to dis Set to enable.	able.					
5	ET2	Cleared to dis	low interrupt able timer 2 ov timer 2 overflo	verflow interrupt				
4	ES		Enable bit able serial port serial port inter					
3	ET1	Cleared to dis	<b>low interrupt</b> able timer 1 ov timer 1 overflo	verflow interrupt				
2	EX1	Cleared to dis	External interrupt 1 Enable bit Cleared to disable external interrupt 1. Set to enable external interrupt 1.					
1	ET0	Cleared to dis	Timer 0 overflow interrupt Enable bit Cleared to disable timer 0 overflow interrupt. Set to enable timer 0 overflow interrupt.					
0	EX0	Cleared to dis	rrupt 0 Enable able external in external interru	nterrupt 0.				

Reset Value = 0000 0000b Bit addressable





#### Table 82. IPL0 Register

IPL0 - Interrupt Priority Register (B8h)

7	6	5	4	3	2	1	0
-	PPCL	PT2L	PSL	PT1L	PX1L	PTOL	PX0L
Bit Number	Bit Mnemonic	Description					
7	-	Reserved The value rea	d from this bit	is indeterminate	e. Do not set thi	s bit.	
6	PPCL	PCA interrup Refer to PPC	<b>t Priority bit</b> H for priority le	vel.			
5	PT2L		flow interrupt				
4	PSL	Serial port 0 Refer to PSH	Priority bit for priority leve	el.			
3	PT1L		flow interrupt I for priority lev	•			
2	PX1L		rrupt 1 Priorit				
1	PTOL		flow interrupt I for priority lev	•			
0	PX0L		rrupt 0 Priorit I for priority lev	•			

Reset Value = X000 0000b Bit addressable

#### Table 83. IPH0 Register

IPH0 - Interrupt Priority High Register (B7h)

7	6	5	4	3	2	1	0
-	PPCH	PT2H	PSH	PT1H	PX1H	PT0H	PX0H
Bit Number	Bit Mnemonic	Description					
7	-	<b>Reserved</b> The value rea	d from this bit i	s indeterminate	e. Do not set this	s bit.	
6	PPCH	PCA interrup           PPCH         PPCL           0         0           1         0           1         1	<b>t Priority high</b> <u>Priority Level</u> Lowest Highest	bit.			
5	PT2H	Timer 2 over           PT2H         PT2L           0         0           1         0           1         1	f <b>low interrupt</b> <u>Priority Level</u> Lowest Highest	Priority High t	it		
4	PSH	Serial port P           PSH         PSL           0         0           0         1           1         0           1         1	riority High bit <u>Priority Level</u> Lowest Highest				
3	PT1H	Timer 1 over           PT1H         PT1L           0         0           1         0           1         1           1         1	f <b>low interrupt</b> <u>Priority Level</u> Lowest Highest	Priority High k	)it		
2	PX1H		<b>rrupt 1 Priority</b> <u>Priority Level</u> Lowest Highest				
1	РТОН	Timer 0 over           PTOH         PTOL           0         0           1         0           1         1	f <b>low interrupt</b> <u>Priority Level</u> Lowest Highest	Priority High b	, it		
0	РХОН	External inte           PX0H         PX0L           0         0           1         0           1         1	<b>rrupt 0 Priorit</b> y <u>Priority Level</u> Lowest Highest	y High bit			

Reset Value = X000 0000b Not bit addressable





#### Table 84. IEN1 Register

IEN1 - Interrupt Enable Register (B1h)

7	6	5	4	3	2	1	0	
-	-	-	-	ES_1	ESPI	ETWI	EKBD	
Bit Number	Bit Mnemonic	Description						
7	-	Reserved						
6	-	Reserved						
5	-	Reserved						
4	-	Reserved						
3	ES_1		Enable bit able serial por serial port inte					
2	ESPI	SPI interrupt Cleared to dis Set to enable	able SPI interr	upt.				
1	ETWI	Cleared to dis	TWI interrupt Enable bit Cleared to disable TWI interrupt. Set to enable TWI interrupt.					
0	EKBD	Cleared to dis	errupt Enable able keyboard keyboard inter	interrupt.				

Reset Value = XXXX 00x0b Bit addressable

#### Table 85. IPL1 Register

IPL1 - Interrupt Priority Register (B2h)

7	6	5	4	3	2	1	0	
-	-	-	-	PSL_1	SPIL	TWIL	KBDL	
Bit Number	Bit Mnemonic	Description						
7	-	Reserved The value rea	d from this bit i	s indeterminate	e. Do not set thi	s bit.		
6	-	Reserved The value rea	d from this bit i	s indeterminate	e. Do not set thi	s bit.		
5	-	Reserved The value rea	d from this bit i	s indeterminate	e. Do not set thi	s bit.		
4	-	<b>Reserved</b> The value rea	d from this bit i	s indeterminate	e. Do not set thi	s bit.		
3	PSL_1	Serial port 1 Refer to PSH	Priority bit _1 for priority le	evel.				
2	SPIL	SPI interrupt Refer to SPIH	Priority bit for priority leve	el.				
1	TWIL	_	TWI interrupt Priority bit Refer to TWIH for priority level.					
0	KBDL		errupt Priority H for priority lev					

Reset Value = XXXX 00X0b Bit addressable





#### Table 86. IPH1 Register

IPH1 - Interrupt Priority High Register (B3h)

7	6	5	4	3	2	1	0
-	-	-	-	PSH_1	SPIH	тwін	KBDH
Bit Number	Bit Mnemonic	Description					
7	-	<b>Reserved</b> The value rea	d from this bit i	is indeterminate	e. Do not set thi	s bit.	
6	-	Reserved The value rea	d from this bit i	is indeterminate	e. Do not set thi	s bit.	
5	-	<b>Reserved</b> The value rea	d from this bit i	is indeterminate	e. Do not set thi	s bit.	
4	-	<b>Reserved</b> The value rea	d from this bit i	is indeterminate	e. Do not set thi	s bit.	
3	PSH_1		Priority High I <u>  Priority Level</u> Lowest Highest				
2	SPIH	SPI interrupt           SPIH         SPIL           0         0           0         1           1         0           1         1	Priority High Priority Level Lowest Highest				
1	TWIH		Priority High Priority Level Lowest Highest				
0	KBDH			Level			

Reset Value = XXXX 00X0b Not bit addressable

#### Interrupt Sources and Vector Addresses

 Table 87. Interrupt Sources and Vector Addresses

Number	Polling Priority	Interrupt Source	Interrupt Request	Vector Address
0	0	Reset		0000h
1	1	INTO	IE0	0003h
2	2	Timer 0	TF0	000Bh
3	3	INT1	IE1	0013h
4	4	Timer 1	IF1	001Bh
5	6	UART0	RI+TI	0023h
6	7	Timer 2	TF2+EXF2	002Bh
7	5	PCA	CF + CCFn (n = 0-4)	0033h
8	8	Keyboard	KBDIT	003Bh
9	9	TWI	TWIIT	0043h
10	10	SPI	SPIIT	004Bh
11	11	UART1	RI_1+TI_1	0053h





## **Power Management**

Introduction	Power power	ower reduction modes are implemented in the AT89C51RE2. The Idle mode and the r-Down mode. These modes are detailed in the following sections. In addition to these reduction modes, the clocks of the core and peripherals can be dynamically divided by 2 the X2 mode detailed in Section "Enhanced Features", page 13.						
Idle Mode	gram periph the pro mode.	ode is a power reduction mode that reduces the power consumption. In this mode, pro- execution halts. Idle mode freezes the clock to the CPU at known states while the erals continue to be clocked. The CPU status before entering Idle mode is preserved, i.e., ogram counter and program status word register retain their data for the duration of Idle The contents of the SFRs and RAM are also retained. The status of the Port pins during ode is detailed in Table 88.						
Entering Idle Mode	Idle m the las	ter Idle mode, set the IDL bit in PCON register (see Table 89). The AT89C51RE2 enters ode upon execution of the instruction that sets IDL bit. The instruction that sets IDL bit is st instruction executed.						
	Note:	If IDL bit and PD bit are set simultaneously, the AT89C51RE2 enters Power-Down mode. Then it does not go in Idle mode when exiting Power-Down mode.						
Exiting Idle Mode	There	are two ways to exit Idle mode:						
	1. Generate an enabled interrupt.							
	_	Hardware clears IDL bit in PCON register which restores the clock to the CPU. Execution resumes with the interrupt service routine. Upon completion of the interrupt service routine, program execution resumes with the instruction immediately following the instruction that activated Idle mode. The general purpose flags (GF1 and GF0 in PCON register) may be used to indicate whether an interrupt occurred during normal operation or during Idle mode. When Idle mode is exited by an interrupt, the interrupt service routine may examine GF1 and GF0.						
	2. Ge	enerate a reset.						
	_	A logic high on the RST pin clears IDL bit in PCON register directly and asynchronously. This restores the clock to the CPU. Program execution momentarily resumes with the instruction immediately following the instruction that activated the Idle mode and may continue for a number of clock cycles before the internal reset algorithm takes control. Reset initializes the AT89C51RE2 and vectors the CPU to address C:0000h.						
	Note:	During the time that execution resumes, the internal RAM cannot be accessed; however, it is pos- sible for the Port pins to be accessed. To avoid unexpected outputs at the Port pins, the instruction immediately following the instruction that activated Idle mode should not write to a Port pin or to the external RAM.						
Power-Down Mode	stops Down data fe	ower-Down mode places the AT89C51RE2 in a very low power state. Power-Down mode the oscillator, freezes all clock at known states. The CPU status prior to entering Power-mode is preserved, i.e., the program counter, program status word register retain their or the duration of Power-Down mode. In addition, the SFR and RAM contents are pred. The status of the Port pins during Power-Down mode is detailed in Table 88. VCC may be reduced to as low as $V_{RET}$ during Power-Down mode to further reduce power dissipation. Take care, however, that VDD is not reduced until Power-Down mode is invoked.						

Entering Power-Down To enter Power-Down mode, set PD bit in PCON register. The AT89C51RE2 enters the Power-Down mode upon execution of the instruction that sets PD bit. The instruction that sets PD bit is the last instruction executed.

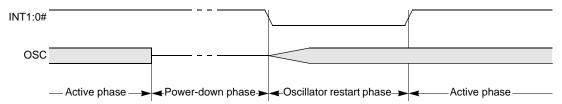
Exiting Power-Down Mode

Note: If VCC was reduced during the Power-Down mode, do not exit Power-Down mode until VCC is restored to the normal operating level.

There are two ways to exit the Power-Down mode:

- 1. Generate an enabled external interrupt.
  - The AT89C51RE2 provides capability to exit from Power-Down using INT0#, INT1#. Hardware clears PD bit in PCON register which starts the oscillator and restores the clocks to the CPU and peripherals. Using INTx# input, execution resumes when the input is released (see Figure 45). Execution resumes with the interrupt service routine. Upon completion of the interrupt service routine, program execution resumes with the instruction immediately following the instruction that activated Power-Down mode.
- Note: The external interrupt used to exit Power-Down mode must be configured as level sensitive (INT0# and INT1#) and must be assigned the highest priority. In addition, the duration of the interrupt must be long enough to allow the oscillator to stabilize. The execution will only resume when the interrupt is deasserted.
- Note: Exit from power-down by external interrupt does not affect the SFRs nor the internal RAM content.

Figure 45. Power-Down Exit Waveform Using INT1:0#



- 2. Generate a reset.
  - A logic high on the RST pin clears PD bit in PCON register directly and asynchronously. This starts the oscillator and restores the clock to the CPU and peripherals. Program execution momentarily resumes with the instruction immediately following the instruction that activated Power-Down mode and may continue for a number of clock cycles before the internal reset algorithm takes control. Reset initializes the AT89C51RE2 and vectors the CPU to address 0000h.
- Note: During the time that execution resumes, the internal RAM cannot be accessed; however, it is possible for the Port pins to be accessed. To avoid unexpected outputs at the Port pins, the instruction immediately following the instruction that activated the Power-Down mode should not write to a Port pin or to the external RAM.
- Note: Exit from power-down by reset redefines all the SFRs, but does not affect the internal RAM content.





Mode	Port 0	Port 1	Port 2	Port 3	Port 4	ALE	PSEN#
Reset	Floating	High	High	High	High	High	High
Idle (internal code)	Data	Data	Data	Data	Data	High	High
Idle (external code)	Floating	Data	Data	Data	Data	High	High
Power-Down (internal code)	Data	Data	Data	Data	Data	Low	Low
Power-Down (external code)	Floating	Data	Data	Data	Data	Low	Low

#### Registers

**Table 89.** PCON RegisterPCON (87:h) Power configuration Register

7	6	5	4	3	2	1	0	
SMOD1	SMOD0	-	POF	GF1	GF0	PD	IDL	
Bit Number	Bit Mnemonic	Description						
7	SMOD1	Serial Port M Set to select o		e in mode 1, 2	or 3.			
6	SMOD0		ode bit 0 ect SM0 bit in S E bit in SCON	0				
5	-	reserved						
4	POF		ognize next re		its nominal volta	age. Can also l	be set by	
3	GF1	General Purp One use is to Idle mode.	•	er an interrupt o	occurred during	normal operat	ion or during	
2	GF0	General Purp One use is to Idle mode.		er an interrupt o	occurred during	normal operat	ion or during	
1	PD	Cleared by ha Set to activate	Power-Down Mode bit Cleared by hardware when an interrupt or reset occurs. Set to activate the Power-Down mode. If IDL and PD are both set, PD takes precedence.					
0	IDL	Set to activate	rdware when a the Idle mode	n interrupt or re D takes preced				

Reset Value= XXXX 0000b





# **Oscillator** To optimize the power consumption and execution time needed for a specific task, an internal prescaler feature has been implemented between the oscillator and the CPU and peripherals.

#### Registers

Table 90. CKRL Register

CKRL - Clock Reload Register (97h)

7		6	5	4	3	2	1	0		
CKRL7	С	KRL6	CKRL5	CKRL4	CKRL3	CKRL2	CKRL1	CKRL0		
Bit Numbe	ər	Mnemonic		Description						
7:0		CKRL		Clock Reload Register Prescaler value						

Reset Value = 1111 1111b Not bit addressable

#### Table 91. PCON Register

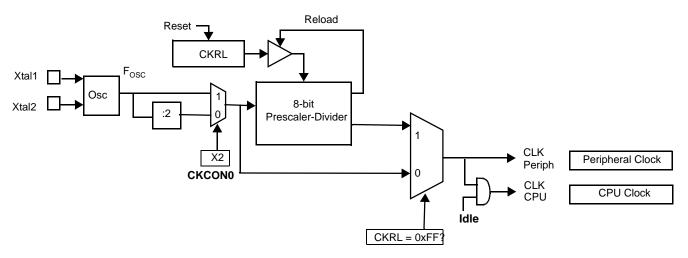
PCON - Power Control Register (87h)

7	6	5	4	3	2	1	0		
SMOD1	SMOD0	-	POF GF1 GF0 PD			IDL			
Bit Numb	er Bit N	Inemonic	Description						
7	S	MOD1	Serial Port Mode bit 1 Set to select double baud rate in mode 1, 2 or 3.						
6	S	MOD0	Serial Port Mode bit 0 Cleared to select SM0 bit in SCON register. Set to select FE bit in SCON register.						
5		-	<b>Reserved</b> The value read from this bit is indeterminate. Do not set this bit.						
4		POF	<b>Power-off Flag</b> Cleared by software to recognize the next reset type. Set by hardware when $V_{CC}$ rises from 0 to its nominal voltage. Can also be set by software.						
3		GF1	General-purpo Cleared by softw Set by software	vare for general		e.			
2		GF0	General-purpose Flag Cleared by software for general-purpose usage. Set by software for general-purpose usage.						
1		PD Power-down Mode bit Cleared by hardware when reset occurs. Set to enter power-down mode.							
0		IDL	Idle Mode bit Cleared by hardware when interrupt or reset occurs. Set to enter idle mode.						

