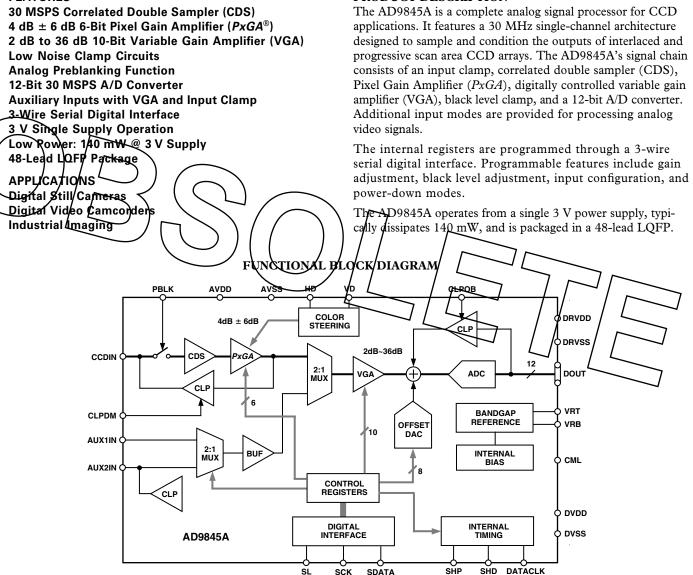


## Complete 12-Bit 30 MSPS CCD Signal Processor

## AD9845A

#### **FEATURES**



PRODUCT DESCRIPTION

PxGA is a registered trademark of Analog Devices, Inc.

#### REV.0

Information furnished by Analog Devices is believed to be accurate and reliable. However, no responsibility is assumed by Analog Devices for its use, nor for any infringements of patents or other rights of third parties which may result from its use. No license is granted by implication or otherwise under any patent or patent rights of Analog Devices.

# AD9845A-SPECIFICATIONS

**GENERAL SPECIFICATIONS** ( $T_{MIN}$  to  $T_{MAX}$ , AVDD = DVDD = 3.0 V,  $f_{DATACLK}$  = 30 MHz, unless otherwise noted.)

Parameter	Min	а Тур	Ν	Max	Unit
TEMPERATURE RANGE					
Operating	-20		4	+85	°C
Storage	-65		+	+150	°C
POWER SUPPLY VOLTAGE					
Analog, Digital, Digital Driver	2.7		3	3.6	V
POWER CONSUMPTION					
Normal Operation	(Spe	ecified Under Ea	hch Mode of C	Operation)	
Power-Down Modes	(0)			operation)	
Fast Recovery Mode		90			mW
Standby		5			mW
Total Power-Down		1			mW
		1			
MAXMUM CLOCK/RATE	30				MHz
A/D/CONVERTER / / / )					
Resolution / / / / /	12				Bits
Differential Nonlinearity (DNL)	$)   \frown$	±0.5	<u>+</u>	±1.0	LSB
No Missing Codes					Bits Guaranteed
Full-Scale Input Voltage		$\left  \right\rangle \left  2.0 \right\rangle$			V
Data Output Coding	> ) /   /	Straight E	Binary /	<b>_</b>	
	/////	++			
VOLTAGE REFERENCE					TT T
Reference Top Voltage (VRT)	$\setminus$				
Reference Bottom Voltage (VRB)			$ - + \square$	$\searrow$	
Specifications subject to change without notice.			$\neg IL$	-	
				<u> </u>	
DIGITAL SPECIFICATIONS (DRVDD	) — 2 7 V C. — 20 nF	unless otherwise	noted )	$ \longrightarrow $	
		1			
Parameter	Symbol	Min	Тур	Max	Unit
LOGIC INPUTS					
High Level Input Voltage	V <sub>IH</sub>	2.1			V
Low Level Input Voltage	V <sub>IL</sub>			0.6	V
High Level Input Current	III		10		μA
Low Level Input Current	IIL		10		μΑ
Input Capacitance	C <sub>IN</sub>		10		pF
LOGIC OUTPUTS					
	1				1
	Vou	2.2			V
High Level Output Voltage, $I_{OH} = 2 \text{ mA}$ Low Level Output Voltage, $I_{OL} = 2 \text{ mA}$	V <sub>OH</sub> V <sub>OL</sub>	2.2		0.5	

Specifications subject to change without notice.

### **CCD-MODE SPECIFICATIONS** (T<sub>MIN</sub> to T<sub>MAX</sub>, AVDD = DVDD = 3.0 V, $f_{DATACLK} = f_{SHD} = 30$ MHz, unless otherwise noted.)

Parameter	Min	Тур	Max	Unit	Notes
POWER CONSUMPTION		140		mW	See TPC 1 for Power Curves
MAXIMUM CLOCK RATE	30			MHz	
CDS Gain Allowable CCD Reset Transient <sup>1</sup> Max Input Range Before Saturation <sup>1</sup> Max CCD Black Pixel Amplitude <sup>1</sup>	1.0	0 500 200		dB mV V p-p mV	See Input Waveform in Footnote 1 <i>PxGA</i> Gain at 4 dB
PIXEL GAIN AMPLIFIER ( <i>PxGA</i> ) Max Input Range Max Output Range Gain Control Resolution Gain Monotonicity Gain Range (Two's Complement Coding) Min Gain ( <i>PxGA</i> Gain Code 32) Max Gain ( <i>PxGA</i> Gain Code 31)	1.0 1.6	64 Guarant -2 10	eed	V p-p V p-p Steps dB dB	See Figure 28 for <i>PxGA</i> Gain Curve
VARIABLE GAIN AMPLIFER (VGA) Max Input Range Max Output Range Gain Control Resolution Gain Monotonicity Gain Range Low Gain (VGA Gain Code 91) Max Gain (VGA Gain Code 1023)	1.6	1024 Guaran 2 36	ted	V p-p V p-p Steps dB dB	See Figure 29 for VGA Gain Curve
BLACK LEVEL CLAMP Clamp Level Resolution Clamp Level Min Clamp Level Max Clamp Level		256 0 255		Steps LSB LSB	Measured at AIPC Output
SYSTEM PERFORMANCE Gain Accuracy (VGA Code 91 to 1023) <sup>2</sup>	-0.5		+0.5		Specifications Include Entire Signal Chain Use Equations on Page 18 to Calculate Gai
<ul> <li>PxGA Gain Accuracy Min Gain (PxGA Register Code 32) Max Gain (PxGA Code 31)</li> <li>Peak Nonlinearity, 500 mV Input Signal Total Output Noise</li> <li>Power Supply Rejection (PSR)</li> </ul>	-1 11	0 12 0.1 0.6 40	+1 13	dB dB % LSB rms dB	VGA Gain Fixed at 2 dB (Code 91) VGA Gain Fixed at 2 dB (Code 91) 12 dB Gain Applied AC Grounded Input, 6 dB Gain Applied Measured with Step Change on Supply
POWER-UP RECOVERY TIME Fast Recovery Mode Reference Standby Mode Total Shutdown Mode Power-Off Condition		0.1 1 3 15		ms ms ms ms	Normal Clock Signals Applied

NOTES

<sup>1</sup>Input Signal Characteristics defined as follows:

soomv TYP RESET TRANSIENT OPTICAL BLACK PIXEL ' ~t 4 dB (Code 63). "ithout notice.

