RENESAS

DATASHEET

ZL2102

6A Digital Integrated Synchronous Step-Down DC/DC Regulator with Auto Compensation

The ZL2102 is an integrated digital power regulator with auto compensation and power management functions in a small package, resulting in a flexible and integrated solution, which can be configured using the PowerNavigator[™] graphical user interface. This synchronous buck converter operates from a 4.5V to 14V input supply and provides from 0.54V to 5.5V output voltage at up to 6A.

The ZL2102 can be configured for most applications using only hardware pin straps to adjust switching frequency, output voltage, UVLO, soft-start ramp/delay settings, sequencing options, and SMBus address. For more advanced configurations, the ZL2102 supports over 70 PMBus commands. Output voltage/current is factory calibrated.

Internal synchronous power MOSFETs enable the ZL2102 to deliver continuous loads up to 6A with high efficiency. An internal Schottky bootstrap diode reduces discrete component count. The ZL2102 also supports phase spreading to reduce system input capacitance.

The ZL2102 uses the SMBus™ with PMBus™ protocol for communication with a host controller and the Intersil's proprietary Digital-DC™ bus for interoperability between other Intersil devices.

Features

- Integrated MOSFET switches
- 6A continuous output current
- Adjustable 0.54V to 5.5V output range
- 4.5V to 14V input range
- Up to 90% efficiency
- Auto compensation for fast transient response
- SMBus compliant serial interface
- Snapshot™ parametric capture
- Internal nonvolatile memory
- Small footprint QFN package (6mmx6mm)

Applications

- Servers/storage equipment
- Telecom/datacom equipment
- Power supplies (memory, DSP, ASIC, FPGA)

Related Literature

- **[AN2010](http://www.intersil.com/content/dam/Intersil/documents/an20/an2010.pdf)** "Thermal and Layout Guidelines for Digital-DC™ Products"
- **[AN2035](http://www.intersil.com/content/dam/Intersil/documents/an20/an2035.pdf)** "Compensation Using CompZL™"
- [TB389](http://www.intersil.com/content/dam/Intersil/documents/tb38/tb389.pdf) "PCB Land Pattern and Surface Mount Guidelines for QFN Packages"

FIGURE 1. TYPICAL APPLICATION DIAGRAM

FN8440 Rev 2.00 November 20, 2014

Table of Contents

FIGURE 2. BLOCK DIAGRAM

Pin Configuration

ZL2102 (36 LD 6x6 QFN) TOP VIEW

Pin Description

Ordering Information

NOTES:

1. Please refer to **[TB347](http://www.intersil.com/data/tb/tb347.pdf)** for details on reel specifications.

2. These Intersil Pb-free plastic packaged products employ special Pb-free material sets, molding compounds/die attach materials, and 100% matte tin plate plus anneal (e3 termination finish, which is RoHS compliant and compatible with both SnPb and Pb-free soldering operations). Intersil Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.

3. For Moisture Sensitivity Level (MSL), please see device information page for [ZL2102](http://www.intersil.com/products/ZL2102#packaging). For more information on MSL, please see tech brief [TB363](http://www.intersil.com/data/tb/tb363.pdf).

Absolute Maximum Ratings Thermal Information

Recommended Operating Conditions

CAUTION: Do not operate at or near the maximum ratings listed for extended periods of time. Exposure to such conditions may adversely impact product reliability and result in failures not covered by warranty.

NOTES:

- 4. θ_{JA} is measured in free air with the device mounted on a multi-layer FR4 test board and the exposed metal pad soldered to a low impedance ground plane using multiple vias.
- 5. For θ_{JC} , the "case temp" location is the center of the exposed metal pad on the package underside.
- 6. Thermal impedance is dependent upon PCB layout.
- 7. Includes margin limits.

Electrical Specifications VDDP = VDDS = 12V, T_A = -40°C to +85°C unless otherwise noted ([Note 9\)](#page-6-2). Typical values are at T_A = +25 °C. Boldface limits apply across the operating temperature range, -40 °C to +85 °C.

Electrical Specifications vDDP = VDDS = 12V, T_A = -40°C to +85°C unless otherwise noted (<u>Note 9</u>). Typical values are at T_A = +25 ° C. Boldface limits apply across the operating temperature range, -40 ° C to +85 ° C. (Continued)

Electrical Specifications VDDP = VDDS = 12V, T_A = -40°C to +85°C unless otherwise noted (Note 9). Typical values are at T_A = +25 °C. Boldface limits apply across the operating temperature range, -40 °C to +85 °C. (Continued)

NOTES:

8. Compliance to datasheet limits is assured by one or more methods: production test, characterization and/or design.

9. Refer to Safe Operating Area in [Figure 8](#page-8-3) and thermal design guidelines in **AN2010**.

10. Does not include margin limits.

11. Switch node current should not exceed IRMS of 6A.

Typical Performance Curves

For some applications, ZL2102 operating conditions (input voltage, output voltage, switching frequency, temperature) may require derating to remain within the Safe Operating Area (SOA). V_{IN} = V_{DDP} = V_{DDS}, T_J = +125°C

FIGURE 3. LOW-SIDE r_{DS(ON)} vs T_J NORMALIZED FOR T_J = +25°C $(V_{\text{DDS}} = 12V, I_{\text{DRAIN}} = 0.3A)$

FIGURE 4. HIGH-SIDE r_{DS(ON)} vs T_J NORMALIZED FOR T_J = +25 °C $(V_{\text{DDS}} = 12V, \overline{BST} - SW = 6.5V, I_{\text{DRAIN}} = 0.3A)$

Typical Performance Curves

For some applications, ZL2102 operating conditions (input voltage, output voltage, switching frequency, temperature) may require derating to remain within the Safe Operating Area (SOA). V_{IN} = V_{DDP} = V_{DDS}, T_J = +125°C (**Continued)**

FIGURE 5. LOW-SIDE r_{DS(ON)} vs V_{DDS} WITH T_J

FIGURE 6. SAFE OPERATING AREA, $T_1 \leq +125$ °C

FIGURE 7. MAXIMUM CONVERSION RATIO, $T_1 \leq +125$ °C

Digital-DC Architecture Overview

The ZL2102 is an innovative mixed-signal power conversion and power management IC based on Intersil patented Digital-DC technology that provides an integrated, high performance stepdown regulator for point of load applications. The ZL2102 integrates all necessary PWM control circuitry as well as low $r_{DS(ON)}$ synchronous power MOSFETs to provide an extremely small solution for supplying load currents up to 6A.

Its unique PWM loop utilizes an ideal mix of analog and digital blocks to enable precise control of the entire power conversion process with no software required, resulting in a very flexible device that is also very easy to use. An extensive set of power management functions are fully integrated and can be configured using simple pin connections. The user configuration can be saved in an internal nonvolatile memory (NVM). Additionally, all functions can be configured and monitored via the SMBus hardware interface using standard PMBus commands, allowing ultimate flexibility.

Once enabled, the ZL2102 is immediately ready to regulate power and perform power management tasks with no programming required. Advanced configuration options and realtime configuration changes are available via the SMBus interface if desired and continuous monitoring of multiple operating parameters is possible with minimal interaction from a host controller. Integrated sub regulation circuitry enables single supply operation from any external supply between 4.5V and 14V with no additional bias supplies needed.

The ZL2102 can be configured by simply connecting its pins according to the tables provided in the following sections. Additionally, a comprehensive set of application notes are available to help simplify the design process. An evaluation board is also available to help the user become familiar with the device. This board can be evaluated as a standalone platform using pin configuration settings. A Windows[™]-based GUI is also provided to enable full configuration and monitoring capability via the SMBus interface using a computer and the included USB cable.

Power Conversion Overview

The ZL2102 operates as a voltage-mode, synchronous buck converter with a selectable constant frequency pulse width modulator (PWM) control scheme. The ZL2102 integrates dual low r_{DS(ON)} synchronous MOSFETs and a high-side driver to minimize the circuit footprint. [Figure 8](#page-8-3) illustrates the basic synchronous buck converter topology showing the primary power train components. This converter is also called a step-down converter, as the output voltage must always be lower than the input voltage.

FIGURE 8. STEP DOWN CONVERTER

The ZL2102 integrates two N-channel power MOSFETs; QH is the top control MOSFET and QL is the bottom synchronous MOSFET. The amount of time that QH is on as a fraction of the total switching period is known as duty cycle D, which is described by **Equation 1:**

$$
D \approx \frac{V_{\text{OUT}}}{V_{\text{IN}}} \tag{EQ. 1}
$$

During time D, QH is on and V_{IN} to V_{OUT} is applied across the inductor. The output current ramps up as shown in **Figure 9.**

FIGURE 9. OUTPUT CURRENT

When QH turns off (time 1-D), the current flowing in the inductor must continue to flow from the ground up through QL, during which the current ramps down. Since the output capacitor C_{OUT} exhibits low impedance at the switching frequency, the AC component of the inductor current is filtered from the output voltage so the load sees nearly a DC voltage.

The maximum conversion ratio is shown in **Figure 7**. Typically, buck converters specify a maximum duty cycle that effectively limits the maximum output voltage that can be realized for a given input voltage and switching frequency. This duty cycle limit ensures that the low-side MOSFET is allowed to turn on for a minimum amount of time during each switching cycle, which enables the bootstrap capacitor to be charged up and provide adequate gate drive voltage for the high-side MOSFET.

The block diagram for the ZL2102 is illustrated in [Figure 2.](#page-2-1) In this circuit, the target output voltage is regulated by connecting the VSEN pin directly to the output regulation point. The VSEN signal is then compared to an internal reference voltage that had been set to the desired output voltage level by the user. The error signal derived from this comparison is converted to a digital value with an analog-to-digital (A/D) converter. The digital signal is also applied to an adjustable digital compensation filter and the compensated signal is used to derive the appropriate PWM duty cycle for driving the internal MOSFETs in a way that produces the desired output.

