# 600 mA Step-Down DC/DC Converter with Built-in Inductor

## FEATURES

- Built-in inductor and transistors
- Operating Input Voltage Range: 2.0 V ~ 6.0 V (A/B/C types) or 1.8 V ~ 6.0 V (G type)
- Output Voltage Range: 0.8 V ~ 4.0 V
- Output Voltage Error: ±2%
- Output Current: 600 mA
- High Efficiency: 90% (V<sub>IN</sub> = 4.2 V, V<sub>OUT</sub> = 3.3V)
- Oscillation Frequency: 3 MHz
- Maximum Duty Cycle: 100%
- **Operating Modes**: PWM, PWM/PFM auto select
- **Functions:** Built-in Current Limit, High Speed Load Capacitor Discharge, Soft start
- Operating Ambient temperature: -40 ~ +85<sup>o</sup>C
- Package Size: 2.5 x 2.0 x 1.0 mm
- EU RoHS Compliant, Pb Free

## APPLICATION

- Mobile Phones
- Bluetooth headsets
- Digital home appliances
- Office automation equipment
- Various portable equipment

## DESCRIPTION

The IXD9205/06 series of converters are synchronous step-down DC/DC converters with an inductor and a control IC in one tiny  $(2.5 \times 2.0 \times 1.0 \text{ mm})$  package. A stable power supply with an output current of 600 mA requires only two capacitors connected externally.

Operating voltage range is from 2.0 V to 6.0 V (1.8 V  $\sim$  6.0 V for IXD920xG version).

Output voltage is internally set in a range from 0.8 V to 4.0 V in 0.05 V increments. The device operates at 3.0 MHz switching frequency, and includes a 0.42  $\Omega$  P-channel switching transistor and a 0.52  $\Omega$  N-channel transistor for synchronous rectification. The IXD9205 series operates in PWM mode, while the IXD9206 series automatically switches between PWM/PFM modes.

Automatic PWM/PFM switching allows fast response to the load changes, low ripple noise, and high efficiency over the full range of loads.

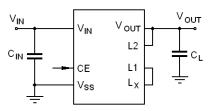
The CE pin allows setting of the device into stand-by mode with a current consumption below 1.0  $\mu$ A.

The built-in Under Voltage Lockout (UVLO) function forces the internal switching transistor OFF when input voltage becomes less than 1.4V

The B and G versions of the IXD9205/06 series have a fast soft start function internally set at 0.25 ms (typ).

The B, C, and G versions of the IXD9205/06 series also have fast load capacitor  $C_{\rm L}$  auto discharge function, which allows for fast  $C_{\rm L}$  discharge through a switch located between the  $L_{\rm X}$  and  $V_{\rm SS}$  pins. When the devices enter stand-by mode, the output voltage quickly returns to the  $V_{\rm SS}$  level because of this function.

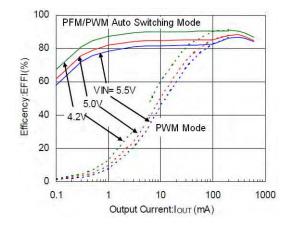
## **TYPICAL APPLICATION CIRCUIT**



Pins L1 –  $L_{X}$  and L2 –  $V_{\text{OUT}}$  should be connected externally

## **TYPICAL PERFORMANCE CHARACTERISTIC**

Efficiency vs. Output Current (f<sub>OSC</sub> = 3.0 MHz, V<sub>OUT</sub> = 3.3 V)



# **ABSOLUTE MAXIMUM RATINGS**

PARAMETER	SYMBOL	RATINGS	UNITS
V <sub>IN</sub> Pin Voltage	V <sub>IN</sub>	- 0.3 ~ 6.5	V
L <sub>x</sub> Pin Voltage	V <sub>LX</sub>	$-0.3 \sim V_{IN} + 0.3^{1}$	V
V <sub>OUT</sub> Pin Voltage	Vout	- 0.3 ~ 6.5	V
FB Pin Voltage	V <sub>FB</sub>	- 0.3 ~ 6.5	V
CE/MODE Pin Voltage	V <sub>CE</sub>	- 0.3 ~ 6.5	V
Lx Pin Current	I <sub>LX</sub>	±1500	mA
Inductor Current at $\Delta T = 40^{\circ}C$	I <sub>LMAX</sub>	1000	mA
Power Dissipation	PD	1000 <sup>2)</sup>	mW
Operating Temperature Range	T <sub>OPR</sub>	- 40 ~ + 85	Do
Storage Temperature Range	T <sub>STG</sub>	- 50 ~ +105	Do

#### NOTE:

- 1.  $L_X$  pin voltage should not exceed  $V_{IN}$  +0.3 V or 6.5 V, which is less.
- 2. Power dissipation shown for a PCB mounted part. Please refer to page 15 for more information.

