



# **Precision Analog-to-Digital Converter (ADC) with 8051 Microcontroller and Flash Memory**

# **FEATURES**

# **ANALOG FEATURES**

- **24 Bits No Missing Codes**
- **22 Bits Effective Resolution at 10Hz − Low Noise: 75nV**
- **PGA From 1 to 128**
- **Precision On-Chip Voltage Reference**
- **8 Differential/Single-Ended Channels**
- **On-Chip Offset/Gain Calibration**
- **Offset Drift: 0.1ppm/**°**<sup>C</sup>**
- **Gain Drift: 0.5ppm/**°**<sup>C</sup>**
- **On-Chip Temperature Sensor**
- **Burnout Sensor Detection**
- **Single-Cycle Conversion**
- **Selectable Buffer Input**

# **DIGITAL FEATURES**

**Microcontroller Core**

- **8051-Compatible**
- **High-Speed Core**
	- **− 4 Clocks per Instruction Cycle**
- **DC to 33MHz**
- **Single Instruction 121ns**
- **Dual Data Pointer**

## **Memory**

- **Up To 32kB Flash Memory**
- **Flash Memory Partitioning**
- **Endurance 1M Erase/Write Cycles, 100 Year Data Retention**
- **In-System Serially Programmable**
- **External Program/Data Memory (64kB)**
- **1,280 Bytes Data SRAM**
- **Flash Memory Security**
- **2kB Boot ROM**
- **Programmable Wait State Control**

**Peripheral Features**

- **34 I/O Pins**
- **Additional 32-Bit Accumulator**
- **Three 16-Bit Timer/Counters**
- **System Timers**
- **Programmable Watchdog Timer**
- **Full-Duplex Dual USARTs**
- **Master/Slave SPI**
- **16-Bit PWM**
- **Power Management Control**
- **Idle Mode Current < 1mA**
- **Stop Mode Current < 1<sup>A</sup>**
- **Programmable Brownout Reset**
- **Programmable Low Voltage Detect**
- **24 Interrupt Sources**
- **Two Hardware Breakpoints**

## **GENERAL FEATURES**

- **Pin-Compatible with MSC1211/12/13/14**
- **Package: TQFP-64**
- **Low Power: 4mW**
- **Industrial Temperature Range: −40**°**C to +125**°**C**
- **Power Supply: 2.7V to 5.25V**

# **APPLICATIONS**

- **Industrial Process Control**
- **Instrumentation**
- **Liquid/Gas Chromatography**
- **Blood Analysis**
- **Smart Transmitters**
- **Portable Instruments**
- **Weigh Scales**
- **Pressure Transducers**
- **Intelligent Sensors**
- **Portable Applications**
- **DAS Systems**

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# **PACKAGE/ORDERING INFORMATION(1)**



(1) For the most current package and ordering information, see the Package Option Addendum at the end of this datasheet, or refer to our web site at www.ti.com.



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

# **ABSOLUTE MAXIMUM RATINGS(1)**



(1) Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to absolute maximum conditions for extended periods may affect device reliability.

## **MSC1210YX FAMILY FEATURES**



(1) All peripheral features are the same on all devices; the flash memory size is the only difference.

(2) The last digit of the part number (N) represents the onboard flash size =  $(2^N)$ kBytes.

ELECTRICAL CHARACTERISTICS: AV<sub>DD</sub> = 5V<br>All specifications from T<sub>MIN</sub> to T<sub>MAX</sub>, DV<sub>DD</sub> = +2.7V to 5.25V, f<sub>MOD</sub> = 15.625kHz, PGA = 1, Buffer ON, f<sub>DATA</sub> = 10Hz, Bipolar, and V<sub>REF</sub> = (REF IN+) – (REF IN–) = +2.5V,<br>unless



(1) Calibration can minimize these errors.

(2) The self-gain calibration cannot have a REF IN+ of more than AV<sub>DD</sub> −1.5V with Buffer ON. To calibrate gain, turn Buffer OFF.

 $^{(3)}$   $\Delta V_{\text{OUT}}$  is change in digital result.

(<sup>4)</sup> 9pF switched capacitor at f<sub>SAMP</sub> clock frequency (see Figure 14).<br><sup>(5)</sup> The input impedance for PGA = 128 is the same as that for PGA = 64 (that is, 7ΜΩ/64).



ELECTRICAL CHARACTERISTICS: AV<sub>DD</sub> = 5V (continued)<br>All specifications from T<sub>MIN</sub> to T<sub>MAX</sub>, DV<sub>DD</sub> = +2.7V to 5.25V, f<sub>MOD</sub> = 15.625kHz, PGA = 1, Buffer ON, f<sub>DATA</sub> = 10Hz, Bipolar, and V<sub>REF</sub> = (REF IN+) – (REF IN–) = + unless otherwise noted.



(1) Calibration can minimize these errors.

(2) The self-gain calibration cannot have a REF IN+ of more than AV<sub>DD</sub> −1.5V with Buffer ON. To calibrate gain, turn Buffer OFF.

(3) ∆VOUT is change in digital result.

(<sup>4)</sup> 9pF switched capacitor at f<sub>SAMP</sub> clock frequency (see Figure 14).<br><sup>(5)</sup> The input impedance for PGA = 128 is the same as that for PGA = 64 (that is, 7ΜΩ/64).

**ELECTRICAL CHARACTERISTICS: AV<sub>DD</sub> = 3V**<br>All specifications from T<sub>MIN</sub> to T<sub>MAX</sub>, DV<sub>DD</sub> = +2.7V to 5.25V, f<sub>MOD</sub> = 15.625kHz, PGA = 1, Buffer ON, f<sub>DATA</sub> = 10Hz, Bipolar, and V<sub>REF</sub> = (REF IN+) – (REF IN–) = +1.25V,<br>unl



(1) Calibration can minimize these errors.

(2) The self-gain calibration cannot have a REF IN+ of more than AV<sub>DD</sub> −1.5V with Buffer ON. To calibrate gain, turn Buffer OFF.

(3) ∆VOUT is change in digital result.

(4) 9pF switched capacitor at f<sub>SAMP</sub> clock frequency (see Figure 14).

(5) The input impedance for PGA = 128 is the same as that for PGA = 64 (that is, 7MΩ/64).



ELECTRICAL CHARACTERISTICS: AV<sub>DD</sub> = 3V (continued)<br>All specifications from T<sub>MIN</sub> to T<sub>MAX</sub>, DV<sub>DD</sub> = +2.7V to 5.25V, f<sub>MOD</sub> = 15.625kHz, PGA = 1, Buffer ON, f<sub>DATA</sub> = 10Hz, Bipolar, and V<sub>REF</sub> = (REF IN+) – (REF IN–) = +



(1) Calibration can minimize these errors.

(2) The self-gain calibration cannot have a REF IN+ of more than AV<sub>DD</sub> −1.5V with Buffer ON. To calibrate gain, turn Buffer OFF.

 $^{(3)}$   $\Delta V_{\text{OUT}}$  is change in digital result.

 $(4)$  9pF switched capacitor at  $f_{SAMP}$  clock frequency (see Figure 14).

(5) The input impedance for PGA = 128 is the same as that for PGA = 64 (that is,  $7M\Omega/64$ ).



# **DIGITAL CHARACTERISTICS: DVDD = 2.7V to 5.25V**

All specifications from  $T_{MIN}$  to  $T_{MAX}$ ,  $f_{OSC}$  = 1MHz, unless otherwise specified.



(1) Digital Brownout Detect disabled (HCR1.2 = 1), Low Voltage Detect disabled (LVDCON.3 =1). Ports configured for CMOS output low. If in External Oscillation mode, the oscillator must be disabled.

# **FLASH MEMORY CHARACTERISTICS: DV<sub>DD</sub> = 2.7V to 5.25V**





# AC ELECTRICAL CHARACTERISTICS<sup>(1)(2)</sup>: DV<sub>DD</sub> = 2.7V to 5.25V



(1) Parameters are valid over operating temperature range, unless otherwise specified.

 $(2)$  Load capacitance for Port 0, ALE, and PSEN = 100pF; load capacitance for all other outputs = 80pF.

(3) These values are characterized but not 100% production tested.

(4) In the MSC1210,  $f_{\text{OSC}} = f_{\text{CLK}}$ ,  $t_{\text{CLK}} = 1/f_{\text{osc}} =$  one oscillator clock period.

 $(5)$  t<sub>MCS</sub> is a time period related to the Stretch MOVX selection. The following table shows the value of t<sub>MCS</sub> for each stretch selection:





## **EXPLANATION OF THE AC SYMBOLS**

Each Timing Symbol has five characters. The first character is always 't' (= time). The other characters, depending on their positions, indicate the name of a signal or the logical status of that signal. The designators are:

- A-Address C-Clock
- D-Input Data
- H-Logic Level HIGH
- I-Instruction (program memory contents)
- L-Logic Level LOW, or ALE
- P-PSEN
- Q-Output Data

R-RD Signal t-Time V-Valid W-WR Signal X-No Longer a Valid Logic Level Z-Float Examples:

(1)  $t_{AVLL}$  = Time for address valid to ALE LOW.

(2)  $t_{LLPL}$  = Time for ALE LOW to  $\overline{PSEN}$  LOW.



**Figure 1. External Program Memory Read Cycle**





# **MSC1210**



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**Figure 3. External Data Memory Write Cycle**



**Figure 4. External Clock Drive CLK**



# **RESET AND POWER-ON TIMING**



**Figure 5. Reset Timing**



# **Figure 6. Parallel Flash Programming Power-On Timing (EA is ignored)**



# **Figure 7. Serial Flash Programming Power-On Timing (EA is ignored)**





## **PIN ASSIGNMENTS**



# **PIN DESCRIPTIONS**





# **PIN DESCRIPTIONS (continued)**



(1) The program mode is changed during the falling edge of the reset signal.



## **TYPICAL CHARACTERISTICS**

AV<sub>DD</sub> = +5V, DV<sub>DD</sub> = +5V, f<sub>OSC</sub> = 8MHz, PGA = 1, f<sub>MOD</sub> = 15.625Hz, Bipolar, Buffer ON, and V<sub>REF</sub> = (REF IN+) − (REF IN-) = +2.5V, unless otherwise specified.





## **TYPICAL CHARACTERISTICS (Continued)**

 $AV_{DD} = +5V$ ,  $DV_{DD} = +5V$ ,  $f_{OSC} = 8MHz$ ,  $PGA = 1$ ,  $f_{MOD} = 15.625Hz$ , Bipolar, Buffer ON, and  $V_{REF} = (REF IN+) - (REF IN-) = +2.5V$ , unless otherwise specified.





## **TYPICAL CHARACTERISTICS (Continued)**

 $AV_{DD} = +5V$ ,  $DV_{DD} = +5V$ ,  $f_{OSC} = 8MHz$ ,  $PGA = 1$ ,  $f_{MOD} = 15.625Hz$ , Bipolar, Buffer ON, and  $V_{REF} = (REF IN+) - (REF IN-) = +2.5V$ , unless otherwise specified.





## **TYPICAL CHARACTERISTICS (Continued)**

 $AV_{DD} = +5V$ ,  $DV_{DD} = +5V$ ,  $f_{OSC} = 8MHz$ ,  $PGA = 1$ ,  $f_{MOD} = 15.625Hz$ , Bipolar, Buffer ON, and  $V_{REF} = (REF IN+) - (REF IN-) = +2.5V$ , unless otherwise specified.





## **TYPICAL CHARACTERISTICS (Continued)**

 $AV_{DD}$  = +5V,  $DV_{DD}$  = +5V, f<sub>OSC</sub> = 8MHz, PGA = 1, f<sub>MOD</sub> = 15.625Hz, Bipolar, Buffer ON, and V<sub>REF</sub> = (REF IN+) – (REF IN−) = +2.5V, unless otherwise specified.



# **DESCRIPTION**

The MSC1210Yx is a completely integrated family of mixed-signal devices incorporating a high-resolution delta-sigma ADC, 8-channel multiplexer, burnout current sources, selectable buffered input, offset DAC (digital-to-analog converter), PGA (programmable gain amplifier), temperature sensor, voltage reference, 8-bit microcontroller, Flash Program Memory, Flash Data Memory, and Data SRAM, as shown in Figure 8.

