

Tsi308™ HyperTransport to PCI/X User Manual

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About this Document

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- "Related Information" on page 15
- "Revision History" on page 15

Scope

The *Tsi308 HyperTransport to PCI/X User Manual* discusses the features, capabilities, and configuration requirements for the Tsi308. It is intended for hardware and software engineers who are designing system interconnect applications with these devices.

Document Conventions

This document uses a variety of conventions to establish consistency and to help you quickly locate information of interest. These conventions are briefly discussed in the following sections.

Non-differential Signal Notation

Non-differential signals are either active-low or active-high. An active-low signal has an active state of logic 0 (or the lower voltage level), and is denoted by a lowercase \hat{m} . An active-high signal has an active state of logic 1 (or the higher voltage level), and is not denoted by a special character. The following table illustrates the non-differential signal naming convention.

Object Size Notation

This document uses the following object size notation:

- A *byte* is an 8-bit object.
- A *word* is a 16-bit object.
- ï A *doubleword* (Dword) is a 32-bit object.

Numeric Notation

- Hexadecimal numbers are denoted by the prefix $0x$. For example, 0x04.
- Binary numbers are denoted by the prefix *0b*. For example, 0b010.
- Registers that have multiple iterations are denoted by $\{x,y\}$ in their names; where *x* is first register and address, and y is the last register and address. For example, $REG{0..3}$ indicates there are four versions of the register at different addresses: REG0, REG1, REG2, and REG3.

Symbols

This symbol indicates a basic design concept or information considered helpful.

This symbol indicates important configuration information or suggestions.

This symbol indicates procedures or operating levels that may result in misuse or damage to the device.

Document Status Information

User manuals are classified as Advance, Preliminary, or Final:

- Advance Contains information that is subject to change, and is available once prototypes are released to customers.
- Preliminary Contains information about a product that is near production-ready, and is revised as required.
- Final Contains information about a final, customer-ready product, and is available once the product is released to production.

Related Information

The following documents contain useful reference information for using this manual:

ï Tsi308 Device Errata and Design Notes

Revision History

80D4000_MA001_04, Formal, September 2009

This document was rebranded as IDT. It does not include any technical changes.

80D4000_MA001_03, Preliminary, January 2007

This is the current release of the user manual. There have been slight modifications throughout the manual.

1. Functional Description

This chapter discusses the following topics about the Tsi308:

- "Overview" on page 18
- "Features" on page 19
- "HyperTransport Interface" on page 20
- "PCI-X Interface" on page 21
- "Interrupt Controller" on page 23
- "Interface Levels" on page 23
- "Clocking" on page 24
- "Reset" on page 24

1.1 Overview

The Tsi308 is HyperTransport^{M}-to-PCI-X Bridge that interfaces the new generation of HT based microprocessors and micro controllers to PCI or PCI-X based peripherals. It also connects HT based hosts to HT based peripherals.

The Tsi308 can be configured to support either single 64-bit PCI-X bus or two 32-bit PCI-X buses. The Tsi308 implements two bi-directional 8-bit HyperTransport™ interfaces that provide 1200 MByte per second of bandwidth in each direction. Up to 31 devices can be daisy-chained to build higher capacity systems with multiple PCI-X busses and HT based peripherals. A fairness algorithm allocates bandwidth among devices, thereby eliminating starvation of bridges at the end of the chain.

The Tsi308 breathes new life into systems that are encumbered by the limits of traditional PCI or PCI-X based fabrics. It reduces the time to market, design complexity and system costs of PCI-X and HT based systems.

The block diagram for Tsi308 is shown [Figure 1 on page 18](#page-17-1).

Figure 1: Tsi308 Block Diagram

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1.2 Features

The following sections describe the features of Tsi308.

1.2.1 General features

- Two Bidirectional 8-bit HyperTransport[™] interfaces
	- ó Up to 600MHz DDR (Double Data Rate) for peak bandwidth of 1200 MB/s simultaneously in each direction
- Complies with Revision 1.05 of HyperTransport^{M} I/O Link Specification
	- $-$ Maximum link width supported is only 8-bits
	- $-$ Supports asymmetric link widths and frequencies
- Tunnels between the two HyperTransport^{M} interfaces
- Can be configured as single-ended cave device with only one link active
- The HT interfaces support double hosted chain (Host CPU on each port)
- Single 64-bit PCI-X bus or two 32-bit PCI-X buses
	- ó Device emulates two HT link devices (with virtual internal tunnel in dual PCI-X mode)
	- $-$ Implements two independent sets of CSRs in dual PCI-X mode
	- $-$ PCI-X bus can also operate in traditional PCI mode
	- ó Operating frequencies and mode of the two PCI-X buses are independently selectable
	- \sim Supports 50, 66, 100 and 133 MHZ in PCI-X mode
	- \sim Supports 25, 33, 50 and 66 MHZ in PCI mode
	- $-$ PCI-X mode complies to Revision 1.0b of PCI-X Addendum to the PCI Local Bus Specification
	- ó PCI mode complies to Revision 2.2 of PCI Local Bus Specification
- Supports Tsi301 software backward compatibility mode through hardware strap setting
- Supports daisy-chaining up to 31 devices. The bandwidth is shared among the devices using a fairness algorithm
- Programmable interrupt controller with up to 10 interrupts per PCI-X port
- Built-in 2-level PCI-X arbiter with support for up to 6 devices
	- $-$ Also supports external arbiter
- Transaction forwarding for the following commands
	- $-$ All I/O and memory commands
	- $-$ Type 1 to Type 1 configuration commands (downstream only)
- *o Type 1 to Type 0 configuration commands (downstream only)*
	- Internal buffers to support high-speed operation, including:
		- ó 1536-bytes HT forwarding (512 bytes each for posted, non-posted and response)
		- $-$ Following buffers are supported for each PCI-X port
			- 1024-bytes upstream write (posted)
			- 2048-bytes upstream read (non-posted), among up to four outstanding requests
			- 512-bytes downstream write (posted)
			- 512-bytes downstream read (non-posted)
	- ï 64-bit memory mapped space and 25-bit I/O space
	- ï 64-bit Address Remapping (downstream) and one DMA Window (upstream)
	- Full UnitID Clumping support
	- Supports 64-bit Address Extension
	- Evaluation board available with firmware and software drivers
	- ï 6 Watt max, 1.8V core, 1.2V HT I/O, 3.3V PCI I/O
	- Optional 5V tolerant PCI I/O in standard PCI mode while operating at 25 or 33 MHZ
	- 388-pin HSBGA package
	- Compatible with x86 systems
	- Supports Online Insertion and Removal
	- Supports Boundary scan
	- Software and Hardware compatibility Revision A $&$ Revision B

1.3 HyperTransport Interface

The Tsi308 HyperTransport-to-PCI-X bridge primary interface is a HyperTransport tunnel. The primary interface is compliant with HyperTransport™ I/O Link Specification, Revision 1.05. The interface contains two HyperTransport links, which allow the connection of multiple bridge, chips in a daisy-chain configuration. As shown in the figure above Tsi308 can be configured to behave as two independent tunnel devices that are connected through a virtual internal tunnel. In this mode each tunnel device can host a 32-bit PCI-X bus. The programming in this mode is pretty much transparent to software in a way that software treats them as if they are independent devices or chips.

Each HyperTransport link has an 8-bit DDR transmit and an 8-bit DDR receive port running at clock speeds up to 600 MHz, allowing for raw bandwidth of 1200 MB/s simultaneously in each direction.

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- For testing or connection to slower devices, the Tsi308 may be programmed to operate at slower link clock rates
- The Tsi308 supports both the synchronous and asynchronous modes of link initialization

1.4 PCI-X Interface

The Tsi308 secondary interface is a 64-bit, 133 MHz capable PCI-X bus that can be configured to have two completely independent 32-bit buses in split bus mode including buffer space and transaction handling. The two PCI-X ports are identical in split bus mode and the subsequent description applies to each port. The PCI-X interface can operate at 50, 66, 100 and 133 MHz, which can also operate at 25, 33, 50 and 66 MHz while operating in traditional PCI mode. Additionally PCI-X bus can be configured for compatibility with 3.3V or 5.0V operation while operating at up to 33 MHz in traditional PCI mode. At higher frequencies of PCI or while in PCI-X mode only 3.3V is supported.

The Tsi308 supports the full 64-bit memory-mapped space and 25-bit I/O space described in HyperTransport I/O Link Specification, Revision 1.05. In addition device supports 64-bit address remapping capability and a single upstream DMA window. PCI dual address cycle (DAC) support is provided both inbound and outbound to support memory-mapped space.

- The Tsi308 supports configuration accesses to devices 0-15, using Address/Data bits 16-31 for IDSEL#.
- ï The Tsi308 implements all parity and error checking features described in *PCI Local Bus Specification, Revision 2.2*.

1.4.1 PCI-X Master

As a PCI-X master, the Tsi308 chip can generate MemRd, MemWr, ConfigRd and ConfigWr cycles.

- The Tsi308 does not implement a cacheline size register and does not prefetch to PCI, so it never generates MemRdLine, MemRdMult or MemWrInv cycles.
- ï The Tsi308 does generate *Memory Read Block* but does not generate *Memory Write Block* cycles in PCI-X mode
- The Tsi308 does not support a Southbridge connection to PCI bus, so it never generates INTA cycles.
- The Tsi308 does not support burst I/O and burst Configuration cycles initiated from Host. These transactions are target aborted inside the chip and does not appear on the PCI-X bus.

PCI-X master cycles that are retried or disconnected on the PCI-X bus are reissued locally by the Tsi308 until they complete. The Tsi308 can track up to two outstanding requests in the Outbound Request Controller, of which one is reserved for posted requests. The other one is used for either posted or non-posted. The reserved posted buffer allows the passage of posted requests in case of blockage of non-posted requests.

In addition to two request-tracking buffers, Tsi308's PCI-X port has 512 byte buffer spaces each for posted and non-posted requests.

1.4.2 PCI-X Slave

As a PCI-X slave, the Tsi308 can respond to all types of memory and I/O cycles. However, the Tsi308 never responds to PCI-X configuration cycles.

- The Tsi308 employs medium DEVSEL# timing.
- All PCI-X slave writes are posted excluding I/O writes which is non-posted.
- A total of 1024 bytes of buffering is provided on chip for posted requests
- All PCI-X slave reads are implemented as delayed requests (PCI) or split (PCI-X), with up to four requests outstanding at once and a maximum of 512 byte buffering is provided for each outstanding request to store the response data received from HT.
- Fast back to back transactions are supported.

Prefetching is supported for all flavors of memory read cycle while operating in standard PCI mode, which separate prefetch controls for each cycle type and a maximum prefetch per read of 512 bytes. Prefetching may be done once at the beginning of each read, or it may be enabled to continuously issue requests as data is drained to PCI. All prefetch data is discarded when the read disconnects on the PCI bus. The bridge chip provides buffer space for a total of 2048 bytes of read prefetch data per PCI-X port.

While operating in PCI-X mode, Tsi308 fetches only enough bytes to satisfy the byte count field that appears in the attribute phase of all PCI-X burst transactions. The Tsi308 can support any sized request up to 4096 bytes as specified in [2]. However since Tsi308 has only 512-byte buffer to store the read data per request, it will continue to fetch data from HyperTransport as the buffer is drained on to PCI-X in chunks of single ADB.

1.4.3 PCI-X Arbiter

The Tsi308 implements an on-chip PCI Arbiter with 6 request/grant pairs. The request/grant pairs include a high-priority set for the on-chip PCI master, and five symmetrical sets for external device use.

All connections to the arbiter are through external pins, to use internal arbiter user has to route request/grant outputs back into chip connecting to any of the six request/grant pairs. So Tsi308 can automatically be configured to interface to external arbiter.

1.5 Interrupt Controller

The Tsi308 implements a HyperTransport interrupt controller. It supports 10 external interrupt sources per PCI port. To program interrupts Tsi308 implements *Interrupt Discovery and Configuration Capability Block* and associated *Interrupt Definition Registers* for each interrupt source. Each interrupt can independently be enabled and programmed to be level or edge-triggered and active high or low. In order for to be software compatible to previous generation Tsi301 chip, Tsi308 also implements an alternate register map to program the interrupts in a non-standard way. However these registers are visible to software only when Tsi308 is operating in Tsi301 software compatible mode by hardware strap settings and this is the only means to program interrupts in Tsi301 mode.

1.6 Interface Levels

A complete pinout of the Tsi308 is provided in the *Signals Chapter*. The grouping of signal types is shown in [Table 1](#page-22-2).

Table 1: HyperTransport PCI-X Bridge Interface Voltages

1.7 Clocking

During functional operation, the Tsi308's reference clocks (P0_CLK and P1_CLK) come from respective PCI-X bus clocks. These clocks are received from same sources that drive clocks to devices on the bridge's PCI-X buses and are nominally in phase with them; although they may be delayed relative to other PCI-X bus clocks. The P1 CLK is only used to clock the PCI-X interface logic of second PCI-X port (PCI_B) while the device is operating in split bus mode. The reference clock frequencies and bus mode (traditional PCI or PCI-X) are indicated by Px_M66EN, Px_PCIX_N and Px_133_N input pins where x denotes PCI-X bus (0 for PCI_A and 1 for PCI $\,$ B). The Px $\,$ PCIX $\,$ N and Px $\,$ 133 $\,$ N are normal TTL level signals derived from standard 3-state add-in card connector pin PCIXCAP. Since Tsi308 does not decode PCIXCAP, user has to implement an external three-level Comparator circuitry to generate Px PCIX N and Px 133 N. A reference circuit can be found in [3]. Though three pins above indicate operating mode (PCI or PCI-X) and frequency group (33MHz or 66 MHz or 133 MHz), Tsi308 needs exact operating frequency of a given bus to generate internal clocks as well as to generate PCI-X Initialization Pattern for devices on PCI-X bus as specified in [3]. This is done through hardware straps. These straps are sampled using combinational logic while warm/cold reset is in progress and used to combinationally generate PCI-X initialization pattern that is sampled by devices on PCI-X bus at the rising edge of PCI reset. Refer Chapter 4 for more details on Clocking and Hardware strap settings.

1.8 Reset

All the internal resets of Tsi308 and resets for secondary PCI-X ports are derived from HyperTransport PWROK and RESET# signals. The combination of these two signals defines ColdReset and WarmReset windows on HyperTransport chain. While PWROK is implemented as input-only, RESET# is implemented as in-out in Tsi308. The asserted state of RESET# is stretched by Tsi308 and released after internal PLLs are locked. The PCI-X and CORE PLLs are only reset upon ColdReset but WarmReset resets HyperTransport PLLs. This way software could re-program link frequencies and issue WarmReset for new frequencies to take effect.

2. Interface Operation

This chapter discusses the following topics about the Tsi308:

- "Overview" on page 25
- "HyperTransport Interface" on page 27
- "Outbound Transactions" on page 34
- "Inbound Transactions" on page 36
- \cdot "PCI-X Arbiter" on page 41
- "Online Insertion and Removal (OIR)" on page 42
- "LDTSTOP# Support" on page 42
- "Power Management" on page 43
- "Reset" on page 43
- "Error Handling" on page 45
- "Test Features" on page 52

2.1 Overview

This chapter details the operation of the HyperTransport-to- PCI-X Bridge chip.

2.2 HyperTransport Interface

The Tsi308 HyperTransport interface consists of two identical link interfaces, each with a HyperTransport transmitter and receiver. Some central reset and error-handling logic is shared between the two links.

In the HyperTransport protocol, all logical packet transfer is between HyperTransport slaves and the host. Direct peer-to-peer communication is not allowed. To support peer-to-peer operations, packets are reflected through the host. Packets issued from the host to a HyperTransport slave are defined to travel downstream on the HyperTransport chain. Packets issued from a HyperTransport slave to the host are defined to travel upstream. Intermediate nodes in the daisy chain forward packets from link to link until they reach their final destination, which accepts the packet.

Link interfaces in the Tsi308 are symmetrical, which allows connection to either bridge link toward a host. The Tsi308 also supports being placed in a double-hosted chain with hosts on both ends.

