\sum National Semiconductor

PRELIMINARY October 1994 DP83256/56-AP/57 PLAYER

DP83256/56-AP/57 $\mathsf{PLAYER} + \mathsf{TM}$ Device (FDDI Physical Layer Controller)

General Description

The DP83256/56-AP/57 Enhanced Physical Layer Controller (PLAYER $+$ device) implements one complete Physical Layer (PHY) entity as defined by the Fiber Distributed Data Interface (FDDI) ANSI X3T9.5 standard.

The PLAYER $+$ device integrates state of the art digital clock recovery and improved clock generation functions to enhance performance, eliminate external components and remove critical layout requirements.

FDDI Station Management (SMT) is aided by Link Error Monitoring support, Noise Event Timer (TNE) support, Optional Auto Scrubbing support, an integrated configuration switch and built-in functionality designed to remove all stringent response time requirements such as PC_React and CF_React.

Features

- Single chip FDDI Physical Layer (PHY) solution
- Integrated Digital Clock Recovery Module provides enhanced tracking and greater lock acquisition range
- Integrated Clock Generation Module provides all necessary clock signals for an FDDI system from an external 12.5 MHz reference
- Alternate PMD Interface (DP83256-AP/57) supports UTP twisted pair FDDI PMDs with no external clock recovery or clock generation functions required
- No External Filter Components
- Connection Management (CMT) Support (LEM, TNE, PC_React, CF_React, Auto Scrubbing)
- Full on-chip configuration switch
- Low Power CMOS-BIPOLAR design using a single 5V supply
- Full duplex operation with through parity
- Separate management interface (Control Bus)
- Selectable Parity on PHY-MAC Interface and Control Bus Interface
- Two levels of on-chip loopback
- 4B/5B encoder/decoder
- Framing logic
- **Elasticity Buffer, Repeat Filter, and Smoother**
- Line state detector/generator
- Supports single attach stations, dual attach stations and concentrators with no external logic
- DP83256 for SAS/DAS single path stations
- DP83257 for SAS/DAS single/dual path stations
- DP83256-AP for SAS/DAS single path stations that require the alternate PMD interface

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1.0 FDDI Chip Set Overview

National Semiconductor's next generation FDDI 2-chip set consists of two components as shown in Figure 1-1. The $PLAYER +$ device integrates the features of the DP83231 CRDTM Clock Recovery Device, DP83241 CDDTM Clock Distribution Device, and DP83251/55 PLAYERTM Physical Layer Controller. In addition, the $PLAYER +$ device contains enhanced SMT support.

National Semiconductor's FDDI TP-PMD Solutions consist of two components-the DP83222 CYCLONE™ Twisted Pair FDDI Stream Cipher Device and the DP83223A TWISTERTM Twisted Pair FDDI Transceiver Device.

For more information on the other devices of the chip set, consult the appropriate datasheets and application notes.

1.1 FDDI 2-CHIP SET

$DP83256/56-AP/57$ PLAYER + Device Physical Layer Controller

The $PLAYER +$ device implements the Physical Layer (PHY) protocol as defined by the ANSI FDDI PHY X3T9.5 standard.

Features

- Single chip FDDI Physical Layer (PHY) solution
- Integrated Digital Clock Recovery Module provides enhanced tracking and greater lock acquisition range
- Integrated Clock Generation Module provides all necessary clock signals for an FDDI system from an external 12.5 MHz reference
- Alternate PMD Interface (DP83256-AP/57) supports UTP twisted pair FDDI PMDs with no external clock recovery or clock generation functions required 3256/56-AP/57 PLAYER additioned in the commutes for higher performance and

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- No External Filter Components
- Gonnection Management (CMT) Support (LEM, TNE, PC_React, CF_React, Auto Scrubbing)
- \blacksquare Full on-chip configuration switch
- Low Power CMOS-BIPOLAR design using a single 5V supply
- \blacksquare Full duplex operation with through parity
- Separate management interface (Control Bus)
- Selectable Parity on PHY-MAC Interface and Control Bus Interface
- Two levels of on-chip loopback
- 4B/5B encoder/decoder
- \blacksquare Framing logic
- Elasticity Buffer, Repeat Filter, and Smoother
- Line state detector/generator
- Supports single attach stations, dual attach stations and concentrators with no external logic
- DP83256/56-AP for SAS/DAS single path stations
- P83257 for SAS/DAS single/dual path stations

In addition, the DP83257 contains the additional PHY Data.request and PHY_Data.indicate ports required for concentrators and dual attach, dual path stations.

DP83266 MACSITM Device Media Access Controller and System Interface

The DP83266 Media Access Controller and System Interface (MACSI) implements the ANSI X3T9.5 Standard Media Access Control (MAC) protocol for operation in an FDDI token ring and provides a comprehensive System Interface.

The MACSI device transmits, receives, repeats, and strips tokens and frames. It produces and consumes optimized data structures for efficient data transfer. Full duplex architecture with through parity allows diagnostic transmission and self testing for error isolation in point-to-point connections.

The MACSI device includes the functionality of both the DP83261 BMAC device and the DP83265 BSI-2 device with additional enhancements for higher performance and reliability.

Features

- Over 9 Kbytes of on-chip FIFO
- 5 DMA Channels (2 Output and 3 Input)
- 12.5 MHz to 33 MHz operation
- Full duplex operation with through parity
- Real-time VOID frame stripping indicator for bridges
- On-chip Address bit swapping capability
- 32-bit wide Address/Data path with byte parity
- Programmable transfer burst sizes of 4 or 8 32-bit words
- Receive frame filtering services
- Frame-per-Page mode controllable on each DMA channel
- Demultiplexed Addresses supported on ABus
- New multicast address matching
- ANSI X3T9.5 MAC standard defined ring service options
- Supports all FDDI Ring Scheduling Classes (Synchronous, Asynchronous, etc.)
- Supports Individual, Group, Short, Long, and External Addressing.
- Generates Beacon, Claim, and Void frames
- \blacksquare Extensive ring and station statistics gathering
- Extension for MAC level bridging
- **Enhanced SBus compatibility**
- \blacksquare Interfaces to DRAMs or directly to system bus
- Supports frame Header/Info splitting
- Programmable Big or Little Endian alignment

DP83222 CYCLONE Twisted Pair FDDI Stream Cipher Device

General Description

The DP83222 CYCLONE Stream Cipher Scrambler/Descrambler Device is an integrated circuit designed to interface directly with the serial bit streams of a Twisted Pair FDDI PMD. The DP83222 is designed to be fully compatible with the National Semiconductor FDDI Chip Sets, including twisted pair FDDI Transceivers, such as the DP83223A Twisted Pair Transceiver (TWISTER). The DP83222 requires a 125 MHz Transmit Clock and corresponding Receive Clock for synchronous data scrambling and descrambling. The DP83222 is compliant with the ANSI X3T9.5 TP-PMD standard and is required for the reduction of EMI emission over unshielded media. The DP83222 is specified to work in conjunction with existing twisted pair transceiver signalling schemes and enables high bandwidth transmission over Twisted Pair copper media.

Features

- Enables 100 Mbps FDDI signalling over Category 5 Unshielded Twisted Pair (UTP) cable and Type 1 Shielded Twisted Pair (STP)
- Reduces EMI emissions over Twisted Pair media
- Gompatible with ANSI X3T9.5 TP-PMD standard
- Requires a single $+5V$ supply
- Transparent mode of operation
- Elexible NRZ and NRZI format options
- Advanced BiCMOS process
- Signal Detect and Clock Detect inputs provided for enhanced functionality
- Suitable for Fiber Optic PMD replacement applications

DP83223A TWISTER High Speed Networking Transceiver Device

General Description

The DP83223A Twisted Pair Transceiver is an integrated circuit capable of driving and receiving either binary or (MLT-3) encoded datastreams. The DP83223A Transceiver is designed to interface directly with standards compliant FDDI, 100BASE-TX or STS-3c ATM chip sets, allowing low cost data links over copper based media. The DP83223A allows links of up to 100 meters over both Shielded Twisted Pair (STP) and datagrade Unshielded Twisted Pair (UTP) or equivalent. The electrical performance of the DP83223A meets or exceeds all performance parameters specified in the ANSI X3T9.5 TP-PMD standard, the IEEE 802.3 100BASE-TX Fast Ethernet Specification and the ATM Forum 155 Mbps Twisted Pair PMD Interface Specification. The DP83223A also provides important features such as baseline restoration, TRI-STATE® capable transmit outputs, and controlled transmit output edge rates (to reduce EMI radiation) for both binary and MLT-3 modes of operation. encion with existing twisted pair transcene of the PMD Interface Specification.

Interface Specification and pair opper media.

Interface Specification of TRI-STATE capable transmit cluding the media and controlled transmi

Features

- Gompliant with ANSI X3T9.5 TP-PMD standard
- Gompliant with IEEE 802.3 100BASE-TX Ethernet draft standard
- Gompliant with ATM Forum 155 Mbps Twisted Pair Specification
- Integrated baseline restoration circuit
- Integrated transmitter and receiver with adaptive equalization circuit
- Programmable binary or MLT-3 operation
- Isolated TX and RX power supplies for minimum noise coupling
- Controlled transmit output edge rates for reduced EMI
- TRI-STATE capable current transmit outputs
- **E** Loopback feature for board diagnostics
- Programmable transmit voltage amplitude

2.0 Architecture Description

2.1 BLOCK OVERVIEW

The PLAYER $+$ device is comprised of six blocks: Clock Recovery, Receiver, Configuration Switch, Transmitter, Station Management (SMT) Support, and Clock Generation Module as shown in *Figure 2-1*.

Clock Recovery

The Clock Recovery Module accepts a 125 Mbps NRZI data stream from the external PMD receiver. It then provides the extracted and synchronized data and clock to the Receiver block.

The Clock Recovery Module performs the following operations:

- Locks to and tracks the incoming NRZI data stream
- Extracts data stream and synchronized 125 MHz clock

Receiver

During normal operation, the Receiver Block accepts serial data as inputs at the rate of 125 Mbps from the Clock Recovery Module. During the Internal Loopback mode of operation, the Receiver Block accepts data directly from the Transmitter Block.

The Receiver Block performs the following operations:

- Optionally converts the incoming data stream from NRZI to NRZ.
- Decodes the data from 5B to 4B coding.
- Converts the serial bit stream into 10-bit bytes composed of 8 bits data, 1 bit parity, and 1 bit control information.
- Compensates for the differences between the upstream station clock and the local clocks.
- Decodes Line States.
- Detects link errors.
- Presents data symbol pairs (bytes) to the Configuration Switch Block.

Configuration Switch

An FDDI station may be in one of three configurations: Isolate, Wrap or Thru. The Configuration Switch supports these configurations by switching the transmitted and received data paths between $PLAYER +$ devices and one or more MACSI devices.

The configuration switch is integrated into the PLAYER + device, therefore no external logic is required for this function.

Setting the Configuration switch can be done explicitly via the Control Bus Interface or it can be set automatically with the CF_React SMT Support feature.

2.0 Architecture Description (Continued)

Transmitter

The Transmitter Block accepts 10-bit bytes composed of 8 bits data, 1 bit parity, and 1 bit control information from the Configuration Switch.

The Transmitter Block performs the following operations:

- Encodes the data from 4B to 5B coding.
- Filters out code violations from the data stream.
- Generates Idle, Master, Halt, Quiet, or other user defined symbol pairs upon request.
- Converts the data stream from NRZ to NRZI format for transmission.
- **Provides smoothing function when necessary.**

During normal operation, the Transmitter Block presents serial data to the PMD transmitter. While in Internal Loopback mode, the Transmitter Block presents serial data to the Receiver Block. While in the External Loopback mode, the Transmitter Block presents serial data to the Clock Recovery Module.

Clock Generation Module

The Clock Generation Module is an integrated phase locked loop that generates all of the required clock signals for the $PLAYER +$ device and an FDDI system from a single 12.5 MHz reference.

The Clock Generation Module features:

- High precision clock timing generated from a single 12.5 MHz reference.
- Multiple precision phased (8 ns/16 ns) 12.5 MHz Local Byte Clocks to eliminate timing skew in large multi-board concentrator configurations.
- LBC timing which is insensitive to loading variations over a wide range (20 pF to 70 pF) of LBC loads.
- A selectable dual frequency system clock.
- Low clock edge jitter, due to high VCO stability.

Station Management (SMT) Support

The Station Management Support Block provides a number of useful features to simplify the implementation of the Connection Management (CMT) portion of SMT.

These features eliminate the time critical CMT response time constraints imposed by PC_React and CF_React times.

Integrated counters and timers eliminate the need for additional external devices.

The following are the CMT features supported:

- PC_React
- \bullet CF $_\$ React
- Auto Scrubbing (TCF Timer)
- Timer, Idle Detection (TID Timer)
- Noise Event Counter (TNE Timer)
- Link Error Monitor (LEM Counter)

2.2 INTERFACES

The PLAYER $+$ device connects to other devices via five functional interfaces: PMD Interface, PHY Port Interface, Control Bus Interface, Clock Interface, and the Miscellaneous Interface.

PMD Interface

The PMD Interface connects the PLAYER $+$ device to a standard FDDI Physical Media Connection such as a fiber optic transceiver or a copper twisted pair transceiver. It is a 125 MHz full duplex serial connection.

The DP83256-AP and DP83257 PLAYER $+$ devices contain two PMD interfaces. The Primary PMD Interface should be used for all PMD implementations that do not require an external scrambler/descrambler function, clock recovery function, or clock generation function, such as a Fiber Optic or Shielded Twisted Pair (SDDI) PMD. The second, Alternate PMD Interface can be used to support Unshielded Twisted Pair (UTP) PMDs that require external scrambling, and allows implementation with no external clock recovery or clock generation functions required.

