

General Description

The DS1874 controls and monitors all functions for SFF, SFP, and SFP+ modules including all SFF-8472 functionality. The combination of the DS1874 with the MAX3798/MAX3799 laser driver/limiting amplifier provides APC loop, modulation current control, and eye safety functionality. The DS1874 continuously monitors for high output current, high bias current, and low and high transmit power to ensure that laser shutdown for eye safety requirements are met without adding external components. Six ADC channels monitor V_{CC}, temperature, and four external monitor inputs (MON1–MON4) that can be used to meet all monitoring requirements. MON3 is differential with support for common mode to V_{CC}. Two digital-to-analog (DAC) outputs with temperature-indexed lookup tables (LUTs) are available for additional monitoring and control functionality.

Applications

SFF, SFP, and SFP+ Transceiver Modules

Pin Configuration

Features

- **Meets All SFF-8472 Control and Monitoring Requirements**
- ♦ **Laser Bias Controlled by APC Loop and Temperature LUT to Compensate for Tracking Error**
- ♦ **Laser Modulation Controlled by Temperature LUT**
- ♦ **Six Analog Monitor Channels: Temperature, VCC, MON1–MON4 MON1–MON4 Support Internal and External Calibration**

Scalable Dynamic Range Internal Direct-to-Digital Temperature Sensor Alarm and Warning Flags for All Monitored Channels

- ♦ **Two 9-Bit Delta-Sigma Outputs with 36 Entry Temperature LUTs**
- ♦ **Digital I/O Pins: Five Inputs, Five Outputs**
- ♦ **Comprehensive Fault-Measurement System with Maskable Laser Shutdown Capability**
- ♦ **Flexible, Two-Level Password Scheme Provides Three Levels of Security**
- ♦ **256 Additional Bytes Located at A0h Slave Address**
- ♦ **I 2C-Compatible Interface**
- ♦ **3-Wire Master to Communicate with the MAX3798/ MAX3799 Laser Driver/Limiting Amplifier**
- ♦ **+2.85V to +3.9V Operating Voltage Range**
- ♦ **-40°C to +95°C Operating Temperature Range**
- ♦ **28-Pin TQFN (5mm x 5mm) Package**

Ordering Information

+Denotes a lead(Pb)-free/RoHS-compliant package. T&R = Tape and reel.

*EP = Exposed pad.

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DS1874

LIST OF TABLES

ABSOLUTE MAXIMUM RATINGS

*Subject to not exceeding +6V.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

RECOMMENDED OPERATING CONDITIONS

 $(T_A = -40^{\circ}C$ to $+95^{\circ}C$, unless otherwise noted.)

DC ELECTRICAL CHARACTERISTICS

(V_{CC} = +2.85V to +3.9V, T_A = -40°C to +95°C, unless otherwise noted.)

DS1874 **DS1874**

DAC1, DAC2 ELECTRICAL CHARACTERISTICS

(V_{CC} = $+2.85V$ to $+3.9V$, T_A = -40° C to $+95^{\circ}$ C, unless otherwise noted.)

ANALOG QUICK-TRIP CHARACTERISTICS

(V_{CC} = $+2.85V$ to $+3.9V$, T_A = -40°C to $+95^{\circ}$ C, unless otherwise noted.)

ANALOG VOLTAGE MONITORING CHARACTERISTICS

(V_{CC} = $+2.85V$ to $+3.9V$, T_A = -40°C to $+95°C$, unless otherwise noted.)

DIGITAL THERMOMETER CHARACTERISTICS

(V_{CC} = +2.85V to +3.9V, T_A = -40°C to +95°C, unless otherwise noted.)

AC ELECTRICAL CHARACTERISTICS

(V_{CC} = $+2.85V$ to $+3.9V$, T_A = -40°C to $+95^{\circ}$ C, unless otherwise noted.)

TIMING CHARACTERISTICS (CONTROL LOOP AND QUICK TRIP)

(V_{CC} = $+2.85V$ to $+3.9V$, T_A = -40°C to $+95°C$, unless otherwise noted.)

3-WIRE DIGITAL INTERFACE SPECIFICATION

(V_{CC} = +2.85V to +3.9V, T_A = -40°C to +95°C, timing referenced to V_{IL(MAX)} and V_{IH(MIN)}, unless otherwise noted. See Figure 15.)

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I2C AC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = +2.85V$ to +3.9V, $T_A = -40^{\circ}$ C to +95°C, timing referenced to V_{IL(MAX)} and V_{IH(MIN)}, unless otherwise noted. See Figure 17.)

NONVOLATILE MEMORY CHARACTERISTICS

 $(V_{\text{CC}} = +2.85V$ to $+3.9V$, unless otherwise noted.)

Note 1: All voltages are referenced to ground. Current into the IC is positive, and current out of the IC is negative.

Note 2: Inputs are at supply rail. Outputs are not loaded.

Note 3: This parameter is guaranteed by design.

Note 4: Full-scale is user programmable.

Note 5: The DACs are the bias and modulation DACs found in the MAX3798/MAX3799 that are controlled by the DS1874.

- **Note 6:** The DS1874 is configured with TXDOUT connected to the MAX3798/MAX3799 DISABLE input.
- **Note 7:** This includes writing to the modulation DAC and the initial step written to the bias DAC.

Note 8: A temperature conversion is completed and the modulation register value is recalled from the LUT and V_{CC} has been measured to be above V_{CC} LO alarm.

Note 9: The timing is determined by the choice of the update rate setting (see Table 02h, Register 88h).

Note 10: This specification is the time it takes from MON3 voltage falling below the LLOS trip threshold to LOSOUT asserted high.

Note 11: This specification is the time it takes from MON3 voltage rising above the HLOS trip threshold to LOSOUT asserted low. **Note 12:** Assuming an appropriate initial step is programmed that would cause the power to exceed the APC set point within four steps, the bias current will be within 3% within the time specified by the binary search time. See the BIAS and MODULA-

TION Control During Power-Up section.

Note 13: I²C interface timing shown is for fast mode (400kHz). This device is also backward compatible with I²C standard mode timing.

Note 14: C_B—the total capacitance of one bus line in pF.

Note 15: EEPROM write begins after a STOP condition occurs.

Typical Operating Characteristics

(V_{CC} = +2.85V to +3.9V, T_A = +25°C, unless otherwise noted.)

Pin Description

Block Diagram

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Typical Operating Circuit

Detailed Description

The DS1874 integrates the control and monitoring functionality required to implement a VCSEL-based SFP or SFP+ system using Maxim's MAX3798/MAX3799 combined limiting amplifier and laser driver. Key components of the DS1874 are shown in the Block Diagram and described in subsequent sections.

MAX3798/MAX3799 DAC Control

The DS1874 controls two 9-bit DACs inside the MAX3798/MAX3799. One DAC is used for laser bias control while the other is used for laser modulation control. The DS1874 communicates with the MAX3798/ MAX3799 over a 3-wire digital interface (see the 3-Wire Master for Controlling the MAX3798/MAX3799 section). The communication between the DS1874 and MAX3798/MAX3799 is transparent to the end user.

BIAS Register/APC Control

The MAX3798/MAX3799 control their laser bias current DAC using the APC loop within the DS1874. The APC loop's feedback to the DS1874 is the monitor diode (MON2) current, which is converted to a voltage using

Table 1. Acronyms

an external resistor. The feedback is sampled by a comparator and compared to a digital set-point value. The output of the comparator has three states: up, down, or no-operation. The no-operation state prevents the output from excessive toggling once steady state is reached. As long as the comparator output is in either the up or down states, the bias is adjusted by writing increment and decrement values to the MAX3798/MAX3799 through the BIASINC register (3-wire address 13h).

The DS1874 has an LUT to allow the APC set point to change as a function of temperature to compensate for tracking error (TE). The TE LUT has 36 entries that determine the APC setting in 4°C windows between -40 $^{\circ}$ C to +100 $^{\circ}$ C.

MODULATION Control

The MAX3798/MAX3799 control the laser modulation using the internal temperature-indexed LUT within the DS1874. The modulation LUT is programmed in 2°C increments over the -40°C to +102°C range to provide temperature compensation for the laser's modulation. The modulation is updated after each temperature conversion using the 3-wire interface that connects to the MAX3798/MAX3799. The MAX3798/MAX3799 include a 9-bit DAC. The modulation LUT is 8 bits.

Figure 1 demonstrates how the 8-bit LUT controls the 9-bit DAC with the use of a temperature control bit (MODTC, Table 02h, Register C6h) and a temperature index register (MODTI, Table 02h, Register C2h).