Figure 3: Single HyperTransport Link Interface Block Diagram

2.2.1 HyperTransport Packet Reception

Packets received from a HyperTransport link are decoded and routed to appropriate destination port where they are stored in Receive (Rx) buffers for subsequent transmission. HyperTransport flow control algorithm guarantees that no packet is received without buffer space to store it. Packet contents are divided into command information (including address) and data, with separate buffers for each.

For packets arriving from HT Links, each destination port (forward HT Link, PCI_A and PCI B) has dedicated command and data buffers for two possible source ports (Both HT links could source data to any of the two PCI-X ports and each of the two HT Link Ports can receive forwarded data from other HT Link Port and reflected traffic) statically partitioned among the three virtual channels (posted, non-posted and response), with each channel allocated space to hold eight commands and eight data packets. However packets originated on PCI-X bus is locally buffered in respective PCI-X Ports and delivered directly to the appropriate HT Link.

2.2.1.1 Packet Decode

As packets arrive from a HT link, the associated commands and addresses are decoded to determine if the Tsi308 is the packet target on HyperTransport chain and routed to the appropriate port where they are locally stored. Packets are checked in the destination port for ordering collisions against other packets resident in the buffers. The decode and collision results are stored in the buffers with the packets. The packet routing is as follows:

- Packets received on the HyperTransport interface may be routed to internal logic (CSRs), including PCI-X interfaces. This is "accepting" a packet.
- Packets may also be routed to the other HyperTransport link interface for transmission to the next device in the chain. This is "forwarding" a packet.
- A particular packet may be accepted, forwarded, or both.

The Tsi308 HyperTransport packet decode first determines whether the incoming packet is traveling upstream or downstream. This determination is based on the packet source information contained in the packet itself, not on which link is the upstream or downstream link.

- Upstream packets are always forwarded toward the host and are never accepted by the Tsi308 chip.
- Downstream RdSized and WrSized request packet addresses are decoded according to the HyperTransport Address Map described in Section 3.3 and are accepted if they match any Tsi308 address ranges of any PCI-X ports or internal CSRs.
- Downstream WrSized and RdSized that do not match any of the Tsi308 address ranges are forwarded to the next device in the chain.
- Broadcast request addresses are also decoded and accepted if they match a Tsi308 address range. However, these packets are also always forwarded.
- Fence and Flush requests are never accepted by Tsi308 and are always forwarded.
- Downstream response packets are accepted if their UnitID field matches the value in the BaseUnitID field of any of the Tsi308 LdtCmd register; otherwise, they are forwarded.

2.2.1.2 Collision Checking and Ordering

Collision checking is performed according to the HyperTransport protocol to determine if incoming packets are required to stay ordered behind packets already in the Rx buffers. Only packets headed to the same accept or forward destination may have ordering requirements. If a packet has an ordering collision, it may not be issued from the Rx buffers until the packet with which it collided has both been issued and reached an appropriate commit point to guarantee ordering. This ordering point varies by destination.

Because the Outbound Request Controller (ORC) can reorder requests, requests accepted by the Tsi308 may not be committed until they are retired by ORC. For accesses to PCI-X bus, this means that the request reached the PCI-X bus and all data was transferred. Packets forwarded from one link to the other are either streamed – meaning that the transmission may start before the whole packet is received or a complete packet is first stored and then forwarded, avoiding possible "holes" on transmitting link if transmitting link bandwidth (link frequency * link width) is an order faster than that of receiving link. Since there is no specification defined way to determine the bandwidth of receiving link, Tsi308 implements a proprietary control bit (StoreForward) per link in its CSR which system software can program that tells the Tsi308 whether to stream (low latency) or not. It is the Rx Command Buffer's responsibility to ensure that required packet ordering is maintained so packets can be committed as soon as they are passed to the link controller.

Packets that do not have any ordering requirements may leave the Rx buffers in a different order than they reached them, both among and within virtual channels. In general, Rx buffers select a packet choosing the oldest non-blocked packet in each channel to a given destination.

HyperTransport Ordering

The [Table 2](#page-29-0) lists ordering implemented in Tsi308 for packets issued from PCI/PCI-X to HyperTransport link and for packets forwarded from one HT link to other HT link.

Row Pass Column	Posted Request	Non-posted Request	Response	
Posted Request	No	Yes	Yes	
Non-posted Request	No	No	No	
Response	No	Yes	No	

Table 2: HyperTransport Ordering

PCI Ordering

The [Table 3](#page-30-1) lists the ordering implemented in Tsi308 for packets traveling to PCI from HT interface.

Row Pass Column?	Posted Memory Write (PMW)	Delayed Read Request (DRR)	Delayed Write Request (DWR)	Delayed Read Completion (DRC)	Delayed Write Completion (DWC)
PMW	No.	Yes	Yes	Yes	Yes
DRR	No.	No.	No	No.	No.
DWR	No.	No.	No	No.	No.
DRC	No	Yes	Yes	No.	No.
DWC	No	Yes	Yes	No.	No.

Table 3: PCI Bus Transaction Ordering

PCI-X Ordering

The [Table 3](#page-30-1) lists the ordering implemented in Tsi308 for packets traveling to PCI-X from HT interface.

Table 4: PCI-X Bus Transaction Ordering

Row Pass Column?	Posted Memory Write (PMW)	Split Read Request (SRR)	Split Write Request (SWR)	Split Read Completion (SRC)	Split Write Completion (SWC)
PMW	No	Yes	Yes	Yes	Yes
SRR	No.	No	No.	No.	No.
SWR	No	No	No.	No.	No.
SRC	No	Yes	Yes	No.	No.
SWC	No	Yes	Yes	No	No.

2.2.2 HyperTransport Address Map

The Tsi308 implements a single flat 64-bit address space for all accesses. All address spaces that can be reached from HyperTransport are mapped into this space. The Tsi308 checks addresses on incoming packets in each space for ranges that it accepts.

2.2.2.1 Memory Mapped Space

The HyperTransport specification places Memory Mapped Space in the address range of 0000_0000_0000_00000 to 0000_00FC_FFFF_FFFFh. The Tsi308 accepts two ranges within this space, as enabled by MemSpaceEn in the Command (Cmd) CSR, consisting of the following:

- Memory Space, defined by the MemBase and MemLimit CSRs.
- Prefetchable Memory Space, defined by the PrefMemBaseUpper/PrefMemBase and PrefMemLimitUpper/PrefMemLimit CSRs.

Setting the VgaEn bit in the Bridge Control CSR creates an additional window of 00_000A_0000h to 00_000B_FFFFh, which is also accepted. The Tsi308 never does prefetching to PCI, so the prefetchable/nonprefetchable attribute of these ranges does not matter. RdSized requests to these ranges result in MemRd requests on PCI-X bus. WrSized requests to these ranges result in MemWr requests on the PCI-X bus. If above 4GB, addresses are passed straight through as a Dual Address Cycle (DAC).

2.2.2.2 I/O Space

The HyperTransport specification places PCI-X I/O space in the address range of 0000 OOFD FC00 0000h to 0000 OOFD FDFF FFFFh. The Tsi308 strips the top 39 bits off of the addresses in this range.

- If enabled by I/OspaceEn in the Cmd CSR, the Tsi308 accepts requests that fall in the range defined by the I/O Base and I/O Range Base Upper, and I/O Limit and I/O Range Limit Upper CSRs.
- If set, the IsaEn bit in Bridge Control CSR creates a series of holes (the top 768 bytes of each 1 KB block in the low 64 KB) in this space that the Tsi308 does not accept.
- Setting the VgaEn bit in the Bridge Control CSR creates an additional set of windows (all addresses in the low 64 KB where the bottom 10 bits are in the ranges $3B0h - 3BBh$ or 3C0h ñ 3DFh), which the Tsi308 accepts. Accepted RdSized requests result in IoRd requests on PCI-X bus, and WrSized requests result in IoWr requests, with the bottom 25 bits of the HyperTransport address passed through. Bits 31:26 are 0.

2.2.2.3 Configuration Space

The HyperTransport specification places PCI-X configuration space in the address range of 0000 OOFD FE00 0000h to 0000 OOFD FFFF FFFFh. Address bit 24 identifies requests as Type 0 or Type 1 configuration requests.

Type 0 requests are accepted and routed to the Tsi308 internal configuration registers if their device number (bits 15:11) matches the value of BaseUnitID in the HyperTransport Command CSR.

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Type 1 requests are accepted if their bus number (bits $23:16$) falls within the range defined by the Secondary Bus Number and Subordinate Bus Number CSRs, inclusive. Type 1 requests are routed to PCI-X, as ConfigRd or ConfigWr cycles.

A Type 1 request with a bus number that exactly matches the Secondary Bus Number CSR becomes a PCI-X Type 0 configuration request, with AD[deviceNumber + 16] set.

Type 1 requests with a bus number greater than Secondary Bus Number but less than or equal to Subordinate Bus number are passed on to PCI-X as Type 1 configuration cycles. Bits 23:2 of the address are left unchanged. The Tsi308 does not support the Type 1 configuration to Special cycle mapping.

2.2.2.4 Interrupt Space

The HyperTransport specification places interrupt space in the address range of 0000_00FD_F800_0000h to 0000_00FD_F8FF_FFFFh. The Tsi308 only accepts End of Interrupt (EOI) requests in this range, which should always be broadcasts. These EOI requests are routed to the interrupt controller.

2.2.3 HyperTransport Address Remap

Since HyperTransport technology is meant to provide a high-bandwidth backbone for I/O systems, which are likely to contain a variety of other buses with varying addressing capabilities, the HyperTransport specification defines mechanism to remap the HyperTransport addresses to locally defined addresses of other buses allowing mapping of the smaller address spaces of individual buses into different locations within the HyperTransport technology address map.

To support this, Tsi308 implements 64-bit Address Remapping Capability as specified in [1] with single upstream DMA window. This DMA window can also be used to set specific attributes in packets that originate on PCI-X and also fall inside the address ranges defined by the DMA window. An example of this attribute is that the user can program Tsi308 to set Isoc bit for all the packets that pass through the DMA window.

Address Remapping is enabled through CSRs and is only applicable to packets traveling from/to PCI-X. It is not applicable to forward packets

2.2.4 HyperTransport Packet Transmission

The HyperTransport packet generator logic is essentially a large arbiter/multiplexer that formats and combines packets from each of the three virtual channels issued from within the Tsi308. The output stream is combined with the stream of packets forwarded through the Tsi308 from the far HyperTransport link. This later multiplexing is also used to insert NOP/buffer release messages to the transmitter on the link's other end.

Packet transmission is paced by the transmit buffer counters maintained in each virtual channel for both command/address and data, as the HyperTransport specification describes. These counters are decremented as packets are transmitted and incremented as buffer release messages are received from the transmitter at the link's other end. Transmit buffer counters can be throttled using the Transmit Buffer Counter Maximum CSRs (C8h and CCh).

2.2.4.1 Packet Insertion

To prevent devices close to the host bridge from starving devices further out in the chain of bandwidth, the Tsi308 implements the packet insertion fairness algorithm described in [1]. This algorithm throttles the insertion rate of packets from Tsi308 relative to packets being forwarded and attempts to balance the packet insertion rates of all devices on the chain.

When Tsi308 is operating in Dual Device Mode(Split PCI-X Bus), fairness is implemented on cumulative basis wherein insert rate is computed for a single device in standard way and then actual insertion rate is doubled to account for two devices in single node/chip.

Insertion of buffer release messages is forced, even when the outgoing transmission stream is busy. Forcing allows traffic to flow through the Tsi308 continuously while maintaining a relatively small number of Rx buffers. The Tsi308 forces a buffer release message as soon as possible when an Rx buffer is freed, subject to the requirements of the HyperTransport protocol. The frequency of buffer release messages is limited under HyperTransport Transmit Control CSR (6Eh) to prevent them from occupying too much bandwidth in a busy stream. Throttling buffer releases clumps the released messages together and raises their efficiency.

In a single-hosted HyperTransport chain, the Tsi308 may be at the end of the chain furthest from the host and therefore have no downstream link connection. In this case, packets are routed to the End of Chain (EOC) logic in the unconnected link interface. The EOC logic drops responses and posted requests and generates Non-Existent Access (NXA) Error responses back into the receiver for non-posted requests. These error responses then get forwarded back to the other HyperTransport link interface's reflect path Rx buffers and back to the requesting device. Error logging for the dropped packets occurs in the Link Control Registers CSRs (44h and 48h) in Tsi301 mode or in the Link Error Register (4Dh and 51h) in standard HyperTransport mode.

2.3 Outbound Transactions

Outbound transactions to the Tsi308 are those accepted from the HyperTransport chain. All outbound requests go first from the HyperTransport link interface on which they are received to the Outbound Request Controller (ORC). This controller is responsible for issuing the request to the appropriate destination functional unit and for tracking the request state while it is outstanding.

The controller has two buffers, which allow state tracking for two outstanding requests. If both requests are outstanding in the controller at once, it rotates them in round-robin fashion to issue or reissue them. When the ORC filo fills, subsequent requests back up to the HyperTransport Rx buffers. Since the ORC doesn't guarantee ordering, the HyperTransport Rx buffers must not issue the second in an ordered pair of transactions until the first has completed.

The ORC is also responsible for managing space in the PCI response data buffer. All outbound non-posted requests, regardless of destination, must be allocated space in the response data buffer before they can be accepted from the Rx buffers through Data Mover (DM) by the ORC. Even though PCI response data buffer can hold two responses, ORC uses only request buffer for receiving non-posted requests from Rx command buffers, reserving the other buffer always for posted requests which in turn, provides the deadlock-avoidance guarantee required by [2] and [3] (non-posted requests are never allowed to block posted requests).

As requests complete at their destinations that fact is signaled back to the ORC (Normally PCI-X Master but PCI-X Target if request was non-posted and Tsi308 is operating in PCI-X mode), which allows the request buffer to be retired. If the request was non-posted, the transaction will require generation of a response to the host. ORC considers posted transactions as complete when the request completes at its destination and the buffer is retired. Non-posted transactions are complete when the response packet is issued to the HyperTransport transmit interface from which the request was received.

2.3.1 PCI-X Outbound Transactions

Outbound requests to PCI-X are handed to the PCI-X interface to be driven out to the bus. Write data comes from the Rx data buffers, Read data is returned from the bus and placed in the PCI Response Data Buffer.

When PCI-X bus A is configured as 64 bit at reset (P0 AD[14]), the interface automatically asserts P0_REQ64_N on all transactions for which it is legal. The PCI-X bus B can only be 32 bit.

The PCI-X interface supports Type 0 PCI configuration cycles to device numbers 0 through 15. Px AD[31:16] $(x = 0$ for PCI-X A, 1 for PCI-X B) are driven with a one-hot encoding during these configuration cycles, with bit 16 asserted for accesses to device number 0. This logic assumes that one Px_AD bit is connected to the IDSEL# pin on each PCI slot through a series resister on the board.

If a request is retried or disconnected on the PCI, that fact is reported back to the ORC. The controller finishes any data movement associated with the disconnected transaction and then reissues the request from the point of disconnection. It continues to reissue a request until it completes or until the retry timer for the request expires. Because the ORC can handle two outstanding requests at a time, transactions repeatedly retried or disconnected may be reordered or interleaved.

2.3.1.1 PCI-X Response Data Buffer

The PCI response data buffer contains read data returned from outbound reads to the PCI interface. This data buffer can hold a total of 64 bytes for one HyperTransport read requests.

Once the response is issued, the buffer is retired. Non-posted write requests still occupy space in the response data buffer, even though they have no read data.

2.3.1.2 End of Interrupt

When an interrupt is configured as level sensitive, upstream interrupt logic must respond to an interrupt request packet with an end of interrupt (EOI) packet. Until the EOI packet is received by the Tsi308, no new interrupt request packets will be generated by that interrupt pin.

2.4 Inbound Transactions

Inbound transactions are requests from the PCI-X bus or internal interrupt controller across HyperTransport to the host bridge. From there, they are routed to destinations behind the host bridge or reflected peer-to-peer back onto the HyperTransport chain. If the request is non-posted, the transaction also includes the response from the host bridge back to the original requesting unit.