PHY Port Interface

The PHY Port Interface connects the $PLAYER +$ device to one or more MAC devices and/or $PLAYER +$ devices. Each PHY Port Interface consists of two byte-wide interfaces, one for PHY Request data input to the $PLAYER +$ device and one for the PHY Indicate data output of the PLAYER $+$ device. Each byte-wide interface consists of a parity bit (odd parity), a control bit, and two 4-bit symbols. We the minimist minimist behavior of the PEAT of the minimist and the properties of the properties and the p

The DP83257 PLAYER $+$ device has two PHY Port Interfaces while the DP83256 has one PHY Port Interface.

Control Bus Interface

The Control Bus Interface connects the $PLAYER +$ device to a wide variety of microprocessors and microcontrollers. The Control Bus is an asynchronous interface which provides access to 64 8-bit registers which monitor and control the behavior of the $PLAYER +$ device.

- The Control Bus Interface allows a user to:
- Configure SMT features.
- Program the Configuration Switch.
- Enable/disable functions within the Transmitter and Receiver Blocks (i.e., NRZ/NRZI Encoder, Smoother, PHY Request Data Parity, Line State Generation, Symbol pair Injection, NRZ/NRZI Decoder, Cascade Mode, etc.).

The Control Bus Interface also can be used to perform the following functions:

- . Monitor Line States received.
- \bullet Monitor link errors detected by the Receiver Block.
- Monitor other error conditions.

Clock Interface

The Clock Interface is used to configure the Clock Generation Module and to provide the required clock signals for an FDDI system.

The following clock signals are generated:

- 5 phase offset 12.5 MHz Local Byte Clocks
- 25 MHz Local Symbol Clock
- 15.625 or 31.25 MHz System Clock

Miscellaneous Interface

The Miscellaneous Interface consists of:

- A reset signal.
- User definable sense signals.
- User definable enable signals.
- \bullet Synchronization for cascading PLAYER + devices (a high-performance non-FDDI mode).
- Device Power and Ground pins.

3.0 Functional Description

The PLAYER $+$ device is comprised of six blocks: Clock Recovery, Receiver, Transmitter, Configuration Switch, Clock Generation, and Station Management Support.

3.1 CLOCK RECOVERY MODULE

The Clock Recovery Module accepts a 125 Mbps NRZI data stream from the external PMD receiver. It then provides the extracted and synchronized data and clock to the Receiver block.

The Clock Recovery Module performs the following operations:

- Locks onto and tracks the incoming NRZI data stream
- Extracts the data stream and the synchronized 125 MHz clock

The Clock Recovery Module is implemented using an advanced digital architecture that replaces sensitive analog blocks with digital circuitry. This allows the PLAYER + device to be manufactured to tighter tolerances since it is less sensitive to processing variations that can adversely affect analog circuits.

The Clock Recovery Module is comprised of 5 main functional blocks:

Digital Phase Detector

Digital Phase Error Processor

Digital Loop Filter

Digital Phase to Frequency Converter

Frequency Controlled Oscillator

See Figure 3-1, Clock Recovery Module Block Diagram.

DIGITAL PHASE DETECTOR

The Digital Phase Detector has two main functions: phase error detection and data recovery.

Phase error detection is accomplished by a digital circuit that compares the input data (PMID) to an internal phaselocked 125 MHz reference clock and generates a pair of error signals. The first signal is a pulse whose width is equal to the phase error between the input data and a reference clock and the second signal is a 4 ns reference pulse. These signals are fed into the Digital Phase Error Processor block.

The data recovery function converts the incoming encoded data stream (PMID) into synchronized data and clock signals. When the circuit is in lock the rising edge of the recovered clock is exactly centered in the recovered data bit cell.

The digital phase detector uses a common path for phase error detection and data recovery so as to minimize clock Static Alignment Error (SAE). Phase error averaging is also included so that phase errors generated by positive and negative PMID edges equally affect the clock recovery circuit. This greatly improves the immunity to Duty Cycle Distortion (DCD) in the data recovery circuit.

DIGITAL PHASE ERROR PROCESSOR

The Digital Phase Error Processor is responsible for sampling the Phase Detector's phase error outputs and producing two digital outputs that indicate to the digital loop filter how to adjust for a difference between the data phase and reference phases.

The Phase Error Processor is designed to eliminate the effects of different clock edge densities between data symbols and the various line state symbols on the PLL's loop gain.

Since the loop gain is held constant regardless of the incoming signal edge density, PLL characteristics such as jitter, acquisition rate, locking range etc., are deterministic and show minimal spread under various operating environments.

The phase error processor also automatically puts the loop in open-loop-mode when the incoming data stream contains abnormal low edge rates. When the PLL is in open-loopmode, no update is made to the PLL's filter variables in the filter block. The PLL can then use the pretrained frequency and phase contents to perform data recovery. Since the loop is implemented digitally, these values (the frequency and phase variables) are retained. The resolution of the frequency variable is about 1.3 ppm of the incoming frequency. The resolution of the phase variable is about 40 ps.

DIGITAL LOOP FILTER

The digital loop filter emulates a 1-pole, 1-zero filter and uses an automatic acquisition speed control circuit to dynamically adjust loop parameters.

The digital loop filter takes the phase error indicator signals Data Valid and Up/Down from the Phase Error processor and accumulates errors over a few cycles before passing on the Data Valid and Up/Down signals to the Phase Error to Frequency converter.

The filter has 4 sets of bandwidth and damping parameters which are switched dynamically by an acquisition control circuit. The input Signal Detect (SD) starts the sequence and, thereafter, no user programming is required to finish the sequence.

At the completion of the locking sequence, the loop has the narrowest bandwidth such that the loop produces minimal recovered clock jitter. The PLL can track an incoming frequency offset of approximately ± 200 ppm. After the acquisition sequence, the equivalent natural frequency of the loop is reduced to about 7 kHz (± 56 ppm) of frequency offset.

The automatic tracking mechanism allows the loop to quickly lock onto the initial data stream for data recovery (typically less than 10 μ s) and yet produce very little recovered clock iitter.

PHASE ERROR TO FREQUENCY CONVERTER $(\emptyset - F)$

The Phase Error to Frequency Converter takes the Data Valid and Up/Down signals modified by the Digital Loop Filter and converts them to triangle waves. The frequency of the triangle waves is then used to control the Frequency Controlled Oscillator's (FCO) 250 MHz oscillations.

Each valid Up or Down signal causes a partial 7-bit counter (using only 96 counts) to increment or decrement at the \varnothing – F converter's clock rate of 15.625 MHz (250 MHz/16). When the Data Valid signal is not asserted, the counter holds count.

The counter value is used to produce 3 triangle waves that are offset in phase by 120 degrees. This is done with a special Pulse Density Modulator waveform synthesizer which takes the place of a traditional Digital-Analog converter. The frequency of the triangle waves tells the Frequency Controlled Oscillator how much to adjust oscillation. The phase relationships (leading or lagging) between the 3 signals indicates the direction of change.

The minimum frequency of the triangle waves is 0 and corresponds to the case when the PLL is in perfect lock with the incoming signal.

The maximum frequency that the \varnothing –F converter can produce determines the locking range of the PLL. In this case the maximum frequency of each triangle wave is 162.76 kHz, which is produced when the \varnothing –F converter gets a continuous count in one direction that is valid every \varnothing – F converter clock cycle of 15.625 MHz (250 MHz/16). The triangle waves have an amplitude resolution of 48 digital steps, so a full rising and falling period takes 96 counts which produces a maximum frequency of 162.76 kHz (1/(1/15.625 kHz * 96)). s filter emulates a 1-point be incoming single. The incoming signal that we can be closed to the REL in this case.

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The 96 digital counts of the triangle waves also lead to a very fine PLL phase resolution of 42 ps (4 ns/96 counts). This high phase resolution is achieved using very low frequency signals, in contrast to a standard PLL which must operate at significantly higher frequencies than the data being tracked to achieve such high phase resolution.

FREQUENCY CONTROLLED OSCILLATOR (FCO)

The frequency controlled oscillator produces a 250 MHz clock that, when divided by 2, is phase locked to the incoming data's clock.

The FCO uses three 250 MHz reference clock signals from the Clock Generation Module and three 0 Hz to 162.76 kHz error clock signals from the Phase Error to Frequency Converter as inputs. Each signal in a triplet is 120 degrees phase shifted from the next.

Each corresponding pair (one 250 MHz and one error signal) of signals is mixed together using an amplitude switching modulator, with the error signal modulating the reference. All of the outputs are then summed together to produce the final 250 MHz $+f_m$ phase locked clock signal, where f_m is the error frequency.

3.2 RECEIVER BLOCK

During normal operation, the Receiver Block accepts serial data input at the rate of 125 Mbps from the Clock Recovery Module. During the Internal Loopback mode of operation, the Receiver Block accepts input data from the Transmitter Block.

The Receiver Block performs the following operations:

- Optionally converts the incoming data stream from NRZI to NRZ.
- Decodes the data from 5B to 4B coding.
- Converts the serial bit stream into the National byte-wide code.
- Compensates for the differences between the upstream station clock and the local clock.
- Decodes Line States.
- Detects link errors.
- . Presents data symbol pairs to the Configuration Switch Block.

The Receiver Block consists of the following functional blocks:

NRZI to NRZ Decoder

Shift Register

Framing Logic

Symbol Decoder

Line State Detector

Elasticity Buffer

Link Error Detector

See Figure 3-2.

NRZI TO NRZ DECODER

The NRZI to NRZ Decoder converts Non-Return-To-Zero-Invert-On-Ones data to Non-Return-To-Zero format.

NRZ format data is the natural data format that the receiver block utilizes internally, so this function is required when the standard NRZI format data is fed into the device. The receiver block can bypass this conversion function in the case where an alternate data source outputs NRZ format data.

This function can be enabled and disabled through bit 7 (RNRZ) of the Mode Register (MR). When the bit is cleared, it converts the incoming bit stream from NRZI to NRZ. This is the normal configuration required. When the bit is set, the incoming NRZ bit stream is passed unchanged.

SHIFT REGISTER

The Shift Register converts the serial bit stream into symbol-wide data for the 5B/4B Decoder.

The Shift Register also provides byte-wide data for the Framing Logic.

FRAMING LOGIC

The Framing Logic performs the Framing function by detecting the beginning of a frame or the Halt-Halt or Halt-Quiet symbol pair.

The J-K symbol pair (11000 10001) indicates the beginning of a frame during normal operation. The Halt-Halt (00100 00100) and Halt-Quiet (00100 00000) symbol pairs are detected for Connection Management (CMT).

Framing may be temporarily suspended (i.e. framing hold), in order to maintain data integrity.

Detecting JK

The JK symbol pair can be used to detect the beginning of a frame during Active Line State (ALS) and Idle Line State (ILS) conditions.

While the Line State Detector indicates Idle Line State the receiver ''reframes'' upon detecting a JK symbol pair and enters the Active Line State.

During Active Line State, acceptance of a JK symbol (reframing) is allowed for any on-boundary JK which is detected at least 1.5 byte times after the previous JK.

During Active Line State, once reframed on a JK, a subsequent off-boundary JK is ignored, even if it is detected beyond 1.5 byte times after the previous JK.

During Active Line State, an Idle or Ending Delimiter (T) symbol will allow reframing on any subsequent JK, if a JK is detected at least 1.5 byte times after the previous JK.

Detecting HALT-HALT AND HALT-QUIET

During Idle Line State, the detection of a Halt-Halt, or Halt-Quiet symbol pair will still allow the reframing of any subsequent on-boundary JK.

Once a JK is detected during Active Line State, off-boundary Halt-Halt, or Halt-Quiet symbol pairs are ignored until the Elasticity Buffer (EB) has an opportunity to recenter. They are treated as violations.

After recentering on a Halt-Halt, or Halt-Quiet symbol pair, all off boundary Halt-Halt or Halt-Quiet symbol pairs are ignored until the EB has a chance to recenter during a line state other than Active Line State (which may be as long as 2.8 byte times).

SYMBOL DECODER

The Symbol Decoder is a two level system. The first level is a 5-bit to 4-bit converter, and the second level is a 4-bit symbol pair to byte-wide code converter.

The first level latches the received 5-bit symbols and decodes them into 4-bit symbols. Symbols are decoded into two types: data and control. The 4-bit symbols are sent to the Line State Detector and the second level of the Symbol Decoder. See Table 3-1 for the 5B/4B Symbol Decoding list.

The second level translates two symbols from the 5B/4B converter and the line state information from the Line State Detector into the National byte-wide code.

LINE STATE DETECTOR

The ANSI X3T9.5 FDDI Physical Layer (PHY) standard specifies eight Line States that the Physical Layer can transmit. These Line States are used in the Connection Management process. They are also used to indicate data within a frame during normal operation.

The Line States are reported through the Current Receive State Register (CRSR), Receive Condition Register A (RCRA), and Receive Condition Register B (RCRB).

Note: V' denotes PHY Invalid or an Elasticity Buffer stuff byte I' denotes Idle symbol in ILS or an Elasticity Buffer stuff byte

LINE STATES DESCRIPTION

Active Line State

The Line State Detector recognizes the incoming data to be in the Active Line State upon the reception of the Starting Delimiter (JK symbol pair).

The Line State Detector continues to indicate Active Line State while receiving data symbols, Ending Delimiter (T symbols), and Frame Status symbols (R and S) after the JK symbol pair.

Idle Line State

The Line State Detector recognizes the incoming data to be in the Idle Line State upon the reception of 2 Idle symbol pairs nominally (plus up to 9 bits of 1 in start up cases).

Idle Line State indicates the preamble of a frame or the lack of frame transmission during normal operation. Idle Line State is also used in the handshake sequence of the PHY Connection Management process.

3.0 Functional Description (Continued) Super Idle Line State

The Line State Detector recognizes the incoming data to be in the Super Idle Line State upon the reception of 8 consecutive Idle symbol pairs nominally (plus 1 symbol pair).

The Super Idle Line State is used to insure synchronization of PCM signalling.