On-chip peripherals include an additional 32-bit accumulator, an SPI-compatible serial port, dual USARTs, multiple digital input/output ports, watchdog timer, low-voltage detect, on-chip power-on reset, 16-bit PWM, and system timers, brownout reset, and three timer/counters.

The device accepts low-level differential or single-ended signals directly from a transducer. The ADC provides 24 bits of resolution and 24 bits of no-missing-code performance using a Sinc3 filter with a programmable sample rate. The ADC also has a selectable filter that allows for high-resolution single-cycle conversion.

The microcontroller core is 8051 instruction set compatible. The microcontroller core is an optimized 8051 core that executes up to three times faster than the standard 8051 core, given the same clock source. That makes it possible to run the device at a lower external clock frequency and achieve the same performance at lower power than the standard 8051 core.

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**TRUMENTS** 

The MSC1210Yx allows the user to uniquely configure the Flash and SRAM memory maps to meet the needs of their application. The Flash is programmable down to 2.7V using both serial and parallel programming methods. The Flash endurance is 1 million Erase/Write cycles. In addition, 1280 bytes of RAM are incorporated on-chip.

The part has separate analog and digital supplies, which can be independently powered from 2.7V to +5.25V. At +3V operation, the power dissipation for the part is typically less than 4mW. The MSC1210Yx is packaged in a TQFP-64 package.

The MSC1210Yx is designed for high-resolution measurement applications in smart transmitters, industrial process control, weigh scales, chromatography, and portable instrumentation.



**Figure 8. Block Diagram**



## **ENHANCED 8051 CORE**

All instructions in the MSC1210 family perform exactly the same functions as they would in a standard 8051. The effect on bits, flags, and registers is the same. However, the timing is different. The MSC1210 family utilizes an efficient 8051 core which results in an improved instruction execution speed of between 1.5 and 3 times faster than the original core for the same external clock speed (4 clock cycles per instruction versus 12 clock cycles per instruction, as shown in Figure 9). The internal system clock is equal to the external oscillator frequency. This translates into an effective throughput improvement of more than 2.5 times, using the same code and same external clock speed.

Therefore, a device frequency of 33MHz for the MSC1210Yx actually performs at an equivalent execution speed of 82.5MHz compared to the standard 8051 core. This allows the user to run the device at slower external clock speeds which reduces system noise and power consumption, but provides greater throughput. This performance difference can be seen in Figure 10. The timing of software loops will be faster with the MSC1210. However, the timer/counter operation of the MSC1210 may be maintained at 12 clocks per increment or optionally run at 4 clocks per increment.

The MSC1210 also provides dual data pointers (DPTRs) to speed block Data Memory moves.

Additionally, it can stretch the number of memory cycles to access external Data Memory from between two and nine instruction cycles in order to accommodate different speeds of memory or devices, as shown in Table 1. The MSC1210 provides an external memory interface with a 16-bit address bus (P0 and P2). The 16-bit address bus makes it necessary to multiplex the low address byte through the P0 port. To enhance P0 and P2 for high-speed memory access, hardware configuration control is provided to configure the ports for external memory/peripheral interface or general-purpose I/O.

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**Figure 10. Comparison of MSC1210 Timing to Standard 8051 Timing**

**Table 1. Memory Cycle Stretching. Stretching of MOVX timing as defined by MD2, MD1, and MD0 bits in CKCON register (address 8Eh).**





**Figure 9. Instruction Timing Cycle**



Furthermore, improvements were made to peripheral features that off-load processing from the core, and the user, to further improve efficiency. For instance, 32-bit accumulation can be done through the summation register to significantly reduce the processing overhead for the multiple byte data from the ADC or other sources. This allows for 32-bit addition and shifting to be accomplished in a few instruction cycles, compared to hundreds of instruction cycles through a software implementation.

## **Family Device Compatibility**

The hardware functionality and pin configuration across the MSC1210 family are fully compatible. To the user the only difference between family members is the memory configuration. This makes migration between family members simple. Code written for the MSC1210Y2 can be executed directly on an MSC1210Y3, MSC1210Y4, or

MSC1210Y5. This gives the user the ability to add or subtract software functions and to freely migrate between family members. Thus, the MSC1210 can become a standard device used across several application platforms.

## **Family Development Tools**

The MSC1210 is fully compatible with the standard 8051 instruction set. This means that the user can develop software for the MSC1210 with their existing 8051 development tools. Additionally, a complete, integrated development environment is provided with each demo board, and third-party developers also provide support.

## **Power Down Modes**

The MSC1210 can power down several of the on-chip peripherals and put the CPU into IDLE. For more information, see the Idle Mode and Stop Mode sections.



**Figure 11. MSC1210 Timing Chain and Clock Control**



# **OVERVIEW**

The MSC1210 ADC structure is shown in Figure 12. The figure lists the components that make up the ADC, along with the corresponding special function register (SFR) associated with each component.

## **ADC INPUT MULTIPLEXER**

The input multiplexer provides for any combination of differential inputs to be selected as the input channel, as shown in Figure 13. If AIN0 is selected as the positive differential input channel, any other channel can be selected as the negative differential input channel. With this method, it is possible to have up to eight fully differential input channels. It is also possible to switch the polarity of the differential input pair to negate any offset voltages.

In addition, current sources are supplied that will source or sink current to detect open or short circuits on the pins.

# **TEMPERATURE SENSOR**

On-chip diodes provide temperature sensing capability. When the configuration register for the input MUX is set to all 1s, the diodes are connected to the input of the ADC. All other channels are open.



**Figure 12. MSC1210 ADC Structure**





**Figure 13. Input Multiplexer Configuration**

## **BURNOUT DETECT**

When the Burnout Detect (BOD) bit is set in the ADC control configuration register (ADCON0 DCh), two current sources are enabled. The current source on the positive input channel sources approximately 2µA of current. The current source on the negative input channel sinks approximately 2µA. This allows for the detection of an open circuit (full-scale reading) or short circuit (small differential reading) on the selected input differential pair. Buffer should be on for sensor burnout detection.

## **ADC INPUT BUFFER**

The analog input impedance is always high, regardless of PGA setting (when the buffer is enabled). With the buffer enabled, the input voltage range is reduced and the analog power-supply current is higher. If the limitation of input voltage range is acceptable, then the buffer is always preferred.

The input impedance of the MSC1210 without the buffer is 7MΩ/PGA. The buffer is controlled by the state of the BUF bit in the ADC control register (ADCON0 DCh).

# **ADC ANALOG INPUT**

When the buffer is not selected, the input impedance of the analog input changes with ACLK clock frequency (ACLK F6h) and gain (PGA). The relationship is:

$$
Impedance (\Omega) = \left(\frac{1}{f_{SAMP} \cdot C_S}\right)
$$

$$
A_{\text{IN}}~\text{Impedance}(\Omega)~=~\left(\frac{10^6}{\text{ACLK}~\text{Frequency}}\right) \cdot \left(\frac{7 M \Omega}{\text{PGA}}\right)
$$

where 
$$
ACLK
$$
 frequency =  $\frac{f_{CLK}}{(ACLK+1)}$ 

and modelk = 
$$
f_{\text{MOD}} = \frac{f_{\text{ACLK}}}{64}
$$
.

NOTE: The input impedance for PGA = 128 is the same as that for PGA = 64 (that is,  $\frac{7M\Omega}{64}$ ).

Figure 14 shows the basic input structure of the MSC1210.



**Figure 14. Analog Input Structure**



# **ADC PGA**

The PGA can be set to gains of 1, 2, 4, 8, 16, 32, 64, or 128. Using the PGA can actually improve the effective resolution of the ADC. For instance, with a PGA of 1 on a ±2.5V full-scale range, the ADC can resolve to 1.5µV. With a PGA of 128 on a ±19mV full-scale range, the ADC can resolve to 75nV, as shown in Table 2.





(1)  $ENOB = Log<sub>2</sub>(FSR/RMS Noise) = Log<sub>2</sub>(2<sup>24</sup>) - Log<sub>2</sub>( $\sigma$ <sub>CODES</sub>)$  $= 24 - \text{Log}_2(\sigma_{\text{CODES}})$ 

# **ADC OFFSET DAC**

The analog input to the PGA can be offset (in bipolar mode) by up to half the full-scale input range of the PGA by using the ODAC register (SFR E6h). The ODAC (Offset DAC) register is an 8-bit value; the MSB is the sign and the seven LSBs provide the magnitude of the offset. Since the ODAC introduces an analog (instead of digital) offset to the PGA, using the ODAC does not reduce the range of the ADC.

## **ADC MODULATOR**

The modulator is a single-loop 2nd-order system. The modulator runs at a clock speed  $(f_{\text{MOD}})$  that is derived from the CLK using the value in the Analog Clock (ACLK) register (SFR F6h). The data rate is:

Data Rate  $=$   $\frac{f_{\text{MOD}}}{\text{Decimation}}$ Decimation Ratio

where 
$$
f_{\text{MOD}} = \frac{f_{\text{CLK}}}{(\text{ACLK} + 1) \cdot 64} = \frac{f_{\text{ACLK}}}{64}
$$

and Decimation Ratio is set in [ADCON3:ADCON2].

## **ADC CALIBRATION**

The offset and gain errors in the MSC1210, or the complete system, can be reduced with calibration. Calibration is controlled through the ADCON1 register (SFR DDh), bits CAL2:CAL0. Each calibration process takes seven  $t<sub>DATA</sub>$  (data conversion time) periods to complete. Therefore, it takes 14  $t<sub>DATA</sub>$  periods to complete both an offset and gain calibration.

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For system calibration, the appropriate signal must be applied to the inputs. The system offset command requires a zero differential input signal. It then computes an offset value that will nullify offsets in the system. The system gain command requires a positive full-scale differential input signal. It then computes a value to nullify gain errors in the system. Each of these calibrations will take seven  $t<sub>DATA</sub>$ periods to complete.

Calibration should be performed after power on. It should also be done after a change in temperature, decimation ratio, buffer, Power Supply, voltage reference, or PGA. The Offset DAC wil affect offset calibration; therefore, the value of the Offset DAC should be zero until prior to performing a calibration.

At the completion of calibration, the ADC Interrupt bit goes HIGH which indicates the calibration is finished and valid data is available.

## **ADC DIGITAL FILTER**

The Digital Filter can use either the Fast Settling, Sinc<sup>2</sup>, or Sinc<sup>3</sup> filter, as shown in Figure 15. In addition, the Auto mode changes the Sinc filter after the input channel or PGA is changed. When switching to a new channel or new PGA value, it will use the Fast Settling filter for the next two conversions (the first of which should be discarded). It will then use the  $Sinc<sup>2</sup>$  followed by the  $Sinc<sup>3</sup>$  filter to improve noise performance.



**Figure 15. Filter Step Responses**

This combines the low-noise advantage of the Sinc<sup>3</sup> filter with the quick response of the Fast Settling Time filter. The frequency response of each filter is shown in Figure 16.

## **VOLTAGE REFERENCE**

The MSC1210 can use either an internal or external voltage reference. The voltage reference selection is controlled via ADC Control Register 0 (ADCON0, SFR DCh). The default power-up configuration for the voltage reference is 2.5V internal.

The internal voltage reference can be selected as either 1.25V or 2.5V. The analog power supply  $(AV_{DD})$  must be within the specified range for the selected internal voltage reference. The valid ranges are:  $V_{RFF}$  = 2.5 internal  $(AV_{DD} = 3.3V$  to 5.25V) and  $V_{REF} = 1.25$  internal (AV<sub>DD</sub> = 2.7V to 5.25V). If the internal V<sub>REF</sub> is selected, then the REFOUT pin must be connected to REFIN+, and AGND must be connected to REFIN−. The REFOUT pin should also have a 0.1µF capacitor connected to AGND, as close as possible to the pin. If the internal  $V_{REF}$  is not used, then  $V_{RFF}$  should be disabled in ADCON0.

