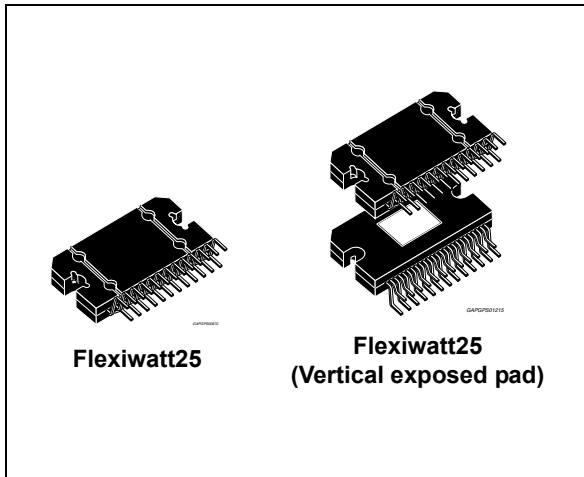


**4 x 41 W quad bridge car radio amplifier**

Datasheet - production data



- Low external component count:
  - Internally fixed gain (26 dB)
  - No external compensation
  - No bootstrap capacitors
- Protections:
  - Output short circuit to GND, to  $V_S$ , across the load
  - Very inductive loads
  - Overtating chip temperature with soft thermal limiter
  - Load dump voltage
  - Fortuitous open GND
  - Reversed battery
  - ESD

**Features**

- High output power capability:
  - 4 x 41 W / 4  $\Omega$  max.
- Low distortion
- Low output noise
- Standby function
- Mute function
- Automute at min. supply voltage detection

**Description**

The TDA7387 is an AB class audio power amplifier, packaged in Flexiwatt 25 and designed for high end car radio applications.

Based on a fully complementary PNP/NPN configuration, the TDA7387 allows a rail to rail output voltage swing with no need of bootstrap capacitors. The extremely reduced boundary components count allows very compact sets.

**Table 1. Device summary**

Order code	Package	Packing
TDA7387	Flexiwatt25	Tube
TDA7387EP	Flexiwatt25 vertical, exposed pad	Tube
TDA7387EPAG	Flexiwatt25 vertical, exposed pad	Tube

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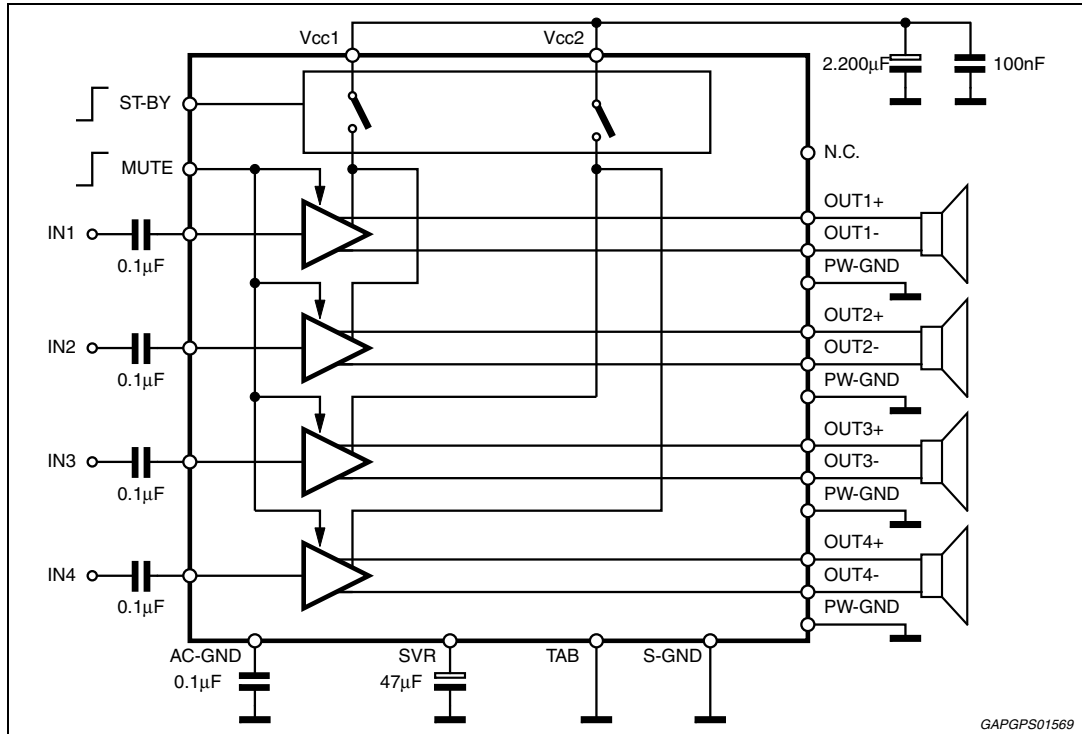
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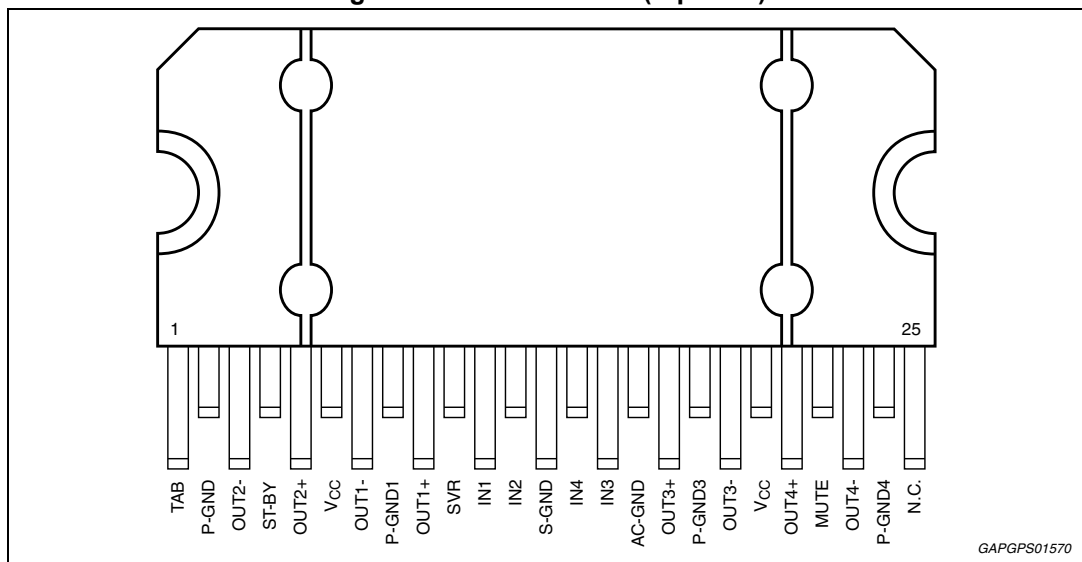
# 1 Block and pin connection diagrams