Figure 1. Modulation LUT Loading to MAX3798/MAX3799 MOD DAC

BIAS and MODULATION Control During Power-Up

The DS1874 has two internal registers, MODULATION and BIAS, that represent the values written to the MAX3798/MAX3799's modulation DAC and bias DAC through the 3-wire interface. On power-up, the DS1874 sets the MODULATION and BIAS registers to 0. When V_{CC} is above POA, the DS1874 initializes the MAX3798/ MAX3799. After a temperature conversion is completed and if the VCC LO alarm is enabled, an additional V_{CC} conversion above the customer-defined VCC LO alarm level is required before the MAX3798/MAX3799 MODU-LATION register is updated with the value determined by the temperature conversion and the modulation LUT.

When the MODULATION register is set, the BIAS register is set to a value equal to ISTEP (see Figure 2). The startup algorithm checks if this bias current causes a feedback voltage above the APC set point, and if not, it continues increasing the BIAS register by ISTEP until the APC set point is exceeded. When the APC set point is exceeded, the device begins a binary search to quickly reach the bias current corresponding to the proper power level. After the binary search is completed, the APC integrator is enabled and single LSB steps are used to tightly control the average power.

The TXP HI, TXP LO, HBAL, and BIAS MAX QT alarms are masked until the binary search is completed. However, the BIAS MAX alarm is monitored during this time to prevent the BIAS register from exceeding IBIASMAX. During the bias current initialization, the BIAS register is not allowed to exceed IBIASMAX. If this occurs during the ISTEP sequence, then the binary search routine is enabled. If IBIASMAX is exceeded during the binary search, the next smaller step is activated. ISTEP or binary increments that would cause the BIAS register to exceed IBIASMAX are not taken. Masking the alarms until the completion of the binary search prevents false positive alarms during startup.

ISTEP is programmed by the customer using Table 02h, Register BBh. During the first steps, the MAX3798/ MAX3799's bias DAC is directly written using SET_IBIAS (3-wire address 09h). ISTEP should be programmed to the maximum safe increase that is allowable during startup. If this value is programmed too low, the DS1874 still operates, but it could take significantly longer for the algorithm to converge and hence to control the average power.

If a fault is detected, and TXD is toggled to reenable the outputs, the DS1874 powers up following a similar sequence to an initial power-up. The only difference is that the DS1874 already has determined the present temperature, so the t_{INIT} time is not required for the DS1874 to recall the APC and MOD set points from EEPROM.

Figure 2. Power-Up Timing

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SFP+ Controller with Digital LDD Interface

BIAS and MODULATION Registers as a Function of Transmit Disable (TXD)

If TXD is asserted (logic 1) during normal operation, the outputs are disabled within tOFF. When TXD is deasserted (logic 0), the DS1874 sets the MODULATION register with the value associated with the present temperature, and initializes the BIAS register using the same search algorithm as done at startup. When asserted, soft TXD (TXDC) (Lower Memory, Register 6Eh) would allow a software control identical to the TXD pin (see Figure 3).

APC and Quick-Trip Timing

As shown in Figure 4, the DS1874's input comparator is shared between the APC control loop and the quicktrip alarms (TXP HI, TXP LO, LOS, and BIAS HI). The comparator polls the alarms in a multiplexed sequence. Five of every eight comparator readings are used for APC loop bias-current control. The other three updates are used to check the HTXP/LTXP (monitor diode voltage), the HBATH (MON1), and LOS (MON3) signals against the internal APC, BIAS, and MON3 reference, respectively. If the last APC comparison was higher than the APC set point, it makes an HTXP comparison, and if it is lower, it makes an LTXP comparison. Depending on the results of the comparison, the corresponding alarms and warnings (TXP HI, TXP LO) are asserted or deasserted.

The DS1874 has a programmable comparator sample time based on an internally generated clock to facilitate a wide variety of external filtering options and time delays resulting from writing values to the MAX3798/ MAX3799's bias DAC. The UPDATE RATE register (Table 02h, Register 88h) determines the sampling time. Samples occur at a regular interval, trep. Table 2 shows the sample rate options available. Any quick-trip alarm that is detected by default remains active until a subsequent comparator sample shows the condition no longer exists. A second bias current monitor (BIAS MAX) compares the MAX3798/MAX3799's BIAS DAC's code to a digital value stored in the IBIASMAX register. This comparison is made at every bias current update to ensure that a high-bias current is quickly detected.

An APC sample that requires an update of the BIAS register causes subsequent APC samples to be

Table 2. Update Rate Timing

Figure 3. TXD Timing

Figure 4. APC Loop and Quick-Trip Sample Timing

ignored until the end of the 3-wire communication that updates the MAX3798/MAX3799's BIAS DAC, plus an additional 16 sample periods (tRFP).

Monitors and Fault Detection

Monitors

Monitoring functions on the DS1874 include five quick-trip comparators and six ADC channels. This monitoring combined with the alarm enables (Table 01h/05h) determines when/if the DS1874 turns off the MAX3798/ MAX3799 DACs and triggers the TXF and TXDOUT outputs. All the monitoring levels and interrupt masks are user programmable.

Five Quick-Trip Monitors and Alarms

Five quick-trip monitors are provided to detect potential laser safety issues and LOS status. These monitor the following:

- 1) High Bias Current (HBATH)
- 2) Low Transmit Power (LTXP)
- 3) High Transmit Power (HTXP)
- 4) Max Output Current (IBIASMAX)
- 5) Loss-of-Signal (LOS LO)

The high-transmit and low-transmit power quick-trip registers (HTXP and LTXP) set the thresholds used to compare against the MON2 voltage to determine if the transmit power is within specification. The HBATH quick trip compares the MON1 input (generally from the MAX3798/MAX3799 bias monitor output) against its threshold setting to determine if the present bias current is above specification. The BIAS MAX quick trip determines if the BIAS register is above specification. The BIAS register is not allowed to exceed the value set in the IBIASMAX register. When the DS1874 detects that the bias is at the limit it sets the BIAS MAX status bit and holds the BIAS register setting at the IBIASMAX level. The bias and power quick trips are routed to the TXF through interrupt masks to allow combinations of these alarms to be used to trigger these outputs. The user can program up to eight different temperatureindexed threshold levels for MON1 (Table 02h, Registers D0h–D7h). The LOS LO quick trip compares the MON3 input against its threshold setting to determine if the present received power is below the specification. The LOS LO quick trip can be used to set the LOSOUT pin. These alarms can be latched using Table 02h, Register 8Ah.

Six ADC Monitors and Alarms

The ADC monitors six channels that measure temperature (internal temp sensor), V_{CC}, and MON1–MON4 using an analog multiplexer to measure them round

robin with a single ADC (see the ADC Timing section). The five voltage channels have a customer-programmable full-scale range and all channels have a customerprogrammable offset value that is factory programmed to default value (see Table 3). Additionally, MON1–MON4 can right-shift results by up to 7 bits before the results are compared to alarm thresholds or read over the I2C bus. This allows customers with specified ADC ranges to calibrate the ADC full scale to a factor of $1/2ⁿ$ of their specified range to measure small signals. The DS1874 can then right-shift the results by n bits to maintain the bit weight of their specification (see the Right-Shifting ADC Result and Enhanced RSSI Monitoring (Dual-Range Functionality) sections).

Table 3. ADC Default Monitor Full-Scale Ranges

The ADC results (after right-shifting, if used) are compared to the alarm and warning thresholds after each conversion, and the corresponding alarms are set, which can be used to trigger the TXF output. These ADC thresholds are user programmable, as are the masking registers that can be used to prevent the alarms from triggering the TXF output.

ADC Timing

There are six analog channels that are digitized in a round-robin fashion in the order shown in Figure 5. The total time required to convert all six channels is t_{RR} (see the Analog Voltage Monitoring Characteristics for details).

Right-Shifting ADC Result

If the weighting of the ADC digital reading must conform to a predetermined full-scale (PFS) value defined by a standard's specification (e.g., SFF-8472), then right-shifting can be used to adjust the PFS analog measurement range while maintaining the weighting of the ADC results. The DS1874's range is wide enough to cover all requirements; when the maximum input value is ≤ 1/2 of the FS value, right-shifting can be used to obtain greater accuracy. For instance, the maximum voltage might be 1/8 the specified PFS value, so only 1/8 the converter's range is effective over this range. An alternative is to calibrate the ADC's full-scale range to 1/8 the readable PFS value and use a right-shift value of 3. With this implementation, the resolution of

Figure 5. ADC Round-Robin Timing

Figure 6. MON3 Differential Input for High-Side RSSI

the measurement is increased by a factor of 8, and because the result is digitally divided by 8 by rightshifting, the bit weight of the measurement still meets the standard's specification (i.e., SFF-8472).