## ELECTRICAL OPERATING CHARACTERISTICS

IXD9205/06 A series,  $Ta = 25^{\circ}C$ 

PARA	METER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNIT	CIRCUIT
Operating Voltag	ge Range	V <sub>IN</sub>		2.0	-	6.0	V	0
Output Voltage		V <sub>OUT</sub>	V <sub>IN</sub> = V <sub>CE</sub> = 5.0 V, I <sub>OUT</sub> = 30 mA	E-1	E-2	E-3	V	0
Maximum Outpu	ut Current	I <sub>OUT_MAX</sub>	$V_{IN} = V_{OUT(E)} + 2.0 \text{ V}, V_{CE} = 1.0 \text{ V}^{8)}$	600			mA	0
UVLO Voltage		V <sub>UVLO</sub>	$V_{CE} = V_{IN}, V_{OUT} = 0^{1), 10}$	1.00	1.40	1.78	V	3
Supply Current	IXD9205		V <sub>IN</sub> = V <sub>CE</sub> = 5.0 V, V <sub>OUT</sub> = V <sub>OUT(E)</sub> x 1.1 V		46	65		
Supply Current	IXD9206	Ι <sub>Q</sub>	VIN - VCE - 5.0 V, VOUT - VOUT(E) X 1.1 V		21	35	μA	2
Standby Current	t	I <sub>STB</sub>	$V_{IN}$ = 5.0 V, $V_{CE}$ = 0 V, $V_{OUT}$ = $V_{OUT(E)}$ x 1.1 V		0	1.0	μA	2
Oscillation Frequ	uency	f <sub>osc</sub>	$V_{IN} = V_{OUT(E)} + 2 V, V_{CE} = 1.0 V, I_{OUT} = 100 \text{ mA}$	2550	3000	3450	kHz	0
PFM Switching	Current	I <sub>PFM</sub> <sup>11)</sup>	$V_{IN} = V_{CE} = V_{OUT(E)} + 2 V$ , , $I_{OUT} = 1 \text{ mA}$	E-4	E-5	E-6	mA	9
P-channel ON ti	me maximum	t <sub>PON_MAX</sub> <sup>11)</sup>	$V_{IN} = V_{CE} = V_{OUT(T)} + 1 V$ , $I_{OUT} = 1 mA$		$2D_{\text{max}}$	$3D_{\text{MAX}}$		0
Maximum Duty	Cycle Ratio	D <sub>MAX</sub>	$V_{IN}$ = $V_{CE}$ = 5.0 V, $V_{OUT}$ = $V_{OUT(E)}$ x 0.9 V	100			%	3
Minimum Duty C	Cycle Ratio	D <sub>MIN</sub>	V <sub>IN</sub> = V <sub>CE</sub> = 5.0 V, V <sub>OUT</sub> = V <sub>OUT(E)</sub> x 1.1 V			0	%	3
Efficiency 2)		EFFI V <sub>IN</sub> = V <sub>CE</sub> = V <sub>OUT(E)</sub> + 1.2 V, I <sub>OUT</sub> = 100 mA			E-7		%	0
L <sub>x</sub> "H" ON Resis	"H" ON Resistance 1 <sup>3)</sup> $R_{LXH1}$ $V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = 0 \text{ V}, I_{LX} = 100 \text{ mA}$			0.35	0.55	Ω	4	
L <sub>x</sub> "H" ON Resis	tance 2 <sup>3)</sup>	R <sub>LXH2</sub>	$V_{IN} = V_{CE} = 3.6 \text{ V}, V_{OUT} = 0 \text{ V}, I_{LX} = 100 \text{ mA}$		0.42	0.67	Ω	4
L <sub>x</sub> "L" ON Resist	tance 1 <sup>4)</sup>	R <sub>LXL1</sub>	$V_{IN} = V_{CE} = 5.0 V$		0.45	0.65	Ω	
L <sub>x</sub> "L" ON Resist	tance 24)	R <sub>LXL2</sub>	$V_{IN} = V_{CE} = 3.6 V$		0.52	0.77	Ω	
L <sub>x</sub> "H" Leakage	Current <sup>5)</sup>	I <sub>LXH</sub>	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = 0 \text{ V}, V_{LX} = 5.0 \text{ V}$		0.01	1.0	μA	5
L <sub>x</sub> "L" Leakage (	Current <sup>5), 15)</sup>	I <sub>LXH</sub>	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = 0 \text{ V}, V_{LX} = 5.0 \text{ V}$		0.01	1.0	μA	5
Current Limit <sup>10)</sup>		I <sub>LIM</sub>	$V_{IN} = V_{CE} = 5.0 \text{ V}, V_{OUT} = V_{OUT(E)} \times 0.9 \text{ V}^{7)}$	900	1050	1350	mA	6
Output Voltage - Characteristics	Temperature	$\frac{\Delta V_{OUT}}{V_{OUT} * \Delta T_{OPR}}$	$-40^{\circ}C \le T_{OPR} \le 85^{\circ}C$ , $I_{OUT} = 30 \text{ mA}$		±100		ppm/ <sup>0</sup> C	0
CE "H" Voltage <sup>1</sup>	3)	V <sub>CEH</sub>	V <sub>OUT</sub> = 0 V			6.0	V	3
CE "L" Voltage <sup>14</sup>	4)	V <sub>CEL</sub>	V <sub>OUT</sub> = 0 V			0.25	V	3
CE "H" Current		I <sub>ENH</sub>	V <sub>IN</sub> = V <sub>CE</sub> = 5.0 V, V <sub>OUT</sub> = 0 V			0.1	μA	5
CE "L" Current		I <sub>ENL</sub>	V <sub>IN</sub> = 5.0 V, V <sub>CE</sub> = 0 V, V <sub>OUT</sub> = 0 V	-0.1		0.1	μA	5
Soft-Start Time	tart Time A, C version tss lour = 1 mA		0.5	0.9	2.5	ms	0	
Latch Time <sup>6)</sup>	B, G versions	t <sub>LAT</sub>	$V_{IN}$ = $V_{CE}$ = 5.0 V, $V_{OUT}$ = 0.8 x $V_{OUT(E)}$ , $L_x$ short with 1 $\Omega$ resistor to ground	1.0	E-11	E-12 20.0	ms	Ø

When the difference between input and output voltage is

current maximizes. If load current increases in this state,

output voltage will decrease because of the voltage drop

Current limit denotes the level of an inductor peak current

10) Voltage, when  $L_X$  pin voltage is "L"=+0.1 V ~ -0.1 V

12)  $V_{OUT}$  voltage at which L<sub>x</sub> pin state changes from "H" to "L"

small, some cycles may be skipped completely before

# ELECTRICAL OPERATING CHARACTERISTICS (CONTINUED)

PARAMETER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNIT	CIRCUIT
Short Protection Threshold Voltage	V <sub>SHORT</sub>	$V_{IN} = V_{CE} = 5.0 \text{ V}, L_X \text{ short with } 1 \Omega \text{ resistor to}$ ground <sup>12</sup>	E-8	E-9	E-10	V	Ø
C <sub>L</sub> Discharge Resistance <sup>13)</sup>	R <sub>CLD</sub>	$V_{IN} = L_X = 5 V$ , $V_{CE} = 0$ , $V_{OUT} = open$	200	300	450	Ω	8
Inductance Value	L	Test frequency 1 MHz		1.5		μH	
Inductor Current Maximum	I <sub>LMAX</sub>	$\Delta T = 40^{\circ}C$		1000		mA	