Reset Value = 00X1 0000b Not bit addressable

#### **Functional Block Diagram**

Figure 46. Functional Oscillator Block Diagram



#### **Prescaler Divider**

- A hardware RESET puts the prescaler divider in the following state:
- CKRL = FFh:  $F_{CLK CPU} = F_{CLK PERIPH} = F_{OSC}/2$  (Standard C51 feature)
- Any value between FFh down to 00h can be written by software into CKRL register in order to divide frequency of the selected oscillator:
  - CKRL = 00h: minimum frequency  $F_{CLK CPU} = F_{CLK PERIPH} = F_{OSC}/1020$  (Standard Mode)  $F_{CLK CPU} = F_{CLK PERIPH} = F_{OSC}/510$  (X2 Mode)
  - CKRL = FFh: maximum frequency  $F_{CLK CPU} = F_{CLK PERIPH} = F_{OSC}/2$  (Standard Mode)  $F_{CLK CPU} = F_{CLK PERIPH} = F_{OSC}$  (X2 Mode)

 $\rm F_{CLK\,CPU}$  and  $\rm F_{CLK\,PERIPH}$ 

In X2 Mode, for CKRL<>0xFF:  $F_{CPU} = F_{CLKPERIPH} = \frac{F_{OSC}}{2 \times (255 - CKRL)}$ 

In X1 Mode, for CKRL<>0xFF then:  $F_{CPU} = F_{CLKPERIPH} = \frac{F_{OSC}}{4 \times (255 - CKRL)}$ 





# Hardware<br/>Watchdog<br/>TimerThe WDT is intended as a recovery method in situations where the CPU may be subjected to<br/>software upset. The WDT consists of a 14-bit counter and the WatchDog Timer ReSeT<br/>(WDTRST) SFR. The WDT is by default disabled from exiting reset. To enable the WDT, user<br/>must write 01EH and 0E1H in sequence to the WDTRST, SFR location 0A6H. When WDT is<br/>enabled, it will increment every machine cycle while the oscillator is running and there is no way<br/>to disable the WDT except through reset (either hardware reset or WDT overflow reset). When<br/>WDT overflows, it will drive an output RESET HIGH pulse at the RST-pin.

**Using the WDT** To enable the WDT, user must write 01EH and 0E1H in sequence to the WDTRST, SFR location 0A6H. When WDT is enabled, the user needs to service it by writing to 01EH and 0E1H to WDTRST to avoid WDT overflow. The 14-bit counter overflows when it reaches 16383 (3FFFH) and this will reset the device. When WDT is enabled, it will increment every machine cycle while the oscillator is running. This means the user must reset the WDT at least every 16383 machine cycle. To reset the WDT the user must write 01EH and 0E1H to WDTRST. WDTRST is a write only register. The WDT counter cannot be read or written. When WDT overflows, it will generate an output RESET pulse at the RST-pin. The RESET pulse duration is 96 x T<sub>CLK PERIPH</sub>, where T<sub>CLK PERIPH</sub> = 1/F<sub>CLK PERIPH</sub>. To make the best use of the WDT, it should be serviced in those sections of code that will periodically be executed within the time required to prevent a WDT reset.

To have a more powerful WDT, a  $2^7$  counter has been added to extend the Time-out capability, ranking from 16ms to 2s @  $F_{OSCA}$  = 12MHz. To manage this feature, refer to WDTPRG register description, Table 92.

 Table 92.
 WDTRST Register

WDTRST - Watchdog Reset Register (0A6h)

7	6	5	4	3	2	1	0
-	-	-	-	-	-	-	-

Reset Value = XXXX XXXXb

Write only, this SFR is used to reset/enable the WDT by writing 01EH then 0E1H in sequence.

#### Table 93. WDTPRG Register

WDTPRG - Watchdog Timer Out Register (0A7h)

7	6	5	4	3	2	1	0
-	-	-	-	-	S2	S1	S0
Bit Number	Bit Mnemonic	Description					
7	-						
6	-						
5	-	Reserved The va	alue read from	this bit is undete	rmined. Do not	trv to set this I	oit.
4	-					,	
3	-						
2	S2	WDT Time-o	ut select bit 2				
1	S1	WDT Time-o	ut select bit 1				
0	S0	WDT Time-o	ut select bit 0				
	-	S2         S1           0         0           0         1           0         1           1         0           1         1           1         1           1         1	<b>S0</b> 0 1 0 1 0 1 0 1	Selected Time- $(2^{14} - 1)$ machin $(2^{15} - 1)$ machin $(2^{16} - 1)$ machin $(2^{17} - 1)$ machin $(2^{18} - 1)$ machin $(2^{19} - 1)$ machin $(2^{20} - 1)$ machin $(2^{21} - 1)$ machin	e cycles, 16. 3 e cycles, 32.7 r e cycles, 65. 5 e cycles, 131 m e cycles, 262 m e cycles, 542 m e cycles, 1.05 s	ns @ $F_{OSCA}$ =1 ms @ $F_{OSCA}$ =1 ns @ $F_{OSCA}$ =12 ns @ $F_{OSCA}$ =12 ns @ $F_{OSCA}$ =12 s @ $F_{OSCA}$ =12 s @ $F_{OSCA}$ =12	2 MHz 2 MHz 2 MHz 2 MHz 2 MHz MHz MHz

Reset value = XXXX X000

#### WDT During Power Down and Idle

In Power Down mode the oscillator stops, which means the WDT also stops. While in Power Down mode the user does not need to service the WDT. There are 2 methods of exiting Power Down mode: by a hardware reset or via a level activated external interrupt which is enabled prior to entering Power Down mode. When Power Down is exited with hardware reset, servicing the WDT should occur as it normally should whenever the AT89C51RE2 is reset. Exiting Power Down with an interrupt is significantly different. The interrupt is held low long enough for the oscillator to stabilize. When the interrupt is brought high, the interrupt is serviced. To prevent the WDT from resetting the device while the interrupt pin is held low, the WDT is not started until the interrupt is pulled high. It is suggested that the WDT be reset during the interrupt service routine.

To ensure that the WDT does not overflow within a few states of exiting of powerdown, it is better to reset the WDT just before entering powerdown.

In the Idle mode, the oscillator continues to run. To prevent the WDT from resetting the AT89C51RE2 while in Idle mode, the user should always set up a timer that will periodically exit Idle, service the WDT, and re-enter Idle mode.





# Reduced EMIThe ALE signal is unal program or dateModeThe ALE signal is unal program or date

The ALE signal is used to demultiplex address and data buses on port 0 when used with external program or data memory. Nevertheless, during internal code execution, ALE signal is still generated. In order to reduce EMI, ALE signal can be disabled by setting AO bit.

The AO bit is located in AUXR register at bit location 0. As soon as AO is set, ALE is no longer output but remains active during MOVX and MOVC instructions and external fetches. During ALE disabling, ALE pin is weakly pulled high.

#### Table 94. AUXR Register

AUXR - Auxiliary Register (8Eh)

7	6	5	4	3	2	1	0
-	-	MO	XRS2	XRS1	XRS0	EXTRAM	AO
Bit Number	Bit Mnemonic	Description					
7	-	Reserved The value rea	ad from this bi	t is indeterminate	e. Do not set th	is bit.	
6	-	Reserved The value rea	ad from this bi	t is indeterminate	e. Do not set th	is bit.	
5	МО	(default).	etch MOVX c		·	lse length is 6 clo ength is 30 clock	
4	XRS2	XRAM Size					
3	XRS1	<u>XRS2 XR</u> 0 0	<u>S1</u> <u>XRS0</u> 0	XRAM size 256 bytes			
2	XRS0	0 0 0 1 0 1 1 0	1 0 1 0	512 bytes 768 bytes(defau 1024 bytes 1792 bytes	lt)		
1	EXTRAM	Set to access Programmed	EXTRAM bit Cleared to access internal XRAM using movx @ Ri/ @ DPTR. Set to access external memory. Programmed by hardware after Power-up regarding Hardware Security Byte (HSB), default setting, XRAM selected.				
0	AO		is emitted at			ator frequency (o OVX or MOVC ir	

Reset Value = XX00 10'HSB. XRAM'0b Not bit addressable

#### Keyboard Interface

The AT89C51RE2 implements a keyboard interface allowing the connection of a 8 x n matrix keyboard. It is based on 8 inputs with programmable interrupt capability on both high or low level. These inputs are available as alternate function of P1 and allow to exit from idle and power down modes.

The keyboard interface interfaces with the C51 core through 3 special function registers: KBLS, the Keyboard Level Selection register (Table 97), KBE, The Keyboard interrupt Enable register (Table 96), and KBF, the Keyboard Flag register (Table 95).

Interrupt The keyboard inputs are considered as 8 independent interrupt sources sharing the same interrupt vector. An interrupt enable bit (KBD in IE1) allows global enable or disable of the keyboard interrupt (see Figure 47). As detailed in Figure 48 each keyboard input has the capability to detect a programmable level according to KBLS. x bit value. Level detection is then reported in interrupt flags KBF. x that can be masked by software using KBE. x bits.

This structure allow keyboard arrangement from 1 by n to 8 by n matrix and allow usage of P1 inputs for other purpose.



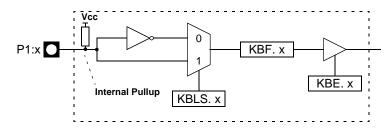
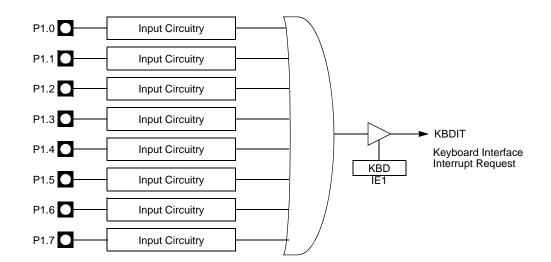


Figure 48. Keyboard Input Circuitry



# Power Reduction Mode

P1 inputs allow exit from idle and power down modes as detailed in Section "Power Management", page 118.





#### Registers

#### Table 95. KBF Register

KBF-Keyboard Flag Register (9Eh)

7	6	5	4	3	2	1	0
KBF7	KBF6	KBF5	KBF4	KBF3	KBF2	KBF1	KBF0
Bit Number	Bit Mnemonic	Description					
7	KBF7	Keyboard inte	are when the Po	ort line 7 detect the KBKBIE. 7			rates a
6	KBF6	Keyboard inte	are when the Po	ort line 6 detect the KBIE. 6 bit			rates a
5	KBF5	Set by hardwa Keyboard inte	<b>Keyboard line 5 flag</b> Set by hardware when the Port line 5 detects a programmed level. It generates a Keyboard interrupt request if the KBIE. 5 bit in KBIE register is set. Must be cleared by software.				
4	KBF4	Set by hardwa Keyboard inte	<b>Keyboard line 4 flag</b> Set by hardware when the Port line 4 detects a programmed level. It generates a Keyboard interrupt request if the KBIE. 4 bit in KBIE register is set. Must be cleared by software.				
3	KBF3	Keyboard inte	are when the Po	ort line 3 detect the KBIE. 3 bit			rates a
2	KBF2	Set by hardwa Keyboard inte	<b>Keyboard line 2 flag</b> Set by hardware when the Port line 2 detects a programmed level. It generates Keyboard interrupt request if the KBIE. 2 bit in KBIE register is set. Must be cleared by software.				rates a
1	KBF1	<b>Keyboard line 1 flag</b> Set by hardware when the Port line 1 detects a programmed level. It generate Keyboard interrupt request if the KBIE. 1 bit in KBIE register is set. Must be cleared by software.					rates a
0	KBF0	Keyboard inte	are when the Po	ort line 0 detect the KBIE. 0 bit		-	rates a

Reset Value= 0000 0000b

#### Table 96. KBE Register

KBE-Keyboard Input Enable Register (9Dh)

7	6	5	4	3	2	1	0
KBE7	KBE6	KBE5	KBE4	KBE3	KBE2	KBE1	KBE0
Bit Number	Bit Mnemonic	Description					
7	KBE7	Cleared to ena	e <b>7 Enable bit</b> able standard I KBF. 7 bit in Kl	/O pin.	enerate an inte	rrupt request.	
6	KBE6	Cleared to ena	<b>Teyboard line 6 Enable bit</b> Ieared to enable standard I/O pin. et to enable KBF. 6 bit in KBF register to generate an interrupt request.				
5	KBE5	Cleared to ena	<b>Keyboard line 5 Enable bit</b> Cleared to enable standard I/O pin. Set to enable KBF. 5 bit in KBF register to generate an interrupt request.				
4	KBE4	Cleared to ena	Keyboard line 4 Enable bit Cleared to enable standard I/O pin. Set to enable KBF. 4 bit in KBF register to generate an interrupt request.				
3	KBE3	Cleared to ena	e 3 Enable bit able standard I KBF. 3 bit in KI	/O pin.	enerate an inte	rrupt request.	
2	KBE2	Cleared to ena	e <b>2 Enable bit</b> able standard I KBF. 2 bit in KI		enerate an inte	rrupt request.	
1	KBE1	Cleared to ena	<b>Keyboard line 1 Enable bit</b> Cleared to enable standard I/O pin. Set to enable KBF. 1 bit in KBF register to generate an interrupt request.				
0	KBE0	Cleared to ena	e <b>0 Enable bit</b> able standard I KBF. 0 bit in Kl	•	enerate an inte	rrupt request.	