Figure 1. Block diagram



GAPGPS01569

Figure 2. Pin connection (top view)



GAPGPS01570

## 2 Electrical specifications

### 2.1 Absolute maximum ratings

Table 2. Absolute maximum ratings

Symbol	Parameter	Value	Unit
$V_S$	Operating supply voltage	18	V
$V_{S(DC)}$	DC supply voltage	28	V
$V_{S(pk)}$	Peak supply voltage (t = 50 ms)	50	V
$I_O$	Output peak current: Repetitive (duty cycle 10 % at f = 10 Hz)	4.5	A
	Non repetitive (t = 100 $\mu$ s)	5.5	A
$P_{tot}$	Power dissipation, (T <sub>case</sub> = 70 °C)	80	W
$T_j$	Junction temperature	150	°C
$T_{stg}$	Storage temperature	- 55 to 150	°C

### 2.2 Thermal data

Table 3. Thermal data

Symbol	Parameter	Value	Unit
$R_{th\ j-case}$	Thermal resistance junction-to-case max. TDA7387	1	°C/W
	TDA7387EP/EPAG	1.3	
$T_{amb}$	Operative temperature range TDA7387EP	-30 to +85	°C
	Operative temperature range TDA7387, TDA7387EPAG	-40 to +105	°C

### 2.3 Electrical characteristics

$V_S = 14.4$  V; f = 1 kHz;  $R_g = 600$   $\Omega$ ;  $R_L = 4$   $\Omega$ ;  $T_{amb} = 25$  °C; Refer to the test and application diagram ([Figure 3](#)), unless otherwise specified.