8)

9)

11)

#### NOTE:

Test conditions: Unless otherwise stated,  $V_{IN}$  = 5.0 V,  $V_{OUT (E)}$  = Nominal Voltage

- 1) Including hysteresis operating voltage range
- EFFI = {(output voltage × output current) / (input voltage × input current)} × 100%
- 3) ON resistance ( $\Omega$ ) = (V<sub>IN</sub> Lx pin measurement voltage) / 100mA
- 4) Design target value
- 5) A 10µA (maximum) current may leak at high temperature
- When V<sub>IN</sub> is less than 2.4 V, current limit may not be reached because of voltage drop across ON resistance

## E-TABLES

#### Output Voltage Error

NOMINAL OUTPUT		V <sub>оит,</sub> (V)	
VOLTAGE	E-1	E-2	E-3
V <sub>OUT(T)</sub> , V	MIN	TYP	MAX
1.00	0.980	1.000	1.020
1.20	1.176	1.200	1.224
1.40	1.372	1.400	1.428
1.50	1.470	1.500	1.530
1.75	1.715	1.750	1.785
1.80	1.764	1.800	1.836
1.90	1.862	1.900	1.938
2.50	2.450	2.500	2.550
2.80	2.744	2.800	2.856
2.85	2.793	2.850	2.907
3.00	2.940	3.000	3.060
3.30	3.234	3.300	3.366

## PFM Switching Current

NOMINAL OUTPUT	I <sub>PFM</sub> , (mA)				
VOLTAGE	E-4	E-5	E-6		
VOLTAGE	MIN	MAX	TYP		
V <sub>OUT(T)</sub> < 1.2V	190	260	350		
1.2 V < V <sub>OUT(T)</sub> ≤1.75 V	180	240	300		
1.8 V< V <sub>OUT(T)</sub>	170	220	270		

across P-channel transistor

IXD9206 series only

13) B, C, and G versions only

within 1 ms

14) Version A only

## Efficiency

## Short Circuit Protection Threshold Voltage

,	NOMINAL			V <sub>SHOR</sub>	<sub>it</sub> , (V)		
	OUTPUT	А,	<b>B, C VERSION</b>	1S		G VERSION	
	VOLTAGE	E-8	E-9	E-10	E-8	E-9	E-10
	V <sub>OUT(T)</sub> , V	MIN	TYP	MAX	MIN	TYP	MAX
	1.00	0.375	0.500	0.625	0.188	0.250	0.313
	1.20	0.450	0.600	0.750	0.225	0.300	0.375
	1.40	0.525	0.700	0.875	0.263	0.350	0.438
	1.50	0.563	0.750	0.938	0.282	0.375	0.469
	1.75	0.656	0.875	1.094	0.328	0.438	0.547
	1.80	0.675	0.900	1.125	0.338	0.450	0.563
	1.90	0.713	0.950	1.188	0.357	0.475	0.594
	2.50	0.938	1.250	1.563	0.469	0.625	0.782
	2.80	1.050	1.400	1.750	0.525	0.700	0.875
	2.85	1.069	1.425	1.781	0.535	0.713	0.891
	3.00	1.125	1.500	1.875	0.563	0.750	0.938
	3.30	1.238	1.650	2.063	0.619	0.825	1.032

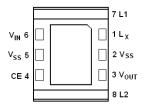
NOMINAL OUTPUT	EFFICIENCY, %
VOLTAGE	E-7
V <sub>OUT(T)</sub> , V	
1.00	79
1.20	82
1.40	83
1.50	84
1.75 – 1.90	85
2.50 - 3.30	86



#### Soft-Start Time (IXD9205/06 B and G versions only)

NOMINAL OUTPUT	t <sub>ss</sub> , (ms)				
VOLTAGE	E-11	E-12			
V <sub>OUT(T)</sub> , V	TYP	MAX			
0.8 V < V <sub>OUT(T)</sub> ≤1.75 V	0.25	0.4			
1.8 V < V <sub>OUT(T)</sub> ≤4.0 V	0.32	0.5			

## **PIN CONFIGURATION**



#### NOTE:

The dissipation pad should be soldered in recommended mount pattern and metal masking to enhance mounting strength and heat release. If the pad needs to be connected to other pins, it should be connected to the  $V_{\rm SS}$  (No 2 and No 5) pins.

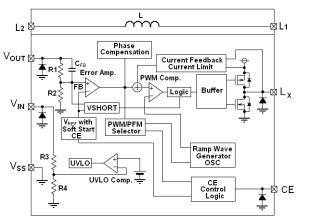
Vss pins (No. 2 and 5) should be tied together and connected to ground plane on the board.

## PIN ASSIGNMENT

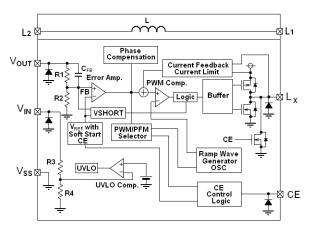
PIN NUMBER	PIN NAME	FUNCTIONS			
1	Lx	Switching Node			
2, 5	Vss	Ground			
3	V <sub>OUT</sub>	Fixed Output Voltage			
4	CE	Enable (Active HIGH)			
6	V <sub>IN</sub>	Power Input			
7	L1	Inductor Connection			
8	L2	Inductor Connection			

## **BLOCK DIAGRAMS**

IXD9205/06 A Series



# IXD9205/06 B/C/G Series



Internal diodes include an ESD protection and a parasitic diode

# **BASIC OPERATION**

The IXD9205/06 series consists of a Reference Voltage source, Ramp Wave Generator, Error Amplifier, PWM Comparator, Phase Compensation circuit, output voltage resistive divider, P-channel switching transistor, N-channel transistor for the synchronous switch, Current Limiter circuit, UVLO circuit, and other features. (See the block diagrams above.)