Reset Value= 0000 0000b





#### Table 97. KBLS Register

KBLS-Keyboard Level Selector Register (9Ch)

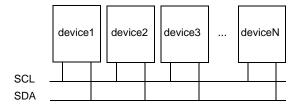
7	6	5	4	3	2	1	0	
KBLS7	KBLS6	KBLS5	KBLS4	KBLS3	KBLS2	KBLS1	KBLS0	
Bit Number	Bit Mnemonic	Description	escription					
7	KBLS7	Cleared to ena		<b>ction bit</b> I detection on P tection on Port				
6	KBLS6	Cleared to en	eyboard line 6 Level Selection bit eared to enable a low level detection on Port line 6. It to enable a high level detection on Port line 6.					
5	KBLS5	Cleared to en	eyboard line 5 Level Selection bit eared to enable a low level detection on Port line 5. et to enable a high level detection on Port line 5.					
4	KBLS4	Cleared to en	Keyboard line 4 Level Selection bit Cleared to enable a low level detection on Port line 4. Set to enable a high level detection on Port line 4.					
3	KBLS3	Cleared to ena		<b>ction bit</b> I detection on P tection on Port				
2	KBLS2	Cleared to ena	Keyboard line 2 Level Selection bit Cleared to enable a low level detection on Port line 2. Set to enable a high level detection on Port line 2.					
1	KBLS1	Cleared to ena	<b>Keyboard line 1 Level Selection bit</b> Cleared to enable a low level detection on Port line 1. Set to enable a high level detection on Port line 1.					
0	KBLS0	Cleared to en		<b>ction bit</b> I detection on P tection on Port				

Reset Value= 0000 0000b

#### 2-wire Interface (TWI)

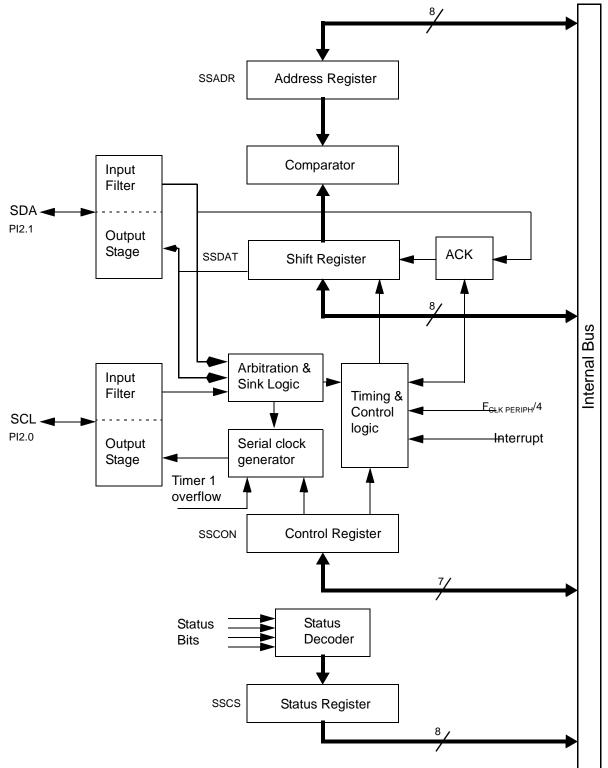
This section describes the 2-wire interface. The 2-wire bus is a bi-directional 2-wire serial communication standard. It is designed primarily for simple but efficient integrated circuit (IC) control. The system is comprised of two lines, SCL (Serial Clock) and SDA (Serial Data) that carry information between the ICs connected to them. The serial data transfer is limited to 400 Kbit/s in standard mode. Various communication configuration can be designed using this bus. Figure 49 shows a typical 2-wire bus configuration. All the devices connected to the bus can be master and slave.











#### Description

The CPU interfaces to the 2-wire logic via the following four 8-bit special function registers: the Synchronous Serial Control register (SSCON; Table 107), the Synchronous Serial Data register (SSDAT; Table 108), the Synchronous Serial Control and Status register (SSCS; Table 109) and the Synchronous Serial Address register (SSADR Table 112).

SSCON is used to enable the TWI interface, to program the bit rate (see Table 100), to enable slave modes, to acknowledge or not a received data, to send a START or a STOP condition on the 2-wire bus, and to acknowledge a serial interrupt. A hardware reset disables the TWI module.

SSCS contains a status code which reflects the status of the 2-wire logic and the 2-wire bus. The three least significant bits are always zero. The five most significant bits contains the status code. There are 26 possible status codes. When SSCS contains F8h, no relevant state information is available and no serial interrupt is requested. A valid status code is available in SSCS one machine cycle after SI is set by hardware and is still present one machine cycle after SI has been reset by software. to Table 106. give the status for the master modes and miscellaneous states.

SSDAT contains a byte of serial data to be transmitted or a byte which has just been received. It is addressable while it is not in process of shifting a byte. This occurs when 2-wire logic is in a defined state and the serial interrupt flag is set. Data in SSDAT remains stable as long as SI is set. While data is being shifted out, data on the bus is simultaneously shifted in; SSDAT always contains the last byte present on the bus.

SSADR may be loaded with the 7-bit slave address (7 most significant bits) to which the TWI module will respond when programmed as a slave transmitter or receiver. The LSB is used to enable general call address (00h) recognition.

Figure 51 shows how a data transfer is accomplished on the 2-wire bus.

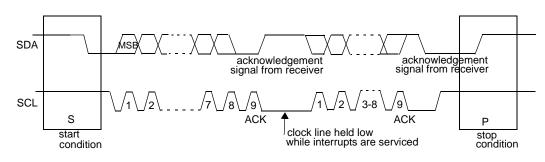


Figure 51. Complete Data Transfer on 2-wire Bus

The four operating modes are:

- Master Transmitter
- Master Receiver
- Slave transmitter
- Slave receiver

Data transfer in each mode of operation is shown in Table to Table 106 and Figure 52. to Figure 55.. These figures contain the following abbreviations:

- S : START condition
- R : Read bit (high level at SDA)





- W: Write bit (low level at SDA)
- A: Acknowledge bit (low level at SDA)

 $\overline{A}$ : Not acknowledge bit (high level at SDA)

Data: 8-bit data byte

P: STOP condition

In Figure 52 to Figure 55, circles are used to indicate when the serial interrupt flag is set. The numbers in the circles show the status code held in SSCS. At these points, a service routine must be executed to continue or complete the serial transfer. These service routines are not critical since the serial transfer is suspended until the serial interrupt flag is cleared by software.

When the serial interrupt routine is entered, the status code in SSCS is used to branch to the appropriate service routine. For each status code, the required software action and details of the following serial transfer are given in Table to Table 106.

Master Transmitter Mode In the master transmitter mode, a number of data bytes are transmitted to a slave receiver (Figure 52). Before the master transmitter mode can be entered, SSCON must be initialised as follows:

Table 98. SSCON Initialization

CR2	SSIE	STA	STO	SI	AA	CR1	CR0
bit rate	1	0	0	0	х	bit rate	bit rate

CR0, CR1 and CR2 define the internal serial bit rate if external bit rate generator is not used. SSIE must be set to enable TWI. STA, STO and SI must be cleared.

The master transmitter mode may now be entered by setting the STA bit. The 2-wire logic will now test the 2-wire bus and generate a START condition as soon as the bus becomes free. When a START condition is transmitted, the serial interrupt flag (SI bit in SSCON) is set, and the status code in SSCS will be 08h. This status must be used to vector to an interrupt routine that loads SSDAT with the slave address and the data direction bit (SLA+W).

When the slave address and the direction bit have been transmitted and an acknowledgement bit has been received, SI is set again and a number of status code in SSCS are possible. There are 18h, 20h or 38h for the master mode and also 68h, 78h or B0h if the slave mode was enabled (AA=logic 1). The appropriate action to be taken for each of these status code is detailed in Table . This scheme is repeated until a STOP condition is transmitted.

SSIE, CR2, CR1 and CR0 are not affected by the serial transfer and are referred to Table 7 to Table 11. After a repeated START condition (state 10h) the TWI module may switch to the master receiver mode by loading SSDAT with SLA+R.

Master Receiver ModeIn the master receiver mode, a number of data bytes are received from a slave transmit-<br/>ter (Figure 53). The transfer is initialized as in the master transmitter mode. When the<br/>START condition has been transmitted, the interrupt routine must load SSDAT with the<br/>7-bit slave address and the data direction bit (SLA+R). The serial interrupt flag SI must<br/>then be cleared before the serial transfer can continue.

When the slave address and the direction bit have been transmitted and an acknowledgement bit has been received, the serial interrupt flag is set again and a number of status code in SSCS are possible. There are 40h, 48h or 38h for the master mode and also 68h, 78h or B0h if the slave mode was enabled (AA=logic 1). The appropriate action to be taken for each of these status code is detailed in Table . This scheme is repeated until a STOP condition is transmitted.

SSIE, CR2, CR1 and CR0 are not affected by the serial transfer and are referred to Table 7 to Table 11. After a repeated START condition (state 10h) the TWI module may switch to the master transmitter mode by loading SSDAT with SLA+W.

Slave Receiver Mode In the slave receiver mode, a number of data bytes are received from a master transmitter (Figure 54). To initiate the slave receiver mode, SSADR and SSCON must be loaded as follows:

Table 99. SSADR: Slave Receiver Mode Initialization

A6	A5	A4	A3	A2	A1	A0	GC
	own slave address						

The upper 7 bits are the address to which the TWI module will respond when addressed by a master. If the LSB (GC) is set the TWI module will respond to the general call address (00h); otherwise it ignores the general call address.

Table 100. SSCON: Slave Receiver Mode Initialization

CR2	SSIE	STA	STO	SI	AA	CR1	CR0
bit rate	1	0	0	0	1	bit rate	bit rate

CR0, CR1 and CR2 have no effect in the slave mode. SSIE must be set to enable the TWI. The AA bit must be set to enable the own slave address or the general call address acknowledgement. STA, STO and SI must be cleared.

When SSADR and SSCON have been initialised, the TWI module waits until it is addressed by its own slave address followed by the data direction bit which must be at logic 0 (W) for the TWI to operate in the slave receiver mode. After its own slave address and the W bit have been received, the serial interrupt flag is set and a valid status code can be read from SSCS. This status code is used to vector to an interrupt service routine. The appropriate action to be taken for each of these status code is detailed in Table . The slave receiver mode may also be entered if arbitration is lost while TWI is in the master mode (states 68h and 78h).

If the AA bit is reset during a transfer, TWI module will return a not acknowledge (logic 1) to SDA after the next received data byte. While AA is reset, the TWI module does not respond to its own slave address. However, the 2-wire bus is still monitored and address recognition may be resume at any time by setting AA. This means that the AA bit may be used to temporarily isolate the module from the 2-wire bus.

Slave Transmitter Mode In the slave transmitter mode, a number of data bytes are transmitted to a master receiver (Figure 55). Data transfer is initialized as in the slave receiver mode. When SSADR and SSCON have been initialized, the TWI module waits until it is addressed by

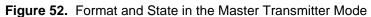


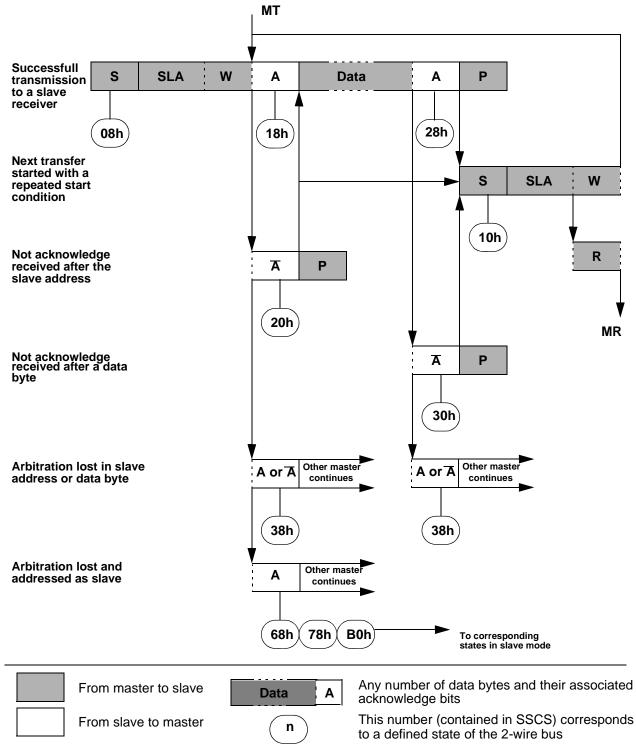
	its own slave address followed by the data direction bit which must be at logic 1 (R) for TWI to operate in the slave transmitter mode. After its own slave address and the R bit have been received, the serial interrupt flag is set and a valid status code can be read from SSCS. This status code is used to vector to an interrupt service routine. The appropriate action to be taken for each of these status code is detailed in Table . The slave transmitter mode may also be entered if arbitration is lost while the TWI module is in the master mode.
	If the AA bit is reset during a transfer, the TWI module will transmit the last byte of the transfer and enter state C0h or C8h. the TWI module is switched to the not addressed slave mode and will ignore the master receiver if it continues the transfer. Thus the master receiver receiver all 1's as serial data. While AA is reset, the TWI module does not respond to its own slave address. However, the 2-wire bus is still monitored and address recognition may be resume at any time by setting AA. This means that the AA bit may be used to temporarily isolate the TWI module from the 2-wire bus.
Miscellaneous States	There are two SSCS codes that do not correspond to a define TWI hardware state (Table 106). These codes are discuss hereafter.
	Status F8h indicates that no relevant information is available because the serial interrupt flag is not set yet. This occurs between other states and when the TWI module is not involved in a serial transfer.
	Status 00h indicates that a bus error has occurred during a TWI serial transfer. A bus error is caused when a START or a STOP condition occurs at an illegal position in the format frame. Examples of such illegal positions happen during the serial transfer of an address byte, a data byte, or an acknowledge bit. When a bus error occurs, SI is set. To recover from a bus error, the STO flag must be set and SI must be cleared. This causes the TWI module to enter the not addressed slave mode and to clear the STO flag (no other bits in SSCON are affected). The SDA and SCL lines are released and no STOP condition is transmitted.
Notes	the TWI module interfaces to the external 2-wire bus via two port pins: SCL (serial clock line) and SDA (serial data line). To avoid low level asserting on these lines when the TWI module is enabled, the output latches of SDA and SLC must be set to logic 1.