## AD9845A-SPECIFICATIONS

### **AUX1-MODE SPECIFICATIONS** ( $T_{MIN}$ to $T_{MAX}$ , AVDD = DVDD = 3.0 V, $f_{DATACLK}$ = 30 MHz, unless otherwise noted.)

Parameter	Min	Тур	Max	Unit
POWER CONSUMPTION		100		mW
MAXIMUM CLOCK RATE	30			MHz
INPUT BUFFER				
Gain		0		dB
Max Input Range	1.0			V p-p
VGA				
Max Output Range	2.0			V p-p
Gain Control Resolution		1023		Steps
Gain (Selected Using VGA Gain Register)				
<u>Min Gain</u>		0		dB
Max Gain		36		dB
AIN2-MANE SPECIFICATIONS 🗔 🤅	. התעת – תתעה ד	_ 2 0 V f 20 N	INT unloss otherwise	notod )
AUX2-MODE SPECIFICATIONS TIME T	- I I I I I I I I I I I I I I I I I I I			
Parameter	T <sub>MAX</sub> , AVDD = DVDD =		IHz, unless otherwise Max	Unit
Parameter Power Consumption	Min			Unit mW
Parameter			Max	Unit
Parameter POWER CONSUMPTION MAXIMUM CLOCK RATE INPUT BUFFER		<b>Typ</b>	Max	Unit mW
Parameter POWER CONSUMPTION MAXIMUM CLOCK RATE INPUT BUFFER VGA	30 Sar	<b>Typ</b>	Max	Unit mW
Parameter POWER CONSUMPTION MAXIMUM CLOCK RATE INPUT BUFFER VGA Max Output Range		Typ 100 me as AUX1-MODE	Max	Unit mW MHz
Parameter POWER CONSUMPTION MAXIMUM CLOCK RATE INPUT BUFFER VGA Max Output Range Gain Control Resolution	30 Sar	<b>Typ</b>	Max	Unit mW
Parameter POWER CONSUMPTION MAXIMUM CLOCK RATE INPUT BUFFER VGA Max Output Range	30 Sar	Typ 100 me as AUX1-MODE	Max	Unit mW MHz
Parameter POWER CONSUMPTION MAXIMUM CLOCK RATE INPUT BUFFER VGA Max Output Range Gain Control Resolution Gain (Selected Using VGA Gain Register)	30 Sar	Typ 100 me as AUX1-MODE 512	Max	Unit mW MHz y p-p- Steps
Parameter POWER CONSUMPTION MAXIMUM CLOCK RATE INPUT BUFFER VGA Max Output Range Gain Control Resolution Gain (Selected Using VGA Gain Register) Min Gain	30 Sar	Typ 100 me as AUX1-MODE 512 0	Max	Unit mW MHz y p-p- Steps dB
Parameter POWER CONSUMPTION MAXIMUM CLOCK RATE INPUT BUFFER VGA Max Output Range Gain Control Resolution Gain (Selected Using VGA Gain Register) Min Gain Max Gain	30 Sar	Typ 100 me as AUX1-MODE 512 0	Max	Unit mW MHz y p-p- Steps dB
Parameter POWER CONSUMPTION MAXIMUM CLOCK RATE INPUT BUFFER VGA Max Output Range Gain Control Resolution Gain (Selected Using VGA Gain Register) Min Gain Max Gain ACTIVE CLAMP	30 Sar	<b>Typ</b> 100 me as AUX1-MODE 512 0 18	Max	Unit mW MHz y p-p Steps dB dB
Parameter POWER CONSUMPTION MAXIMUM CLOCK RATE INPUT BUFFER VGA Max Output Range Gain Control Resolution Gain (Selected Using VGA Gain Register) Min Gain Max Gain ACTIVE CLAMP Clamp Level Resolution	30 Sar	<b>Typ</b> 100 me as AUX1-MODE 512 0 18	Max	Unit mW MHz y p-p Steps dB dB

Specifications subject to change without notice.

## **TIMING SPECIFICATIONS** $(C_L = 20 \text{ pF}, f_{SAMP} = 30 \text{ MHz}, CCD-Mode Timing in Figures 10 and 11, AUX-Mode Timing in Figure 7. Serial Timing in Figures 26–29.)$

Parameter	Symbol	Min	Тур	Max	Unit
SAMPLE CLOCKS					
DATACLK, SHP, SHD Clock Period	t <sub>CONV</sub>	32	33		ns
DATACLK Hi/Low Pulsewidth	t <sub>ADC</sub>	13	16.7		ns
SHP Pulsewidth	t <sub>SHP</sub>	5	8.3		ns
SHD Pulsewidth	t <sub>SHD</sub>	5	8.3		ns
CLPDM Pulsewidth	t <sub>CDM</sub>	4	10		Pixels
CLPOB Pulsewidth <sup>1</sup>	t <sub>COB</sub>	2	20		Pixels
SHP Rising Edge to SHD Falling Edge	t <sub>S1</sub>	0	8.3		ns
SHP Rising Edge to SHD Rising Edge	t <sub>S2</sub>	13	16.7		ns
Internal Clock Delay	t <sub>ID</sub>		3.0		ns
Inhibited Clock Period	t <sub>INH</sub>	10			ns
ATAQUTPUT					
Output Delay	t <sub>OD</sub>		14.5	16	ns
Output Hold Time	t <sub>H</sub>	7.0	7.6		ns
Pipeline Delay			9		Cycle
SERIAL INTERFACE					
Maximum \$CK Frequency		7 10			MHz
SL to SCK Setup Time		/ 10	$\sim$		ns
SCK to SL Hold Time		/ 10			ns
SDATA Valid to SCK Rising Edge Setup	$\sqrt{f_{DS}}//$	/ 10 /			ns
SCK Falling Edge to SDATA Valid Hold		10 /			
SCK Falling Edge to SDATA Valid Read		10 /	$\sim$ 1		ns
NOTES			$t \sim 1$		+ -
Minimum CLPOB pulsewidth is for functional operation only. W	vider typical pulses are re-	commended to ac	chieve low noise c	lamp performance.	
Specifications subject to change without notice.	JI I				$\sim 1$

