



## 4 Channel Level Translating Fast-Mode Plus I2C-bus/SMbus Repeater

## **Features**

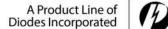
- 4 channel, bidirectional buffer isolates capacitance and allows 540pF on either side of the device at 1 MHz and up to 4000 pF at lower speeds
- Voltage level translation from 0.6V to 5.5V and from 2.2V to 5.5V
- Port A operating supply voltage range of 0.6V to 5.5V with normal levels( 0.4VCC(A) + 0.8 V ≤ VCC(B) )
- Port B operating supply voltage range of 2.2V to 5.5V with static offset level
- 5V tolerant I<sup>2</sup>C-bus and enable pins
- 0 Hz to 1 MHz clock frequency (the maximum system operating frequency may be less than 1MHz because of the delays added by the repeater)
- Active HIGH repeater enable input referenced to VCC(B)
- Open-drain input/outputs
- Latching free operation
- Supports arbitration and clock stretching across the repeater
- Accommodates Standard-mode, Fast-mode and Fastmode Plus I2C-bus devices, SMBus (standard and high power mode), PMBus and multiple masters
- Powered-off high-impedance I<sup>2</sup>C-bus pins
- ESD protection exceeds 8000V HBM per JESD22-A114
- Package: TQFN-16L 4x4 and QSOP-16L

## **Description**

The PI6ULS5V9627A is a CMOS integrated circuit intended for Fast-mode Plus (Fm+) I<sup>2</sup>C-bus or SMBus applications. It can provide level shifting between low voltage (down to 0.6V) and higher voltage (2.2V to 5.5V) in mixed-mode applications. The PI6ULS5V9627A enables the system designer to isolate two halves of a bus for both voltage and capacitance, accommodating more I<sup>2</sup>C devices or longer trace length. It also permits extension of the I<sup>2</sup>C-bus by providing bidirectional buffering for both the data (SDA) and the clock (SCL) lines, thus enabling two buses of 540 pF at 1 MHz or up to 4000 pF at lower speeds. The SDA and SCL pins are overvoltage tolerant and are high-impedance when the PI6ULS5V9627A is unpowered.

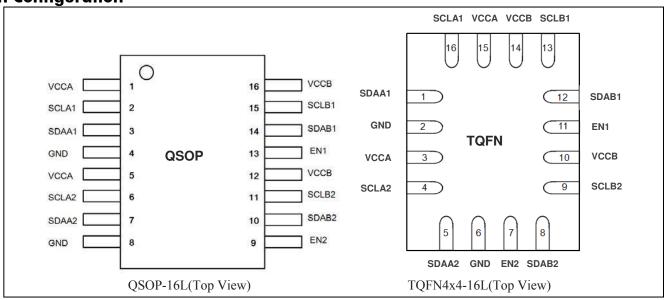
The 2.2V to 5.5V bus port B drivers have the static level offset, while the adjustable voltage bus port A drivers eliminate the static offset voltage. This results in a LOW on the port B translating into a nearly 0V LOW on the port A which accommodates the smaller voltage swings of lower voltage logic. The EN pin is referenced to  $V_{\rm CC(B)}$  and can also be used to turn the drivers on and off under system control.







**Pin Configuration** 



# **Pin Description**

Pir	ı No.	D! N	D 1.0		
QSOP	TQFN	Pin Name	Description		
3	1	SDAA1	serial data port A bus 1		
4	2	GND <sup>(1)</sup>	supply ground (0 V)		
5	3	VCCA <sup>(2)</sup>	port A supply voltage (0.6 V to 5.5V)		
6	4	SCLA2	serial clock port A bus 2		
7	5	SDAA2	serial data port A bus 2		
8	6	GND <sup>(1)</sup>	supply ground (0 V)		
9	7	EN2	active HIGH repeater enable input 2		
10	8	SDAB2	serial data port B bus 2		
11	9	SCLB2	serial clock port B bus 2		
12	10	VCCB <sup>(3)</sup>	port B supply voltage (2.2 V to 5.5 V)		
13	11	EN1	active HIGH repeater enable input 1		
14	12	SDAB1	serial data port B bus 1		
15	13	SCLB1	serial clock port B bus 1		
16	14	VCCB <sup>(3)</sup>	port B supply voltage (2.2 V to 5.5 V)		
1	15	VCCA <sup>(2)</sup>	port A supply voltage (0.6 V to 5.5 V)		
2	16	SCLA1	serial clock port A bus 1		

### Note:

- (1) The two GND pins need to be connected to the ground, can't be floating.
- (2) The two VCCA pins need to be connected to the power supply, can't be floating.
- (3) The two VCCB pins need to be connected to the power supply, can't be floating.