			Bit Freque	ency ( kHz)	
CR2	CR1	CR0	F <sub>OSCA</sub> = 12 MHz	F <sub>OSCA</sub> = 16 MHz	F <sub>OSCA</sub> divided by
0	0	0	47	62.5	256
0	0	1	53.5	71.5 224	
0	1	0	62.5	83	192
0	1	1	75	100	160
1	0	0	-	-	Unused
1	0	1	100	133.3	120
1	1	0	200	266.6	60
1	1	1	0.5 <. < 62.5	0.67 <. < 83	96 · (256 - reload valueTimer 1) (reload value range: 0-254 in mode 2)

 Table 101. Bit Frequency Configuration

# AT89C51RE2





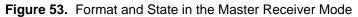


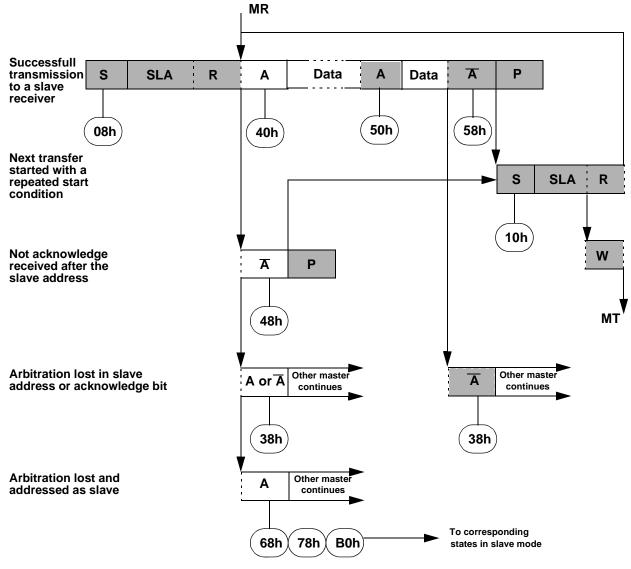


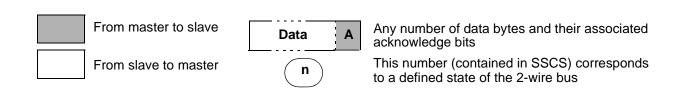
#### Table 102. Status in Master Transmitter Mode

		Application software response			nse			
Status	Status of the Two-			To SSC				
Code SSSTA	wire Bus and Two- wire Hardware	To/From SSDAT	SSSTA	SSSTO	SSI	SSAA	Next Action Taken by Two-wire Hardware	
08h	A START condition has been transmitted	Write SLA+W	x	0	0	х	SLA+W will be transmitted.	
101	A repeated START	Write SLA+W	x	0	0	х	SLA+W will be transmitted.	
10h	condition has been transmitted	Write SLA+R	x	0	0	х	SLA+R will be transmitted. Logic will switch to master receiver mode	
		Write data byte	0	0	0	х	Data byte will be transmitted. Repeated START will be transmitted.	
18h	SLA+W has been	No SSDAT action	1	0	0	X X	STOP condition will be transmitted and SSSTO flag	
1011	transmitted; ACK has been received	NO SSDAT action	0	1	0	^	will be reset.	
		No SSDAT action	1	1	0	х	STOP condition followed by a START condition will be transmitted and SSSTO flag will be reset.	
		Write data byte	0	0	0	х	Data byte will be transmitted.	
	SLA+W has been transmitted; NOT ACK has been received	No SSDAT action	1	0	0	Х	Repeated START will be transmitted. STOP condition will be transmitted and SSSTO flag	
20h		No SSDAT action	0	1	0	Х	will be reset.	
has been received		No SSDAT action	1	1	0	х	STOP condition followed by a START condition will be transmitted and SSSTO flag will be reset.	
		Write data byte	0	0	0	х	Data byte will be transmitted.	
	Data byte has been transmitted; ACK has	No SSDAT action	1	0	0	Х	Repeated START will be transmitted.	
28h		No SSDAT action	0	1	0	Х	STOP condition will be transmitted and SSSTO flag will be reset.	
been received		No SSDAT action	1	1	0	x	STOP condition followed by a START condition will be transmitted and SSSTO flag will be reset.	
		Write data byte	0	0	0	х	Data byte will be transmitted.	
	Data byte has been transmitted; NOT ACK has been received	No SSDAT action	1	0	0	х	Repeated START will be transmitted.	
		No SSDAT action	0	1	0	х	STOP condition will be transmitted and SSSTO flag will be reset.	
		No SSDAT action	1	1	0	x	STOP condition followed by a START condition will be transmitted and SSSTO flag will be reset.	
0.01	Arbitration lost in	No SSDAT action	0	0	0	x	Two-wire bus will be released and not addressed slave mode will be entered.	
38h	SLA+W or data bytes	No SSDAT action	1	0	0	х	A START condition will be transmitted when the bus becomes free.	

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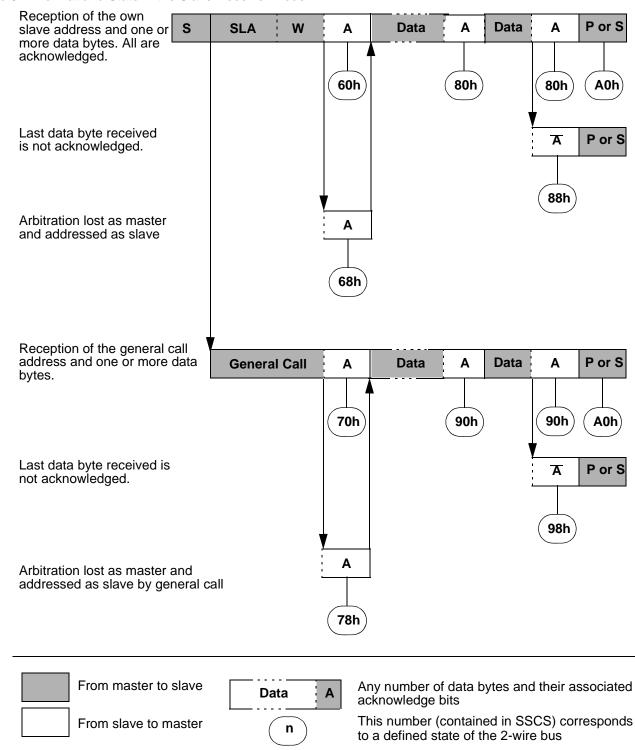




#### Table 103. Status in Master Receiver Mode

		Application software response					
Status Status of the Two- Code wire Bus and Two-				To SSC			
SSSTA	wire Hardware	To/From SSDAT	SSSTA	SSSTO	SSI	SSAA	Next Action Taken by Two-wire Hardware
08h	A START condition has been transmitted	Write SLA+R	х	0	0	x	SLA+R will be transmitted.
	A repeated START	Write SLA+R	х	0	0	х	SLA+R will be transmitted.
10h	condition has been transmitted	Write SLA+W	х	0	0	x	SLA+W will be transmitted. Logic will switch to master transmitter mode.
Arbitration lost in 38h SLA+R or NOT ACK bit		No SSDAT action	0	0	0	х	Two-wire bus will be released and not addressed slave mode will be entered.
		No SSDAT action	1	0	0	x	A START condition will be transmitted when the bus becomes free.
40h SLA+R has been transmitted; ACK has been received		No SSDAT action	0	0	0	0	Data byte will be received and NOT ACK will be returned.
		No SSDAT action	0	0	0	1	Data byte will be received and ACK will be returned.
SLA+R has been		No SSDAT action	1	0	0	x	Repeated START will be transmitted. STOP condition will be transmitted and SSSTO flag
48h	transmitted; NOT ACK	No SSDAT action	0	1	0	Х	will be reset.
has been received		No SSDAT action	1	1	0	х	STOP condition followed by a START condition will be transmitted and SSSTO flag will be reset.
50h	Data byte has been received; ACK has	Read data byte	0	0	0	0	Data byte will be received and NOT ACK will be returned.
	been returned	Read data byte	0	0	0	1	Data byte will be received and ACK will be returned.
	Data buta kao kao -	Read data byte	1	0	0	х	Repeated START will be transmitted.
58h	Data byte has been received; NOT ACK	Read data byte	0	1	0	Х	STOP condition will be transmitted and SSSTO flag will be reset.
	has been returned	Read data byte	1	1	0	x	STOP condition followed by a START condition will be transmitted and SSSTO flag will be reset.

Figure 54. Format and State in the Slave Receiver Mode





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	Table 104.	Status in	Slave	Receiver	Mode
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		Application S	Software	e Respo				
Status		To/from SSDAT		To SS	CON			
Code (SSCS)	Status of the 2-wire bus and 2-wire hardware		STA	STO	SI	AA	Next Action Taken By 2-wire Software	
Own SLA+W has been		No SSDAT action or	х	0	0	0	Data byte will be received and NOT ACK will be returned	
60h	received; ACK has been returned	No SSDAT action	х	0	0	1	Data byte will be received and ACK will be returned	
69h	Arbitration lost in SLA+R/W as master; own SLA+W has been	No SSDAT action or	х	0	0	0	Data byte will be received and NOT ACK will b returned	
68h	received; ACK has been returned	No SSDAT action	х	0	0	1	Data byte will be received and ACK will be returned	
70h	General call address has been received: ACK has been	No SSDAT action or	х	0	0	0	Data byte will be received and NOT ACK will b returned	
7011	returned	No SSDAT action	х	0	0	1	Data byte will be received and ACK will be returned	
Arbitration lost in SLA+R/W as master; general call address		No SSDAT action or	х	0	0	0	Data byte will be received and NOT ACK will b returned	
/8n	has been received; ACK has been returned	No SSDAT action	х	0	0	1	Data byte will be received and ACK will be returned	
0.01	Previously addressed with own SLA+W; data has been	No SSDAT action or	х	0	0	0	Data byte will be received and NOT ACK will b returned	
80h	received; ACK has been returned	No SSDAT action	х	0	0	1	Data byte will be received and ACK will be returned	
		Read data byte or	0	0	0	0	Switched to the not addressed slave mode; no recognition of own SLA or GCA Switched to the not addressed slave mode; ov	
		Read data byte or	0	0	0	1	SLA will be recognised; GCA will be recognised GC=logic 1	
XXN	Previously addressed with own SLA+W; data has been received; NOT ACK has been returned	Read data byte or	1	0	0	0	Switched to the not addressed slave mode; no recognition of own SLA or GCA. A START condition will be transmitted when the bus becomes free	
		Read data byte	1	0	0	1	Switched to the not addressed slave mode; ov SLA will be recognised; GCA will be recognised; GC=logic 1. A START condition will be transmitted when the bus becomes free	
90h	Previously addressed with general call; data has been	Read data byte or	х	0	0	0	Data byte will be received and NOT ACK will b returned	
90(1	received; ACK has been returned	Read data byte	х	0	0	1	Data byte will be received and ACK will be returned	

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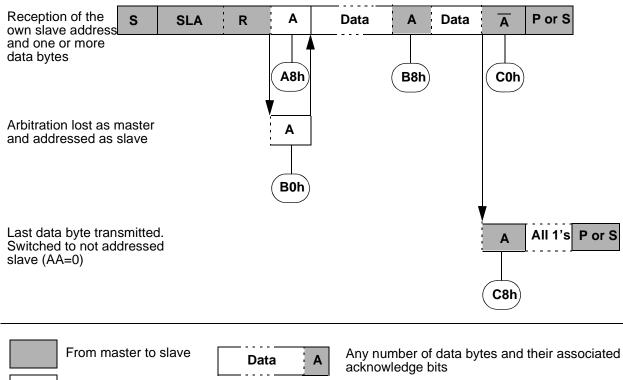
Table 104.	Status in	Slave Recei	ver Mode	(Continued)
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		Application Software Response						
Status		To/from SSDAT	To SSCON					
Code (SSCS)	Status of the 2-wire bus and 2-wire hardware		STA	ѕто	SI	АА	Next Action Taken By 2-wire Software	
		Read data byte or	0	0	0	0	Switched to the not addressed slave mode; no recognition of own SLA or GCA Switched to the not addressed slave mode; own	
		Read data byte or	0	0	0	1	SLA will be recognised; GCA will be recognised if GC=logic 1	
98h	Previously addressed with general call; data has been received; NOT ACK has been returned	Read data byte or	1	0	0	0	Switched to the not addressed slave mode; no recognition of own SLA or GCA. A START condition will be transmitted when the bus becomes free	
		Read data byte	1	0	0	1	Switched to the not addressed slave mode; own SLA will be recognised; GCA will be recognised if GC=logic 1. A START condition will be transmitted when the bus becomes free	
		No SSDAT action or	0	0	0	0	Switched to the not addressed slave mode; no recognition of own SLA or GCA Switched to the not addressed slave mode; own SLA will be recognised; GCA will be recognised if	
	A STOP condition or repeated START condition has been received while still addressed as slave	No SSDAT action or	0	0	0	1	GC=logic 1	
A0h		No SSDAT action or	1	0	0	0	Switched to the not addressed slave mode; no recognition of own SLA or GCA. A START condition will be transmitted when the bus becomes free	
		No SSDAT action	1	0	0	1	Switched to the not addressed slave mode; own SLA will be recognised; GCA will be recognised if GC=logic 1. A START condition will be transmitted when the bus becomes free	





#### Figure 55. Format and State in the Slave Transmitter Mode



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This number (contained in SSCS) corresponds to a defined state of the 2-wire bus

#### Table 105. Status in Slave Transmitter Mode

From slave to master

		Application Software Response			nse		
Status		To/from SSDAT To SSCON					
Code (SSCS)	Status of the 2-wire bus and 2-wire hardware		STA	ѕто	SI	АА	Next Action Taken By 2-wire Software
A 01-	Own SLA+R has been	Load data byte or	х	0	0	0	Last data byte will be transmitted and NOT ACK will be received
A8h	received; ACK has been returned	Load data byte	х	0	0	1	Data byte will be transmitted and ACK will be received
Doh	Arbitration lost in SLA+R/W as master; own SLA+R has been	Load data byte or	х	0	0	0	Last data byte will be transmitted and NOT ACK will be received
B0h	received; ACK has been returned	Load data byte	х	0	0	1	Data byte will be transmitted and ACK will be received
Doh	Data byte in SSDAT has been	Load data byte or	х	0	0	0	Last data byte will be transmitted and NOT ACK will be received
B8h	transmitted; NOT ACK has been received	Load data byte	х	0	0	1	Data byte will be transmitted and ACK will be received

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		Application S	Software	e Respo	nse		
Status		To/from SSDAT		To SS	CON		
Code (SSCS)	Status of the 2-wire bus and 2-wire hardware		STA	ѕто	SI	АА	Next Action Taken By 2-wire Software
		No SSDAT action or No SSDAT action or	0 0	0 0	0 0	0 1	Switched to the not addressed slave mode; no recognition of own SLA or GCA Switched to the not addressed slave mode; own SLA will be recognised; GCA will be recognised if GC=logic 1
C0h	C0h Data byte in SSDAT has been transmitted; NOT ACK has been received	No SSDAT action or	1	0	0	0	Switched to the not addressed slave mode; no recognition of own SLA or GCA. A START condition will be transmitted when the bus becomes free
		No SSDAT action	1	0	0	1	Switched to the not addressed slave mode; own SLA will be recognised; GCA will be recognised if GC=logic 1. A START condition will be transmitted when the bus becomes free
		No SSDAT action or	0	0	0	0	Switched to the not addressed slave mode; no recognition of own SLA or GCA Switched to the not addressed slave mode; own SLA will be recognised; GCA will be recognised if
C8h	Last data byte in SSDAT has been transmitted (AA=0); ACK has been received	No SSDAT action or	1	0	0	0	GC=logic 1 Switched to the not addressed slave mode; no recognition of own SLA or GCA. A START condition will be transmitted when the bus becomes free
		No SSDAT action	1	0	0	1	Switched to the not addressed slave mode; own SLA will be recognised; GCA will be recognised if GC=logic 1. A START condition will be transmitted when the bus becomes free

#### Table 105. Status in Slave Transmitter Mode (Continued)

#### Table 106. Miscellaneous Status

	Application Software R					se		
		To/from To SSCON						
Status Code (SSCS)	Status of the 2-wire bus and 2-wire hardware	SSDAT	STA	ѕто	SI	АА	Next Action Taken By 2-wire Software	
F8h	No relevant state information available; SI= 0	No SSDAT action	No SSCON action			ion	Wait or proceed current transfer	
00h	Bus error due to an illegal START or STOP condition	No SSDAT action	0	0 1 0 X		х	Only the internal hardware is affected, no STOP condition is sent on the bus. In all cases, the bus is released and STO is reset.	





### Registers

#### Table 107. SSCON Register

SSCON - Synchronous Serial Control register (93h)

7	6	5	4	3	2	1	0	
CR2	SSIE	STA	STO	SI	AA	CR1	CR0	
Bit Number	Bit Mnemonic	Description						
7	CR2	Control Rate See Table 10						
6	SSIE	Clear to disa	Synchronous Serial Interface Enable bit Clear to disable the TWI module. Set to enable the TWI module.					
5	STA	Start flag Set to send a	a START conc	lition on the b	us.			
4	ST0	Stop flag Set to send a	Stop flag Set to send a STOP condition on the bus.					
3	SI	Set by hardw	Synchronous Serial Interrupt flag Set by hardware when a serial interrupt is requested. Must be cleared by software to acknowledge interrupt.					
2	AA	Assert Acknowledge flag Clear in master and slave receiver modes, to force a not acknowledge (high level on SDA). Clear to disable SLA or GCA recognition. Set to recognise SLA or GCA (if GC set) for entering slave receiver or transmitter modes. Set in master and slave receiver modes, to force an acknowledge (low level on SDA). This bit has no effect when in master transmitter mode.						
1	CR1	Control Rate bit 1 See Table 101.						
0	CR0		Control Rate bit 0 See Table 101.					