#### ABSOLUTE MAXIMUM RATINGS

	With Respect			
Parameter	То	Min	Max	Unit
AVDD1, AVDD2	AVSS	-0.3	+3.9	V
DVDD1, DVDD2	DVSS	-0.3	+3.9	V
DRVDD	DRVSS	-0.3	+3.9	V
Digital Outputs	DRVSS	-0.3	DRVDD + 0.3	V
SHP, SHD, DATACLK	DVSS	-0.3	DVDD + 0.3	V
CLPOB, CLPDM, PBLK	DVSS	-0.3	DVDD + 0.3	V
SCK, SL, SDATA	DVSS	-0.3	DVDD + 0.3	V
VRT, VRB, CMLEVEL	AVSS	-0.3	AVDD + 0.3	V
BYP1-4, CCDIN	AVSS	-0.3	AVDD + 0.3	V
Junction Temperature			150	°C
Lead Temperature			300	°C
(10 sec)				

#### **ORDERING GUIDE**

Model	Temperature	Package	Package	
	Range	Description	Option	
AD9845AJST	–20°C to +85°C	Thin Plastic Quad Flatpack (LQFP)	ST-48	

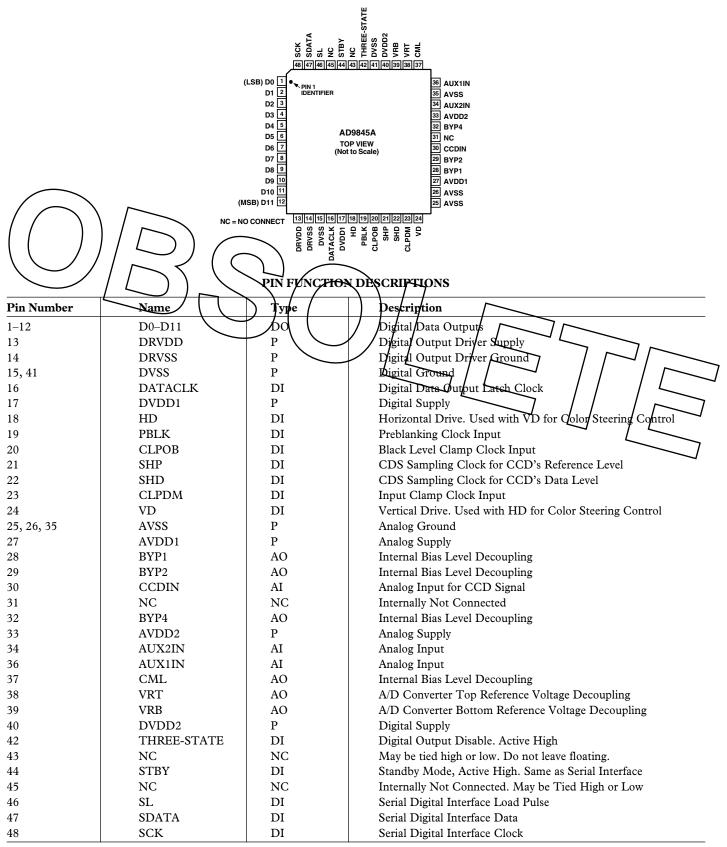
#### THERMAL CHARACTERISTICS Thermal Resistance 48-Lead LQFP Package $\theta_{IA} = 92^{\circ}C$

CAUTION \_

ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000 V readily accumulate on the human body and test equipment and can discharge without detection. Although the AD9845A features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high-energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.



#### PIN CONFIGURATIONS



TYPE: AI = Analog Input, AO = Analog Output, DI = Digital Input, DO = Digital Output, P = Power.

#### DEFINITIONS OF SPECIFICATIONS DIFFERENTIAL NONLINEARITY (DNL)

An ideal ADC exhibits code transitions that are exactly 1 LSB apart. DNL is the deviation from this ideal value. Thus every code must have a finite width. No missing codes guaranteed to 12-bit resolution indicates that all 4096 codes, respectively, must be present over all operating conditions.

#### PEAK NONLINEARITY

Peak nonlinearity, a full signal chain specification, refers to the peak deviation of the output of the AD984x from a true straight line. The point used as "zero scale" occurs 1/2 LSB before the first code transition. "Positive full scale" is defined as a Level 1, 1/2 LSB beyond the last code transition. The deviation is measured from the middle of each particular output code to the true straight line. The error is then expressed as a percentage of the 2 V ADC full-scale signal. The input signal is always appropriately gained up to fill the ADC's full-scale range.

TOTAL OUTPUT NOISE

The rms output noise is measured using histogram rechniques. The standard deviation of the ADC output codes is calculated

### EQUIVALENT INPUT CIRCUITS

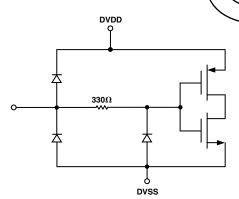


Figure 1. Digital Inputs—SHP, SHD, DATACLK, CLPOB, CLPDM, HD, VD, PBLK, SCK, SL

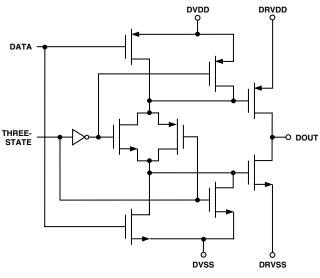


Figure 2. Data Outputs – D0–D11

in LSB, and represents the rms noise level of the total signal chain at the specified gain setting. The output noise can be converted to an equivalent voltage, using the relationship 1 LSB = (ADC Full Scale/2<sup>N</sup> codes) when N is the bit resolution of the ADC. For the AD9845A, 1 LSB is 500  $\mu$ V.

#### POWER SUPPLY REJECTION (PSR)

The PSR is measured with a step change applied to the supply pins. This represents a very high frequency disturbance on the AD984x's power supply. The PSR specification is calculated from the change in the data outputs for a given step change in the supply voltage.

#### **INTERNAL DELAY FOR SHP/SHD**

ACVE

木

ACVSS

The internal delay (also called aperture delay) is the time delay that occurs from when a sampling edge is applied to the AD984x until the actual sample of the input signal is held. Both SHP and SHD sample the input signal during the transition from low to high, so the internal delay is measured from each clock's rising edge to the instant the actual internal sample is taken.