Table 4. Electrical characteristics

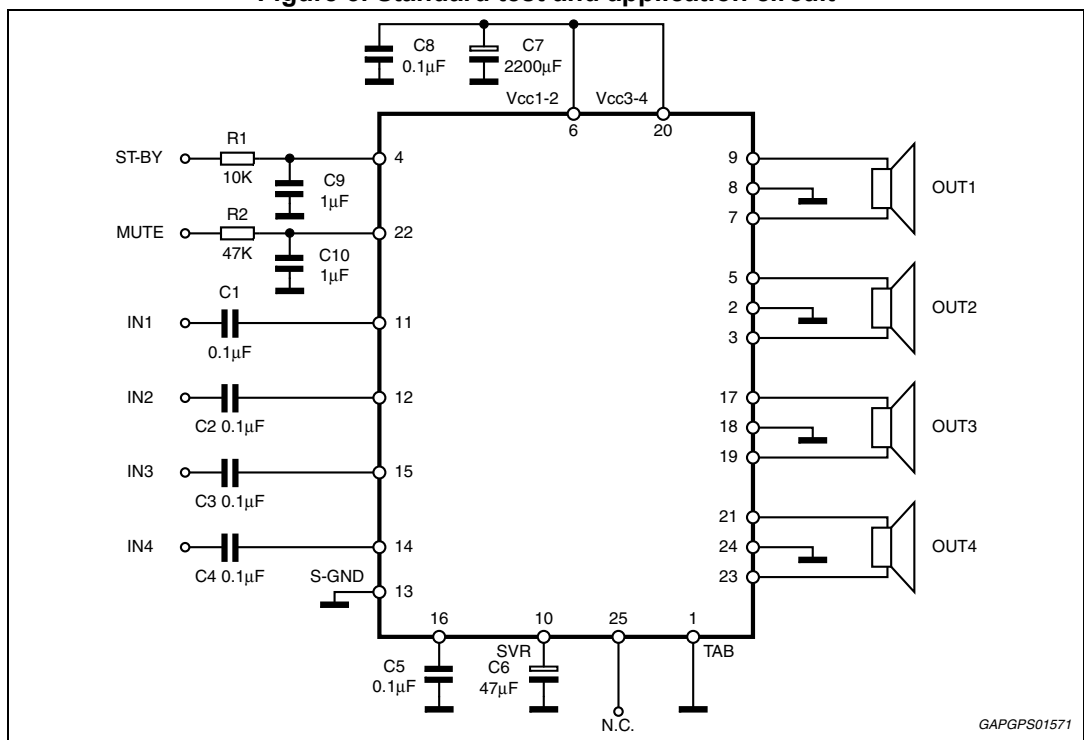
Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
$I_{q1}$	Quiescent current	-	-	180	300	mA
$V_{OS}$	Output offset voltage	-	-	-	100	mV
$G_v$	Voltage gain	-	25	26	27	dB
$P_o$	Output power	THD = 10%	20	22	-	W
		THD = 1%	-	18	-	

Table 4. Electrical characteristics (continued)

Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
P <sub>O max.</sub>	Max. output power (1)	V <sub>S</sub> = 14.4 V	33	37	-	W
		V <sub>S</sub> = 15.2 V	-	41	-	
THD	Distortion	P <sub>O</sub> = 4 W	-	0.04	0.3	%
e <sub>No</sub>	Output noise	"A" Weighted; Bw = 20 Hz to 20 kHz	-	50 65	150	μV μV
SVR	Supply voltage rejection	f = 100 Hz	50	65	-	dB
f <sub>cl</sub>	Low cut-off frequency	-	-	20	-	Hz
f <sub>ch</sub>	High cut-off frequency	-	75	-	-	kHz
R <sub>i</sub>	Input impedance	-	70	100	-	kΩ
C <sub>T</sub>	Cross talk	f = 1 kHz	50	70	-	dB
I <sub>SB</sub>	Standby current consumption	V <sub>standby</sub> = 0 V	-	-	15	μA
V <sub>SB out</sub>	Standby out threshold voltage	(Amp: on)	3.5	-	-	V
V <sub>SB in</sub>	Standby in threshold voltage	(Amp: off)	-	-	1.5	V
A <sub>M</sub>	Mute attenuation	V <sub>O</sub> = 1Vrms	80	90	-	dB
V <sub>M out</sub>	Mute out threshold voltage	(Amp: play)	3.5	-	-	V
V <sub>M in</sub>	Mute in threshold voltage	(Amp: mute)	-	-	1.5	V
I <sub>m (L)</sub>	Muting pin current	V <sub>MUTE</sub> = 1.5 V (source current)	5	10	16	μA

1. Saturated square wave output.

Figure 3. Standard test and application circuit



## 2.4 PCB and component layout

Refer to [Figure 3: Standard test and application circuit](#).