#### Table 108. SSDAT (95h) - Syncrhonous Serial Data register (read/write)

SD7	SD6	SD5	SD4	SD3	SD2	SD1	SD0
7	6	5	4	3	2	1	0

Bit Number	Bit Mnemonic	Description
7 SD7 Address bit 7 or Data bit 7.		Address bit 7 or Data bit 7.
6	SD6	Address bit 6 or Data bit 6.
5	SD5	Address bit 5 or Data bit 5.
4	SD4	Address bit 4 or Data bit 4.
3	SD3	Address bit 3 or Data bit 3.
2	SD2	Address bit 2 or Data bit 2.

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Bit Number	Bit Mnemonic	Description
1	SD1	Address bit 1 or Data bit 1.
0	SD0	Address bit 0 (R/W) or Data bit 0.

Table 109. SSCS (094h) read - Synchronous Serial Control and Status Register

7	6	5	4	3	2	1	0
SC4	SC3	SC2	SC1	SC0	0	0	0

#### Table 110. SSCS Register: Read Mode - Reset Value = F8h

Bit Number	Bit Mnemonic	Description
0	0	Always zero
1	0	Always zero
2	0	Always zero
3	SC0	Status Code bit 0 See to Table 106.
4	SC1	Status Code bit 1 See to Table 106.
5	SC2	Status Code bit 2 See to Table 106.
6	SC3	Status Code bit 3 See to Table 106.
7	SC4	Status Code bit 4 See to Table 106.

Table 111. SSADR (096h) - Synchronus Serial Address Register (read/write)

7	6	5	4	3	2	1	0
A7	A6	A5	A4	A3	A2	A1	A0

#### Table 112. SSADR Register - Reset value = FEh

Bit Number	Bit Mnemonic	Description
7	A7	Slave Address bit 7
6	A6	Slave Address bit 6
5	A5	Slave Address bit 5
4	A4	Slave Address bit 4
3	A3	Slave Address bit 3
2	A2	Slave Address bit 2
1	A1	Slave Address bit 1





Bit Number	Bit Mnemonic	Description
0	GC	General Call bit Clear to disable the general call address recognition. Set to enable the general call address recognition.

### Serial Port Interface (SPI)

The Serial Peripheral Interface Module (SPI) allows full-duplex, synchronous, serial communication between the MCU and peripheral devices, including other MCUs.

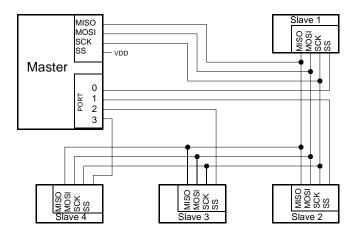
Features

#### Features of the SPI Module include the following:

- Full-duplex, three-wire synchronous transfers
- Master or Slave operation
- Six programmable Master clock rates in master mode
- Serial clock with programmable polarity and phase
- Master Mode fault error flag with MCU interrupt capability

**Signal Description** Figure 56 shows a typical SPI bus configuration using one Master controller and many Slave peripherals. The bus is made of three wires connecting all the devices.

#### Figure 56. SPI Master/Slaves Interconnection



The Master device selects the individual Slave devices by using four pins of a parallel port to control the four  $\overline{SS}$  pins of the Slave devices.

- Master Output Slave<br/>Input (MOSI)This 1-bit signal is directly connected between the Master Device and a Slave Device. The MOSI<br/>line is used to transfer data in series from the Master to the Slave. Therefore, it is an output sig-<br/>nal from the Master, and an input signal to a Slave. A Byte (8-bit word) is transmitted most<br/>significant bit (MSB) first, least significant bit (LSB) last.
- Master Input Slave<br/>Output (MISO)This 1-bit signal is directly connected between the Slave Device and a Master Device. The MISO<br/>line is used to transfer data in series from the Slave to the Master. Therefore, it is an output sig-<br/>nal from the Slave, and an input signal to the Master. A Byte (8-bit word) is transmitted most<br/>significant bit (MSB) first, least significant bit (LSB) last.
- **SPI Serial Clock (SCK)** This signal is used to synchronize the data transmission both in and out of the devices through their MOSI and MISO lines. It is driven by the Master for eight clock cycles which allows to exchange one Byte on the serial lines.
- Slave Select (SS) Each Slave peripheral is selected by one Slave Select pin (SS). This signal must stay low for any message for a Slave. It is obvious that only one Master (SS high level) can drive the network. The Master may select each Slave device by software through port pins (Figure 57). To prevent bus conflicts on the MISO line, only one slave should be selected at a time by the Master for a transmission.





In a Master configuration, the  $\overline{SS}$  line can be used in conjunction with the MODF flag in the SPI Status register (SPSCR) to prevent multiple masters from driving MOSI and SCK (see Error conditions).

A high level on the  $\overline{SS}$  pin puts the MISO line of a Slave SPI in a high-impedance state.

The  $\overline{SS}$  pin could be used as a general-purpose if the following conditions are met:

- The device is configured as a Master and the SSDIS control bit in SPCON is set. This kind
  of configuration can be found when only one Master is driving the network and there is no
  way that the SS pin could be pulled low. Therefore, the MODF flag in the SPSCR will never
  be set<sup>(1)</sup>.
- The Device is configured as a Slave with CPHA and SSDIS control bits set<sup>(2)</sup>. This kind of configuration can happen when the system includes one Master and one Slave only. Therefore, the device should always be selected and there is no reason that the Master uses the SS pin to select the communicating Slave device.
- Note: 1. Clearing SSDIS control bit does not clear MODF.
  - 2. Special care should be taken not to set SSDIS control bit when CPHA ='0' because in this mode, the  $\overline{SS}$  is used to start the transmission.

Baud RateIn Master mode, the baud rate can be selected from a baud rate generator which is controlled by<br/>three bits in the SPCON register: SPR2, SPR1 and SPR0. The Master clock is selected from one<br/>of seven clock rates resulting from the division of the internal clock by 4, 8, 16, 32, 64 or 128.

Table 113 gives the different clock rates selected by SPR2:SPR1:SPR0.

In Slave mode, the maximum baud rate allowed on the SCK input is limited to F<sub>svs</sub>/4

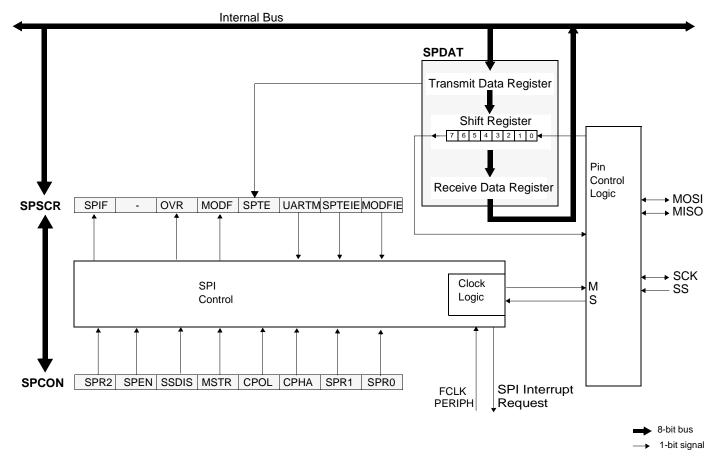
SPR2	SPR1	SPR0	Clock Rate	Baud Rate Divisor (BD)
0	0	0	Don't Use	No BRG
0	0	1	F <sub>CLK PERIPH</sub> /4	4
0	1	0	F <sub>CLK PERIPH</sub> /8	8
0	1	1	F <sub>CLK PERIPH</sub> /16	16
1	0	0	F <sub>CLK PERIPH</sub> /32	32
1	0	1	F <sub>CLK PERIPH</sub> /64	64
1	1	0	F <sub>CLK PERIPH</sub> /128	128
1	1	1	Don't Use	No BRG

Table 113. SPI Master Baud Rate Selection

#### Functional Description

Figure 57 shows a detailed structure of the SPI Module.

Figure 57. SPI Module Block Diagram



#### **Operating Modes**

The Serial Peripheral Interface can be configured in one of the two modes: Master mode or Slave mode. The configuration and initialization of the SPI Module is made through two registers:

- The Serial Peripheral Control register (SPCON)
- The Serial Peripheral Status and Control Register (SPSCR)

Once the SPI is configured, the data exchange is made using:

The Serial Peripheral DATa register (SPDAT)

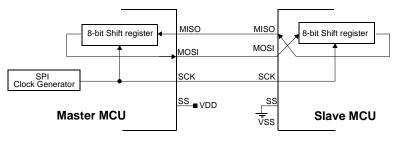
During an SPI transmission, data is simultaneously transmitted (shifted out serially) and received (shifted in serially). A serial clock line (SCK) synchronizes shifting and sampling on the two serial data lines (MOSI and MISO). A Slave Select line  $(\overline{SS})$  allows individual selection of a Slave SPI device; Slave devices that are not selected do not interfere with SPI bus activities.





When the Master device transmits data to the Slave device via the MOSI line, the Slave device responds by sending data to the Master device via the MISO line. This implies full-duplex transmission with both data out and data in synchronized with the same clock (Figure 58).

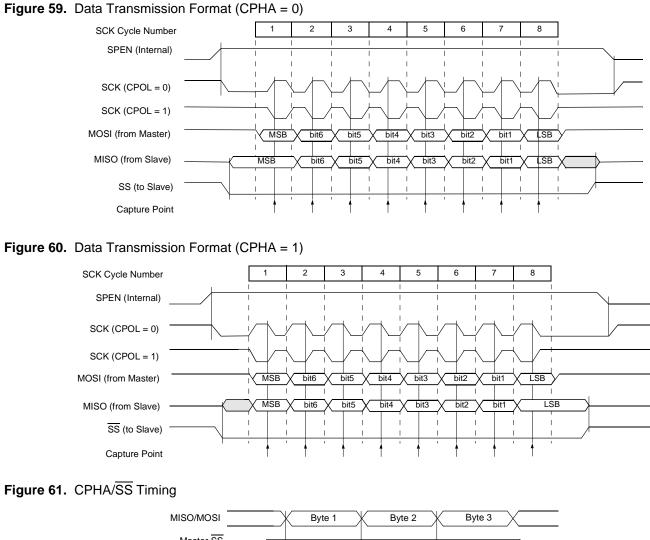
#### Figure 58. Full-Duplex Master-Slave Interconnection



- Master Mode The SPI operates in Master mode when the Master bit, MSTR<sup>(1)</sup>, in the SPCON register is set. Only one Master SPI device can initiate transmissions. Software begins the transmission from a Master SPI Module by writing to the Serial Peripheral Data Register (SPDAT). If the shift register is empty, the Byte is immediately transferred to the shift register. The Byte begins shifting out on MOSI pin under the control of the serial clock, SCK. Simultaneously, another Byte shifts in from the Slave on the Master's MISO pin. The transmission ends when the Serial Peripheral transfer data flag, SPIF, in SPSCR becomes set. At the same time that SPIF becomes set, the received Byte from the Slave is transferred to the receive data register in SPDAT. Software clears SPIF by reading the Serial Peripheral Status register (SPSCR) with the SPIF bit set, and then reading the SPDAT.
- *Slave Mode* The SPI operates in Slave mode when the Master bit, MSTR <sup>(2)</sup>, in the SPCON register is cleared. Before a data transmission occurs, the Slave Select pin, SS, of the Slave device must be set to'0'. SS must remain low until the transmission is complete.

In a Slave SPI Module, data enters the shift register under the control of the SCK from the Master SPI Module. After a Byte enters the shift register, it is immediately transferred to the receive data register in SPDAT, and the SPIF bit is set. To prevent an overflow condition, Slave software must then read the SPDAT before another Byte enters the shift register <sup>(3)</sup>. A Slave SPI must complete the write to the SPDAT (shift register) at least one bus cycle before the Master SPI starts a transmission. If the write to the data register is late, the SPI transmits the data already in the shift register from the previous transmission.

- **Transmission Formats** Software can select any of four combinations of serial clock (SCK) phase and polarity using two bits in the SPCON: the Clock Polarity (CPOL <sup>(4)</sup>) and the Clock Phase (CPHA<sup>4</sup>). CPOL defines the default SCK line level in idle state. It has no significant effect on the transmission format. CPHA defines the edges on which the input data are sampled and the edges on which the output data are shifted (Figure 59 and Figure 60). The clock phase and polarity should be identical for the Master SPI device and the communicating Slave device.
  - 1. The SPI Module should be configured as a Master before it is enabled (SPEN set). Also, the Master SPI should be configured before the Slave SPI.
  - 2. The SPI Module should be configured as a Slave before it is enabled (SPEN set).
  - 3. The maximum frequency of the SCK for an SPI configured as a Slave is the bus clock speed.
  - Before writing to the CPOL and CPHA bits, the SPI should be disabled (SPEN ='0').





As shown in Figure 59, the first SCK edge is the MSB capture strobe. The refore, the Slave must begin driving its data before the first SCK edge, and a falling edge on the SS pin is used to start the transmission. The SS pin must be toggled high and then low between each Byte transmitted (Figure 61).

Figure 60 shows an SPI transmission in which CPHA is'1'. In this case, the Master begins driving its MOSI pin on the first SCK edge. Therefore, the Slave uses the first SCK edge as a start transmission signal. The SS pin can remain low between transmissions (Figure 61). This format may be preferred in systems having only one Master and only one Slave driving the MISO data line.

*Queuing transmission* For an SPI configured in master or slave mode, a queued data byte must be transmitted/received immediately after the previous transmission has completed.



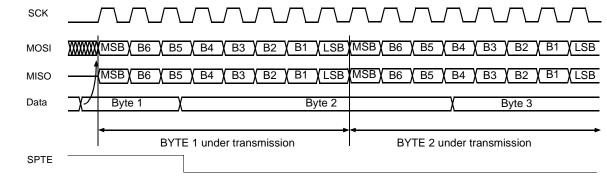


When a transmission is in progress a new data can be queued and sent as soon as transmission has been completed. So it is possible to transmit bytes without latency, useful in some applications.

The SPTE bit in SPSCR is set as long as the transmission buffer is free. It means that the user application can write SPDAT with the data to be transmitted until the SPTE becomes cleared.

Figure 62 shows a queuing transmission in master mode. Once the Byte 1 is ready, it is immediately sent on the bus. Meanwhile an other byte is prepared (and the SPTE is cleared), it will be sent at the end of the current transmission. The next data must be ready before the end of the current transmission.





In slave mode it is almost the same except it is the external master that start the transmission.

Also, in slave mode, if no new data is ready, the last value received will be the next data byte transmitted.

**Error Conditions** The following flags in the SPSCR register indicate the SPI error conditions:

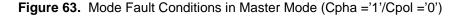
Mode Fault Error (MODF) Mode Fault error in Master mode SPI indicates that the level on the Slave Select (SS) pin is inconsistent with the actual mode of the device.