No Signal Detect

The Line State Detector recognizes the incoming data to be in the No Signal Detect state upon the deassertion of the Signal Detect signal or lack of internal clock detect from the Clock Recovery Module, and reception of 8 Quiet symbol pairs nominally. No Signal Detect indicates that the incoming link is inactive. This is the same as receiving Quiet Line State (QLS).

Master Line State

The Line State Detector recognizes the incoming data to be in the Master Line State upon the reception of eight consecutive Halt-Quiet symbol pairs nominally (plus up to 2 symbol pairs in start up cases).

The Master Line State is used in the handshaking sequence of the PHY Connection Management process.

Halt Line State

The Line State Detector recognizes the incoming data to be in the Halt Line State upon the reception of eight consecutive Halt symbol pairs nominally (plus up to 2 symbol pairs in start up cases).

The Halt Line State is used in the handshaking sequence of the PHY Connection Management process.

Quiet Line State

The Line State Detector recognizes the incoming data to be in the Quiet Line State upon the reception of eight consecutive Quiet symbol pairs nominally (plus up to 9 bits of 0 in start up cases).

The Quiet Line State is used in the handshaking sequence of the PHY Connection Management process.

Noise Line State

The Line State Detector recognizes the incoming data to be in the Noise Line State upon the reception of 16 noise symbol pairs without entering any known line state.

The Noise Line State indicates that data is not being received correctly.

Line State Unknown

The Line State Detector recognizes the incoming data to be in the Line State Unknown state upon the reception of 1 inconsistent symbol pair (i.e. data that is not expected). This may signify the beginning of a new line state.

Line State Unknown indicates that data is not being received correctly. If the condition persists the Noise Line State (NLS) may be entered.

ELASTICITY BUFFER

The Elasticity Buffer performs the function of a ''variable depth'' FIFO to compensate for phase and frequency clock skews between the Receive Clock ($\mathsf{RXC}\pm$) and the Local Byte Clock (LBC).

Bit 5 (EBOU) of the Receive Condition Register B (RCRB) is set to 1 to indicate an error condition when the Elasticity Buffer cannot compensate for the clock skew.

The Elasticity Buffer will support a maximum clock skew of 50 ppm with a maximum packet length of 4500 bytes.

To make up for the accumulation of frequency disparity between the two clocks, the Elasticity Buffer will insert or delete Idle symbol pairs in the preamble. Data is written into the byte-wide registers of the Elasticity Buffer with the Receive Clock, while data is read from the registers with the Local Byte Clock.

The Elasticity Buffer will recenter (i.e. set the read and write pointers to a predetermined distance from each other) upon the detection of a JK or every four byte times during PHY Invalid (i.e. MLS, HLS, QLS, NLS, NSD) and Idle Line State. The Elasticity Buffer is designed such that a given register cannot be written and read simultaneously under normal operating conditions. To avoid metastability problems, the EB overflow event is flagged and the data is tagged before the over/under run actually occurs.

LINK ERROR DETECTOR

The Link Error Detector provides continuous monitoring of an active link (i.e. during Active and Idle Line States) to insure that it does not exceed the maximum Bit Error Rate requirement as set by the ANSI standard for a station to remain on the ring.

Upon detecting a link error, the internal 8-bit Link Error Monitor Counter is decremented. The start value for the Link Error Monitor Counter is programmed through the Link Error Threshold Register (LETR). When the Link Error Monitor Counter reaches zero, bit 4 (LEMT) of the Interrupt Condition Register (ICR) is set to 1. The current value of the Link Error Monitor Counter can be read through the Current Link Error Count Register (CLECR). For higher error rates the current value is an approximate count because the counter rolls over. The Sate Studient recognizes the incoming data b.to worl/where the alleged and the data is agged belows that the Studient point exception of eight consecuently.

Maximum is the consecuently the alternative interest in the

There are two ways to monitor Link Error Rate: polling and interrupt.

Polling

The Link Error Monitor Counter can be set to a large value, like FF. This will allow for the greatest time between polling the register. This start value is programmed through the Link Error Threshold Register (LETR).

Upon detecting a link error, the Line Error Monitor Counter is decremented.

The Host System reads the current value of the Link Error Monitor Counter via the Current Link Error Count Register (CLECR). The Counter is then reset to FF.

Interrupt

The Link Error Monitor Counter can be set to a small value, like 5 to 10. This start value is programmed through the Link Error Threshold Register (LETR).

Upon detecting a link error, the Line Error Monitor Counter is decremented. When the counter reaches zero, bit 4 (LEMT) of the Interrupt Condition Register (ICR) is set to 1, and the interrupt signal goes low, interrupting the Host System.

Miscellaneous Items

When bit 0 (RUN) of the Mode Register (MR) is set to zero, or when the PLAYER $+$ device is reset through the Reset pin (\sim RST), the internal signal detect line is internally forced to zero and the Line State Detector is set to Line State Unknown and No Signal Detect.

3.3 TRANSMITTER BLOCK

The Transmitter Block accepts 10-bit bytes consisting of 8 bits data, 1 bit parity, and 1 bit control information, from the Configuration Switch.

The Transmitter Block performs the following operations:

- Encodes the data from 4B to 5B coding.
- Filters out code violations from the data stream.
- Is capable of generating Idle, Master, Halt, Quiet, or other user defined symbol pairs.
- Converts the data stream from NRZ to NRZI for transmission.
- Serializes data.

During normal operation, the Transmitter Block presents serial data to a PMD transmitter.

While in Internal Loopback mode, the Transmitter Block presents serial data to the Receiver Block. While in the External Loopback mode, the Transmitter Block presents serial data to the Clock Recovery Module.

The Transmitter Block consists of the following functional blocks:

Data Registers Parity Checker 4B/5B Encoder Repeat Filter **Smoother** Line State Generator Injection Control Logic Shift Register NRZ to NRZI Encoder

See Figure 3-3, Transmitter Block Diagram.

DATA REGISTERS

Data from the Configuration Switch is stored in the Data Registers. The 10-bit byte-wide data consists of a parity bit, a control bit, and two 4-bit data symbols as shown below.

FIGURE 3-4. Byte-Wide Data

The parity is odd parity. The control bit determines whether the Data bits represent Data or Control information. When the control bit is 0 the Data field is interpreted as data and when it is 1 the field is interpreted as control information according to the National Semiconductor control codes.

PARITY CHECKER

The Parity Checker verifies that the parity bit in the Data Register represents odd parity (i.e. odd number of 1s).

The parity is enabled and disabled through bit 6 (PRDPE) of the Current Transmit State Register (CTSR).

If a parity error occurs, the Parity Checker will set bit 0 (DPE) in the Interrupt Condition Register (ICR) and report the error to the Repeat Filter.

4B/5B ENCODER

The 4B/5B Encoder converts the two 4-bit data symbols from the Configuration Switch into their respective 5-bit codes.

See Table 3-2 for the Symbol Encoding list.

Note: The upper group of symbols are sent with the Control/Data pin set to Data, while the bottom grouping of symbols are sent with the Control/Data pin set to Control.

REPEAT FILTER

The Repeat Filter is used to prevent the propagation of code violations to the downstream station.

Upon receiving violations in data frames, the Repeat Filter replaces them with two Halt symbol pairs followed by Idle symbols. Thus the code violations are isolated and recovered at each link and will not be propagated throughout the entire ring.

 $\mathsf I$

TABLE 3-3. Abbreviations used in the Repeat Filter State Diagram

The Repeat Filter complies with the FDDI standard by observing the following (see Figure 3-5):

- 1. In Repeat State, violations cause transitions to Halt State and two Halt symbol pairs are transmitted (unless JK or Ix occurs) followed by transition to Idle State.
- 2. When Ix is encountered, the Repeat Filter goes to the Idle State, during which Idle symbol pairs are transmitted until a JK is encountered.
- 3. The Repeat Filter goes to the Repeat State following a JK from any state.

The END State, which is not part of the FDDI PHY standard, allows an R or S prior to a T within a frame to be recognized as a violation. It also allows NT to end a frame as opposed to being treated as a violation.

SMOOTHER

The Smoother is used to keep the preamble length of a frame to a minimum of 6 Idle symbol pairs.

Idle symbols in the preamble of a frame may have been added or deleted by each station to compensate for the difference between the Receive Clock and its Local Clock. The preamble needs to be maintained at a minimum length to allow stations enough time to complete processing of one frame and prepare to receive another. Without the Smoother function, the minimum preamble length (6 Idle symbol pairs) cannot be maintained as several stations may consecutively delete Idle symbols.

The Smoother attempts to keep the number of Idle symbol pairs in the preamble at 7 by:

. Deleting an Idle symbol pair in preambles which have more than 7 Idle symbol pairs

and/or

Inserting an idle symbol pair in preambles which have less than 7 idle symbol pairs (i.e. Extend State).

The Smoother Counter starts counting upon detecting an Idle symbol pair. It stops counting upon detecting a JK symbol pair.

Figure 3-6 describes the Smoother state diagram.

LINE STATE GENERATOR

The Line State Generator allows the transmission of the PHY Request data and can also generate and transmit Idle, Master, Halt, or Quiet symbol pairs which can be used to implement the Connection Management procedures as specified in the FDDI Station Management (SMT) standard document.

The Line State Generator is programmed through Transmit bits 0 to 2 (TM \le 2:0 $>$) of the Current Transmit State Register (CTSR).

Based on the setting of these bits, the Transmitter Block operates in a Transmit Mode where the Line State Generator overwrites the Repeat Filter and Smoother outputs.

See INJECTION CONTROL LOGIC section for a listing of the injection Transmit Modes.

Table 3-4 describes the Transmit Modes.

INJECTION CONTROL LOGIC

The Injection Control Logic replaces the data stream with a programmable symbol pair. This function is used to transmit data other than the normal data frame or Line States. The injection modes can be used for station diagnostic software.

The Injection Symbols overwrite the Line State Generator (Transmit Modes) and the Repeat Filter and Smoother outputs.

These programmable symbol pairs are stored in the Injection Symbol Register A (ISRA) and Injection Symbol Register B (ISRB). The Injection Threshold Register (IJTR) determines where the Injection Symbol pair will replace the data symbols.

The Injection Control Logic is programmed through the bits 0 and 1 (IC \le 1:0 $>$) of the Current Transmit State Register (CTSR) to one of the following Injection Modes (see Figure $3 - 7$:

- 1. No Injection (i.e. normal operation)
- 2. One Shot
- 3. Periodic
- 4. Continuous

In the No Injection mode, the data stream is transmitted unchanged.

In the One Shot mode, ISRA and ISRB are injected once on the nth byte after a JK, where n is the programmed value specified in the Injection Threshold Register.

In the Periodic mode, ISRA and ISRB are injected every nth symbol.

In the Continuous mode, all data symbols are replaced with the content of ISRA and ISRB. This is the same as periodic mode with $IJTR=0$.

SHIFT REGISTER

The Shift Register converts encoded parallel data to serial data. The parallel data is clocked into the Shift Register by the Local Byte Clock (LBC1), and clocked out by the Transmit Bit Clock $(TXC \pm)$ (externally available on the DP83257.)

NRZ TO NRZI ENCODER

The NRZ to NRZI Encoder converts the serial Non-Return-To-Zero data to Non-Return-To-Zero-Invert-On-One format.

This function can be enabled and disabled through bit 6 (TNRZ) of the Mode Register (MR). When programmed to "0", it converts the bit stream from NRZ to NRZI. When programmed to ''1'', the bit stream is transmitted NRZ.

3.4 CONFIGURATION SWITCH

The Configuration Switch consists of a set of multiplexers and latches which allow the $PLAYER +$ device to configure the data paths without any external logic. The Configuration Switch is controlled through the Configuration Register (CR).

The Configuration Switch has four internal buses: the A_Request bus, the B_Request bus, the Receive bus, and the PHY_Invalid bus. The two Request buses can be driven by external input data connected to the external PHY Port interface. The Receive bus is internally connected to the Receive Block of the $PLAYER +$ device, while the PHY_Invalid bus has a fixed 10-bit SMT PHY Invalid connection (LSU) pattern (1 0011 1010), which is useful during the connection process.

The configuration switch also has three internal multiplexers, each can select any of the four buses to connect to its respective data path. The first two are PHY Port interface output data paths, A_Indicate and B_Indicate, that can drive output data paths of the external PHY Port interface. The third output data path is connected internally to the Transmit Block.

The Configuration Switch is the same on the DP83256 device, the DP83256-AP device, and the DP83257 device. However, the DP83257 has two PHY Port interfaces connected to the Configuration Switch, whereas the DP83256 and DP83256-AP have one set of PHY port interfaces. The DP83257 uses the A_Request and A_Indicate paths as one PHY Port interface and the B_Request and B_Indicate paths as the other PHY Port interface (See Figure 3-8). The DP83256 and DP83256-AP, having one port interface, use the B_Request and A_Indicate paths as its external port. The A_Request and B_Indicate paths of the DP83256 and DP83256-AP are null connections and are not used by the device (See Figure 3-9).

STATION CONFIGURATIONS

Single Attach Station (SAS)

The Single Attach Station can be connected to either the Primary or Secondary ring via a Concentrator. Only 1 MAC is needed in a SAS.

The DP83256, DP83256-AP, and DP83257 can be used in a Single Attach Station. The DP83256 and DP83256-AP can be connected to the MAC via its only PHY Port interface. The DP83257 can be connected to the MAC via either one of its 2 PHY Port Interfaces.

See Figure 3-10 and Figure 3-11.

Dual Attach Station(DAS)

A Dual Attach Station can be connected directly to the dual ring, or, optionally to a concentrator. There are two types of Dual Attach Stations: DAS with a single MAC and DAS with two MAC layers. See Figure 3-12 and Figure 3-13.

Two DP83256 or DP83256-AP parts can be connected together to build a Dual Attach Station, however this configuration does not support the optional Thru_B configuration. When the optional Thru_B configuration is desired, it is recommended that the DP83257 be used.