Figure 3. CCDIN (Pin 30)

ACVSS

60Ω

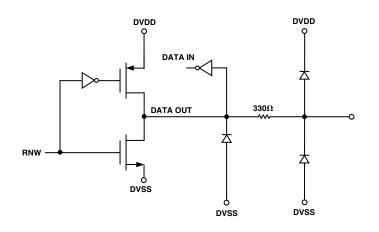
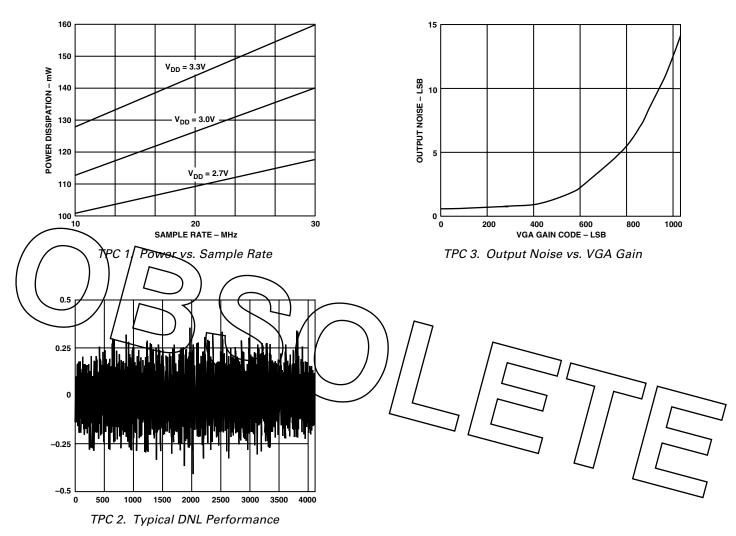
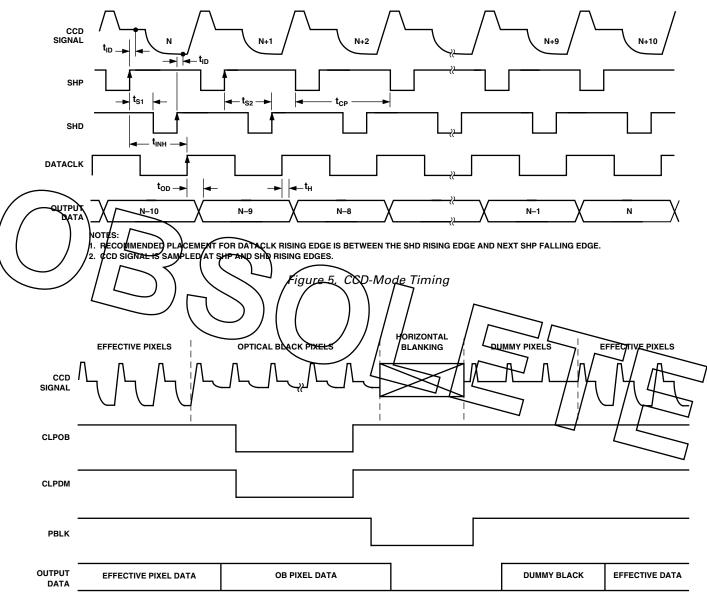


Figure 4. SDATA (Pin 47)

### AD9845A–Typical Performance Characteristics

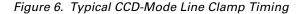


#### CCD-MODE AND AUX MODE TIMING



NOTES:

1. CLPOB AND CLPDM WILL OVERWRITE PBLK. PBLK WILL NOT AFFECT CLAMP OPERATION IF OVERLAPPING CLPDM AND/OR CLPOB. 2. PBLK SIGNAL IS OPTIONAL. 3. DIGITAL OUTPUT DATA WILL BE ALL ZEROS DURING PBLK. OUTPUT DATA LATENCY IS 9 DATACLK CYCLES.



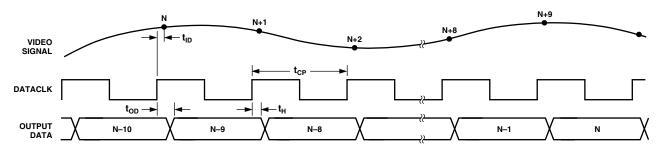


Figure 7. AUX-Mode Timing

#### PIXEL GAIN AMPLIFIER (PxGA) TIMING

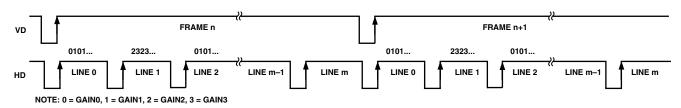


Figure 8. PxGA Mode 1 (Mosaic Separate) Frame/Line Gain Register Sequence

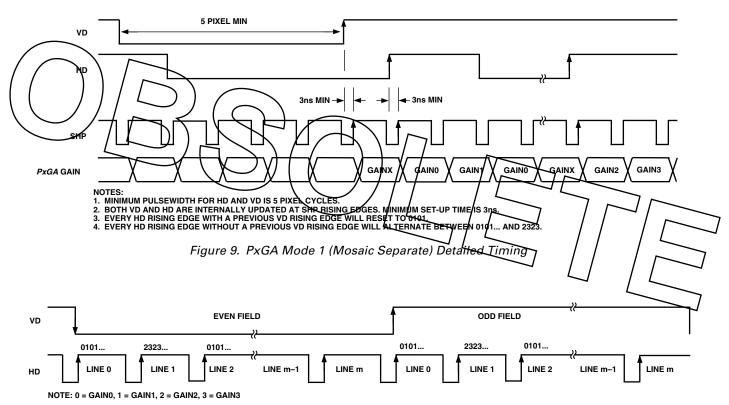


Figure 10. PxGA Mode 2 (Interlace) Frame/Line Gain Register Sequence

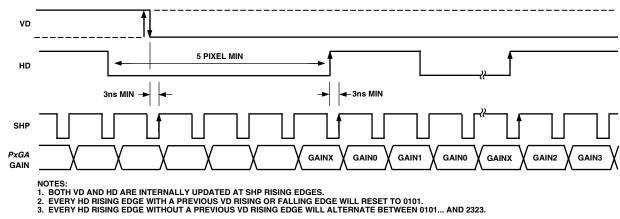
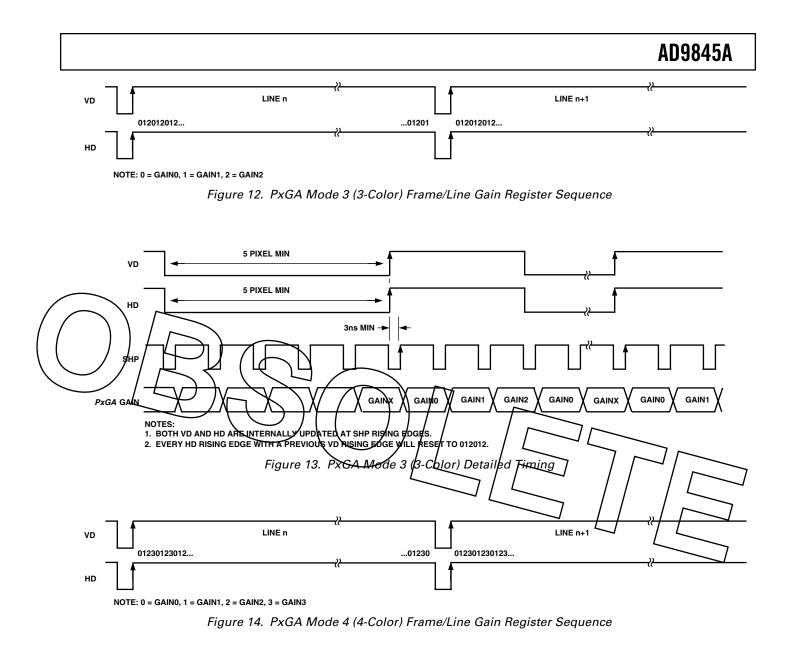