Figure 4. Components and top copper layer

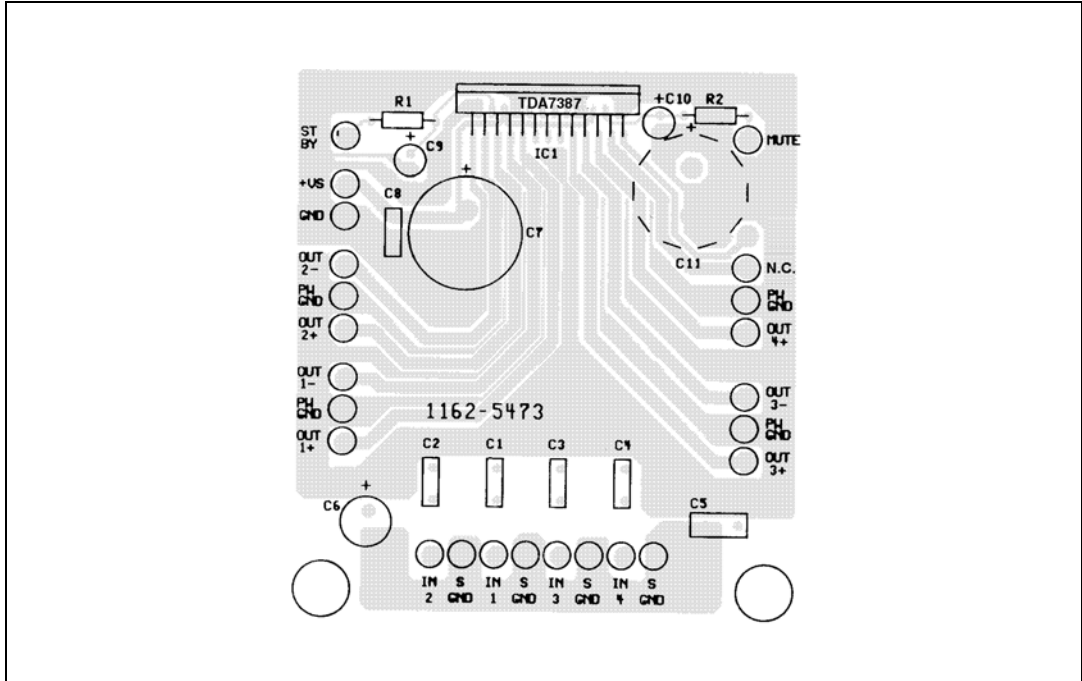
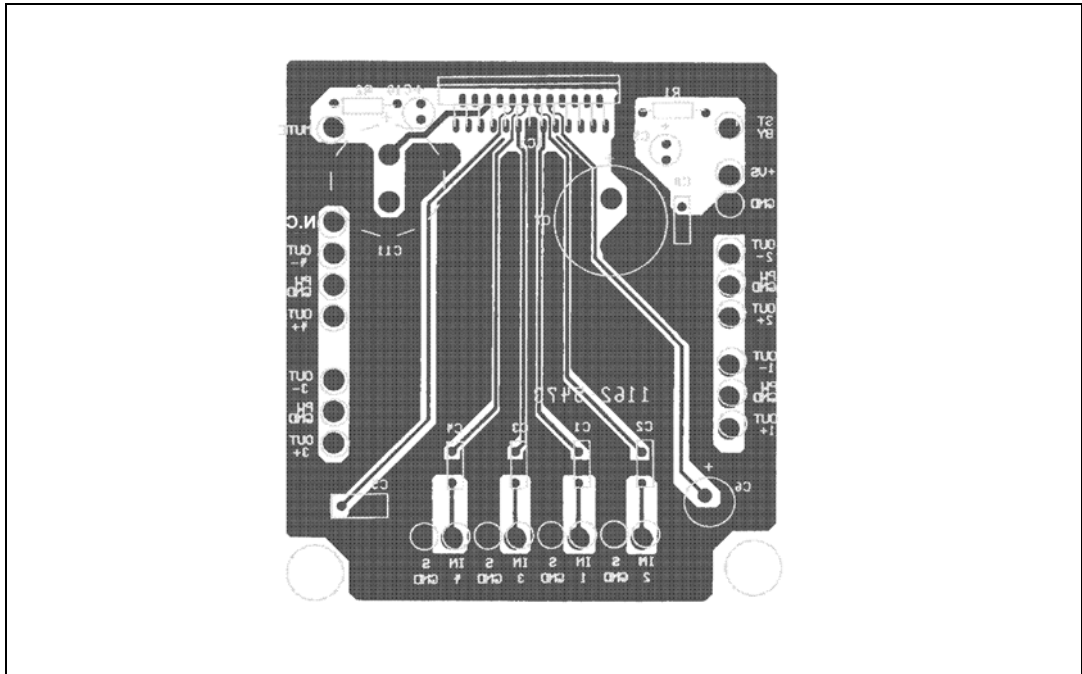


Figure 5. Bottom copper layer





## 2.5 Electrical characteristics curves

Figure 6. Quiescent current vs. supply voltage      Figure 7. Quiescent output voltage vs. supply voltage

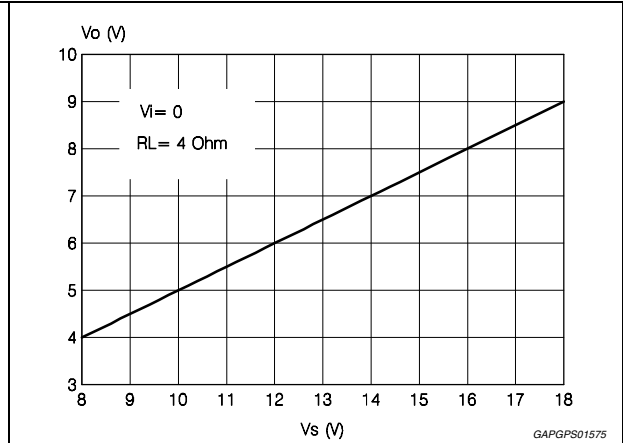
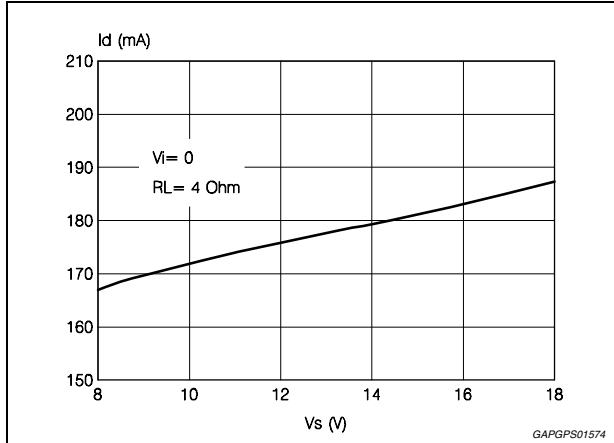