The right-shift operation on the ADC result is carried out based on the contents of right-shift control registers (Table 02h, Registers 8Eh–8Fh) in EEPROM. Four analog channels, MON1–MON4, each have 3 bits allocated to set the number of right-shifts. Up to seven right-shift operations are allowed and are executed as a part of every conversion before the results are compared to the high-alarm and low-alarm levels, or loaded into their corresponding measurement registers (Lower Memory, Registers 64h–6Bh). This is true during the setup of internal calibration as well as during subsequent data conversions.

Enhanced RSSI Monitoring (Dual-Range Functionality)

The DS1874 offers a feature to improve the accuracy and range of MON3, which is most commonly used for monitoring RSSI. The accuracy of the RSSI measurements is increased at the small cost of reduced range (of input signal swing). The DS1874 eliminates this trade-off by offering "dual range" calibration on the MON3 channel (see Figure 6). This feature enables right-shifting (along with its gain and offset settings) when the input signal is below a set threshold (within the

range that benefits using right-shifting) and then automatically disables right-shifting (recalling different gain and offset settings) when the input signal exceeds the threshold. Also, to prevent "chattering," hysteresis prevents excessive switching between modes in addition to ensuring that continuity is maintained. Dual-range operation is enabled by default (factory programmed in EEPROM). However, it can easily be disabled through the RSSI_FC and RSSI_FF bits, which are described in the Register Descriptions section. When dual-range operation is disabled, MON3 operates identically to the other MON channels, although featuring a differential input.

Dual-range functionality consists of two modes of operation: fine mode and coarse mode. Each mode is calibrated for a unique transfer function, hence the term, dual range. Table 5 highlights the registers related to MON3. Fine mode is equivalent to the other MON channels. Fine mode is calibrated using the gain, offset, and right-shifting registers at locations shown in Table 5 and is ideal for relatively small analog input voltages. Coarse mode is automatically switched to when the input exceeds a threshold (to be discussed in a subsequent paragraph). Coarse mode is calibrated using different gain and offset registers, but lacks right-shifting (since coarse mode is only used on large input signals). The gain and offset registers for coarse mode are also shown in Table 5. Additional information for each of the registers can be found in the Register Descriptions section.

Dual-range operation is transparent to the end user. The results of MON3 analog-to-digital conversions are still stored/reported in the same memory locations (68h–69h, Lower Memory) regardless of whether the conversion was performed in fine mode or coarse mode. The only way to tell which mode generated the digital result is by reading the RSSIR bit.

When the DS1874 is powered up, analog-to-digital conversions begin in a round-robin fashion. Every MON3 timeslice begins with a fine mode analog-to-digital conversion (using fine mode's gain, offset, and right-shifting settings). See the flowchart in Figure 7 for more details.

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Figure 7. RSSI Flowchart

Then, depending on whether the last MON3 timeslice resulted in a coarse-mode conversion and also depending on the value of the current fine conversion, decisions are made whether to use the current fine-mode conversion result or to make an additional conversion (within the same MON3 timeslice), using coarse mode (using coarse mode's gain and offset settings and no rightshifting) and reporting the coarse-mode result. The flowchart in Figure 7 also illustrates how hysteresis is

Table 4. MON3 Hysteresis Threshold Values

*This is the minimum reported coarse-mode conversion.

Table 5. MON3 Configuration Registers

implemented. The fine-mode conversion is compared to one of two thresholds. The actual threshold values are a function of the number of right-shifts being used. With the use of right-shifting, the fine mode full-scale is programmed to (1/2nth) of the coarse mode full-scale. The DS1874 now auto ranges to choose the range that gives the best resolution for the measurement. Hysteresis is applied to eliminate chatter when the input resides at the boundary of the two ranges. See Figure 7 for details. Table 4 shows the threshold values for each possible number of right-shifts.

The RSSI FF and RSSI FC bits are used to force finemode or coarse-mode conversions, or to disable the dual-range functionality. Dual-range functionality is enabled by default (both RSSI_FC and RSSI_FF are factory programmed to 0 in EEPROM). It can be disabled by setting RSSI_FC to 0 and RSSI_FF to 1. These bits are also useful when calibrating MON3. For additional information, see Figure 19.

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Low-Voltage Operation

The DS1874 contains two power-on reset (POR) levels. The lower level is a digital POR (POD) and the higher level is an analog POR (POA). At startup, before the supply voltage rises above POA, the outputs are disabled, all SRAM locations are set to their defaults, shadowed EEPROM (SEE) locations are zero, and all analog circuitry is disabled. When V_{CC} reaches POA, the SEE is recalled, and the analog circuitry is enabled. While V_{CC} remains above POA, the device is in its normal operating state, and it responds based on its nonvolatile configuration. If during operation V_{CC} falls below POA, but is still above POD, then the SRAM retains the SEE settings from the first SEE recall, but the device analog is shut down and the outputs disabled. If the supply voltage recovers back above POA, then the device immediately resumes normal operation. If the supply voltage falls below POD, then the device SRAM is placed in its default state and another SEE recall is required to reload the nonvolatile settings. The EEPROM recall occurs the next time V_{CC} exceeds POA. Figure 8 shows the sequence of events as the voltage varies.

Any time V_{CC} is above POD, the I²C interface can be used to determine if V_{CC} is below the POA level. This is accomplished by checking the RDYB bit in the STATUS (Lower Memory, Register 6Eh) byte. RDYB is set when V_{CC} is below POA; when V_{CC} rises above POA, RDYB

is timed (within 500µs) to go to 0, at which point the part is fully functional.

For all device addresses sourced from EEPROM (Table 02h, Register 8Ch), the default device address is A2h until V_{CC} exceeds POA, allowing the device address to be recalled from the EEPROM.

Power-On Analog (POA)

POA holds the DS1874 in reset until V_{CC} is at a suitable level (V_{CC} > POA) for the device to accurately measure with its ADC and compare analog signals with its quicktrip monitors. Because V_{CC} cannot be measured by the ADC when V_{CC} is less than POA, POA also asserts the VCC LO alarm, which is cleared by a V_{CC} ADC conversion greater than the customer-programmable VCC alarm low ADC limit. This allows a programmable limit to ensure that the headroom requirements of the transceiver are satisfied during a slow power-up. The TXF output does not latch until there is a conversion above V_{CC} low limit. The POA alarm is nonmaskable. The TXF output is asserted when V_{CC} is below POA. See the Low-Voltage Operation section for more information.

Delta-Sigma Outputs (DAC1 and DAC2)

Two delta-sigma outputs are provided, DAC1 and DAC2. With the addition of an external RC filter, these outputs provide two 9-bit resolution analog outputs with the full-scale range set by the input REFIN. Each output

Figure 8. Low-Voltage Hysteresis Example

Figure 9. Recommended RC Filter for DAC1/DAC2

Figure 10. Delta-Sigma Outputs

is either manually controlled or controlled using a temperature-indexed LUT. A delta-sigma is a digital output using pulse-density modulation. It provides much lower output ripple than a standard digital PWM output given the same clock rate and filter components. Before t_{INIT}, the DAC1 and DAC2 outputs are high impedance.

The external RC filter components are chosen based on ripple requirements, output load, delta-sigma frequency, and desired response time. A recommended filter is shown in Figure 9.

The DS1874's delta-sigma outputs are 9 bits. For illustrative purposes, a 3-bit example is provided. Each possible output of this 3-bit delta-sigma DAC is given in Figure 10.

In LUT mode, DAC1 and DAC2 are each controlled by a separate 8-bit, 4°C-resolution, temperature-addressed LUT. The delta-sigma outputs use a 9-bit structure. The 8-bit LUTs are either loaded directly into the MSBs (8:1) or the LSBs (7:0). This is determined by DAC1TI (Table 02h, Register C3h), DAC2TI (Table 02h, Register C4h), DAC1TC (Table 02h, Register C6h, bit 6), and DAC2TC (Table 02h, Register C6h, bit 5). See Figure 11 for more details. The DAC1 LUT (Table 07h) and DAC2 LUT (Table 08h) are nonvolatile and password-2 protected.

The reference input, REFIN, is the supply voltage for the output buffer of DAC1 and DAC2. The voltage connected to REFIN must be able to support the edge rate requirements of the delta-sigma outputs. In a typical application, a 0.1µF capacitor should be connected between REFIN and ground.

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Figure 11. DAC1/DAC2 LUT Assignments

Digital I/O Pins

Five digital input and five digital output pins are provided for monitoring and control.

LOS, LOSOUT

By default (LOSC = 1, Table 02h, Register 89h), the LOS pin is used to convert a standard comparator output for loss of signal (LOS) to an open-collector output. This means the mux shown in the Block Diagram by default selects the LOS pin as the source for the LOSOUT output transistor. The output of the mux can be read in the STATUS byte (Lower Memory, Register 6Eh) as the RXL bit. The RXL signal can be inverted (INV $LOS = 1$) before driving the open-drain output transistor using the XOR gate provided. Setting $LOSC = 0$ configures the mux to be controlled by LOS LO, which is driven by the output of the LOS quick trip (Table 02h, Registers BEh and BFh). The mux setting (stored in EEPROM) does not take effect until V_{CC} > POA, allowing the EEPROM to recall.