Figure 11. PxGA Mode 2 (Interlace) Detailed Timing



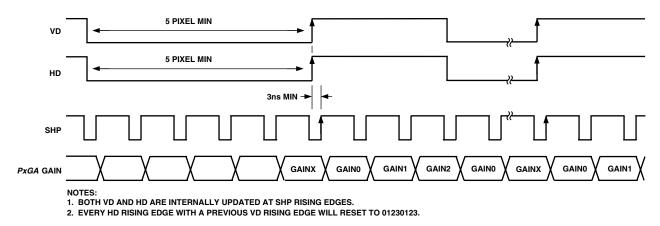


Figure 15. PxGA Mode 4 (4-Color) Detailed Timing

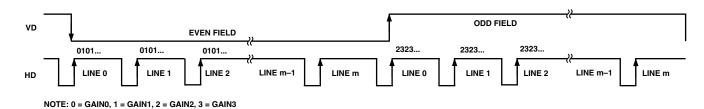


Figure 16. PxGA Mode 5 (VD Selected) Frame/Line Gain Register Sequence

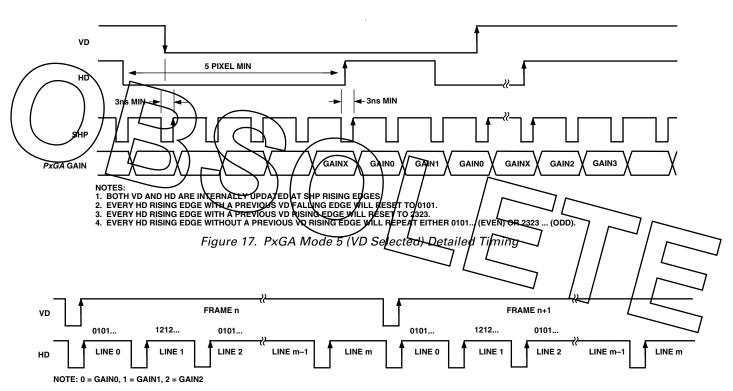
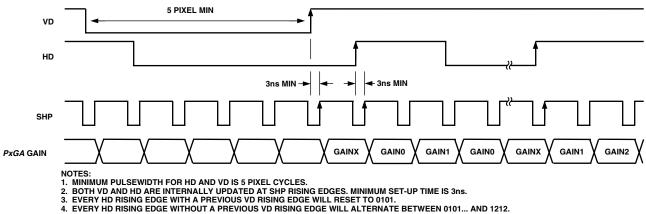
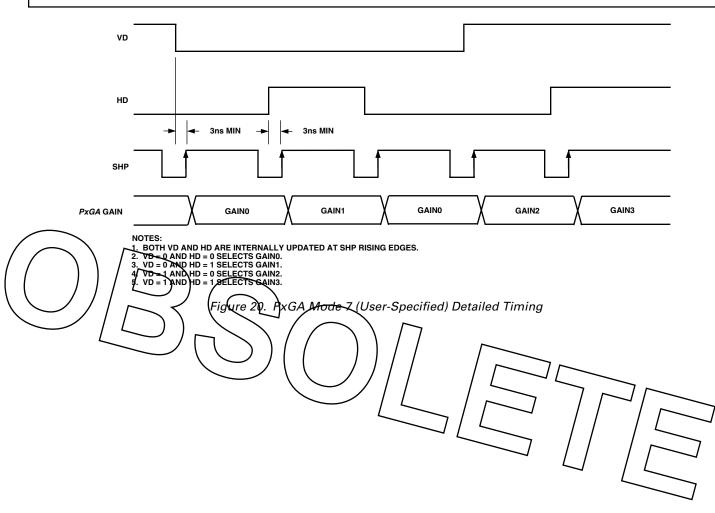


Figure 18. PxGA Mode 6 (Mosaic Repeat) Frame/Line Gain Register Sequence

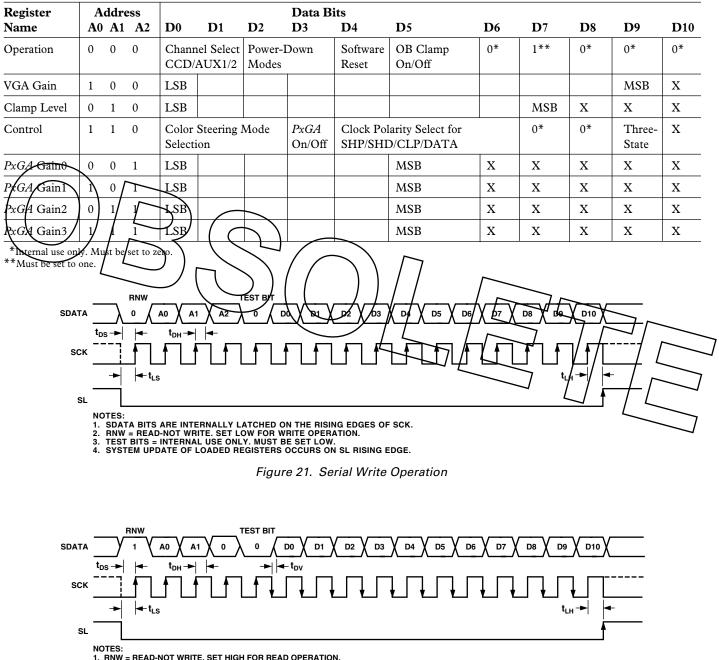


3. 4.