If the external voltage reference is selected, it can be used as either a single-ended input or differential input, for ratiometric measures. When using an external reference, it is important to note that the input current will increase for  $V_{RFF}$  with higher PGA settings and with a higher modulator frequency. The external voltage reference can be used over the input range specified in the Electrical Characteristics section.

For applications requiring higher performance than that obtainable from the internal reference, use an external precision reference such as the REF50xx. The internal reference performance can be observed in the noise (and ENOB) versus input signal graphs in the Typical Characteristics section. All the rest of the ENOB plots are obtained with the inputs shorted together. By shorting the inputs, the inherent noise performance of only the ADC can be determined and displayed. When the inputs are not shorted, the extra noise comes from the reference. As can be seen in the ENOB vs Input Signal graph, the external reference adds about 0.7 bits of noise, whereas the internal reference adds about 2.3 bits of noise. This ENOB performance of 19.4 represents 21.16 bits of noise. With an LSB of 298nV, that translates to 6.3µV, or a peak−to−peak noise of almost 42µV. An external reference provides the best noise, drift, and repeatability performance for high−precision applications.





**Figure 16. Filter Frequency Responses**



## **RESET**

The device can be reset from the following sources:

- Power-on reset
- External reset
- Software reset
- Watchdog timer reset
- Brownout reset

An external reset is accomplished by taking the RST pin high for two t<sub>OSC</sub> periods, followed by taking the RST pin low. A software reset is accomplished through the System Reset register (SRTST, 0F7h). A watchdog timer reset is enabled and controlled through Hardware Configuration Register 0 (HCR0) and the Watchdog Timer register (WDTCON, 0FFh). A brownout reset is enabled through Hardware Configuration Register 1 (HCR1). External reset, software reset, and watchdog timer reset complete after 217 clock cycles. A brownout reset completes after 215 clock cycles.

All sources of reset cause the digital pins to be pulled high from the initiation of the reset. For an external reset, taking the RST pin high stops device operation, crystal oscillation, and causes all digital pins to be pulled high from that point. Taking the RST pin low initiates the reset procedure.

A recommended external reset circuit is shown in Figure 17. The serial 10kΩ resistor is recommended for any external reset circuit configuration.



**Figure 17. Typical Reset Circuit**

## **POWER-ON RESET**

The on-chip power-on reset (POR) circuitry releases the device from reset at approximately  $DV_{DD} = 2.0V$ . The POR accommodates power-supply ramp rates as slow as 1V/10ms. To ensure proper operation, the power supply should ramp monotonically. Note that as the device is released from reset and program execution begins, the device current consumption may increase, which may result in a power-supply voltage drop. If the power supply ramps at a slower rate, is not monotonic, or a brownout condition occurs (where the supply does not drop below the 2.0V threshold), then improper device operation may occur. The on-chip brownout reset may provide benefit in these conditions.

# **BROWNOUT RESET**

The brownout reset (BOR) is enabled through Hardware Configuration Register 1 (HCR1). If the conditions for proper POR are not met or the device encounters a brownout condition that does not generate a POR, the BOR can be used to ensure proper device operation. The BOR will hold the state of the device when the power supply drops below the threshold level programmed in HCR1, and then generate a reset when the supply rises above the threshold level. Note that as the device is released from reset, and program execution begins, the device current consumption may increase, which may result in a power-supply voltage drop, which may initiate another brownout condition.

The BOR level should be chosen to match closely with the application. For example, with a high external clock frequency, the BOR level should match the minimum operating voltage range for the device, or improper operation may still occur.

Note that  $AV<sub>DD</sub>$  must rise above 2.0V for the Analog Brownout Reset function to be disabled; otherwise, it will be enabled and hold the device in reset.

The BOR voltage is not calibrated until the end of the reset cycle; therefore, the actual BOR voltage will be approxiamtely 25% higher than the selected voltage. This can create a condition where the reset never ends (for example, when selecting a 4.5V BOR voltage for a 5V power supply).



## **IDLE MODE**

Idle mode is entered by setting the IDLE bit in the Power Control register (PCON, 087h). In Idle mode, the CPU, Timer0, Timer1, and USARTs are stopped, but all other peripherals and digital pins remain active. The device can be returned to active mode via an active internal or external interrupt. This mode is typically used for reducing power consumption between ADC samples.

By configuring the device prior to entering Idle mode, further power reductions can be achieved (while in Idle mode). These reductions include powering down peripherals not in use in the PDCON register (0F1h).

## **STOP MODE**

Stop mode is entered by setting the STOP bit in the Power Control register (PCON, 087h). In Stop mode, all internal clocks are halted. This mode has the lowest power consumption. The device can be returned to active mode only via an external or power-on reset.

By configuring the device prior to entering Stop mode, further power reductions can be achieved (while in Stop mode). These power reductions include halting the external clock into the device, configuring all digital I/O pins as open drain with low output drive, disabling the ADC buffer, disabling the internal  $V_{REF}$ , and setting PDCON to 0FFh to power down all peripherals.

In Stop mode, if the brownout reset is enabled, there is approximately 25µA of draw from the power supply. To achieve zero current ( $\approx$  100nA) in Stop mode, disable the brownout reset via HCR1.

In Stop mode, all digital pins retain their values.

## **POWER CONSUMPTION CONSIDERATIONS**

The following suggestions will reduce current consumption:

- 1. Use the lowest supply voltage that will work in the application for both  $AV<sub>DD</sub>$  and  $DV<sub>DD</sub>$ .
- 2. Use the lowest clock frequency that will work in the application.
- 3. Use Idle mode and the system clock divider whenever possible. Note that the system clock divider also affects the ADC clock.
- 4. Avoid using 8051-compatible I/O mode on the I/O ports. The internal pull-up resistors will draw current when the outputs are low.
- 5. Use the delay line for Flash Memory control by setting the FRCM bit in the FMCON register (SFR EEh)
- 6. Power down peripherals when they are not needed. Refer to SFR PDCON, LVDCON, and ADCON0.

## **MEMORY MAP**

The MSC1210 contains on-chip SFR, Flash Memory, Scratchpad SRAM Memory, Boot ROM, and SRAM. THe SFR registers are primarily used for control and status. The standard 8051 features and additional peripheral features of the MSC1210 are controlled through the SFR. Reading from an undefined SFR and writing to undefined SFR registers is not recommended, and will have indeterminate effects.

Flash Memory is used for both Program Memory and Data Memory. The user has the ability to select the partition size of Program and Data Memories. The partition size is set through hardware configuration bits, which are programmed through either the parallel or serial programming methods. Both Program and Data Flash Memories are erasable and writable (programmable) in User Application mode (UAM). However, program execution can only occur from Program Memory. As an added precaution, a lock feature can be activated through the hardware configuration bits, which disables erase and writes to 4kB of Program Flash Memory or the entire Program Flash Memory in UAM.

The MSC1210 includes 1kB of SRAM on-chip. SRAM starts at address 0 and is accessed through the MOVX instruction. This SRAM can also be located to start at 8400h and can be accessed as both Program and Data Memory.



The page size for Flash memory is 128 bytes. The respective page must be erased before it can be written to, regardless of whether it is mapped to Program or Data Memory space. The MSC1210 uses a memory addressing scheme that separates Program Memory (FLASH/ROM) from Data Memory (FLASH/RAM). Each area is 64kB beginning at address 0000h and ending at FFFFh, as shown in Figure 18. The program and data segments can overlap since they are accessed in different ways. Program Memory is fetched by the microcontroller automatically. There is one instruction (MOVC) that is used to explicitly read the program area. This is commonly used to read lookup tables.

The Data Memory area is accessed explicitly using the MOVX instruction. This instruction provides multiple ways of specifying the target address. It is used to access the 64kB of Data Memory. The address and data range of devices with on-chip Program and Data Memory overlap the 64kB memory space. When on-chip memory is enabled, accessing memory in the on-chip range will cause the device to access internal memory. Memory accesses beyond the internal range will be addressed externally via Ports 0 and 2.

The MSC1210 has two Hardware Configuration registers (HCR0 and HCR1) that are programmable only during Flash Memory Programming mode.



**Figure 18. Memory Map**

The MSC1210 allows the user to partition the Flash Memory between Program Memory and Data Memory. For instance, the MSC1210Y5 contains 32kB of Flash Memory on-chip. Through the HW configuration registers, the user can define the partition between Program Memory (PM) and Data Memory (DM), as shown in Table 3 and Table 4. The MSC1210 family offers four memory configurations, as shown.







It is important to note that the Flash Memory is readable and writable by the user through the MOVX instruction when configured as either Program or Data Memory (via the MXWS bit in the MWS, SFR 8Fh). This means that the user may partition the device for maximum Flash Program Memory size (no Flash Data Memory) and use Flash Program Memory as Flash Data Memory. This may lead to undesirable behavior if the PC points to an area of Flash Program Memory that is being used for data storage. Therefore, it is recommended to use Flash partitioning when Flash Memory is used for data storage. Flash partitioning prohibits execution of code from Data Flash Memory. Additionally, the Program Memory erase/write can be disabled through hardware configuration bits (HCR0), while still providing access (read/write/erase) to Data Flash Memory.

The effect of memory mapping on Program and Data Memory is straightforward. The Program Memory is decreased in size from the top of internal Program Memory. Therefore, if the MSC1210Y5 is partitioned with 31kB of Flash Program Memory and 1kB of Flash Data Memory, external Program Memory execution will begin at 7C00h (versus 8000h for 32kB). The Flash Data Memory is added on top of the SRAM memory. Therefore, access to Data Memory (through MOVX) will access SRAM for addresses 0000h−03FFh and access Flash Memory for addresses 0400h−07FFh.

## **Data Memory**

The MSC1210 can address 64kB of Data Memory. The MOVX instruction is used to access the Data SRAM Memory. This includes 1,024 bytes of on-chip Data SRAM Memory. The data bus values do not appear on Port 0 (during data bus timing) for internal memory access.

The MSC1210 also has on-chip Flash Data Memory which is readable and writable (depending on Memory Write Select register) during normal operation (full  $V_{DD}$  range). This memory is mapped into the external Data Memory space directly above the SRAM.

The MOVX instruction is used to write the flash memory. Flash memory must be erased before it can be written. Flash memory is erased in 128 byte pages.





## **CONFIGURATION MEMORY**

The MSC1210 Configuration Memory consists of 128 bytes. In UAM, all Configuration Memory is readable using the faddr data read Boot ROM routine, and the CADDR and CDATA registers. In UAM, however, none of the Configuration Memory is writable.

In serial or parallel programming mode, all Configuration Memory is readable. Most locations are also writable, except for addresses 8070h through 8079h, which are read-only.

The two hardware configuration registers reside in configuration memory at 807Eh (HCR1) and 807Fh (HCR0).

Figure 19 shows the configuration memory mapping for programming mode and UAM. Note that reading/writing configuration memory in Flash Programming mode (FPM) requires 16-bit addressing, whereas reading configuration memory in User Application mode (UAM) requires only 8-bit addressing.



**Figure 19. Configuration Memory Map**

## **REGISTER MAP**

The Register Map is illustrated in Figure 20. It is entirely separate from the Program and Data Memory areas mentioned before. A separate class of instructions is used to access the registers. There are 256 potential register locations. In practice, the MSC1210 has 256 bytes of Scratchpad RAM and up to 128 SFRs. This is possible, since the upper 128 Scratchpad RAM locations can only SBAS203J − MARCH 2002 − REVISED JANUARY 2008

be accessed indirectly. Thus, a direct reference to one of the upper 128 locations must be an SFR access. Direct RAM is reached at locations 0 to 7Fh (0 to 127).



**Figure 20. Register Map**

SFRs are accessed directly between 80h and FFh (128 to 255). The RAM locations between 128 and 255 can be reached through an indirect reference to those locations. Scratchpad RAM is available for general-purpose data storage. It is commonly used in place of off-chip RAM when the total data contents are small. When off-chip RAM is needed, the Scratchpad area will still provide the fastest general-purpose access. Within the 256 bytes of RAM, there are several special-purpose areas.