Figure 8. Output power vs. supply voltage (4Ω)

Figure 9. Distortion vs. output power

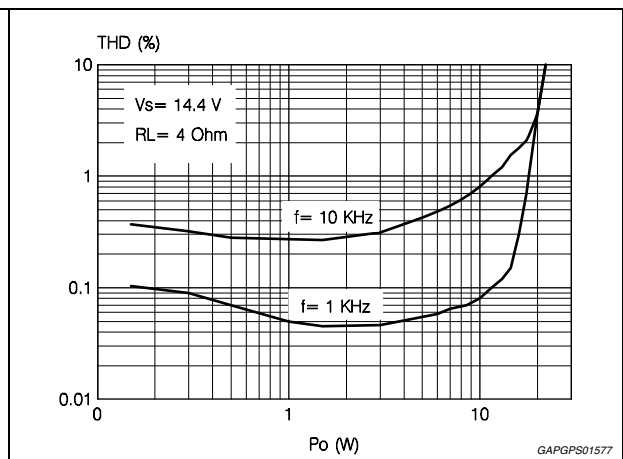
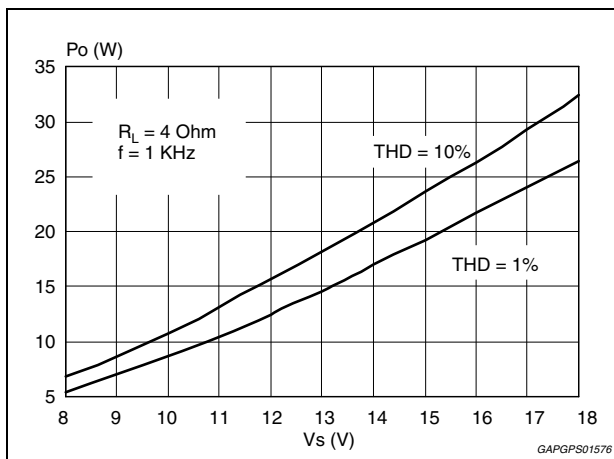
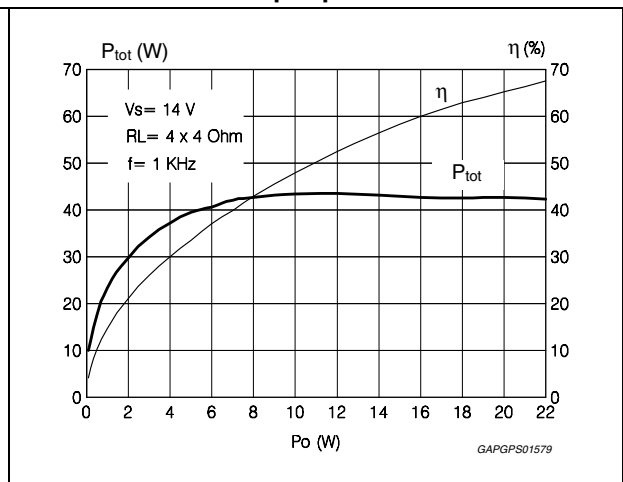
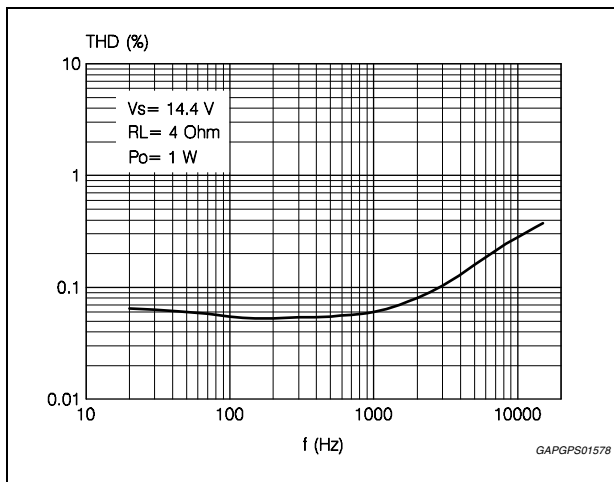


Figure 10. Distortion vs. frequency

Figure 11. Power dissipation and efficiency vs. output power



### 3 Application hints

Refer to [Figure 3](#).