Figure 19. PxGA Mode 6 (Mosaic Repeat) Detailed Timing



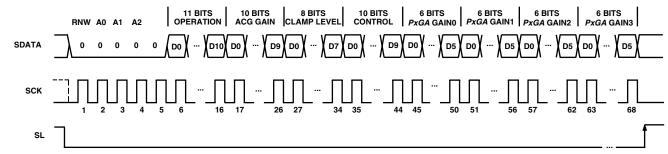
### SERIAL INTERFACE TIMING AND INTERNAL REGISTER DESCRIPTION



#### Table I. Internal Register Map

INGLES: 1. RNW = READ-NOT WRITE. SET HIGH FOR READ OPERATION. 2. TEST BITS = INTERNAL USE ONLY. MUST BE SET LOW. 3. SERIAL DATA FROM THE SELECTED REGISTER IS VALID STARTING AFTER THE 5TH SCK FALLING EDGE, AND IS UPDATED ON SCK FALLING EDGES.

Figure 22. Serial Readback Operation



NOTES: 1. ANY NUMBER OF ADJACENT REGISTERS MAY BE LOADED SEQUENTIALLY, BEGINNING WITH THE LOWEST ADDRESS AND INCREMENTING ONE ADDRESS AT A TIME. 2. WHEN SEQUENTIALLY LOADING MULTIPLE REGISTERS, THE EXACT REGISTER LENGTH (SHOWN ABOVE) MUST BE USED FOR EACH REGISTER. 3. ALL LOADED REGISTERS WILL BE SIMULTANEOUSLY UPDATED WITH THE RISING EDGE OF SL. Figure 23. Continuous Serial Write Operation to All Registers PxGA GAIN3 PxGA GAIN2 PxGA GAIN0 Т PxGA GAIN1 Т 1 RŃW D0 D1 D2 D3 D4 D5 D0 D5 0 D2 10 D4 D5 D0 n SCK 1 2 3 4 13 14 15 17 29 2 23 SL Figure 24. Continuous Serial Write Operation to Alt PxGA Gain Regist ers



D10	D9	D8	<b>D</b> 7	D6	Optical Black Clamp D5	Reset D4	Power-Down Modes D3 D2	Channel Selection D1 D0
0*	0*	0*	1**	0*	<ol> <li>Enable Clamping</li> <li>Disable Clamping</li> </ol>	<ul><li>0 Normal</li><li>1 Reset All Registers to Default</li></ul>	<ol> <li>0 Normal Power</li> <li>0 1 Fast Recovery</li> <li>1 0 Standby</li> <li>1 Total Power-Down</li> </ol>	0 0 CCD Mode 0 1 AUX1 Mode 1 0 AUX2 Mode 1 1 Test Only

\*Must be set to zero.

\*\*Set to one.

#### Table III. VGA Gain Register Contents (Default Value x096)

D10	MSB D9	<b>D</b> 8	<b>D</b> 7	D6	D5	<b>D</b> 4	D3	D2	D1	LSB D0	Gain (dB)
X	0	0	0	1	0	1	1	1	1	1	2.0
					•						•
					•						•
					•						•
	1	1	1	1	1	1	1	1	1	0	35.965
	1	1	1	1	1	1	1	1	1	1	36.0

<b>D</b> 10	D9	<b>D</b> 8	MSB D7	D6	D5	<b>D</b> 4	D3	D2	D1	LSB D0	Clamp Level (LSB)
x	Х	Х	0	0	0	0	0	0	0	0	0
			0	0	0	0	0	0	0	1	1
			0	0	0	0	0	0	1	0	2
						•					•
						•					•
						•					•
			1	1	1	1	1	1	1	0	254
			1	1	1	1	1	1	1	1	255

Table IV. Clamp Level Register Contents (Default Value x080)

	I able V. Control Register Contents (Default Value x000)									
	ata Qut	D8_D7	DATACLK D6	C D	LP/PBLK 95	SHP/SHD D4	<i>PxGA</i> D3**	Color Steer D2 D1 D0	-	
X 0	Enable Three-State	0*_0*	0 Hising Edge Ti 1 Falling Edge T			0 Active Low 1 Active High	0 Disable 1 Enable	$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	Steering Disabled Mosaic Separate Interlace 3-Color 4-Color VD Selected Mosaic Repeat User Splecified	
	De set to zero. D3 = 0 ( $PxGA$ d	,	PxGA gain is fixed to 4		or Gain0, Gai	n1, Gain2, Gain	3 (Default )	/alue x000) /		
<b>D</b> 10	D9	D8	D7 D6	MSI D5	3 D4	D3 D	02 D1	LSB D0	Gain (dB)*	
X	Х	Х	X X	0	1	1 1	1	1	+10.0	
				01	0 1	0 0 1 1	0 1	0 1	+4.3 +4.0	
				1	0	• • 0 0	0	0	• • -2.0	

Table V. Control Register Contents (Default Value x000)

\*Control Register Bit D3 must be set High (PxGA Enable) to use the PxGA Gain Registers.

#### CIRCUIT DESCRIPTION AND OPERATION

The AD9845A signal processing chain is shown in Figure 25. Each processing step is essential in achieving a high-quality image from the raw CCD pixel data.

#### **DC Restore**

To reduce the large dc offset of the CCD output signal, a dcrestore circuit is used with an external 0.1  $\mu$ F series coupling capacitor. This restores the dc level of the CCD signal to approximately 1.5 V, to be compatible with the 3 V single supply of the AD9845A.

#### **Correlated Double Sampler**

The CDS circuit samples each CCD pixel twice to extract the video information and reject low-frequency noise. The timing shown in Figure 5 illustrates how the two CDS clocks, SHP and SHD, are used to sample the reference level and data level of the CCD signal respectively. The CCD signal is sampled on the rising edges of SHP and SHD. Placement of these two clock signals is critical in achieving the best performance from the CCD. An internal SHP/SHD delay  $(t_{\rm ID})$  of 3 ns is caused by internal propagation delays.

#### Input Clamp

A line-rate input clamping arcuit is used to remove the CCD's optical black offset. This offset exists in the CCD's shielded black reference pixels. Unlike some AFE architectures, the AD9845A removes this offset in the input stage to minimize the effect of a

gain change on the system black level, usually called the "gain step." Another advantage of removing this offset at the input stage is to maximize system headroom. Some area CCDs have large black level offset voltages, which, if not corrected at the input stage, can significantly reduce the available headroom in the internal circuitry when higher VGA gain settings are used.

Horizontal timing is shown in Figure 6. It is recommended that the CLPDM pulse be used during valid CCD dark pixels. CLPDM may be used during the optical black pixels, either together with CLPOB or separately. The CLPDM pulse should be a minimum of 4 pixels wide.