## **Bit Addressable Locations**

In addition to direct register access, some individual bits are also accessible. These are individually addressable bits in both the RAM and SFR area. In the Scratchpad RAM area, registers 20h to 2Fh are bit addressable. This provides 128 (16  $\times$  8) individual bits available to software. A bit access is distinguished from a full-register access by the type of instruction. In the SFR area, any register location ending in a 0 or 8 is bit addressable. Figure 21 shows details of the on-chip RAM addressing including the locations of individual RAM bits.

## **Working Registers**

As part of the lower 128 bytes of RAM, there are four banks of Working Registers, as shown in Figure 21. The Working Registers are general-purpose RAM locations that can be addressed in a special way. They are designated R0

through R7. Since there are four banks, the currently selected bank will be used by any instruction using R0—R7. This allows software to change context by simply switching banks. This is controlled via the Program Status Word register (PSW; 0D0h) in the SFR area described below. Registers R0 and R1 also allow their contents to be used for indirect addressing of the upper 128 bytes of RAM. Thus, an instruction can designate the value stored in R0 (for example) to address the upper RAM. The 16 bytes immediately above the R0—R7 registers are bit addressable; any of the 128 bits in this area can be directly accessed using bit addressable instructions.



**Figure 21. Scratchpad Register Addressing**

## **Stack**

Another use of the Scratchpad area is for the programmer's stack. This area is selected using the Stack Pointer (SP; 81h) SFR. Whenever a call or interrupt is invoked, the return address is placed on the Stack. It also is available to the programmer for variables, etc., since the Stack can be moved and there is no fixed location within the RAM designated as Stack. The Stack Pointer will default to 07h on reset. The user can then move it as needed. A convenient location would be the upper RAM area (> 7Fh) since this is only available indirectly. The SP will point to the last used value. Therefore, the next value placed on the Stack is put at SP + 1. Each PUSH or CALL will increment the SP by the appropriate value. Each POP or RET will decrement as well.

## **Program Memory**

After reset, the CPU begins execution from Program Memory location 0000h. The selection of where Program Memory execution begins is made by tying the  $\overline{EA}$  pin to  $DV_{DD}$  for internal access, or DGND for external access. When EA is tied to  $DV_{DD}$ , any PC fetches outside the internal Program Memory address occur from external memory. If  $\overline{EA}$  is tied to DGND, then all PC fetches address external memory. The standard internal Program Memory size for MSC1210 family members is shown in Table 5. If enabled the Boot ROM will appear from address F800h to FFFFh.





## **Boot ROM**

There is a 2kB Boot ROM that controls operation during serial or parallel programming. The Boot ROM routines can be accessed during the user mode if it is enabled. The Boot ROM routines are listed in Table 6. When enabled, the Boot ROM routines will be located at memory addresses F800h−FFFFh during user mode. In program mode the Boot ROM is located in the first 2kB of Program Memory. For additional information, refer to Application Note SBAA085, MSC1210 ROM Routines, available for download from the TI web site (www.ti.com).

## **Table 6. MSC1210 Boot ROM Routines**





## **ACCESSING EXTERNAL MEMORY**

If external memory is used, P0 and P2 can be configured as address and data lines. If external memory is not used, P0 and P2 can be configured as general-purpose I/O lines through the Hardware Configuration Register.

To enable access to external memory, bits 0 and 1 of the HCR1 register must be set to 0. When these bits are enabled all memory addresses for both internal and external memory will appear on ports 0 and 2. During the data portion of the cycle for internal memory, Port 0 will be zero for security purposes.

Accesses to external memory are of two types: accesses to external Program Memory and accesses to external Data Memory. Accesses to external Program Memory use signal PSEN (program store enable) as the read strobe. Accesses to external Data Memory use RD or WR (alternate functions of P3.7 and P3.6) to strobe the memory.

External Program Memory and external Data Memory may be combined if desired by applying the  $\overline{RD}$  and  $\overline{PSEN}$ signals to the inputs of an AND gate and using the output of the gate as the read strobe to the external Program/Data Memory.

Program fetches from external Program Memory always use a 16-bit address. Accesses to external Data Memory can use either a 16-bit address (MOVX @DPTR) or an 8-bit address (MOVX @R<sub>I</sub>).

If Port 2 is selected for external memory use (HCR1, bit 0), it cannot be used as general-purpose I/O. This bit (or Bit 1 of HCR1) also forces bits P3.6 and P3.7 to be used for WR and RD instead of I/O. Port 2, P3.6, and P3.7 should all be written to '1.'

If an 8-bit address is being used (MOVX  $@R<sub>l</sub>$ ), the contents of the MPAGE (92h) SFR remain at the Port 2 pins throughout the external memory cycle. This will facilitate paging.

In any case, the low byte of the address is time-multiplexed with the data byte on Port 0. The ADDR/DATA signals use CMOS drivers in the Port 0, Port 2, WR, and RD output buffers. Thus, in this application the Port 0 pins are not open-drain outputs, and do not require external pull-ups for high-speed access. Signal ALE (Address Latch Enable) should be used to capture the address byte into an external latch. The address byte is valid at the negative transition of ALE. Then, in a write cycle, the data byte to be written appears on Port 0 just before  $\overline{\text{WR}}$  is activated, and remains there until after  $\overline{WR}$  is deactivated. In a read cycle, the incoming byte is accepted at Port 0 just before the read strobe is deactivated.

The functions of Port 0 and Port 2 are selected in Hardware Configuration Register 1. This can only be changed during the Flash Program mode. There is no conflict in the use of these registers; they will either be used as general-purpose I/O or for external memory access. The default state is for Port 0 and Port 2 to be used as general-purpose I/O. If an external memory access is attempted when they are configured as general-purpose I/O, the values of Port 0 and Port 2 will not be affected.

External Program Memory is accessed under two conditions:

- 1. Whenever signal  $\overline{EA}$  is LOW during reset, then all future accesses are external; or
- 2. Whenever the Program Counter (PC) contains a number that is outside of the internal Program Memory address range, if the ports are enabled.

If Port 0 and Port 2 are selected for external memory, all 8 bits of Port 0 and Port 2, as well as P3.6 and P3.7, are dedicated to an output function and may not be used for general-purpose I/O. During external program fetches, Port 2 outputs the high byte of the PC.

## **Programming Flash Memory**

There are four sections of Flash Memory for programming:

- 1. 128 configuration bytes.
- 2. Reset sector (4kB) (not to be confused with the 2kB Boot ROM).
- 3. Program Memory.
- 4. Data Memory.

## **Flash Programming Mode**

There are two programming modes: parallel and serial. The programming mode is selected by the state of the ALE and PSEN signals during power-on reset. Serial programming mode is selected with  $\overline{PSEN} = 0$  and ALE = 1. Parallel programming mode is selected with  $\overline{PSEN} = 1$  and  $ALE = 0$ (see Figure 22). If they are both HIGH, the MSC1210 will operate in normal user mode. Both signals LOW is a reserved mode and is not defined. Programming mode is exited with a reset (BOR, WDT, software, or POR) and the normal mode selected.





**Figure 22. Parallel Programming Configuration**

The MSC1210 is shipped with Flash Memory erased (all 1s). Parallel programming methods typically involve a third-party programmer. Serial programming methods typically involve in-system programming. UAM allows Flash Program and Data Memory programming. The actual code for Flash programming cannot execute from Flash. That code must execute from the Boot ROM, internal (von Neumann) RAM or external memory.

Figure 23 shows the serial programming conection.

Serial programming mode works through USART0, and has special protocols, which are discussed at length in Application Note SBAA076, Programming the MSC1210, available for download at www.ti.com. The serial programming mode works at a maximum baud rate determined by  $f_{\rm OSC}$ .



**Figure 23. Serial Programming Connection**



## **INTERRUPTS**

The MSC1210 uses a three-priority interrupt system. As shown in Table 7, each interrupt source has an independent priority bit, flag, interrupt vector, and enable (except that nine interrupts share the Auxiliary Interrupt [AI] at the highest priority). In addition, interrupts can be globally enabled or disabled. The interrupt structure is compatible with the original 8051 family. All of the standard interrupts are available.

# **HARDWARE CONFIGURATION MEMORY**

The 128 configuration bytes can only be written during the program mode. The bytes are accessed through SFR registers CADDR (SFR 93h) and CDATA (SFR 94h). Two of the configuration bytes control Flash partitioning and system control. If the security bit is set, these bits can not be changed except with a Mass Erase command that erases all of the Flash Memory including the 128 configuration bytes.



## **Table 7. Interrupt Summary**

(1) These interrupts set the AI flag (EICON.4) and are enabled by EAI (EICON.5).

(2) If edge-triggered, cleared automatically by hardware when the service routine is vectored to. If level-triggered, the flag follows the state of the pin.

(3) Cleared automatically by hardware when interrupt vector occurs.

(4) Globally enabled by EA (IE.7).


# **Hardware Configuration Register 0 (HCR0)—Accessed Using SFR Registers CADDR and CDATA.**



**NOTE:** HCR0 is programmable only in Flash Programming mode, but can be read in User Application mode using the CADDR and CDATA SFRs or the **faddr\_data\_read** Boot ROM routine.

#### **EPMA Enable Programming Memory Access (Security Bit).**

bit 7 0: After reset in programming modes, Flash Memory can only be accessed in UAM until a mass erase is done. 1: Fully Accessible (default)

## PML Program Memory Lock (PML has Priority Over RSL).

- bit 6 0: Enable all Flash Programming modes in program mode, can be written in UAM. 1: Enable read-only for program mode; cannot be written in UAM (default).
- **RSL Reset Sector Lock.** The reset sector can be used to provide another method of Flash Memory programming. This bit 5 will allow Program Memory updates without changing the jumpers for in-circuit code updates or program development. The code in this boot sector would then provide the monitor and programming routines with the ability to jump into the main Flash code when programming is finished.
	- 0: Enable Reset Sector Writing
	- 1: Enable Read-Only Mode for Reset Sector (4kB) (default)
- **EBR** Enable Boot ROM. Boot ROM is 2kB of code located in ROM, not to be confused with the 4kB Boot Sector located bit 4 in Flash Memory.
	- 0: Disable Internal Boot ROM
		- 1: Enable Internal Boot ROM (default)

### **EWDR Enable Watchdog Reset.**

- bit 3 0: Disable Watchdog Reset
	- 1: Enable Watchdog Reset (default)

#### **DFSEL Data Flash Memory Size (see Table 3 and Table 4).**

- bits 2−0 000: Reserved
	- 001: 32kB, 16kB, 8kB, or 4kB Data Flash Memory
	- 010: 16kB, 8kB, or 4kB Data Flash Memory
	- 011: 8kB or 4kB Data Flash Memory
	- 100: 4kB Data Flash Memory
	- 101: 2kB Data Flash Memory
	- 110: 1kB Data Flash Memory
	- 111: No Data Flash Memory (default)



# **Hardware Configuration Register 1 (HCR1)**



**NOTE:** HCR1 is programmable only in Flash Programming mode, but can be read in User Application mode using the CADDR and CDATA SFRs or the **faddr\_data\_read** Boot ROM routine.

# **DBLSEL Digital Brownout Level Select** bits 7−6 00: 4.5V 01: 4.2V 10: 2.7V 11: 2.5V (default) **ABLSEL Analog Brownout Level Select** bits 5−4 00: 4.5V 01: 4.2V 10: 2.7V 11: 2.5V (default) **DAB Disable Analog Power-Supply Brownout Reset** bit 3 0: Enable Analog Brownout Reset 1: Disable Analog Brownout Reset (default) (will not disable unless  $AV<sub>DD</sub> > 2.0V$ ) **DDB Disable Digital Power-Supply Brownout Reset** bit 2 0: Enable Digital Brownout Reset 1: Disable Digital Brownout Reset (default) **EGP0 Enable General-Purpose I/O for Port 0** bit 1 0: Port 0 is Used for External Memory, P3.6 and P3.7 Used for  $\overline{\text{WR}}$  and  $\overline{\text{RD}}$ . 1: Port 0 is Used as General-Purpose I/O (default) **EGP23 Enable General-Purpose I/O for Ports 2 and 3** bit 0 0: Port 2 is Used for External Memory, P3.6 and P3.7. Used for  $\overline{\text{WR}}$  and  $\overline{\text{RD}}$ . 1: Port 2 and Port3 are Used as General-Purpose I/O (default)

# **Configuration Memory Programming**

Certain key functions such as Brownout Reset and Watchdog Timer are controlled by the hardware configuration bits. These bits are nonvolatile and can only be changed through serial and parallel programming. Other peripheral control and status functions, such as ADC configuration, timer setup, and Flash control, are controlled through the SFRs.