The Tsi308 operates as a PCI-X target for requests from external PCI-X devices. All accepted requests are forwarded to the HyperTransport link interface leading to the host. Reads go through the delayed request buffers and are handled on the PCI-X bus as delayed requests when in standard PCI mode or as split requests when in PCI-X mode. All writes, except IO writes are posted into the posted request queue and allowed to immediately complete on the PCI-X bus.

2.4.1 PCI-X Address Map

Accesses on PCI-X bus are checked against the following ranges to determine whether the Tsi308 is the target of the access and should assert Px_DEVSEL_N to accept the request. The Tsi308 makes this determination with medium DEVSEL# timing. When PCI-X A bus is configured as 64-bit target at power up, the Tsi308 asserts P0_ACK64_N in response to P0_REQ64_N for requests it accepts.

Memory Mapped Cycles. The Tsi308 implements a 64-bit space for memory mapped accesses and decodes DAC accesses for addresses above 4 GB. While operating in Tsi301 compatible mode address bits above 39 are ignored and result in the 40-bit space aliasing through PCI's 64-bit memory mapped space.
Memory mapped addresses are compared to the range defined by the Memory Range Base Addr and Memory Range Limit Addr CSRs; and the range defined by the Prefetchable Memory Range Base Upper and Prefetchable Memory Range Base Addr, and Prefetchable Memory Range Limit Upper and Prefetchable Memory Range Limit Addr CSRs. Addresses that donít fall into any of these ranges are accepted for forwarding to HyperTransport as long as the MasterEn bit in Command CSR is set and bits $[39:32] \leq FCh$.

ï **I/O Cycles**. The Tsi308 implements a 64-bit space for I/O accesses. While operating in Tsi301 compatible mode, address bits above bit 24 are ignored and result in 25-bit space aliasing through PCI's 32-bit I/O space.

I/O addresses are compared to the range defined by I/O Range Base Upper and I/O Base, and I/O Range Limit Upper and I/O Limit CSRs. Accesses that miss the range are accepted for forwarding to HyperTransport, as long as MasterEn bit in the Command CSR is set.

ï **Configuration and Special Cycles**. The Tsi308 never acts as a target for configuration or special cycles on PCI-X bus.

2.4.2 PCI-X Posted Write Queue

The Tsi308 responds as a PCI-X write target to PCI-X Memory Write, Memory Write Invalidate, and I/O Write commands. All of these writes are posted to the HyperTransport chain except I/O writes which is non-posted. The Tsi308 never responds to Configuration Writes. A total of 1024 bytes of buffering for posted data is provided per PCI-X bus.

Memory Write and Memory Write Invalidate commands stream data into the chip, disconnecting either on 4-KB boundaries or when all of the internal buffer space is filled. The Tsi308 generates the largest HyperTransport write operations possible, issuing them continuously as the data for each write is received from PCI-X.

As the bandwidth of HyperTransport exceeds the bandwidth of PCI-X, it is expected that the internal buffers will not fill and memory writes will proceed continuously at the full bandwidth of PCI-X bus.

2.4.3 PCI-X Delayed/Split Request Buffers

The Tsi308 acts as a PCI-X target for PCI-X Memory Read, Memory Read Line (PCI), Memory Read Multiple (PCI), Memory Read Block (PCI-X), I/O Write and I/O Read commands. The Tsi308 never responds to configuration read or interrupt acknowledge accesses. All supported read transactions are implemented as delayed requests (PCI) or split requests (PCI-X).

Incoming requests are assigned to a delayed request buffer. There are four delayed request buffers, enabled under CSR control, allowing up to four PCI-X read requests to be in progress at one time. If no delayed request buffers are free, incoming requests are retried until one is available. Once the request is assigned to a buffer, the interface continues to retry it on the PCI bus while read requests are issued to the HyperTransport interface.

I/O writes are not allowed to stream and always disconnect after a single data beat on the PCI-X (32-bits). Each I/O write is issued to HyperTransport as an independent request.

2.4.4 Prefetching (PCI mode only)

While operating in standard PCI mode, the Tsi308 supports a variety of prefetching options configured under CSR control using two Read Control CSRs, Read Control 1 at 62h:60h and Read Control 2 at 5Eh:5Ch. Read Control 2 is applied for requests passing through DMA window and is reserved in Tsi301 compatible mode. However:

- I/O reads are never prefetchable.
- MemRdLines and MemRdMult may have prefetching individually configured.
- For systems in which MemRds are known to be side-effect free, MemReadPrefEn can be set to enable prefetching behavior for MemReads using the same parameters as MemRdLines.
- PrefEn can be used to globally enable or disable all prefetching.
- Nonprefetchable reads always request only the bytes required to satisfy the initial data beat of four or eight bytes on the PCI bus, which may result in either one or two HyperTransport requests.

Transactions for which prefetching is enabled issue a HyperTransport read for the remainder of the 64-byte aligned block containing the original request. These transactions also issue HyperTransport reads for the zero to seven complete 64-byte blocks following, as determined by the Read Control CSRs. The total number of reads that may be outstanding to HyperTransport at one time is limited by the Outbound Data Buffers.

When multiple reads to HyperTransport are issued for a single PCI read request due to prefetching or due to clear byte enables in a 64-bit nonprefetchable read on a 64-bit bus, each HyperTransport request is referred to as a subrequest of the PCI request. Each Delayed Request Buffer can track up to 8 subrequests at once. The total number of configured subrequests (number of enabled delayed request buffers * (the maximum number of subrequests each, rounded up to the next power of 2)) must not exceed the number of entries in the Outbound Data **Buffers**

2.4.5 Memory Read Block (PCI-X mode only)

MemRdBlk command of PCI-X is analogous to MemRdLine or MemRdMultiple of PCI.

Since PCI-X request provides the byte count to be satisfied during the attribute phase of a PCI-X read transaction, read data for PCI-X requests are never prefetched, instead data is read just enough to satisfy the byte count of the original request.

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Similar to in PCI mode, Tsi308 can handle up to four PCI-X read requests and they are assigned to Delayed Request Buffers as usual, however a single PCI-X read request can request up to 4K bytes. The Tsi308 implements a maximum of 512 byte buffering per request. For reads requesting greater than 512 bytes, Tsi308 throttles the requests issued on HyperTransport by issuing a subrequest on HyperTransport if at least 64 bytes of buffer space is available in Outbound Data Buffer allotted for that particular request until the byte count is satisfied. As the data arrives from HyperTransport into Outbound Data Buffer, PCI-X Master behaving as Split Completer connects on PCI-X bus and transfers the data to the original requester in chunks of 128 bytes disconnecting at naturally aligned 128-byte boundary or Allowable Disconnect Boundary (ADB).

2.4.6 SrcTags

The SrcTag for each HyperTransport read request is formed by concatenating the delayed request buffer number with the number of the HyperTransport subrequest being issued by that buffer.

2.4.7 Sequences

HyperTransport subrequests that are part of the same PCI-X request must be tagged with a matching nonzero SeqID to guarantee ordering at the target. This 4-bit SeqID is formed by concatenating a leading 1 (guaranteeing a nonzero result) with the 2-bit delayed request buffer number and one bit that toggles for each occupation of the delayed request buffer. The concatenation prevents consecutive PCI-X reads from being issued with the same SeqID and appearing to have HyperTransport ordering requirements.

2.4.8 Read Responses

As the read responses return from HyperTransport, the data is stored in the Outbound Data Buffers. Even though sequenced requests are guaranteed to reach the target in order, responses may be received from the target out of order. When all the data from the first HyperTransport requests is received (the amount required is controlled by the InitCount fields of the Read Control CSRs), the PCI interface ceases retrying the request. Read data is supplied from the buffers when the request is next reissued. Data streams to the PCI bus until the transaction is disconnected by the PCI master or until the next data required is not present in the Outbound Data Buffers.

The difference in operation while operating in PCI-X mode is that:

- Original read request is split first time while latching request and requestor information.
- Tsi308 connects itself with original requester on PCI-X as Split Completer as opposed to waiting for the request to be reissued.
- Data is transferred in ADBs as PCI-X master is not allowed to disconnect arbitrarily.

ï Tsi308 generates one or more split completion transactions until the byte count of the original request is satisfied.

2.4.9 Continuous Prefetching (PCI mode only)

If continuous prefetching is enabled in the Read Control CSR, the Tsi308 issues further ascending read requests to fill buffers as they drain (up to a 4-KB page boundary) in an effort to make sure required data is always available. Otherwise, the transaction is disconnected as soon as all data from the initial reads is returned to the PCI bus.

2.4.10 Transaction Disconnects

Read transactions may be disconnected by the bridge when required data is not available in time. The Delayed Request buffer remains in use until all outstanding HyperTransport prefetch requests receive their responses; then it is retired and any leftover data is discarded (PCI mode only). Each delayed request buffer also has an associated discard timer loaded with one of the two values determined by the Secondary Discard Timer bit (9) of the Bridge Control CSR (3Eh) when the data is received from HyperTransport. If this timer expires before the data is called for by the PCI master, the data is discarded and the buffer is retired.

2.4.11 Outbound Data Buffer

The Tsi308 contains a central data buffer (four 512-byte entries) for the accumulation of read response data to return to the PCI-X bus. These four buffers correspond to four read requests in Delayed Request Buffers. Data returning from HyperTransport is loaded into the buffer based on the comparison between request information stored in the corresponding delayed request number and information contained in response header, and drained out to the PCI-X bus when the delayed request reconnects (PCI) or when the Tsi308's split completer connects.

2.4.12 Interrupt Generation

The Tsi308 interrupt controller supports 10 interrupts per PCI-X port. Each of these interrupts are enabled and configured independently through its Interrupt Definition Register CSR. However Tsi308 while operating in Tsi301 compatible mode, these interrupts are partitioned into different groups, each group consists of four interrupts or less and interrupts are configured in groups. The interrupt configuration options include: edge versus level sensitivity, polarity, and vector ID.

An incoming interrupt is first synchronized to the core clock domain and masked with its CSR enable bit to set the interrupt's bid in the Interrupt Request Register (IRR). A round-robin arbiter will then examine each IRR bit and forward requests to the primary bus logic, setting its In Service Register (ISR) bit. New interrupts from that pin will not be accepted as long as the ISR register is set.

For edge sensitive interrupts, the ISR bit is cleared as soon as the primary interface logic accepts the inbound interrupt request

- For level interrupts, the ISR bit is not cleared until an End of Interrupt (EOI) packet is received with a vector ID matching the ISR. Software may also write a 1 to $63rd$ bit of Interrupt Definition Register to clear the interrupt without an EOI.
- All interrupts and interrupt blocks are disabled at reset and must be enabled by software
- The ISR can also be monitored via the CSRs

2.4.12.1 Interrupt Diagnostic Mode

To simplify debugging interrupt software and hardware, each interrupt may be stimulated and monitored via CSR reads and writes (see the Interrupt Diagnostic Register for details). Writing a 1 to the Initiate field of the CSR will generate an interrupt on the interrupt line in the Pin Number field. The ISR bit may be observed by monitoring the Active field of the same CSR.

The Active field is always for the CSR indicated by the Pin Number field

2.5 PCI-X Arbiter

The Tsi308 includes a PCI-X arbiter for each PCI-X port. The arbiter is an independent unit. The Tsi308's internal PCI-X request and PCI-X grant signals are connected to pins as Px_REQ_OUT_N and Px_GNT_IN_N (x = 0 for PCI A and x = 1 for PCI B). It is possible to either use this arbiter or to bypass it and use an external arbiter.

The Tsi308's internal arbiter contains two round-robin arbitration groups: Px_REQ0_N and Px_REQ1_N through Px_REQ5_N. Within each group, arbitration is shared equally between the requests. Generally Px_REQ_OUT_N would be attached to Px_REQ0_N and Px_GNT_IN_N attached to Px_GNT0_N.

When there are no requests present, the arbiter either parks the bus at the last grant or (based on a CSR bit) at Px GNT0 N, which is presumed to be the Tsi308's internal requester.

The PCI Control CSR ParkMaster bit allows the configuration of the internal PCI-X Arbiter parking.

2.6 Online Insertion and Removal (OIR)

The Tsi308 device provides a hook to help users implement their own proprietary hot plugging outside of the device. The device implements an input pin $(PX$ OIR DISCON EVENT) in each of the two PCI-X interface logic which is asserted by the user when an attached PCI-X card is about to be removed. When asserted, Tsi308's PCI-X master finishes its current transfer if any progress at the time on PCI-X and stops generating any further traffic. Tsi308 starts sending dummy responses to HyperTransport for all the pending read requests (that are within the chip queue yet to be executed on PCI-X as well as the ones that have been already presented on PCI-X and are pending on PCI-X), and silently drops all the posted requests from HyperTransport meant for the affected PCI-X port.

This condition persists until software resets the PCI-X by writing a 1 to SecBusReset (bit 6) of Bridge Control Register.

Tsi308 neither logs this event in CSR nor generates an interrupt of this event. It is up to the user to notify the host of this event and initiating appropriate action if any.

For more information, see [Section A on page 233.](#page-232-0)

2.7 LDTSTOP# Support

Tsi308 implements LDTSTOP# pin to support HT Link Disconnect/Re-connect sequence. When LDTSTOP# is asserted, Tsi308's transmitter finishes sending the current packet and then continues to send disconnect NOP packets through the end of current CRC window and continuing through the transmission of the CRC bits for the current window, the transmitter continues to drive disconnect NOP packets on the link for 64-bit times, after which point transmitter waits for the corresponding receiver on the same link to complete its disconnect sequence, and then disables its drives if LDTSTOP# tri-state enable bit is set. During disconnection of HT link, Tsi308's receiver continues to operate normally through the end of current CRC window until it receives the CRC bits for the current window and then disables its receivers if LDTSTOP# tri-state enable bit is set.

When LDTSTOP# is de-asserted, Tsi308's transmitter goes through link initialization sequence just like it does coming out of warm reset. Tsi308's receiver waits for 1 us before enabling its inputs after LDTSTOP# is de-asserted.

If system software re-programs Tsi308's link width and link frequency, it will take effect after HT Link Disconnect and Re-connect sequence.

Tsi308 registers buffer status prior to the disconnection. No new buffer release messages need to be sent after re-connection of HT links. Tsi308 does not flush out its internal buffers when LDTSTOP $#$ is asserted. It just stores the packet and resumes the packet transmission after its link is re-connected to HT chain.

2.8 Power Management

Tsi308 does power saving when operating in Single PCI-X mode by gating the clock to PCI-B logic internal to it. Tsi308 powers down P1_AD pads during Single PCI-X 32-bit mode operation. Tsi308 does not do any other power savings even though it has implemented PCI power management capability registers and SMAF field in CSR space. These registers are implemented to be compatible with x86 systems.

2.9 Reset

2.9.1 Cold Reset

Cold Reset of Tsi308 is caused by the deassertion of L_POWER_OK pin. At power-on, L_POWER_OK must remain deasserted until power and clocks are stable for at least 1 ms. L_RST_N must be asserted before L_POWER_OK asserts and remains asserted for at least 1 ms following.

Cold reset results in the initialization of all internal state, including the loading of internal registers from strapped PCI bus pins and the SRI (Serial ROM Interface). The PCI-X bus is held in reset until the deassertion of L_RST_N.

2.9.2 Warm Reset

Warm reset of the Tsi308 is caused by the assertion of L_RST_N while leaving L_POWER_OK asserted. Once asserted, L_RST_N must remain asserted for at least 1 ms.

Warm reset results in the initialization of most internal state, with the exception of state loaded from PCI-X bus sampling or the external ROM interface, and persistent error state. The PCI-X bus is held in reset until the deassertion of L_RST_N.

Tsi308 implements L_RST_N as in-out. It extends the asserted state of L_RST_N beyond 1 ms until internal PLLs are locked and configuration information from SROM is loaded

2.9.3 Reset Configuration

Configuration information is loaded into the Tsi308 at reset from PCI-X AD bus, P0_AD[31:0].

2.9.4 HyperTransport Link Initialization

On the deassertion of L_RST_N as part of a warm or cold reset sequence, the Tsi308 attempts to initialize both of its HyperTransport links as described in the HyperTransport specification.