#### **PxGA**

The *PxGA* provides separate gain adjustment for the individual color pixels. A programmable gain amplifier with four separate values, the *PxGA* has the capability to "multiplex" its gain value on a pixel-to-pixel basis. This allows lower output color pixels to be gained up to match higher output color pixels. Also, the *PxGA* may be used to adjust the colors for white balance, reducing the amount of digital processing that is needed. The four different gain values are switched according to the "Color Steering" circuitry. Seven different color steering modes for different types of CCD color filter arrays are programmed in the AD984x's Control Register. For example, Mosaic Separate steering mode accommodates the popular "Barer" arrangement of Red, Green, and Blue filters (see Figure 26).

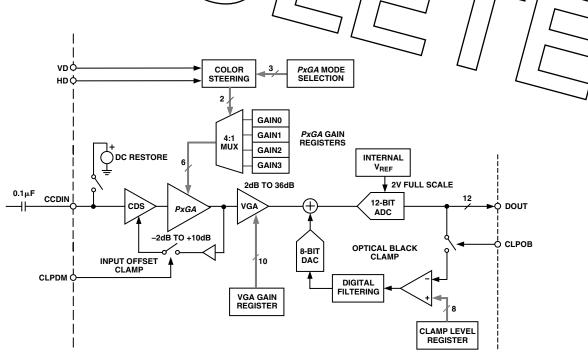


Figure 25. CCD-Mode Block Diagram

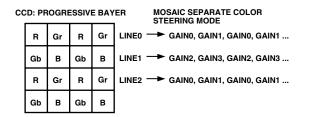


Figure 26. CCD Color Filter Example: Progressive Scan

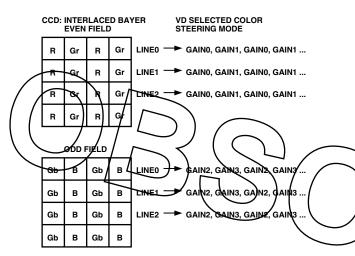


Figure 27. CCD Color Filter Example: Interlaced

The same Bayer pattern can also be interlaced, and the VD Selected mode should be used with this type of CCD (see Figure 27). The Color Steering performs the proper multiplexing of the R, G, and B gain values (loaded into the *PxGA* gain registers), and is synchronized by the user with vertical (VD) and horizontal (HD) sync pulses. For more detailed information, see the *PxGA* Timing section. The *PxGA* gain for each of the four channels is variable from -2 dB to +10 dB, controlled in 64 steps through the serial interface. The *PxGA* gain curve is shown in Figure 28.

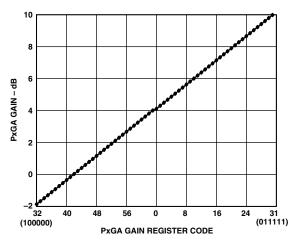


Figure 28. PxGA Gain Curve

#### Variable Gain Amplifier

The VGA stage provides a gain range of 2 dB to 36 dB, programmable with 10-bit resolution through the serial digital interface. Combined with 4 dB from the PxGA stage, the total gain range for the AD9845A is 6 dB to 40 dB. The minimum gain of 6 dB is needed to match a 1 V input signal with the ADC full-scale range of 2 V. When compared to 1 V full-scale systems (such as ADI's AD9803), the equivalent gain range is 0 dB to 34 dB.

The VGA gain curve is divided into two separate regions. When the VGA Gain Register code is between 0 and 511, the curve follows a (1 + x)/(1 - x) shape, which is similar to a "linear-indB" characteristic. From code 512 to code 1023, the curve follows a "linear-in-dB" shape. The exact VGA gain can be calculated for any Gain Register value by using the following two equations:

Code Range	Gain Equation (dB)
------------	--------------------

 $\begin{array}{ll} 0-511 & Gain = 20 \log_{10} \left( [658 + code] / [658 - code] \right) - 0.4 \\ 512 - 1023 & Gain = (0.0354)(code) - 0.4 \end{array}$ 

As shown in the CCD Mode Specifications, only the VGA gain range from 2 dB to 36 dB has tested and guaranteed accuracy. This corresponds to a VGA gain code range of 91 to 1023. The Gain Accuracy Specifications also include the PxGA gain of 4 dB, for a total gain range of 6 dB to 40 dB.

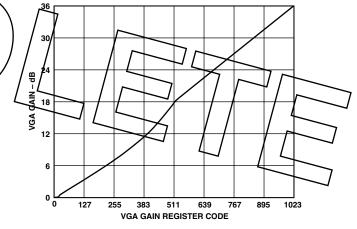


Figure 29. VGA Gain Curve (Gain from PxGA Not Included)

#### **Optical Black Clamp**

The optical black clamp loop is used to remove residual offsets in the signal chain, and to track low-frequency variations in the CCD's black level. During the optical black (shielded) pixel interval on each line, the ADC output is compared with a fixed black level reference, selected by the user in the Clamp Level Register. Any value between 0 LSB and 255 LSB may be programmed, with 8-bit resolution. The resulting error signal is filtered to reduce noise, and the correction value is applied to the ADC input through a D/A converter. Normally, the optical black clamp loop is turned on once per horizontal line, but this loop can be updated more slowly to suit a particular application. If external digital clamping is used during the post processing, the AD9845A optical black clamping may be disabled using Bit D5 in the Operation Register (see Serial Interface Timing and Internal Register Description section). When the loop is disabled, the Clamp Level Register may still be used to provide programmable offset adjustment.

Horizontal timing is shown in Figure 6. The CLPOB pulse should be placed during the CCD's optical black pixels. It is recommended that the CLPOB pulse duration be at least 20 pixels wide to minimize clamp noise. Shorter pulsewidths may be used, but clamp noise may increase, and the ability to track low-frequency variations in the black level will be reduced.

#### A/D Converter

The AD9845A uses high-performance ADC architecture, optimized for high speed and low power. Differential Nonlinearity (DNL) performance is typically better than 0.5 LSB, as shown in TPC 2. Instead of the 1 V full-scale range used by the earlier AD9801 and AD9803 products from Analog Devices, the AD9845A's ADC uses a 2 V input range. Better noise performance results from using a larger ADC full-scale range (see TPC 3).

#### AUX1 Mode

For applications that do not require CDS, the AD9845A can be configured to sample ac-coupled waveforms. Figure 30 shows the circuit configuration for using the AUX1 channel input (Pin 36). A single 0.1  $\mu$ F ac-coupling capacitor is needed between the input signal driver and the AUX1IN pin. An on-chip dc-bias circuit sets the average value of the input signal to approximately 0.4 V, which is referenced to the midscale code of the ADC. The VGA Gain register provides a gain range of 0 dB to 36 dB in this mode of operation (see VGA Gain Curve, Figure 29). The VGA gains up the signal level with respect to the 0.4 V bias level. Signal levels above the bias level will be further increased to a higher ADC code, while signal levels below the bias level will be further decreased to a lower ADC code.