# **SFR Definitions** (**Boldface** definitions indicate that the register is unique to the MSC1210Yx)





**SFR Definitions (continued)** (**Boldface** definitions indicate that the register is unique to the MSC1210Yx)





# **SFR Definitions (continued)** (**Boldface** definitions indicate that the register is unique to the MSC1210Yx)











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**POWER**



FMCON EEh Flash Memory Control (XX) and the state of FTCON EFh Flash Memory Timing Control X X X

# **Table 8. Special Function Register Cross Reference (continued)**



# **Table 8. Special Function Register Cross Reference (continued)**





# **Port 0 (P0)**



**P0.7−0** Port 0. This port functions as a multiplexed address/data bus during external memory access, and as a generalbits 7−0 purpose I/O port when external memory access is not needed. During external memory cycles, this port will contain the LSB of the address when ALE is HIGH, and Data when ALE is LOW. When used as a general-purpose I/O, this port drive is selected by P0DDRL and P0DDRH (ACh, ADh). Whether Port 0 is used as general-purpose I/O or for external memory access is determined by the Flash Configuration Register (HCR1.1)

# **Stack Pointer (SP)**



**SP.7−0 Stack Pointer.** The stack pointer identifies the location where the stack will begin. The stack pointer is incremented bits 7−0 before every PUSH or CALL operation and decremented after each POP or RET/RETI. This register defaults to 07h after reset.

# **Data Pointer Low 0 (DPL0)**



**DPL0.7−0 Data Pointer Low 0.** This register is the low byte of the standard 8051 16-bit data pointer. DPL0 and DPH0 are bits 7−0 used to point to non-scratchpad data RAM. The current data pointer is selected by DPS (SFR 86h).

### **Data Pointer High 0 (DPH0)**



**DPH0.7−0 Data Pointer High 0.** This register is the high byte of the standard 8051 16-bit data pointer. DPL0 and DPH0 are bits 7−0 used to point to non-scratchpad data RAM. The current data pointer is selected by DPS (SFR 86h).

### **Data Pointer Low 1 (DPL1)**



**DPL1.7−0 Data Pointer Low 1.** This register is the low byte of the auxiliary 16-bit data pointer. When the SEL bit (DPS.0, bits 7−0 SFR 86h) is set, DPL1 and DPH1 are used in place of DPL0 and DPH0 during DPTR operations.



# **Data Pointer High 1 (DPH1)**



**DPH1.7−0 Data Pointer High.** This register is the high byte of the auxiliary 16-bit data pointer. When the SEL bit (DPS.0, bits 7−0 SFR 86h) is set, DPL1 and DPH1 are used in place of DPL0 and DPH0 during DPTR operations.

## **Data Pointer Select (DPS)**



**SEL Data Pointer Select.** This bit selects the active data pointer.

bit 0 0: Instructions that use the DPTR will use DPL0 and DPH0.

1: Instructions that use the DPTR will use DPL1 and DPH1.

# **Power Control (PCON)**



# **Timer/Counter Control (TCON)**





### **Timer Mode Control (TMOD)**



#### **GATE** Timer 1 Gate Control. This bit enables/disables the ability of Timer 1 to increment.

bit 7 0: Timer 1 will clock when  $TR1 = 1$ , regardless of the state of pin  $\overline{\text{INT1}}$ .

1: Timer 1 will clock only when  $TR1 = 1$  and pin  $\overline{INT1} = 1$ .

# **C/T Timer 1 Counter/Timer Select.**

bit 6 0: Timer is incremented by internal clocks.

1: Timer is incremented by pulses on T1 pin when TR1 (TCON.6, SFR 88h) is 1.

#### **M1, M0 Timer 1 Mode Select.** These bits select the operating mode of Timer 1.

bits 5−4



- **GATE** Timer 0 Gate Control. This bit enables/disables the ability of Timer 0 to increment.
- bit 3 0: Timer 0 will clock when  $TR0 = 1$ , regardless of the state of pin  $\overline{INT0}$  (software control). 1: Timer 0 will clock only when  $TR0 = 1$  and pin  $\overline{INT0} = 1$  (hardware control).

# **C/T Timer 0 Counter/Timer Select.**

bit 2 0: Timer is incremented by internal clocks.

1: Timer is incremented by pulses on pin T0 when TR0 (TCON.4, SFR 88h) is 1.

**M1, M0 Timer 0 Mode Select.** These bits select the operating mode of Timer 0.

#### bits 1−0



### **Timer 0 LSB (TL0)**



**TL0.7−0 Timer 0 LSB.** This register contains the least significant byte of Timer 0.

bits 7−0

# **Timer 1 LSB (TL1)**



**TL1.7−0 Timer 1 LSB.** This register contains the least significant byte of Timer 1.

bits 7−0

# **Timer 0 MSB (TH0)**



**TH0.7−0 Timer 0 MSB.** This register contains the most significant byte of Timer 0.

bits 7−0

# **Timer 1 MSB (TH1)**



**TH1.7−0 Timer 1 MSB.** This register contains the most significant byte of Timer 1.

bits 7−0

# **Clock Control (CKCON)**



**T2M Timer 2 Clock Select.** This bit controls the division of the system clock that drives Timer 2. This bit has no effect when bit 5 the timer is in baud rate generator or clock output mode. Clearing this bit to 0 maintains 8051 compatibility. This bit has no effect on instruction cycle timing.

0: Timer 2 uses a divide-by-12 of the crystal frequency.

1: Timer 2 uses a divide-by-4 of the crystal frequency.

- **T1M** Timer 1 Clock Select. This bit controls the division of the system clock that drives Timer 1. Clearing this bit to 0 bit 4 maintains 8051 compatibility. This bit has no effect on instruction cycle timing.
	- 0: Timer 1 uses a divide-by-12 of the crystal frequency.

1: Timer 1 uses a divide-by-4 of the crystal frequency.

- **T0M** Timer 0 Clock Select. This bit controls the division of the system clock that drives Timer 0. Clearing this bit to 0 bit 3 maintains 8051 compatibility. This bit has no effect on instruction cycle timing.
	- 0: Timer 0 uses a divide-by-12 of the crystal frequency.
	- 1: Timer 0 uses a divide-by-4 of the crystal frequency.
- **MD2, MD1, MD0 Stretch MOVX Select 2−0.** These bits select the time by which external MOVX cycles are to be stretched. This bits 2−0 allows slower memory or peripherals to be accessed without using ports or manual software intervention. The width of the RD or WR strobe will be stretched by the specified interval, which will be transparent to the software except for the increased time to execute the MOVX instruction. All internal MOVX instructions on devices containing MOVX SRAM are performed at the 2 instruction cycle rate.



(1) For applications without external memory, no extra cycle is needed. To increase speed, set MD2, MD1, and MD0 to '000'.



### **Memory Write Select (MWS)**



#### **MXWS MOVX Write Select. This allows writing to the internal Flash program memory.**

bit 0 0: No writes are allowed to the internal Flash program memory.

1: Writing is allowed to the internal Flash program memory, unless PML (HCR0) or RSL (HCR0) are on.

## **Port 1 (P1)**



**P1.7−0** General-Purpose I/O Port 1. This register functions as a general-purpose I/O port. In addition, all the pins have an bits 7−0 alternative function listed below. Each of the functions is controlled by several other SFRs. The associated Port 1 latch bit must contain a logic '1' before the pin can be used in its alternate function capacity. To use the alternate function, set the appropriate mode in P1DDRL (SFR AEh), P1DDRH (SFR AFh).

**INT5/SCK External Interrupt 5.** A falling edge on this pin will cause an external interrupt 5 if enabled.

- bit 7 **SPI Clock.** The master clock for SPI data transfers.
- **INT4/MISO External Interrupt 4.** A rising edge on this pin will cause an external interrupt 4 if enabled.
- bit 6 **Master In Slave Out.** For SPI data transfers, this pin receives data for the master and transmits data from the slave.
- **INT3/MOSI External Interrupt 3.** A falling edge on this pin will cause an external interrupt 3 if enabled.
- bit 5 **Master Out Slave In.** For SPI data transfers, this pin transmits master data and receives slave data.
- **INT2/SS External Interrupt 2.** A rising edge on this pin will cause an external interrupt 2 if enabled**.** bit 4 **Slave Select.** During SPI operation, this pin provides the select signal for the slave device but does not control the output drive of MISO.
- **TXD1** Serial Port 1 Transmit. This pin transmits the serial Port 1 data in serial port modes 1, 2, 3, and emits the synchrobit 3 hizing clock in serial port mode 0.
- **RXD1** Serial Port 1 Receive. This pin receives the serial Port 1 data in serial port modes 1, 2, 3, and is a bidirectional data bidirectional data transfer pin in serial port mode 0.
- **T2EX Timer 2 Capture/Reload Trigger.** A 1 to 0 transition on this pin will cause the value in the T2 registers to be bit 1 transferred into the capture registers, if enabled by EXEN2 (T2CON.3, SFR C8h). When in auto-reload mode, a 1 to 0 transition on this pin will reload the Timer 2 registers with the value in RCAP2L and RCAP2H if enabled by EXEN2 (T2CON.3, SFR C8h).
- **T2** Timer 2 External Input. A 1 to 0 transition on this pin will cause Timer 2 to increment.

bit 0

# **External Interrupt Flag (EXIF)**



**IE5** External Interrupt 5 Flag. This bit will be set when a falling edge is detected on INT5. This bit must be cleared bit 7 manually by software. Setting this bit in software will cause an interrupt if enabled.

**IE4** External Interrupt 4 Flag. This bit will be set when a rising edge is detected on INT4. This bit must be cleared bit 6 manually by software. Setting this bit in software will cause an interrupt if enabled.

**IE3** External Interrupt 3 Flag. This bit will be set when a falling edge is detected on INT3. This bit must be cleared bit 5 manually by software. Setting this bit in software will cause an interrupt if enabled.

**IE2** External Interrupt 2 Flag. This bit will be set when a rising edge is detected on INT2. This bit must be cleared bit 4 manually by software. Setting this bit in software will cause an interrupt if enabled.

### **Memory Page (MPAGE)**



**MPAGE** The 8051 uses Port 2 for the upper 8 bits of the external data memory access by MOVX A, @Ri and MOVX @Ri, A bits 7−0 instructions. The MSC1210 uses register MPAGE instead of Port 2. To access external data memory using the MOVX A,@Ri and MOVX @Ri,A instructions, the user should preload the upper byte of the address into MPAGE (versus preloading into P2 for the standard 8051).

### **Configuration Address Register (CADDR) (write-only)**



**CADDR Configuration Address Register.** This register supplies the address for reading bytes in the 128 bytes of Flash bits 7−0 Configuration memory. This is a write-only register.

**CAUTION: If this register is written to while executing from Flash Memory, the CDATA register will be incorrect. The faddr\_data\_read routine in the Boot ROM can be used for this purpose.**

#### **Configuration Data Register (CDATA) (read-only)**



**CDATA Configuration Data Register.** This register will contain the data in the 128 bytes of Flash Configuration memory that bits 7−0 are located at the last written address in the CADDR register. This is a read-only register.



## **Memory Control (MCON)**



#### **BPSEL Breakpoint Address Selection**

bit 7 Write: Select one of two Breakpoint registers: 0 or 1.

0: Select breakpoint register 0.

1: Select breakpoint register 1.

Read: Provides the Breakpoint register that created the last interrupt: 0 or 1.

#### **RAMMAP Memory Map 1kB extended SRAM.**

- bit 0 0: Address is: 0000h—03FFh (default) (Data Memory)
	- 1: Address is 8400h—87FFh (Data and Program Memory)

# **Serial Port 0 Control (SCON0)**



**SM0−2 Serial Port 0 Mode.** These bits control the mode of serial Port 0. Modes 1, 2, and 3 have 1 start and 1 stop bit in bits 7−5 addition to the 8 or 9 data bits.