IN1, RSEL, OUT1, RSELOUT

The digital input IN1 and RSEL pins primarily serve to meet the rate-select requirements of SFP and SFP+. They also serve as general-purpose inputs. OUT1 and

RSELOUT are driven by a combination of the IN1, RSEL, and logic dictated by control registers in the EEPROM (Figure 13). The levels of IN1 and RSEL can be read using the STATUS register (Lower Memory, Register 6Eh). The open-drain output OUT1 can be controlled and/or inverted using the CNFGB register (Table 02h, Register 8Ah). The open-drain RSELOUT output is software-controlled and/or inverted through the Status register and CNFGA register (Table 02h, Register 89h). External pullup resistors must be provided on OUT1 and RSELOUT to realize high logic levels.

TXF, TXD, TXDOUT

TXDOUT is generated from a combination of TXF, TXD, and the internal signal FETG. A software control identical to TXD is available (TXDC, Lower Memory, Register 6Eh). A TXD pulse is internally extended (TXD_{EXT}) by time t_{INITR1} to inhibit the latching of low alarms and warnings related to the APC loop to allow for the loop to stabilize. The nonlatching alarms and warnings are TXP LO, LOS LO, and MON1–MON4 LO alarms and warnings. In addition, TXP LO is disabled from creating FETG. TXF is both an input and an output (Figure 12). See the Transmit Fault (TXF) Output section for a detailed explanation of TXF. Figure 12 shows that the

Figure 12. Logic Diagram 1

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Figure 13. Logic Diagram 2

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same signals and faults can also be used to generate the internal signal FETG (Table 01h/05h, Registers FAh and FBh). FETG is used to send a fast "turn-off" command to the laser driver. The intended use is a direct connection to the MAX3798/MAX3799's TXD input if this is desired. When V_{CC} < POA, TXDOUT is high impedance.

Transmit Fault (TXF) Output

TXF can be triggered by all alarms, warnings, and quick trips (Figure 12). The six ADC alarms, warnings, and the LOS quick trips require enabling (Table 01h/05h, Registers F8h and FDh). See Figures 14a and 14b for nonlatched and latched operation. Latching of the alarms is controlled by the CNFGB and CNFGC registers (Table 02h, Registers 8Ah–8Bh).

Die Identification

The DS1874 has an ID hardcoded in its die. Two registers (Table 02h, Registers CEh–CFh) are assigned for this feature. The CEh register reads 74h to identify the part as the DS1874, while the CFh register reads the current device version.

3-Wire Master for Controlling the MAX3798/MAX3799

The DS1874 controls the MAX3798/MAX3799 over a proprietary 3-wire interface. The DS1874 acts as the master, initiating communication with and generating the clock for the MAX3798/MAX3799. It is a 3-pin interface consisting of SDAOUT (a bidirectional data line), an SCLOUT clock signal, and a CSELOUT chip-select output (active high).

Protocol

The DS1874 initiates a data transfer by asserting the CSELOUT pin. It then starts to generate a clock signal

Figure 14a. TXF Nonlatched Operation

Figure 14b. TXF Latched Operation

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after the CSELOUT has been set to 1. Each operation consists of 16-bit transfers (15-bit address/data, 1-bit RWN). All data transfers are MSB first.

Write Mode (RWN = 0): The master generates 16 clock cycles at SCLOUT in total. It outputs 16 bits (MSB first) to the SDAOUT line at the falling edge of the clock. The master closes the transmission by setting the CSELOUT to 0.

Read Mode (RWN = 1): The master generates 16 clock cycles at SCLOUT in total. It outputs 8 bits (MSB first) to the SDAOUT line at the falling edge of the clock. The SDAOUT line is released after the RWN bit has been transmitted. The slave outputs 8 bits of data (MSB first) at the rising edge of the clock. The master samples SDAOUT at the falling edge of SCLOUT. The master closes the transmission by setting the CSELOUT to 0.

3-Wire Interface Timing

Figure 15 shows the 3-wire interface timing. Figure 16 shows the 3-wire state machine. See the 3-Wire Digital Interface Specification table for more information.

DS1874 and MAX3798/MAX3799 Communication

Normal Operation

The majority of the communication between the two devices consists of bias adjustments for the APC loop. After each temperature conversion, the laser modulation setting must be updated. Status registers TXSTAT1 and TXSTAT2 are read between temperature updates at a regular interval: tRR (see the Analog Voltage Monitoring Characteristics table). The results are stored in TXSTAT1 and TXSTAT2 (Table 02h, FCh–FDh).

Manual Operation

The MAX3798/MAX3799 are manually controllable using four registers in the DS1874: 3WCTRL, ADDRESS, WRITE, and READ. Commands can be manually issued while the DS1874 is in normal operation mode. It is also possible to suspend normal 3-wire commands so that only manual operation commands are sent (3WCTRL, Table 02h, Register F8h).

Figure 15. 3-Wire Timing

Figure 16. 3-Wire State Machine

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Initialization

During initialization, the DS1874 transfers all its 3-wire EEPROM control registers to the MAX3798/MAX3799.

The 3-wire control registers include the following:

- RXCTRL1
- RXCTRL2
- SET_CML
- SET_LOS
- TXCTRL
- IMODMAX
- IBIASMAX
- • SET_PWCTRL
- SET_TXDE

The control registers are first written when V_{CC} exceeds POA. They are also written if the MAX3798/MAX3799 TX_POR bit is set high (visible in 3W TXSTAT1, bit 7). In the MAX3798/MAX3799, this bit is "sticky" (latches high and is cleared on a read). They are also updated on a rising edge of TXD. Any time one of these events occurs, the DS1874 reads and updates TXSTAT1 and TXSTAT2 and sets SET_IBIAS and SET_IMOD in the MAX3798/MAX3799 to 0.

MAX3798/MAX3799 Register Map and DS1874 Corresponding Location

Note: This register is not present in the DS1874. To access this register, use manual operation (see the Manual Operation section).

I2C Communication I2C Definitions

The following terminology is commonly used to describe I2C data transfers.

Master device: The master device controls the slave devices on the bus. The master device generates SCL clock pulses and START and STOP conditions.

Slave devices: Slave devices send and receive data at the master's request.

Bus idle or not busy: Time between STOP and START conditions when both SDA and SCL are inactive and in their logic-high states.

START condition: A START condition is generated by the master to initiate a new data transfer with a slave. Transitioning SDA from high to low while SCL remains high generates a START condition. See Figure 17 for applicable timing.

STOP condition: A STOP condition is generated by the master to end a data transfer with a slave. Transitioning SDA from low to high while SCL remains high generates a STOP condition. See Figure 17 for applicable timing.

Repeated START condition: The master can use a repeated START condition at the end of one data transfer to indicate that it will immediately initiate a new data transfer following the current one. Repeated STARTs are commonly used during read operations to identify a specific memory address to begin a data transfer. A repeated START condition is issued identically to a normal START condition. See Figure 17 for applicable timing.

Bit write: Transitions of SDA must occur during the low state of SCL. The data on SDA must remain valid and unchanged during the entire high pulse of SCL plus the setup and hold time requirements (Figure 17). Data is shifted into the device during the rising edge of the SCL.

Bit read: At the end a write operation, the master must release the SDA bus line for the proper amount of setup time (Figure 17) before the next rising edge of SCL during a bit read. The device shifts out each bit of data on SDA at the falling edge of the previous SCL pulse and the data bit is valid at the rising edge of the current SCL pulse. Remember that the master generates all SCL clock pulses, including when it is reading bits from the slave.

Acknowledgement (ACK and NACK): An acknowledgement (ACK) or not acknowledge (NACK) is always the ninth bit transmitted during a byte transfer. The device receiving data (the master during a read or the slave during a write operation) performs an ACK by transmitting a zero during the ninth bit. A device performs a NACK by transmitting a one during the 9th bit. Timing (Figure 17) for the ACK and NACK is identical to all other bit writes. An ACK is the acknowledgment that the device is properly receiving data. A NACK is used to terminate a read

Figure 17. I ²C Timing

sequence or as an indication that the device is not receiving data.

Byte write: A byte write consists of 8 bits of information transferred from the master to the slave (most significant bit first) plus a 1-bit acknowledgement from the slave to the master. The 8 bits transmitted by the master are done according to the bit-write definition and the acknowledgement is read using the bit-read definition.