#### AUX2 Mode

For sampling video-type waveforms, such as NTSC and PAL signals, the AUX2 channel provides black level clamping, gain adjustment, and A/D conversion. Figure 31 shows the circuit configuration for using the AUX2 channel input (Pin 34). A external 0.1  $\mu$ F blocking capacitor is used with the on-chip video clamp circuit, to level-shift the input signal to a desired reference level. The clamp circuit automatically senses the most negative portion of the input signal, and adjusts the voltage across the input capacitor. This forces the black level of the input signal to be equal to the value programmed into the Clamp Level register (see Serial Interface Register Description). The VGA provides gain adjustment from 0 dB to 18 dB. The same VGA Gain register is used, but only the 9 MSBs of the gain register are used (see Table VII.)

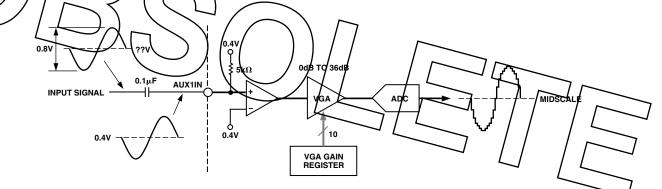


Figure 30. AUX1 Circuit Configuration

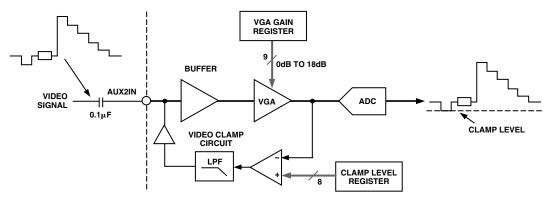


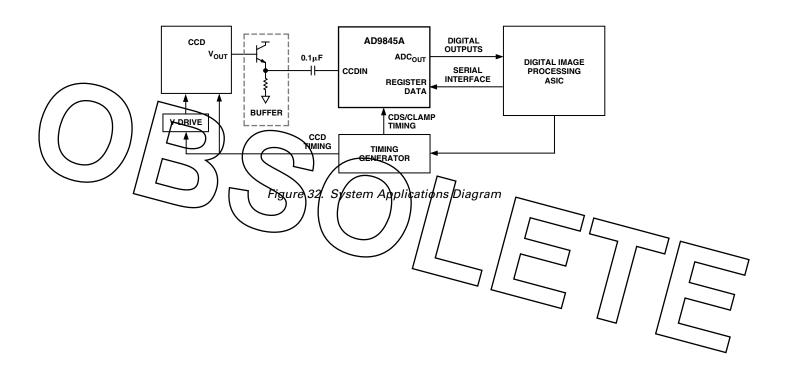
Figure 31. AUX2 Circuit Configuration

Table VII.	VGA Gain	Register	Used for	r AUX2-Mode
1 4010 1 111	v on oam	Register	0304 10	I HOME MOUL

D10	MSB D9	<b>D</b> 8	<b>D</b> 7	D6	D5	<b>D</b> 4	D3	D2	<b>D</b> 1	LSB D0	Gain (dB)
X	0	X	X	X	X	X	X	X	X	X	0.0
	1	0	0	0	0	0	0	0	0	0	0.0
					•						•
					•						•
					•						•
	1	1	1	1	1	1	1	1	1	1	18.0

#### APPLICATIONS INFORMATION

The AD9845A is a complete Analog Front End (AFE) product for digital still camera and camcorder applications. As shown in Figure 32, the CCD image (pixel) data is buffered and sent to the AD9845A analog input through a series input capacitor. The AD9845A performs the dc restoration, CDS, gain adjustment, black level correction, and analog-to-digital conversion. The AD9845A's digital output data is then processed by the image processing ASIC. The internal registers of the AD984x—used to control gain, offset level, and other functions—are programmed by the ASIC or microprocessor through a 3-wire serial digital interface. A system timing generator provides the clock signals for both the CCD and the AFE.



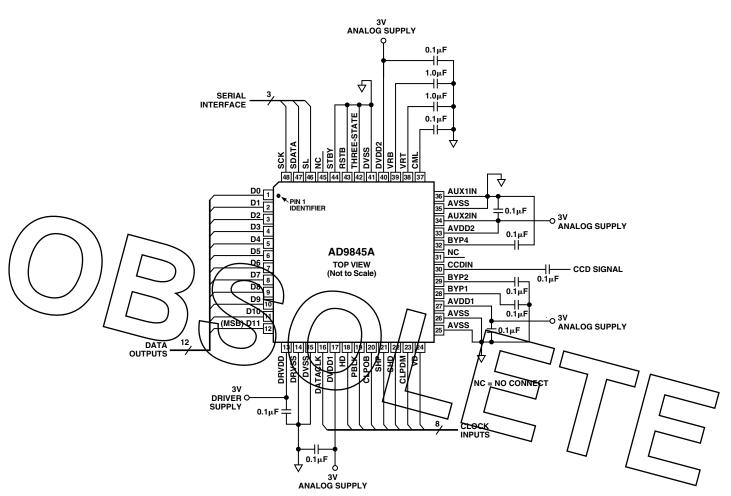


Figure 33. Recommended Circuit Configuration for CCD-Mode

#### Internal Power-On Reset Circuitry

After power-on, the AD9845A will automatically reset all internal registers and perform internal calibration procedures. This takes approximately 1 ms to complete. During this time, normal clock signals and serial write operations may occur. However, serial register writes will be ignored until the internal reset operation is completed. Pin 43 (formerly RSTB on the AD984x non-A products) is no longer used for the reset operation. Toggling Pin 43 in the AD9845A will have no effect.

#### Grounding and Decoupling Recommendations

As shown in Figure 33, a single ground plane is recommended for the AD9845A. This ground plane should be as continuous as possible, particularly around Pins 25 through 39. This will ensure that all analog decoupling capacitors provide the lowest possible impedance path between the power and bypass pins and their respective ground pins. All decoupling capacitors should be located as close as possible to the package pins. A single clean power supply is recommended for the AD9845A, but a separate digital driver supply may be used for DRVDD (Pin 13). DRVDD should always be decoupled to DRVSS (Pin 14), which should be connected to the analog ground plane. Advantages of using a separate digital driver supply include using a lower voltage (2.7 V) to match levels with a 2.7 V ASIC, reducing digital power dissipation, and reducing potential noise coupling. If the digital outputs (Pins 3–12) must drive a load larger than 20 pF, buffering is recommended to reduce digital code transition noise. Alternatively, placing series resistors close to the digital output pins may also help reduce noise.

#### **OUTLINE DIMENSIONS**

Dimensions shown in inches and (mm).