(2) For modes 1 and 3, the selection of Timer 1 or 2 for baud rate is specified via the T2CON (C8h) register.

(3) RI\_0 will only be activated when a valid STOP is received. (4)  $RI_0$  will not be activated if bit  $9 = 0$ .

**REN\_0 Receive Enable.** This bit enables/disables the serial Port 0 received shift register.

bit 4 0: Serial Port 0 reception disabled.

1: Serial Port 0 received enabled (modes 1, 2, and 3). Initiate synchronous reception (mode 0).

- **TB8\_0 9th Transmission Bit State.** This bit defines the state of the 9th transmission bit in serial Port 0 modes 2 and 3.
- bit 3
- **RB8\_0 9th Received Bit State.** This bit identifies the state of the 9th reception bit of received data in serial Port 0 modes bit 2 2 and 3. In serial port mode 1, when SM2\_0 = 0, RB8\_0 is the state of the stop bit. RB8\_0 is not used in mode 0.
- **TI\_0 Transmitter Interrupt Flag.** This bit indicates that data in the serial Port 0 buffer has been completely shifted out. In serial bit 1 port mode 0, TI\_0 is set at the end of the 8th data bit. In all other modes, this bit is set at the end of the last data bit. This bit must be manually cleared by software.
- **RI\_0 Receiver Interrupt Flag.** This bit indicates that a byte of data has been received in the serial Port 0 buffer. In serial bit 0 port mode 0, RI\_0 is set at the end of the 8th bit. In serial port mode 1, RI\_0 is set after the last sample of the incoming stop bit subject to the state of SM2\_0. In modes 2 and 3, RI\_0 is set after the last sample of RB8\_0. This bit must be manually cleared by software.

## **Serial Data Buffer 0 (SBUF0)**



**SBUF0** Serial Data Buffer 0. Data for Serial Port 0 is read from or written to this location. The serial transmit and receive bits 7−0 buffers are separate registers, but both are addressed at this location.

# **SPI Control (SPICON). Any change resets the SPI interface, counters, and pointers. PDCON controls which is enabled.**



**SCK Selection.** Selection of t<sub>CLK</sub> divider for generation of SCK in Master mode.

bits 7−5



#### **ORDER Set Bit Order for Transmit and Receive.**

- bit 3 0: Most Significant Bits First
	- 1: Least Significant Bits First

### **MSTR SPI Master Mode.**

- bit 2 0: Slave Mode
	- 1: Master Mode

### **CPHA Serial Clock Phase Control.**

bit 1 0: Valid data starting from half SCK period before the first edge of SCK 1: Valid data starting from the first edge of SCK

# **CPOL Serial Clock Polarity.**

bit 0 0: SCK idle at logic LOW 1: SCK idle at logic HIGH

### **SPI Data Register (SPIDATA)**



**SPIDATA SPI Data Register.** Data for SPI is read from or written to this location. The SPI transmit and receive buffers are bits 7−0 separate registers, but both are addressed at this location. Read to clear the receive interrupt and write to clear the transmit interrupt.



## **SPI Transmit Control Register (SPITCON)**



### **CLK\_EN SCK Driver Enable.**

- bit 5 0: Disable SCK Driver (Master Mode) 1: Enable SCK Driver (Master Mode)
- **DRV\_DLY Drive Delay.** (Refer to DRV\_EN bit) bit 4 0: Drive output immediately 1: Drive output after current byte transfer

#### **DRV\_EN Drive Enable.**

bit 3



### **Port 2 (P2)**



**P2** Port 2. This port functions as an address bus during external memory access, and as a general-purpose I/O port. bits 7−0 During external memory cycles, this port will contain the MSB of the address. Whether Port 2 is used as general-purpose I/O or for external memory access is determined by the Flash Configuration Register (HCR1.0).

## **PWM Control (PWMCON)**



**PPOL Period Polarity.** Specifies the starting level of the PWM pulse.

bit 5 0: ON Period. PWM Duty register programs the ON period.

1: OFF Period. PWM Duty register programs the OFF period.

**PWMSEL PWM Register Select.** Select which 16-bit register is accessed by PWMLOW/PWMHIGH.

- bit 4 0: Period (must be 0 for TONE mode) 1: Duty
- **SPDSEL Speed Select.** bit 3 0: 1MHz (the USEC Clock) 1: SYSCLK
- **TPCNTL Tone Generator/Pulse Width Modulation Control.**

bits 2−0





# **Tone Low (TONELOW)/PWM Low (PWMLOW)**



**TDIV7-0 Tone Divisor.** The low order bits that define the half-time period. For staircase mode the output is high impedance bits 7-0 for the last 1/4 of this period. for the last 1/4 of this period.

**PWMLOW Pulse Width Modulator Low Bits.** These 8 bits are the least significant 8 bits of the PWM register. bits 7−0

## **Tone High (TONEHI)/PWM High (PWMHI)**



**TDIV15−8 Tone Divisor.** The high order bits that define the half time period. For staircase mode the output is high impedance bits 7–0 for the last 1/4 of this period.

**PWMHI** Pulse Width Modulator High Bits. These 8 bits are the high order bits of the PWM register. bits 7−0

# **Pending Auxiliary Interrupt (PAI)**



**PAI** Pending Auxiliary Interrupt Register. The results of this register can be used as an index to vector to the bits 3−0 appropriate interrupt routine. All of these interrupts vector through address 0033h.





# **Auxiliary Interrupt Enable (AIE)**



Interrupts are enabled by EICON.4 (SFR D8H). The other interrupts are controlled by the IE and EIE registers.

# **ESEC Enable Seconds Timer Interrupt (lowest priority auxiliary interrupt).** bit  $7$  Write: Set mask bit for this interrupt  $0 =$  masked,  $1 =$  enabled. Read: Current value of **Seconds Timer Interrupt** before masking. **ESUM Enable Summation Interrupt.** bit 6 Write: Set mask bit for this interrupt  $0 =$  masked,  $1 =$  enabled. Read: Current value of **Summation Interrupt** before masking. **EADC Enable ADC Interrupt.** bit 5 Write: Set mask bit for this interrupt  $0 =$  masked,  $1 =$  enabled. Read: Current value of **ADC Interrupt** before masking. **EMSEC Enable Millisecond System Timer Interrupt.** bit 4 Write: Set mask bit for this interrupt  $0 =$  masked,  $1 =$  enabled. Read: Current value of **Millisecond System Timer Interrupt** before masking. **ESPIT Enable SPI Transmit Interrupt.** bit 3 Write: Set mask bit for this interrupt  $0 =$  masked,  $1 =$  enabled. Read: Current value of **SPI Transmit Interrupt** before masking. **ESPIR Enable SPI Receive Interrupt.** bit 2 Write: Set mask bit for this interrupt  $0 =$  masked,  $1 =$  enabled. Read: Current value of **SPI Receive Interrupt** before masking. **EALV Enable Analog Low Voltage Interrupt.** bit 1 Write: Set mask bit for this interrupt  $0 =$  masked,  $1 =$  enabled. Read: Current value of **Analog Low Voltage Interrupt** before masking. **EDLVB Enable Digital Low Voltage or Breakpoint Interrupt (highest priority auxiliary interrupt).**

bit 0 Write: Set mask bit for this interrupt  $0 =$  masked,  $1 =$  enabled. Read: Current value of **Digital Low Voltage** or **Breakpoint Interrupt** before masking.

# **Auxiliary Interrupt Status Register (AISTAT)**



# **Interrupt Enable (IE)**



# **Breakpoint Control (BPCON)**



Writing to register sets the breakpoint condition specified by MCON, BPL, and BPH.

**BP** Breakpoint Interrupt. This bit indicates that a break condition has been recognized by a hardware breakpoint register(s). bit 7 Read: Status of Breakpoint Interrupt. Will indicate a breakpoint match for any of the breakpoint registers.

Write: 0: No effect.

1: Clear Breakpoint 1 for breakpoint register selected by MCON (SFR 95h).

**PMSEL Program Memory Select.** Write this bit to select memory for address breakpoints of register selected in bit 1 MCON (SFR 95h).

0: Break on address in data memory.

1: Break on address in program memory.

**EBP Enable Breakpoint.** This bit enables this breakpoint register. Address of breakpoint register selected by bit 0 MCON (SFR 95h).

0: Breakpoint disabled.

1: Breakpoint enabled.

# **Breakpoint Low (BPL) Address for BP Register Selected in MCON (95h)**



**BPL.7−0 Breakpoint Low Address.** The low 8 bits of the 16-bit breakpoint address.

bits 7−0

### **Breakpoint High Address (BPH) Address for BP Register Selected in MCON (95h)**



**BPH.7−0 Breakpoint High Address.** The high 8 bits of the 16-bit breakpoint address**.**

bits 7−0



# **Port 0 Data Direction Low Register (P0DDRL)**



#### **P0.3 Port 0 Bit 3 Control.**

bits 7−6



## **P0.2 Port 0 Bit 2 Control.**

bits 5−4



## **P0.1 Port 0 Bit 1 Control.**

bits 3−2



# **P0.0 Port 0 Bit 0 Control.**

bits 1−0



NOTE: Port 0 also controlled by  $\overline{\mathsf{EA}}$  and Memory Access Control HCR1.1.



# **Port 0 Data Direction High Register (P0DDRH)**



#### **P0.7 Port 0 Bit 7 Control.**

bits 7−6



# **P0.6 Port 0 Bit 6 Control.**

bits 5−4



# **P0.5 Port 0 Bit 5 Control.**

bits 3−2



# **P0.4 Port 0 Bit 4 Control.**

bits 1−0



NOTE: Port 0 also controlled by  $\overline{\mathsf{EA}}$  and Memory Access Control HCR1.1.



# **Port 1 Data Direction Low Register (P1DDRL)**



#### **P1.3 Port 1 Bit 3 Control.**

bits 7−6



## **P1.2 Port 1 Bit 2 Control.**

bits 5−4



# **P1.1 Port 1 Bit 1 Control.**

bits 3−2



# **P1.0 Port 1 Bit 0 Control.**

bits 1−0





# **Port 1 Data Direction High Register (P1DDRH)**



#### **P1.7 Port 1 Bit 7 Control.**

bits 7−6



# **P1.6 Port 1 Bit 6 Control.**

bits 5−4



# **P1.5 Port 1 Bit 5 Control.**

bits 3−2



# **P1.4 Port 1 Bit 4 Control.**

bits 1−0



## **Port 3 (P3)**



**P3.7−0** General-Purpose I/O Port 3. This register functions as a general-purpose I/O port. In addition, all the pins have an bits 7−0 alternative function listed below. Each of the functions is controlled by several other SFRs. The associated Port 3 latch bit must contain a logic '1' before the pin can be used in its alternate function capacity.

- **WR External Data Memory Write Strobe.** This pin provides an active low write strobe to an external memory device. bit 6 If Port 0 or Port 2 is selected for external memory in the HCR1 register, this function will be enabled even if a '1' is not written to this latch bit. When external memory is selected, the settings of P3DRRH are ignored.
- **T1 Timer/Counter 1 External Input.** A 1 to 0 transition on this pin will increment Timer 1.
- bit 5

bit 4

bit 3

bit 2

**T0 Timer/Counter 0 External Input.** A 1 to 0 transition on this pin will increment Timer 0.

- **INT1 External Interrupt 1.** A falling edge/low level on this pin will cause an external interrupt 1 if enabled.
- **INT0 External Interrupt 0.** A falling edge/low level on this pin will cause an external interrupt 0 if enabled.
- **TXD0 Serial Port 0 Transmit.** This pin transmits the serial Port 0 data in serial port modes 1, 2, 3, and emits the bit 1 synchronizing clock in serial port mode 0.
- **RXD0 Serial Port 0 Receive.** This pin receives the serial Port 0 data in serial port modes 1, 2, 3, and is a bidirectional data bidirectional data bidirectional data transfer pin in serial port mode 0.