Byte read: A byte read is an 8-bit information transfer from the slave to the master plus a 1-bit ACK or NACK from the master to the slave. The 8 bits of information that are transferred (most significant bit first) from the slave to the master are read by the master using the bit-read definition, and the master transmits an ACK using the bit-write definition to receive additional data bytes. The master must NACK the last byte read to terminate communication so the slave returns control of SDA to the master.

Slave address byte: Each slave on the I²C bus responds to a slave address byte sent immediately following a START condition. The slave address byte contains the slave address in the most significant 7 bits and the R/\overline{W} bit in the least significant bit.

The DS1874 responds to two slave addresses. The auxiliary memory always responds to a fixed 12° C slave address, A0h. The Lower Memory and Tables 00h–08h respond to I2C slave addresses that can be configured to any value between 00h–FEh using the DEVICE ADDRESS byte (Table 02h, Register 8Ch). The user also must set the ASEL bit (Table 02h, Register 89h) for this address to be active. By writing the correct slave address with $R/\overline{W} = 0$, the master indicates it will write data to the slave. If R/\overline{W} $= 1$, the master reads data from the slave. If an incorrect slave address is written, the DS1874 assumes the master is communicating with another I 2C device and ignores the communications until the next START condition is sent. If the main device's slave address is programmed to be A0h, access to the auxiliary memory is disabled.

Memory address: During an I²C write operation to the DS1874, the master must transmit a memory address to identify the memory location where the slave is to store the data. The memory address is always the second byte transmitted during a write operation following the slave address byte.

I2C Protocol

Writing a single byte to a slave: The master must generate a START condition, write the slave address byte (R/\overline{W} = 0), write the memory address, write the byte of data, and generate a STOP condition. Remember the master must read the slave's acknowledgement during all byte-write operations.

Writing multiple bytes to a slave: To write multiple bytes to a slave, the master generates a START condition, writes the slave address byte $(R/\overline{W} = 0)$, writes the memory address, writes up to 8 data bytes, and generates a STOP condition. The DS1874 writes 1 to 8 bytes (one page or row) with a single write transaction. This is internally controlled by an address counter that allows data to be written to consecutive addresses without transmitting a memory address before each data byte is sent. The address counter limits the write to one 8-byte page (one row of the memory map). Attempts to write to additional pages of memory without sending a STOP condition between pages results in the address counter wrapping around to the beginning of the present row.

For example, a 3-byte write starts at address 06h and writes 3 data bytes (11h, 22h, and 33h) to three "consecutive" addresses. The result is that addresses 06h and 07h would contain 11h and 22h, respectively, and the third data byte, 33h, would be written to address 00h.

To prevent address wrapping from occurring, the master must send a STOP condition at the end of the page, then wait for the bus-free or EEPROM write time to elapse. Then the master can generate a new START condition and write the slave address byte (\overline{RW} = 0) and the first memory address of the next memory row before continuing to write data.

Acknowledge polling: Any time a EEPROM page is written, the DS1874 requires the EEPROM write time (tW) after the STOP condition to write the contents of the page to EEPROM. During the EEPROM write time, the DS1874 will not acknowledge its slave address because it is busy. It is possible to take advantage of that phenomenon by repeatedly addressing the DS1874, which allows the next page to be written as soon as the DS1874 is ready to receive the data. The alternative to acknowledge polling is to wait for maximum period of tw to elapse before attempting to write again to the DS1874.

EEPROM write cycles: When EEPROM writes occur, the DS1874 writes the whole EEPROM memory page, even if only a single byte on the page was modified. Writes that do not modify all 8 bytes on the page are allowed and do not corrupt the remaining bytes of memory on the same page. Because the whole page is written, bytes on the page that were not modified during the transaction are still subject to a write

Figure 18. Example l²C Timing

cycle. This can result in a whole page being worn out over time by writing a single byte repeatedly. Writing a page one byte at a time wears the EEPROM out eight times faster than writing the entire page at once. The DS1874's EEPROM write cycles are specified in the Nonvolatile Memory Characteristics table. The specification shown is at the worst-case temperature. It can handle approximately ten times that many writes at room temperature. Writing to SRAMshadowed EEPROM memory with SEEB = 1 does not count as an EEPROM write cycle when evaluating the EEPROM's estimated lifetime.

Reading a single byte from a slave: Unlike the write operation that uses the memory address byte to define where the data is to be written, the read operation occurs at the present value of the memory address counter. To read a single byte from the slave, the master generates a START condition, writes the slave address byte with $R/\overline{W} = 1$, reads the data byte with a NACK to indicate the end of the transfer, and generates a STOP condition.

Manipulating the address counter for reads: A dummy write cycle can be used to force the address pointer to a particular value. To do this, the master generates a START condition, writes the slave

address byte (R/\overline{W} = 0), writes the memory address where it desires to read, generates a repeated START condition, writes the slave address byte (R/\overline{W}) = 1), reads data with ACK or NACK as applicable, and generates a STOP condition.

Memory Organization

The DS1874 features nine separate memory tables that are internally organized into 8-byte rows.

The **Lower Memory** is addressed from 00h to 7Fh and contains alarm and warning thresholds, flags, masks, several control registers, password entry area (PWE), and the table-select byte.

Table 01h primarily contains user EEPROM (with PW1 level access) as well as alarm and warning-enable bytes.

Table 02h is a multifunction space that contains configuration registers, scaling and offset values, passwords, interrupt registers as well as other miscellaneous control bytes.

Table 04h contains a temperature-indexed LUT for control of the modulation voltage. The modulation LUT can be programmed in 2°C increments over the -40°C to +102°C range.

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Table 05h is empty by default. It can be configured to contain the alarm- and warning-enable bytes from Table 01h, Registers F8h–FFh with the MASK bit enabled (Table 02h, Register 89h). In this case Table 01h is empty.

Table 06h contains a temperature-indexed LUT that allows the APC set point to change as a function of temperature to compensate for tracking error (TE). The APC LUT has 36 entries that determine the APC setting in 4°C windows between -40°C and +100°C.

Table 07h contains a temperature-indexed LUT for control of DAC1. The LUT has 36 entries that determine the DAC setting in 4°C windows between -40°C and +100°C.

Table 08h contains a temperature-indexed LUT for control of DAC2. The LUT has 36 entries that determine the DAC setting in 4°C windows between -40°C and +100°C.

Auxiliary Memory (device A0h) contains 256 bytes of EE memory accessible from address 00h–FFh. It is selected with the device address of A0h.

See the Register Descriptions section for more complete details of each byte's function, as well as for read/write permissions for each byte.

Shadowed EEPROM

Many NV memory locations (listed within the Register Descriptions section) are actually shadowed EEPROM that are controlled by the SEEB bit in Table 02h, Register 80h.

The DS1874 incorporates shadowed-EEPROM memory locations for key memory addresses that can be written many times. By default the shadowed-EEPROM bit, SEEB, is not set and these locations act as ordinary EEPROM. By setting SEEB, these locations function like SRAM cells, which allow an infinite number of write cycles without concern of wearing out the EEPROM. Setting SEEB also eliminates the requirement for the EEPROM write time, tw. Because changes made with SEEB enabled do not affect the EEPROM, these changes are not retained through power cycles. The power-on value is the last value written with SEEB disabled. This function can be used to limit the number of EEPROM writes during calibration or to change the monitor thresholds periodically during normal operation helping to reduce the number of times EEPROM is written. Figure 19 indicates which locations are shadowed EEPROM.

Figure 19. Memory Map

Register Descriptions

The register maps show each byte/word (2 bytes) in terms of its row in the memory. The first byte in the row is located in memory at the row address (hexadecimal) in the leftmost column. Each subsequent byte on the row is one/two memory locations beyond the previous byte/word's address. A total of 8 bytes are present on each row. For more information about each of these bytes see the corresponding register description.

Lower Memory Register Map

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Table 01h Register Map

The ALARM ENABLE bytes (Registers F8h–FFh) can be configured to exist in Table 05h instead of here at Table 01h with the MASK bit (Table 02h, Register 89h). If the row is configured to exist in Table 05h, then these locations are empty in Table 01h.

The access codes represent the factory default values of PW_ENA and PW_ENB (Table 02h, Registers C0h–C1h).

Table 02h Register Map

*The final result must be XORed with BB40h before writing to this register.

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Table 04h Register Map

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Table 05h Register Map

Table 05h is empty by default. It can be configured to contain the alarm and warning-enable bytes from Table 01h, Registers F8h–FFh with the MASK bit enabled (Table 02h, Register 89h). In this case Table 01h is empty.

Table 06h Register Map

The access codes represent the factory default values of PW_ENA and PW_ENB (Table 02h, Registers C0h–C1h).

Table 07h Register Map

Table 08h Register Map

Auxiliary A0h Memory Register Map

The access codes represent the factory default values of PW_ENA and PW_ENB (Table 02h, Registers C0h–C1h).