A DAS with a single MAC and two paths can be configured as follows (see Figure 3-12):

- B Indicate data of PHY_A is connected to A Request input of PHY_B. B_Request input of PHY_A is connected to A Indicate output of PHY_B.
- The MAC can be connected to either the A Request input and the A Indicate output of PHY_A or the B Request input and the B Indicate output of PHY_B.

A DAS with a single MAC and one path using the DP83256 or DP83256-AP can be configured as follows (see Figure 3- 13 :

- B_Request input of PHY_A is connected to A Indicate output of PHY_B.
- . The MAC is connected to the B Request input of PHY_B and the A_Indicate output of PHY_A.

A DAS with dual MACs can be configured as follows (see Figure 3-14):

- B Indicate data of PHY_A is connected to A Request input of PHY_B. B_Request input of PHY_A is connected to A Indicate output of PHY_B.
- MAC₁ is connected to the B_Indicate output and the B_Request Input of PHY_B.
- MAC₂ is connected to the A_Indicate output and the A_Request Input of PHY_A.

CONCENTRATOR CONFIGURATIONS

There are 2 types of concentrators: Single Attach and Dual Attach. These concentrators can be designed with or without MAC(s). The configuration is determined based upon its type and the number of active MACs in the concentrator.

Using the PLAYER $+$ device, a concentrator can be built with many different configurations without any external logic.

The DP83256, DP83256-AP, and DP83257 can be used to build a Single Attach concentrator.

See Application Note AN-675, Designing FDDI concentrators and Application Note AN-741, Differentiating FDDI concentrators for further information.

Concepts

A concentrator is comprised of 2 parts: the Dual Ring Connect portion and the Master Ports.

The Dual Ring Connection portion connects the concentrator to the dual ring directly or to another concentrator. If the concentrator is connected directly to the dual ring, it is a part of the ''Dual Ring of Trees''. If the concentrator is connected to another concentrator, it is a ''Branch'' of the ''Dual Ring of Trees''. entative is completed of 2 parts. the Dual Ring Con-

For the Master Ports.

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The Master Ports connect the concentrator to its ''Slaves'', or S-class, Single Attach connections. A slave could be a Single Attach Station or another concentrator (thus forming another Branch of the Dual Ring Tree).

When a MAC in a concentrator is connected to the primary or secondary ring, it is required to be situated at the exit port of that ring (i.e. its PH IND is connected to the IND Interface of the last Master Port in the concentrator (PHY_M n) that is connected to that ring).

A concentrator can have two MACs, one connected to the primary ring and one to the secondary ring. In addition, roving MACs can be included in the concentrator configuration. A roving MAC can be used to test the stations connected to the concentrator before allowing them to join the dual ring.

This may require external multiplexers, if used in conjunction with two other MAC layers.

Single Attach Concentrator

A Single Attach concentrator is a concentrator that has only one PHY at the dual ring connect side. It cannot, therefore, be connected directly to the dual ring. A Single Attach concentrator is a branch to the dual ring tree. It is connected to the ring as a slave of another concentrator.

Multiple Single Attach concentrators can be connected together hierarchically to build a multiple levels of branches in a dual ring.

The Single Attach concentrator can be connected to either the primary or secondary ring depending on the connection with its concentrator (the concentrator that it is connected to as a slave).

Figure 3-15 shows a Single Attach concentrator with a single MAC.

Dual Attach Concentrator

A Dual Attach concentrator is a concentrator that has two PHYs on the dual ring connect side. It is connected directly to the dual ring and is a part of the dual ring tree.

The Dual Attach concentrator is connected to both the primary and secondary rings.

Dual Attach Concentrator with Single MAC

Figure 3-16 shows a Dual Attach concentrator with a single MAC.

Because the concentrator has one MAC, it can only transmit and receive frames on the ring to which the MAC is connected. The concentrator can only repeat frames on the other ring.

Dual Attach Concentrator with Dual MACs

Figure 3-17 shows a Dual Attach concentrator with dual MACs.

Because the concentrator has two MACs, it can transmit and receive frames on both the primary and secondary rings.

3.5 CLOCK GENERATION MODULE

The Clock Generation Module is an integrated phase locked loop that generates all of the required clock signals for the $PLAYER +$ device and the rest of an FDDI system from a single 12.5 MHz reference.

The Clock Generation Module features:

- . High precision clock timing generated from a single 12.5 MHz reference.
- Multiple precision phased (8 ns/16 ns) 12.5 MHz Local Byte Clocks to eliminate timing skew in large multi-board concentrator configurations.
- LBC timing which is insensitive to loading variations over a wide range (20 pF to 70 pF) of LBC loads.
- A selectable dual frequency system clock.
- Low clock edge jitter, due to high VCO stability.

The Clock Generation Module is comprised of 6 main functional blocks:

- Reference Selector
- Phase Comparator
- Loop Filter

250 MHz Voltage Controlled Oscillator

Output Phasing and Divide by 10

See Figure 3-18, Clock Generation Module Block Diagram.

REFERENCE SELECTOR

The Reference Selector block allows the user to choose between 2 sources for the Clock Generation Module's 12.5 MHz reference clock.

The simplest reference clock source option is to use an external 12.5 MHz reference signal fed into the REF IN input. This input can come from a crystal oscillator module or from a Local Byte Clock generated by another $PLAYER +$ device. Using the appropriate crystal oscillator ensures correct operating frequency without having to adjust any discrete components.

Using an LBC clock from another $PLAYER +$ device allows one PLAYER $+$ device to create a master clock to which other $PLAYER +$ devices in a system can be synchronized.

Another reference clock source option is a local 12.5 MHz crystal circuit. An example crystal circuit with component values is shown in *Figure 3-19*. This circuit is designed to operate with a crystal that has a C_L of 15 pF. The capacitor values may need to be slightly adjusted for an individual application to accomodate differences in parasitic loading.

PHASE COMPARATOR

The Phase Comparator uses two signal inputs: the selected 12.5 MHz reference from the Reference Select Block and a Local Byte Clock that has been selected for the feedback input, FBK_IN. Typically, LBC1 is used as the feedback clock.

The Phase Comparator generates a pulse of current that is proportional to the phase difference between the two signals. The current pulses are used to charge and discharge a control voltage on the internal Loop Filter. This control voltage is used to minimize the phase difference between the two signals.

LOOP FILTER

The Loop Filter is a simple internal filter made up of one capacitor in parallel with a serial capacitor and resistor combination. One end of the filter is connected to Ground and the other node is driven by the Phase Comparator and controls the internal 250 MHz Voltage Controlled Oscillator. This node can be examined for diagnostic purposes on the LPFLTR pin when the FLTREN bit of the CGMREG register is enabled. The LPFLTR pin is provided for diagnostic purposes only and should not be connected in any application.

The voltage on the Loop Filter is set by the current pulses generated by the Phase Comparator. The voltage on the Loop Filter node controls the frequency of the 250 MHz VCO.

250 MHZ VOLTAGE CONTROLLED OSCILLATOR (VCO)

The internal Voltage Controlled Oscillator is a low gain VCO whose primary frequency of oscillation centers around 250 MHz. The VCO produces little clock jitter due to its exceptional stability under all circumstances.

The VCO's output frequency is proportional to the voltage on the Loop Filter node.

OUTPUT PHASING

The Output Phasing block is a precision clock division circuit that produces clock signals of 4 distinct frequencies. Within the 12.5 MHz frequency, 5 clock signals with selectable 8 ns or 16 ns phase difference are produced.

The following clock signals are produced:

System Clock (CLK16/CLK32)

Local Symbol Clock (LSC)

Local Byte Clocks 1–5 (LBCn) (Divide by 10) System Clock (CLK16/CLK32)

The System Clock is provided as an extra set of clock frequencies that may be used as a clock for non-FDDI chipset portions of a system or as a higher frequency System Interface clock for the MACSI device. This clock is derived by dividing the 125 MHz clock by 8 or 4 times.

The frequency is selectable through the CLKSEL bit of the MODE2 register. The output has built-in glitch suppression so that changing the CLKSEL bit will not result in glitches appearing at the output.

Local Symbol Clock (LSC)

The Local Symbol Clock is a 40% HIGH/60% LOW duty cycle clock provided for use by the MACSI device and any external logic that needs to be synchronized to the Symbol timing.

This clock is derived by dividing the 125 MHz clock by 5.

Local Byte Clocks 1–5 (LBCn)

The Local Byte Clocks are provided for use by the MACSI device, by any external logic that needs to be synchronized to the Byte timing, and for use in concentrators to synchronize the timing between multiple $PLAYER +$ devices.

These clocks are derived by dividing the 125 MHz clock by 10. The different phase relationships between the LBCs are achieved by tapping off of different outputs of a Johnson counter inside the Output Phasing block.

The phase relationship (separation by 8 ns or 16 ns) of the LBCs is selected using the PH_SEL pin.

One of the LBCs must be used as the source of the feedback input, FBK IN, which requires a 12.5 MHz frequency. When the PLAYER + device is using a crystal as a reference it does not matter which LBC is used as the feedback input. Typically the least loaded LBC is used. However, when using an external reference that is supplied by another PLAYER $+$ device, it is important to select the LBC that keeps your system properly synchronized. Typically, all devices will use LBC1 as the feedback input.

3.6 STATION MANAGEMENT SUPPORT

The Station Management Support Block provides a number of useful features to simplify the implementation of the Connection Management (CMT) portion of SMT.

These features eliminate the most severe CMT response time constraints imposed by the PC_React and CF_React times. The many integrated counters and timers also eliminate the need for additional external devices.

The following CMT features are supported:

- \bullet PC React
- \bullet CF $_$ React
- Auto Scrubbing (TCF Timer)
- Timer, Idle Detection (TID Timer)
- Noise Event Counter (TNE Timer)
- Link Error Monitor (LEM Counter)

PC_REACT

PC_React is one of the timing restrictions imposed by Connection Management (CMT). It is one of the two most critical timing restrictions imposed (the other being CF_React.)

The ANSI SMT standard states that "PC_React is the maximum time for PCM [Physical Connection Management] to make a state transition to PC_Break when QLS, a fault condition, or PC_Start signal is present. This maximum time also places a limit on the time to react to a PC_Stop signal. This limitation does not apply to any other PCM transitions." PC_React puts a sharp time limit on how long it takes to transition to the PC_Break state and transmit the correct line state when a PCBreak transition is required. supery, Solox signals with selectable 8 ns

• Link Error Monder (LEM Counter)

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The range for the timer is PC_React ≤ 3.0 ms and has a default value equal to 3.0 ms.

The PLAYER $+$ device contains a Trigger Definition Register and a set of CMT Condition Registers that can be used to satisfy the PC_React timing.

The Trigger Definition Register (TDR) controls two functions. First, it allows the selection of the line state(s) on which to trigger (SILS, MLS, HLS \dots). For PC $_\$ React, the line states used would be the ones that caused a transition to the PC_Break state from the current PCM state

Second, it allows specification of a line state to be transmitted when the trigger condition is met. For PC_React, this is the line state that needs to be transmitted when a transition to the PC_Break state occurs, which is Quiet Line State (QLS).

The set of CMT Condition registers controls interrupt generation when a trigger condition occurs. The CMT Condition Register set includes a CMT Condition Register (CMTCR), a CMT Condition Comparison Register (CMTCCR), and a CMT Condition Mask Register (CMTCMR).

Line state triggering for PC_React is enabled by selecting line states to trigger on from the Trigger Definition Register (TDR) bits 3-7.

The Trigger Condition Occurred (TCO) bit of the CMTCR is automatically set when the trigger condition specified by the TDR register is met.

The line state specified by the Trigger Definition Register (TDR) bits 0–2 is then loaded into the Current Transmit Mode Register (CTSR), causing the line state to be transmitted.

If the TCO Mask (TCOM) bit of the CMTCMR is set, then whenever the CMTCR.TCO bit becomes set the Receive Condition Register B's Connection Service Event (RCRB.CSE) bit will be set. This allows an interrupt to be generated for the trigger event.

As an example, suppose the PCM state machine is in the ACTIVE state. From this state, if a Halt Line State (HLS) or Quiet Line State (QLS) is detected, or the Noise Threshold is reached, the state machine must move to the PC_Break state and begin transmitting QLS. To implement this behavior when the PC_ACTIVE state is entered, set TDR.TTM2–0 to 110 (Quiet Transmit), set TDR.TOHLS, TDR.TOQLS, and TDR.TONT and reset all other bits (TO-SILS and TOMLS). Also set CMTCMR.TCOM if an interrupt is desired.

CF REACT

CF_DReact is one of the timing restrictions imposed by Connection Management (CMT). It is one of the two most critical timing restrictions imposed (the other being PC_React).

The ANSI SMT standard states that "CF_React is the maximum time for CFM [Configuration Management] to reconfigure to remove a non-Active connection from the token path.''

The range for the timer is CF_React ≤ 3.0 ms and has a default value equal to 3.0 ms.

The PLAYER $+$ device contains a Trigger Transition Configuration Register and a set of CMT Condition Registers that can be used to satisfy the CF_React timing.

he Trigger Transition Configuration Register (TTCR) holds the new configuration switch settings to be loaded into the Configuration Register (CR) when a trigger condition occurs.

Enabling line state triggering with the Trigger Definition Register (TDR) bits 3-7 also enables the CF_React response. This means that whenever trigger conditions are actively used for PC_React, the value of the TTCR register will be used also. This implies that it either must always then be loaded with the current configuration setting, causing no change to the CR, or it must be loaded with the appropriate value to accommodate the CF_React function.

The Trigger Transition Configuration Register (TTCR) must be set the configuration desired when the trigger condition occurs. When the trigger condition occurs the value of this register is loaded into the Configuration Register (CR). During this time writes to the CR are inhibited.