Power Management Overview

The ZL2102 incorporates a wide range of configurable power management features that are simple to implement with no external components. Additionally, the ZL2102 includes circuit protection features that continuously safeguard the device and load from damage due to unexpected system faults. The ZL2102 can continuously monitor input voltage, output voltage/current, and internal temperature. A power-good output signal is also included to enable power-on reset functionality for an external processor.

All power management functions can be configured using either pin configuration techniques described in this document or via the SMBus interface using PMBus commands. Monitoring parameters can also be preconfigured to provide alerts for specific conditions. "PMBus Command Summary" on page 19 contains a listing of all the PMBus commands supported by the ZL2102 and a detailed description of the use of each of these commands.

Functional Description and Configuration

INTERNAL BIAS REGULATORS AND INPUT SUPPLY CONNECTIONS

The ZL2102 employs three internal LDO regulators, allowing operation from a single input supply from 4.5V to 14V. The regulators are as follows:

- VR is derived from VDDS and provides a 7V bias supply for the internal high-side MOSFET driver circuit. A 4.7µF capacitor is required for the VR pin.
- VRA is derived from VDDS and provides a 5V bias supply for the internal analog circuitry. A 4.7µF capacitor is required for the VR pin.
- V2P5 is derived from VRA and provides a 2.5V bias supply for the digital circuitry. A 10µF capacitor is required at the V2P5 pin.

Input voltage ranges and connections are shown in **Figure 10**.

FIGURE 10. INPUT SUPPLY CONNECTIONS

The internal bias regulators, VR and VRA, are not designed to be outputs for powering other circuitry. Do not attach external loads to any of these pins. Only the multimode pins may be connected to the V2P5 pin for logic HIGH settings.

MULTIMODE PINS

In order to simplify circuit design, the ZL2102 incorporates patented multimode pins that allow the user to easily configure many aspects of the device with no programming. Most power management features can be configured using these pins. The multimode pins can respond to four different connections as shown in [Table 1](#page-9-2). These pins are sampled once when power is applied or by issuing a PMBus Restore command.

Pin Strap Settings: This is the simplest implementation method, as no external components are required. Using this method, each pin can take on one of three possible states: LOW, OPEN, or HIGH. These pins can be connected to the V25 pin for logic HIGH settings. Using a single pin, one of three settings can be selected.

Resistor Settings: This method allows a greater range of adjustability when connecting a finite value resistor (in a specified range) between the multimode pin and SGND. Standard 1% resistor values are used, and only every fourth E96 resistor value is used so the device can reliably recognize the value of resistance connected to the pin while eliminating the error associated with the resistor accuracy. Up to 31 unique selections are available using a single resistor.

FIGURE 11. PIN STRAP AND RESISTOR SETTING EXAMPLES

SMBus: Most ZL2102 functions/parameters can be configured via the SMBus interface using standard PMBus commands. "PMBus Command Summary" on page 19 explains the use of the available PMBus commands in detail.

CONFIGURABLE PINS

Many operating parameters can be set using the multimode pin setup method: SMBus address (SA), output voltage (VSET), clock synchronization and sequencing options (CFG), switching frequency (SYNC), soft-start delay, soft-start ramp, input undervoltage lock-out (SS), and automatic loop compensation settings (FC). These pins are checked once during start-up only. Changes to the settings of these pins will not be read until the device's power supply has been cycled off and on.

The device's SMBus address is the only parameter that must be set by the multimode pins. All others are configurable using PMBus commands.

SMBus Device Address Selection (SA)

The ZL2102 provides an SMBus digital interface that enables the user to configure all aspects of the device operation as well as monitor the input and output parameters. The ZL2102 is compatible with SMBus version 2.0 and includes an SALRT line to help mitigate bandwidth limitations related to continuous fault monitoring.

When communicating with multiple devices using the SMBus interface, each device must have its own unique address so the host can distinguish between the devices. The device address can be set according to the pin strap options listed in [Table 2](#page-10-2). The SMBus address cannot be changed with a PMBus command.

TABLE 2. PIN STRAP OPTIONS

Output Voltage and VOUT_MAX Selection (VSET)

The output voltage may be set to any voltage between 0.6V and 5.5V provided that the input voltage is higher than the desired output voltage by an amount sufficient to prevent the device from exceeding its maximum duty cycle specification. V_{OUT} can be set to any of the pin strap options shown in $Table 3. V_{OUT}$ $Table 3. V_{OUT}$ can also be set using the VOUT_COMMAND PMBus command.

The maximum accepted value of V_{OUT} is limited by VOUT_MAX. The default value of VOUT_MAX is 110% of the VSET pin strap setting, but it can also be set using the VOUT_MAX PMBus command.

TABLE 3. (Continued)

In addition to the VOUT_COMMAND and VOUT_MAX settings, this pin strap setting is also used to set several other V_{OUT} related settings including:

- VOUT_UV_FAULT_LIMIT = 85% of VSET
- \cdot POWER_GOOD_ON = 90% of VSET
- VOUT_MARGIN_LOW = 95% of VSET
- VOUT_MARGIN_HIGH = 105% of VSET
- VOUT_OV_FAULT_LIMIT = 115% of VSET

The above parameters are automatically adjusted by the VSET pin strap selection. If the value of VOUT_COMMAND is adjusted via PMBus, the values of these commands may also need to be adjusted to compensate for the V_{OUT} change. The configured voltage relationships must follow: VOUT_UV_FAULT_LIMIT < POWER_GOOD_ON < VOUT_MARGIN_LOW < VOUT_COMMAND < VOUT_MARGIN_HIGH < VOUT_OV_FAULT_LIMIT.

Automatic Loop Compensation (FC)

The ZL2102 has an automatic loop compensation feature that measures the characteristics of the power train and calculates the proper PID tap coefficients. Auto compensation is configured using the FC pin as shown in [Table 4](#page-10-4).

When Auto Comp is enabled, it will run once each time the part is enabled. Auto Comp runs a series of tests on the output and calculates the optimal tap coefficients immediately after the V_{OUT} ramp completes. The calculated tap coefficients are stored in the Auto Comp Store and may be read back through the PID TAPS PMBus command. If auto compensation is disabled, the device will use the tap coefficients that are stored in the USER_STORE.

If the PG Assert parameter is set to "Use PG Delay", the PG pin will be asserted according to the POWER_GOOD_DELAY command. When Auto Comp is enabled and the "Use PG Delay" option is selected, the user must ensure that the output perturbation from the Auto Comp test cycle is acceptable before PG is asserted. If PG Assert is set to "After Auto Comp", PG will be asserted immediately after the Auto Comp cycle completes (POWER_GOOD_DELAY will be ignored in this case).

The Auto Comp Gain control scales the Auto Comp results to allow a trade-off between transient response and steady-state duty cycle jitter. A setting of 100% will provide the fastest transient response while a setting of 10% will produce the lowest jitter. For best results, V_{IN} must be stable before Auto Comp begins as shown in **Equation 2:**

The auto compensation function can also be configured via the AUTO_COMP_CONFIG PMBus command and controlled using the AUTO_COMP_CONTROL PMBus command. Compensation values can be programmed manually by disabling Auto Comp and writing preferred values to the PID_TAPS PMBus command.

Synchronization and Sequencing Configuration Settings (CFG)

The ZL2102 supports several options of clock synchronization and output sequencing. The ZL2102's configuration settings can be set using the CFG pin strap options shown in [Table 5](#page-11-3). The operation of these functions is covered in their respective sections of this document.

Switching Frequency Setting (SYNC)

The ZL2102's switching frequency can be set from 200kHz to 1000kHz using the SYNC pin strap options shown in [Table 6](#page-11-4).

TABLE 6.

The switching frequency can also be set to any value between 200kHz and 1MHz using the FREQUENCY_SWITCH PMBus command. The available frequencies below 1MHz are defined by $f_{SW} = 8MHz/N$, where $8 \le N \le 40$.

If a value other than $f_{SW} = 8MHz/N$ is entered using the PMBus command, the device will select the switching frequency value using N as a whole number to achieve a value nearest to the entered value. For example, if 810kHz is entered, the device will select 800 k Hz (N = 10).

Note: The switching frequency read back using the appropriate PMBus command may differ slightly from the programmed value. The difference is due to hardware quantization.

The SYNC pin can also be configured to perform synchronization between devices. The CFG pin is used to configure the SYNC pin

as an input, an output, or auto-detect mode. The ZL2102 incorporates an internal phase-locked loop (PLL) to clock the internal circuitry. The PLL can be driven by an external clock source connected to the SYNC pin, or it can be configured to drive the internal clock out of the SYNC pin to other devices.

SYNC OUTPUT: When the SYNC pin is configured as an output, the device will run from its internal oscillator and will drive the selected switching frequency onto the SYNC pin so that other devices can be synchronized to it.

SYNC INPUT: When the SYNC pin is configured as an input, the device will check for an external clock signal on the SYNC pin each time the output is enabled. The internal oscillator will then synchronize with the rising edge of the external clock. The incoming clock signal must be in the range of 200kHz to 1MHz and must be stable when the enable pin is asserted. The clock signal must have a minimum width of 150ns, and it must stay within 10% of its initial value. In the event of a loss of the external clock signal, will automatically switch to its internal oscillator and switch at a frequency close to the previous incoming frequency. The output voltage may show a transient overshoot/undershoot if this occurs.

SYNC AUTO DETECT: When the SYNC pin is configured in auto-detect mode, the device will check for a clock signal on the SYNC pin each time EN is asserted. If there is a valid clock, the pin will run as a sync input.

If no incoming clock signal is present, the device will switch at the selected internal clock rate.

Soft-Start and UVLO Settings (SS)

The ZL2102 supports variable turn-on/off delay times, turn-on/off ramp rates, and input undervoltage lockout (UVLO) functions. These features may be used as part of an overall in-rush current management strategy or to precisely control how fast a load is turned on. The ZL2102 provides several options for precisely and independently controlling both the delay and ramp time periods.