Lower Memory Register Descriptions

Lower Memory, Register 00h–01h: TEMP ALARM HI Lower Memory, Register 04h–05h: TEMP WARN HI

Temperature measurement updates above this two's complement threshold set corresponding alarm or warning bits. Temperature measurement updates equal to or below this threshold clear alarm or warning bits.

Lower Memory, Register 02h–03h: TEMP ALARM LO Lower Memory, Register 06h–07h: TEMP WARN LO

Temperature measurement updates below this two's complement threshold set corresponding alarm or warning bits. Temperature measurement updates equal to or above this threshold clear alarm or warning bits.

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Lower Memory, Register 08h-09h: V_{CC} ALARM HI **Lower Memory, Register 0Ch–0Dh: VCC WARN HI Lower Memory, Register 10h–11h: MON1 ALARM HI Lower Memory, Register 14h–15h: MON1 WARN HI Lower Memory, Register 18h–19h: MON2 ALARM HI Lower Memory, Register 1Ch–1Dh: MON2 WARN HI Lower Memory, Register 20h–21h: MON3 ALARM HI Lower Memory, Register 24h–25h: MON3 WARN HI Lower Memory, Register 28h–29h: MON4 ALARM HI Lower Memory, Register 2Ch–2Dh: MON4 WARN HI**

Voltage measurement updates above this unsigned threshold set corresponding alarm or warning bits. Voltage measurements equal to or below this threshold clear alarm or warning bits.
Lower Memory, Register 0Ah–0Bh: VCC ALARM LO Lower Memory, Register 0Eh–0Fh: VCC WARN LO Lower Memory, Register 12h–13h: MON1 ALARM LO Lower Memory, Register 16h–17h: MON1 WARN LO Lower Memory, Register 1Ah–1Bh: MON2 ALARM LO Lower Memory, Register 1Eh–1Fh: MON2 WARN LO Lower Memory, Register 22h–23h: MON3 ALARM LO Lower Memory, Register 26h–27h: MON3 WARN LO Lower Memory, Register 2Ah–2Bh: MON4 ALARM LO Lower Memory, Register 2Eh–2Fh: MON4 WARN LO

Voltage measurement updates below this unsigned threshold set corresponding alarm or warning bits. Voltage measurements equal to or above this threshold clear alarm or warning bits.

Lower Memory, Register 30h–5Fh: EE

PW2 level access-controlled EEPROM.

Lower Memory, Register 60h–61h: TEMP VALUE

Signed two's complement direct-to-temperature measurement.

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Lower Memory, Register 62h–63h: VCC VALUE Lower Memory, Register 64h–65h: MON1 VALUE Lower Memory, Register 66h–67h: MON2 VALUE Lower Memory, Register 68h–69h: MON3 VALUE Lower Memory, Register 6Ah–6Bh: MON4 VALUE

Left-justified unsigned voltage measurement.

Lower Memory, Register 6Ch–6Dh: RESERVED

These registers are reserved. The value when read is 00h.

Lower Memory, Register 6Eh: STATUS

Lower Memory, Register 6Fh: UPDATE

Lower Memory, Register 70h: ALARM3

Lower Memory, Register 71h: ALARM2

Lower Memory, Register 72h: ALARM1

Lower Memory, Register 73h: ALARM0

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Lower Memory, Register 74h: WARN3

Lower Memory, Register 75h: WARN2

Lower Memory, Register 76h–7Ah: RESERVED MEMORY

These registers are reserved. The value when read is 00h.

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Lower Memory, Register 7Bh–7Eh: Password Entry (PWE)

There are two passwords for the DS1874. Each password is 4 bytes long. The lower level password (PW1) has all the access of a normal user plus those made available with PW1. The higher level password (PW2) has all the access of PW1 plus those made available with PW2. The values of the passwords reside in EEPROM inside PW2 memory. At power-up, all PWE bits are set to 1. All reads at this location are 0.

Lower Memory, Register 7Fh: Table Select (TBL SEL)

The upper memory tables of the DS1874 are accessible by writing the desired table value in this register. The power-on value of this register is defined by the value written to TBLSELPON (Table 02h, Register C7h).

EEPROM for PW1 and/or PW2 level access.

Table 01h, Register F8h: ALARM EN3

Layout is identical to ALARM₃ in Lower Memory, Register 70h. Enables alarms to create TXFINT (Lower Memory, Register 71h) logic. The MASK bit (Table 02h, Register 89h) determines whether this memory exists in Table 01h or 05h.

Table 01h, Register F9h: ALARM EN2

Layout is identical to ALARM₂ in Lower Memory, Register 71h. Enables alarms to create TXFINT (Lower Memory, Register 71h) logic. The MASK bit (Table 02h, Register 89h) determines whether this memory exists in Table 01h or 05h.

Table 01h, Register FAh: ALARM EN1

Layout is identical to ALARM₁ in Lower Memory, Register 72h. Enables alarms to create internal signal FETG (see Figure 12) logic. The MASK bit (Table 02h, Register 89h) determines whether this memory exists in Table 01h or 05h.

Table 01h, Register FBh: ALARM EN0

Layout is identical to ALARM₀ in Lower Memory, Register 73h. The MASK bit (Table 02h, Register 89h) determines whether this memory exists in Table 01h or 05h.

Table 01h, Register FCh: WARN EN3

Layout is identical to WARN₃ in Lower Memory, Register 74h. Enables warnings to create TXFINT (Lower Memory, Register 71h) logic. The MASK bit (Table 02h, Register 89h) determines whether this memory exists in Table 01h or 05h.

Table 01h, Register FDh: WARN EN2

Layout is identical to WARN₂ in Lower Memory, Register 75h. Enables warnings to create TXFINT (Lower Memory, Register 71h) logic. The MASK bit (Table 02h, Register 89h) determines whether this memory exists in Table 01h or 05h.

Table 01h, Register FEh–FFh: RESERVED

These registers are reserved.

Table 02h Register Descriptions

Table 02h, Register 80h: MODE

Table 02h, Register 81h: Temperature Index (TINDEX)

Holds the calculated index based on the temperature measurement. This index is used for the address during lookup of Tables 04h, 06h–08h. Temperature measurements below -40°C or above +102°C are clamped to 08h and C7h, respectively. The calculation of TINDEX is as follows:

$$
TINDER = \frac{Temp_Value + 40°C}{2°C} + 80h
$$

For the temperature-indexed LUTs, the index used during the lookup function for each table is as follows:

Table 02h, Register 82h–83h: MODULATION REGISTER

The digital value used for MODULATION and recalled from Table 04h at the adjusted memory address found in TINDEX. This register is updated at the end of the temperature conversion.

Table 02h, Register 84h–85h: DAC1 VALUE

The digital value used for DAC1 and recalled from Table 07h at the adjusted memory address found in TINDEX. This register is updated at the end of the temperature conversion.

$$
V_{\text{DAC1}} = \frac{\text{REFIN}}{512} \times \text{DAC1 VALUE}
$$

Table 02h, Register 86h–87h: DAC2 VALUE

The digital value used for DAC2 and recalled from Table 08h at the adjusted memory address found in TINDEX. This register is updated at the end of the temperature conversion.

$$
V_{\text{DAC2}} = \frac{\text{REFIN}}{512} \times \text{DAC2 VALUE}
$$

Table 02h, Register 88h: SAMPLE RATE

Table 02h, Register 89h: CNFGA

Table 02h, Register 8Ah: CNFGB

Table 02h, Register 8Bh: CNFGC

Table 02h, Register 8Ch: DEVICE ADDRESS

This value becomes the I²C slave address for the main memory when the ASEL (Table 02h, Register 89h) bit is set. If A0h is programmed to this register, the auxiliary memory is disabled.

Table 02h, Register 8Dh: RESERVED

This register is reserved.

Table 02h, Register 8Eh: RIGHT-SHIFT1 (RSHIFT1)

Allows for right-shifting the final answer of MON1 and MON2 voltage measurements. This allows for scaling the measurements to the smallest full-scale voltage and then right-shifting the final result so the reading is weighted to the correct LSB.

Table 02h, Register 8Fh: RIGHT-SHIFT0 (RSHIFT0)

Allows for right-shifting the final answer of MON3 and MON4 voltage measurements. This allows for scaling the measurements to the smallest full-scale voltage and then right-shifting the final result so the reading is weighted to the correct LSB. The MON3 right-shifting is only available for the fine mode of operation. The coarse mode does not right-shift.

Table 02h, Register 90h–91h: RESERVED

These registers are reserved.

Table 02h, Register 92h–93h: VCC SCALE Table 02h, Register 94h–95h: MON1 SCALE Table 02h, Register 96h–97h: MON2 SCALE Table 02h, Register 98h–99h: MON3 FINE SCALE Table 02h, Register 9Ah–9Bh: MON4 SCALE Table 02h, Register 9Ch–9Dh: MON3 COARSE SCALE

Controls the scaling or gain of the FS voltage measurements. The factory-calibrated value produces an FS voltage of 6.5536V for V_{CC}; 2.5V for MON1, MON2, MON4; and 0.3125V for MON3 fine.