The Error Amplifier compares output voltage divided by internal resistors  $R_1/R_2$  with the internal reference voltage. Amplified difference between these two signals applies to the one input of the PWM Comparator, while ramp voltage from the Ramp Wave Generator applies to the second input. The resulting PWM pulse determines switching transistor ON time. It goes through the buffer and it appears at the gate of the internal P-channel switching transistor. This continuous process stabilizes output voltage.



The Current Feedback circuit monitors the current of the P-channel transistor at each switching cycle, and modulates output signal from the Error Amplifier to provide additional feedback. This guarantees a stable converter operation even with low ESR ceramic load capacitor.

#### **Reference Voltage Source**

The Reference Voltage Source provides the reference voltage to ensure stable output voltage of the DC/DC converter.

#### Ramp Wave Generator

The Ramp Wave Generator produces ramp waveform signal needed for PWM operation, and signals to synchronize all the internal circuits. It operates at an internally fixed 1.2 MHz or 3.0 MHz frequency.

#### **Error Amplifier**

The Error Amplifier monitors output voltage through a resistive divider connected to  $V_{OUT}$  (FB) pin. If output voltage falls below preset value and the Error Amplifier's input signal becomes less than the internal reference voltage, the Error Amplifier/s output signal increases. This results in a wider PWM pulse and respectively longer ON time for the switching transistor to increase output voltage. The gain and frequency characteristics of the error amplifier output are fixed internally to optimize IC performance.

#### Current Limiter

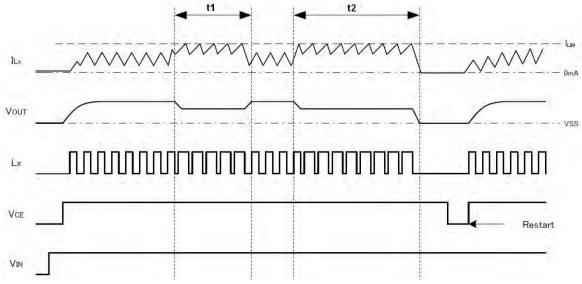
The Current Limiter circuit monitors the current flowing through the P-channel transistor connected to the Lx pin, and combines the function of the current limit and operation suspension.

When the transistor's current is greater than a specified level, the Current Limiter turns off the P-channel transistor immediately. After that, the Current Limiter turns off too, returning to monitoring mode.

The driver transistor turns on at the next cycle, but the Current Limiter turns it off immediately if an over- current exists. When the over current state is eliminated, the IC resumes its normal operation.

The IC waits for the end of the over current state, repeating the steps above (t1 in the figure below). If an overcurrent state continues for a few ms with IC repeatedly performing these steps, the Current Limiter latches the Pchannel transistor in OFF state, and the IC suspends operations (t2 in the figure below). To restart IC operation after this condition, either the CE pin should be toggled H - L - H, or the V<sub>IN</sub> pin voltage should be set below UVLO to resume operations from soft start.

The suspension mode is not a standby mode. In the suspension mode, pulse output is suspended; however, internal circuitries remain in operation mode, consuming power.



#### **Short-Circuit Protection**

Short-circuit protection monitors the  $R_{FB1}/R_{FB2}$  divider voltage (FB point in the block diagram). If output is accidentally shorted to the ground, FB voltage starts falling. When this voltage becomes less than half of the reference voltage ( $V_{REF}$ ) and the P-channel switching transistor's current is more than the  $I_{LIM}$  threshold, Short-Circuit Protection turns off and quickly latches the P-channel transistor.



At D/E/F/G series, Short Circuit Protection starts once FB voltage becomes less than 0.25 of reference voltage ( $V_{REF}$ ), irrespective of the transistor's current.

To restart IC operation after this condition, either the EN pin should be toggled H - L - H, or the V<sub>IN</sub> pin voltage should be set below UVLO to resume operations from soft start.

The sharp load transients creating a voltage drop at the  $V_{OUT}$  propagate to the FB point through  $C_{FB}$ ; that may result in Short Circuit protection operating at voltages higher than 1/2  $V_{REF}$  voltage.

## **UVLO Circuit**

When the  $V_{IN}$  pin voltage becomes 1.4V or lower, the P-channel transistor is forced OFF to prevent false pulse output caused by unstable operation of the internal circuitry. When the  $V_{IN}$  pin voltage becomes 1.8 V or higher, switching operations resume with the soft start. The soft start function operates even when the  $V_{IN}$  voltage falls below the UVLO threshold for a very short time. The UVLO circuit does not cause a complete shutdown of the IC, but causes pulse output to be suspended; therefore, the internal circuitry remains in operation.

## **PFM Switch Current**

In PFM mode, the IC keeps the P-channel transistor on until inductor current reaches a specified level (I<sub>PFM</sub>).

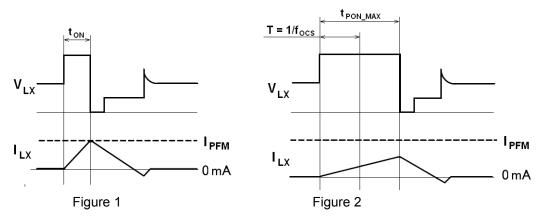
P-channel transistor's ON time is equal

$$t_{ON} = L \times I_{PFM} / (V_{IN} - V_{OUT}), \mu s,$$

where L is an inductance in  $\mu H$ , and  $I_{PFM}$  is a current limit in A.

#### PFM Duty Limit

In PFM mode, P-channel ON time maximum ( $t_{PON_MAX}$ ) is set to  $2D_{MAX}$ , i.e. two periods of the switching frequency. Therefore, under conditions when the ON time increases (i.e. step-down ratio is small), it is possible that the P-channel transistor is turned off, even when the inductor current does not reach  $I_{PFM}$ . (See Figures 1 and 2 below)



## C<sub>L</sub> High Speed Discharge

The IXD9205/06B, C, and G series can quickly discharge the output capacitor ( $C_L$ ) to avoid application malfunction, when the CE pin set logic LOW to disable IC.