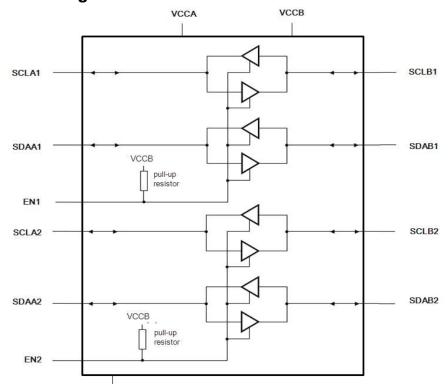


GND



PI6ULS5V9627A

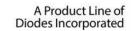
# **Block Diagram**



EN1	EN2	Function
Н	X	SCLA1 = SCLB1 SDAA1= SDAB1
X	Н	SCLA2 = SCLB2 SDAA2=SDAB2
L	L	disabled

Figure 1: Block Diagram







## **Maximum Ratings**

Storage Temperature	-55°C to +125°C
Supply Voltage port B	-0.5V to +6.0V
Supply Voltage port A	-0.5V to+6.0V
DC Input Voltage	-0.5V to +6.0V
Control Input Voltage (EN)	-0.5V to+6.0V
Total Power Dissipation	100mW
Input/Output Current (port A&B)	50mA
Input current (EN, VCCA, VCCB, GND)	50mA
ESD: HBM Mode	8000V

### Note:

Stresses greater than those listed under MAXIMUM RATINGS may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

## **DC Electrical Characteristics**

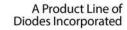
 $V_{CC(A)} = 0.6V$  to  $5.5V^{(1)}$ ;  $V_{CC(B)} = 2.2V$  to 5.5V; GND = 0V;  $T_A = -40$ °C to +85°C; unless otherwise noted.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>CC(B)</sub>	supply voltage port B	-	2.2	-	5.5	V
$V_{CC(A)}$	supply voltage port A	-	0.6	-	5.5	V
T	supply current on pin	$V_{CC(A)} = 0.95V$	-	-	16	۸
$I_{CC(A)}$	$V_{CC(A)}$	$V_{CC(A)} = 5.5V$	=	-	100	μΑ
$I_{\text{CCH(B)}}$	port B HIGH-level supply current	$V_{CC(B)} = 5.5 \text{ V}$ $SDAn = SCLn = V_{CC(n)}$	-	3	5	mA
I <sub>CCL(B)</sub>	port B LOW-level supply current	$V_{CC(B)} = 5.5 \text{ V}$ ; one SDA and one SCL = GND; other SDA and SCL open (with pull-up resistors)	-	3.4	5.8	mA
I <sub>QVC(B)</sub>	Quiescent current on $V_{CC(B)}$	EN=GND;V <sub>CC(B)</sub> =5.5V	-	0.8	1	mA

## Note:

<sup>(1)</sup>  $V_{CC(A)}$  may be as high as 5.5 V for over voltage tolerance but  $0.4V_{CC(A)} + 0.8 \text{ V} \leq V_{CC(B)}$  for the channels to be enabled and functional normally.







## **DC Electrical Characteristics**

 $V_{CC(A)} = 0.6V$  to  $5.5V^{(5)}$ ;  $V_{CC(B)} = 2.2V$  to 5.5V; GND = 0V;  $T_A = -40^{\circ}C$  to  $+85^{\circ}C$ ; Typical values measured with  $V_{CC(A)} = 0.95V$  and  $V_{CC(B)} = 2.5V$ , unless otherwise noted.