To continue the example from the PC_React description, suppose that when in the ACTIVE state for the PCM state machine, the CFM state machine is also in the THRU_A state. If trigger conditions are enabled via the CMTCMR.TCOM bit and it is desired to not implement CF React, TTCR must be set to the present value of CR. If it is desired to not implement CF_React then TTCR should be set to the value which would change the configuration to the WRAP state. The wrap conditions WRAP_A or WRAP_B depend on which PHY gets reconfigured.

AUTO SCRUBBING

Auto Scrubbing is an additional CMT feature that further enhances the automatic configuration switch setting in order to meet the CF_React timing. When enabled, Auto Scrubbing causes 2 PHY_Invalid symbols followed by Scrub Symbol pairs (Idles) to be sourced for a user selectable duration (the scrubbing time) after a trigger condition (the same one used for PC_React and CF_React) occurs and prior to a change in the configuration switch setting on all indicate ports that will be changed.

Auto Scrubbing is enabled by setting the Enable Scrubbing on Trigger Conditions (ESTC) bit of Mode Register 2 (MODE2).

The Scrub Timer Threshold Register (STTR) defines the duration of the scrubbing, which can last up to approximately 10ms. The Scrub Timer Value Register (STVR) can be used to examine a snapshot of the upper 8 bits of the STTR register.

TIMER, IDLE DETECTION

The Idle Detection Timer is required to flag the continued presence of the Idle Line State for a duration of 8 Idle Symbol pairs plus 1 symbol pair.

This feature is implemented in the Receiver Block by the Super Idle Line State (SILS).

NOISE EVENT COUNTER

The Noise Event Counter can be used to time the duration between Noise Events (which are described in detail below) and to count frame sizes. The first feature is the most often recognized, but the second is often overlooked and can lead to potential difficulty if not properly set.

The Noise Event Counter is implemented as a pair of down counters: one the actual Noise Counter and the other a Noise Counter Prescaling value. The Noise Threshold Register (NTR) and the Noise Prescale Threshold Register (NPTR) can be programmed to the counter's initial value while the Current Noise Count Register (CNCR) and the Current Noise Prescale Count Register (CNPCR) provide a snapshot of the actual counter.

The Noise Event Counter decrements whenever a Noise Line State (NLS), Line State Unknown (LSU), or Active Line State (ALS) is received and has its start value reloaded whenever it receives Halt Line State (HLS), Idle Line State (ILS), Master Line State (MLS), Quiet Line State (QLS), or No Signal Detect (NSD). The Noise Event Counter is also reset for a Start or End Delimiter. This means the Noise counter increments for bad events as well as for every data symbol in a frame. Should the Noise Counter expire, it indicates that a new line state (including ALS) has not been entered for NT_MAX time. This indicates that either a frame is too long or that noise is being received. EACT the termin grastictions imposed by Con-

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For this reason it is important to choose a value for the counter that is larger than the longest frame of 4500 bytes. The ANSI SMT specification recommends a value for NT_MAX of 1.3ms for the noise threshold.

A Noise Event is defined as follows:

A noise event is a noisebyte, or a byte of data which is not in line with the current line state, indicating error or corruption.

Link Error Monitoring is accomplished in the PLAYER $+$ device through the Link Error Monitor Counter. The initial value of this down counter is set using the Link Error Threshold Register (LETR). A snapshot of the counter can be taken

3.7 PHY-MAC INTERFACE

NATIONAL BYTE-WIDE CODE

The PLAYER $+$ device outputs the National byte-wide code from its PHY Port Indicate Output to the MAC device. Each National byte-wide code may contain data or control codes or the line state information of the connection. Table 3-7 lists all the possible outputs.

During Active Line State all data and control symbols are being repeated to the PHY Port Indicate Output with the exception of data in data-control mixture bytes. That data symbol is replaced by zero. If only one symbol in a byte is a control symbol, the data symbol will be replaced by 0000 and the whole byte will be presented as control code. Note that the Line State Detector recognizes the incoming data to be in the Active Line State upon reception of the Starting Delimiter (JK symbol pair).

During Idle Line State any non Idle symbols will be reflected as the code I'uILS. If both symbols received during Idle Line State are Idle symbols, then the Symbol Decoder generates I'kILS as its output. Note the coded Known/Unknown Bit (b3) and the Last Known Line State (b2–0). The Receive State is 4 bits long and it represents either the PHY Invalid (0011) or the Idle Line State (1011) condition. The Known/ Unknown Bit shows if the symbols received match the line state information in the last 3 bits.

During any line state other than Idle Line State or Active Line State, the Symbol Decoder generates the code V'kLS if the incoming symbols match the current line state. The symbol decoder generates V'uLS if the incoming symbols do not match the current line state.

TABLE 3-7. National Byte Wide Code

National Byte-Wide Code Example

*Assume the receiver is in the Idle Line State.

3.8 PMD INTERFACE

The PMD Interface connects the PLAYER $+$ device to a standard FDDI Physical Media Connection such as a fiber optic transceiver or a copper twisted pair transceiver. It is a 125 MHz full duplex serial connection.

The DP83256 PLAYER $+$ device contains one PMD interface. This PMD Interface should be used for all PMD implementations that do not require an external scrambler/ descrambler function, clock recovery function, or clock generation function, such as a Fiber Optic or Shielded Twisted Pair (SDDI) PMDs.

The DP83256-AP and DP83257 PLAYER $+$ devices contain two PMD interfaces. The PMD Interface should be used for all PMD implementations that do not require an external scrambler/descrambler function, clock recovery function, or clock generation function, such as a Fiber Optic or Shielded Twisted Pair (SDDI) PMDs. The second, Alternate PMD Interface can be used to support Unshielded Twisted Pair (UTP) PMDs that require external scrambling, and allows implementation with no external clock recovery or clock generation functions required. See Figure 3-21.

$PLAYER + TO PMD CONNECTIONS$

The following figures illustrate how the $PLAYER +$ device can be connected to various types of PMDs.

Figure 3-20 shows how the DP83256, DP83256-AP, or D P83257 PLAYER + device is connected to a Fiber Optic or Shielded Twisted Pair (SDDI) PMD using the Primary PMD Interface.

Figure 3-21 shows how the DP83256-AP or DP83257 $PLAYER +$ device is connected to an Unshielded Twisted Pair (UTP) PMD using the Alternate PMD Interface.

3.0 Functional Description (Continued) INTERFACE ACTIVATION

The Primary PMD Interface is always enabled.

The Alternate PMD Interface is enabled by programming a $PLAYER + register bit.$ To enable the interface, write a 1 to the APMDEN bit in the APMDREG register. The interface is off by default and should be left that way unless it is being used.

It will also probably be necessary to enable the Transmit Clocks when using the Alternate PMD Interface. The Transmit Clocks (TXC) are enabled by writing a 1 to the TXCE bit in the CGMREG register. The transmit clocks are disabled by default and should be left that way unless it is being used.

Note that when the Alternate PMD Interface is active, the Primary PMD Interface can not be used without the Alternate PMD Interface connections. Also note that the Long Internal Loopback (LILB) can not be used when the Alternate PMD Interface is activated.

Obsolete

4.0 Modes of Operation

The PLAYER $+$ device can operate in 4 basic modes: RUN, STOP, LOOPBACK, and CASCADE.

4.1 RUN MODE

RUN is the normal mode of operation.

In this mode, the PLAYER $+$ device is configured to be connected to the media via the PMD transmitter and PMD receiver at the PMD Interface. It is also connected to any other $PLAYER + device(s)$ and/or MACSI device(s) via the Port A and Port B Interfaces.

While operating in the RUN mode, the $PLAYER +$ device receives and transmits Line States (Quiet, Halt, Master, Idle) and frames (Active LIne State).

4.2 STOP MODE

The PLAYER $+$ device operates in the STOP mode while it is being initialized or configured.

The PLAYER $+$ device is also reset to the STOP mode automatically when the \sim RST pin is set to ground.

When in STOP mode, the $PLAYER +$ device performs the following functions:

- Resets the Repeat Filter.
- Resets the Smoother.
- Resets the Receiver Block Line State Counters.
- Resets the Clock Recovery Module
- Flushes the Elasticity Buffer.
- **Forces Line State Unknown in the Receiver Block.**
- Outputs PHY Invalid condition symbol pairs through the PHY Data Indicate pins (AIP, AIC, AID<7:0>, BIP, BIC, $BID < 7:0$), when port is enabled.
- Outputs Quiet symbol pairs through the PMD Data Request pins (PMRD \pm).

4.3 LOOPBACK MODE

The PLAYER $+$ device provides 3 types of loopback tests: Configuration Switch Loopback, Short Internal Loopback, and Long Internal Loopback. These Loopback modes can be used to test different portions of the device.

Configuration Switch Loopback

The Configuration Switch Loopback can be used to test the data paths of the MACSI device(s) that are connected to the $PLAYER +$ device before transmitting and receiving data through the network.

In the Configuration Switch Loopback mode, the PLAYER $+$ device Configuration Register (CR) can be programmed to perform the following functions:

- Select Port A PHY Request Data, Port B PHY Request Data, or PHY Invalid to connect to Port A PHY Indicate Data via the A_NIND Mux.
- Select Port A PHY Request Data, Port B PHY Request Data, or PHY Invalid to connect to Port B PHY Indicate Data via the B_IND Mux.
- Connect data from the Receiver Block to the Transmitter Block via the Transmitter $_M$ Mux. (The PLAYER + device is repeating incoming data from the media in the Configuration Switch Loopback mode.)

See Figure 4-1 and Figure 4-2.

Short Internal Loopback

The Short Internal Loopback mode can be used to test the functionality of the $PLAYER +$ device, not including the Clock Recovery function, and to test the data paths between the $PLAYER +$ device and MACSI devices before ring insertion.

When in the Short Internal Loopback mode, the PLAYER \pm device performs the following functions:

- Directs the output data of the Transmitter Block to the input of the Receiver Block through an internal path.
- \bullet Ignores the PMD Data Indicate pins (PMID \pm),
- Outputs Quiet symbols through the PMD Data Request pins (PMRD \pm).

The level of the Quiet symbols transmitted through the $PMRD[±]$ pins during loopback is automatically set to the transmitter off level.

If both Short Internal Loopback and Long Internal Loopback modes are selected, Long Internal Loopback mode will have priority over Short Internal Loopback mode. This is the longest loopback path within the $PLAYER +$ device. See Figure 4-3, Short Internal Loopback.

Long Internal Loopback

The Long Internal Loopback mode implements the longest loopback path that is completely within the PLAYER + device.

The Long Internal Loopback mode can be used to test the f unctionality of the PLAYER + device, including the Clock Recovery function, and to test the data paths between the $PLAYER +$ device and MACSI devices before ring insertion. When in the Long Internal Loopback mode, the PLAYER $+$ device performs the following functions:

• Directs the output data of the Transmitter Block to the input of the Clock Recovery Module through an internal path.

- \bullet Ignores the PMD Data Indicate pins (PMID \pm),
- Outputs Quiet symbols through the PMD Data Request pins (PMRD \pm).

The level of the Quiet symbols transmitted through the $PMRD \pm$ pins during loopback is automatically set to the transmitter off level.

If both Short Internal Loopback and Long Internal Loopback modes are selected, Long Internal Loopback mode will have priority over Short Internal Loopback mode. This is the longest loopback path within the $PLAYER +$ device.

Note that the LILB path is disconnected and should not be used when the Alternate PMD Interface is active. See Figure 4-4, Long Internal Loopback.

4.4 DEVICE RESET

The revision B PLAYER $+$ device has five different levels of device Reset-Power Up Reset, Hardware Reset, Player Reset, Reference Select Reset, and Stop Mode. The Resets can be used to return the whole device or a portion of the device to its default configuration.

Power Up Reset begins automatically when power is first applied to the $PLAYER +$ device and reaches a certain voltage level. Power Up Reset affects all of the modules in the $PLAYER +$ device, specifically the Clock Generation Module (CGM), Clock Recovery Module (CRM), and the Player Module, returning each module to its default configuration. This reset begins by waiting for the crystal to stabilize, then the CGM PLL proceeds to lock to the crystal and the rest of the PLAYER $+$ device is reset. This reset takes the longest amount of time at approximately 10 ms from the time the $PLAYER +$ device's power supply reaches 4.4V. Even though the Power Up Reset is usually effective, due to the variation in the start-up conditions of a systems power supply, the Power Up Reset trigger can not be guaranteed to operate correctly. Therefore, a Hardware Reset should always be performed on the PLAYER $+$ after waiting a minimum of 10 ms for the Power Up Reset to complete its reset attempt.

Hardware Reset occurs at the rising edge of $PLAYER +$ device's \sim RST pin. Hardware Reset affects all of the modules in the PLAYER + device, specifically the CGM, CRM and the Player Module, returning each module to its default configuration. During Hardware Reset it is not necessary to force the Clock Generation Module to wait for the crystal to settle again at this time because it has settled in the time since the initial reset at power up. This reset takes the second longest amount of time at approximately 1 ms from the rising edge of \sim RST. at approximately 10 ms from the time the original revision A PLAYEEH device restal togic and the original relationship of the original relationship of the stationary of the stationary of the stationary of the stationary of

Player Reset is activated by writing a 1 to the PHYRST bit in Mode Register 2. Player Reset only affects the Player Module. This reset is the shortest and only takes about 3 μ s from the completion of the register write. The device should not be accessed by the Control Bus during this reset.

Reference Select Reset occurs when the $PLAYER + de$ vice's REF_SEL pin is switched from using the REF_IN input to using a crystal with the XTAL_IN and XTAL_OUT pins. This is the same as a Power Up Reset and is done because the crystal is going from a dead stop to an active state when REF_SEL is switched. This reset, like the Power Up Reset, takes about 10 ms from the falling edge of REF_{SEL.}

Stop Mode is activated by writing a 0 to the RUN bit in the Mode Register. Stop Mode is a selective reset that resets the Clock Recovery Module and portions of the Player Module.

Changes from Revision A to Revision B:

The previous descriptions describe the reset logic in the revision B PLAYER $+$ device. Two changes were made to the original revision A PLAYER + device reset logic.