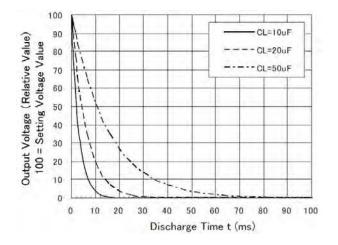
 $C_L$  Discharge Time is proportional to the resistance (R) of the N-channel transistor located between the  $L_X$  pin and ground and the output  $C_L$  capacitance, as shown below.

$$t_{DSH} = RC_L x Ln (V_{OUT(E)} / V)$$
, where

 $\label{eq:V-Output} \begin{array}{l} V \mbox{- Output voltage after discharge} \\ V_{\text{OUT(E)}} \mbox{- Output voltage} \\ R \mbox{= } 300 \ \Omega \ (Typical value) \end{array}$ 



### **Output Voltage Discharge Characteristics**



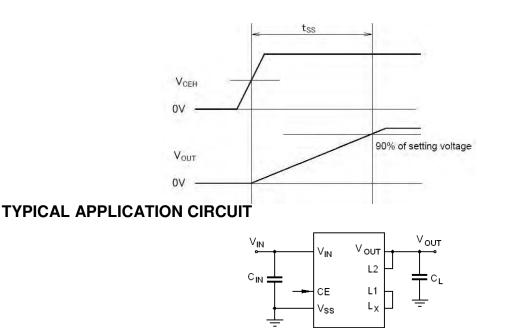
#### **CE Pin Function**

The IXD9205/06 series enters shut down mode when a LOW logic-level signal applies to the CE pin. In the shutdown mode, IC current consumption is ~0  $\mu$ A (Typical value), with the Lx and V<sub>OUT</sub> pins at high impedance state. The IC starts its operation when a HIGH logic-level signal applies to the CE.

## Soft Start

Soft start time is available in two options via product selection.

The soft-start time of the IXD9205/06 series is optimized by using internal circuits. The definition of the soft-start time is the time when the output voltage goes up to the 90% of nominal output voltage after the IC is enabled by CE "H" signal.



#### **EXTERNAL COMPONENTS**

f <sub>osc</sub>	3.0 MHz
C <sub>IN</sub> , μF	4.7
C <sub>L</sub> , µF	10

Capacitors should be X7R or X5R series to minimize power losses.



# Product Specification IXD9205/IXD9206

## LAYOUT AND USE CONSIDERATIONS

- 1. Wire external components as close to the IC as possible and use thick, short connecting traces to reduce the circuit impedance. Pay special attention to the  $V_{IN}$  and GND wiring. Switching noise, which occurs from the GND, may cause the instability of the IC, so, position  $V_{IN}$  and  $V_{CL}$  capacitors as close to IC as possible (See recommended layout on the right).
- Transitional voltage drops or voltage rising phenomenon could make the IC unstable if ratings are exceeded.
- 3. The IXD9205/06 series is designed to work with ceramic output capacitors. However, if the difference between input and output voltages is too high, a ceramic capacitor may fail to absorb the resulting high switching energy and oscillation could occur. In this case, connect an electrolytic capacitor in parallel to ceramic one to compensate for insufficient capacitance.
- 4. In PWM mode, IC generates very narrow pulses, and there is a possibility that some cycles will be skipped completely, if the difference between V<sub>IN</sub> and V<sub>OUT</sub> is high.
- 5. If the difference between V<sub>IN</sub> and V<sub>OUT</sub> is small, IC generates very wide pulses, and there is a possibility that some cycles will be skipped completely at the heavy load current.
- 6. When dropout voltage or load current is high, Current Limit may activate prematurely that will lead to IC instability. To avoid this condition, choose inductor's value to set peak current below Current Limit threshold. Calculate the peak current according to the following formula:

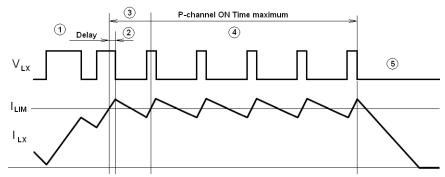
$$I_{PK} = (V_{IN} - V_{OUT}) \times D / (2 \times L \times f_{OSC}) + I_{OUT}$$
, where

L - Inductance

fosc -- Oscillation Frequency

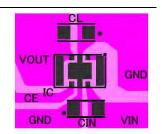
D - Duty cycle

7. Inductor's rated current exceeds Current Limit threshold of 1050 mA to avoid damage, which may occur until P-channel transistor turns off after Current Limiter activates, but pulse current may exceed this



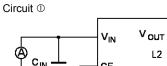
value(see figure below).

- $\odot$  Current flows into P-channel transistor reaches the current limit (I\_{LIM}).
- <sup>®</sup> Current is more than I<sub>LIM</sub> due the circuit's delay time from the current limit detection to the P-channel transistor OFF.
- $\ensuremath{\textcircled{}}$  ] The inductor's current time rate becomes quite small.
- IC generates very narrow pulses for several milliseconds.
- ⑤ The circuit latches, stopping operation.
- 8. If V<sub>IN</sub> voltage is less than 2.4 V, current limit threshold may not be reached due to voltage drop caused by the switching transistor's ON resistance.
- 9. Latch time may become longer or latch may not work due electrical noise. To avoid this effect, the board should be laid out so that input capacitors are placed as close to the IC as possible.
- 10. Use of the IC at voltages below recommended voltage range may lead to instability.
- 11. At high temperature, output voltage may increase up to input voltage level at no load, because of the leakage current of the driver transistor.
- 12. High step-down ratio and very light load may be cause of intermittent oscillations in PWM mode.
- 13. In PWM/PFM automatic switching mode, IC may become unstable during transition to continuous mode. Please verify with actual components.