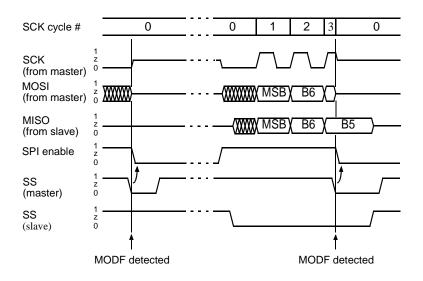
Mode fault detection in Master mode:

MODF is set to warn that there may be a multi-master conflict for system control. In this case, the SPI system is affected in the following ways:

- An SPI receiver/error CPU interrupt request is generated
- The SPEN bit in SPCON is cleared. This disables the SPI
- The MSTR bit in SPCON is cleared

Clearing the MODF bit is accomplished by a read of SPSCR register with MODF bit set, followed by a write to the SPCON register. SPEN Control bit may be restored to its original set state after the MODF bit has been cleared.





- Note: When SS is discarded (SS disabled) it is not possible to detect a MODF error in master mode because the SPI is internally unselected and the SS pin is a general purpose I/O.
- Mode fault detection in Slave mode

In slave mode, the MODF error is detected when SS goes high during a transmission.

A transmission begins when SS goes low and ends once the incoming SCK goes back to its idle level following the shift of the eighteen data bit.

A MODF error occurs if a slave is selected (SS is low) and later unselected (SS is high) even if no SCK is sent to that slave.

At any time, a'1' on the SS pin of a slave SPI puts the MISO pin in a high impedance state and internal state counter is cleared. Also, the slave SPI ignores all incoming SCK clocks, even if it was already in the middle of a transmission. A new transmission will be performed as soon as SS pin returns low.





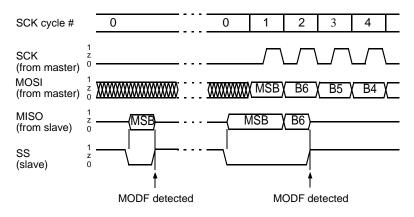


Figure 64. Mode Fault Conditions in Slave Mode

Note: when SS is discarded (SS disabled) it is not possible to detect a MODF error in slave mode because the SPI is internally selected. Also the SS pin becomes a general purpose I/O.

*OverRun Condition* This error mean that the speed is not adapted for the running application:

An OverRun condition occurs when a byte has been received whereas the previous one has not been read by the application yet.

The last byte (which generate the overrun error) does not overwrite the unread data so that it can still be read. Therefore, an overrun error always indicates the loss of data.

Interrupts Three SPI status flags can generate a CPU interrupt requests:

#### Table 114. SPI Interrupts

Flag	Request
SPIF (SPI data transfer)	SPI Transmitter Interrupt Request
MODF (Mode Fault)	SPI mode-fault Interrupt Request
SPTE (Transmit register empty)	SPI transmit register empty Interrupt Request

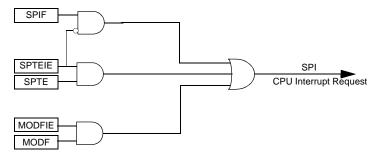
Serial Peripheral data transfer flag, SPIF: This bit is set by hardware when a transfer has been completed. SPIF bit generates transmitter CPU interrupt request only when SPTEIE is disabled.

Mode Fault flag, MODF: This bit is set to indicate that the level on the  $\overline{SS}$  is inconsistent with the mode of the SPI (in both master and slave modes).

Serial Peripheral Transmit Register empty flag, SPTE: This bit is set when the transmit buffer is empty (other data can be loaded is SPDAT). SPTE bit generates transmitter CPU interrupt request only when SPTEIE is enabled.

Note: While using SPTE interruption for "burst mode" transfers (SPTEIE='1'), the user software application should take care to clear SPTEIE, during the last but one data reception (to be able to generate an interrupt on SPIF flag at the end of the last data reception).

#### Figure 65. SPI Interrupt Requests Generation



Three registers in the SPI module provide control, status and data storage functions. These reg-

#### Registers

Serial Peripheral Control Register (SPCON)

- The Serial Peripheral Control Register does the following:
- Selects one of the Master clock rates

isters are describe in the following paragraphs.

- Configure the SPI Module as Master or Slave
- Selects serial clock polarity and phase
- Enables the SPI Module
- Frees the SS pin for a general-purpose

Table 115 describes this register and explains the use of each bit

#### Table 115. SPCON Register

SPCON - Serial Peripheral Control Register (0D4H)

7	6	5	<b>4 3 2 1 0</b>					
SPR2	SPEN	SSDIS	MSTR	CPOL	СРНА	SPR1	SPR0	
Bit Number	Bit Mn	emonic I	Description					
7	SF	PR2	Serial Peripheral Rate 2 Bit with SPR1 and SPR0 define the clock rate (See bits SPR1 and SPR0 fo detail).					
6	SF	PEN	Serial Peripheral Enable Cleared to disable the SPI interface (internal reset of the SPI). Set to enable the SPI interface.					
5	ss	SS Disable         Cleared to enable SS in both Master and Slave modes.         SSDIS         Set to disable SS in both Master and Slave modes. In Slave mode, has no effect if CPHA ='0'. When SSDIS is set, no MODF interrupt is generated.						
4	MS	STR (	Serial Peripheral Master Cleared to configure the SPI as a Slave. Set to configure the SPI as a Master.					





Bit Number	Bit Mnemonic	Descri	ption					
3	CPOL	Cleare	<b>Clock Polarity</b> Cleared to have the SCK set to'0' in idle state. Set to have the SCK set to'1' in idle state.					
2	СРНА	Cleare CPOL) Set to	Clock Phase Cleared to have the data sampled when the SCK leaves the idle state (see CPOL). Set to have the data sampled when the SCK returns to idle state (see CPOL).					
1	SPR1	<b>SPR2</b> 0 0 0	<b>SPR1</b> 0 0 1	<mark>SPR0</mark> 0 1 0	Invalid F <sub>CLK PERIPH</sub> /4			
0	SPR0	0 1 1 1 1	1 0 0 1 1	1 0 1 0 1	F <sub>CLK PERIPH</sub> /16 F <sub>CLK PERIPH</sub> /32 F <sub>CLK PERIPH</sub> /64 F <sub>CLK PERIPH</sub> /128 Invalid			

Reset Value = 0001 0100b

Not bit addressable

Serial Peripheral Status Register and Control (SPSCR) The Serial Peripheral Status Register contains flags to signal the following conditions:

- Data transfer complete
- Write collision
- Inconsistent logic level on SS pin (mode fault error)

#### Table 116. SPSCR Register

SPSCR - Serial Peripheral Status and Control register (C4H)

7	6	5	4	3	2	1	0		
SPIF	-	OVR	MODF	SPTE	UARTM	SPTEIE	MODFIE		
Bit Number	Bit Mnemonic	Description	Description						
7	SPIF	Cleared by har clearing seque Set by hardwa	Serial Peripheral Data Transfer Flag Cleared by hardware to indicate data transfer is in progress or has been approved by a clearing sequence. Set by hardware to indicate that the data transfer has been completed. This bit is cleared when reading or writing SPDATA after reading SPSCR.						
6	-	Reserved The value read from this bit is indeterminate. Do not set this bit.							
5	OVR	Overrun Error Flag - Set by hardware when a byte is received whereas SPIF is set (the previous received data is not overwritten). - Cleared by hardware when reading SPSCR							

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Bit Number	Bit Mnemonic	Description
4	MODF	Mode Fault         - Set by hardware to indicate that the SS pin is in inappropriate logic level (in both master and slave modes).         - Cleared by hardware when reading SPSCR         When MODF error occurred:         - In slave mode: SPI interface ignores all transmitted data while SS remains high. A new transmission is perform as soon as SS returns low.         - In master mode: SPI interface is disabled (SPEN=0, see description for SPEN bit in SPCON register).
3	SPTE	<ul> <li>Serial Peripheral Transmit register Empty</li> <li>Set by hardware when transmit register is empty (if needed, SPDAT can be loaded with another data).</li> <li>Cleared by hardware when transmit register is full (no more data should be loaded in SPDAT).</li> </ul>
2	UARTM	Serial Peripheral UART mode Set and cleared by software: - Clear: Normal mode, data are transmitted MSB first (default) - Set: UART mode, data are transmitted LSB first.
1	SPTEIE	Interrupt Enable for SPTE Set and cleared by software: - Set to enable SPTE interrupt generation (when SPTE goes high, an interrupt is generated). - Clear to disable SPTE interrupt generation Caution: When SPTEIE is set no interrupt generation occurred when SPIF flag goes high. To enable SPIF interrupt again, SPTEIE should be cleared.
0	MODFIE	Interrupt Enable for MODF Set and cleared by software: - Set to enable MODF interrupt generation - Clear to disable MODF interrupt generation

Reset Value = 00X0 XXXXb

Not Bit addressable

Serial Peripheral DATa Register (Table 117) is a read/write buffer for the receive data register (SPDAT) The Serial Peripheral Data Register (Table 117) is a read/write buffer for the receive data register. A write to SPDAT places data directly into the shift register. No transmit buffer is available in this model.

A Read of the SPDAT returns the value located in the receive buffer and not the content of the shift register.

Table 117. SPDAT Register

SPDAT - Serial Peripheral Data Register (C5H)

7	6	5	4	3	2	1	0
R7	R6	R5	R4	R3	R2	R1	R0

Reset Value = Indeterminate

R7:R0: Receive data bits





SPCON, SPSTA and SPDAT registers may be read and written at any time while there is no ongoing exchange. However, special care should be taken when writing to them while a transmission is on-going:

- Do not change SPR2, SPR1 and SPR0
- Do not change CPHA and CPOL
- Do not change MSTR
- Clearing SPEN would immediately disable the peripheral
- Writing to the SPDAT will cause an overflow.

# AT89C51RE2

# **Power Monitor** The POR/PFD function monitors the internal power-supply of the CPU core memories and the peripherals, and if needed, suspends their activity when the internal power supply falls below a safety threshold. This is achieved by applying an internal reset to them.

By generating the Reset the Power Monitor insures a correct start up when AT89C51RE2 is powered up.

**Description** In order to startup and maintain the microcontroller in correct operating mode, V<sub>CC</sub> has to be stabilized in the V<sub>CC</sub> operating range and the oscillator has to be stabilized with a nominal amplitude compatible with logic level VIH/VIL.

These parameters are controlled during the three phases: power-up, normal operation and power going down. See Figure 66.

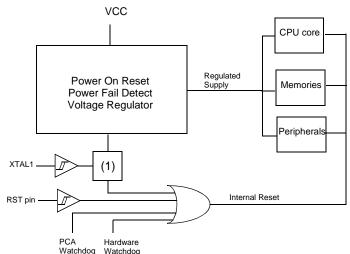


Figure 66. Power Monitor Block Diagram

Note: 1. Once XTAL1 High and low levels reach above and below VIH/VIL. a 1024 clock period delay will extend the reset coming from the Power Fail Detect. If the power falls below the Power Fail Detect threshold level, the Reset will be applied immediately.

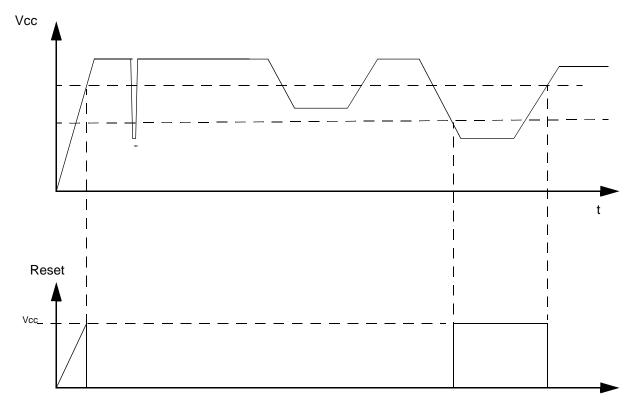
The Voltage regulator generates a regulated internal supply for the CPU core the memories and the peripherals. Spikes on the external Vcc are smoothed by the voltage regulator.





The Power fail detect monitor the supply generated by the voltage regulator and generate a reset if this supply falls below a safety threshold as illustrated in the Figure 67 below.

Figure 67. Power Fail Detect



When the power is applied, the Power Monitor immediately asserts a reset. Once the internal supply after the voltage regulator reach a safety level, the power monitor then looks at the XTAL clock input. The internal reset will remain asserted until the Xtal1 levels are above and below VIH and VIL. Further more. An internal counter will count 1024 clock periods before the reset is de-asserted.

If the internal power supply falls below a safety level, a reset is immediately asserted.

### **Power-off Flag**

The power-off flag allows the user to distinguish between a "cold start" reset and a "warm start" reset.

A cold start reset is the one induced by  $V_{CC}$  switch-on. A warm start reset occurs while  $V_{CC}$  is still applied to the device and could be generated for example by an exit from power-down.

The power-off flag (POF) is located in PCON register (Table 118). POF is set by hardware when  $V_{CC}$  rises from 0 to its nominal voltage. The POF can be set or cleared by software allowing the user to determine the type of reset.

#### Table 118. PCON Register

PCON - Power Control Register (87h)

7	6	5	4	3	2	1	0	
SMOD1	SMOD0	-	POF	GF1	GF0	PD	IDL	
Bit Number	Bit Mnemonic	Description	vescription					
7	SMOD1	Serial port Me Set to s		aud rate in mod	de 1, 2 or 3.			
6	SMOD0	Cleared to sel	erial port Mode bit 0 eared to select SM0 bit in SCON register. et to select FE bit in SCON register.					
5	-	Reserved The value rea	Reserved The value read from this bit is indeterminate. Do not set this bit.					
4	POF	Cleared to rec	<b>Power-Off Flag</b> Cleared to recognize next reset type. Set by hardware when V <sub>CC</sub> rises from 0 to its nominal voltage. Can also be set by software.					
3	GF1	Cleared by us	<b>General purpose Flag</b> Cleared by user for general purpose usage. Set by user for general purpose usage.					
2	GF0	Cleared by us	<b>General purpose Flag</b> Cleared by user for general purpose usage. Set by user for general purpose usage.					
1	PD	Cleared by ha	Power-Down mode bit Cleared by hardware when reset occurs. Set to enter power-down mode.					
0	IDL							

Reset Value = 00X1 0000b Not bit addressable

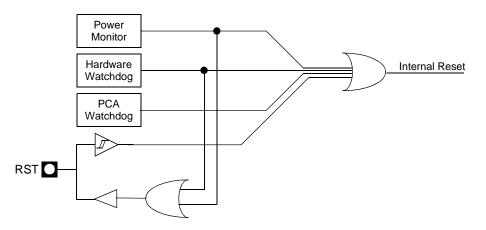




### Reset

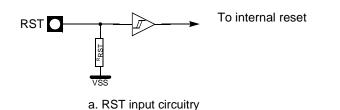
Introduction The reset sources are: Power Management, Hardware Watchdog, PCA Watchdog and Reset input.

Figure 68. Reset schematic

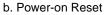


**Reset Input** The Reset input can be used to force a reset pulse longer than the internal reset controlled by the Power Monitor. RST input has a pull-down resistor allowing power-on reset by simply connecting an external capacitor to V<sub>CC</sub> as shown in Figure 69. Resistor value and input characteristics are discussed in the Section "DC Characteristics" of the AT89C51RE2 datasheet.