Symbol	Parameter	Test Conditions	Min	Тур	Max	Unit		
Input and	Input and output SDAB and SCLB							
V <sub>IH</sub>	HIGH-level input voltage	-	0.7V <sub>CC(B)</sub>	-	5.5	V		
$V_{\rm IL}$	LOW-level input voltage	-	-0.5	ı	+0.4	V		
$V_{IK}$	Input clamping voltage	$I_I = -18 \text{ mA}$	-1.2	-	-	V		
$I_{LI}$	Input leakage current	$V_I = 5.5V$	-	-	±1	μΑ		
$I_{\rm IL}$	LOW-level input current	SDA, SCL; $V_I = 0.2 \text{ V}$	-	í	10	μΑ		
$V_{OL}$	LOW-level output voltage	$I_{OL} = 150 \mu A$ at $V_{CC(B)} = 2.2 V^{(1)}$	0.47	-	-	V		
	D'CC LOWL LOW	$I_{OL} = 13 \text{mA} \text{ at } V_{CC(B)} = 2.2 V^{(2)}$	-	0.54	0.60	V		
$V_{\text{OL}}$ - $V_{\text{IL}}$	Difference between LOW-level output and LOW-level input voltage contention	$V_{OL}$ at $I_{OL} = 1$ mA; guaranteed by design	60	90	160	mV		
$C_{io}$	Input/output capacitance	$V_I = 3V \text{ or } 0V; V_{CC(B)} = 3.3V;$ EN = LOW	-	7	10	pF		
	1 1	$V_I = 3V$ or $0V$ ; $V_{CC(B)} = 0V$	-	7	10	pF		
Input and	Input and output SDAA and SCLA							
$V_{\mathrm{IH}}$	HIGH-level input voltage	-	$0.7V_{CC(A)}$	-	5.5	V		
$V_{ m IL}$	LOW-level input voltage	-	$-0.5^{(3)}$	-	$+0.25V_{CC(A)}^{(4)}$	V		
$V_{IK}$	Input clamping voltage	$I_I = -18 \text{ mA}$	-1.2	-	-	V		
${ m I_{LI}}$	Input leakage current	$V_I = 5.5V$	-	-	±1	μΑ		
${ m I}_{ m IL}$	LOW-level input current	SDA, SCL; $V_I = 0.2 \text{ V}$	-	í	10	μΑ		
$V_{\mathrm{OL}}$	LOW-level output voltage	$I_{OL} = 13$ mA at $V_{CC(B)} = 2.2$ V	-	0.1	0.2	V		
$C_{io}$	Input/output capacitance	$V_I = 3V \text{ or } 0V; V_{CC(B)} = 3.3V;$ EN = LOW	-	7	10	pF		
	1 1	$V_I = 3V$ or $0V$ ; $V_{CC(B)} = 0V$	-	7	10	pF		
Enable								
$V_{\mathrm{IH}}$	HIGH-level input voltage	-	$0.7V_{CC(B)}$	ı	5.5	V		
$V_{\rm IL}$	LOW-level input voltage	-	-0.5	ı	+0.3V <sub>CC(B)</sub>	V		
$I_{LI}$	Input leakage current	$V_{I} = V_{CC(B)}$	-1	-	+1	μΑ		
${ m I}_{ m IL}$	LOW-level input current	$V_I = 0.2V$ , EN; $V_{CC(B)} = 2.2V$ ;	-18	-7	-	μΑ		
Ci	Input capacitance	$V_{I} = V_{CC(B)}$	-	6	-	pF		

### Note

- (1) Pull-up should result in  $I_{OL} \ge 150 \mu A$ .
- (2) Guaranteed by design and characterization.
- (3)  $V_{IL}$  for port A with envelope noise must be below  $0.3V_{CC(A)}$  for stable performance.
- (4) When  $V_{CC(A)}$  is less than 1V, care is required to make certain that the system ground offset and noise is minimized such that there is reasonable difference between the  $V_{IL}$  present at the PI6ULS5V9627 A-side input and the  $0.25V_{CC(A)}$  input threshold.
- (5)  $V_{CC(A)}$  may be as high as 5.5 V for over-voltage tolerance but  $0.4V_{CC(A)} + 0.8 \text{ V} \leq V_{CC(B)}$  for the channels to be enabled and functional normally.





**Dynamic characteristics**  $V_{CC(A)} = 0.6V$  to  $5.5V^{(8)}$ ;  $V_{CC(B)} = 2.2V$  to 5.5V; GND = 0V;  $T_A = -40^{\circ}C$  to  $+85^{\circ}C$ ; Typical values measured with  $V_{CC(A)} = 0.95V$  and  $V_{CC(B)} = 2.5V$ , unless otherwise noted. (1)(2)