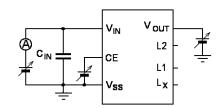




## **TEST CIRCUITS**

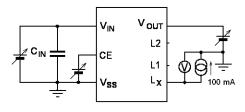


Circuit 2

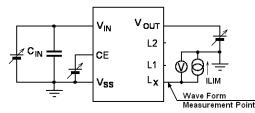


Circuit ④

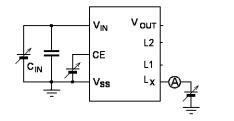
Circuit



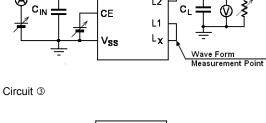
ON Resistance =  $(V_{IN} - V_{OUT}/100 \text{ mA})$ 



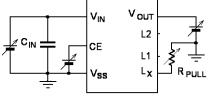




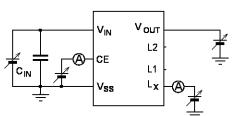
Circuit



Vout

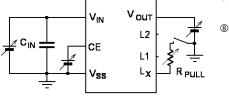


Circuit (5)



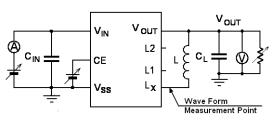


6





Circuit (9)

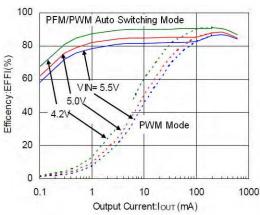


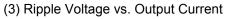
**External Components**  $C_{IN} = 4.7 \ \mu F$  (ceramic),  $C_L = 10 \ \mu F$  (ceramic) L = 1.5 μH R<sub>PULL</sub> = 200 Ω

# **TYPICAL PERFORMANCE CHARACTERISTICS**

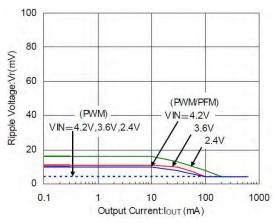
## (1) Efficiency vs. Output Current

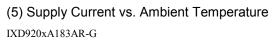
IXD920xA183AR-G  $C_{IN}$  = 4.7 µF,  $C_L$ = 10 µF

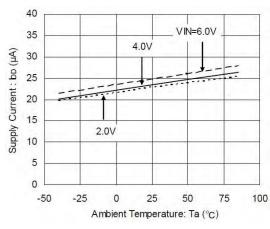




IXD920xA183AR-G

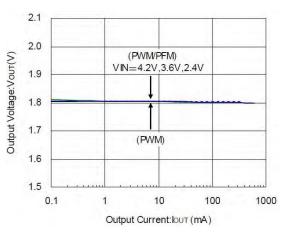






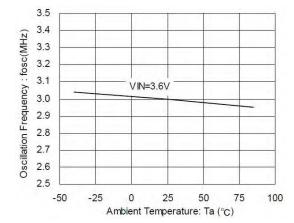
# (2) Output Voltage vs. Output Current

IXD920xA183AR-G

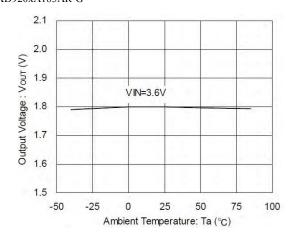


(4) Oscillation Frequency vs. Ambient Temperature

IXD920xA183AR-G



(6) Output Voltage vs. Ambient Temperature IXD920xA183AR-G

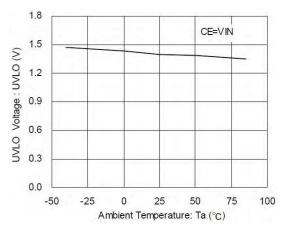


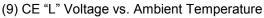


# **TYPICAL PERFORMANCE CHARACTERISTICS (Continued)**

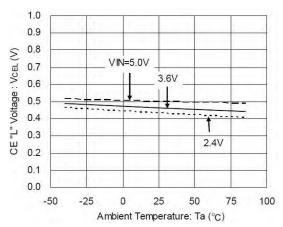
(7) UVLO Voltage vs. Ambient Temperature

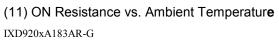
IXD920xA183AR-G

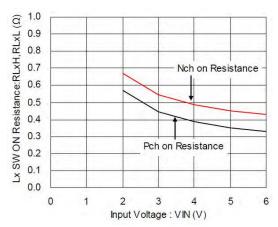




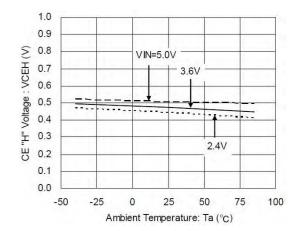
IXD920xA183AR-G

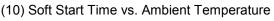




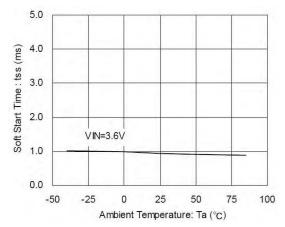


(8) CE "H" Voltage vs. Ambient Temparature IXD920xA183AR-G



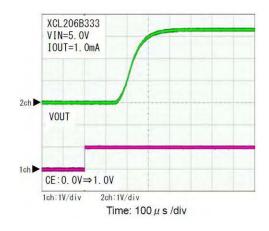


IXD920xA183AR-G



(12) Start Wave Form

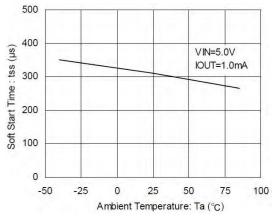
IXD920xB333AR-G



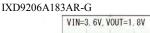


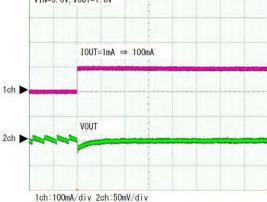
## (13) Soft Start Time vs. Ambient Temperature