Figure 69. Reset Circuitry and Power-On Reset

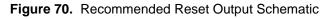


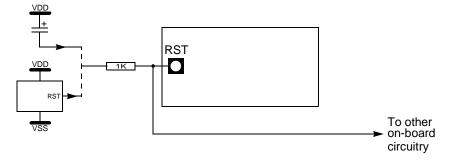




#### **Reset Output**

As detailed in Section "Hardware Watchdog Timer", page 124, the WDT generates a 96-clock period pulse on the RST pin. In order to properly propagate this pulse to the rest of the application in case of external capacitor or power-supply supervisor circuit, a 1 k $\Omega$  resistor must be added as shown Figure 70.









### **Electrical Characteristics**

### **Absolute Maximum Ratings**

I = industrial	40°C to 85°C
Storage Temperature	65°C to + 150°C
Voltage on V <sub>CC</sub> to V <sub>SS</sub> (standard voltage	e)0.5V to + 6.5V
Voltage on V <sub>CC</sub> to V <sub>SS</sub> (low voltage)	0.5V to + 4.5V
Voltage on Any Pin to V <sub>SS</sub>	0.5V to V <sub>CC</sub> + 0.5V
Power Dissipation	1 W <sup>(2)</sup>

Note: Stresses at or above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions may affect device reliability.

Power dissipation is based on the maximum allowable die temperature and the thermal resistance of the package.

### **DC Parameters**

Symbol	Parameter	Min	Тур	Max	Unit	Test Conditions
V <sub>IL</sub>	Input Low Voltage	-0.5		0.2 V <sub>CC</sub> - 0.1	V	
V <sub>IH</sub>	Input High Voltage except RST, XTAL1	0.2 V <sub>CC</sub> + 0.9		V <sub>CC</sub> + 0.5	V	
V <sub>IH1</sub>	Input High Voltage RST, XTAL1	0.7 V <sub>CC</sub>		V <sub>CC</sub> + 0.5	V	
V <sub>OL</sub>	Output Low Voltage, ports 1, 2, 3, 4 <sup>(6)</sup>			0.3 0.45 1.0	v v v	$\begin{split} V_{CC} &= 4.5V \text{ to } 5.5V \\ I_{OL} &= 100 \ \mu A^{(4)} \\ I_{OL} &= 1.6 \ m A^{(4)} \\ I_{OL} &= 3.5 \ m A^{(4)} \end{split}$
				0.45	v	$V_{CC} = 2.7 V \text{ to } 5.5 V$ $I_{OL} = 0.8 \text{ mA}^{(4)}$
V <sub>OL1</sub>	Output Low Voltage, port 0, ALE, PSEN <sup>(6)</sup>			0.3 0.45 1.0	v v v	$\begin{split} V_{CC} &= 4.5 V \text{ to } 5.5 V \\ I_{OL} &= 200 \ \mu A^{(4)} \\ I_{OL} &= 3.2 \ m A^{(4)} \\ I_{OL} &= 7.0 \ m A^{(4)} \end{split}$
				0.45	v	$V_{CC} = 2.7V \text{ to } 5.5V$ $I_{OL} = 1.6 \text{ mA}^{(4)}$
V <sub>OH</sub>	Output High Voltage, ports 1, 2, 3, 4	V <sub>CC</sub> - 0.3 V <sub>CC</sub> - 0.7 V <sub>CC</sub> - 1.5			V V V	$\begin{split} V_{CC} &= 5V \pm 10\% \\ I_{OH} &= -10 \; \mu A \\ I_{OH} &= -30 \; \mu A \\ I_{OH} &= -60 \; \mu A \end{split}$
		0.9 V <sub>CC</sub>			V	$V_{CC} = 2.7V \text{ to } 5.5V$ $I_{OH} = -10 \ \mu\text{A}$
V <sub>OH1</sub>	Output High Voltage, port 0, ALE, PSEN	V <sub>CC</sub> - 0.3 V <sub>CC</sub> - 0.7 V <sub>CC</sub> - 1.5			V V V	$V_{CC} = 5V \pm 10\%$ $I_{OH} = -200 \ \mu A$ $I_{OH} = -3.2 \ m A$ $I_{OH} = -7.0 \ m A$
		0.9 V <sub>CC</sub>			V	$V_{CC}$ = 2.7V to 5.5V $I_{OH}$ = -10 $\mu$ A

Symbol	Parameter	Min	Тур	Мах	Unit	Test Conditions
R <sub>RST</sub>	RST Pull-down Resistor	50	200 <sup>(5)</sup>	250	kΩ	
I <sub>IL</sub>	Logical 0 Input Current ports 1, 2, 3, 4 and 5			-50	μA	V <sub>IN</sub> = 0.45V
I <sub>LI</sub>	Input Leakage Current			±10	μA	$0.45V < V_{IN} < V_{CC}$
I <sub>TL</sub>	Logical 1 to 0 Transition Current, ports 1, 2, 3, 4			-650	μA	V <sub>IN</sub> = 2.0V
C <sub>IO</sub>	Capacitance of I/O Buffer			10	pF	$F_{C} = 3 \text{ MHz}$ $T_{A} = 25^{\circ}\text{C}$
I <sub>PD</sub>	Power-down Current		75	150	μA	$2.7 < V_{CC} < 5.5 V^{(3)}$
I <sub>CCOP</sub>	Power Supply Current on normal mode			0.4 x Frequency (MHz) + 5	mA	$V_{\rm CC} = 5.5 V^{(1)}$
	Power Supply Current on idle mode			0.3 x Frequency (MHz) + 5	mA	$V_{\rm CC} = 5.5 V^{(2)}$
ICCWRITE	Power Supply Current on flash write			0.8 x Frequency (MHz) + 15	mA	$V_{CC} = 5.5V$
t <sub>WRITE</sub>	Flash programming time		7	10	ms	2.7 < V <sub>CC &lt;</sub> 5.5V

 $T_A = -40^{\circ}C$  to  $+85^{\circ}C$ ;  $V_{SS} = 0V$ ;  $V_{CC} = 2.7V$  to 5.5V; F = 0 to 40 MHz (Continued)

Notes: 1. Operating I<sub>CC</sub> is measured with all output pins disconnected; XTAL1 driven with T<sub>CLCH</sub>, T<sub>CHCL</sub> = 5 ns (see Figure 74), V<sub>IL</sub> = V<sub>SS</sub> + 0.5V, V<sub>IH</sub> = V<sub>CC</sub> - 0.5V; XTAL2 N.C.; EA = RST = Port 0 = V<sub>CC</sub>. I<sub>CC</sub> would be slightly higher if a crystal oscillator used (see Figure 71).

2. Idle I<sub>CC</sub> is measured with all output pins disconnected; XTAL1 driven with T<sub>CLCH</sub>, T<sub>CHCL</sub> = 5 ns, V<sub>IL</sub> = V<sub>SS</sub> + 0.5V, V<sub>IH</sub> = V<sub>CC</sub> - 0.5V; XTAL2 N.C; Port 0 = V<sub>CC</sub>;  $\overline{EA}$  = RST = V<sub>SS</sub> (see Figure 72).

Power-down I<sub>CC</sub> is measured with all output pins disconnected; EA = V<sub>CC</sub>, PORT 0 = V<sub>CC</sub>; XTAL2 NC.; RST = V<sub>SS</sub> (see Figure 73).

4. Capacitance loading on Ports 0 and 2 may cause spurious noise pulses to be superimposed on the V<sub>OLS</sub> of ALE and Ports 1 and 3. The noise is due to external bus capacitance discharging into the Port 0 and Port 2 pins when these pins make 1 to 0 transitions during bus operation. In the worst cases (capacitive loading 100 pF), the noise pulse on the ALE line may exceed 0.45V with maxi V<sub>OL</sub> peak 0.6V. A Schmitt Trigger use is not necessary.

5. Typical values are based on a limited number of samples and are not guaranteed. The values listed are at room temperature and 5V.

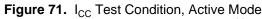
 Under steady state (non-transient) conditions, I<sub>OL</sub> must be externally limited as follows: Maximum I<sub>OL</sub> per port pin: 10 mA

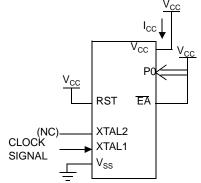
Maximum I<sub>OL</sub> per 8-bit port:

Port 0: 26 mA Ports 1, 2 and 3: 15 mA

Maximum total  $I_{OL}$  for all output pins: 71 mA

If  $I_{OL}$  exceeds the test condition,  $V_{OL}$  may exceed the related specification. Pins are not guaranteed to sink current greater than the listed test conditions.



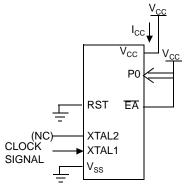


All other pins are disconnected.



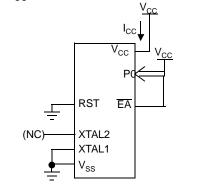


Figure 72. I<sub>CC</sub> Test Condition, Idle Mode

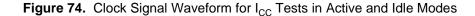


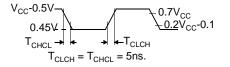
All other pins are disconnected.





All other pins are disconnected.





#### **AC Parameters**

**Explanation of the AC** Symbols Each timing symbol has 5 characters. The first character is always a "T" (stands for time). The other characters, depending on their positions, stand for the name of a signal or the logical status of that signal. The following is a list of all the characters and what they stand for.

 $\begin{array}{l} \mbox{Example:} T_{AVLL} = \mbox{Time for Address Valid to ALE Low.} \\ T_{LLPL} = \mbox{Time for ALE Low to PSEN Low.} \end{array}$ 

(Load Capacitance for port 0, ALE and PSEN = 100 pF; Load Capacitance for all other outputs = 80 pF.)

Table 119 Table 122, and Table 125 give the description of each AC symbols.

Table 120, Table 121, Table 123 and Table 126 gives the range for each AC parameter.

Table 120, Table 121 and Table 127 give the frequency derating formula of the AC parameter for each speed range description. To calculate each AC symbols. Take the x value in the corresponding column (-M or -L) and use this value in the formula.

Example:  $T_{LLIU}$  for -M and 20 MHz, Standard clock. x = 35 ns T 50 ns  $T_{CCIV}$  = 4T - x = 165 ns

External Program Memory Characteristics Table 119. Symbol Description

Symbol	Parameter
Т	Oscillator clock period
T <sub>LHLL</sub>	ALE pulse width
T <sub>AVLL</sub>	Address Valid to ALE
T <sub>LLAX</sub>	Address Hold After ALE
T <sub>LLIV</sub>	ALE to Valid Instruction In
T <sub>LLPL</sub>	ALE to PSEN
T <sub>PLPH</sub>	PSEN Pulse Width
T <sub>PLIV</sub>	PSEN to Valid Instruction In
T <sub>PXIX</sub>	Input Instruction Hold After PSEN
T <sub>PXIZ</sub>	Input Instruction Float After PSEN
T <sub>AVIV</sub>	Address to Valid Instruction In
T <sub>PLAZ</sub>	PSEN Low to Address Float

Table 120. AC Parameters for a Fix Clock

Symbol	-N	(1)	-L	Units	
	Min	Max	Min	Max	
Т	25		25		ns
T <sub>LHLL</sub>	35		35		ns
T <sub>AVLL</sub>	5		5		ns
T <sub>LLAX</sub>	5		5		ns
T <sub>LLIV</sub>		n 65		65	ns
T <sub>LLPL</sub>	5		5		ns
T <sub>PLPH</sub>	50		50		ns
T <sub>PLIV</sub>		30		30	ns
T <sub>PXIX</sub>	0		0		ns
T <sub>PXIZ</sub>		10		10	ns
T <sub>AVIV</sub>		80		80	ns
T <sub>PLAZ</sub>		10		10	ns

Notes: 1. '-L ' refers to 2V - 5.5V version.

2. '-M' refers to 4.5V to 5.5V version.





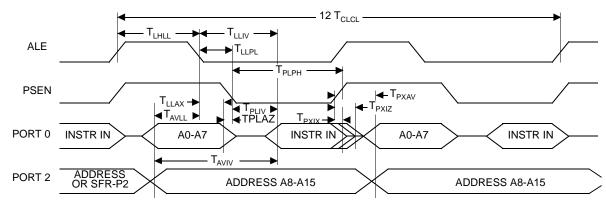
Symbol	Туре	Standard Clock	X2 Clock	X parameter for -M <sup>(1)</sup> range	X parameter for -L <sup>(2)</sup> range	Units
T <sub>LHLL</sub>	Min	2 T - x	T - x	15	15	ns
T <sub>AVLL</sub>	Min	T - x	0.5 T - x	20	20	ns
T <sub>LLAX</sub>	Min	T - x	0.5 T - x	20	20	ns
T <sub>LLIV</sub>	Max	4 T - x	2 T - x	35	35	ns
T <sub>LLPL</sub>	Min	T - x	0.5 T - x	15	15	ns
T <sub>PLPH</sub>	Min	3 T - x	1.5 T - x	25	25	ns
T <sub>PLIV</sub>	Max	3 T - x	1.5 T - x	45	45	ns
T <sub>PXIX</sub>	Min	х	х	0	0	ns
T <sub>PXIZ</sub>	Max	T - x	0.5 T - x	15	15	ns
T <sub>AVIV</sub>	Max	5 T - x	2.5 T - x	45	45	ns
T <sub>PLAZ</sub>	Max	х	х	10	10	ns

#### Table 121. AC Parameters for a Variable Clock

Notes: 1. '-L ' refers to 2V - 5.5V version.

2. '-M' refers to 4.5V to 5.5V version.

#### External Program Memory Read Cycle



#### External Data Memory Characteristics

 Table 122.
 Symbol Description

Symbol	Parameter
T <sub>RLRH</sub>	RD Pulse Width
T <sub>WLWH</sub>	WR Pulse Width
T <sub>RLDV</sub>	RD to Valid Data In
T <sub>RHDX</sub>	Data Hold After RD
T <sub>RHDZ</sub>	Data Float After RD
T <sub>LLDV</sub>	ALE to Valid Data In
T <sub>AVDV</sub>	Address to Valid Data In
T <sub>LLWL</sub>	ALE to WR or RD
T <sub>AVWL</sub>	Address to WR or RD
T <sub>QVWX</sub>	Data Valid to WR Transition
T <sub>QVWH</sub>	Data Set-up to WR High
T <sub>WHQX</sub>	Data Hold After WR
T <sub>RLAZ</sub>	RD Low to Address Float
T <sub>WHLH</sub>	RD or WR High to ALE high





Table 123.	AC Parameters for a Fix Clock
------------	-------------------------------

	-M <sup>(1)</sup>			-L <sup>(2)</sup>	
Symbol	Min	Мах	Min	Мах	Units
T <sub>RLRH</sub>	125		125		ns
T <sub>WLWH</sub>	125		125		ns
T <sub>RLDV</sub>		95		95	ns
T <sub>RHDX</sub>	0		0		ns
T <sub>RHDZ</sub>		25		25	ns
T <sub>LLDV</sub>		155		155	ns
T <sub>AVDV</sub>		160		160	ns
T <sub>LLWL</sub>	45	105	45	105	ns
T <sub>AVWL</sub>	70		70		ns
T <sub>QVWX</sub>	5		5		ns
T <sub>QVWH</sub>	155		155		ns
T <sub>WHQX</sub>	10		10		ns
T <sub>RLAZ</sub>	0		0		ns
T <sub>WHLH</sub>	5	45	5	45	ns

Notes: 1. '-L ' refers to 2V - 5.5V version. 2. '-M ' refers to 4.5V to 5.5V version.