Table 02h, Register 9Eh–A1h: RESERVED

These registers are reserved.

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WRITE ACCESS PW2 or (PW1 and RWTBL246) MEMORY TYPE Nonvolatile (SEE)

A2h, A4h, A6h, A8h, AAh, ACh S | S | 2^{15} | 2^{14} | 2^{13} | 2^{12} | 2^{11} | 2^{10} A3h, A5h, A7h, A9h, ABh, ADh 29 | 28 | 27 | 26 | 25 | 24 | 23 | 22

BIT 7 BIT 0

Allows for offset control of these voltage measurements if desired. This number is two's complement.

Table 02h, Register AEh–AFh: INTERNAL TEMP OFFSET

Allows for offset control of temperature measurement if desired. The final result must be XORed with BB40h before writing to this register. Factory calibration contains the desired value for a reading in degrees Celsius.

Table 02h, Register B0h–B3h: PW1

The PWE value is compared against the value written to this location to enable PW1 access. At power-on, the PWE value is set to all ones. Thus, writing these bytes to all ones grants PW1 access on power-on without writing the password entry. All reads of this register are 00h.

Table 02h, Register B4h–B7h: PW2

The PWE value is compared against the value written to this location to enable PW2 access. At power-on, the PWE value is set to all ones. Thus, writing these bytes to all ones grants PW2 access on power-on without writing the password entry. All reads of this register are 00h.

Table 02h, Register B8h: LOS RANGING

This register controls the full-scale range of the quick-trip monitoring for the differential input's of MON3.

Table 02h, Register B9h: COMP RANGING

The upper nibble of this byte controls the full-scale range of the quick-trip monitoring for BIAS. The lower nibble of this byte controls the full-scale range for the quick-trip monitoring of the APC reference as well as the closed-loop monitoring of APC.

Table 02h, Register BAh: RESERVED

This register is reserved.

Table 02h, Register BBh: ISTEP

The initial step value used at power-on or after a TXD pulse to control the BIAS register. At startup, this value plus $2⁰$ = 1 is continuously added to the BIAS register value until the APC feedback (MON2) is greater than its threshold. At that time, a binary search is used to complete the startup of the APC closed loop. If the resulting math operation is greater than IBIASMAX (Table 02h, Register EEh), the result is not loaded into the BIAS register, but the binary search is begun to complete the initial search for APC. During startup, the BIAS register steps causing a higher bias value than IBIASMAX do not create the BIAS MAX alarm. The BIAS MAX alarm detection is enabled at the end of the binary search.

Table 02h, Register BCh: HTXP

Fast-comparison DAC threshold adjust for high TXP. This value is added to the APC DAC value recalled from Table 06h. If the sum is greater than 0xFF, 0xFF is used. Comparisons greater than V_{HTXP}, compared against VMON2, create a TXP HI alarm. The same ranging applied to the APC DAC should be used here.

$$
V_{\text{HTXP}} = \frac{\text{Full Scale}}{255} \times (\text{HTXP} + \text{APC DAC})
$$

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Table 02h, Register BDh: LTXP

Fast-comparison DAC threshold adjust for low TXP. This value is subtracted from the APC DAC value recalled from Table 06h. If the difference is less than 0x00, 0x00 is used. Comparisons less than VLTXP, compared against V_{MON2}, create a TXP LO alarm. The same ranging applied to the APC DAC should be used here.

$$
V_{LTXP} = \frac{Full Scale}{255} \times (APC DAC - LTXP)
$$

Table 02h, Register BEh: HLOS

Fast-comparison DAC threshold adjust for high LOS. The combination of HLOS and LLOS creates a hysteresis comparator. As RSSI falls below the LLOS threshold, the LOS LO alarm bit is set to 1. The LOS alarm remains set until the RSSI input is found above the HLOS threshold setting, which clears the LOS LO alarm bit and sets the LOS HI alarm bit. At power-on, both LOS LO and LOS HI alarm bits are 0 and the hysteresis comparator uses the LLOS threshold setting.

$$
V_{\text{HLOS}} = \frac{\text{Full Scale}}{255} \times \text{HLOS}
$$

Table 02h, Register BFh: LLOS

Fast-comparison DAC threshold adjust for low LOS. See HLOS (Table 02h, Register BEh) for functional description.

$$
V_{LLOS} = \frac{Full Scale}{255} \times LLOS
$$

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Table 02h, Register C0h: PW_ENA

Table 02h, Register C1h: PW_ENB

Table 02h, Register C2h: MODTI

The modulation temperature index defines the TempCo boundary for the MODULATION LUT. The MODTC bit (Table 02h, Register C6h) defines the polarity of the TempCo.

$$
MODTI = \frac{Temp_Value + 40°C}{2°C} + 80h
$$

Table 02h, Register C3h: DAC1TI

DAC1 temperature index (DAC1TI) defines the TempCo boundary for the DAC1 LUT. The DAC1TC bit (Table 02h, Register C6h) defines the polarity of the TempCo. This value is compared with the adjusted memory address used during the LUT recall, not the value in the TINDEX register (Table 02h, Register 81h).

> $\text{DAC1TI} = \frac{\text{Temp}}{100} = \frac{\text{Value} + 40^{\circ}\text{C}}{100}$ $\frac{4^{\circ}C}{4^{\circ}C}$ + 80h
Table 02h, Register C4h: DAC2TI

DAC2 temperature index defines the TempCo boundary for the DAC2 LUT. The DAC2TC bit (Table 02h, Register C6h) defines the polarity of the TempCo. This value is compared with the adjusted memory address used during the LUT recall, not the value in the TINDEX register (Table 02h, Register 81h).

$$
DAC2TI = \frac{Temp_Value + 40°C}{4°C} + 80h
$$

Table 02h, Register C5h: RESERVED

This register is reserved.

Table 02h, Register C6h: LUTTC

Table 02h, Register C7h: TBLSELPON

Chooses the initial value for the table-select byte (Lower Memory, Register 7Fh) at power-on.

Table 02h, Register C8h–C9h: MAN BIAS

When BIAS EN (Table 02h, Register 80h) is written to 0, writes to these bytes control the BIAS register, which then updates the MAX3798/MAX3799 SET_IBIAS register.

Table 02h, Register CAh: MAN_CNTL

When BIAS EN (Table 02h, Register 80h) is written to 0, MAN_CLK controls the updates of the MAN BIAS value to the BIAS register. This new value is sent through the 3-wire interface. The values of MAN BIAS must be written with a separate write command. Setting MAN_CLK to a 1 clocks the MAN BIAS value to the BIAS register, which then updates the MAX3798/MAX3799 SET_IBIAS register.

1) Write the MAN BIAS value with a write command.

2) Set the MAN_CLK bit to a 1 with a separate write command.

3) Clear the MAN_CLK bit to a 0 with a separate write command.

Table 02h, Register CBh–CCh: BIAS REGISTER

The digital value used for BIAS and resolved from the APC. This register is updated after each decision of the APC loop.

Table 02h, Register CDh: APC DAC

The digital value used for APC reference and recalled from Table 06h at the adjusted memory address found in TINDEX. This register is updated at the end of the temperature conversion.

Table 02h, Register CEh: DEVICE ID

Hardwired connections to show the device ID.

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Table 02h, Register CFh: DEVICE VER

Hardwired connections to show the device version.

Table 02h, Register D0h–D7h: HBATH

High-Bias Alarm Threshold (HBATH) is a digital clamp used to ensure that the DAC setting for BIAS currents does not exceed a set value. The table below shows the range of temperature for each byte's location. The table shows a rising temperature; for a falling temperature there is 1°C of hysteresis.

Table 02h, Register D8h–E7h: EMPTY

These registers do not exist.

Table 02h, Register E8h: RXCTRL1

MAX3798/MAX3799 register. After either V_{CC} exceeds POA (after a POR event), the MAX3798/MAX3799 TX_POR bit is set high (visible in 3W TXSTAT1, Bit 7) or on a rising edge of TXD, this value is written to the MAX3798/ MAX3799 through the 3-wire interface.

Table 02h, Register E9h: RXCTRL2

MAX3798/MAX3799 register. After either V_{CC} exceeds POA (after a POR event), the MAX3798/MAX3799 TX_POR bit is set high (visible in 3W TXSTAT1, Bit 7) or on a rising edge of TXD, this value is written to the MAX3798/ MAX3799 through the 3-wire interface.