#### 3.1 Biasing and SVR

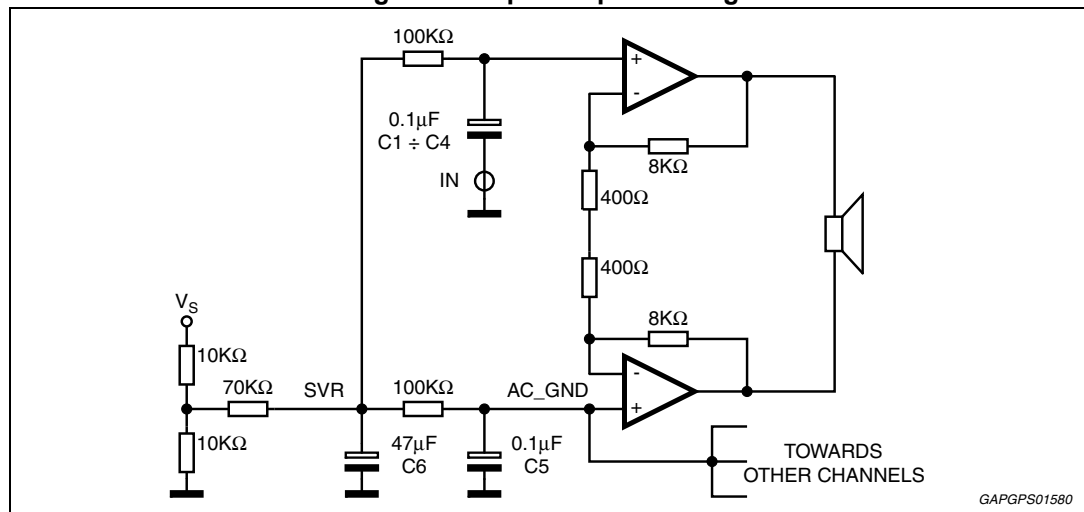
As shown in [Figure 12](#), all the TDA7387's main sections, such as Inputs, Outputs AND AC-GND (pin 16) are internally biased at half supply voltage level ( $V_s/2$ ), coming from the Supply Voltage Rejection (SVR) block. In this way no current flows through the internal feedback network. The AC-GND is common to all the 4 amplifiers and represents the connection point of all the inverting inputs.

Both individual inputs and AC-GND are connected to  $V_s/2$  (SVR) by means of 100 kΩ resistors.

To ensure proper operation and high supply voltage rejection, it is of fundamental importance to provide a good impedance matching between Inputs and AC-GROUND terminations. This implies that **C<sub>1</sub>, C<sub>2</sub>, C<sub>3</sub>, C<sub>4</sub>, C<sub>5</sub> capacitors have to carry the same nominal value and their tolerance should never exceed ± 10 %**.

Besides its contribution to the ripple rejection, the SVR capacitor controls the turn ON/OFF time sequence and, consequently, plays an essential role in the pop optimization during ON/OFF transients. To conveniently serve both needs, **its minimum recommended value is 10μF**.

Figure 12. Input/output biasing.



#### 3.2 Input stage

The TDA7387's inputs are ground-compatible and can stand very high input signals ( $\pm 8$  Vpk) without any performances degradation.

If the standard value for the input capacitors (0.1 μF) is adopted, the low frequency cut-off turns out to be 16 Hz.

### 3.3 Standby and muting

Standby and muting facilities are both CMOS-compatible. If unused, a straight connection to Vs of their respective pins would be admissible. Conventional low-power transistors can be used to drive muting and stand-by pins in absence of true CMOS ports or microprocessors. R-C cells have always to be used in order to smooth down the transitions for preventing any audible transient noises.

Since a DC current of about 10  $\mu\text{A}$  normally flows out of pin 22, the maximum allowable muting-series resistance ( $R_2$ ) is 70 k $\Omega$ , which is sufficiently high to allow a muting capacitor reasonably small (about 1  $\mu\text{F}$ ).

If  $R_2$  is higher than recommended, the involved risk is that the voltage at pin 22 may rise to above the 1.5 V threshold voltage and the device is consequently fails to turn OFF when the mute line is brought down.

About the stand-by, the time constant to be assigned in order to obtain a virtually pop-free transition has to be slower than 2.5 V/ms.

### 3.4 Stability and layout considerations

If properly layouted and hooked to standard car-radio speakers, the TDA7387 is intrinsically stable with no need of external compensations such as output R-C cells. Due to the high number of channels involved, this translates into a very remarkable components saving if compared to similar devices on the market.

To simplify pc-board layout design, each amplifier stage has its own power ground externally accessible (pins 2,8,18,24) and one supply voltage pin for each couple of them. Even more important, this makes it possible to achieve the highest possible degree of separation among the channels, with remarkable benefits in terms of cross-talk and distortion features.

About the layout grounding, it is particularly important to connect the AC-GND capacitor ( $C_5$ ) to the signal GND, as close as possible to the audio inputs ground: this guarantees high rejection of any common mode spurious signal.

The SVR capacitor ( $C_6$ ) has also to be connected to the signal GND.

Supply filtering elements ( $C_7$ ,  $C_8$ ) have naturally to be connected to the power-ground and located as close as possible to the Vs pins.