The soft-start delay period begins when the EN pin is asserted and ends when the delay time expires. The soft-start ramp timer enables a precisely controlled monotonic ramp to the nominal V_{OUT} value that begins once the delay period has expired. This process is also followed for ramp down after the EN pin has been deasserted.

The input undervoltage lockout (UVLO) prevents the ZL2102 from operating when the input falls below a preset threshold, indicating the input supply is out of its specified range.

The ZL2102's TON/TOFF delay time, TON/TOFF ramp time, and UVLO functions can be configured using the SS pin strap options shown in [Table 7](#page-12-2).

 \blacksquare

These functions can also be set independently using TON_DELAY, TON_RISE, TOFF_DELAY, and TOFF_FALL PMBus commands. The UVLO threshold can be adjusted using the VIN_UV_FAULT_LIMIT PMBus command. The ramp down function is disabled by default, but it can be enabled using the ON_OFF_CONFIG or OPERATION PMBus commands depending on the desired device enable method.

Start-up Procedure

The ZL2102 follows a specific internal start-up procedure after power is applied. [Figure 12](#page-13-5) describes the start-up sequence.

FIGURE 12. START-UP SEQUENCE

If the device is to be synchronized to an external clock source, the clock frequency must be stable prior to asserting the EN pin. The device requires approximately 35ms to check for specific values stored in its internal memory. If the user has stored values in memory, those values will be loaded. If Auto Comp is enabled, start-up time increases to ~55ms.

Once this process is completed, the device is ready to accept commands via the serial interface and the device is ready to be enabled. Once enabled, the device requires approximately 5ms before its output voltage may be allowed to start its ramp-up process. If a soft-start delay period less than 5ms has been configured (using PMBus commands), the device will default to a 5ms delay period. If a delay period greater than 5ms is configured, the device will wait for the configured delay period prior to starting to ramp its output.

After the delay period has expired, the output will begin to ramp towards its target voltage according to the preconfigured soft-start ramp time.

Power-Good

The ZL2102 provides a Power-Good (PG) signal that indicates the output voltage is within a specified tolerance of its target level and no fault condition exists. By default, the PG pin will assert if the output reaches 90% of the target voltage. The limit and drive configuration of the pin may be changed using the POWER_GOOD_ON and MFR_CONFIG PMBus commands.

A PG delay period is defined as the time from when all conditions for asserting PG are met to when the PG pin is actually asserted. This feature is commonly used instead of using an external reset controller to control external digital logic. By default, the ZL2102 PG delay is set equal to 1ms. The PG delay can be adjusted using the POWER_GOOD_DELAY PMBus command.

Power Management Function **Description**

Input Undervoltage Lockout

The input undervoltage lockout feature (UVLO) prevents the ZL2102 from operating when the input falls below a preset threshold, indicating the input supply is out of its specified range. The UVLO threshold can be set to 4.5V, 5.5V, or 7.5V using the SS pin. The UVLO voltage can also be adjusted using the VIN_UV_FAULT_LIMIT PMBus command.

The default response from a UVLO fault is an immediate shutdown of the device during the fault and an automatic restart when the fault condition has cleared. The UVLO fault response can be configured using the VIN_UV_FAULT_RESPONSE PMBus command.

Output Overvoltage Protection

The ZL2102 offers an internal output overvoltage protection circuit that can be used to protect sensitive load circuitry from being subjected to a voltage higher than its prescribed limits. A hardware comparator is used to compare the actual output voltage (seen at the VSEN pin) to a threshold set above the target output voltage. The default setting is 115% of the pin strap setting for VOUT, but this value can be adjusted using the VOUT_OV_FAULT_LIMIT PMBus command. If the VSEN voltage exceeds this threshold, the PG pin will deassert and the device response can be set using the VOUT_OV_FAULT_RESPONSE PMBus command. The default response is an immediate shutdown of the device during the fault and an automatic restart when the fault has cleared. For continuous overvoltage protection when operating from an external clock, the only allowed response is an immediate shutdown with no automatic restart.

Output Prebias Protection

An output prebias condition exists when an externally applied voltage is present on a power supply's output before the power supply's control IC is enabled. Certain applications require that the converter not be allowed to sink current during start-up if a prebias condition exists at the output. The ZL2102 provides prebias protection by sampling the output voltage prior to initiating an output ramp.

If a prebias voltage exists after the preconfigured delay period has expired, the ramp start voltage will be set to match the existing prebias voltage and the output will be enabled. The output voltage will then ramp to the final regulation value at the ramp rate set by the SS pin strap setting or the TON_RISE PMBus command.

The actual time the output will take to ramp up from the prebias voltage to the target voltage will vary depending on the prebias voltage, but the total time from enable to when the output reaches its target value will match the preconfigured delay and ramp times. When ramping down from a prebias higher than the target voltage, the device will wait until after SS ramp time is complete and will then ramp down to the target voltage at an approximate rate of 0.1V/ms (see [Figure 13](#page-14-5)).

VPREBIAS > VTARGET

FIGURE 13. OUTPUT RESPONSES TO PREBIAS VOLTAGES

If the prebias voltage is higher than the output overvoltage limit set in VOUT_OV_FAULT_LIMIT, the device will declare an overvoltage fault condition and it will respond based on the output overvoltage response method that has been selected in the VOUT_OV_FAULT_RESPONSE PMBus command.

Output Overcurrent Protection

The ZL2102 incorporates a patented "lossless" current sensing method across the internal low-side MOSFET that is independent of r_{DS(ON)} variations, including temperature. This current limiting mechanism is used to protect the power supply from damage and prevent excessive current from being drawn from the input supply in the event that the output is overloaded or shorted to ground. Current limiting is accomplished by sensing the current through the circuit during a portion of the duty cycle. The current limit threshold is set to 7.2A by default, but it can be adjusted using the IOUT_AVG_OC_FAULT_LIMIT PMBus command. The default response of an overcurrent fault is an immediate shutdown of the device during the fault and an automatic restart when the fault has cleared, but it can be adjusted using the MFR_IOUT_OC_RESPONSE PMBus command.

Thermal Overload Protection

The ZL2102 includes an on-chip thermal sensor that continuously measures the internal temperature of the die and shuts down the device when the temperature exceeds the preset limit. The default temperature limit is set to +125°C in the factory, but it can be adjusted using the OT_FAULT_LIMIT PMBus command. The default response to an over-temperature fault is

an immediate shutdown of the device during the fault and an automatic restart when the fault has cleared, but the response can be adjusted using the OT_FAULT_RESPONSE PMBus command.

If the device is configured to restart, it will wait 250ms and will then check the device temperature. Once the temperature has dropped below the over-temperature warning limit, the device will attempt to restart. The default value of the over-temperature warning limit is +110°C, providing +15°C of hysteresis, but the value can be adjusted using the OT_WARN_LIMIT PMBus command.

Voltage Margining

The ZL2102 provides a simple method to vary its output higher or lower than its nominal voltage setting in order to determine whether the load device is capable of operating over its specified supply voltage range. The voltage margin range can be set by driving the MGN pin or using the OPERATION PMBus command. The MGN pin is a tri-level input that is continuously monitored and can be driven directly by a processor I/O pin or other logic-level output.

The ZL2102's output will be forced higher than its nominal set point when the MGN command is set HIGH, and the output will be forced lower than its nominal set point when the MGN command is set LOW. The default margin settings are ±5% of the pin strapped value of V_{OUT} , but the margin settings can be adjusted using the VOUT_MARGIN_HIGH and VOUT_MARGIN_LOW PMBus commands. The default transition rate between the nominal output voltage and either margin limit is 0.5V/ms, but it can be adjusted using the VOUT_TRANSITION_RATE PMBus command.

Digital-DC Bus

The Digital-DC Communications (DDC) bus is used to communicate between Intersil Digital-DC devices. This dedicated bus provides the communication channel between devices for features such as sequencing and fault spreading. The DDC pin on all Digital-DC devices in an application should be connected together. A pull-up resistor is required on the DDC bus in order to guarantee the rise time as follows: Rise time = $R_{PU} * C_{LOAD} \approx 1 \mu s$, where R_{PI} is the DDC bus pull-up resistance and C_{LOAD} is the bus loading. The pull-up resistor may be tied to VR or to an external 3.3V or 5V supply as long as this voltage is present prior to or during device power-up. As a rule of thumb, each device connected to the DDC bus presents approximately 10pF of capacitive loading, and each inch of FR4 PCB trace introduces approximately 2pF. The ideal design will use a central pull-up resistor that is well-matched to the total load capacitance. In power module applications, the user should consider whether to place the pull-up resistor on the module or on the PCB of the end application. The minimum pull-up resistance should be limited to a value that enables any device to assert the bus to a voltage that will ensure a logic 0 (typically 0.8V at the device monitoring point) given the pull-up voltage (5V if tied to VR5) and the pull-down current capability of the ZL2102 (nominally 4mA).

Phase Spreading

When multiple power converters share a common DC input supply, it is desirable to adjust the clock phase offset of each device, such that not all devices start to switch simultaneously.

Setting each converter to start its switching cycle at a different point in time can dramatically reduce input capacitance requirements and efficiency losses. Since the peak current drawn from the input supply is effectively spread out over a period of time, the peak current drawn at any given moment is .
reduced and the power losses proportional to the I_{RMS}² are reduced dramatically.

In order to enable phase spreading, all converters must be synchronized to the same switching clock. This can be accomplished by setting the CFG pin strap or by using the INTERLEAVE PMBus command.

Selecting the phase offset for the device is accomplished by selecting a device address according to **[Equation 3](#page-15-5):**

Phase offset = device address \times 45 \degree (EQ. 3)

For example:

- A device address of 0x20 would configure no phase offset
- A device address of 0x21 would configure 45 $^{\circ}$ of phase offset
- A device address of 0x22 would configure 90 $^{\circ}$ of phase offset

The phase offset of each device may also be set to any value between 0° and 360° in 22.5° increments via the INTERLEAVE PMBus command.