When link initialization is done, the InitDone CSR bit is set. Software polls this CSR bit to determine that the link is live and may be included in fabric initialization. InitDone remains clear if the Tsi308 is unable to initialize the link because of any of the following:

- There is no device at the other end.
- Communication with the device at the other end is not possible.
- The link is disabled because of a previous failure.

2.9.5 HyperTransport Fabric Initialization

Once hardware initialization of the individual links in the HyperTransport chain completes, host software is responsible for initializing the HyperTransport fabric. A sample initialization sequence proceeds according to the following example. If starting from the host, initialization proceeds recursively for each link in the chain.

Example initialization sequence:

- 1. Read the InitDone bit for the link to determine whether the link is live. Also, read the various link error bits to determine if the link has taken errors since reset that prevent it from functioning correctly. If the link is not live, or is taking fatal errors, fabric sizing is complete and you can proceed to Step 6.
- 2. All devices assume a HyperTransport UnitID of 0 at reset. Therefore, a configuration access to PCI device number 0 is accepted by the first uninitialized device on the chain. This is the device at the far end of the link currently being sized.
- 3. Performing a write to the HyperTransport Command register, without changing any fields, causes initialization of the master host bit. This indicates the HyperTransport link that connects toward the host bridge. Polling the error bits for that link determines whether the node on the far end is detecting any fatal errors on the link. If so, this link is not to be used, fabric sizing is complete, and you can proceed to Step 5.
- 4. Software reads the Class Code, Vendor ID, and Device ID from the device at the end of the current link to determine what type of device it is talking to.
- 5. Software writes the BaseUnitID register in the device with the next free HyperTransport unitID value, starting with 1 for the first device. It reads the unitCount register to determine how many unitID values the current device requires and increments the next free unitID value appropriately.
- 6. Return to Step 1 to size the next link in the chain. Because step 3 wrote the base unitID of the last sized device to a nonzero value, it no longer accepts accesses to device 0. Instead, it forwards them to the next device in the chain.
- 7. When sizing completes to the last live link in the chain, set the EndOfChain and TransmitOff bits for the outgoing link of the last device. This prevents the unused link from being driven and enables proper handling of HyperTransport packets that traverse the HyperTransport link without finding their target node.

2.9.6 Secondary Bus Reset

The PCI-X bus may be placed and held in reset while the HyperTransport interface remains live. When the reset pin of the PCI-X bus (Px_RST_N) is asserted, all internal PCI-X buffers are flushed. Inbound writes that are in progress during the reset may be completed, depending on how far into the pipeline they are. Inbound reads are retired as their responses return from HyperTransport and the data dropped. Outbound operations to PCI-X are dropped and error status maintained and returned as if the operations had master aborted on the PCI-X bus. Software is responsible for coping with any transfers that were interrupted or dropped as a result of the reset. The interrupt controller is not affected by secondary bus reset.

Only the Tsi308 can initiate PCI-X bus reset. The PCI-X bus is placed and held in reset under CSR control

2.10 Error Handling

The Tsi308 provides a variety of error checking, logging, and containment functions to ensure correct operation and diagnose failures.

2.10.1 Reporting

The Tsi308 logs all errors it detects in CSRs that are persistent through a warm reset. CSR enables are used to mask and control routing of error notification. Not all signaling methods are available for all error types.

When the Tsi308 takes an error, it signals the system in one of three ways. The error-signaling methods are listed below in order of increasing severity:

For errors detected by the Tsi308 as a transaction target, errors may be signaled in the bridge chip response. The method of signaling in the response depends on the protocol of the bus on which the error is detected.

- 1. Transaction errors on non-posted HyperTransport requests may be indicated by an Error response.
- 2. Transaction errors on PCI may be indicated by a target abort, PERR# assertion, or premature disconnection before the data in error is transferred.

In each of these cases, the protocol on the given bus continues to run, and it is the requester's responsibility to take appropriate action on receipt of the error. Transmission of HyperTransport error responses (without NXA) sets the SigdTgtAbort bit in the Status CSR. Transmission of a target abort on PCI sets the SigdTgtAbort bit in the Secondary Bus Status CSR.

Errors may be signaled to the system by the error interrupt pins. Two pins, FATAL_ERR_N and NONFATAL_ERR_N, allow division of errors into two different priority classes.

These pins can be directly connected to input pins on the Tsi308 interrupt controller or to an external interrupt controller. They are active-low, open drain outputs, which allows the pins to be wire-ORedwith other interrupt sources. They also provide edge-triggered interrupts, pulsing when an error is detected. Only in Tsi301 mode the SerrEn bit in the Command CSR serves as a master enable for the error interrupts, in addition to the enables for individual error conditions. Assertion of either error interrupt sets the SigdSerr bit in the Status CSR.

The HyperTransport transmitters can flood the link with synchronization packets. These propagate along the length of the chain and are detected by the host bridge. The host bridge is then responsible for taking appropriate action. The link has to pass through a warm reset sequence before it can be re-enabled, and all transactions in progress are lost. This option is only available for catastrophic

HyperTransport errors that render the chain untrustworthy. Any error that causes sync flooding also sets the LinkFail bit in the HyperTransport Link 0/1 Control CSR for the link on which the error was detected. The setting of LinkFail will cause that link to not be re-initialized on the next warm reset event. The SigdSerr bit in the Status CSR is also set.

2.10.2 HyperTransport Errors

2.10.2.1 Link Errors

HyperTransport link errors are detectable at the lowest levels of the HyperTransport protocol and indicate a basic failing of the HyperTransport link. These errors are not localizable to individual transactions, and all can bring down the link through sync flooding.

CRC is checked by all of the HyperTransport link receivers in accordance with the HyperTransport specification.

Basic protocol checks are performed for proper switching of the CTL signal and legal command encodings. An overflow error is detected if the HyperTransport flow control mechanism breaks down and packets are received with no space for them in the Rx buffers.

[Table 5](#page-46-0) indicates the CSR bits used to log and enable reporting of each HyperTransport link error type.

Error	Log Bit	Sync Flood	Fatal Interrupt	NonFatal Interrupt
Bad CRC	LinkCtrl/	LinkCtrl/	ErrCtrl/	ErrCtrl/
	CrcErr[0]	CrcSyncFloodEn	CrcFatalEn	CrcNonFatalEn
Protocol Error	LinkCtrl/	ErrCtrl/	ErrCtrl/	ErrCtrl/
	ProtErr	ProtSyncFloodEn	ProtFatalEn	ProtNonFatalEn
Rx Overflow	LinkCtrl/	ErrCtrl/	ErrCtrl/	ErrCtrl/
	OvfErr	OvfSyncFloodEn	OvfFatalEn	OvfNonFatalEn

Table 5: HyperTransport Link Error CSR Bits

In addition to these HyperTransport link errors, the Tsi308 also propagates sync packets out of its transmitters if sync flooding is detected on either receiver. This is not logged or reported. This condition represents propagation of an error report from another device, not error detection by the Tsi308.

2.10.2.2 Transmission Errors

End of Chain (EOC) accesses occur when the Tsi308 tries to transmit a packet from PCI, or the other HyperTransport link, on a link that is not live (there is nowhere to send the packet).

- 1. If the outgoing packet was a non-posted request, the Tsi308 generates a matching response with the error and NXA bits set and sends the response back to the requester. This is equivalent to a master abort on PCI and requires no logging or further action by the Tsi308.
- 2. If the outgoing packet was a broadcast, it is silently dropped (it has traversed the whole chain).
- 3. If the outgoing packet was not a non-posted request (either posted request or response) or a broadcast, then there is no in-band way to signal the error. The packet is dropped and may be signaled as an error, as shown in [Table 6](#page-46-1).

Table 6: HyperTransport Forwarding Error CSR Bits

Tsi308 supports Drop on Uninitialized Link, if this bit is set then a transmitter with its Initialization complete bit clear will always act as if the End of Chain bit were set.

Tsi308 also supports 64 Bit Addressing Enable, If this bit is set, requests that access addresses above FF_FFFF_FFFFh can be issued or forwarded by this link interface with the Address Extension command. If this bit is clear, then any access above FF_FFFF_FFFFh will be master aborted as if the end of chain was reached.

2.10.2.3 Master Errors

The Tsi308 accepts received responses that match its unitId and compares the response srcTags to the bridge's outstanding request srcTags. When the response does not match an outstanding request, it is called a Response Match error.

If the response does match an outstanding request, it is routed back to the PCI bus. The response may contain an error assertion, indicated by the Error bit. In this case, the NXA bit indicates whether the error was caused by the access failing to reach a target (value of 1) or signaled by the target (value of 0).

- An NXA bit value of 1 is roughly equivalent to a PCI master abort.
- An NXA bit value of 0 is equivalent to a PCI target abort.

In general, these errors are signaled to the PCI bus in the same way as in a standard PCI-PCI bridge.

Tsi308 HyperTransport master error settings are described in [Table 7](#page-48-0).

Because the Tsi308 prefetches read data, it is possible that the error occurred on a location that the PCI device did not intend to access. If error responses are received for prefetch requests, the associated data is dropped. The prefetching stops at the point of the error, and the PCI transaction is disconnected when it reaches that point. If the PCI device then requests the data again, the request goes back to HyperTransport. If the error recurs, the response error is passed to PCI.

Once a response reaches the delayed request buffers, it must wait there until the requesting PCI master reconnects. As soon as the initial HyperTransport request completes and places its data in the buffer, a timer starts. The timer may be initialized to one of two values, as configured by the SecDiscardTimer bit in the Bridge Control CSR. If the timer expires before the data is called for, that may be treated as an error.

Table 7: HyperTransport Master Errors CSR Bits

2.10.2.4 Slave Errors

The Tsi308 CSR master only supports accesses within a 32-bit aligned block. Accesses that span more than one 32-bit block receive HyperTransport error responses, equivalent to a PCI target abort. No other action is taken. Error responses may also be signaled to HyperTransport because of errors taken when the request was issued to PCI.

2.10.3 PCI Errors

2.10.3.1 PCI System Errors

PCI devices may assert an unrecoverable system error by asserting SERR# on the secondary PCI bus.

Settings for this error are described in [Table 8](#page-49-0).

Table 8: PCI System Error CSR Bits

2.10.3.2 PCI Master Errors

PCI master errors refers to errors detected by the Tsi308 when acting as a master on the PCI bus. Master and Target Abort are defined in the *PCI Local Bus Specification, Revision 2.*2.

 TRDY# timeout refers to a violation of the target latency requirements, as given by the Error/TrdyTimer CSR. Retry timeout refers to an excessive number of retries and/or disconnects, as given by the Error/Retry-Timer CSR.

All PCI requests issued are forwarded through the Tsi308 from HyperTransport. If the HyperTransport request was non-posted, error status may be returned to the HyperTransport requester in the response.

If it was posted, the error may only be signaled by the error interrupts. A single set of error-reporting controls is used for all posted requests, regardless of the specific error taken.

Table 9: PCI Master Errors CSR Bits

Table 9: PCI Master Errors CSR Bits

All of the above errors return all 1s data for read requests, whether or not the response Error bit is set.

2.10.3.3 PCI Parity Errors

All PCI devices are required to drive even parity on P_PAR when they are driving the bottom half of the P_AD bus, and on P_PAR64 when they are driving the top half.

The Tsi301 checks parity on the P_AD bus during command/address phases and data phases when it receives data. The Tsi308 then logs bad parity in the DetParErr bit of the Secondary Bus Status CSR. Other action is taken only if enabled by the ParErrRespEn bit in the Bridge Control CSR. The action taken depends on the type of information being transferred at the time of the error and in which direction the transfer was occurring.

[Table 10](#page-51-0) indicates the CSR bits used to log and enable reporting of each PCI parity error.

Table 10: PCI Parity Errors CSR Bits

Error in	Fatal Interrupt	NonFatal Interrupt	PCI
Command/Address	Error/ CmdPerrFatalEn	Error/ CmdPerrNonFatalEn	If decode has caused the $Tsi308$ to drive P DEVSEL_N, Target Abort
Write Data to Tsi308	Not Supported	Not Supported	Assert P PERR N
Read Data to Tsi308	Not Supported	Not Supported	Set SecStatus/MstrDParErr. return HyperTransport error response, and assert P PERR N

The Tsi308 may also sample P_PERR_N, asserted when it is driving write data out, indicating that a parity error was detected by the target of the write. If the ParErrRespEn bit is set and the request was a non-posted write, it receives an error response. If the request was a posted write and the PostFatalEn or NonPostFatalEn bits in the Error Control CSR are enabled, the error is signaled by one of the error interrupts.

2.11 Test Features

The following test features are included in the Tsi308 to facilitate testing of the chip.

2.11.1 JTAG

The Tsi308sís JTAG interface is compliant to IEEE 1149.1 standard. The basic test operation is controlled through five pins namely TCK, TMS, TDI, TDO and TRST_N. In Tsi308, JTAG interface is used for Memory BIST, JTAG CSR Read/Write, Force Tri-state, Boundary Scan and Core clock frequency selection during BIST.

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2.11.1.1 Memory BIST Controller

JTAG Memory BIST controller is the interface between JTAG and Memory BIST controllers in each of the four ports. It also decodes JTAG-MBIST instruction and generates BIST enable for the four ports. After BIST operation is over is collects the status information from all the four ports and sends out through JTAG controller.

2.11.1.2 CSR Read/Write Controller

JTAG CSR controller decodes the JTAG-CSR instruction and generates required signals for JTAG CSR read/write operation. It also returns JTAG CSR read data to the JTAG controller.

2.11.1.3 LDT BIST

LDT BIST controller generates Tri-State, Reset, PowerOK signals if JTAG-RUNBIST instruction is executed.

2.11.1.4 Boundary Scan

Boundary Scan is a DFT technique for testing chips and inter-connectivity among chips on printed circuit board. With JTAG insertion to the IC design circuitry, only few signals are needed to control the test activity of the chip instead of the ad-hoc Bed-of-Nails technique.

Boundary Scan JTAG instruction SAMPLE/PRELOAD access the boundary scan register via a data scan operation to take a sample of the functional data entering and leaving the device. This instruction is also used to preload test data into the boundary scan register prior to loading and EXTEST instruction.

The EXTEST instruction places the device into an external boundary test mode and selects the boundary scan register to be connected between TDI and TDO. During this instruction, the boundary scan cells associated with outputs are preloaded with test patterns to test downstream devices. The input boundary cells are set up to capture the input data for later analysis.

2.11.1.5 Signal setting for BSD

Following are the signal values to be set during BSD test.

TMODE

This signal need to be 1'b0 so that MODE1 and for tristate will be generated based on JTAG instructions. In the design, these two signals are forced to 1 'b0 when TMODE is high (i.e., during ATPG).

Power Down

For all pads to be active during Boundary Scan, the power down signals generated inside the design have been forced to Low.

2.11.1.6 Boundary Scan Chain Order

The following is the Boundary Scan Chain Order with the relevant information.

2.11.2 SCAN and ATPG

The Tsi308 implements a total of 21 scan chains. All of these 21 scan inputs and scan outputs are shared with various functional pins. The scan operation is controlled by the input signals, SCAN EN and TMODE.

[Scan Input and Output Pins](#page-63-0) lists the mapping of scan input and output pins onto functional pins.

Table 12: Scan Input and Output Pins

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3. Clock Frequency and Mode Selection Hardware Straps

This chapter discusses the following topics about the Tsi308:

- "Overview" on page 67
- "Core Clock Frequency Selection in RevC mode" on page 69
- "PCI Bus A Frequency Selection in RevC mode" on page 70
- "PCI Bus B Frequency Selection in RevC mode" on page 70
- "PCI Bus A and Core Clock Frequency Selection in non-RevC mode" on page 71
- "PCI Bus B Frequency Selection in non-RevC mode" on page 72
- "Link Frequency Selection (Tsi301 mode only)" on page 72
- "Miscellaneous Straps" on page 74

3.1 Overview

Tsi308 has five PLLs: HT0_PLL_0, HT1_PLL_1, CORE_PLL, PCI_A_PLL and PCI_B_PLL.