Symbol	Parameter	Conditions	Min	Typ <sup>[3]</sup>	Max	Unit
$t_{\rm PLH}$	LOW-to-HIGH propagation delay	B-side to A-side	-	-52	-103	ns
$t_{\mathrm{PLH2}}^{[4]}$	LOW-to-HIGH propagation delay2	B-side to A-side	-	94	130	ns
${ m t_{PHL}}$	HIGH-to-LOW propagation delay	B-side to A-side	-	76	152	ns
$\mathfrak{t}_{\mathrm{TLH}}^{[5]}$	LOW-to-HIGH transition time	A-side	-	60	1	ns
SRf	Falling slew rate	port A; 0.7V <sub>CC(A)</sub> to 0.3V <sub>CC(A)</sub>	-	0.037	-	V/ns
$t_{\mathrm{PLH}}^{[6]}$	LOW-to-HIGH propagation delay	A-side to B-side	-	45	102	ns
$\mathfrak{t}_{\mathrm{PHL}}^{[6]}$	HIGH-to-LOW propagation delay	A-side to B-side	-	50	173	ns
$t_{ m TLH}$	LOW-to-HIGH transition time	B-side	-	60	-	ns
$t_{ m THL}$	HIGH-to-LOW transition time	B-side	-	5	-	ns
$t_{\rm en}^{[7]}$	Enable time	Quiescent -0.3 V; EN HIGH to enable;	-	-	100	ns
t <sub>dis</sub> <sup>[7]</sup>	Disable time	quiescent + 0.3 V; EN LOW to disable;	-	-	100	ns

### Note:

- (1) Times are specified with loads of 1.35 kΩ pull-up resistance and 50 pF load capacitance on port A and port B, and a falling edge slew rate of 0.05 V/ns input signals.
- (2) Pull-up voltages are  $V_{\text{CC}(A)} \text{ on port } A \text{ and } V_{\text{CC}(B)} \text{ on port } B.$
- (3) Typical values were measured with  $V_{CC(A)} = 0.95 \text{ V}, V_{CC(B)} = 2.5 \text{ V}$  at  $T_A = 25 ^{\circ}\text{C}$ , unless otherwise noted.
- (4) The  $t_{PLH2}$  delay data from port B to port A is measured at 0.45 V on port B to  $0.5V_{CC(A)}$  on port A.
- (5) The  $t_{TLH}$  of the bus is determined by the pull-up resistance (1.35 k  $\Omega$ ) and the total capacitance (50 pF).
- (6) The proportional delay data from port A to port B is measured at 0.5V<sub>CC(A)</sub> on port A to 0.5V<sub>CC(B)</sub> on port B.
- (7) The enable pin EN, should only change state when the global bus and the repeater port are in an idle state.
- (8)  $V_{CC(A)}$  may be as high as 5.5 V for over-voltage tolerance but  $0.4V_{CC(A)} + 0.8 \text{ V} \leq V_{CC(B)}$  for the channels to be enabled and functional normally.

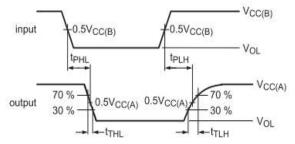


Figure 2: Propagation Delay and Transition Times B→A

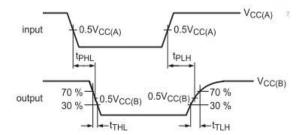
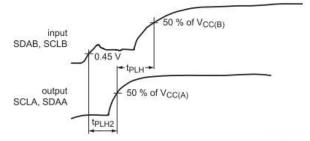


Figure 3: Propagation Delay and Transition Times A→B



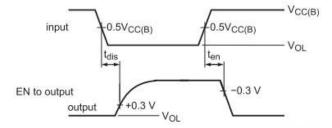
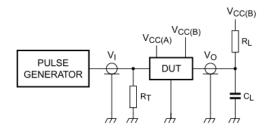


Figure 4: Propagation Delay and Enable and disable time





 $R_L = load resistor; 1.35 k\Omega on port B$ 

 $C_{\text{\tiny L}}$  = load capacitance includes jig and probe capacitance; 50 pF

 $R_T$  = termination resistance should be equal to  $Z_0$  of pulse generators

Figure 5: Test Circuit

## **Functional Description**

The PI6ULS5V9627A enables I $^2$ C-bus or SMBus translation down to  $V_{CC(A)}$  as low as 0.6 V without degradation of system performance. The PI6ULS5V9627A contains two bidirectional open-drain buffers specifically designed to support up-translation/down-translation between the low voltage (as low as 0.6 V) and a 2.5 V, 3.3 V or 5 V I $^2$ C-bus or SMBus. All inputs and I/Os are overvoltage tolerant to 5.5 V even when the device is unpowered ( $V_{CC(B)}$ ) and/or  $V_{CC(A)}$  = 0 V).