Output Sequencing

Intersil devices may be configured to power-up as a group in a predetermined sequence. This feature is especially useful when powering advanced processors, FPGA's and ASIC's that require one supply to reach its operating voltage prior to another supply reaching its operating voltage in order to avoid latch-up from occurring. Multidevice sequencing can be achieved by configuring each device through the SEQUENCE PMBus command or by using Intersil patented autonomous sequencing mode.

Autonomous sequencing mode configures sequencing by using events transmitted between devices over the DDC bus. The sequencing order is determined using each device's SMBus address. Using autonomous sequencing mode (configured using the CFG pin), the devices must be assigned sequential SMBus addresses with no missing addresses in the chain, the enable pins must be tied together, and the sync pins of all devices must be tied together. The first device in the sequence will become the clock master, and the other devices will sync to this clock. This mode will also constrain each device to have a phase offset according to its SMBus address as described in the "Phase Spreading" on page 15.

The enable (EN) line is driven high to initiate a sequenced turn-on of the group. Enable is driven low to initiate a sequenced turnoff of the group. During enable, the sequencing group will enable in order starting with the device with the lowest SMBus address and will continue through to turn on each device in the address chain until all devices connected have been turned on. During disable, the device with the highest SMBus address will turn off first followed in reverse order by the other devices in the group. The PG threshold is used to determine when the following device is enabled.

Fault Spreading

Digital-DC devices can be configured to broadcast a fault event over the DDC bus to the other devices in the group. When a fault occurs and the device is configured to shut down on a fault, the device will shut down and broadcast the fault event over the DDC bus. The other devices on the DDC bus can be configured to respond to the broadcast in several ways including group shutdown/restart. Fault spreading mode can be enabled through the USER_CONFIG PMBus command, and the fault spreading group can be defined through the DDC_GROUP PMBus command.

Monitoring via SMBus

A system controller can monitor a wide variety of different ZL2102 parameters through the SMBus interface. The device can monitor for fault conditions by monitoring the SALRT pin, which will be asserted when any number of preconfigured fault conditions occur.

The device can also be monitored continuously for many power conversion parameters including input voltage, output voltage, output current, internal junction temperature, switching frequency, duty cycle, fault status information.

The PMBus Host should respond to SALRT as follows:

- ZL device pulls SALRT Low.
- PMBus Host detects that SALRT is now low, performs transmission with Alert Response Address to find which ZL device is pulling SALRT low.
- PMBus Host talks to the ZL device that has pulled SALRT low. The actions that the host performs are up to the System Designer.

If multiple devices are faulting, SALRT will still be low after doing the above steps and will require transmission with the Alert Response Address repeatedly until all faults are cleared.

Nonvolatile Memory

The ZL2102 has internal nonvolatile memory where user settings are stored. There are two storage locations; the factory store and the user store. During the initialization process, the ZL2102 checks for any stored values in the user store. If user programmed settings are found, they will be used for those parameters. If there are no user settings stored, the factory settings and pin strap values are used for those parameters.

Snapshot™ Parametric Capture

The ZL2102 offers a special feature that enables the user to capture parametric data during normal operation or following a fault. The Snapshot function can be enabled through the MISC CONFIG PMBus command, and the data can be read back as a block read transfer using the SNAPSHOT PMBus command.

The SNAPSHOT_CONTROL PMBus command enables the user to store the snapshot parameters to NV memory in response to a fault as well as to read the stored data from NV memory after a fault has occurred.

Automatic writes to NV memory following a fault are triggered when any fault threshold level is exceeded, provided that the

specific fault's response is to shut down. Writing to NV memory is not allowed if the device is configured to retry following the specific fault condition.

The device's VDD voltage must remain valid during the time when the device is writing the data to NV memory; a process that requires up to 2ms.

Power Train Component Selection

The ZL2102 is a synchronous buck converter with MOSFETs that uses an external inductor and capacitors to perform the power conversion process. The proper selection of the external components is critical for optimized performance.

To select the appropriate external components for the desired performance goals, the power supply requirements listed in [Table 8](#page-16-4) must be known.

TABLE 8. TABLE SUPPLY REQUIREMENTS

Design Goal Trade-offs

The design of the buck power stage requires several compromises among size, efficiency, and cost. The inductor core loss increases with frequency, so there is a trade-off between a small output filter made possible by a higher switching frequency and getting better power supply efficiency. Size can be decreased by increasing the switching frequency at the expense of efficiency. Cost can be minimized by using through-hole inductors and capacitors; however these components are physically large.

To start the design, select a switching frequency based on [Table 9.](#page-16-7) This frequency is a starting point and may be adjusted as the design progresses.

Inductor Selection

The output inductor selection process must include several trade-offs. A high inductance value will result in a low ripple current (I_{opp}), which will reduce output capacitance and produce a low output ripple voltage, but may also compromise output transient load performance. Therefore, a balance must be struck between output ripple and optimal load transient performance. A good starting point is to select the output inductor ripple equal to the expected load transient step magnitude (I_{oster}): $I_{opp} = I_{oster}$.

Now the output inductance can be calculated using **[Equation 4](#page-16-5)**, where V_{INM} is the maximum input voltage:

$$
L_{OUT} = \frac{V_{OUT} \times \left(1 - \frac{V_{OUT}}{V_{INM}}\right)}{f_{SW} \times I_{opp}}
$$
 (EQ. 4)

The average inductor current is equal to the maximum output current. The peak inductor current $(I_{Lp}k)$ is calculated using **[Equation 5,](#page-16-6) where** I_{OUT} **is the maximum output current:**

$$
I_{Lpk} = I_{OUT} + \frac{I_{opp}}{2}
$$
 (EQ. 5)

Select an inductor rated for the average DC current with a peak current rating above the peak current computed in [Equation 5.](#page-16-6)

In overcurrent or short-circuit conditions, the inductor may have currents greater than 2x the normal maximum rated output current. It is desirable to use an inductor that still provides some inductance at maximum current to protect the load and the internal MOSFETs from damaging currents in this situation.

Once an inductor is selected, the DCR and core losses in the inductor are calculated. Use the DCR specified in the inductor manufacturer's datasheet.

$$
P_{LDCR} = DCR \times I_{Lrms}^2
$$
 (EQ. 6)

I_{Lrms} is given by:

$$
I_{\text{Lrms}} = \sqrt{I_{\text{OUT}}^2 + \frac{(I_{\text{opp}})^2}{12}}
$$
 (EQ. 7)

where I_{OUT} is the maximum output current. Next, calculate the core loss of the selected inductor. Since this calculation is specific to each inductor and manufacturer, refer to the chosen inductor datasheet. Add the core loss and the ESR loss and compare the total loss to the maximum power dissipation recommendation in the inductor datasheet.

Output Capacitor Selection

Several trade-offs must also be considered when selecting an output capacitor. Low ESR values are needed to have a small output deviation during transient load steps (V_{osag}) and low output voltage ripple (V_{orip}). However, capacitors with low ESR, such as semi-stable (X5R and X7R) dielectric ceramic capacitors, also have relatively low capacitance values. Many designs can use a combination of high capacitance devices and low ESR devices in parallel.

For high ripple currents, a low capacitance value can cause a significant amount of output voltage ripple. Likewise, in high transient load steps, a relatively large amount of capacitance is

needed to minimize the output voltage deviation while the inductor current ramps up or down to the new steady state output current value.

As a starting point, apportion one-half of the output ripple voltage to the capacitor ESR and the other half to capacitance, as shown in the following equations:

$$
C_{OUT} = \frac{I_{opp}}{8 \times f_{SW} \times \frac{V_{orig}}{2}}
$$
 (EQ. 8)

$$
ESR = \frac{V_{\text{orip}}}{2 \times l_{\text{opp}}} \tag{EQ.9}
$$

Use these values to make an initial capacitor selection, using a single capacitor or several capacitors in parallel.

After a capacitor has been selected, the resulting output voltage ripple can be calculated using **[Equation 10:](#page-17-2)**

$$
V_{\text{orip}} = I_{\text{opp}} \times ESR + \frac{I_{\text{opp}}}{8 \times f_{\text{SW}} \times C_{\text{OUT}}}
$$
 (EQ. 10)

Because each part of this equation was made to be less than or equal to half of the allowed output ripple voltage, the V_{orip} should be less than the desired maximum output ripple.

Input Capacitor

It is highly recommended that dedicated input capacitors be used in any point of load design, even when the supply is powered from a heavily filtered 5V or 12V "bulk" supply from an off-line power supply. This is because of the high RMS ripple current that is drawn by the buck converter topology. This ripple (I_{CINrms}) can be determined from **[Equation 11:](#page-17-3)**

$$
I_{\text{CINrms}} = I_{\text{OUT}} \times \sqrt{\text{D} \times (1-\text{D})}
$$
 (EQ. 11)

Without capacitive filtering near the power supply circuit, this current would flow through the supply bus and return planes, coupling noise into other system circuitry. The input capacitors should be rated at 1.2x the ripple current calculated above to avoid overheating of the capacitors due to the high ripple current, which can cause premature failure. Ceramic capacitors with X7R or X5R dielectric with low ESR and 1.1x the maximum expected input voltage are recommended.

BOOTSTRAP CAPACITOR SELECTION

The high-side driver boost circuit utilizes an internal Schottky diode (DB) and an external bootstrap capacitor (CB) to supply sufficient gate drive for the high-side MOSFET driver. CB should be a 0.1µF ceramic type rated for at least 10V.

CV2P5 SELECTION

This capacitor is used to both stabilize and provide noise filtering for the 2.5V internal power supply. It should be between 4.7µF and 10µF, should use a semi-stable X5R or X7R dielectric ceramic with a low ESR (less than 10mΩ) and should have a rating of 4V or more.

CVR SELECTION

This capacitor is used to both stabilize and provide noise filtering for the 7V reference supply. It should be 4.7µF, should use a

semi-stable X5R or X7R dielectric ceramic capacitor with a low ESR (less than 10mΩ) and should have a rating of 10V or more.