First, the Hardware Reset was shortened by eliminating the requirement of having to wait for the crystal to settle before letting the Clock Generation Module try to lock to the crystal. This behavior is correct because the $PLAYER +$ device has already waited for the crystal to settle once during the Power Up Reset. The revision A PLAYER $+$ follows a Power Up Reset cycle when Hardware Reset is activated.

Second, a full Power Up Reset is now done when the clock reference is switched to the crystal. This is necessary to allow the crystal time to start up when it is switched to from the REF_IN input. This reset is not performed on the revision A $PLAYER +$.

Recommendations:

The following are some recommendations for using the reset mechanisms of the $PLAYER +$ most effectively:

- 1. Always wait a minimum of 10 ms after power-up before doing anything to the $PLAYER +$ device. 10 ms is a minimum, it may be desirable to wait longer if the system power supply or clock reference has not stabilized by this time.
- 2. Always use the Hardware Reset to reset the PLAYER + device after Power Up. This should be done after the initial Power Up waiting period of at least 10 ms.

4.5 CASCADE MODE

The PLAYER $+$ device can operate in the Cascade (parallel) mode (Figure 4-5) which is used in high bandwidth, point-to-point data transfer applications. This is a non-FDDI mode of operation. This is only available on the DP83257 device.

Concepts

In the Cascade mode, multiple $PLAYER +$ devices are connected together to provide data transfer at multiples of the FDDI data rate. Two cascaded $PLAYER +$ devices provide a data rate twice the FDDI data rate; three cascaded $PLAYER +$ devices provide a data rate three times the FDDI data rate, etc.

Multiple data streams are transmitted in parallel over each pair of cascaded PLAYER $+$ devices. All data streams start simultaneously and begin with the JK symbol pair on each $PLAYER + device.$

Data is synchronized at the receiver of each $PLAYER + de$ vice by the JK symbol pair. Upon receiving a JK symbol pair, a PLAYER $+$ device asserts the Cascade Ready signal to indicate the beginning of data reception.

The Cascade Ready signals of all $PLAYER +$ devices are open drain ANDed together to create the Cascade Start signal. The Cascade Start signal is used as the input to indicate that all $PLAYER +$ devices have received the JK symbol pair. Data is now being received at every PLAYER $+$ device and can be transferred from the cascaded $PLAYER +$ devices to the host system.

See Figure 4-6 for more information.

Operating Rules

When the PLAYER $+$ device is operating in Cascade mode, the following rules apply:

- 1. Data integrity can be guaranteed if the worst case PMD transmission skew between parallel media is less than 40 ns. For example, this amounts to about 785 meters of fiber optic cable, assuming a 1% worst case variance.
- 2. Even though this is a non-FDDI application, the general rules for FDDI frames must be obeyed.
- . Data frames must be a minimum of three bytes long (including the JK symbol pair). Smaller frames will cause Elasticity Buffer errors.
- Data frames must have a maximum size of 4500 bytes, with a JK starting delimiter and a T or R or S ending delimiter.
- 3. Due to the different clock rates, the JK symbol pair may arrive at different times at each $PLAYER +$ device. The total skew between the fastest and slowest cascaded $PLAYER +$ devices receiving the JK starting delimiter must not exceed 80 ns.
- 4. The first PLAYER + device to receive a JK symbol pair will present it to the host system and release the Cascade Ready signal. The $PLAYER +$ device will present one more JK as it waits for the other $PLAYER +$ devices to recognize their JK. The maximum number of consecutive JKs that can be presented to the host is 2.
- 5. The Cascade Start signal is set to 1 when all the cascaded PLAYER + devices release their Cascade Ready signals.
- 6. Bit 4 (CSE) of the Receive Condition Register B (RCRB) is set to 1 if the Cascade Start signal (CS) is not set before the second falling edge of clock signal LBC from when Cascade Ready (CR) was released. CS has to be set approximately within 80 ns of CR release. This condition signifies that not all cascaded $PLAYER +$ devices have received their respective JK symbol pair with the allowed skew range. cascade PLAYER+1 exvious All cascade Text who conserves that it is even in the U.K. The maximum number of conserved in the case of PLAYER+1 device and the presented to the mask that the reselver of each PLAYER+1 device and
	- 7. PLAYER + devices may not report a Cascaded Synchronization Error if the JK symbols are corrupted in the pointto-point links.
	- 8. To guarantee integrity of the interframe information, the user must put at least 8 Idle symbol pairs between frames. The PLAYER $+$ device will function properly with only 4 Idle symbol pairs, however the interframe symbols may be corrupted with random non-JK symbols.

The MACSI device could be used to provide the required framing and optional FCS support.

5.0 Registers

The PLAYER + device can be initialized, configured, and monitored using 64 8-bit registers. These registers are accessible through the Control Bus Interface.

The following tables summarize each register's attributes.

Note: RESERVED Registers may be read at any time, although the values read are not specified. The results of RESERVED Register writes are not specified, and
may have adverse implications. The user should not write to RES

5.2 CONFIGURATION REGISTER (CR)

The Configuration Register controls the Configuration Switch Block and enables/disables both the A and B ports. The CR can be used to create a number of Configuration Loopback paths.

The CR is conditionally writable because the TTCR can be writing a new value into the register if this feature is enabled. Note that the A_Request and B_Indicate port are offered only on the DP83257, and not in the DP83256. For further information, refer to section 3.4, CONFIGURATION SWITCH.

5.3 INTERRUPT CONDITION REGISTER (ICR)

The Interrupt Condition Register records the occurrence of an internal error event, the detection of Line State, an unsuccessful write by the Control Bus Interface, the expiration of an internal counter, or the assertion of one or more of the User Definable Sense pins.

The Interrupt Condition Register will assert the Interrupt pin (\sim INT) when one or more bits within the register are set to 1 and the corresponding mask bits in the Interrupt Condition Mask Register (ICMR) are also set to 1.

5.4 INTERRUPT CONDITION MASK REGISTER (ICMR)

The Interrupt Condition Mask Register allows the user to dynamically select which events will generate an interrupt. The Interrupt pin will be asserted (i.e. \sim INT = GND) when one or more bits within the Interrupt Condition Register (ICR) are set to 1 and the corresponding mask bits in this register are also set to 1.

This register is cleared (i.e. set to 0) and all interrupts are initially masked during the reset process.

5.5 CURRENT TRANSMIT STATE REGISTER (CTSR)

The Current Transmit State Register can program the Transmitter Block to internally generate and transmit Idle, Master, Halt, Quiet, or user programmable symbol pairs, in addition to the normal transmission of incoming PHY Request data. The Smoother and PHY Request Data Parity are also enabled and disabled through this register.

When the Trigger Definition register (TDR) is used, the CTSR can automatically be set to a preprogrammed line state when a trigger condition occurs. This capability can be used to implement both PC_React and CF_React.

The Transmit Modes have priority over the Repeat Filter and Smoother outputs. The Injection Symbols have priority over the Transmit Modes.

During the reset process (i.e. \sim RST = GND) the Transmit Mode is set to Off (TM < 2:0 > = 010), the Smoother is enabled (i.e. SE is set to 1), and the Reserved bit (b7) is set to 1. All other bits of this register are cleared (i.e. set to 0) during the reset process. When the TDR register is used to respond to trigger conditions the CTSR will be blocked when the TDR register transmit mode is copied into the CTSR. The Write Reject bit of the ICR will be set if any writes are attempted at this time. Note: This register has no effect while the device is in Stop Mode.

5.6 INJECTION THRESHOLD REGISTER (IJTR)

The Injection Threshold Register, in conjunction with the Injection Control bits (IC<1:0>) in the Current Transmit State Register (CTSR), set the frequency at which the contents of the Injection Symbol Register A (ISRA) and Injection Symbol Register B (ISRB) are inserted into the data stream. It contains the start value for the Injection Counter.

The Injection Threshold Register value is loaded into the Injection Counter when the counter reaches zero or during every Control Bus Interface write-cycle of this register.

The Injection Counter is an 8-bit down-counter which decrements every 80 ns. It's current value is read for CIJCR.

The counter is active only during One Shot or Periodic Injection Modes (i.e. Injection Control < 1:0 > bits (IC < 1:0 >) of the Current Transmit State Register (CTSR) are set to either 01 or 10). The Transmitter Block will replace a data symbol pair with ISRA and ISRB when the counter reaches 0 and the Injection Mode is either One Shot or Periodic.

If the Injection Threshold Register is set to 0 during the One Shot mode, the JK will be replaced with ISRA and ISRB. If the Injection Threshold Register is set to 0 during the Periodic mode, all data symbols are replaced with ISRA and ISRB.

The counter is initialized to 0 during the reset process (i.e. \sim RST = GND).

For further information, see the INJECTION CONTROL LOGIC section.

5.7 INJECTION SYMBOL REGISTER A (ISRA)

The Injection Symbol Register A, along with Injection Symbol Register B, contains the programmable value (already in 5B code) that can be inserted to replace the data symbol pairs.

In One Shot mode, ISRA and ISRB are injected n bytes after a JK, where n is the programmed value of the Injection Threshold Register. In the Periodic mode, ISRA and ISRB are injected every n-th symbol pair. In the Continuous mode, all data symbols are replaced with ISRA and ISRB.

5.8 INJECTION SYMBOL REGISTER B (ISRB)

The Injection Symbol Register B, along with Injection Symbol Register A, contains the programmable value (already in 5B code) that will replace the data symbol pairs.

In One Shot mode, ISRA and ISRB are injected n bytes after a JK, where n is the programmed value of the Injection Threshold Register. In the Periodic mode, ISRA and ISRB are injected every n-th symbol pair. In the Continuous mode, all data symbols are replaced with ISRA and ISRB.

5.9 CURRENT RECEIVE STATE REGISTER (CRSR)

The Current Receive State Register represents the current line state being detected by the Receiver Block. When the Receiver Block recognizes a new Line State, the bits corresponding to the previous line state are cleared, and the bits corresponding to the new line state are set.

During the reset process (\sim RST = GND), the Receiver Block is forced to Line State Unknown (i.e. the Line State Unknown bit (LSU) is set to 1).

Note: Users are discouraged from writing to this register. An attempt to write into this register will cause the PLAYER + device to ignore the Control Bus write cycle
and set the Control Bus Write Command Reject bit (CCR

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5.10 RECEIVE CONDITION REGISTER A (RCRA)

The Receive Condition Register A maintains a historical record of the Line States recognized by the Receiver Block. When a new Line State is entered, the bit corresponding to that line state is set to 1. The bits corresponding to the previous Line States are not cleared by the PLAYER + device, thereby maintaining a record of the Line States detected.

The Receive Condition A bit (RCA) of the Interrupt Condition Register (ICR) will be set to 1 when one or more bits within the Receive Condition Register A is set to 1 and the corresponding mask bit(s) in Receive Condition Mask Register A (RCMRA) is also set to 1.

5.11 RECEIVE CONDITION REGISTER B (RCRB)

The Receive Condition Register B maintains a historical record of the Lines States recognized by the Receiver Block. When a new Line State is entered, the bit corresponding to that line state is set to 1. The bits corresponding to the previous Line States are not cleared, thereby maintaining a record of the Line States detected.

The Receive Condition B bit (RCB) of the Interrupt Condition Register (ICR) will be set to 1 when one or more bits within the Receive Condition Register B is set to 1 and the corresponding mask bit(s) in Receive Condition Mask Register B (RCMRB) is also set to 1.

5.12 RECEIVE CONDITION MASK REGISTER A (RCMRA)

The Receive Condition Mask Register A allows the user to dynamically select which events will generate an interrupt. The Receive Condition A bit (RCA) of the Interrupt Condition Register (ICR) will be set to 1 when one or more bits within the Receive Condition Register A (RCRA) is set to 1 and the corresponding mask bit(s) in this register is also set to 1. Since this register is cleared (i.e. set to 0) during the reset process, all interrupts are initially masked.

5.13 RECEIVE CONDITION MASK REGISTER B (RCMRB)

The Receive Condition Mask Register B allows the user to dynamically select which events will generate an interrupt. The Receive Condition B bit (RCB) of the Interrupt Condition Register (ICR) will be set to 1 when one or more bits within the Receive Condition Register B (RCRA) is set to 1 and the corresponding mask bits in this register is also set to 1. Since this register is cleared (i.e. set to 0) during the reset process, all interrupts are initially masked.

5.14 NOISE THRESHOLD REGISTER (NTR)

The Noise Threshold Register contains the start value for the Noise Timer. This threshold register is used in conjunction with the Noise Prescale Threshold register for setting the maximum allowable time between entry to ILS, HLS, MLS, ALS, or NSD line states. The Noise timer is used to implement the TNE timing requirement of PCM. The Noise timer decrements by one for every $80 \times (NPTR+1)$ ns in case of Noise events. As a result, the internal noise counter takes the following amount of time to reach zero:

$((NPTR+1) \times NTR + NPTR) \times 80$ ns

The threshold values for the Noise Counter and Noise Prescale Counter are simultaneously loaded into both counters if one of the following conditions is true:

1. Both the Noise Counter and Noise Prescale Counter reach zero and the current Line State is either Noise Line State, Active Line State, or Line State Unknown.

or

2. The current Line State is either Halt Line State, Idle Line State, Master Line State, Quiet Line State, or No Signal Detect. or

3. The Noise Threshold Register or Noise Prescale Threshold Register goes through a Control Bus Interface write cycle. In addition, the value of the Noise Prescale Threshold register is loaded into the Noise Prescale Counter if the Noise Prescale Counter reaches zero.

The Noise Counter and Noise Prescale Counter will continue to count, without resetting or reloading the threshold values, if a Line State change occurs and the new line state is either Noise Line State, Active Line State, or Line State Unknown. When both the Noise Threshold Counter and Noise Counter both reach zero, the Noise Threshold bit of the Receive Condition

Register A will be set.