**RD External Data Memory Read Strobe.** This pin provides an active low read strobe to an external memory device.<br>
bit 7 **Form** of Port 2 is selected for external memory in the HCR1 register this function will be enabled If Port 0 or Port 2 is selected for external memory in the HCR1 register, this function will be enabled even if a '1' is not written to this latch bit. When external memory is selected, the settings of P3DRRH are ignored.



# **Port 2 Data Direction Low Register (P2DDRL)**



### **P2.3 Port 2 Bit 3 Control.**

bits 7−6



# **P2.2 Port 2 Bit 2 Control.**

bits 5−4



# **P2.1 Port 2 Bit 1 Control.**

bits 3−2



# **P2.0 Port 2 Bit 0 Control.**

bits 1−0



NOTE: Port 2 also controlled by  $\overline{\mathsf{EA}}$  and Memory Access Control HCR1.1.



# **Port 2 Data Direction High Register (P2DDRH)**



### **P2.7 Port 2 Bit 7 Control.**

bits 7−6



## **P2.6 Port 2 Bit 6 Control.**

bits 5−4



# **P2.5 Port 2 Bit 5 Control.**

bits 3−2



# **P2.4 Port 2 Bit 4 Control.**

bits 1−0



NOTE: Port 2 also controlled by  $\overline{\mathsf{EA}}$  and Memory Access Control HCR1.1.



# **Port 3 Data Direction Low Register (P3DDRL)**



### **P3.3 Port 3 Bit 3 Control.**

bits 7−6



# **P3.2 Port 3 Bit 2 Control.**

bits 5−4



# **P3.1 Port 3 Bit 1 Control.**

bits 3−2



# **P3.0 Port 3 Bit 0 Control.**

bits 1−0





# **Port 3 Data Direction High Register (P3DDRH)**



### **P3.7 Port 3 Bit 7 Control.**

bits 7−6



NOTE: Port 3.7 also controlled by  $\overline{EA}$  and Memory Access Control HCR1.1.

#### **P3.6 Port 3 Bit 6 Control.**

bits 5−4



NOTE: Port 3.6 also controlled by  $\overline{EA}$  and Memory Access Control HCR1.1.

# **P3.5 Port 3 Bit 5 Control.**

bits 3−2



# **P3.4 Port 3 Bit 4 Control.**

bits 1−0





# **Interrupt Priority (IP)**





#### **Serial Port 1 Control (SCON1)**



**SM0−2 Serial Port 1 Mode.** These bits control the mode of serial Port 1. Modes 1, 2, and 3 have 1 start and 1 stop bit bits 7−5 in addition to the 8 or 9 data bits.



 $\vert$  (3) RI\_0 will not be activated if bit 9 = 0.

**REN 1** Receive Enable. This bit enables/disables the serial Port 1 received shift register.

bit 4 0 = Serial Port 1 reception disabled.

1 = Serial Port 1 received enabled (modes 1, 2, and 3). Initiate synchronous reception (mode 0).

### **TB8\_1 9th Transmission Bit State.** This bit defines the state of the 9th transmission bit in serial Port 1 modes 2 and 3. bit 3

**RB8\_1 9th Received Bit State.** This bit identifies the state of the 9th reception bit of received data in serial Port 1 modes bit 2 2 and 3. In serial port mode 1, when  $SM2_1 = 0$ , RB8 1 is the state of the stop bit. RB8 1 is not used in mode 0.

**TI\_1 Transmitter Interrupt Flag.** This bit indicates that data in the serial Port 1 buffer has been completely shifted out. bit 1 In serial port mode 0, TI 1 is set at the end of the 8th data bit. In all other modes, this bit is set at the end of the last data bit. This bit must be cleared by software to transmit the next byte.

**RI\_1 Receiver Interrupt Flag.** This bit indicates that a byte of data has been received in the serial Port 1 buffer. In serial bit 0 port mode 0, RI\_1 is set at the end of the 8th bit. In serial port mode 1, RI\_1 is set after the last sample of the incoming stop bit subject to the state of SM2\_1. In modes 2 and 3, RI\_1 is set after the last sample of RB8\_1. This bit must be cleared by software to receive the next byte.

# **Serial Data Buffer 1 (SBUF1)**



**SBUF1.7−0 Serial Data Buffer 1.** Data for serial Port 1 is read from or written to this location. The serial transmit and receive bits 7−0 buffers are separate registers, but both are addressed at this location.

### **Enable Wake Up (EWU) Waking Up from IDLE Mode**



Auxiliary interrupts will wake up from IDLE. They are enabled with EAI (EICON.5, SFR D8h).

#### **EWUWDT Enable Wake Up Watchdog Timer.** Wake using watchdog timer interrupt.

bit 2  $0 = Don't$  wake up on watchdog timer interrupt.

1 = Wake up on watchdog timer interrupt.

**EWUEX1 Enable Wake Up External 1.** Wake using external interrupt source 1.

bit 1 0 = Don't wake up on external interrupt source 1.

1 = Wake up on external interrupt source 1.

#### **EWUEX0 Enable Wake Up External 0.** Wake using external interrupt source 0.

bit  $0 = 0$  = Don't wake up on external interrupt source 0.

1 = Wake up on external interrupt source 0.

# **Timer 2 Control (T2CON)**



1 = Timer 2 captures will occur when a falling edge is detected on T2EX if EXEN2 = 1.

# **Timer 2 Capture LSB (RCAP2L)**



**RCAP2L** Timer 2 Capture LSB. This register is used to capture the TL2 value when Timer 2 is configured in capture mode.<br>bits 7–0 RCAP2L is also used as the LSB of a 16-bit reload value when Timer 2 is configured in auto-r RCAP2L is also used as the LSB of a 16-bit reload value when Timer 2 is configured in auto-reload mode.
## **Timer 2 Capture MSB (RCAP2H)**



**RCAP2H Timer 2 Capture MSB.** This register is used to capture the TH2 value when Timer 2 is configured in capture mode. bits 7−0 RCAP2H is also used as the MSB of a 16-bit reload value when Timer 2 is configured in auto-reload mode.

## **Timer 2 LSB (TL2)**



**TL2** Timer 2 LSB. This register contains the least significant byte of Timer 2.

bits 7−0

## **Timer 2 MSB (TH2)**



**TH2** Timer 2 MSB. This register contains the most significant byte of Timer 2.

bits 7−0

## **Program Status Word (PSW)**



**CY Carry Flag.** This bit is set when the last arithmetic operation resulted in a carry (during addition) or a borrow (during bit 7 subtraction). Otherwise, it is cleared to 0 by all arithmetic operations.

**AC Auxiliary Carry Flag.** This bit is set to 1 if the last arithmetic operation resulted in a carry into (during addition), or bit 6 a borrow (during subtraction) from the high-order nibble. Otherwise, it is cleared to 0 by all arithmetic operations.

**F0** User Flag 0. This is a bit-addressable, general-purpose flag for software control.

bit 5

**RS1, RS0 Register Bank Select 1−0.** These bits select which register bank is addressed during register accesses.

bits 4−3



**OV Overflow Flag.** This bit is set to 1 if the last arithmetic operation resulted in a carry (addition), borrow (subtraction), bit 2 or overflow (multiply or divide). Otherwise it is cleared to 0 by all arithmetic operations.

**F1** User Flag 1. This is a bit-addressable, general-purpose flag for software control.

bit 1

**P** Parity Flag. This bit is set to 1 if the modulo-2 sum of the 8 bits of the accumulator is 1 (odd parity); and cleared to bit 0 0 on even parity.



## **ADC Offset Calibration Register Low Byte (OCL)**



**OCL ADC Offset Calibration Register Low Byte.** This is the low byte of the 24-bit word that contains the ADC offset bits 7−0 calibration. A value that is written to this location will set the ADC offset calibration value.

## **ADC Offset Calibration Register Middle Byte (OCM)**



**OCM ADC Offset Calibration Register Middle Byte.** This is the middle byte of the 24-bit word that contains the ADC offset bits 7−0 calibration. A value that is written to this location will set the ADC offset calibration value.

## **ADC Offset Calibration Register High Byte (OCH)**



**OCH ADC Offset Calibration Register High Byte.** This is the high byte of the 24-bit word that contains the ADC offset bits 7−0 calibration. A value that is written to this location will set the ADC offset calibration value.

## **ADC Gain Calibration Register Low Byte (GCL)**



**GCL** ADC Gain Calibration Register Low Byte. This is the low byte of the 24-bit word that contains the ADC gain bits 7−0 calibration. A value that is written to this location will set the ADC gain calibration value.

## **ADC Gain Calibration Register Middle Byte (GCM)**



**GCM ADC Gain Calibration Register Middle Byte.** This is the middle byte of the 24-bit word that contains the ADC gain bits 7−0 calibration. A value that is written to this location will set the ADC gain calibration value.

## **ADC Gain Calibration Register High Byte (GCH)**



**GCH ADC Gain Calibration Register High Byte.** This is the high byte of the 24-bit word that contains the ADC gain bits 7−0 calibration. A value that is written to this location will set the ADC gain calibration value.



## **ADC Multiplexer Register (ADMUX)**



## **INP3−0 Input Multiplexer Positive Channel.** This selects the positive signal input.

bits 7−4



#### **INN3−0 Input Multiplexer Negative Channel.** This selects the negative signal input.

bits 3−0





#### **Enable Interrupt Control (EICON)**



1 = Double baud rate for Port 1.

- **EAI Enable Auxiliary Interrupt.** The Auxiliary Interrupt accesses nine different interrupts which are masked and bit 5 identified by SFR registers PAI (SFR A5h), AIE (SFR A6h), and AISTAT (SFR A7h).
	- 0 = Auxiliary Interrupt disabled (default).
	- 1 = Auxiliary Interrupt enabled.
- **AI Auxiliary Interrupt Flag.** AI must be cleared by software before exiting the interrupt service routine, after the source bit 4 of the interrupt is cleared. Otherwise, the interrupt occurs again. Setting AI in software generates an Auxiliary Interrupt, if enabled.
	- 0 = No Auxiliary Interrupt detected (default).

1 = Auxiliary Interrupt detected.

**WDTI Watchdog Timer Interrupt Flag.** WDTI must be cleared by software before exiting the interrupt service routine. bit 3 Otherwise, the interrupt will occur again. Setting WDTI in software generates a watchdog time interrupt, if enabled. The Watchdog timer can generate an interrupt or reset. The interrupt is available only if the reset action is disabled in HCR0.

0 = No Watchdog Timer Interrupt detected (default).

1 = Watchdog Timer Interrupt detected.

## **ADC Results Register Low Byte (ADRESL)**



**ADRESL The ADC Results Low Byte.** This is the low byte of the 24-bit word that contains the ADC converter results. bits 7−0 Reading from this register clears the ADC interrupt. However, AI in EICON (SFR D8h) must also be cleared.

## **ADC Results Register Middle Byte (ADRESM)**



**ADRESM The ADC Results Middle Byte.** This is the middle byte of the 24-bit word that contains the ADC converter results. bits 7−0

## **ADC Results Register High Byte (ADRESH)**



**ADRESH The ADC Results High Byte.** This is the high byte of the 24-bit word that contains the ADC converter results. bits 7−0



## **ADC Control Register 0 (ADCON0)**



## **BOD** Burnout Detect. When enabled this connects a positive current source to the positive channel and a negative bit 6 current source to the negative channel. If the channel is open circuit then the ADC results will be full-scale. 0 = Burnout Current Sources Off (default).

1 = Burnout Current Sources On.

## **EVREF Enable Internal Voltage Reference.** If the internal voltage reference is not used, it should be turned off to savepower bit 5 and reduce noise.

0 = Internal Voltage Reference Off.

1 = Internal Voltage Reference On (default). **NOTE:** REFIN− must be connected to AGND, and REFOUT to REFIN+.

## **VREFH Voltage Reference High Select.** The internal voltage reference can be selected to be 2.5V or 1.25V.

bit 4  $0 =$  REFOUT is 1.25V.

 $1 = REFOUT$  is 2.5V (default).

## **EBUF Enable Buffer.** Enable the input buffer to provide higher input impedance but limits the input voltage range and bit 3 dissipates more power.

 $0 =$  Buffer disabled (default).

 $1 =$  Buffer enabled.