CVRA SELECTION

This capacitor is used to both stabilize and provide noise filtering for the analog 5V reference supply. It should be 4.7µF, should use a semi-stable X5R or X7R dielectric ceramic capacitor with a low ESR (less than $10 \text{m}\Omega$) and should have a rating of 6.3V or more.

THERMAL CONSIDERATIONS

In typical applications, the ZL2102's high efficiency will limit the internal power dissipation inside the package. However, in applications that require a high ambient operating temperature the user must perform some thermal analysis to ensure that the ZL2102's maximum junction temperature is not exceeded.

The ZL2102 has a maximum junction temperature limit of +125°C, and the internal over-temperature limiting circuitry will force the device to shut down if its junction temperature exceeds this threshold. In order to calculate the maximum junction temperature, the user must first calculate the power dissipated inside the IC (PQ) as expressed in **[Equation 12](#page-17-4):**

$$
P_Q = (I_{LOAD}^2)[(r_{DS(ON)QH})(D) + (r_{DS(ON)QL})(1 - D)]
$$
 (EQ. 12)

The operating junction temperature can then be calculated using **[Equation 13:](#page-17-5)**

$$
T_i = T_{pcb} + P_Q \times \theta_{JC}
$$
 (EQ. 13)

Where T_{nch} is the printed circuit board temperature (under the package) and θ_{JC} is the junction-to-case thermal resistance for the ZL2102 package.

PCB Layout Recommendation

The PCB layout is a very important step to make sure the designed converter works well. For ZL2102, the power system is composed of the input capacitor, VDDP pins, output inductor, SWITCH pins, output capacitor, and the PGND pins. It is necessary to group these connections as closely as possible and the connecting traces among them should be direct, short and wide. The switching node of the ZL2102 should connect directly to the inductor with minimal distance. The ZL2102 PGND pins, input and output capacitors should be connected as closely as possible to each other on the power GND plane. The input capacitor should be tightly coupled to the VIN pin. The VSEN voltage feedback trace should be routed to avoid the switch node, and should be connected directly to the pad of the output capacitor. The thermal pad is connected to SGND, and PGND must be isolated from SGND aside from a single connection point at the SGND pin.

The heat of the IC is mainly dissipated through the thermal pad. It is recommended to pass the heat from the thermal pad through the PCB using a large number of thermal vias. Connect as much copper as possible to these thermal vias throughout the PCB layers to help dissipate the heat, but do not route this SGND connection underneath any of the power components. The bottom layer is the best conductor of heat, so it should have at least 25mm² of copper connected to these vias. Any traces that must be routed in this copper area should be radial in nature so

that the thermal path from the thermal vias outward is not interrupted.

PMBus Command Summary

PMBus Command Summary (Continued)

PMBus Command Summary (Continued)

PMBus Data Formats

LINEAR-11 (L11)

L11 data format uses 5-bit two's compliment exponent (N) and 11-bit two's compliment mantissa (Y) to represent real world decimal value (X).

7 6 5 4 3 2 1 0 7 6 5 4 3 2 1 0 Data Byte High Data Byte Low

Exponent (N) Mantissa (Y)

Relation between real world decimal value (X) , N and Y is: $X = Y \cdot 2^N$

Linear-16 Unsigned (L16u)

L16u data format uses a fixed exponent (hardcode to N = -13) set by VOUT_MODE Command and 16-bit unsigned integer mantissa (Y) to represent real world decimal value (X).

Relation between real world decimal value (X), N and Y is: $X = Y \cdot 2^{-13}$

Bit Field (BIT)

Break down of Bit field is provided in "PMBus Command Summary" on page 19.

Custom (CUS)

Break down of Custom data format is provided in "PMBus Command Summary" on page 19. A combination of Bit field and integer are common type of Custom data format.

ASCII (ASC)

A variable length string of text characters uses ASCII data format.

PMBus Command Detail

OPERATION (01h)

Definition: Enable/disable command and VOUT Margin settings. The MGN pin has priority over the margin state of the device. Data values of OPERATION that force margin high or low only take effect when the MGN pin is left open (i.e., in the NOMINAL margin state). When the MGN pin has been set either high or low, bits 5:4 only report the margin state. When ON_OFF_CONFIG command is set for pin enable, Bits 7:6 only report the enable/disable status that has been set by the enable pin. All margin settings are "Act on Fault" type.

Data Length in Bytes: 1

Data Format: BIT

Type: R/W

Default Value: 04h (immediate off, nominal margin)

Units: N/A

ON_OFF_CONFIG (02h)

Definition: Configures the interpretation and coordination of the OPERATION command and the ENABLE pin (EN).

Data Length in Bytes: 1

Data Format: BIT

Type: R/W

Default Value: 17h (ENABLE pin control, active high)

Units: N/A

Reference: N/A

CLEAR_FAULTS (03h)

Definition: Clears all fault bits in all registers and releases the SALRT pin (if asserted) simultaneously. If a fault condition still exists, the bit will reassert immediately. This command will not restart a device if it has shut down, it will only clear the faults.

Data Length in Bytes: 0

Data Format: N/A

Type: Write only

Default Value: N/A

Units: N/A

STORE_USER_ALL (15h)

Definition: Stores all PMBus settings from the operating memory to the nonvolatile USER store memory. To clear the USER store, perform a RESTORE_FACTORY then STORE_USER_ALL. To add to the USER store, perform a RESTORE_USER_ALL, write commands to be added, then STORE_USER_ALL. This command can be used during device operation, but the device will be unresponsive for up to 20ms while storing values.

Data Length in Bytes: 0

Data Format: N/A

Type: Write only

Default Value: N/A

Units: N/A

Reference: N/A

RESTORE_USER_ALL (16h)

Definition: Restores all PMBus settings from the USER store memory to the operating memory. This command is performed automatically at power-up. This command can be used during device operation, but the device will be unresponsive for up to 20ms while storing values.

Data Length in Bytes: 0

Data Format: N/A

Type: Write only

Default Value: N/A

Units: N/A

Reference: N/A

VOUT_MODE (20H)

Definition: Reports the V_{OUT} mode and provides the exponent used in calculating several V_{OUT} settings.

Data Length in Bytes: 1

Data Format: BIT

Type: Read-only

Default Value: 13h (Linear Mode, Exponent = -13)

Units: N/A

VOUT_COMMAND (21h)

Definition: This command sets or reports the target output voltage. The integer value is multiplied by 2 raised to the power of VOUT_MODE. This command cannot be set to be higher than the value of VOUT_MAX. This command can be written while the device is enabled, but the adjusted value must be within 10% of the value that was selected prior to enable.

Data Length in Bytes: 2

Data Format: L16u.

Type: R/W

Default Value: Pin strap setting (VSET)

Units: Volts

Equation: V_{OUT} = VOUT_COMMAND×2⁻¹³

Range: 0.6V to 5.5V or the value of VOUT_MAX

Reference: N/A

Example: VOUT MODE = 13h (Since the value is 5-bit signed $13h = -13$)

VOUT_COMMAND = 699Ah = 27,034

Target voltage equals: $27034 \times 2^{-13} = 3.3V$

VOUT_MAX (24h)

Definition: The VOUT_ MAX command sets an upper limit on the output voltage the unit can command regardless of any other commands or combinations. The intent of this command is to provide a safeguard against a user accidentally setting the output voltage to a possibly destructive level rather than to be the primary output overprotection. If a VOUT_COMMAND is sent with a value higher than VOUT_MAX, the device will set the output voltage to VOUT_MAX. The initial value of VOUT_MAX is 110% of the pin strap value of VOUT.

Data Length in Bytes: 2

Data Format: L16u

Type: R/W

Default Value: 1.10 x VOUT_COMMAND pin strap setting

Units: Volts

Equation: V_{OUT} max = VOUT_MAX×2⁻¹³

Range: 0V to 5.5V

VOUT_MARGIN_HIGH (25h)

Definition: Sets the value of the VOUT when OPERATION or the MGN pin is set for "margin high". This command can be written while the device is enabled, but the adjusted value must be within 10% of the value of VOUT that was selected prior to enable.

Data Length in Bytes: 2

Data Format: L16u

Type: R/W

Default value: 1.05 x VOUT_COMMAND pin strap setting

Units: Volts

Equation: VOUT_MARGIN_HIGH = $Y \times 2^N$

Range: 0.54V to the value of VOUT_MAX

Reference: N/A

VOUT_MARGIN_LOW (26h)

Definition: Sets the value of the VOUT when OPERATION or the MGN pin is set for "margin low". This command can be written while the device is enabled, but the adjusted value must be within 10% of the value of VOUT that was selected prior to enable.

Data Length in Bytes: 2

Data Format: L16u

Type: R/W

Default value: 0.95 x VOUT_COMMAND pin strap setting

Units: Volts

Equation: VOUT_MARGIN_LOW = $Y \times 2^N$

Range: 0.54V to the value of VOUT_MAX

VOUT_TRANSITION_RATE (27h)

Definition: Sets the rate at which the output should change voltage when the device receives an OPERATION command (Margin High, Margin Low) or VOUT_COMMAND command that causes the output voltage to change. The maximum possible positive value of the two data bytes indicates that the device should make the transition as quickly as possible.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default value: B200h (0.5V/ms)

Units: V/ms

Equation: VOUT_TRANSITION_RATE = Y×2^N

Range: 0.1V/ms to 2V/ms

Reference: N/A

FREQUENCY_SWITCH (33h)

Definition: Sets the switching frequency of the device. Initial default value is defined by a pin strap and this value can be overridden by writing this command. If an external SYNC is utilized, this value should be set as close as possible to the external clock value. The output must be disabled when writing this command.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: Pin strap setting (SYNC)

Units: kHz

Equation: FREQUENCY_SWITCH = $Y \times 2^N$

Range: 200kHz to 1MHz

INTERLEAVE (37h)

Definition: Configures the phase offset of a device that is sharing a common SYNC clock with other devices. Note that a value of 0 for the Number in Group field is interpreted as 16, to allow for phase spreading groups of up to 16 devices.