The PI6ULS5V9627A includes a power-up circuit that keeps the output drivers turned off until  $V_{CC(B)}$  is above 2.2 V and until after the internal reference circuits have settled at about 400  $\mu$ s, and the  $V_{CC(A)}$  is above 0.6 V.  $V_{CC(B)}$  and  $V_{CC(A)}$  can be applied in any sequence at power-up.

The PCA9627A includes a VCC(A) over-voltage disable that turns the channel off if 0.4VCC(A) + 0.8 V > VCC(B). The PCA9627A logic and all I/Os are powered by the VCC(B) pins.

The B-side drivers operate from 2.2V to 5.5V. The output low level of port B internal buffer is approximately 0.55 V, while the input voltage must be 90mV lower (0.45V) or even more lower. The nearly 0.5V low signal is called a buffered low. When the B-side I/O is driven low internally, the low is not recognized as a low by the input. This feature prevents a lockup condition from occurring when the input low condition is released. This type of design on B port prevents it from being used in series with another PI6ULS5V9627A (B side) or similar devices, because they don't recognize buffer low signals as a valid low.

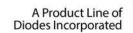
The A-side drivers operate from 0.6V to 5.5V. The output low level of port A internal buffer is nearly 0V, while the input low level is set at  $0.35V_{CC(A)}$  to accommodate the need for a lower LOW level in systems where the low voltage side supply voltage is as low as 0.6~V. Port A of two or more PI6ULS5V9627As can be connected together to allow a star topography with port A on the common bus. And port A can be connected directly to any other buffer with static or dynamic offset voltage. Multiple PI6ULS5V9627As can be connected in series, port A to port B, with no build-up in offset voltage with only time of flight delays to consider.

After power-up and with the EN HIGH, a LOW level on port A (below  $0.3V_{\rm CC(A)}$ ) turns the corresponding port B driver (either SDA or SCL) on and drives port B down to about 0.55V. When port A rises above  $0.3V_{\rm CC(A)}$ , the port B pull-down driver is turned off and the external pull-up resistor pulls the pin HIGH. When port B falls first and goes below 0.4 V, the port A driver is turned on and port A pulls down to about 0 V. The port A pull-down is not enabled unless the port B voltage goes below 0.4 V. If the port B low voltage goes below 0.4 V, the port B pull-down driver is enabled and port B will only be able to rise to 0.55 V until port A rises above  $0.3V_{\rm CC(A)}$ , then port B will continue to rise being pulled up by the external pull-up resistor. The  $V_{\rm CC(A)}$  is only used to provide the  $0.35V_{\rm CC(A)}$  reference to the port A input comparators and for the power good detect circuit. The PI6ULS5V9627A logic and all I/Os are powered by the  $V_{\rm CC(B)}$  pin.

The EN pin is active HIGH with thresholds referenced to  $V_{\text{CC(B)}}$  and an internal pull-up to  $V_{\text{CC(B)}}$  that maintains the device active unless the user selects to disable the repeater to isolate a badly behaved slave on power-up until after the system power-up reset. It should never change state during an I<sup>2</sup>C-bus operation because disabling during a bus operation will hang the bus and enabling part way through a bus cycle could confuse the I<sup>2</sup>C-bus parts being enabled. The enable does not switch the internal reference circuits so the 400 $\mu$ s delay is only seen when  $V_{\text{CC(B)}}$  comes up. The enable pin should only change state when the global bus and the repeater port are in an idle state to prevent system failures.

As with the standard I<sup>2</sup>C-bus system, pull-up resistors are required to provide the logic HIGH levels on the buffered bus (standard







open-collector configuration of the I<sup>2</sup>C-bus). The size of these pull-up resistors depends on the system, but each side of the repeater must have a pull-up resistor. This part designed to work with Standard mode, Fast-mode and Fast-mode Plus I<sup>2</sup>C-bus devices in addition to SMBus devices. Standard mode and Fast-mode I<sup>2</sup>C-bus devices only specify 3mA output drive; this limits the termination current to 3mA in a generic I<sup>2</sup>C-bus system where Standard-mode devices, Fast-mode devices and multiple masters are possible. When only Fast-mode Plus devices are used with 30mA at 5V drive strength, then lower value pull-up resistors can be used. The B-side RC should not be less than 67.5ns because shorter RCs increase the turnaround bounce when the B-side transitions from being externally driven to pulled down by its offset buffer.