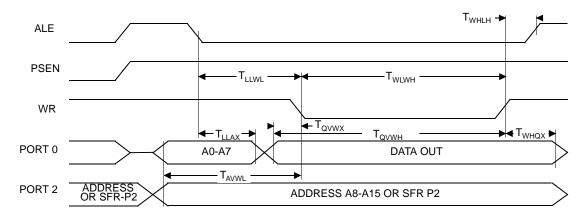
Symbol	Туре	Standard Clock	X2 Clock	X parameter for -M <sup>(1)</sup> range	X parameter for -L <sup>(2)</sup> range	Units
T <sub>RLRH</sub>	Min	6 T - x	3 T - x	25	25	ns
T <sub>WLWH</sub>	Min	6 T - x	3 T - x	25	25	ns
T <sub>RLDV</sub>	Max	5 T - x	2.5 T - x	30	30	ns
T <sub>RHDX</sub>	Min	х	х	0	0	ns
T <sub>RHDZ</sub>	Max	2 T - x	T - x	25	25	ns
T <sub>LLDV</sub>	Max	8 T - x	4T -x	45	45	ns
T <sub>AVDV</sub>	Max	9 T - x	4.5 T - x	65	65	ns
T <sub>LLWL</sub>	Min	3 T - x	1.5 T - x	30	30	ns
T <sub>LLWL</sub>	Max	3 T + x	1.5 T + x	30	30	ns
T <sub>AVWL</sub>	Min	4 T - x	2 T - x	30	30	ns
T <sub>QVWX</sub>	Min	T - x	0.5 T - x	20	20	ns
T <sub>QVWH</sub>	Min	7 T - x	3.5 T - x	20	20	ns
T <sub>WHQX</sub>	Min	T - x	0.5 T - x	15	15	ns
T <sub>RLAZ</sub>	Max	х	х	0	0	ns
T <sub>WHLH</sub>	Min	T - x	0.5 T - x	20	20	ns
T <sub>WHLH</sub>	Max	T + x	0.5 T + x	20	20	ns

Table 124	<b>AC</b> Parameters	for a	Variahle	Clock
	AC Falameters	iui a	valiable	

Notes: 1. '-L' refers to 2V - 5.5V version.

2. '-M' refers to 4.5V to 5.5V version.

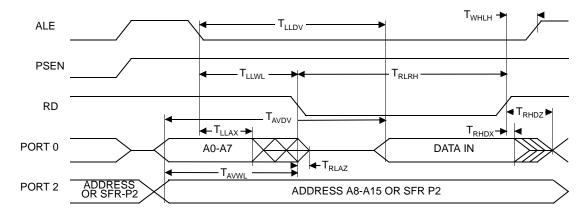
#### External Data Memory Write Cycle







#### **External Data Memory Read Cycle**



#### Serial Port Timing -Shift Register Mode

#### Table 125. Symbol Description

Symbol	Parameter
T <sub>XLXL</sub>	Serial port clock cycle time
T <sub>QVHX</sub>	Output data set-up to clock rising edge
T <sub>XHQX</sub>	Output data hold after clock rising edge
T <sub>XHDX</sub>	Input data hold after clock rising edge
T <sub>XHDV</sub>	Clock rising edge to input data valid

#### Table 126. AC Parameters for a Fix Clock

-M <sup>(</sup>		(1)		(2)	
Symbol	Min	Мах	Min	Мах	Units
T <sub>XLXL</sub>	300		300		ns
T <sub>QVHX</sub>	200		200		ns
T <sub>XHQX</sub>	30		30		ns
T <sub>XHDX</sub>	0		0		ns
T <sub>XHDV</sub>		117		117	ns

Notes: 1. '-L ' refers to 2V - 5.5V version.

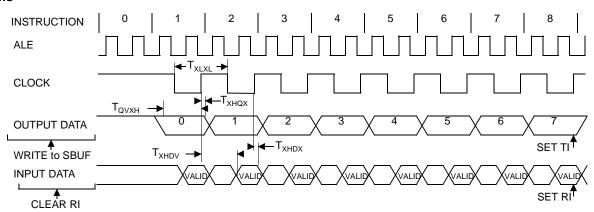
2. '-M' refers to 4.5V to 5.5V version.

Symbol	Туре	Standard Clock	X2 Clock	X Parameter For -M <sup>(1)</sup> Range	X Parameter For -L <sup>(2)</sup> Range	Units
T <sub>XLXL</sub>	Min	12 T	6 T			ns
T <sub>QVHX</sub>	Min	10 T - x	5 T - x	50	50	ns
T <sub>XHQX</sub>	Min	2 T - x	Т - х	20	20	ns
T <sub>XHDX</sub>	Min	х	х	0	0	ns
T <sub>XHDV</sub>	Max	10 T - x	5 T- x	133	133	ns

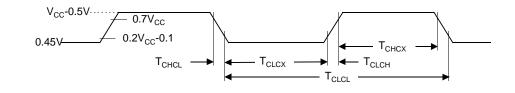
Notes: 1. '-L ' refers to 2V - 5.5V version.

2. '-M' refers to 4.5V to 5.5V version.

# Shift Register Timing Waveforms

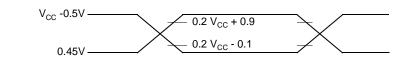


# External Clock Drive Waveforms



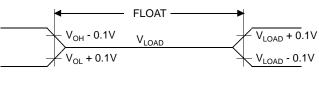
AC Testing Input/Output Waveforms

INPUT/OUTPUT



AC inputs during testing are driven at V<sub>CC</sub> - 0.5 for a logic "1" and 0.45V for a logic "0". Timing measurement are made at V<sub>IH</sub> min. for a logic "1" and V<sub>IL</sub> max for a logic "0".

**Float Waveforms** 

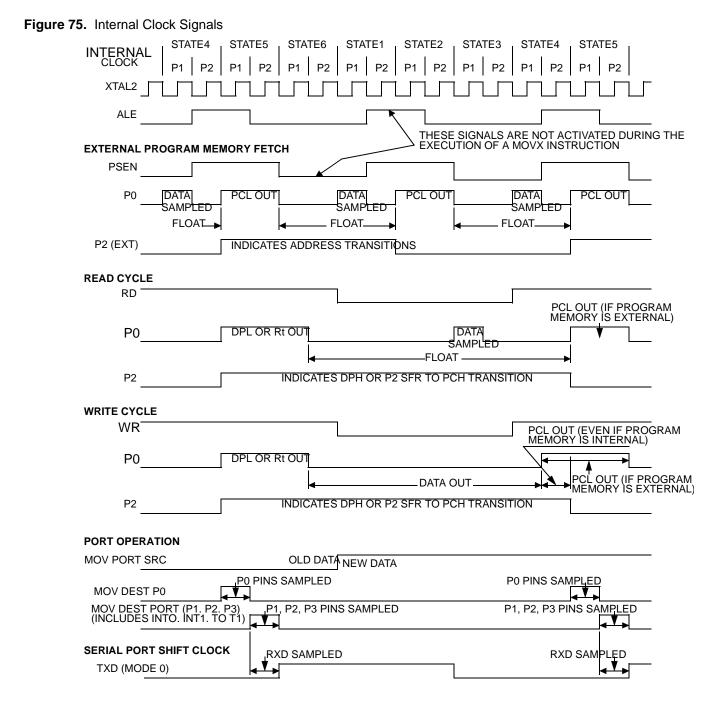






For timing purposes as port pin is no longer floating when a 100 mV change from load voltage occurs and begins to float when a 100 mV change from the loaded V<sub>OH</sub>/V<sub>OL</sub> level occurs. I<sub>OL</sub>/I<sub>OH</sub>  $\geq \pm$  20 mA.

Clock Waveforms Valid in normal clock mode. In X2 mode XTAL2 must be changed to XTAL2/2.



This diagram indicates when signals are clocked internally. The time it takes the signals to propagate to the pins, however, ranges from 25 to 125 ns. This propagation delay is dependent on variables such as temperature and pin loading. Propagation also varies from output to output and component. Typically though ( $T_A = 25^{\circ}C$  fully loaded) RD and WR propagation delays are approximately 50 ns. The other signals are typically 85 ns. Propagation delays are incorporated in the AC specifications.

#### **Flash Memory**

#### Table 128. Timing Symbol Definitions

Signals					
S (Hardware condition)	PSEN#,EA				
R	RST				
В	FBUSY flag				

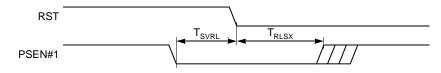
Conditions		
L	Low	
V	Valid	
X	No Longer Valid	

#### Table 129. Memory AC Timing

VDD = 3V to 5.5V, TA = -40 to +85°C

Symbol	Parameter	Min	Тур	Max	Unit
T <sub>SVRL</sub>	Input PSEN# Valid to RST Edge	50			ns
T <sub>RLSX</sub>	Input PSEN# Hold after RST Edge	50			ns
T <sub>BHBL</sub>	Flash Internal Busy (Programming) Time		10		ms
N <sub>FCY</sub>	Number of Flash Erase/Write Cycles	100 000			cycles
T <sub>FDR</sub>	Flash Retention Time	10			years

#### Figure 76. Flash Memory – ISP Waveforms



#### Figure 77. Flash Memory – Internal Busy Waveforms







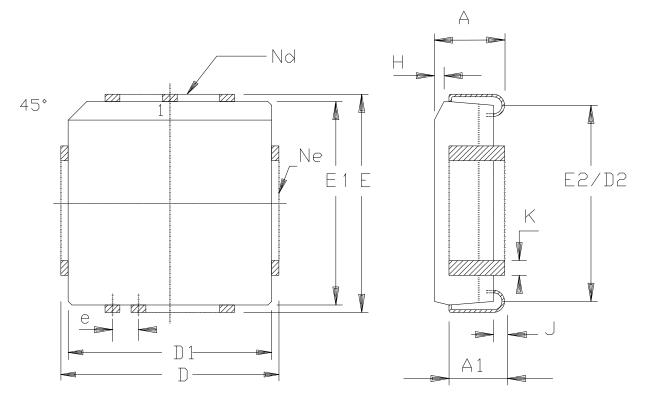
# **Ordering Information**

Table 130.	Possible Order	- Entries
	1 0001010 01001	

Part Number	Supply Voltage	Temperature Range	Package
AT89C51RE2-SLSUM	2.7V-5.5V	Industrial & Green	PLCC44
AT89C51RE2-RLTUM			VQFP44
AT89C51RE2-SLSEM	2.7V-5.5V	Engineering Samples	PLCC44
AT89C51RE2-RLTEM			VQFP44

# **Packaging Information**

## PLCC44



	MM		IN	СН
A	4, 20	4. 57	, 165	, 180
A1	2, 29	3. 04	, 090	. 120
D	17.40	17,65	, 685	, 695
D 1	16.44	16, 66	, 647	, 656
D2	14.99	16.00	. 590	, 630
E	17.40	17.65	, 685	, 695
E1	16.44	16, 66	, 647	, 656
E2	14.99	16.00	. 590	, 630
e	1, 27	BSC	, 050	BSC
Н	1.07	1. 42	. 042	, 056
J	0, 51	-	, 020	Ι
К	0, 33	0, 53	. 013	. 021
Nd	1 1		1	1
Ne	1 1		1	1
P	KG STD	00		





#### **STANDARD NOTES FOR PLCC**

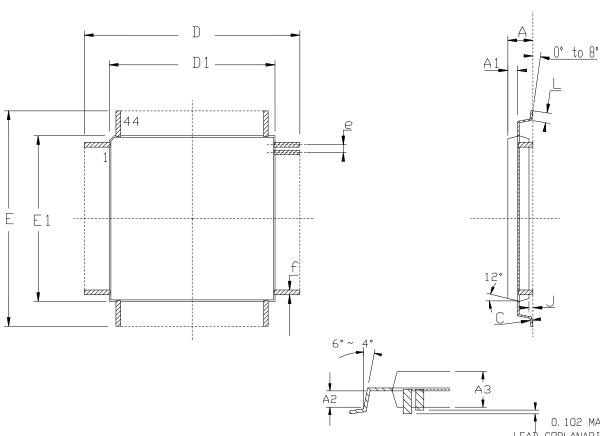
#### 1/ CONTROLLING DIMENSIONS : INCHES

#### 2/ DIMENSIONING AND TOLERANCING PER ANSI Y 14.5M - 1982.

3/ "D" AND "E1" DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTUSIONS. MOLD FLASH OR PROTUSIONS SHALL NOT EXCEED 0.20 mm (.008 INCH) PER SIDE.

## 180 AT89C51RE2

#### VQFP44



44 PINS SHRINK QUAD FLAT PACK SQUARE GULL WING (1, 4 mm)

> 0.102 MAX. LEAD COPLANARITY

	М	M	IN	СН	
	Min	Ma×	Min	Max	
A	_	1,60	_	, 063	
A 1	0.64 REF		. 0	25 REF	
A2	0, 64 REF		. 0	, 025 REF	
A3	1.35	1.45	, 053	. 057	
D	11.90	12.10	, 468	. 476	
D 1	9, 90	10.10	, 390	. 398	
E	11,90	12.10	, 468	, 476	
E 1	9, 90	10.10	, 390	. 398	
J	0.05	_	, 002	_	
L	0, 45	0, 75	, 018	, 030	





#### STANDARD NOTES FOR PQFP/ VQFP / TQFP / DQFP

1/ CONTROLLING DIMENSIONS : INCHES

2/ ALL DIMENSIONING AND TOLERANCING CONFORM TO ANSI Y 14.5M - 1982.

3/ "D1 AND E1" DIMENSIONS DO NOT INCLUDE MOLD PROTUSIONS. MOLD PROTUSIONS SHALL NOT EXCEED 0.25 mm (0.010 INCH). THE TOP PACKAGE BODY SIZE MAY BE SMALLER THAN THE BOTTOM PACKAGE BODY SIZE BY AS MUCH AS 0.15 mm.

4/ DATUM PLANE "H" LOCATED AT MOLD PARTING LINE AND COINCIDENT WITH LEAD, WHERE LEAD EXITS PLASTIC BODY AT BOTTOM OF PARTING LINE.

5/ DATUM "A" AND "D" TO BE DETERMINED AT DATUM PLANE H.

6/ DIMENSION " f " DOES NOT INCLUDE DAMBAR PROTUSION ALLOWABLE DAMBAR PROTUSION SHALL BE 0.08mm/.003" TOTAL IN EXCESS OF THE " f " DIMENSION AT MAXIMUM MATERIAL CONDITION . DAMBAR CANNOT BE LOCATED ON THE LOWER RADIUS OR THE FOOT.

### Document Revision History

Changes from	1.	Modified ordering information.
7663B to 7663C	2.	Various grammatical corections throughout document.
Changes from 7663C to 7663D	1.	TWI interface added.
Changes from	1.	Removed 64 and 68 pins package product version.
7663D to 7663E	2.	Minor correction on Table 69 on page 102.





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