Data Length in Bytes: 2

Data Format: BIT

Type: R/W

Default Value:

Default Group Number: 0 (00h)

Default Number in Group: 16 (00h)

Default Position in Group: Four LSB's of SMBus address

Units: N/A

Reference: [AN2034](http://www.intersil.com/content/dam/Intersil/documents/an20/an2034.pdf) - Configuring Current Sharing on the ZL2004 and ZL2006.

VOUT_OV_FAULT_LIMIT (40h)

Definition: Sets the VOUT overvoltage fault threshold.

Data Length in Bytes: 2

Data Format: L16u

Type: R/W

Default Value: 1.15 x VOUT_COMMAND pin strap setting

Units: Volts

Equation: VOUT OV fault limit = VOUT_OV_FAULT_LIMIT×2⁻¹³

Range: 0V to 6V

VOUT_OV_FAULT_RESPONSE (41h)

Definition: Configures the V_{OUT} overvoltage fault response. Only two settings are valid: 80h (immediate shutdown until commanded to restart) and BFh (immediate shutdown, 80ms delay, and then automatic restart once the fault condition has cleared).

Data Length in Bytes: 1

Data Format: BIT

Type: R/W

Default Value: BFh (Disable and retry continuously)

Units: N/A

Reference: N/A

VOUT_UV_FAULT_LIMIT (44h)

Definition: Sets the VOUT undervoltage fault threshold. This fault is masked during ramps (when PG is not set).

Data Length in Bytes: 2

Data Format: L16u

Type: R/W

Default Value: 0.85 x VOUT_COMMAND pin strap setting

Units: Volts

Equation: V_{OUT} UV fault limit = VOUT_UV_FAULT_LIMIT×2⁻¹³

Range: 0V to 6V

VOUT_UV_FAULT_RESPONSE (45h)

Definition: Configures the VOUT undervoltage fault response. Only two settings are valid: 80h (immediate shutdown until commanded to restart) and BFh (immediate shutdown, 80ms delay, and then automatic restart once the fault condition has cleared).

Data Length in Bytes: 1

Data Format: BIT

Type: R/W

Default Value: BFh (Disable and retry continuously)

Units: N/A

Reference: N/A

IOUT_OC_FAULT_LIMIT (46h)

Definition: Sets the IOUT peak overcurrent fault threshold. For down-slope sensing, this corresponds to the first current sample after the Current Sense Blanking Time has expired during the (1-D) time interval. For up-slope sensing, this corresponds to the last current sample of the D time interval. This feature shares the OC fault bit operation (in STATUS_IOUT) and OC fault response with IOUT_AVG_OC_FAULT_LIMIT.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: D240h (9A)

Units: A

Equation: $IOUT_OC_FAULT_LIMIT = Y \times 2^N$

Range: 0A to 9A

IOUT_UC_FAULT_LIMIT (4Bh)

Definition: Sets the IOUT valley undercurrent fault threshold. For down-slope sensing, this corresponds to the last current sample of the (1-D) time interval. For up-slope sensing, this corresponds to the first current sample of the D time interval, excluding the Current Sense Blanking time (which occurs at the beginning of the D interval). This feature shares the UC fault bit operation (in STATUS_IOUT) and UC fault response with IOUT_AVG_UC_FAULT_LIMIT.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: D5C0h (-9A)

Units: A

Equation: IOUT_OC_FAULT_LIMIT = Y×2^N

Range: 0A to -9A

Reference: N/A

OT_FAULT_LIMIT (4Fh)

Definition: Sets the over temperature fault threshold. Note that the temperature must drop below OT_WARN_LIMIT before the device will automatically restart.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: EBE8h (+125°C)

Units: Celsius

Equation: OT _FAULT_LIMIT = $Y \times 2^N$

Range: 0°C to +125°C

OT_FAULT_RESPONSE (50h)

Definition: Configures the over temperature fault response. Only two settings are valid: 80h (immediate shutdown until commanded to restart) and BFh (immediate shutdown, 250ms delay, and then automatic restart once the fault condition has cleared). The temperature must drop below OT_WARN_LIMIT before the device will restart.

Data Length in Bytes: 1

Data Format: BIT

Type: R/W

Default Value: BFh (Disable and retry continuously)

Units: N/A

Reference: N/A

OT_WARN_LIMIT (51h)

Definition: Sets the over temperature warning alarm threshold. In response to the threshold being exceeded, the device:

• Sets the TEMPERATURE bit in STATUS_WORD,

- Sets the OT_WARNING bit in STATUS_TEMPERATURE, and
- Notifies the host by setting the SALRT pin.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: EB70h (+110°C)

Units: Celsius

Equation: $OT_WARN_LIMIT = Y \times 2^N$

Range: 0°C to +125°C

UT_WARN_LIMIT (52h)

Definition: Sets the under-temperature warning alarm threshold. In response to the threshold being exceeded, the device:

- Sets the TEMPERATURE bit in STATUS_WORD,
- Sets the UT_WARNING bit in STATUS_TEMPERATURE, and
- Notifies the host by setting the SALRT pin.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: E4E0h (-50°C)

Units: Celsius

Equation: UT_WARN_LIMIT = $Y \times 2^N$

Range: -55°C to +25°C

Reference: N/A

UT_FAULT_LIMIT (53h)

Definition: Sets the under-temperature fault threshold. Note that the temperature must rise above UT_WARN_LIMIT before the device will automatically restart.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: E490h (-55°C)

Units: Celsius

Equation: UT _{FAULT}_LIMIT = $Y \times 2^N$

Range: -55°C to +25°C

Reference: N/A

UT_FAULT_RESPONSE (54h)

Definition: Configures the under-temperature fault response. Only two settings are valid: 80h (immediate shutdown until commanded to restart) and BFh (immediate shutdown, 250ms delay, and then automatic restart once the fault condition has cleared). The temperature must rise above UT_WARN_LIMIT before the device will automatically restart.

Data Length in Bytes: 1

Data Format: BIT Type: R/W Default Value: BFh (Disable and retry continuously) Units: N/A

VIN_OV_FAULT_LIMIT (55h)

Definition: Sets the VIN overvoltage fault threshold.

Data Length in Bytes: 2

Data Format: L11 Type: R/W Default Value: D380h (14V) Units: Volts Equation: $VIN_OV_FAULT_LIMIT = Y \times 2^N$ Range: 4.5V to 16V Reference: N/A

VIN_OV_FAULT_RESPONSE (56h)

Definition: Configures the V_{IN} overvoltage fault response as defined by the following table. Only two settings are valid: 80h (immediate shutdown until commanded to restart) and BFh (immediate shutdown, 80ms delay, and then automatic restart once the fault condition has cleared).

Data Length in Bytes: 1 Data Format: BIT Type: R/W Default Value: BFh (Disable and retry continuously) Units: N/A Reference: N/A

VIN_OV_WARN_LIMIT (57h)

Definition: Sets the V_{IN} overvoltage warning threshold as defined by the following table. In response to the OV_WARN_LIMIT being exceeded, the device:

- Sets the NONE OF THE ABOVE and INPUT bits in STATUS_WORD
- Sets the VIN_OV_WARNING bit in STATUS_INPUT, and
- Notifies the host by setting the SALRT pin.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: D360h (13.5V)

Units: Volts

Equation: VIN_OV_FAULT_LIMIT = Y×2^N

Range: 4.5V to 16V

VIN_UV_WARN_LIMIT (58h)

Definition: Sets the VIN undervoltage warning threshold. If a VIN_UV_FAULT occurs, the input voltage must rise above VIN_UV_WARN_LIMIT to clear the fault, which provides hysteresis to the fault threshold. In response to the UV_WARN_LIMIT being exceeded, the device:

- Sets the NONE OF THE ABOVE and INPUT bits in STATUS_WORD,
- Sets the VIN_UV_WARNING bit in STATUS_INPUT, and
- Notifies the host by setting the SALRT pin.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: 1.03 x VIN_UV_FAULT_LIMIT pin strap setting

Units: V

Equation: $VIN_UV_WARN_LIMIT = Y \times 2^N$

Range: 4.5V to 16V

Reference: N/A

VIN_UV_FAULT_LIMIT (59h)

Definition: Sets the VIN undervoltage fault threshold. Also referred to as undervoltage lockout (UVLO).

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: Pin strap setting (SS)

Units: Volts

Equation: $VIN_UV_FAULT_LIMIT = Y \times 2^N$

Range: 4.5V to 16V

Reference: N/A

VIN_UV_FAULT_RESPONSE (5Ah)

Definition: Configures the V_{IN} undervoltage fault response as defined by the following table. Only two settings are valid: 80h (immediate shutdown until commanded to restart) and BFh (immediate shutdown, 80ms delay, and then automatic restart once the fault condition has cleared).

Data Length in Bytes: 1

Data Format: BIT

Type: R/W

Default Value: BFh (Disable and retry continuously)

Units: N/A

POWER_GOOD_ON (5Eh)

Definition: Sets the voltage threshold for Power-Good indication. Power-Good asserts when the output voltage exceeds POWER_GOOD_ON and deasserts when the output voltage is less than VOUT_UV_FAULT_LIMIT.

Data Length in Bytes: 2

Data Format: L16u

Type: R/W

Default Value: 0.9 x VOUT_COMMAND pin strap setting

Units: Volts

Equation: Power-good on threshold = $POWER_GOOD_ON \times 2^{-13}$

Range: 0V to 5V

TON_DELAY (60h)

Definition: Sets the delay time from when the device is enabled to the start of V_{OUT} rise.

Data Length in Bytes: 2 Data Format: L11 Type: R/W Default Value: Pin strap setting (SS) Units: ms Equation: TON _{DELAY} = $Y \times 2^N$ Range: 5ms to 30s Reference: N/A

TON_RISE (61h)

Definition: Sets the rise time of V_{OUT} after ENABLE and TON_DELAY.