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Table 02h, Register EAh: SETCML

EAh 2⁷ 2⁶ 2⁵ 2⁴ 2³ 2² 2² 2¹ 2⁰ BIT 7 BIT 0

> MAX3798/MAX3799 register. After either V_{CC} exceeds POA (after a POR event), the MAX3798/MAX3799 TX_POR bit is set high (visible in 3W TXSTAT1, Bit 7) or on a rising edge of TXD, this value is written to the MAX3798/ MAX3799 through the 3-wire interface.

Table 02h, Register EBh: SETLOS

MAX3798/MAX3799 register. After either V_{CC} exceeds POA (after a POR event), the MAX3798/MAX3799 TX_POR bit is set high (visible in 3W TXSTAT1, Bit 7) or on a rising edge of TXD, this value is written to the MAX3798/ MAX3799 through the 3-wire interface.

Table 02h, Register ECh: TXCTRL

MAX3798/MAX3799 register. After either V_{CC} exceeds POA (after a POR event), the MAX3798/MAX3799 TX_POR bit is set high (visible in 3W TXSTAT1, Bit 7) or on a rising edge of TXD, this value is written to the MAX3798/ MAX3799 through the 3-wire interface.

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Table 02h, Register EDh: IMODMAX

MAX3798/MAX3799 register. After either V_{CC} exceeds POA (after a POR event), the MAX3798/MAX3799 TX_POR bit is set high (visible in 3W TXSTAT1, Bit 7) or on a rising edge of TXD, this value is written to the MAX3798/ MAX3799 through the 3-wire interface.

Table 02h, Register EEh: IBIASMAX

MAX3798/MAX3799 register. After either V_{CC} exceeds POA (after a POR event), the MAX3798/MAX3799 TX_POR bit is set high (visible in 3W TXSTAT1, Bit 7) or on a rising edge of TXD, this value is written to the MAX3798/ MAX3799 through the 3-wire interface. In addition, this value defines the maximum DAC value allowed for the upper 8 bits of BIAS output during APC closed-loop operations. During the intial step and binary search, this value does not cause an alarm but still clamps the BIAS register value. After the startup seqence (or normal APC operations), if the APC loop tries to create a BIAS value greater than this setting, it is clamped and creates a MAX BIAS alarm.

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Table 02h, Register EFh: SETPWCTRL

BIT 7 BIT 0

MAX3798/MAX3799 register. After either V_{CC} exceeds POA (after a POR event), the MAX3798/MAX3799 TX_POR bit is set high (visible in 3W TXSTAT1, Bit 7) or on a rising edge of TXD, this value is written to the MAX3798/ MAX3799 through the 3-wire interface.

Table 02h, Register F0h: SETTXDE

MAX3798/MAX3799 register. After either V_{CC} exceeds POA (after a POR event), the MAX3798/MAX3799 TX_POR bit is set high (visible in 3W TXSTAT1, Bit 7) or on a rising edge of TXD, this value is written to the MAX3798/ MAX3799 through the 3-wire interface.

Table 02h, Register F1h–F7h: RESERVED

These registers are reserved.

Table 02h, Register F8h: 3WCTRL

Table 02h, Register F9h: ADDRESS

BIT 7 BIT 2 BIT 0 BIT 0

This byte is used during manual 3-wire communication. When a manual read or write is initiated, this register contains the address for the operation.

Table 02h, Register FAh: WRITE

This byte is used during manual 3-wire communication. When a manual write is initiated, this register contains the data for the operation.

Table 02h, Register FBh: READ

This byte is used during maunual 3-wire communication. When a manual read is initiated, the return data is stored in this register.

Table 02h, Register FCh: TXSTAT1

MAX3798/MAX3799 register. This value is read from the MAX3798/MAX3799 with the 3-wire interface every tRR (see the MAX3798/MAX3799 electrical characteristics).

Table 02h, Register FDh: TXSTAT2

MAX3798/MAX3799 register. This value is read from the MAX3798/MAX3799 with the 3-wire interface every tRR (see the MAX3798/MAX3799 electrical characteristics).

Table 02h, Register FEh–FFh: RESERVED

These registers are reserved.

Table 04h Register Description

Table 04h, Register 80h–C7h: MODULATION LUT

The digital value for the modulation DAC output.

The MODULATION LUT is a set of registers assigned to hold the temperature profile for the MODULATION REGISTER. The values in this table determine the set point for the modulation voltage. The temperature measurement is used to index the LUT (TINDEX, Table 02h, Register 81h) in 2°C increments from -40°C to +102°C, starting at 80h in Table 04h. Register 80h defines the -40°C to -38°C MOD output, Register 81h defines the -38°C to -36°C MOD output, and so on. Values recalled from this EEPROM memory table are written into the MODULATION REGISTER (Table 02h, Register 82h–83h) location that holds the value until the next temperature conversion. The DS1874 can be placed into a manual mode (MOD EN bit, Table 02h, Register 80h), where the MODULATION REGISTER is directly controlled for calibration. If the temperature compensation functionality is not required, then program the entire Table 04h to the desired modulation setting.

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Table 06h Register Descriptions

Table 06h, Register 80h–A3h: APC LUT

BIT 7 BIT 0

The APC LUT is a set of registers assigned to hold the temperature profile for the APC reference DAC. The values in this table combined with the APC bits in the COMP RANGING register (Table 02h, Register B9h) determine the set point for the APC loop. The temperature measurement is used to index the LUT (TINDEX, Table 02h, Register 81h) in 4°C increments from -40°C to +100°C, starting at Register 80h in Table 06h. Register 80h defines the -40°C to -36°C APC reference value, Register 81h defines the -36°C to -32°C APC reference value, and so on. Values recalled from this EEPROM memory table are written into the APC DAC (Table 02h, Register CDh) location that holds the value until the next temperature conversion. The DS1874 can be placed into a manual mode (APC EN bit, Table 02h, Register 80h), where the APC DAC can be directly controlled for calibration. If TE temperature compensation is not required by the application, program the entire LUT to the desired APC set point.

Table 06h, Register A4h–A7h: RESERVED

These registers are reserved.

Table 07h Register Descriptions

The DAC1 LUT is a set of registers assigned to hold the PWM profile for DAC1. The values in this table determine the set point for DAC1. The temperature measurement is used to index the LUT (TINDEX, Table 02h, Register 81h) in 4°C increments from -40°C to +100°C, starting at Register 80h in Table 07h. Register 80h defines the -40°C to -36°C DAC1 value, Register 81h defines -36°C to -32°C DAC1 value, and so on. Values recalled from this EEPROM memory table are written into the DAC1 VALUE (Table 02h, Registers 84h–85h) location, which holds the value until the next temperature conversion. The part can be placed into a manual mode (DAC1 EN bit, Table 02h, Register 80h), where DAC1 can be directly controlled for calibration. If temperature compensation is not required by the application, program the entire LUT to the desired DAC1 set point.

Table 07h, Register A4h–A7h: RESERVED

Table 07h, Register 80h–A3h: DAC1 LUT

These registers are reserved.

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Table 08h Register Descriptions

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Table 08h, Register 80h–A3h: DAC2 LUT

The DAC2 LUT is set of registers assigned to hold the PWM profile for DAC2. The values in this table determine the set point for DAC2. The temperature measurement is used to index the LUT (TINDEX, Table 02h, Register 81h) in 4°C increments from -40°C to +100°C, starting at Register 80h in Table 08h. Register 80h defines the -40°C to -36°C DAC2 value, Register 81h defines -36°C to -32°C DAC2 value, and so on. Values recalled from this EEPROM memory table are written into the DAC2 VALUE (Table 02h, Registers 86h–87h) location that holds the value until the next temperature conversion. The DS1874 can be placed into a manual mode (DAC2 EN bit, Table 02h, Register 80h), where DAC2 can be directly controlled for calibration. If temperature compensation is not required by the application, program the entire LUT to the desired DAC2 set point.

Table 08h, Register A4h–A7h: RESERVED

These registers are reserved.

Auxiliary Memory A0h Register Descriptions

Auxiliary Memory A0h, Register 00h–FFh: EEPROM

Accessible with the slave address A0h.

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Applications Information

Power-Supply Decoupling

To achieve best results, it is recommended that the power supply is decoupled with a 0.01µF or a 0.1µF capacitor. Use high-quality, ceramic, surface-mount capacitors, and mount the capacitors as close as possible to the V_{CC} and GND pins to minimize lead inductance.

SDA and SCL Pullup Resistors

SDA is an open-collector output on the DS1874 that requires a pullup resistor to realize high logic levels. A master using either an open-collector output with a pullup resistor or a push-pull output driver can be utilized for SCL. Pullup resistor values should be chosen to ensure that the rise and fall times listed in the I^2C AC Electrical Characteristics table are within specification.

Package Information

For the latest package outline information and land patterns, go to **www.maxim-ic.com/packages**.

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