The recommended default value for the NTR register is 40h and for the NPTR register is F9h which corresponds to 1.3 ms as specified in the ANSI standard.

5.15 NOISE PRESCALE THRESHOLD REGISTER (NPTR)

The Noise Prescale Threshold Register contains the start value for the Noise Prescale Timer. This threshold register is used in conjunction with the Noise Threshold register for setting the maximum allowable time between entry to ILS, HLS, MLS, ALS, or NSD. The Noise timer is used to implement the TNE timing requirement of PCM. The Noise Prescale threshold controls how often the Noise timer is decremented. When the Noise Prescale Timer reaches zero, it reloads the count with the contents of the Noise Prescale Threshold Register and also causes the Noise Timer to decrement.

The threshold values for the Noise Counter and Noise Prescale Counter are simultaneously loaded into both counters if one of the following conditions is true:

1. Both the Noise Counter and Noise Prescale Counter reach zero and the current Line State is either Noise Line State, Active Line State, or Line State Unknown.

or

2. The Current Line State is either Halt Line State. Idle Line State, Master Line State, Quiet Line State, or No Signal Detect or

3. The Noise Threshold Register or Noise Prescale Threshold Register goes through a Control Bus Interface write cycle.

In addition, the value of the Noise Prescale Threshold Register is loaded into the Noise Prescale Counter if the Noise Prescale Counter reaches zero.

The Noise Counter and Noise Prescale Counter will continue to count, without resetting or reloading the threshold values, if a Line State change occurs and the new line state is either Noise Line State, Active Line State, or Line State Unknown.

When both the Noise Threshold Counter and Noise Counter both reach zero, the Noise Threshold bit of the Receive Condition Register A will be set.

See the NTR register description for default value recommendations.

5.16 CURRENT NOISE COUNT REGISTER (CNCR)

The Current Noise Count Register takes a snap-shot of the Noise Timer during every Control Bus Interface read cycle of this register.

During a Control Bus Interface write cycle, the Control Bus Write Command Reject bit (CCR) of the Interrupt Condition Register (ICR) will be set to 1 and will ignore a write cycle.

5.18 STATE THRESHOLD REGISTER (STR)

The State Threshold Register contains the start value for the State Timer. This timer is used in conjunction with the State Prescale Timer to count the Line State duration. The State Timer will decrement every 80 ns if the State Prescale Timer is zero and the current Line State is Halt Line State, Idle Line State, Master Line State, Quiet Line State, or No Signal Detect. The State Timer takes

$((SPTR + 1) \times STR + SPTR) \times 80$ ns

to reach zero during a continuous line state condition.

The threshold values for the State Timer and State Prescale Timer are simultaneously loaded into both counters if one of the following conditions is true:

1. Both the State Timer and State Prescale Timer reach zero and the current Line State is Halt Line State, Idle Line State, Master Line State, Quiet Line State, or No Signal Detect.

or

2. A line state change occurs and the new Line State is Halt Line State, Idle Line State, Master Line State, Quiet Line State, or No Signal Detect.

or

3. The State Threshold Register or State Prescale Threshold Register goes through a Control Bus Interface write cycle. In addition, the value of the State Prescale Threshold Register is loaded into the State Prescale Counter if the State Prescale Timer reaches zero.

The State Timer and State Prescale Timer will reset by reloading the threshold values, if a Line State change occurs and the new Line State is Halt Line State, Idle Line State, Master Line State, Quiet Line State, or No Signal Detect. On detection of ALS, NLS, or LSU the timer will not decrement.

5.19 STATE PRESCALE THRESHOLD REGISTER (SPTR)

The State Prescale Threshold Register contains the start value for the State Prescale Timer. The State Prescale Timer is a down counter. It is used in conjunction with the State Timer to count the Line State duration.

The threshold values for the State Timer and State Prescale Timer are simultaneously loaded into both timers if one of the following conditions is true:

1. Both the State Timer and State Prescale Timer reach zero and the current Line State is Halt Line State, Idle Line State, Master Line State, Quiet Line State, or No Signal Detect.

or

2. A Line State change occurs and the new Line State is Halt Line State, Idle Line State, Master Line State, Quiet Line State, or No Signal Detect.

or

3. The State Threshold Register or State Prescale Threshold Register goes through a Control Bus Interface write cycle. The State Prescale Timer will decrement every 80 ns if the current Line State is Halt Line State, Idle Line State, Master Line State, Quiet Line State, or No Signal Detect.

5.20 CURRENT STATE COUNT REGISTER (CSCR)

The Current State Count Register takes a snap-shot of the State Counter during every Control Bus Interface read cycle of this register.

During a Control Bus Interface write cycle, the Control Bus Write Command Reject bit (CCR) of the Interrupt Condition Register (ICR) will be set to 1 and will ignore a write cycle.

5.22 LINK ERROR THRESHOLD REGISTER (LETR)

The Link Error Threshold Register contains the start value for the Link Error Monitor Counter. It is an 8-bit down-counter which decrements if link errors are detected.

When the Counter reaches 0, the Link Error Monitor Threshold Register value is loaded into the Link Error Monitor Counter and the Link Error Monitor Threshold bit (LEMT) of the Interrupt Condition Register (ICR) is set to one.

The Link Error Monitor Threshold Register value is also loaded into the Link Error Monitor Counter during every Control Bus Interface write cycle of LETR.

The counter is initialized to 0 during the reset process (i.e. \sim RST=GND).

5.24 USER DEFINABLE REGISTER (UDR)

The User Definable Register is used to monitor and control events which are external to the PLAYER + device. The value of the Sense Bits reflect the asserted/deasserted state of their corresponding Sense pins. On the other hand, the Enable bits assert/deassert the Enable pins.

Note: SB2 and EB2 are only effective for the DP83257.

5.25 DEVICE ID REGISTER (IDR)

The Device ID Register contains the binary equivalent of the revision number for this device. It can be used to ensure proper software and hardware versions are matched.

During a Control Bus Interface write cycle, the Control Bus Write Command Register bit (CCR) of the Interrupt Condition Register (ICR) will be set to 1, and will ignore write cycle.

REVISION TABLE

5.26 CURRENT INJECTION COUNT REGISTER (CIJCR)

The Current Injection Count Register takes a snap-shot of the Injection Counter during every Control Bus Interface read cycle of this register.

During a Control Bus Interface write cycle, the Control Bus Write Command Reject bit (CCR) of the Interrupt Condition Register (ICR) will be set to 1 and will ignore a write cycle.

The Injection Counter is an 8-bit down-counter which decrements every 80 ns.

The counter is active only during One Shot or Periodic Injection Modes (i.e. Injection Control<1:0> bits (IC<1:0>) of the Current Transmit State Register (CTSR) are set to either 01 or 10).

The Injection Threshold Register (IJTR) value is loaded into the Injection Counter when the counter reaches zero and during every Control Bus Interface write cycle of IJTR.

The counter is initialized to 0 during the reset process (i.e. \sim RST=GND).

5.27 INTERRUPT CONDITION COMPARISON REGISTER (ICCR)

The Interrupt Condition Comparison Register ensures that the Control Bus must first read a bit modified by the PLAYER + device before it can be written to by the Control Bus Interface.

The current state of the Interrupt Condition Register (ICR) is automatically written into the Interrupt Condition Comparison Register (i.e. ICCR=ICR) during a Control Bus Interface read-cycle of ICR.

During a Control Bus Interface write cycle, the PLAYER + device will set the Conditional Write Inhibit bit (CWI) of the Interrupt Condition Register (ICR) to 1 and disallow the setting or clearing of a bit within ICR when the value of a bit in ICR differs from the value of the corresponding bit in the interrupt Condition Comparison Register.

5.28 CURRENT TRANSMIT STATE COMPARISON REGISTER (CTSCR)

The Current Transmit State Comparison Register ensures that the Control Bus must first read a bit modified by the PLAYER + device before it can be written to by the Control Bus Interface.

The current state of the Current Transmit State Register (CTSR) is automatically written into the Current Transmit State Comparison Register A (i.e. CTSCR = CTSR) during a Control Bus Interface read cycle of CTSR.

During a Control Bus Interface write cycle, the PLAYER + device will set the Conditional Write Inhibit bit (CWI) of the Interrupt Condition Register (ICR) to 1 and disallow the setting or clearing of a bit within the CTSR when the value of a bit in the CTSR differs from the value of the corresponding bit in the Current Transmit State Comparison Register.

5.29 RECEIVE CONDITION COMPARISON REGISTER A (RCCRA)

The Receive Condition Comparison Register A ensures that the Control Bus must first read a bit modified by the PLAYER+ device before it can be written to by the Control Bus Interface.

The current state of RCRA is automatically written into the Receive Condition Comparison Register A (i.e. RCCRA=RCRA) during a Control Bus Interface read cycle of RCRA.

During a Control Bus Interface write cycle, the PLAYER + device will set the Conditional Write Inhibit bit (CWI) of the Interrupt Condition Register (ICR) to 1 and prevent the setting or clearing of a bit within RCRA when the value of a bit in RCRA differs from the value of the corresponding bit in the Receive Condition Comparison Register A.

5.30 RECEIVE CONDITION COMPARISION REGISTER B (RCCRB)

The Receive Condition Comparison Register B ensures that the Control Bus must first read a bit modified by the PLAYER + device before it can be written to by the Control Bus Interface.

The current state of RCRB is automatically written into the Receive Condition Comparison Register B (i.e. RCCRB=RCRB) during a Control Bus Interface read cycle RCRB.

During a Control Bus Interface write cycle, the PLAYER + device will set the Conditional Write Inhibit bit (CWI) of the Interrupt Condition Register (ICR) to 1 and prevent the setting or clearing of a bit within RCRB when the value of a bit in RCRB differs from the value of the corresponding bit in the Receive Condition Comparison Register B.

5.32 CMT CONDITION COMPARISON REGISTER (CMTCCR)

The CMT Condition Comparison Register (CMTCR) ensures that the Control Bus must first read a bit modified by the PLAYER + device before it can be written to by the Control Bus Interface.

The current state of the CMT Condition Register (CMTCR) is automatically written into the CMT Condition Comparison Register (CMTCR) (i.e. CMTCCR = CMTCR) during a Control Bus Interface read-cycle of CMTCR.

During a Control Bus Interface write cycle, the PLAYER + device will set the Conditional Write Inhibit bit (CWI) of the Interrupt Control Register (ICR) to 1 and disallow the setting or clearing of a bit within the CMTCR when the value of a bit in the CMTCR differs from the value of the corresponding bit in the CMT Condition Comparison Register.

5.33 CMT CONDITION REGISTER (CMTCR)

The CMT Condition Register maintains a history of all CMT events and actions performed. The corresponding CMT Condition Mask Register (CMTCMR) can be used to generate an interrupt. When the bits in both the CMTCMR and CMTCR are set, the Receive Condition Register B's Connection Service Event (RCRB.CSE) bit will be set.

5.0 Registers (Continued) 5.36 SCRUB TIMER THRESHOLD REGISTER (STTR) This is the threshold value of the internal scrub timer. It has a resolution of 40.96 μ s and a maximum value of \sim 10 ms. When the scrub timer reaches zero, the Scrub Timer Expired (CMTCR.STE) bit is set. Scrubbing is initiated when MODE2.ESTC=1 and a trigger condition occurs. Writing to STTR during scrubbing will not affect the scrubbing action. ACCESS RULES ADDRESS READ WRITE 24h | Always | Always D7 D6 D5 D4 D3 D2 D1 D0 STT7 | STT6 | STT5 | STT4 | STT3 | STT2 | STT1 | STT0 Bit Symbol in the Symbol Description $\overline{D0-D7}$ STT0–STT7 SCRUB TIMER THRESHOLD BIT < 0-7>: Scrub Timer threshold. STT0 is the Least Significant Bit (LSB). Description

STTD is the Least Significant Bit (LSB).

STTD is the Least Significant Bit (LSB).

Contact Significant Bit (LSB).

5.38 TRIGGER DEFINITION REGISTER (TDR)

This register determines which events cause a trigger transition and which transmit mode is entered when a trigger transition is detected. The trigger transmit modes are the same as those found in the Current Transmit State Register (CTSR), and are loaded from the TDR into the CTSR when any of the selected trigger conditions occur. When a trigger condition occurs CMTCR.TCO is set.

The Trigger Definition Register is useful to implement the strict PC_React time requirement.

5.39 TRIGGER TRANSITION CONFIGURATION REGISTER (TTCR)

The Trigger Transition Configuration Register holds the configuration switch setting to be loaded into the Configuration Register (CR) when a trigger transition takes place. When scrubbing is enabled, scrubbing is performed for a period of time indicated by the Scrub Timer Threshold Register (STTR).The register bit descriptions for the Configuration Register and, therefore, the Trigger Transition Configuration Register are reprinted below.

5.41 CLOCK GENERATION MODULE REGISTER (CGMREG)

This register is used to enable or disable the 125 MHz ECL Transmit clock outputs. These outputs are not required for use in a standard FDDI board implementation and are disabled by default to reduce high frequency noise.

These TXC outputs are included for support of alternate FDDI PMDs, such as unshielded twisted pair copper cable.

DO NOT WRITE TO RESERVED REGISTER BITS. Writes to reserved register bits could prevent proper device operation. Therefore, read the register first, and then write it back with the non-reserved bits set to the desired value.

5.42 ALTERNATE PMD REGISTER (APMDREG)

This register is used to enable or disable the Alternate PMD inputs and ouputs. These signals are not required for use in FDDI board implementations that do not require a scrambler that is external to the PLAYER $+$ device. The actual interface consists of the signal pairs RXC_OUT, RXD_OUT, RXC_IN, and RXD_IN.

The interface is disabled by default and should only be enabled if it is being used. Note that Long Internal Loopback should not be used when the Alternate PMD Interface is enabled.

DO NOT WRITE TO RESERVED REGISTER BITS. Writes to reserved register bits could prevent proper device operation. Therefore, read the register first, and then write it back with the non-reserved bits set to the desired value.