#### IXD920xB333AR-G

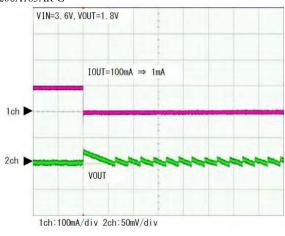


#### (15) Load Transient Response MODE: PWM/PFM Auto Switching



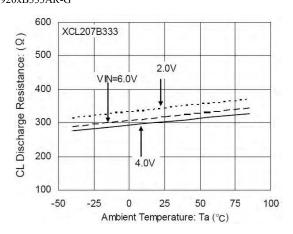


 $Ch1 - I_{\text{OUT}}, \, Ch2 - V_{\text{OUT}} \, \text{50 mV/div}, \, \text{Time} - 100 \, \mu\text{s/div} \, IXD9206A183AR-G$ 

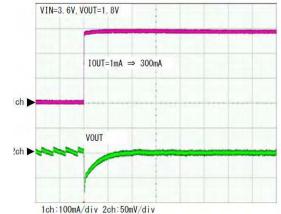


 $Ch1 - I_{OUT}, \, Ch2 - V_{OUT} \, 50 \, \, mV/div, \, Time - 100 \, \mu s/div$ 

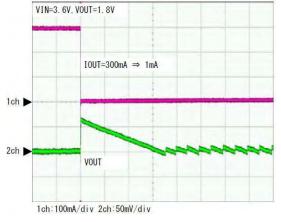
## (14) C<sub>L</sub> Discharge Time vs. Ambient Temperature IXD920xB333AR-G







 $\label{eq:ch1_lout_ch2} \begin{array}{c} Ch1-I_{OUT}, Ch2-V_{OUT} \text{ 50 mV/div}, \text{ Time}-100 \ \mu\text{s/div}\\ IXD9206A183AR-G \end{array}$ 



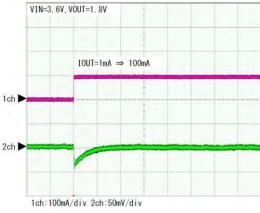
Ch1 –  $I_{\text{OUT}},\,Ch2$  –  $V_{\text{OUT}}$  50 mV/div, Time – 100  $\mu s/div$ 



# TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

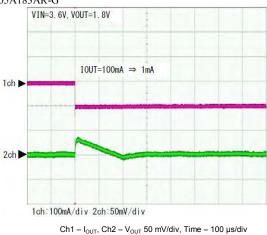
(15) Load Transient Response (Continued) MODE: PWM

#### IXD9205A183AR-G



Ch1 – I<sub>OUT</sub>, Ch2 – V<sub>OUT</sub> 50 mV/div, Time – 100 µs/div

IXD9205A183AR-G

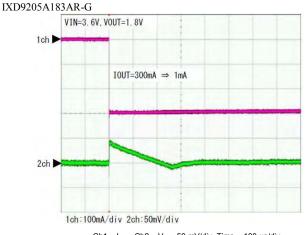


## ORDERING INFORMATION

IXD92050@3@56-⑦ PWM Mode only IXD92060@3@56-⑦ PFM/PWM Mode auto switching

	VIN=3. 6V, V	OUT=1.8V		
		IOUT=1mA	⇒ 300mA	
1ch 🕨	-		-	 
2ch ►		-		

Ch1 –  $I_{\text{OUT}},$  Ch2 –  $V_{\text{OUT}}$  50 mV/div, Time – 100  $\mu s/div$ 



Ch1 –  $I_{\text{OUT}},$  Ch2 –  $V_{\text{OUT}}$  50 mV/div, Time – 100  $\mu s/div$ 

DESIGNATOR	DESCRIPTION	SYMBOL	DESCRIPTION	
		A	$V_{IN} \ge 2 V$ , No C <sub>L</sub> auto discharge, standard soft start	
0	Type of DC/DC Controller	В	$V_{IN} \ge 2 V, C_{L}$ auto discharge, fast soft start	
U		С	$V_{IN} \ge 2 V$ , $C_L$ auto discharge, standard soft start	
		G	$V_{IN} \ge 1.8 \text{ V}, \text{ C}_{L}$ auto discharge, fast soft start	
			② - integer part, ③ - decimal part, i.e.	
			$V_{OUT} = 2.8 \text{ V} - 2 = 2, 3 = 8$	
23	Fixed Output Voltage, V <sup>2)</sup>	Fixed Output Voltage, V <sup>2)</sup>	08 - 40	0.05 V increments: 0.05 = A, 0.15 = B, 0.25 = C. 0.35 = D, 0.45 = E, 0.55 = F,
			0.65 = H, 0.75 = K, 0.85 = L, 0.95 = M	
			V <sub>OUT</sub> = 2.85 V - ② = 2, ③ = L	
4	Oscillation Frequency	3	3.0 MHz	
\$6-Ø <sup>1)</sup>	Packages (Order Limit)	AR-G	CL-2025 (3000/reel)	

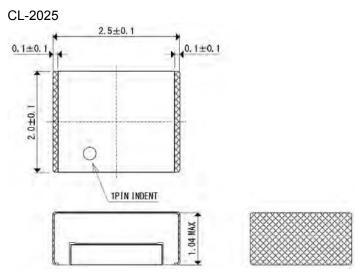
#### NOTE:

- 1) The "-G" suffix denotes halogen and antimony free, as well as being fully RoHS compliant.
- 2) Standard output voltages are: 1.0, 1.2, 1.4, 1.5, 1.75, 1.8, 1.9, 2.5, 2.8, 2.85, 3.0, and 3.3 V. Contact local representative for more information if other voltages in the range from 0.8 to 4.0 V require.

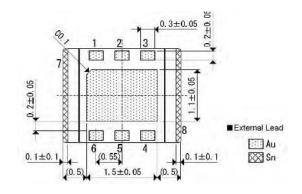


# PACKAGE DRAWING AND DIMENSIONS

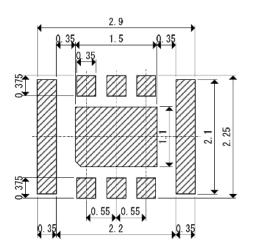
(Units: mm)



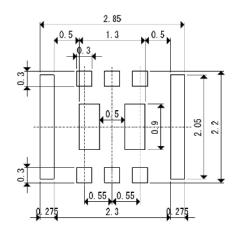
Bottom View



Reference Pattern Layout



Reference Metal Mask Design





## PACKAGE POWER DISSIPATION

## **CL-2025 Power Dissipation**

The power dissipation varies with the mount board conditions. Please use this data as a reference only.