The reference clock for the first three PLLs is REFCLK_C in RevC mode and P0_CLK in non-RevC mode. The reference clock for PCI_A_PLL is P0_CLK and the reference clock for PCI B PLL is P1_CLK.

- HT0 PLL generates the HyperTransport transmit clock for Link 0.
- HT1 PLL generates the HyperTransport transmit clock for Link 1.
- CORE PLL generates the clock for core logic.
- PCI A PLL generates the clock (PCI CLK A) for PCI bus A interface logic.
- PCI_B_PLL generates the clock(PCI_CLK_B) for PCI bus B interface logic.

Tsi308 Revision C is backward pin compatible with previous revisions $(A \& B)$. Tsi308 Revision C device can be dropped into the boards made for previous revisions. AD[15] is used to select between RevC or non-RevC modes. $AD[15] = 1$ for RevC mode and $AD[15] = 0$ for non-RevC mode.

Figure 4: Primary clock inputs to Tsi308 **PLLs**

When operating in 'non-RevC' mode, P0 CLK is fed to an internal PLL that generates clock for the PCI_A interface logic as well as core clock. The P0_CLK is also fed to two other internal PLLs that generate HyperTransport transmit clocks for both the linksThe P1_CLK is fed to yet another PLL that generates clock for the PCI_B interface logic.

If the Tsi308 is to perform synchronous link initialization with HyperTransport devices on either side of it, the reference clock (P0_CLK) must be derived from the same base frequency source. If not, asynchronous link initialization must be used. No phase relationship is required in either mode.

When operating in 'RevC mode', Tsi308 derives core clock from REFCLK C. RevC mode is set through hardware strap option. A CORE PLL is used to generate core clock. REFCLK C is also fed to internal PLLs that generate HyperTransport transmit clocks for both the links. P0_CLK and P1_CLK are fed to corresponding internal PLLs that generate clocks for the PCI_A and PCI_B iterface logic respectively. If the Tsi308 is to perform synchronous link initialization with HyperTransport devices on either side of it, the reference clock (REFCLK_C) must be derived from the same base frequency source.

For debug and test purposes, the Tsi308 allows bypassing of HT PLLs. It provides separate bypass clock inputs to both the HT links. All bypass clocks must be derived from the same base frequency source. The PCI_A ,PCI_B and core PLLs also can be bypassed. When PCI PLLs are bypassed PCI A interface is run directly at P0 CLK and PCI B interface is run directly at P1 CLK. When CORE PLL is bypassed, core runs directly at REFCLK C in 'RevC mode' and runs directly at P0_CLK at other modes.

For all the straps that use P0_AD bus, logic 1 assumes that the signal is pulled high to 3.3v supply through a 4.7k ohm resister and logic 0 assumes that the signal is pulled low to ground through a 4.7k ohm resister.

Following sections describe the strap options.

3.2 Core Clock Frequency Selection in RevC mode

All the inputs to Core PLL are taken from the straps. These strap values are taken during cold reset period. [Table 13](#page-68-1) shows the valid strap combinations for Core PLL.

P0_AD[23,22,21,20]	REFCLK_C (MHz)	Core Clock (MHz)
0001	25	100
0010	25	200
0011	33	133
0100	33	200
0101	50	100
0110	50	200
0111	66	133
1000	66	200
1001	100	200
1010	133	200
All other values are reserved		

Table 13: Core Clock Frequency Selection Straps in RevC mode

3.3 PCI Bus A Frequency Selection in RevC mode

[Table 14](#page-69-2) shows the valid strap combinations for PCI-A PLL.

Table 14: PCI-A Clock Frequency Selection Straps in RevC mode

P0_133_N	P0_PCIX_N	P0_M66EN	MODE	P0_AD[6]	PCI_CLK_A (MHz)
		$\pmb{0}$	PCI		25
		$\pmb{0}$	PCI	0	33
			PCI		50
			PCI	$\mathbf 0$	66
	Ω	N/A	PCI-X		50
	Ω	N/A	PCI-X	$\mathbf 0$	66
Ω	$\mathbf 0$	N/A	PCI-X		100
Ω	$\mathbf 0$	N/A	PCI-X	$\mathbf 0$	133

3.4 PCI Bus B Frequency Selection in RevC mode

[Table 15](#page-69-3) shows the valid strap combinations for PCI-A PLL.

Table 15: PCI-B Clock Frequency Selection Straps

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3.5 PCI Bus A and Core Clock Frequency Selection in non-RevC mode

[Table 16](#page-70-1) shows all the valid strap combinations for setting PCI_CLK_A exact operating frequency that is also used to generate CoreClock.

3.6 PCI Bus B Frequency Selection in non-RevC mode

[Table 17](#page-71-2) lists all the valid strap combinations that indicate operating frequency of P1_CLK that is also used to generate PCI_CLK_B.

3.7 Link Frequency Selection (Tsi301 mode only)

[Table 18](#page-71-3) lists the valid strap combinations to preload HyperTransport Link 0 and Link 1 frequency.

This table is valid only when Tsi308 is operating in Tsi301 compatible mode. The link frequencies are initialized as specified in HyperTransport specification in normal mode.

Table 18: Link Transmit Clock Frequency Selection Straps

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Table 18: Link Transmit Clock Frequency Selection Straps

For proper functioning of Tsi308, Core clock should be greater than or equal to $1/4^{\text{th}}$ HT Link Frequency(DDR).

3.8 Miscellaneous Straps

[Table 18](#page-71-0) lists the other miscellaneous mode selection straps.

Table 19: Miscellaneous Pin Straps

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Table 19: Miscellaneous Pin Straps

Table 19: Miscellaneous Pin Straps

4. Register Descriptions

This chapter discusses the following topics about Tsi308's registers:

- "Configuration Registers" on page 78
- "Summary of Configuration Registers" on page 81
- "64-bit Address Remapping Capability Indices" on page 85

4.1 Configuration Registers

This chapter provides details about the configuration registers or control and status registers (CSRs) that Tsi308 implements for software to access and program. The Tsi308 supports various modes of operation and the following section outlines these operating modes.

4.1.1 Operating Modes

From the software's viewpoint Tsi308 implements two major operating modes; Tsi301 backward compatible mode or Tsi301 mode and standard HT 1.05 compliant Tsi308 mode. Each of these two modes in turn support either a single 64-bit PCI-X (PCI) device or dual 32-bit PCI-X (PCI) devices.

4.1.1.1 Single Tsi301 (SP) Mode

In Single Tsi301 mode, Tsi308 behaves just like previous generation Tsi301 chip from software's viewpoint. It implements Tsi301 software compatible register set as specified in [5], so that existing software driver that is written for Tsi301 will work seamlessly. In this mode device may not support some of the features that are defined in HT 1.05 specification. However this mode provides some features that are not existent in Tsi301 chip in software transparent way, for example this mode supports HT link speeds up to 600 MHz beyond the 400 MHz supported by Tsi301 chip. This mode supports both sharing and non-sharing dual hosted chain implementation and allowed End Of Chain (EOC) bit to be set in any of the two links.

Figure 5: Single Tsi301 Mode

4.1.1.2 Dual Tsi301 (DSP) Mode

The Dual Tsi301 Mode is same as Single Tsi301 Mode from software point of view, except that this mode implements two independently configured 32-bit PCI devices. The software views this mode as two independent 8-bit HyperTransport tunnel devices (nodes) each bridging to a 32-bit PCI. It implements two identical sets of CSR one for each of the two devices, Device A and Device B. The connection between two devices is through the emulation of a software transparent internal virtual link. This mode supports both sharing and non-sharing dual hosted chain implementation and also allows the logical End Of Chain (EOC) to be set in any of the two links of either device.

Figure 6: Dual Tsi301 Mode

4.1.1.3 Tsi308 **Single PCI-X Mode (GSP)**

This mode implements standard HT 1.05 compliant tunnel device bridging to a single 64-bit PCI-X bus that can be operated in traditional PCI mode as well. The PCI-X mode is compliant to [3] and the standard PCI mode is compliant to [2]. The device A implements desired PCI-X functionality and device B is disabled in this mode. This mode supports both sharing and non-sharing dual hosted chain implementation and also allows End Of Chain to be set in either of the two links.

Figure 7: Tsi308 **Single PCI-X Mode**

4.1.1.4 Tsi308 **Dual PCI-X Mode (GDP)**

This mode implements standard HT 1.05 compliant tunnel device bridging to two independently configured 32-bit PCI-X buses. The software views this mode as two independent 8-bit HyperTransport tunnel devices (nodes) each bridging to a 32-bit PCI-X. It implements two independently configured CSR sets one for each of the two devices, Device A and Device B. The connection between two devices is through the emulation of a software transparent internal virtual link. This mode supports both sharing and non-sharing dual hosted chain implementation and also allows the logical End Of Chain (EOC) to be set in any of the two links of either device.

Figure 8: Tsi308 **Dual PCI-X Mode**

4.1.2 Configuration Mechanism

Configuration accesses are accepted from HyperTransport to the Tsi308 internal CSRs if they are Type 0 accesses with a device number equal to the value in the BaseUnitId field of the HyperTransport Command CSR (CSR 0 for Device A, CSR 1 for Device B if enabled).

4.2 Summary of Configuration Registers

[Table 20](#page-81-0) summarizes the Tsi308 configuration register fields and offsets for various modes of operation. Since Tsi308 supports different modes of operation, some of the fields exist in more than one mode and may take different meaning. The following sub-sections describe the usage model.

4.2.1 Register Access Definitions

Access types are indicated as follows:

- $R Read$. A read of this register returns the field.
- W Write. A write of this register loads the value.
- $C Clear$. A write of 1 to this field clears the field.

4.2.2 Register Access Rules

The following rules apply to Tsi308 CSR accesses.

- Reads to undefined fields return undefined data.
- Read Only fields are blocked from being written and return default values when read.

4.2.3 Mode Encodings

As described in previous sections the following abbreviations are used to denote various operating modes of Tsi308.

- **SP Single Tsi301 mode:** Implements a 8-bit tunnel device (Device A) bridging to a single 64-bit PCI. It uses CSR 0 only and Device B and corresponding CSR 1 is not visible to the software. This mode supports both 32 and 64 bit PCI. CSR is software compatible to previous generation Tsi301 chip.
- **DSP Dual Tsi301 mode:** Behaves as if two Tsi301 software compliant devices in single chip. CSR 0 is used by Device A and CSR 1 is used by Device B. This mode supports only 32-bit PCI devices.
- GSP Tsi308 Single PCI-X Mode: Implements standard HT 1.05 compliant tunnel device with bridge to single 64-bit PCI-X. Only Device A is active and uses CSR 0. Device B and corresponding CSR 1 is not visible to software in this mode.
- **GDP Tsi308 Dual PCI-X Mode:** Implements two standard HT1.05 compliant tunnel devices each bridging to a 32-bit PCI-X. Device A uses CSR 0 and Device B uses CSR 1.

4.2.4 CSR Layout

[Table 20](#page-81-0) lists the layout of all the fields including offsets. The fields are color-coded to denote:

- ï Fields are valid in GSP, GDP and reserved in SP and DSP.
- Fields are multifunctional. They have different meaning in each mode.
- Fields are valid in SP and DSP modes. Reserved in GSP and GDP modes.
- Uncolored fields are valid in all the modes

Since Tsi308 implements two identical sets of CSR in dual device mode, only one set is described below and it has:

- Standard 256 byte space (00h-FFh) that can be accessed using the corresponding Device ID.
- Indirectly accessed 64-bit Address Remapping Registers
- Indirectly accessed Interrupt Definition Registers
- Indirectly accessed SROM Registers

Table 20: Tsi308 **CSR Header**

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Table 20: Tsi308 **CSR Header**

Table 20: Tsi308 **CSR Header**

4.3 64-bit Address Remapping Capability Indices

[Table 21](#page-84-1) lists the fields of 64-bit Address Remap Registers as specified in [1]. These fall under 64-bit Address Remapping Capability Block (Capability 4). The block is located at offset E0h-ECh, the registers under this block are accessed through the Index register at offset E4h and Data register at offset E8h-E9h. As shown in [Table 21](#page-84-1), the Tsi308 implements 64-bit Address Remapping Capability and a single upstream DMA window.

4.3.1 ISOC Bit Setting

Upstream requests that pass through the DMA window defined in [Table 21,](#page-84-1) will have Isoc bit set if bit 2 (Isochronous) of DMA Window Control Register (DMACtrl0) is set. Note that it will only affect the bit setting and Tsi308 still sends traffic through same three base virtual channels. It doesn't implement dedicated Isochronous Virtual Channel buffers.

4.3.2 Read Control 2 Register

The controls (prefetch length, prefetch count etc.) set in Read Control 2 register (offset 5Eh-5Ch) are used for upstream traffic that is passing through the address ranges of DMA window defined in [Table 21](#page-84-1).

4.3.3 Interrupt Definition Registers

The Interrupt Definition Registers fall under Interrupt Discovery and Configuration Capability Block (Capability 2) located at offsets 78h-7Ch. The fields of Interrupt Definition Registers are as shown in [Table 22,](#page-86-0) these are accessed indirectly through Index register located at offset 7Ah and Data register located at offset 7Ch-7Fh. The Tsi308 implements 10 interrupt sources per device, and each one of these interrupt sources implement a 64-bit definition register shown below. The register for Interrupt 0 would occupy indexes 10h and 11h, Interrupt 1 uses 12h and 13h, etc. Bits 31:0 are accessed through the lower (even) index and bits 63:32 are accessed through the high (odd) index.

Interrupts are programmed through Interrupt Definition Registers in RevB mode whereas they are programmed through IOAPIC Registers in RevC mode. So, Interrupt Definition Registers are valid only in RevB mode and they are reserved but may return non-zero value when read in Rev.C mode. This is because these registers are used in IOAPIC mode also.

Table 22: Interrupt Definition Registers

Bit	R/W Access	Initial Value	Field Name and Description
63	R/C	Ω	Waiting for EOI: If RQEOI is 1, then this bit is set by hardware when an interrupt request is sent and cleared by hardware when the EOI is returned. Software may write a 1 to this bit to clear it without an EOI.
62	R	1	PassPW: When 1, interrupt messages will be sent with the PassPW bit set and no ordering of the message with other upstream cycles is guaranteed. When 0, interrupt messages will be sent with PassPW clear, and the device must guarantee that the interrupt message will not pass upstream posted cycles within its queues. If a device supports only one of these behaviors, this bit is read-only and indicates which behavior is supported. For Tsi308, this bit is hardcoded to 1.
61:56	R	Ω	Reserved
55:32	R/W	Ω	IntrInfo[55:32]
31:24	R/W	F8h	Intrinfo[31:24]: Must default to F8h for compatibility with HT technology 1.01 and earlier devices. Values of F9 or above must not be used or conflicts with non-interrupt address spaces will result. Some hosts only recognize interrupts with this field set to F8h.
23:8	R/W	Ω	Intrinfo[23:8]
$\overline{7}$	R	Ω	Intrinfo[7] For Tsi308, this bit is hardcoded to 0.

Table 22: Interrupt Definition Registers

4.3.4 SRI Indices

This section defines the registers that are accessed indirectly through Sri Index located at offset D4h and Data register at offset D8h. All the offsets listed are accessed indirectly through Sri Index.

4.3.4.1 GG-Link0 Impedance Control Registers

These registers are in addition to the Link0 Impedance Control (73h-70h) registers. These extra registers are needed because impedance calibrator circuit used in Tsi308 is different from Tsi301.

Register Offset: 14h (indirect through Sri Index located at offset D4h)

4.3.4.2 GG-Link1 Impedance Control Registers

These registers are in addition to the Link1 Impedance Control(77h-74h) registers. These extra registers are needed because impedance calibrator circuit used in Tsi308 is different from Tsi301.

Register Offset: 2Ch (indirect through Sri Index located at offset D4h).

4.3.4.3 GG-Link0 Impedance Calibration Registers

Register Offset: 30h (indirect through Sri Index located at offset D4h)

4.3.4.4 GG-Link0 Impedance CalOut Registers

Register Offset : 34h (indirect through Sri Index located at offset D4h)

4.3.4.5 GG-Link1 Impedance Calibration Registers

Register Offset: 38h (indirect through Sri Index located at offset D4h).