Pin 1, which is mechanically attached to the device's tab, needs to be tied to the cleanest power ground point in the pc-board, which is generally near the supply filtering capacitors.

The exposed pad package doesn't require any particular care compared to the ST standard flexiwatt package. For particular PCB configurations, in order to maximize the rejection against any disturbances coming from the battery line (SVR), it is suggested to use one of the following IC metal slug (heat-sink) connections:

- leave the slug simply electrically isolated from the PCB ground;
- in case of 2 layers board, connect the slug to the PCB power ground (P-GND) and not to the signal ground (S-GND);
- in case of a PCB with a layer dedicated to grounding (wide / diffused GND area with no distinction between P-GND and S-GND) connect the slug to the common board ground.

# 4 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK® packages, depending on their level of environmental compliance. ECOPACK® specifications, grade definitions and product status are available at: [www.st.com](http://www.st.com).

ECOPACK® is an ST trademark.

**Figure 13. Flexiwatt25 mechanical data and package dimensions**

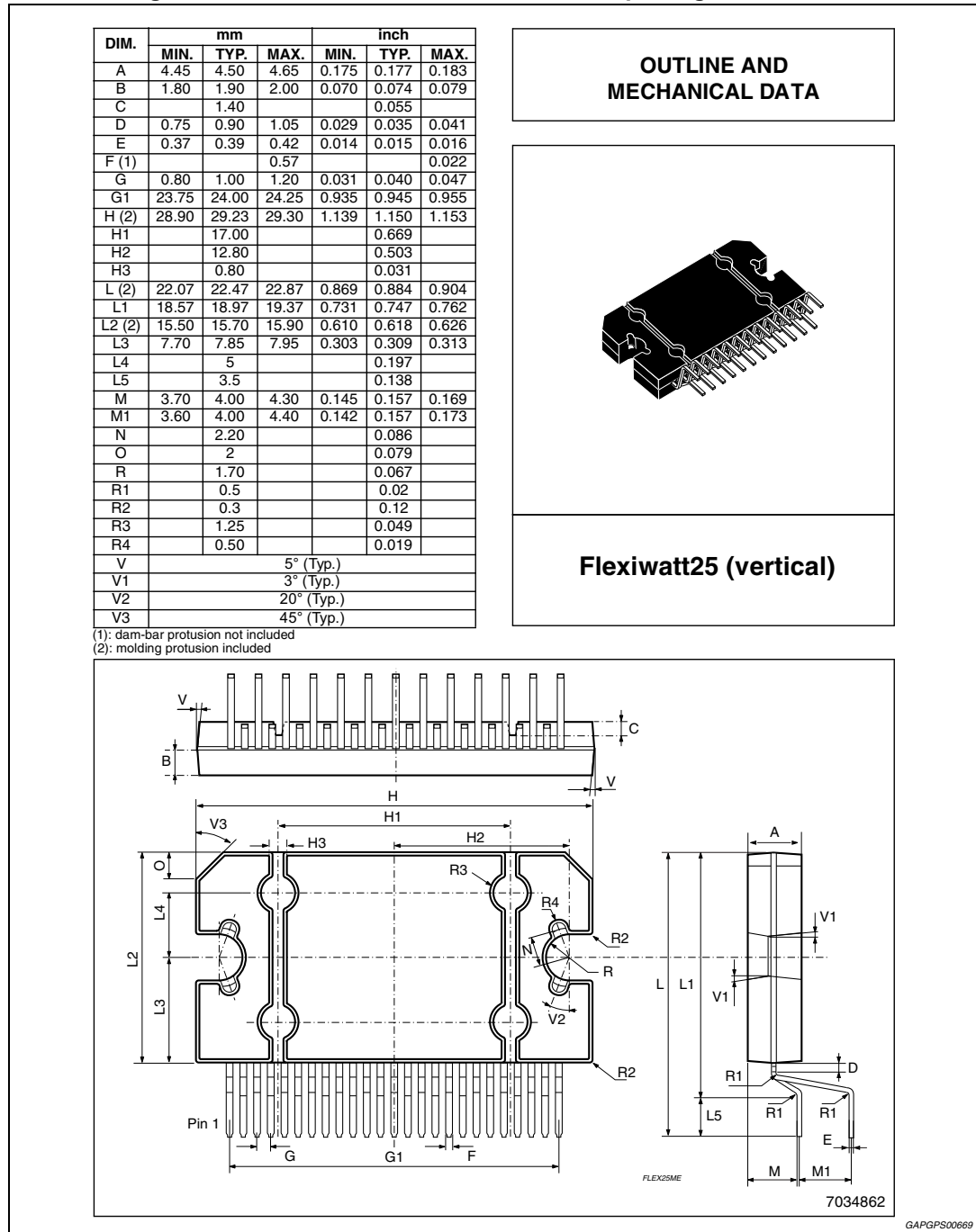
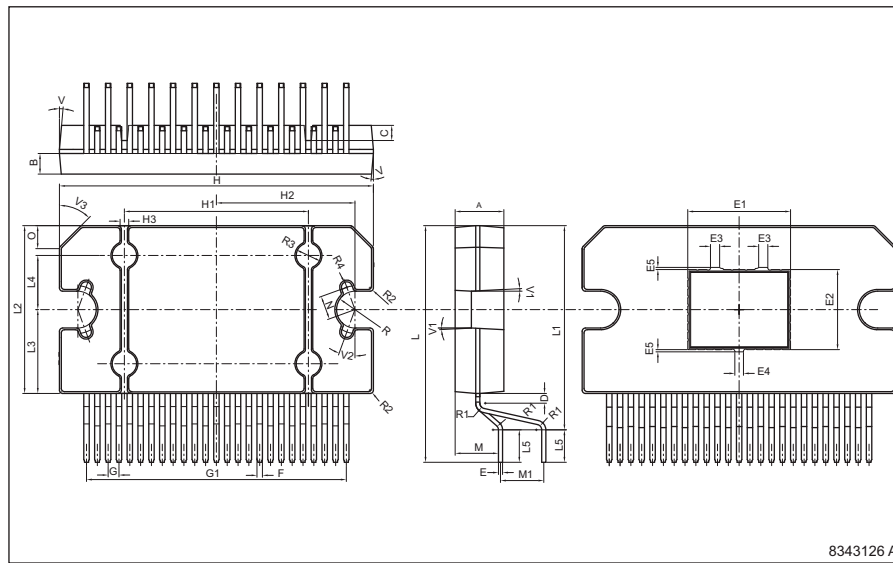
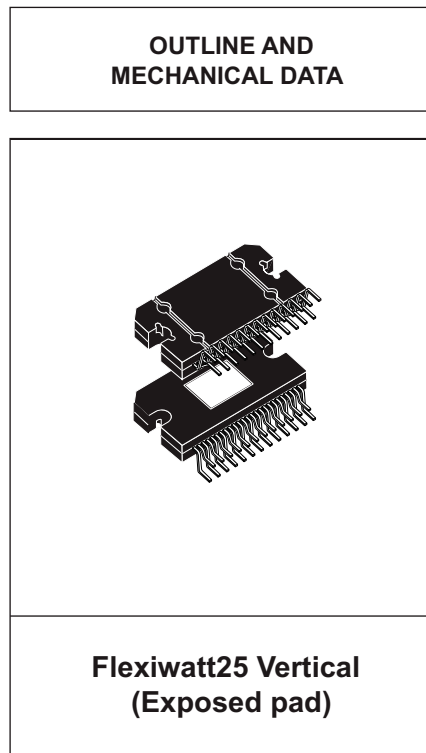