## **Application Information**

A typical application is shown in Figure 6. In this example, the system master is running on a 3.3V I<sup>2</sup>C-bus while the slave is connected to a 1.2V bus. Both buses run at 1MHz. Master devices can be placed on either bus.

The PI6ULS5V9627A is 5V tolerant, so it does not require any additional circuitry to translate between 0.6V to 5.5V bus voltages and 2.2V to 5.5V bus voltages.

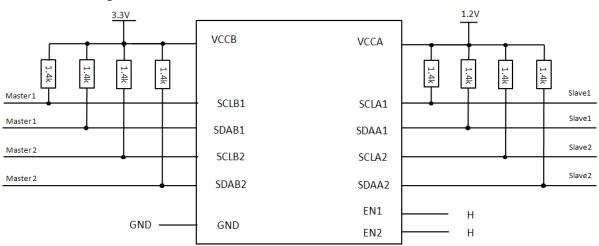


Figure 6: Typical Application

When port A of the PI6ULS5V9627A is pulled LOW by a driver on the  $I^2C$ -bus, a comparator detects the falling edge when it goes below  $0.3V_{CC(A)}$  and causes the internal driver on port B to turn on, causing port B to pull down to about 0.5 V. When port B of the PI6ULS5V9627A falls, first a CMOS hysteresis type input detects the falling edge and causes the internal driver on port A to turn on and pull the port A pin down to ground. In order to illustrate what would be seen in a typical application, refer to Figure 9 and Figure 10. If the bus master in Figure 6 were to write to the slave through the PI6ULS5V9627A, waveforms shown in Figure 9 would be observed on the A bus. This looks like a normal  $I^2C$ -bus transmission except that the HIGH level may be as low as 0.6 V, and the turn on and turn off of the acknowledge signals are slightly delayed.

On the B bus side of the PI6ULS5V9627A, the clock and data lines would have a positive offset from ground equal to the  $V_{OL}$  of the PI6ULS5V9627A. After the eighth clock pulse, the data line will be pulled to the  $V_{OL}$  of the slave device which is very close to ground in this example. At the end of the acknowledge, the level rises only to the LOW level set by the driver in the PI6ULS5V9627A for a short delay while the A bus side rises above  $0.3V_{CC(A)}$  then it continues HIGH. It is important to note that any arbitration or clock stretching events require that the LOW level on the B bus side at the input of the PI6ULS5V9627A ( $V_{IL}$ ) be at or below 0.4~V to be recognized by the PI6ULS5V9627A and then transmitted to the A bus side.

Multiple PI6ULS5V9627A port A sides can be connected in a star configuration (Figure 7), allowing all nodes to communicate with each other.

Multiple PI6ULS5V9627As can be connected in series as long as port A is connected to port B (Figure 8). I<sup>2</sup>C-bus slave devices can be connected to any of the bus segments. The number of devices that can be connected in series is limited by repeater delay/time-of-flight considerations on the maximum bus speed requirements.



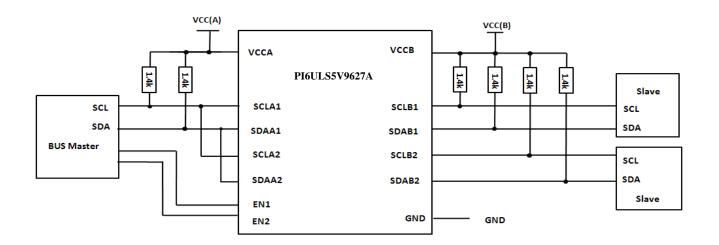


Figure 7: Typical Star Application

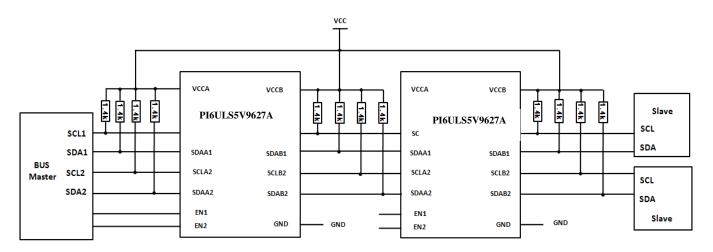
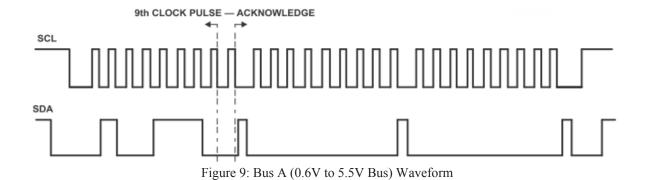