4.3.4.6 GG-Link1 Impedance CalOut Registers

Register Offset : 3Ch (indirect through Sri Index located at offset D4h)

4.3.5 Tsi308 **Registers**

This section describes the bit position, type, default value and description of each register in Tsi308.

Behavior of the register in Tsi301 mode and Tsi308 mode is described explicitly when there is a conflict.

4.3.5.1 VendorID Register

Register Offset : 01h-00h

4.3.5.2 Device ID Register

Register Offset : 03h-02h

4.3.5.3 Command Register

Register Offset : 05h-04h

4.3.5.4 Status Register

Register Offset : 07h-06h

4.3.5.5 Revision ID Register

Register Offset : 08h

4.3.5.6 Class Code Register

Register Offset: 0B-09h

4.3.5.7 CacheLineSize Register

Register Offset : 0Ch

4.3.5.8 Primary Latency Timer Register

Register Offset : 0Dh

4.3.5.9 Header Type Register

Register Offset : 0Eh

4.3.5.10 BIST Register

Register Offset : 0Fh

4.3.5.11 Base Address Register 0

Register Offset : 13h-10h

4.3.5.12 Base Address Register 1

Register Offset : 17h-14h

4.3.5.13 Primary Bus Number Register

Register Offset : 18h

4.3.5.14 Secondary Bus Number Register

Register Offset : 19h

4.3.5.15 Subordinate Bus Number Register

Register Offset : 1Ah

4.3.5.16 Secondary Latency Timer Register

Register Offset : 1Bh

4.3.5.17 I/O Base Address Register

Register Offset : 1Ch

4.3.5.18 I/O Limit Address Register

Register Offset : 1Dh

4.3.5.19 Secondary Bus Status Register

Register Offset : 1Fh-1Eh

4.3.5.20 Memory Range Base Address Register

Register Offset : 21h-20h

4.3.5.21 Memory Range Limit Address Register

Register Offset 23h-22h

4.3.5.22 Prefetchable Memory Range Base Address Register

Register Offset : 25h-24h

4.3.5.23 Prefetchable Memory Range Limit Address

Register Offset : 27h-26h

4.3.5.24 Prefetch Memory Range Base Upper 32-Bit Address

Register Offset : 2Bh-28h

4.3.5.25 Prefetch Memory Range Limit Upper 32-Bit Address

Register Offset : 2Fh-2Ch

4.3.5.26 I/O Range Base Upper 16 Bits Register

Register Offset : 31h-30h

4.3.5.27 I/O Range Limit Upper 16 Bits Register

Register Offset : 33h-32h

4.3.5.28 Capability 1 Register

Register Offset : 34h

4.3.5.29 Reserved Register

Register Offset : 37-35h

4.3.5.30 Expansion ROM Register

Register Offset : 3Bh-38h

4.3.5.31 Interrupt Line Register

Register Offset : 3Ch

4.3.5.32 Interrupt Pin Register

Register Offset : 3Dh

4.3.5.33 Bridge Control Register

Register Offset : 3Fh-3Eh

4.3.5.34 HyperTransport Capability ID Register

Register Offset : 40h

4.3.5.35 Capability 2 Register

Register Offset : 41h

4.3.5.36 HyperTransport Command Register

Register Offset : 43h-42h

4.3.5.37 HyperTransport Link 0 Control Register

Register Offset : 45h-44h

4.3.5.38 Link 0 Width Control(SP)/Link 0 Configuration Register(Tsi308**)**

Register Offset : 47h-46h

4.3.5.39 HyperTransport Link 1 Control Register

Tsi308 uses *HyperTransport Link 0 Control* of CSR0 for Link0 and *HyperTransport Link 0 Control* of CSR1 for Link1in **GDP** and **DSP** modes.

HyperTransport Link 1 Control of CSR0 is used for Link1 in **SP** and **GSP** modes.

CSR1 is not visible in these modes.

Register Offset : 49h-48h

4.3.5.40 Link 1 Width Control(SP)/Link 1 Configuration Register(Tsi308**)**

Tsi308 uses *Link0 Width Control(SP)/Link0 Configuration Register(*Tsi308*)* of CSR0 for Link0 and *Link0 Width Control(SP)/Link0 Configuration Register(*Tsi308*)* of CSR1 for Link1in **GDP** and **DSP** modes.

*Link 1 Width Control(SP)/Link 1 Configuration Register(*Tsi308*)*of CSR0 is used for Link1 in **SP** and **GSP** modes.

CSR1 is not visible in these modes.

Register Offset : 4Bh-4Ah

4.3.5.41 HyperTransport Revision ID Register

Register Offset: 4Ch

4.3.5.42 Link0 Frequency & Link0 Error Registers

Register Offset : 4Dh

4.3.5.43 Link0 Frequency Capability Register

Register Offset : 4E-4Fh

4.3.5.44 Feature Capability Register(Tsi308**)/Reserved(SP/DSP) Register**

Register Offset : 50h

4.3.5.45 Link1 Frequency & Link1 Error Registers.

Tsi308 uses *Link0 Frequency & Link0 Error Registers* of CSR0 for Link0 and *Link0 Frequency & Link0 Error Register* of CSR1 for Link1in **GDP** mode.

Link1 Frequency & Link1 Error Registers of CSR0 is used for Link1 in **GSP** mode.

CSR1 is not visible in these modes.These registers are reserved in SP/DSP modes and reads 0.

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Register Offset : 51h

4.3.5.46 Link1 Frequency Capability Register

Tsi308 uses *Link0 Frequency Capability Registers* of CSR0 for Link0 and *Link0 Frequency Capability Register*s of CSR1 for Link1in **GDP** mode.

Link1 Frequency Capability Registers of CSR0 is used for Link1 in **GSP** mode.

CSR1 is not visible in these modes.

These registers are reserved in SP/DSP modes and reads 0.

Register Offset : 52-53h

4.3.5.47 Enumeration Scratchpad Register(GSP/GDP)/Reserved(SP/DSP) Register

Register Offset : 54-55h

4.3.5.48 Error Handling Register **(GSP/GDP)/Reserved(SP/DSP) Register**

Register Offset : 56-57h

4.3.5.49 Memory Base Upper & Memory Limit Upper(GSP/GDP)/Reserved(SP/DSP) Register

Register Offset : 58-59h

4.3.5.50 Reserved Register

Register Offset : 5A-5Bh

4.3.5.51 Read Control 2 Register

These registers are reserved in SP/DSP modes. But they are valid in GSP and GDP modes. The attributes of Read Control 2 registers are applied for upstream packets instead of Read Control 1(62h-60h) if the an address of upstream request falls in DMA window of Address Remapping Indices.

Register Offset : 5Eh-5Ch

4.3.5.52 Reserved Register

Register Offset 5Fh
4.3.5.53 Read Control 1 Register

Register Offset : 62h-60h

4.3.5.54 PCI Control Register

Register Offset : 63h

4.3.5.55 Error Control Register

Register Offset : 67h-64h

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4.3.5.56 HyperTransport Error Control(SP/DSP)/Reserved(GSP/GDP) Register

Register Offset : 69h-68h

4.3.5.57 Parity Error Reporting Enable

This register is reserved in SP/DSP modes. But, they are valid in GSP and GDP modes.

Register Offset: 6Ah

4.3.5.58 Reserved Register

Register Offset: 6Bh

4.3.5.59 HyperTransport Rx Data Buffer Allocation Register

Register Offset : 6Dh-6Ch

4.3.5.60 HyperTransport Transmit Control Register

Register Offset : 6Eh

4.3.5.61 Link Impedance Control 0 Register

Link Impedance Control 0 Register of CSR0 is used for Port0 and Link Impedance Control 0 Register of

CSR1 is used for Port1 in DSP/GDP modes.

Register Offset : 73h-70h

4.3.5.62 Link Impedance Control 1 Register

These registers are not used in DSP and GDP modes. But, they are valid in SP and GSP modes.

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Register Offset : 77h-74h

4.3.5.63 Interrupt Capability Registers

These registers are reserved in SP/DSP modes. But, they are valid in GSP/GDP modes.

Register Offset : 7Bh-78h

4.3.5.64 Interrupt Capability Registers

These registers are reserved in SP/DSP modes.But, they are valid in GSP/GDP modes.

Register Offset : 7Fh-7Ch

4.3.5.65 Reserved Registers

Register Offset : 9Fh-80h

4.3.5.66 Blockx Interrupty Register

These registers are valid only in SP/DSP modes. They are reserved in other modes.

Register Offset : AFh-A0h (x=0 & 1; y=0,1,2 & 3)

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4.3.5.67 Blockx Interrupt 4 Register

Register Offset : B3h-B0h $(x=0 & 1)$

4.3.5.68 Unit ID Clumping Capability Register

These registers are valid only in GSP/GDP modes.

Register Offset : B7h-B4h

4.3.5.69 Unit ID Clumping Support Register

These registers are valid only in GSP/GDP modes

Register Offset : BCh-B8h

4.3.5.70 Unit ID Clumping Enable Register

These registers are valid only in GSP/GDP modes

Register Offset : BFh-BDh

4.3.5.71 PCI Power Management Capability ID Register

Register Offset: C0h

4.3.5.72 PCI Power Management Next Capability Pointer Register

Register Offset: C1h

4.3.5.73 Interrupt Diagnostics Register

These registers are multifunctional. The following definitions are valid only in SP/DSP modes.

Register Offset : C2h

4.3.5.74 Interrupt Block Level0 Register

These registers are multifunctional. The following definitions are valid only in SP/DSP modes.

Register Offset : C3h

4.3.5.75 Power Management Capabilities Register

These registers are multifunctional. The following definitions are valid only in GSP/GDP modes.

Register Offset: C3h-C2h

4.3.5.76 Interrupt Block Level1 Register

These registers are multifunctional. The following definitions are valid only in SP/DSP modes.

Register Offset : C4h

4.3.5.77 Interrupt Block Level2 Register

These registers are multifunctional. The following definitions are valid only in SP/DSP modes.

Register Offset : C5h

4.3.5.78 Power Management Control & Status Registers

These registers are multifunctional. The following definitions are valid only in GSP/GDPmodes.

Register Offset : C5h-C4h

4.3.5.79 Bridge Support Extensions Register

Register Offset : C6h

4.3.5.80 Reserved

These are reserved in all modes.

Register Offset : C7h

4.3.5.81 Transmit Buffer Counter Maximum Count0 Register

Buffer releases in excess of these thresholds are discarded to allow throttling of traffic.

Register Offset : CAh-C8h

4.3.5.82 Transmit Buffer Counter Maximum Count1 Register

Register Offset : CEh-CCh

4.3.5.83 Test Port Register

These registers are valid only in GSP/GDP modes. They are Reserved in other modes.

Register Offset : D0h

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4.3.5.84 StoreForward Registe

Register Offset : D1h

4.3.5.85 SMAF field

These register is valid only in GSP/GDP modes. They are reserved in all other modes.

Register Offset : D2h

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4.3.5.86 Reserved

Offset D3h is reserved and reads 0.

4.3.5.87 Sri Index Register

These registers are valid only in GSP/GDP modes**.**

Register Offset : D7h-D4h

4.3.5.88 Sri Data Register

These registers are valid only in GSP/GDP modes.

Register Offset : DBh-D8h

4.3.5.89 Diagnostics Link 0 Received CRC Expected Register

These registers are valid only in SP/ DSP modes. They are reserved in other modes.

Register Offset : DFh-DCh

4.3.5.90 HT Capability ID Register

These registers are valid only in GSP/GDP modes.

Register Offset : E0h

4.3.5.91 Capability 5 Register

These registers are valid only in GSP/GDP modes.

Register Offset : E1h

4.3.5.92 DMA Map Register

These registers are valid only in GSP/GDP modes.

Register Offset : E3h-E2h

4.3.5.93 Index Register

These registers are valid only in GSP/GDP modes.

Register Offset : E7h-E4h

4.3.5.94 Data Lower Register

These registers are valid only in GSP/GDP modes.

Register Offset : EBh-E8h

4.3.5.95 Data Upper Register

These registers are valid only in GSP/GDP modes.

Register Offset : EFh-ECh

4.3.5.96 PCI-X Capability ID Register

These registers (FFh-F0h) are multifunctional. The following tables show the value they hold in GSP/GDP modes

Register Offset : F0h

4.3.5.97 Capability 6 Register

Register Offset : F1h

4.3.5.98 PCI-X Secondary Status Register

Register Offset : F3-F2h

4.3.5.99 PCI-X Bridge Status Register

Register Offset : F7-F4h

4.3.5.100 Upstream Split Transaction Control Register

Register Offset : FBh-F8h

4.3.5.101 Downstream Split Transaction Control Register

Register Offset : FFh-FCh

4.3.5.102 Diagnostics Link 0 Receive CRC Received Register

These registers are multifunctional and they hold the following values in SP/DSP modes.

Register Offset : F3h-F0h

4.3.5.103 Diagnostics Link 1 Receive CRC Expected Register

These registers are multifunctional and they hold the following values in SP/DSP modes.

Register Offset : F7h-F4h

4.3.5.104 Diagnostics Link 1 Receive CRC Received Register

These registers are multifunctional and they hold the following values in SP/DSP modes.

Register Offset : FCh-F8h

4.3.5.105 Scratch Register

These registers are multifunctional and they hold the following values in SP/DSP modes.

Register Offset : FFh-FCh

4.3.6 CSR Layout for IOAPIC

To be compatible with x86 systems, Tsi308 provides a memory-mapped version of the interrupt discovery and configuration register set. The memory-mapped register set is comparable to a standard IOAPIC register set and the redirection table entries would have the layout shown in Section 5.4.3.7. These registers are available in the Function1 pci configuration space for PCI-A & PCI-B. These registers are valid only in "GSP/GDP and RevC" mode. They are not visible in other modes as Tsi308 will be a single function device in those modes. In "GSP/GDP and RevC" mode, Interrupts can be programmed only by IOAPIC register set whereas in other modes it can be programmed by Interrupt Discovery registers.

4.3.6.1 Vendor ID and Device ID

Register Offset : 03h-00h

4.3.6.2 Command and Status

Register Offset : 07h-04h

4.3.6.3 IOAPIC Revision ID and Class Code

Register Offset : 0Bh-08h

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4.3.6.4 IOAPIC BAR

Register Offset : 17h-10h

4.3.7 IOAPIC Registers

Each IOAPIC register set support 10 interrupts. The IOAPIC registers are accessed by an indirect addressing scheme using two registers (IOAPIC INDEX and IOAPIC DATA) that are located in memory space specified by BAR0 in function 1 CSR. Memory Mapped Registers for accessing IOAPIC registers are given below:

4.3.7.1 IOAPIC INDEX

Register: 03-00h

4.3.7.2 IOAPIC DATA

Register: 13-10h

4.3.7.3 IOAPIC Registers

4.3.7.4 IOAPIC Version

4.3.7.5 IOAPIC Arbitration ID

4.3.7.6 Redirection Table

4.3.7.7 INTERRUPT Definition Registers

5. Electrical Characteristics

This chapter defines the electrical characteristics of the Golden Gate HyperTransport-PCI/PCI-X bridge.