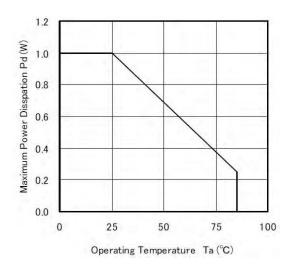
1. Measurement Conditions:

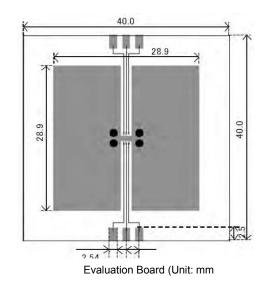
Condition: Ambient: Soldering:	Mount on a board Natural convection Lead (Pb) free
Board:	Dimensions 40×40 mm (1600 mm <sup>2</sup> in one side) Copper (Cu) traces occupy 50% of the board area on top and bottom layers Package heat sink tied to the copper traces.
Material:	Glass Epoxy (FR-4)
Thickness:	1.6 mm
Through-hole:	4 x 0.8 Diameter

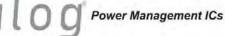
2. Power Dissipation vs. Ambient Temperature

Board Mount (Tjmax = 125 <sup>0</sup>C)

Ambient Temperature, ⁰C	Power Dissipation Pd, mW	Thermal Resistance, <sup>°</sup> C/W
25	1000	80
85	250	80





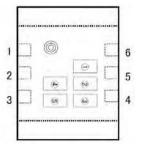


An IXYS Company

# MARKING

CL-2025

## ① Represents product series



MARK	PRODUCT SERIES
4	IXD9205Axxxxx-G
С	IXD9205Bxxxxx-G
	IXD9205Gxxxxx-G
K	IXD9205Cxxxxx-G
5	IXD9206Axxxxx-G
D	IXD9206Bxxxxx-G
	IXD9206Gxxxxx-G
L	IXD9206Cxxxxx-G

2 Represents integral part of the voltage value

OUTPUT VOLTAGE,	MARK	
V	IXD920xA/B/C	IXD920xG
0.x	F	U
1.x	Н	V
2.x	К	Х
3.x	L	Y
4.x	М	Z

③ Represents decimal part of the Voltage value

V <sub>out</sub> , V	MARK
x.00	0
x.05	A
x.10	1
x.15	В
x.20	2
x.25	2 C
x.30	3
x.35	D
x.40	4
x.45	E
x.50	5
x.55	F
x.60	6
x.65	Н
x.70	7
x.75	K
x.80	8
x.85	L
x.90	9
x.95	М

@ S Represents production lot number 01~09, 0A~0Z, 11~9Z, A1~A9, AA~AZ, B1~ZZ in order (G, I, J, O, Q, and W excluded)



# **Customer Support**

To share comments, get your technical questions answered, or report issues you may be experiencing with our products, please visit Zilog's Technical Support page at <a href="http://support.zilog.com">http://support.zilog.com</a>. To learn more about this product, find additional documentation, or to discover other fac-ets about Zilog product offerings, please visit the Zilog Knowledge Base at <a href="http://zilog.com/kb">http://zilog.com/kb</a> or consider participating in the Zilog Forum at <a href="http://zilog.com/kb">http://zilog.com/kb</a> or consider participating in the Zilog Forum at <a href="http://zilog.com/kb">http://zilog.com/kb</a> or consider participating in the Zilog Forum at <a href="http://zilog.com/kb">http://zilog.com/kb</a> or consider participating in the Zilog Forum at <a href="http://zilog.com/kb">http://zilog.com/kb</a> or consider participating in the Zilog Forum at <a href="http://zilog.com/kb">http://zilog.com/kb</a> or consider participating in the Zilog Forum at <a href="http://zilog.com/kb">http://zilog.com/kb</a> or consider participating in the Zilog Forum at <a href="http://zilog.com/kb">http://zilog.com/kb</a> or consider participating in the Zilog Forum at <a href="http://zilog.com/kb">http://zilog.com/kb</a> or consider participating in the Zilog Forum at <a href="http://zilog.com/kb">http://zilog.com/kb</a> or consider participating in the Zilog website at <a href="http://www.zilog.com">http://zilog.com</a>.

## Warning: DO NOT USE THIS PRODUCT IN LIFE SUPPORT SYSTEMS.

**LIFE SUPPORT POLICY** ZILOG'S PRODUCTS ARE NOT AUTHORIZED FOR USE AS CRITICAL COMPONENTS IN LIFE SUPPORT DEVICES OR SYSTEMS WITHOUT THE EXPRESS PRIOR WRITTEN APPROVAL OF THE PRESIDENT AND GENERAL COUNSEL OF ZILOG CORPORATION.

**As used herein** Life support devices or systems are devices which (a) are intended for surgical implant into the body, or (b) support or sustain life and whose failure to perform when properly used in accordance with instructions for use provided in the labeling can be reasonably expected to result in a significant injury to the user. A critical component is any component in a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system or to affect its safety or effectiveness.

**Document Disclaimer** ©2015 Zilog, Inc. All rights reserved. Information in this publication concerning the devices, applications, or technology described is intended to suggest possible uses and may be superseded. ZILOG, INC. DOES NOT ASSUME LIABILITY FOR OR PROVIDE A REPRESENTATION OF ACCURACY OF THE INFORMATION, DEVICES, OR TECHNOLOGY DESCRIBED IN THIS DOCUMENT. ZILOG ALSO DOES NOT ASSUME LIABILITY FOR INTELLECTUAL PROPERTY INFRINGEMENT RELATED IN ANY MANNER TO USE OF INFORMATION, DEVICES, OR TECHNOLOGY DESCRIBED HEREIN OR OTHERWISE. The information contained within this document has been verified according to the general principles of electrical and mechanical engineering.