Figure 8: Typical Series Application





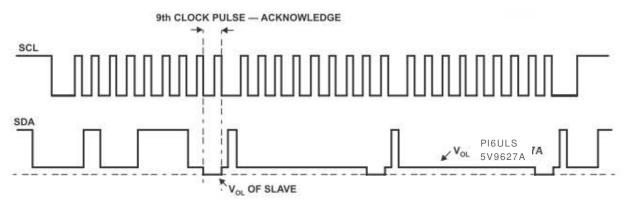
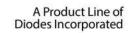


Figure:10: Bus B (2.2V to 5.5V Bus) Waveform

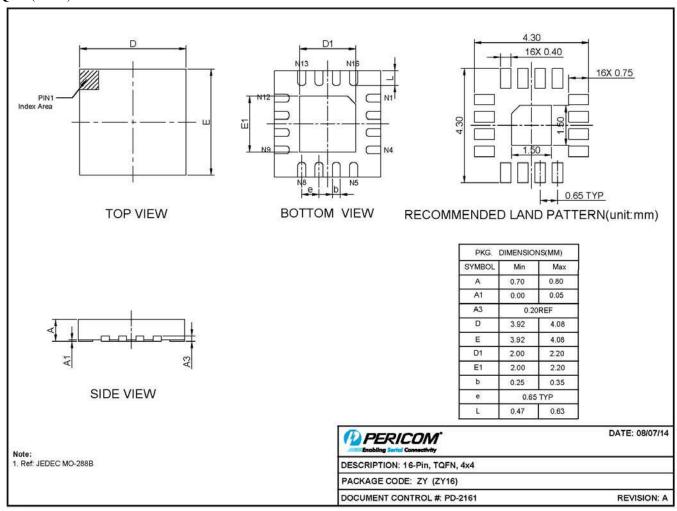






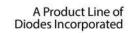
## **Mechanical Information**

## TQFN (ZY16)



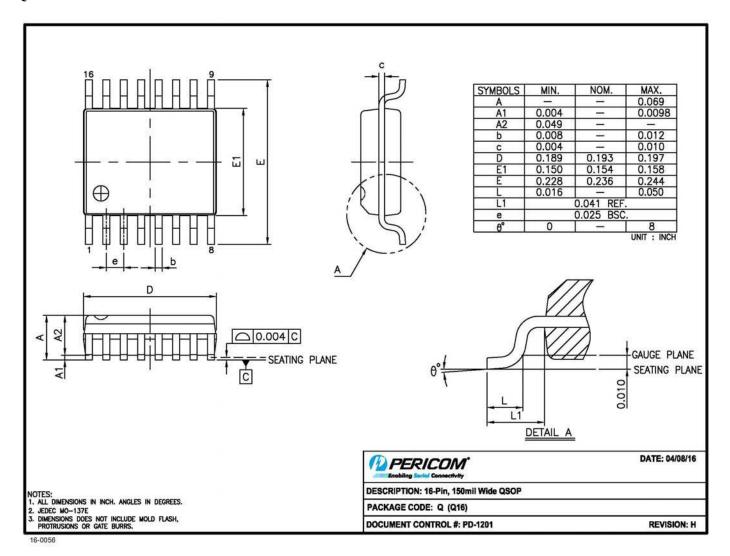
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**QSOP-16L** 



## For latest package info.

 $please\ check: http://www.diodes.com/design/support/packaging/pericom-packaging/packaging-mechanicals-and-thermal-characteristics/packaging/packaging-mechanicals-and-thermal-characteristics/packaging-packaging-mechanicals-and-thermal-characteristics/packaging-pack$ 

## **Ordering Information**

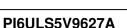
Part No.	Package Code	Package
PI6ULS5V9627AZYEX	ZY	16-pin, 4x4 (TQFN), Tape & Reel
PI6ULS5V9627AQEX Q		16-pin, 150mil Wide (QSOP), Tape & Reel

### Notes:

- Thermal characteristics can be found on the company web site at www.diodes.com/design/support/packaging/
- E = Pb-free and Green
- X suffix = Tape/Reel







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