- "AC Timing Definitions" on page 184
- "Clock Parameters" on page 187
- "HyperTransport Output Timing Characteristics" on page 188
- "HyperTransport Input Timing Characteristics" on page 190
- "HyperTransport Interconnect Timing Characteristics" on page 192
- "HyperTransport Transfer Timing Characteristics" on page 193
- "HyperTransport Impedance Requirements" on page 195
- "HyperTransport DC Electrical Characteristics" on page 197
- "Reset Timing" on page 198
- "Power Consumption" on page 200
- "Thermal Data" on page 200
- "Thermal Recommendations" on page 201
- " Power Sequencing" on page 202
- "Supply Operatiing Ranges" on page 202

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5.1 AC Timing Definitions

Figure 9: Timing Definitions Waveform

Table 23: AC Timing Definitions

Table 23: AC Timing Definitions

5.1.1 AC Timing Values

Table 24: Typical AC Timing Values

Table 24: Typical AC Timing Values

5.2 Clock Parameters

5.2.1 Input Clock

Figure 10: Input Clock Parameters Waveform

 $(Ta = 0^{\circ} C \text{ to } +70^{\circ} C \text{ Commercial}, \text{Vec} = +3.3 \text{ V})$

	25 MHz		33.33 MHz		50 MHz		66.66 MHz		100 MHz		133.33 MHz		
Symbol	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Units
Frequency	24.95	25.05	33.27	33.40	49.9	50.1	66.53	66.8	99.8	100.2	133.07	133.6	MHz
T_{CLK}	39.92	40.08	29.94	30.06	19.96	20.04	14.97	15.03	9.98	10.02	7.49	7.51	ns
T_{CLKH}	12	28	9	21	6	14	4.5	10.5	3	7	2.25	5.25	ns
TCLKL	12	28	9	21	6	14	4.5	10.5	3	7	2.25	5.25	ns
T_{RISE}		3.0		3.0	۰	3.0	٠	3.0	$\overline{}$	3.0	$\overline{}$	3.0	ns
T _{FALL}		3.0	-	3.0	Ξ.	3.0	$\overline{}$	3.0	$\overline{}$	3.0	$\overline{}$	3.0	ns

Table 25: Input Clock Parameters

5.3 HyperTransport Output Timing Characteristics

5.3.1 Differential Output Skew

TODIFF defines the allowable output differential skew as defined by the time difference measured in a single-ended fashion at the midpoint of the transition of the true signal and the midpoint of the transition of the complement signal.

Differential output skew is limited primarily by DV_{OCM} such that at the given minimum output edge rate differential skew would cause a violation of DV_{OCM} before violating the output differential skew specification.

Figure 11: TODIFF

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5.3.2 TCADV (TCADValid)

 T_{CADV} defines the TX_CAD/CTL valid time from TX_CAD/CTL to TX_CLK or from TX CLK to TX CAD/ CTL and is simultaneously an aggregate measurement of the accuracy of the transmitter to place the TX_CAD/CTL edges relative to TX_CLK edge, the minimum TX CLK bit-time and, the TX CAD/CTL group skew.

Nominally, TX_CLK is driven delayed by one-half of a bit-time from the TX_CAD/CTL transitions. This delay provides required setup and hold time to and from the TX_CLK edge at the receiver and therefore allows for simple data recovery. $T_{\text{CADV MIN}}$ is measured at the device pins from the crossing point of either the latest TX_CAD/CTL transition to the crossing point of the TX_CLK transition or the TX_CLK transition to the earliest TX_CAD/CTL transition. $T_{CADVMAX}$ is measured at the device pins from either the crossing point of the earliest TX_CAD/CTL transition to the crossing point of the TX_CLK transition or the TX_CLK transition to the latest TX_CAD/CTL transition.

Because T_{CADV} is an aggregate measure of different uncertainties, it must be measured over a large number of samples and under conditions defined to maximize TX_CAD/CTL group skew, TX CLK edge placement error, and TX CLK phase compression.

Figure 12: T_{CADV}

5.4 HyperTransport Input Timing Characteristics

5.4.1 Input Differential Skew

 T_{IDIFF} defines the allowable input differential skew as defined by the time difference measured in a single-ended fashion at the midpoint of the transition of the true signal and the midpoint of the transition of the complement signal.

Differential input skew is limited primarily by DV_{ICM} such that at the given minimum output edge rate differential skew would cause a violation of DV_{ICM} before violating the output differential skew specification.

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$5.4.2$ T_{SU} and T_{HD}

 $T_{\rm{S}U}$ defines the receiver is required input setup time $T_{\rm{S}U}$ is measured from the crossing point of the last RX_CAD transition to the RX_CLK transition crossing point. $T_{\rm{SII}}$ accounts for receiver package skew, distribution skew, and device input setup time. .

 T_{HD} defines the receiverís required input hold time. $_{HD}$ is measured from the crossing point of the earliest RX_CAD transition to the RX_CLK transition crossing point. T_{HD} accounts for receiver package skew, distribution skew, and device input hold time. T_{SU} and T_{HD} do not necessarily cover the required time to attain $V_{ID-MIN} (AC)$ at the specified minimum input edge rates. $_T$

In the following figure, $T_{\text{SU MAX}}$ represents the maximum setup time that the device can require. This corresponds to the minimum setup time that the system can provide to the device input.

Figure 14: T_{SU} and T_{HD}

5.5 HyperTransport Interconnect Timing Characteristics

5.5.1 T_{CADVRS/RH}

 $T_{\text{CADVRS/RH}}$ defines the remaining RX_CAD valid times to RX_CLK (T_{CADVRS}) and from RX_CLK to RX_CAD (T_{CADVRH}) measured at the receiver inputs. $T_{\text{CADVRS/RH}}$ are used as an aggregate and accumulative measure of the timing uncertainty composed of device output skew, clock edge placement error, and interconnect skew at the device inputs. As such, $T_{\text{CADVRS/RH}}$ must be measured over a large number of samples and conditions which will maximize device output skew, interconnect skew, and clock edge placement error.

 T_{CADVRS} is measured from the crossing point of the last transitioning RX CAD signal to the crossing point of the RX CLK transitioning signal at the receiver. T_{CADVRH} is measured from the RX_CLK transitioning signal to the first RX_CAD signal at the receiver.

Figure 15: TCADVRS / TCADVRH

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5.6 HyperTransport Transfer Timing Characteristics

The following table defines the allowed values for the transfer timing characteristics.

Table 26: HyperTransport Link Transfer Timing Specifications

Parameter	Description	Link Speed	Min	Max	Units
TODIFF	Output Differential Skew	400 MT/s		70	ps
		600 MT/s		70	ps
		800 MT/s		70	ps
		1000 MT/s		60	ps
		1200 MT/s		60	ps
TIDIFF	Input Differential Skew	400 MT/s	695	1805	ps
		600 MT/s	467	1200	ps
		800 MT/s	345	905	ps
		1000 MT/s	280	720	ps
		1200 MT/s	234	600	ps
T_{CADVRS}	Receiver input CADIN valid time to CLKIN	400 MT/s	460		ps
		600 MT/s	312		ps
		800 MT/s	225		ps
		1000 MT/s	194		ps
		1200 MT/s	166		ps
T CADVRS	Receiver input CADIN valid time to CLKIN	400 MT/s	460		ps
		600 MT/s	312		ps
		800 MT/s	225		ps
		1000 MT/s	194		ps
		1200 MT/s	166		ps

Table 26: HyperTransport Link Transfer Timing Specifications

5.7 HyperTransport Impedance Requirements

 is the value of the differential input impedance of the receiver under DC conditions implemented with an on-die differential terminating resistor. This specification must be supported by any compensation technique used within the receiver across all device specific process, voltage, and temperature operating points. The R_{TT} value is defined to match the Z_{OD} of the coupled transmission lines and to provide a slightly overdamped single-ended termination.

R_{ON} is the driver output impedance under DC conditions. This range must be maintained over the valid V_{OD} range. This specification must be supported by any compensation technique used within the output driver across all device specific process, voltage, and temperature operating points. The R_{ON} value is defined to match one-half of the Z_{OD} of the coupled transmission lines.

DR_{ON} (pull-up) is the allowable difference in the driver output impedance between the true and complement when driving a logic $\tilde{e}0i$ and when driving a logic $\tilde{e}1i$ (additionally defined as when true is driven high and when complement is driven high). DR_{ON} (pull-up) is defined to limit differences in both output rising edge slew rate and the resulting differential skew and crossing point shift.

DR_{ON} (pull-down) is the allowable difference in the driver output impedance between the true and complement when driving a logic $\ddot{\text{e}}$ and when driving a logic $\ddot{\text{e}}$ 0í (additionally defined as when true is driven low and when complement is driven low). DR_{ON} (pull-down) is defined to limit the differences in both output falling-edge slew rate and the resulting differential skew and crossing point shift.

The following table gives the DC specifications for these parameters.

Parameter	Description	Min	Typ	Max	Units
R_{TT}	Differential Termination	90	100	110	W
R_{ON}	Driver Output Impedance	45	50	55	W
R_{ON} (pull-up)	High-Drive Impedance Magnitude Change	0		5	%
R_{ON} (pull-down)	Low-Drive Impedance Magnitude Change	0		5	%

Table 27: R_{TT} and R_{ON} DC Specifications

5.8 HyperTransport Signal AC Specifications

Figure 16: Output Loading for AC Timing

Table 28: HyperTransport Link Differential Signal AC Specifications

a. Minimum values assume $V_{LDT} = V_{LDT}$ (min) as a measurement condition.

b. Typical values assume $V_{LDT} = V_{LDT}$ (typ) as a measurement condition.

C. Maximum values assume $V_{LDT} = V_{LDT}$ (max) as a measurement condition.

d. Input edge rates are measured in a differential fashion.

5.9 HyperTransport DC Electrical Characteristics

Figure 17: Output Reference System Load

The following table defines the allowed values for each of the DC characteristics.

Table 29: HT Link Differential Signal DC Specifications

a. Minimum values assume $V_{LDT} = V_{LDT}$ (min) as a measurement condition.

b. Typical values assume $V_{\text{LDT}} = V_{\text{LDT}}$ (typ) as a measurement condition.

C. Maximum values assume $V_{LDT} = V_{LDT}$ (max) as a measurement condition.

5.10 Reset Timing

The following figure shows the reset timing of the Tsi308.

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5.11 Power Consumption

Tsi308 power consumption can be accurately calculated for specific configurations with a spreadsheet available upon request from Alliance Semiconductor.

5.12 Thermal Data

Table 30: Recommended Operating Temperature

Table 31: Thermal Maximum

Table 32: Thermal Characteristics

a. Recommended heat sink is Aavid 374524B00032.

5.13 Thermal Recommendations

The worst-case scenario for the **Tsi308** device is to maintain a junction temperature of 125°C with an ambient temperature of 70° C and a power dissipation of 7 watts. The required $_{14}$ to accomplish this is 7.9°C/W or less.

The actual, measured power dissipation of the device is 3.79 watts. A dissipation of 7 watts was a simulated value.

[Table 32](#page-199-2) shows the thermal performance possible by using the recommended Aavid 374524B00032 heat sink with 1 m/s or greater air flow.

Applications with a lower ambient temperature and/or less power dissipation may require less air flow and/or no heat sink.

Figure 19: Recommended Heat Sink for Tsi308

5.14 Power Sequencing

The preferred power-on sequence for the **Tsi308** is from low voltage to high: +1.2V first, +1.8V second, then +3.3V. Tracking between the power supplies is not required.

The **Tsi308** can be powered up with the higher voltages first, but there will be more current drawn during the power-up period until the lower voltages reach their steady-state levels.

5.15 Supply Operatiing Ranges

Table 33: Supply Operating Ranges

5.16 Absolute Maximum Ratings

Table 34: Absolute Maximum Ratings

Stresses beyond those listed above may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated in the Operating Conditions table is not implied.

6. Packaging

This chapter discusses the following topics about Tsi308's packaging:

- "Package Specification" on page 204
- "Package Diagram" on page 232

6.1 Package Specification

6.1.1 Pins Sorted by Name

Table 35: Tsi308 Sorted by Name

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a. Multiplexed pin - see Table 37 on [page 226](#page-225-0)

b. Multiplexed pin – see Table 37 on [page 226](#page-225-0)

6.1.2 Pins Sorted by Number

Table 36: Tsi308 Sorted by Number

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a. Multiplexed pin - see [Table 37 on page 226](#page-225-0)

b. Multiplexed pin – se[eTable 37 on page 226](#page-225-0)

c. Multiplexed pin – see [Table 37 on page 226](#page-225-0)

d. Multiplexed pin – see [Table 37 on page 226](#page-225-0)

6.1.3 Multiplexed Pins

Table 37: Tsi308 Multiplexed Pins

Pin	Default Pin Name	Multiplexed Pin Name	Multiplex Select Pin
F ₂₃	L1 CCLK TEST	P1 TSTMS	PLL TESTENB
F ₂₄	L1 DCLK TEST	P1 TSTMOD1	PLL TESTENB
AA3	LO CCLK TEST	P1 TSTDIO	PLL TESTENB
AA4	LO DCLK TEST	P1 TSTDI	PLL TESTENB
AC ₁	PLL SELDIV2	P1 TSTCLK	PLL TESTENB
AC ₂	PLL SEL BK	P1 TSTMOD0	PLL_TESTENB
AD1	TX BYPASS CLK E	P1 TSTDO	PLL TESTENB
AF23	PCIB CLK TEST	SCAN CLK	PLL TESTENB

6.1.4 Power Pins

The tables in this section list the power pins for +1.2V HyperTransport power, +1.8V core power, +1.8V analog PLL power and ground, and +3.3V PCI and HT receive power.

6.1.4.1 +1.2V HyperTransport Power Pins

Table 38: Tsi308 +1.2V HyperTransport Power

6.1.4.2 +1.8V Core Power Pins

Table 39: Tsi308 +1.8V Core Power

Table 39: Tsi308 +1.8V Core Power

6.1.4.3 +1.8V Analog PLL Power and Ground

Table 40: Tsi308 1.8V Analog PLL Power and Ground

6.1.4.4 +3.3V PCI Core and I/O Power and HT Receive Power

Table 41: Tsi308 +3.3V PCI Core and I/O Power and HT Receive Power

Table 41: Tsi308 +3.3V PCI Core and I/O Power and HT Receive Power

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6.1.5 Ground Pins

This table lists the common ground pins for +1.2V HyperTransport power, +1.8V core power, and +3.3V PCI and HT receive power. The PLL ground pins are listed in [Section 6.1.4.3 on](#page-227-0) [page 228.](#page-227-0)

Table 42: Tsi308 Ground Pins

6.2 Package Diagram

6.2.1 Package Handling Procedures

- Moisture Classification Level: Level 3
- Storage: Parts may be stored in unopened vacuum-packed antistatic bag up to a minimum of 12 months at < 40°C and 90% humidity. Parts may be stored outside of bag indefinitely at 20% humidity.
- Floor Life: Packages will absorb moisture after opening the bag. Parts must be mounted on PCB within 48 hours after bag is opened; otherwise, baking is required. Floor life conditions are 30°C and 60% humidity.
- Baking Time: 12 hours $@$ 125°C.

A. Online Insertion and Removal

This chapter discusses the following topics about Tsi308's online insertion options:

- "Overview" on page 233
- "Insertion and Removal Sequence" on page 233

A.1 Overview

Tsi308 supports Online Insertion and Removal of PCI/X cards on its secondary side. Switching of modes (PCI to PCIX and vice versa) and change of frequency is possible, upon insertion of the new card in RevC mode. Tsi308 uses P0_OIR_DISCON_EVENT and P1_OIR_DISCON_EVENT to support OIR operation on PCI-A and PCI-B ports respectively.

A.2 Insertion and Removal Sequence

The following sequences explain the actions taken by Tsi308 after the detection of the OIR event.