## **PGA2−0 Programmable Gain Amplifier.** Sets the gain for the PGA from 1 to 128.

bits 2−0





## **ADC Control Register 1 (ADCON1)**



POL Polarity. Polarity of the ADC result and Summation register.

bit 6  $0 =$  Bipolar.

1 = Unipolar. The LSB size is 1/2 the size of bipolar (twice the resolution).



### **SM1−0 Settling Mode.** Selects the type of filter or auto select which defines the digital filter settling characteristics.

bits 5−4



## **CAL2−0 Calibration Mode Control Bits.**

bits 2−0 Writing to these bits starts ADC calibration.



NOTE: Read Value—000b.

## **ADC Control Register 2 (ADCON2)**



#### **DR7−0 Decimation Ratio LSB.**

bits 7−0

## **ADC Control Register 3 (ADCON3)**



**DR10−8 Decimation Ratio Most Significant 3 Bits.** The output data rate = f<sub>CLK</sub>/[(ACLK + 1) • 64 • Decimation Ratio]. bits 2−0



## **Accumulator (A or ACC)**



**ACC.7−0 Accumulator.** This register serves as the accumulator for arithmetic and logic operations.

bits 7−0

## **Summation/Shifter Control (SSCON)**



The Summation register is powered down when the ADC is powered down. If all zeroes are written to this register the 32-bit SUMR3−0 registers will be cleared. The Summation registers will do sign extend if Bipolar is selected in ADCON1.

#### **SSCON1−0 Summation/Shift Count.**

bits 7−6



**SCNT2−0 Summation Count.** When the summation is complete an interrupt will be generated unless masked. Reading the bits 5−3 SUMR0 register clears the interrupt.



## **SHF2−0 Shift Count.**

bits 2−0





## **Summation Register 0 (SUMR0)**



**SUMR0** Summation Register 0. This is the least significant byte of the 32-bit summation register or bits 0 to 7. bits 7−0 Write: Will cause values in SUMR3−0 to be added to the summation register.

Read: Will clear the Summation Count Interrupt. AI in EICON (SFR D8h) must also be cleared.

## **Summation Register 1 (SUMR1)**



**SUMR1 Summation Register 1.** These are bits 8−15 of the 32-bit summation register.

bits 7−0

## **Summation Register 2 (SUMR2)**



#### **SUMR2 Summation Register 2.** These are bits 16−23 of the 32-bit summation register.

bits 7−0

## **Summation Register 3 (SUMR3)**



**SUMR3 Summation Register 3.** This is the most significant byte of the 32-bit summation register or bits 24−31. bits 7−0

## **Offset DAC Register (ODAC)**



**ODAC Offset DAC Register.** This register will shift the input by up to half of the ADC full-scale input range. The offset DAC bits 7−0 value is summed with the ADC input prior to conversion. Writing 00h or 80h to ODAC turns off the offset DAC. bit 7 Offset DAC Sign bit.

 $0 =$  Positive

 $1$  = Negative

bit 6–0 Offset = 
$$
\frac{-V_{REF}}{2 \cdot PGA} \cdot \left(\frac{ODAC[6:0]}{127}\right) \cdot (-1)^{bit7}
$$

**NOTE:** ODAC cannot be used to offset the input so that the buffer can be used for AGND signals. Offset DAC should be cleared before offset calibration, since the offset DAC output is applied directly to the ADC input.



## **Low Voltage Detect Control (LVDCON)**



**NOTE:** By default, both analog and digital low-voltage detections are enabled, which causes approximately 25µA of current consumption from the power supply. To minimize this power consumption, both low-voltage detections should be disabled before entering Stop mode.

## **ALVDIS Analog Low Voltage Detect Disable.**

bit 7 0 = Enable Detection of Low Analog Supply Voltage

1 = Disable Detection of Low Analog Supply Voltage

#### **ALVD2−0 Analog Voltage Detection Level.**

bits 6−4



1.2V

## **DLVDIS Digital Low Voltage Detect Disable.**

bit 3 0 = Enable Detection of Low Digital Supply Voltage 1 = Disable Detection of Low Digital Supply Voltage

## **DLVD2−0 Digital Voltage Detection Level.**

bits 2−0





#### **Extended Interrupt Enable (EIE)**



## **Hardware Product Code Register 0 (HWPC0) (read-only)**



## **HWPC1.7−0 Hardware Product Code LSB.** Read-only.

bits 7−0



## **Hardware Product Code Register 1 (HWPC1) (read-only)**



## HWPC1.7-0 Hardware Product Code MSB. Read-only. bits 7−0



## **Hardware Version Register (HDWVER)**



## **Flash Memory Control (FMCON)**



## **PGERA Page Erase.** bit  $6$  0 = MOVX to Flash will perform a byte write operation 1 = MOVX to Flash will perform a page erase operation



 $1 = Busv$ 

## **Flash Memory Timing Control Register (FTCON)**



Refer to Flash Timing Characteristics.



bits 7−4 A minimum of 10ms is needed for industrial temperature range. A minimum of 4ms is needed for commercial temperature range.

**FWR3-0** Set Write. Flash Write Time =  $(1 + FWR) \bullet (USEC + 1) \bullet 5 \bullet t_{CLK}$ .

bits 3−0 Write time should be 30−40µs.

## **B Register (B)**



**B.7−0 B Register.** This register serves as a second accumulator for certain arithmetic operations. bits 7−0



## **Power-Down Control Register (PDCON)**



Turning peripheral modules off puts the MSC1210 in the lowest power mode.



## **PSEN/ALE Select (PASEL)**



## **PSEN2−0 PSEN Mode Select.**

bits 5−3



## **ALE1−0 ALE Mode Select.**

bits 1−0



**NOTE:** For power-saving purposes, it is recommended that the PSEN and ALE pins be set to low or high mode when external memory is not used.



## **Analog Clock (ACLK)**



**FREQ6−0 Clock Frequency − 1.** This value + 1 divides the system clock to create the ADC clock.

bit 6 $-0$  f

$$
f_{\text{ACK}} = \frac{f_{\text{CLK}}}{\text{FREQ} + 1} = \frac{f_{\text{CLK}}}{\text{ACLK} + 1}
$$

$$
f_{\text{MOD}} = \frac{f_{\text{CLK}}}{(\text{ACLK} + 1) \cdot 64}
$$

Output Data Rate **Decimation** 

## **System Reset Register (SRST)**



**RSTREQ Reset Request.** Setting this bit to 1 and then clearing to 0 will generate a system reset. bit 0

## **Extended Interrupt Priority (EIP)**



**PWDI Watchdog Interrupt Priority.** This bit controls the priority of the watchdog interrupt. bit 4  $0 =$  The watchdog interrupt is low priority.  $1 =$  The watchdog interrupt is high priority. **PX5** External Interrupt 5 Priority. This bit controls the priority of external interrupt 5. bit  $3 \qquad 0 =$  External interrupt  $5$  is low priority.  $1 =$  External interrupt 5 is high priority. **PX4** External Interrupt 4 Priority. This bit controls the priority of external interrupt 4. bit 2  $0 =$  External interrupt 4 is low priority.  $1 =$  External interrupt 4 is high priority. **PX3** External Interrupt 3 Priority. This bit controls the priority of external interrupt 3. bit 1  $0 =$  External interrupt 3 is low priority. 1 = External interrupt 3 is high priority. **PX2** External Interrupt 2 Priority. This bit controls the priority of external interrupt 2. bit 0  $0 =$  External interrupt 2 is low priority. 1 = External interrupt 2 is high priority.



#### **Seconds Timer Interrupt (SECINT)**



This system clock is divided by the value of the 16-bit register MSECH:MSECL. Then, the 1ms timer tick is divided by the register HMSEC that provides the 100ms signal used by this seconds timer. Therefore, the seconds timer can generate an interrupt that occurs from 100ms to 12.8 seconds. Reading this register clears the Seconds Interrupt. This Interrupt can be monitored in the AIE register.

**WRT Write Control.** Determines whether to write the value immediately or wait until the current count is finished. bit  $7$  Read = 0.

0 = Delay Write Operation. The SEC value is loaded when the current count expires.

1 = Write Immediately. The counter is loaded once the CPU completes the write operation.

**SECINT6−0 Seconds Count.** Normal operation uses 100ms as the clock interval, and would equal: (SEC + 1)/10 seconds. bits 6–0 Seconds Interrupt =  $(1 + \text{SEC}) \cdot (\text{HMSEC} + 1) \cdot (\text{MSEC} + 1) \cdot t_{\text{CI K}}$ 

#### **Milliseconds Interrupt (MSINT)**



The clock used for this timer is the 1ms clock, which results from dividing the system clock by the values in registers MSECH:MSECL. Reading this register clears the milliseconds interrupt. AI in EICON (SFR D8h) must also be cleared.

**WRT Write Control.** Determines whether to write the value immediately or wait until the current count is finished.  $bit 7$  Read = 0.

0 = Delay Write Operation. The MSINT value is loaded when the current count expires.

1 = Write Immediately. The MSINT counter is loaded once the CPU completes the write operation.

**MSINT6−0 Seconds Count.** Normal operation would use 1ms as the clock interval.

bits 6–0 MS Interrupt Interval =  $(1 + MSINT) \cdot (MSEC + 1) \cdot t_{CJK}$ 

#### **One Microsecond Register (USEC)**



**FREQ4−0 Clock Frequency − 1.** This value + 1 divides the system clock to create a 1µs clock.

bits 4−0 USEC = CLK/(FREQ + 1). This clock is used to set Flash write time. See FTCON (SFR EFh).

#### **One Millisecond Low Register (MSECL)**



**MSECL7−0 One Millisecond Low.** This value in combination with the next register is used to create a 1ms clock.

bits 7–0 1ms = (MSECH • 256 + MSECL + 1) • t<sub>CLK</sub>. This clock is used to set Flash erase time. See FTCON (SFR EFh).



## **One Millisecond High Register (MSECH)**



**MSECH7−0 One Millisecond High.** This value in combination with the previous register is used to create a 1ms clock. bits 7–0  $1 \text{ms} = (MSECH \cdot 256 + MSECL + 1) \cdot t_{CLK}$ 

## **One Hundred Millisecond Register (HMSEC)**



**HMSEC7−0 One Hundred Millisecond.** This clock divides the 1ms clock to create a 100ms clock.

bits 7–0  $100 \text{ms} = (MSECH \cdot 256 + MSECL + 1) \cdot (HMSEC + 1) \cdot t_{CLK}$ 

## **Watchdog Timer Register (WDTCON)**



#### **EWDT Enable Watchdog (R/W).**

bit 7 Write 1/Write 0 sequence sets the Watchdog Enable Counting bit.

### **DWDT Disable Watchdog (R/W).**

bit 6 Write 1/Write 0 sequence clears the Watchdog Enable Counting bit.

## **RWDT Reset Watchdog (R/W).**

bit 5 Write 1/Write 0 sequence restarts the Watchdog Counter.

#### **WDCNT4−0 Watchdog Count (R/W).**

bits 4−0 Watchdog expires in (WDCNT + 1) • HMSEC to (WDCNT + 2) • HMSEC, if the sequence is not asserted. There is an uncertainty of 1 count.

# **Revision History**



NOTE: Page numbers for previous revisions may differ from page numbers in the current version.



## **PACKAGING INFORMATION**



**(1)** The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

<sup>(2)</sup> RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

**(3)** MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

**(4)** There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

**(5)** Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.



# **PACKAGE OPTION ADDENDUM**

<sup>(6)</sup> Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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**TEXAS** 

## **TAPE AND REEL INFORMATION**

**STRUMENTS** 



\*All dimensions are nominal



#### **QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**







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# **PACKAGE MATERIALS INFORMATION**



\*All dimensions are nominal



# **MECHANICAL DATA**

MTQF006A – JANUARY 1995 – REVISED DECEMBER 1996

**PAG (S-PQFP-G64) PLASTIC QUAD FLATPACK**



NOTES: A. All linear dimensions are in millimeters.

B. This drawing is subject to change without notice.

C. Falls within JEDEC MS-026



# **LAND PATTERN DATA**



Β. This drawing is subject to change without notice.

- С. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.
- D. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



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