Data Length in Bytes: 2 Data Format: L11 Type: R/W Default Value: Pin strap setting (SS) Units: ms Equation: $TON_RISE = Y \times 2^N$ Range: 5ms to 200ms Reference: N/A

TOFF_DELAY (64h)

Definition: Sets the delay time from DISABLE to start of V_{OUT} fall.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: 1 x TON_DELAY pin strap value

Units: ms

Equation: TON_DELAY = Y×2^N

Range: 5ms to 30s

TOFF_FALL (65h)

Definition: Sets the fall time for V_{OUT} after DISABLE and TOFF_DELAY.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: 1 x TON_RISE pin strap setting

Units: ms

Equation: TOFF_FALL = $Y \times 2^N$

Range: 5ms to 200ms

STATUS_WORD (79h)

Definition: Returns fault condition status information. Based on the information in these bytes, the host can get more information by reading the appropriate status registers. Status bits are only cleared by a forced restart or by writing to the CLEAR_FAULTS PMBus command.

Data Length in Bytes: 2

Data Format: BIT

Type: Read-only

Default Value: N/A

Units: N/A

Reference: N/A

NOTE:

12. If the POWER_GOOD# bit is set, this indicates that the POWER_GOOD signal, if present, is signaling that the output power is not good.

STATUS_VOUT (7Ah)

Definition: Returns output voltage status information. Status bits are only cleared by a forced restart or by writing to the CLEAR_FAULTS PMBus command.

Data Length in Bytes: 1

Data Format: BIT

Type: Read-only

Default Value: N/A

Units: N/A

Reference: N/A

STATUS_IOUT (7Bh)

Definition: Returns the output current status information. Status bits are only cleared by a forced restart or by writing to the CLEAR_FAULTS PMBus command.

Data Length in Bytes: 1

Data Format: BIT

Type: Read-only

Default Value: N/A

Units: N/A

STATUS_INPUT (7Ch)

Definition: Returns input voltage and input current status information. Status bits are only cleared by a forced restart or by writing to the CLEAR_FAULTS PMBus command.

Data Length in Bytes: 1

Data Format: BIT

Type: Read-only

Default Value: N/A

Units: N/A

Reference: N/A

STATUS_TEMPERATURE (7Dh)

Definition: Returns temperature related status information. Status bits are only cleared by a forced restart or by writing to the CLEAR_FAULTS PMBus command.

Data Length in Bytes: 1

Data Format: BIT

Type: Read-only

Default Value: N/A

Units: N/A

STATUS_CML (7Eh)

Definition: Returns Communications, Logic and/or Memory status information. Status bits are only cleared by a forced restart or by writing to the CLEAR_FAULTS PMBus command.

Data Length in Bytes: 1

Data Format: BIT

Type: Read-only

Default Value: N/A

Units: N/A

Reference: N/A

STATUS_MFR_SPECIFIC (80h)

Definition: Returns clock synchronization status. Only bit 3 is used on this command for this device. Status bits are only cleared by a forced restart or by writing to the CLEAR_FAULTS PMBus command.

Data Length in Bytes: 1

Data Format: BIT

Type: Read only

Default value: N/A

Units: N/A

READ_VIN (88h)

Definition: Returns the input voltage reading.

Data Length in Bytes: 2 Data Format: L11 Type: Read-only Default Value: N/A Units: Volts Equation: READ_VIN = Y×2^N Range: N/A Reference: N/A

READ_VOUT (8Bh)

Definition: Returns the output voltage reading.

Data Length in Bytes: 2

Data Format: L16u

Type: Read-only

Default Value: N/A

Units: Volts

Equation: Read V_{OUT} = READ_VOUT×2⁻¹³

Reference: N/A

READ_IOUT (8Ch)

Definition: Returns the output current reading.

Data Length in Bytes: 2

Data Format: L11

Type: Read-only

Default Value: N/A

Units: A

Equation: READ_IOUT = Y×2^N

Range: N/A

READ_TEMPERATURE_1 (8Dh)

Definition: Returns the temperature reading internal to the device.

Data Length in Bytes: 2

Data Format: L11

Type: Read-only

Default Value: N/A

Units: °C

Equation: READ_TEMPERATURE_1 = $Y \times 2^N$

Range: N/A

Reference: N/A

READ_DUTY_CYCLE (94h)

Definition: Reports the actual duty cycle of the converter during the enable state.

Data Length in Bytes: 2 Data Format: L11 Type: Read only Default Value: N/A Units: % **Equation: READ_DUTY_CYCLE =** $Y \times 2^N$ Range: N/A Reference: N/A

READ_FREQUENCY (95h)

Definition: Reports the actual switching frequency of the converter during the enable state.

Data Length in Bytes: 2

Data Format: L11

Type: Read only

Default Value: N/A

Units: kHz

Equation: READ_FREQUENCY = Y×2^N

Range: N/A

PMBUS_REVISION (98h)

Definition: The PMBUS_REVISION command returns the revision of the PMBus specification to which the device is compliant.

Data Length in Bytes: 1

Data Format: BIT

Type: Read only

Default Value: 01h (Part 1 Revision 1.0, Part 2 Revision 1.1)

Units: N/A

Reference: N/A

MFR_ID (99h)

Definition: Sets a user defined identification. The sum total of characters in MFR_ID and USER_DATA_00 plus one byte per command cannot exceed 128 characters. This limitation includes multiple writes of this command before a STORE command. To clear multiple writes, perform a RESTORE, write this command then perform a STORE/RESTORE.

Data Length in Bytes: user defined

Data Format: ASC

Type: Block R/W

Default Value: <null>

Units: N/A

Reference: N/A

IC_DEVICE_ID (ADH)

Definition: Reports device identification information.

Data Length in Bytes: 4

Data Format: CUS

Type: Block Read

Default Value: 49A01200h

Units: N/A

IC_DEVICE_REV (AEH)

Definition: Reports device revision information.

Data Length in Bytes: 4

Data Format: CUS

Type: Block Read

Default Value:

Units: N/A

Reference: N/A

USER_DATA_00 (B0h)

Definition: Sets a user defined data. The sum total of characters in MFR_ID and USER_DATA_00 plus one byte per command cannot exceed 128 characters. This limitation includes multiple writes of this command before a STORE command. To clear multiple writes, perform a RESTORE, write this command then perform a STORE/RESTORE.

Data Length in Bytes: user defined

Data Format: ASC

Type: Block R/W

Default Value: <null>

Units: N/A

Reference: N/A

AUTO_COMP_CONFIG (BCh)

Definition: Configures the auto compensation algorithm.

Data Length in Bytes: 1

Data Format: BIT

Type: R/W

Default Value: Pin strap setting (FC)

Units: N/A

AUTO_COMP_CONTROL (BDh)

Definition: Writing the AUTO_COMP_CONTROL command will initiate the auto compensation algorithm, provided that it has been enabled in AUTO_COMP_CONFIG.

Data Length in Bytes: 0 Byte

Type: Write only

Default Value: N/A

Units: N/A

Reference: N/A

MFR_CONFIG (D0h)

Definition: Configures several manufacturer-level features.

Data Length in Bytes: 2

Data Format: BIT

Type: R/W word

Default Value: 4801h

Units: N/A

Range: N/A

USER_CONFIG (D1h)

Definition: Configures several user-level features.

Data Length in Bytes: 2

Data Format: BIT

Type: R/W

Default Value: Pin strap setting

Units: N/A

DDC_CONFIG (D3h)

Definition: Configures the DDC bus.

Data Length in Bytes: 2

Data Format: BIT

Type: R/W

Default Value: Broadcast Group: 0; DDC ID: Lowest five bits of the SMBus Address.

Units: N/A

POWER_GOOD_DELAY (D4h)

Definition: Sets the delay applied between the output exceeding the PG threshold (POWER_GOOD_ON) and asserting the PG pin. The delay time can range from 1ms up to 30s. When auto comp is enabled and is set to assert the PG pin after completion, this command will be ignored.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: BA00h (1ms)

Units: ms

Equation: POWER_GOOD_DELAY = $(Y \times 2^N)$

Range: 1ms to 30s

Reference: N/A

PID_TAPS (D5h)

Definition: This command configures the control loop compensator coefficients. The PID algorithm implements the following Z-domain:

 $A + Bz^{-1} + Cz^{-2}$ $1 - z^{-1}$ $\frac{A + BZ + CZ}{4}$ (EQ. 14)

The coefficients A, B, and C are represented using a pseudo-floating point format similar to the VOUT parameters (with the addition of a sign bit), defined as **Equation 15:**

$$
(A = (-1)^S) \bullet 2^E \bullet M
$$

 $(EQ. 15)$

where M is a two-byte unsigned mantissa, S is a sign-bit, and E is a 7-bit two's-complement signed integer. The 9 byte data field is defined in the following table. S is stored as the MSB of the E byte.

NOTE: Data bytes are transmitted on the PMBus in the order of Byte 0 through Byte 8.

Data Length in Bytes: 9

Data Format: CUS

Type: Block R/W

Default Value:

Auto Comp Off, taps stored $-$ (A, B, C) = stored values

Auto Comp Off, no taps stored - (A, B, C) correspond to (G, Q, fn) = (20dB, 2, $f_{SW}/10$)

Auto Comp On - (A, B, C) = Auto Comp results

Units: N/A

Reference: **[AN2035](http://www.intersil.com/content/dam/Intersil/documents/an20/an2035.pdf)** - Compensation Using CompZL[™]

SEQUENCE (E0h)

Definition: Identifies the Rail DDC ID of the prequel and sequel rails when performing multi-rail sequencing. The device will enable its output when its EN or OPERATION enable state, as defined by ON_OFF_CONFIG, is set and the prequel device has issued a Power-Good event on the DDC bus. The device will disable its output (using the programmed delay values) when the sequel device has issued a Power-Down event on the DDC bus.