- 1. Initial state of Tsi308 is assumed to be up and running.
- 2. The OIR logic in Tsi308 monitors the assertion of P0_OIR_DISCON_EVENT for PCI-A and P1_OIR_DISCON_EVENT for PCI-B.
- 3. Upon hot removal of PCI/X card, corresponding OIR_DISCON_EVENT gets asserted. The design requirement for the PCI/X card is such that, OIR_DISCON_EVENT signal is asserted first and then after few milli seconds the PCI/X card can be disconnected.
- 4. Tsi308 completes the ongoing transaction on the PCI port. Tsi308 asserts the REQ# to prevent other PCI masters from taking the bus. No further transactions happen on the PCI bus.
- 5. All the subsequent HT to PCI writes are dropped. HT to PCI read request results in the return of 0x0 data back to HT. Any pending PCI to HT writes are dropped internal to the Tsi308. PCI to HT reads are completed on the HT side but dropped on the PCI side. Flushing all the buffers and sending responses back to CPU takes 40us, after the assertion of OIR_DISCON_EVENT.
- 6. CPU writes to the SecBusReset (bit 6 of the Bridge Control Register) to reset the PCI port. The reference clock for the PCI port is assumed to be stable for few clock cycles (10 clock cycles) after the port is reset. Then the clocks can be shut off if needed. The reset bit is written only after the host CPU receives all the responses and there are no outstanding requests for the OIRed PCI port.
- 7. Switching of modes from PCI to PCIX or vice versa can be done by setting proper strap values when the PCI side is in reset.
- 8. New PCI/X card can be inserted and powered up.
- 9. Clock for the PCI/X card is restarted or adjusted to a new frequency.
- 10. CPU writes the new PCI frequency values in CSR that are routed to PCI PLL. The standard frequencies for PCI mode are 25, 33, 50 and 66 MHz. Standard frequencies for PCI-X modes are 50, 66, 100 and 133 MHz. However the PCI frequencies can be adjusted to any value between the standard frequencies for each mode. Refer definitions for bits[3:2] in section 5.4.1.89 for CSR values, their corresponding PCI frequency ranges and when to program these values.
- 11. PCI PLL is allowed to lock to the new frequency. PCI PLL takes about 5ms to lock.
- 12. After the PCI PLLís are locked, CPU writes to the SecBusReset signal to deassert the reset bit. This pulls the PCI port out of reset.
- 13. Software initializes the PCI/X card through the Tsi308 device.

B. Typical Applications

This chapter discusses the following topics about Tsi308's configuration options:

- "Recommendations for Use" on page 235
- "PCB Layout Guidelines" on page 237
- "Power Distribution" on page 243
- "AS90L10208 Die Pad-to-Ball Trace Length Information" on page 245
- "Example PCB Stackup for HyperTransport" on page 248

B.1 Recommendations for Use

B.1.1 Unused HyperTransport CAD, CLK, and CTL Inputs

If using only one of the two available HyperTransport link interfaces, then the unused Lx RX CAD[7:0], Lx CLK inputs must be pulled to their logic low levels. This means that Lx CAD[x] L must be pulled high and Lx CAD[x] H must be pulled low. A 50 Ω pull-up / pull-down resistor value is recommended as this will hold Lx $CAD[x]$ L at +0.9V and Lx $CAD[x]$ H at +0.3V, both of which are correct HT logic levels. [Figure 21](#page-235-0) shows how the input terminations work.

This scheme would also be used to terminate unused inputs where a 2-bit or 4-bit HT interface was being used.

B.1.2 Analog PLL Power Filtering

For optimal performance of the AS90L10208 on-board PLLs, it is important to provide adequate power filtering for the dedicated PLL power pins, listed in § 8.3.3.

The recommended power filtering for each of the PLL power pins is a series ferrite bead followed by a low-ESR 10 μ F capacitor and a 0.1 μ F ceramic capacitor to GND. A diagram is shown in [Figure 22](#page-235-1). Each PLL power pin should have its own filter.

It is recommended that the selected ferrite bead have a value of 600Ω @ 100 MHz and be capable of sustaining 200 mA current. It is not recommended that a resistor be used in place of the ferrite bead as the current drawn by the PLL will cause an unacceptable drop in voltage.

B.1.3 Decoupling Capacitor Recommendations

IDT recommends that each power pin $(+1.2V, +1.8V,$ and $+3.3V)$ of the Tsi308 have one 0.1 or 0.01μ F ceramic capacitor to GND. The use of surface-mount capacitor packs makes it easier to place these decoupling capacitors near and directly underneath the Tsi308 package.

In addition, several bulk capacitors are recommended. Tsi308 reference platforms have eight 10 μ F X5R dielectric ceramic capacitors for +3.3V, eight of the same capacitors for +1.8V, and one each for +1.2V L0_VLDT and L1_VLDT. These bulk capacitors are placed directly adjacent to the Tsi308 package.

B.2 PCB Layout Guidelines

B.2.1 Tsi308 **HyperTransport Interface Layout Guidelines**

The Tsi308 is a PCI/PCI-X to HyperTransport bridge, that also acts as a HyperTransport tunnel. This layout guide focuses on the HyperTransport interface only. The AS90L10208 conforms to draft 1.05 of the HyperTransport specification.

HyperTransport is a parallel, unidirectional protocol with differential, DDR signaling on both the transmit and receive interfaces. The AS90L10208 supports link widths of 8, 4 and 2 bits and link frequencies of 800, 600, 500, 400, 300 and 200 MHz. The bus width can be independent between transmit and receive interfaces, permitting efficient allocation of system board resources, whereby a bigger bus width can be used for data intensive interfaces and a smaller bus width used for non-intensive applications.

HyperTransport uses a specialized version of LVDS I/Os, with a V_{LDT} of 1.2V. Low-voltage signaling combined with very high data speeds require strict adherence to good layout practices and power circuitry design on any board using the AS90L10208. The AS90L10208 supports a bus width of 2, 4, or 8 bits on both the transmit and receive interfaces on both of the HyperTransport ports.

These are the signals on the AS90L10208 HyperTransport interface:

- $L_TX_CAD_H/L[7:0]$
- L_TX_CLK_H/L
- L_TX_CTL_H/L
- $L_RX_CAD_H/L[7:0]$
- L_RX_CLK_H/L
- L RX CTL_H/L
- L_RST_N
- L_POWER_OK
- LDTSTOP_N

The RST_N, POWER OK and LDTSTOP_N signals are not subject to the same stringent layout requirements of the CAD, CTL, and CLK signals.

The ball assignments for the pins listed above on the AS90L10208 conform to the recommendations outlined in the *Interface Design Guide* published by the HyperTransport Technology Consortium, thereby allowing for easy point-to-point routing between conformant devices on the system board. Conformant devices have their ball out assigned in a manner that permits easier breakout for clean point-to-point routing without additional layer changes.

If the die pad-to-ball trace length matching of the device that interfaces to the AS90L10208 does not fall within the range recommended by the HyperTransport consortium in the *Interface Design Guide* or any layout constraints prevent straight point to point routing on the system board, then trace-by-trace compensation on the system board will be required to meet the constraints laid out in the Interface Design Guide for end to end trace length variations between different signals.

When using trace-by-trace compensated matching, the system board trace length must compensate for the mismatch in the package trace lengths on a trace-by-trace basis. Effectively, the length of the system board trace must be lengthened or shortened to compensate for short or long traces on the transmitter or receiver package. The following steps would produce this type of trace matching:

- 1. Gather trace-by-trace length data for both transmitter and receiver packages.
- 2. Identify the longest system board trace within each clock group. (Use Manhattan distances for approximation.)
- 3. Route the longest signal (differential pair) so the differential skew specification is met, and determine effective pad-to-pad signal length (adding transmitter and receiver package trace electrical lengths).
- 4. Assign length rules to each remaining signal trace such that the system board trace length plus the sum of the package trace electrical lengths equals the effective pad-to-pad signal length of the longest signal within the clock group.
- 5. Route the rest of the clock group.
- 6. Extract resulting system board trace lengths and add the sum of the package electrical lengths to determine if skew control has been maintained.

The die pad-to-ball trace length data for the AS90L10208 is included in the AS90L10208 die pad-to-ball trace length information at the end of this document. Similar information for other devices that interface to the AS90L10208 should be available from the respective vendors.

B.2.2 Layout Guidelines

- All signals are ground referenced differential pairs.
- All HyperTransport signals are referenced to V_{SS} or V_{DD} (never to V_{LDT}).
- $V_{LDT} = +1.2 V$.
- No termination is required on the board. All termination is provided on-die.
- Ideally, all nets in a clock/data group break out on the same layer and route together. If a clock/data group is split between top and bottom routing layers, trace length must be inserted on top layer nets to approximately match the extra delay due to the vias for the bottom layer.
- Decoupling is required in the vicinity of a layer change.
- Routing is to be 20/5/5/5/20 mils for all HyperTransport nets. Route 50/5/5/5/50 mils when serpentining where the 50 mil clearance is any net to itself.
- Group electrical lengths are matched piecewise.

B.2.3 AS90L10208 Board Trace Electrical Specification

- $V_{LDT} = 1.2 V \pm 60 mV$
- Maximum $R_{DC} = 0.29$ (Ω /in)
- Maximum R_{AC} = 1.50 (Ω /in) @ 1 GHz
- Maximum $G_{AC} = 0.32$ (S/in) @ 1 GHz
- Z_{OD} = 100Ω ± 10% (90 110Ω), (* Z_O can be 50 70Ω)
- Bus length vs. frequency:
- 400 Mb/s, 600 Mb/s, 800 Mb/sLength = 1.0 to 24.0 inches
- 1000 Mb/s, 1200 Mb/sLength = 1.0 to 12.0 inches
- 1600 Mb/s Length = 1.0 to 6.0 inches
- $T_{\rm PD}$ = 150 to 170 ps/in
- $T_{\rm PD-ODD}$ = 135 to 155 ps/in

B.2.4 Routing Rules for Individual Signal Groups

B.2.4.1 CAD_H/L, CTL_H/L, CLK_H/L

- All signals are routed as a differential pair.
- All signals are routed 20/5/5/5/20 mils trace width/spacing.

The actual width/space/width may vary slightly depending on PCB variables such as stack-up and dielectric constant (E_r) . The Z_{OD} is the governing factor. Each true signal is routed to within 25 mils of its compliment.

• If the total bus length is longer than 12.00 inches, the majority of the bus must be routed on a single layer.

The complete requirement for length matching and skew compensation is detailed in [Table 43.](#page-239-0)

Table 43: System Board Design Rules

Table 43: System Board Design Rules

B.2.4.2 L_POWER_OK, L_RST_N and LDTSTOP_N

Open-drain signals require a pull-up resistor to VDDIO, +2.5V or +3.3 V, depending on the other devices in the HT chain.

B.2.4.3 HyperTransport Trace Referencing to VSS/VDD

- HyperTransport is a ground-referenced differential bus. Differential pairs are weakly coupled and reference V_{SS} more than each other.
- Ideally, all HyperTransport signals reference V_{SS}/V_{DD} .
- If traces cross two reference planes, the plane separation should be 15 mils or less and should use one 0.01 uF 0603 or 0402 ceramic capacitor to bridge the boundary between the planes for every four diff-pairs.
- All signal groups must match the length requirement.

B.2.4.4 HyperTransport Trace Layer Changes

- In general, layer changes are discouraged for bus speeds > 400 Mb/s with the exception of vias for breakout under the package.
- If a clock/data group must change layers, place a pair's vias as close as possible in order to prevent a major Z_{OD} discontinuity. It is all right for the via anti-pad plane clearances to overlap in this instance.
- If a clock/data group is split between top and bottom routing layers, trace length must be inserted on the top layer nets to approximately match the extra delay due to the vias for the bottom layer routing.
- One 0.01 uF 0603 or 0402 ceramic decoupling capacitor is in the vicinity of layer transitions for every four differential pairs to minimize plane bounce.
- All signal groups must match the length requirement.

B.3 Power Distribution

AS90L10208 power distribution is split into V_{LDT} (+1.2 V for HT PHY), V_{CORE} (+1.8 V for the core), V_{ANALOG} (+1.8 V for the PLL circuitry), and VDDIO (+3.3 V for the I/O).

- V_{ANALOG} should be derived from V_{CORE} using a ferrite bead.
- V_{CORE} should be supplied from a dedicated plane or an island in the power plane.
- V_{ANALOG} and VDDIO should be supplied from a dedicated island located close to the pins supplied by it.

HyperTransport supports a very high data-rate transaction. Careful design practices must be exercised to ensure a low DC and AC impedance path from the regulator that supplies V_{LDF} .

VLDT to each HT link must be delivered by an independent interconnect in order to provide maximum isolation between the individual links. This will minimize the high frequency noise on the individual HyperTransport supply due to the switching of high data-rate signals on other HyperTransport links. AS90L10208 has dedicated V_{LDT} pins for each of its two, HyperTransport ports. It is recommended that a separate regulator be used for each port and that power is supplied from a dedicated island for each V_{LDT} in the power plane. Detailed information on V_{LDT} layout is provided in the following sections.

B.3.1 Number of Layers

The number of layers on the board dictates if V_{LDT} is routed on the board as a trace or as section of a plane. A larger number of layers alleviates some of the signal routing constraints allowing for placing a greater number of discrete decoupling capacitors closer to the AS90L10208 packages or on the back of the board. Ideally, power should be supplied from a dedicated island for each V_{LDT} in the power plane.

B.3.2 VLDT Layout

Regardless of the number of layers in the board, the HyperTransport voltage regulator should be as close to the AS90L10208 package as possible. The maximum recommended distance is 1.5 to 2.0 inches. If possible, a section of a plane on one of the inner signal layers should be used to deliver V_{LDT} to AS90L10208, as is done with 6- or 8-layer motherboards. This plane should be large enough to accommodate the placement of all V_{LDT} regulator components and to make connections to every V_{LDT} ball on the AS90L10208 package. The plane should be subdivided into islands, with each island used to supply an individual link.

If a V_{LDT} plane is not available, the minimum width of the trace should be 0.2 inches or greater whenever possible.

B.3.3 Decoupling

The use of decoupling capacitors on the board is recommended. They should be placed as close to the package as possible without interfering with the signal routing. This can be implemented with an array of 0603 or 0402-size capacitors. If the application permits, the capacitors should be mounted on the back of the board directly under the package.

The best way to select the number of capacitors and their values is through SPICE simulations, if the circuit model for the power delivery system is available. If the circuit model is not available, the following guidelines should be used for component selection. There should be at least eight ceramic capacitors to ground of 0.01 uF or 0.1 uF capacitance - one for each V_{LDT} power pin. All capacitors should be of X7R dielectric or better.

B.3.4 Bulk Decoupling

Proper bulk decoupling is required for stable operation of the regulator and to ensure low-power delivery system impedance at low frequencies. It is recommended that low-ESR ceramic (X5R dielectric or better), OsCon, or aluminum polymer type capacitors be used for bulk decoupling.

B.3.5 Multiple HyperTransport Links

A single V_{LDT} regulator can be used to supply power to two HyperTransport links. The distance from the regulator to the AS90L10208 package should not exceed 1.5 to 2.0 inches. This scheme is recommended only for applications where board space is constrained so much that it is impossible to accommodate individual regulators for each port.

It is important to note that although a single voltage regulator can be used to feed two independent HyperTransport links, the board routing must consist of two separate traces leading from the regulator to the package. The traces should be 0.2 inches wide and should run separately for a minimum length of at least 1.0 inch. The minimum spacing between them should not be less than 0.1 inch. If these layout recommendations are not followed, significant coupling can be expected between the V_{LDT} links if they are connected to the same regulator. Regardless of which feeding scheme is used to deliver the power to the AS90L10208, it is recommended that decoupling capacitors be placed at the point where the traces connect to the AS90L10208.

Please note that it is not recommended that a single voltage regulator be used to feed more than two HyperTransport links in any application.

B.4 AS90L10208 Die Pad-to-Ball Trace Length Information

[Table 44](#page-244-1) shows the pad-to-ball trace length inside the AS90L10208 package, in both millimeters and inches, for the HyperTransport differential signals. For convenience, the pad-to-ball propagation delay is also included. The propagation delay for each signal is 10 ps/mm.

Table 44: Pad-to-Ball Trace Length Information

Table 44: Pad-to-Ball Trace Length Information

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Table 44: Pad-to-Ball Trace Length Information

Table 44: Pad-to-Ball Trace Length Information

B.5 Example PCB Stackup for HyperTransport

The following example PCB stack-up will provide a 60-ohm single-ended impedance or a 100-ohm differential impedance if a 5 mil wide trace is used with 5 mil spacing on any signal layer.

Figure 23: PCB Stackup for HyperTransport

Notes on the Stackup Design:

- 1. The stackup is based on 5 mil nominal trace width on both internal and external layers.
- 2. Impedance is nominally 60Ω with the dielectrics shown. All internal signal layers are ½ oz copper.
- 3. All internal reference planes are 1 oz copper.
- 4. Overall board thickness is 91.6 mils measured between outer resin surfaces.

C. Ordering Information

The following table contains ordering information for the Tsi308.

Table 45: Ordering Information

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