Figure 14. Flexiwatt25 mechanical data and package dimensions

DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A	4.450	4.500	4.650	0.1752	0.1772	0.1831
B	1.800	1.900	2.000	0.0709	0.0748	0.0787
C		1.400			0.0551	
D	0.750	0.900	1.050	0.0295	0.0354	0.0413
E	0.370	0.390	0.420	0.0146	0.0154	0.0165
E1	9.480	9.500	9.530	0.3732	0.3740	0.3752
E2	7.380	7.400	7.430	0.2906	0.2913	0.2925
E3	0.850			0.0335		
E4	0.500			0.0197		
E5			0.200			0.0079
F (*)			0.570			0.0224
G	0.800	1.000	1.200	0.0315	0.0394	0.0472
G1	23.750	24.000	24.250	0.9350	0.9449	0.9547
H (**)	28.900	29.230	29.300	1.1378	1.1508	1.1535
H1		17.000			0.6693	
H2		12.800			0.5039	
H3		0.800			0.0315	
L (**)	22.070	22.470	22.870	0.8689	0.8846	0.9004
L1	18.570	18.970	19.370	0.7311	0.7469	0.7626
L2 (**)	15.500	15.700	15.900	0.6102	0.6181	0.6260
L3	7.700	7.850	7.950	0.3031	0.3091	0.3130
L4		5.000			0.1969	
L5		3.500			0.1378	
M	3.700	4.000	4.300	0.1457	0.1575	0.1693
M1	3.600	4.000	4.400	0.1417	0.1575	0.1732
N		2.200			0.0866	
O		2.000			0.0787	
R		1.700			0.0669	
R1		0.500			0.0197	
R2		0.300			0.0118	
R3		1.250			0.0492	
R4		0.500			0.0197	
V5				5° (Typ.)		
V1				3° (Typ.)		
V2				20° (Typ.)		
V3				45° (Typ.)		

(\*) Dimension "F" dam-bar protrusion not included.  
 (\*\*) Dimensions H, L, L2: molding protrusion included.



8343126 A

GAPGPS01225

## 5 Revision history

**Table 5. Document revision history**

<b>Date</b>	<b>Revision</b>	<b>Changes</b>
09-Apr-2013	1	Initial release.
18-Sep-2013	2	Updated Disclaimer.

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