The data field is a two-byte value. The most-significant byte contains the 5-bit Rail DDC ID of the prequel device. The least-significant byte contains the 5-bit Rail DDC ID of the sequel device. The most significant bit of each byte contains the enable of the prequel or sequel mode. This command overrides the corresponding sequence configuration set by the CONFIG pin settings.

Data Length in Bytes: 2

Data Format: CUS

Type: R/W

Default Value: Pin strap setting (CFG)

Units: N/A

DDC_GROUP (E2h)

Definition: Sets which rail DDC IDs are monitored for fault spreading information. The data sent is a 4-byte, 32-bit, bit vector where every bit represents a rail's DDC ID. Setting a bit to 1 will include that rail's DDC ID in the group. All DDC ID's that are selected will be monitored. If fault spread mode is enabled in USER_CONFIG (Bit 8 set to 1), the rail will respond to any fault spreading events within the group. The device will immediately shut down if one of its DDC_GROUP members fails. The device/rail will attempt its configured restart only after all devices/rails within the DDC_GROUP have cleared their faults.

Note: The device/rail's own DDC ID should not be set within the DDC_GROUP command for that device/rail.

All devices in a current share rail must shutdown for the rail to report a shutdown.

If fault spread mode is disabled in USER_CONFIG (Bit 8 cleared to 0), the device will perform a sequenced shutdown as defined by the SEQUENCE command setting. The rails/devices in a sequencing set only attempt their configured restart after all faults have cleared within the DDC_GROUP.

Data Length in Bytes: 4

Data Format: BIT

Type: R/W

Default Value: 00000000h Units: N/A

Reference: N/A

DEVICE_ID (E4h)

Definition: Returns the 16-byte (character) device identifier string.

Data Length in Bytes: 16

Data Format: ASC

Type: Block Read

Default Value: <part number/die revision/firmware revision>

Units: N/A

Reference: N/A

MFR_IOUT_OC_FAULT_RESPONSE (E5h)

Definition: Configures the I_{OUT} overcurrent fault response as defined by the following table. Only two settings are valid: 80h (immediate shutdown until commanded to restart) and BFh (immediate shutdown, 80ms delay, and then automatic restart once the fault condition has cleared). The command format is the same as the PMBus standard fault responses except that it sets the overcurrent status bit in STATUS IOUT.

Data Length in Bytes: 1

Data Format: BIT

Type: R/W

Default Value: BFh (Disable and retry continuously)

Units: N/A

Reference: N/A

November 20, 2014

MFR_IOUT_UC_FAULT_RESPONSE (E6h)

Definition: Configures the IOUT undercurrent fault response as defined by the following table. Only two settings are valid: 80h (immediate shutdown until commanded to restart) and BFh (immediate shutdown, 80ms delay, and then automatic restart once the fault condition has cleared). The command format is the same as the PMBus standard fault responses except that it sets the undercurrent status bit in STATUS_IOUT.

Data Length in Bytes: 1

Data Format: BIT

Type: R/W

Default Value: BFh (Disable and retry continuously)

Units: N/A

IOUT_AVG_OC_FAULT_LIMIT (E7h)

Definition: Sets the IOUT average overcurrent fault threshold. For down-slope sensing, this corresponds to the average of all the current samples taken during the (1-D) time interval, excluding the Current Sense Blanking time (which occurs at the beginning of the 1-D interval). For up-slope sensing, this corresponds to the average of all the current samples taken during the D time interval, excluding the Current Sense Blanking time (which occurs at the beginning of the D interval). This feature shares the OC fault bit operation (in STATUS_IOUT) and OC fault response with IOUT_ OC_FAULT_LIMIT.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: CB99h (7.2A)

Units: Amperes

Equation: IOUT_AVG_OC_FAULT_LIMIT = Y×2^N

Range: 0A to 9A

Reference: N/A

IOUT_AVG_UC_FAULT_LIMIT (E8h)

Definition: Sets the IOUT average undercurrent fault threshold. For down-slope sensing, this corresponds to the average of all the current samples taken during the (1-D) time interval, excluding the Current Sense Blanking time (which occurs at the beginning of the 1-D interval). For up-slope sensing, this corresponds to the average of all the current samples taken during the D time interval, excluding the Current Sense Blanking time (which occurs at the beginning of the D interval). This feature shares the UC fault bit operation (in STATUS_IOUT) and UC fault response with IOUT_ UC_FAULT_LIMIT.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: CC67h (-7.2A)

Units: Amperes

Equation: $10UT_AVG_UC_FAULT_LIMIT = Y \times 2^N$

Range: 0A to -9A

MISC_CONFIG (E9h)

Definition: Sets options pertaining to advanced features.

Data Length in Bytes: 2

Data Format: BIT

Type: R/W

Default Value: 00h

Units: N/A

Reference: N/A

SNAPSHOT (EAh)

Definition: The SNAPSHOT command is a 32-byte read-back of parametric and status values. It allows monitoring and status data to be stored to NV memory either during a fault condition or via a system-defined time using the SNAPSHOT_CONTROL command.

- 1. To use the snapshot feature, it must first be enabled by setting bit 1 (Snapshot) in MISC_CONFIG to 1 (Enabled).
- 2. By default, snapshot is continuously updated in RAM and can be read using the SNAPSHOT command.
- 3. When a fault occurs, the latest snapshot in RAM is stored to NV memory. After this, one can read back the snapshot stored in NV memory by writing a 01h to the SNAPSHOT_CONTROL command, then reading SNAPSHOT. This step must be performed while the device's operation is disabled, or when snapshot is temporarily disabled (via MISC_CONFIG).

Data Length in Bytes: 32

Data Format: CUS

Type: Block Read

Default Value: N/A

Units: N/A

BLANK_PARAMS (EBh)

Definition: Returns a 16-byte string that indicates which parameter values were either retrieved by the last RESTORE operation or have been written since that time. Reading BLANK_PARAMS immediately after a restore operation allows the user to determine which parameters are stored in that store. A one indicates the parameter is not present in the store and has not been written since the RESTORE operation. This command is used internally to determine if pin strap values should be used. Contact the factory for the BLANK_PARAMS bit-map if needed.

Data Length in Bytes: 16

Data Format: BIT

Type: Block Read

Default Value: FF...FFh

Units: N/A

Reference: N/A

SNAPSHOT_CONTROL (F3h)

Definition: Writing a 1h will cause the device to copy the current SNAPSHOT values from NV memory to the 32-byte SNAPSHOT command parameter. Writing a 2h will cause the device to write the current SNAPSHOT values to NV memory. All other values will be ignored. Output must be disabled when writing to NV memory.

Data Length in Bytes: 1

Data Format: BIT

Type: R/W

Default Value: 00h

Units: N/A

RESTORE_FACTORY (F4h)

Definition: Restores the device to the hard-coded Default values and pin strap definitions. The device retains the USER store for restoring. Output must be disabled when writing this command.

Data Length in Bytes: 0

Data Format: N/A Type: Write only Default Value: N/A Units: N/A Reference: N/A

Revision History

The revision history provided is for informational purposes only and is believed to be accurate, but not warranted. Please go to the web to make sure that you have the latest revision.

About Intersil

Intersil Corporation is a leading provider of innovative power management and precision analog solutions. The company's products address some of the largest markets within the industrial and infrastructure, mobile computing and high-end consumer markets.

For the most updated datasheet, application notes, related documentation and related parts, please see the respective product information page found at [www.intersil.com.](www.intersil.com)

You may report errors or suggestions for improving this datasheet by visiting [www.intersil.com/ask](http://www.intersil.com/en/support/support-faqs.html?p_page=ask.php&p_prods=679&p_icf_7=ZL2102).

Reliability reports are also available from our website at [www.intersil.com/support](http://www.intersil.com/en/support/qualandreliability.html#reliability)

© Copyright Intersil Americas LLC 2013-2014. All Rights Reserved. All trademarks and registered trademarks are the property of their respective owners.

For additional products, see [www.intersil.com/en/products.html](http://www.intersil.com/en/products.html?utm_source=Intersil&utm_medium=datasheet&utm_campaign=disclaimer-ds-footer)

[Intersil products are manufactured, assembled and tested utilizing ISO9001 quality systems as noted](http://www.intersil.com/en/products.html?utm_source=Intersil&utm_medium=datasheet&utm_campaign=disclaimer-ds-footer) in the quality certifications found at [www.intersil.com/en/support/qualandreliability.html](http://www.intersil.com/en/support/qualandreliability.html?utm_source=Intersil&utm_medium=datasheet&utm_campaign=disclaimer-ds-footer)

Intersil products are sold by description only. Intersil may modify the circuit design and/or specifications of products at any time without notice, provided that such modification does not, in Intersil's sole judgment, affect the form, fit or function of the product. Accordingly, the reader is cautioned to verify that datasheets are current before placing orders. Information furnished by Intersil is believed to be accurate and reliable. However, no responsibility is assumed by Intersil or its subsidiaries for its use; nor for any infringements of patents or other rights of third parties which may result from its use. No license is granted by implication or otherwise under any patent or patent rights of Intersil or its subsidiaries.

For information regarding Intersil Corporation and its products, see [www.intersil.com](http://www.intersil.com?utm_source=intersil&utm_medium=datasheet&utm_campaign=disclaimer-ds-footer)

Package Outline Drawing

L36.6x6A

36 LEAD QUAD FLAT NO-LEAD PLASTIC PACKAGE Rev 1, 9/09

NOTES:

- **Dimensions in () for Reference Only. 1. Dimensions are in millimeters.**
- **Dimensioning and tolerancing conform to AMSE Y14.5m-1994. 2.**
- **Unless otherwise specified, tolerance : Decimal ± 0.05 3.**
- **between 0.15mm and 0.30mm from the terminal tip. Dimension b applies to the metallized terminal and is measured 4.**
- **Tiebar shown (if present) is a non-functional feature. 5.**
- **located within the zone indicated. The pin #1 indentifier may be The configuration of the pin #1 identifier is optional, but must be 6. either a mold or mark feature.**
- **7. Compliant to JEDEC MO-220VJJD.**