Note: The Alternate PMD Interface pins are only available on the 100-pin DP83256-AP and 160-pin DP83257 PLAYER + devices. The Alternate PMD Interface is disabled on reset.

5.43 GAIN REGISTER (GAINREG)

The Gain Register contains the settings for the CGM's on-chip programmable loop filter. For optimal jitter performance on the revision A and B PLAYER + device's Filter Position 4 should be used. The user should check that the IDR register is equal to revision A or B (10h or 11h) before changing the filter setting as later revisions will default to the correct setting which may be a different filter position number.

Pseudo Code Programming Example:

Care must be taken when changing the settings of the on-chip programmable loop filter. The filter should only be set to the recommended value and the additional bits in the Gain Register must not be altered. Alteration of the reserved bits in the Gain Register may result in improper $PLAYER +$ device operation.

The following pseudo code outlines the proper procedure for setting the Gain Register loop filter settings to the correct value.

```
// Register names and constants are all in UPPERCASE
//
//
#define REV_B 0x11
#define REV A 0x10
#define LOOP_MASK OxlF
#define NEW LOOP 0x40
```
if (IDR \leq REV_B) $\{$

 $temp = GAMN_REG$ $temp = temp$ & $LOOP_MASK$ $temp = temp | NEW_LOOP$ $GAIN_REG = temp$

Ó

else {Do Nothing}

6.0 Signal Descriptions

6.1 DP83256VF PIN DESCRIPTIONS

The pin descriptions for the DP83256VF are divided into 5 functional interfaces: PMD Interface, PHY Port Interface, Control Bus Interface, Clock Interface, and Miscellaneous Interface.

For a Pinout Summary list, refer to Table 8-1 and Figure 8-1, DP83256VF 100-Pin JEDEC Metric PQFP Pinout.

PMD INTERFACE

The PMD Interface consists of I/O signals used to connect the PLAYER + device to the Physical Medium Dependant (PMD) sublayer.

PHY PORT INTERFACE

The PHY Port Interface consists of I/O signals used to connect the PLAYER + device to the Media Access Control (MAC) sublayer or other PLAYER + device. The DP83256 Device has two PHY Port Interfaces. The A_Indicate path from one PHY Port Interface and the B_Request path from the second PHY Port Interface. Each path consists of an odd parity bit, a control bit, and two 4-bit symbols.

Refer to section 3.3, the Configuration Switch, for more information.

CONTROL BUS INTERFACE

The Control Bus Interface consists of I/O signals used to connect the PLAYER + device to Station Management (SMT). The Control Bus is an asynchronous interface between the PLAYER + device and a general purpose microprocessor or other controller. It provides access to 64 8-bit internal registers.

In the PLAYER + device the Control Bus address range has been expanded by 1-bit to 6 bits of address space.

CLOCK INTERFACE

The Clock Interface consists of 12.5 MHz and 25 MHz clocks supplied by the PLAYER + device as well as reference and feedback inputs.

MISCELLANEOUS INTERFACE

The Miscellaneous Interface consist of a reset signal, user definable sense signals, and user definable enable signals.

POWER AND GROUND

All power pins should be connected to a single +5V power supply using the recommended filtering. All ground pins should be connected to a common 0V ground supply. Bypassing and filtering requirements are given in a separate User Information Document.

6.2 DP83256VF-AP SIGNAL DESCRIPTIONS

The pin descriptions for the DP83256VF-AP are divided into five functional interfaces; PMD Interface, PHY Port Interface, Control Bus Interface, Clock Interface, and Miscellaneous Interface.

For a Pinout Summary List, refer to Table 8-2 and Figure 8-2, DP83256VF-AP 100-Pin JEDEC Metric PQFP Pinout.

PMD INTERFACE

The PMD Interface consists of I/O signals used to connect the PLAYER + device to the Physical Medium Dependant (PMD) sublayer.

The DP83256VF-AP PLAYER + device actually has two PMD interfaces. The Primary PMD Interface and the Alternate PMD Interface.

The Primary PMD Interface should be used for all PMD implementations that do not require an external scrambler/descrambler function, clock recovery function, or clock generation function, such as a Fiber Optic or Shielded Twisted Pair (SDDI) PMD. The second, Alternate PMD Interface can be used to support Unshielded Twisted Pair (UTP) PMDs that require external scrambling, with no external clock recovery or clock generation functions required.

Section 3.8 describes how the PLAYER + can be connected to the PMD and how the Alternate PMD can be enabled.

Note that when the Alternate PMD Interface is not being used, the pins that make up the interface must be connected in the specific way described in the following Alternate PMD Interface table.

PHY PORT INTERFACE

The PHY Port Interface consists of I/O signals used to connect the PLAYER + device to the Media Access Control (MAC) sublayer or other PLAYER+ device. The DP83256 Device has two PHY Port Interfaces. The A \Box Indicate path from one PHY Port Interface and the B_Request path from the second PHY Port Interface. Each path consists of an odd parity bit, a control bit, and two 4-bit symbols.

Refer to section 3.3, the Configuration Switch, for more information.

CONTROL BUS INTERFACE

The Control Bus Interface consists of I/O signals used to connect the PLAYER+ device to Station Management (SMT). The Control Bus is an asynchronous interface between the PLAYER $+$ device and a general purpose microprocessor or other controller. It provides access to 64 8-bit internal registers.

In the PLAYER + device the Control Bus address range has been expanded by 1-bit to 6 bits of address space.

CLOCK INTERFACE

The Clock Interface consists of 12.5 MHz and 25 MHz clocks supplied by the PLAYER + device as well as reference and feedback inputs.

MISCELLANEOUS INTERFACE

The Miscellaneous Interface consist of a reset signal and user definable enable signals.

POWER AND GROUND

All power pins should be connected to a single +5V power supply using the recommended filtering. All ground pins should be connected to a common 0V ground supply. Bypassing and filtering requirements are given in a separate User Information Document.

SPECIAL CONNECT PINS

These are pins that have special connection requirements.

No Connect (N/C) pins should not be connected to anything. This means not to power, not to ground, and not to each other. Reserved_0 (RES_0) pins must be connected to ground. These pins are not used to supply device power so they do not need to be filtered or bypassed.

Reserved_1 (RES_1) pins must be connected to power. These pins are not used to supply device power so they do not need to be filtered or bypassed.

6.3 DP83257VF SIGNAL DESCRIPTIONS

The pin descriptions for the DP83257VF are divided into five functional interfaces; PMD Interface, PHY Port Interface, Control Bus Interface, Clock Interface, and Miscellaneous Interface.

For a Pinout Summary List, refer to Table 8-3 and Figure 8-3, DP83257VF 160-Pin JEDEC Metric PQFP Pinout.

PMD INTERFACE

The PMD Interface consists of I/O signals used to connect the PLAYER + device to the Physical Medium Dependant (PMD) sublayer.

The DP83257 PLAYER + device actually has two PMD interfaces. The Primary PMD Interface and the Alternate PMD Interface. The Primary PMD Interface should be used for all PMD implementations that do not require an external scrambler/descrambler function, clock recovery function, or clock generation function, such as a Fiber Optic or Shielded Twisted Pair (SDDI) PMD. The second, Alternate PMD Interface can be used to support Unshielded Twisted Pair (UTP) PMDs that require external scrambling, with no external clock recovery or clock generation functions required.

Section 3.8 describes how the PLAYER + can be connected to the PMD and how the Alternate PMD can be enabled. Note that when the Alternate PMD Interface is not being used, the pins that make up the interface must be connected in the specific way described in the following Alternate PMD Interface table. Primary PMD Interface

PHY PORT INTERFACE

The PHY Port Interface consists of I/O signals used to connect the PLAYER+ device to the Media Access Control (MAC) sublayer or other PLAYER+ device. The DP83257 Device has two PHY Port Interfaces. The A_Request and A_Indicate paths from one PHY Port Interface and the B_Request and B_Indicate paths from the second PHY Port Interface. Each path consists of an odd parity bit, a control bit, and two 4-bit symbols.

Refer to section 3.3, the Configuration Switch, for more information.

CONTROL BUS INTERFACE

The Control Bus Interface consists of I/O signals used to connect the PLAYER+ device to Station Management (SMT). The Control Bus is an asynchronous interface between the PLAYER $+$ device and a general purpose microprocessor or other controller. It provides access to 64 8-bit internal registers.

In the PLAYER + device the Control Bus address range has been expanded by 1-bit to 6 bits of address space.

CLOCK INTERFACE

The Clock Interface consists of 12.5 MHz and 25 MHz clocks supplied by the PLAYER + device as well as reference and feedback inputs.

 $\overline{}$

MISCELLANEOUS INTERFACE

The Miscellaneous Interface consist of a reset signal, user definable sense signals, and user definable enable signals.

6.0 Signal Descriptions (Continued)

POWER AND GROUND

All power pins should be connected to a single +5V power supply using the recommended filtering. All ground pins should be connected to a common 0V ground supply. Bypassing and filtering requirements are given in a separate User Information Document.

SPECIAL CONNECT PINS

These are pins that have special connection requirements.

No Connect (N/C) pins should not be connected to anything. This means not to power, not to ground, and not to each other. Reserved_0 (RES_0) pins must be connected to ground. These pins are not used to supply device power so they do not need to be filtered or bypassed.

Reserved_1 (RES_1) pins must be connected to power. These pins are not used to supply device power so they do not need to be filtered or bypassed.

DC Electrical Characteristics for All FDDI Clock Outputs

The following signals are covered: Local Byte Clocks (LBC1–LBC5), and Local Symbol Clock (LSC). These outputs are designed to drive capacitive loads from 20 pF to 60 pF.

DC Electrical Characteristics for All Clock Reference Inputs

The following signals are covered: Reference In (REF_IN) and Feedback In (FBK_IN).

DC Electrical Characteristics for Crystal Inputs and Outputs

The following signals are covered: Crystal In (XTAL_IN) and Crystal Out (XTAL_OUT).

Note A: This parameter is presented as a typical value to provide enough information to design an appropriate crystal network.

DC Electrical Characteristics for All Open Drain Outputs

The following signals are covered: Interrupt (\sim INT), Acknowledge (\sim ACK), and Cascade Ready (CR).

DC Electrical Characteristics for All 100K ECL Compatible Inputs

The following signals are covered: PMD Indicate Data (PMID), Receive Clock In (RXC_IN), Receive Data In (RXD_IN), and Signal Detect (SD).

Note 1: Both inputs of each differential pair are tested together. These specifications guarantee that the inputs are compatible with standard 100K ECL voltage level outputs.

Note 2: V_{CM} is measured from the crossover point of the 300 mV differential test input.

DC Electrical Characteristics for 100K ECL Compatible Outputs

The following signals are covered: PMD Request Data (PMRD) and Transmit Clock (TXC).

DC Electrical Characteristics for Alternate PMD ECL Outputs

The following signals are covered: Receive Clock Out (RXC_OUT) and Receive Data Out (RXD_OUT).

Note 3: It is recommended that RXC_OUT+ and RXC_OUT- always be used together as a differential pair. It is recommended that RXD_OUT+ and RXD _OUT $-$ always be used together as a differential pair.

Supply Current Electrical Characteristics

*Note: The PLAYER+ device has multiple pairs of differential ECL outputs that need to be terminated. The additional current needed for this termination is not $included$ in the PLAYER + 's total supply current, but can be calculated as follows:

V_{OH}__max = V_{CC} - 0.88V
V_{OL}__max = V_{CC} - 1.62V

Since the outputs are differential, the average output level is V_{CC} - 1.25V. The test load per output is 50 Ω at V_{CC} - 2V, therefore the external load current through the 50 Ω resistor is:

 $I_{\text{LOAD}} = \frac{[(V_{\text{CC}} - 1.25) - (V_{\text{CC}} - 2)]}{50}$
= 0.015A

 $= 15$ mA

As a result, the termination for each pair of active ECL outputs typically consumes 30 mA, time averaged.

AC Characteristics for the Alternate PMD Interface

The following input signals are covered: PMD Indicate Data (PMID), Signal Detect (SD), Receive Data In (RXD_IN), Receive Clock In (RXC_IN).

The following output signals are covered: PMD Request Data (PMRD), Transmit Clock (TXC), Recovered Data Out (RXD_OUT), Recovered Clock Out (RXC_OUT).

Note: The Alternate PMD Interface is only available on the 160 pin DP83257 PLAYER+ Device and the 100 pin DP83256-AP Device. The Transmit Clock is
enabled by the CGMREG.TXCE bit. The rest of the Alternate PMD Interface i

Note 1: This parameter is not tested, but is assured by correlation with characterization data.

FIGURE 7-11. ECL Rise and Fall Times

TL/F/11708–52

AC Characteristics for the PMD Interface Inputs (ANSI Specifications)

The following input signals are covered: PMD Indicate Data (PMID), Receive Data In (RXD_IN), Receive Clock In (RXC_IN). Note: The Alternate PMD Interface is only available on the 160 pin DP83257 PLAYER + Device and the 100 pin DP83256-AP Device. The Transmit Clock is enabled by the CGMREG.TXCE bit. The rest of the Alternate PMD Interface is enabled by the APMDREG.APMDEN bit.

All comments in square brackets are section references to the ANSI documents where these specifications can be found.

AC Characteristics for the PMD Interface Outputs (ANSI Specifications)

The following output signals are covered: PMD Request Data (PMRD), Transmit Clock (TXC), Recovered Data Out (RXD_OUT), Recovered Clock Out (RXC_OUT).

Note: The Alternate PMD Interface is only available on the 160 pin DP83257 PLAYER+ Device and the 100 pin DP83256-AP Device. The Transmit Clock is
enabled by the CGMREG.TXCE bit. The rest of the Alternate PMD Interface i

Comments in square brackets are section references to the ANSI documents where these specifications can be found.

Note 1: This parameter is not tested, but is assured through characterization data and periodic testing of sample units.

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